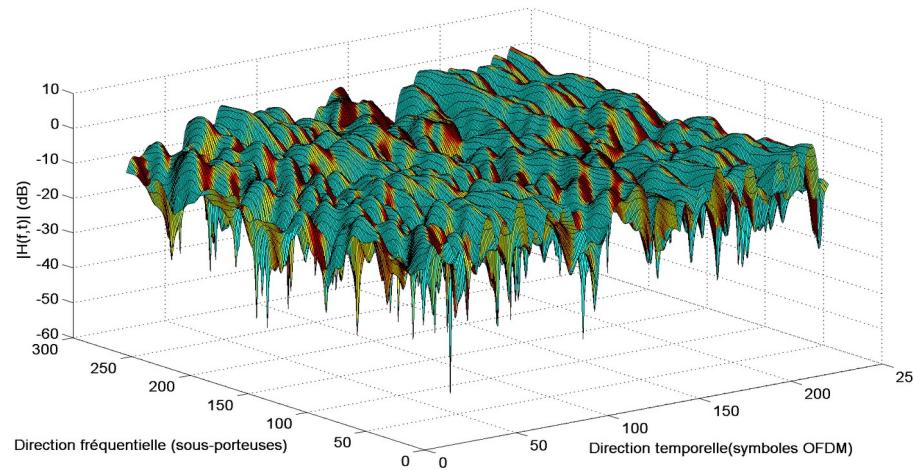


The mobile radio channel: a review



European **GNU Radio** Days 2022



Hervé Boeglen F4JET – SDRA 2022

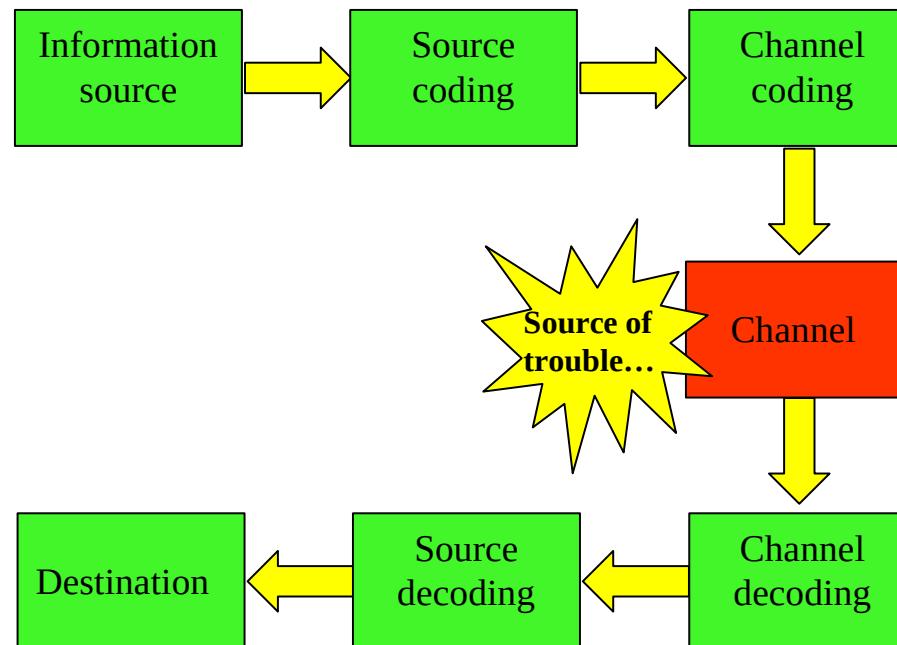


Outline

- 1. Introduction
- 2. Modeling the mobile radio channel
- 3. Simulation techniques
- 4. Conclusion

1. Introduction

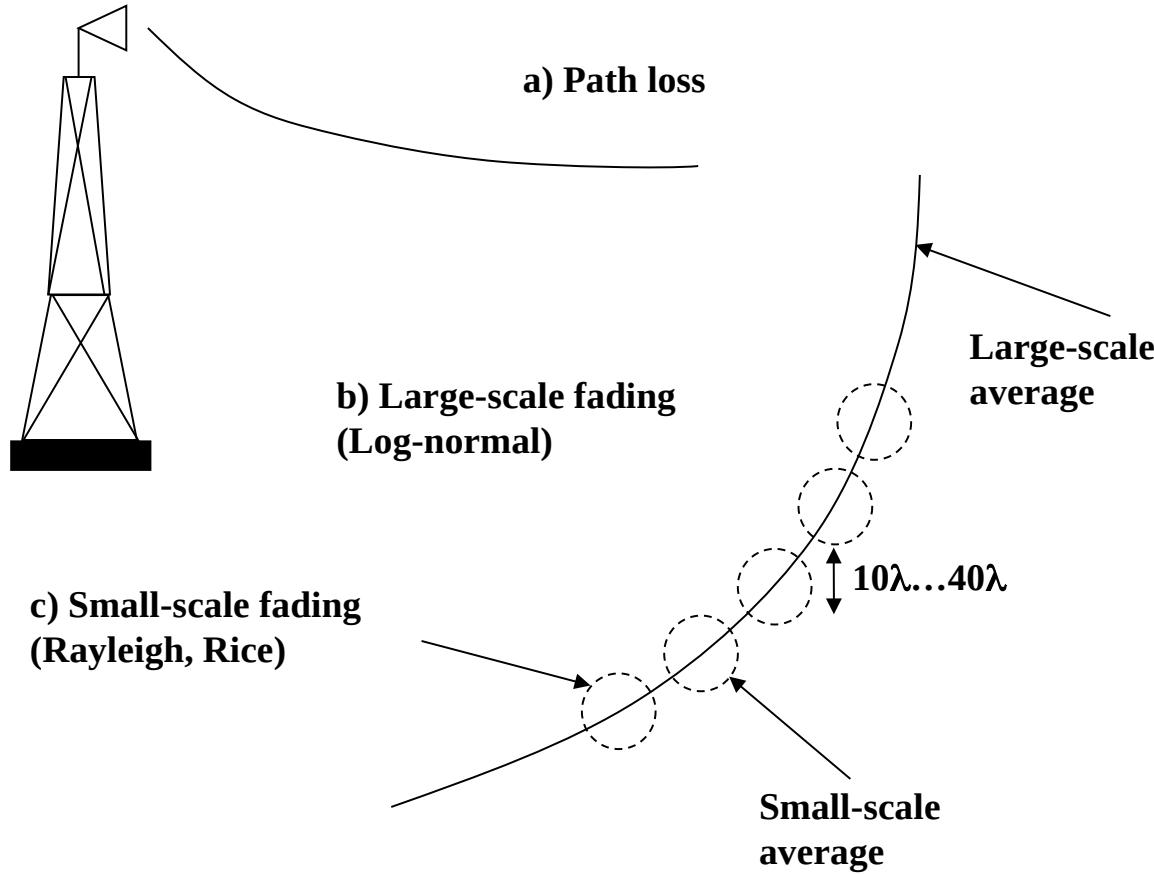
- 1948 [C. Shannon](#) « *A Mathematical Theory of Communication* » :



- Channel coding techniques (e.g. OFDM) depend on channel behaviour (e.g. to fight against frequency selectivity).

1. Introduction

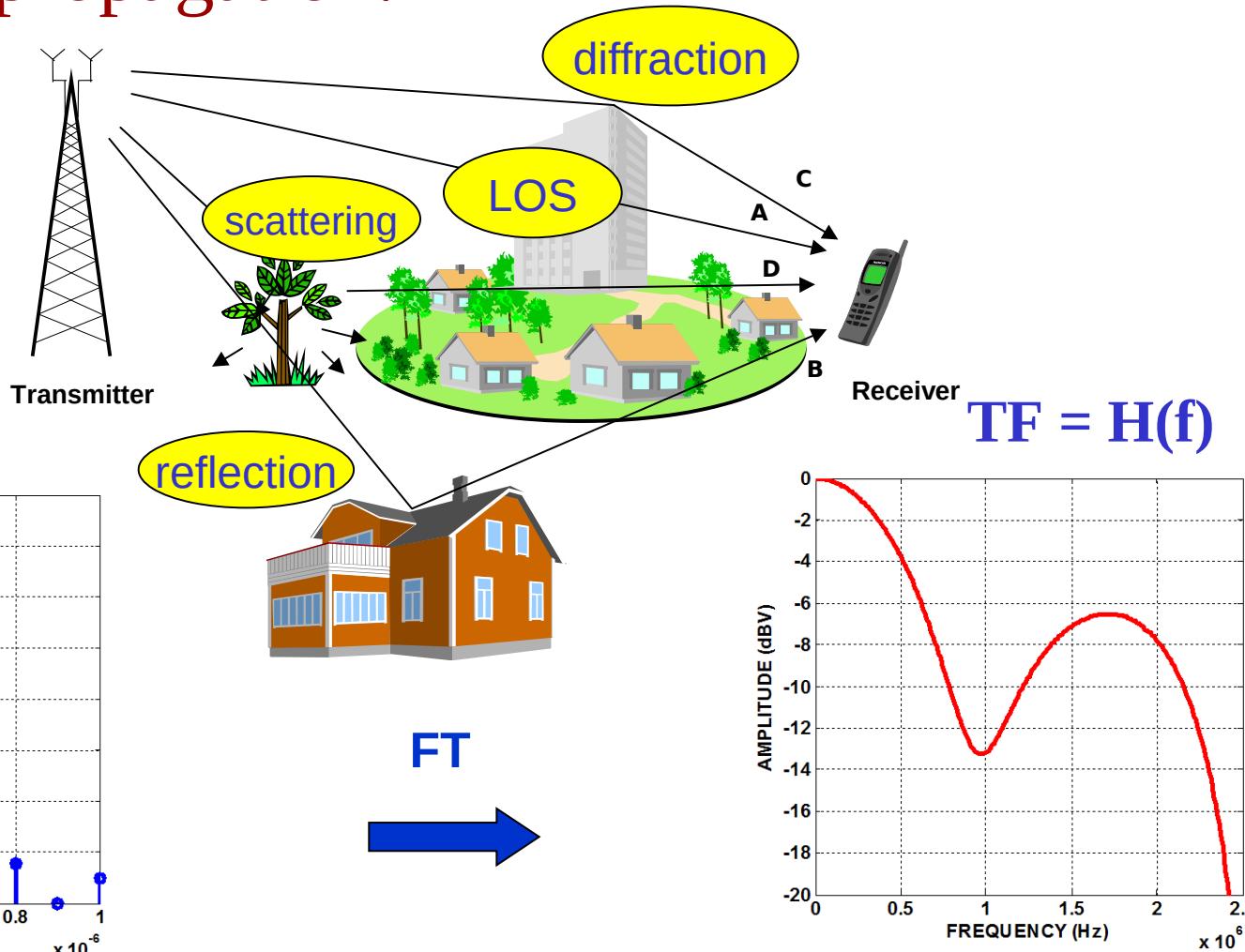
- Types of receive power variations:



- We are interested in the small-scale behaviour

2. Modeling the mobile radio channel

➤ Multipath propagation:



→ Spectrum distortion of the transmitted signal 5

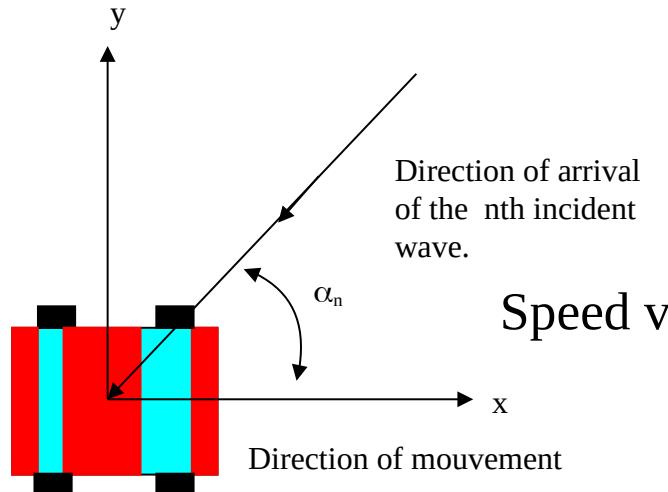
2. Modeling the mobile radio channel

➤ Doppler effect (T and/or R are moving) :

$$f_n = f_{\max} \cdot \cos(\alpha_n)$$

$$f_{\max} = \frac{v}{c_0} \cdot f_0$$

f_0 = carrier frequency



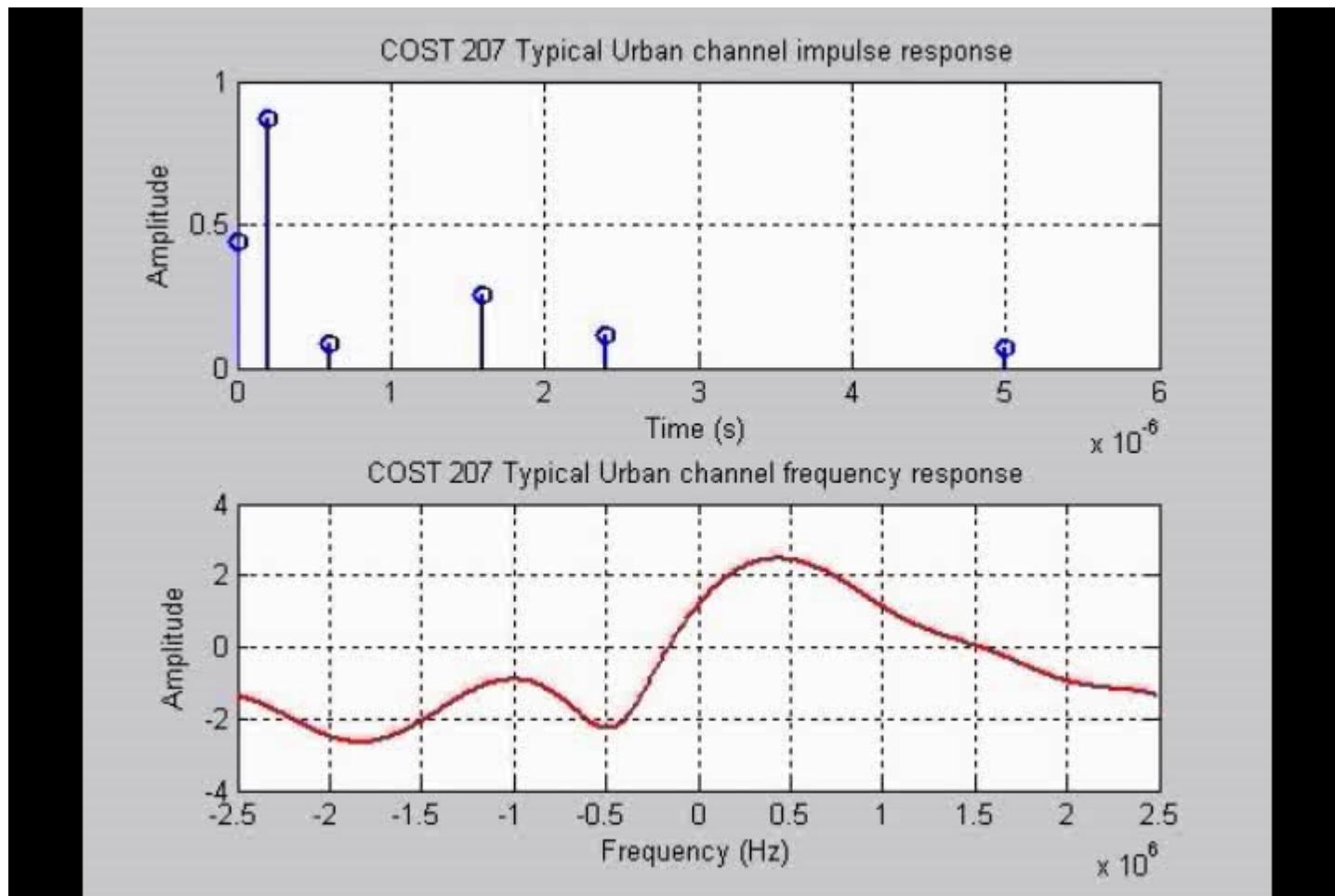
The transmitted signal spectrum undergoes a frequency expansion



The CIR becomes time dependant

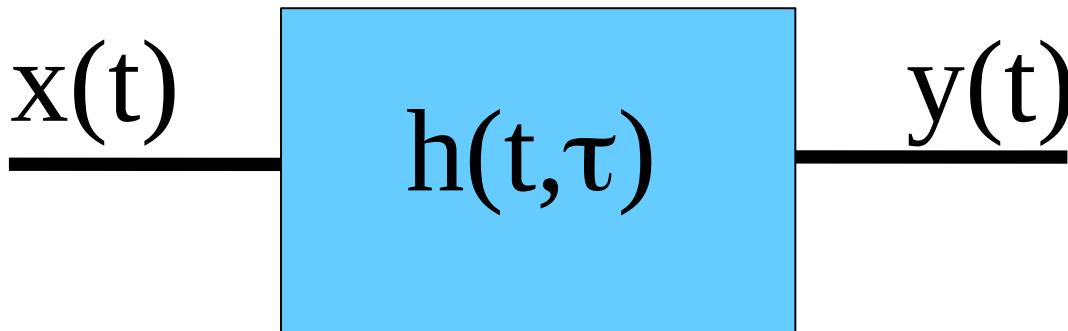
2. Modeling the mobile radio channel

- Doppler effect (T and/or R are moving) :



2. Modeling the mobile radio channel

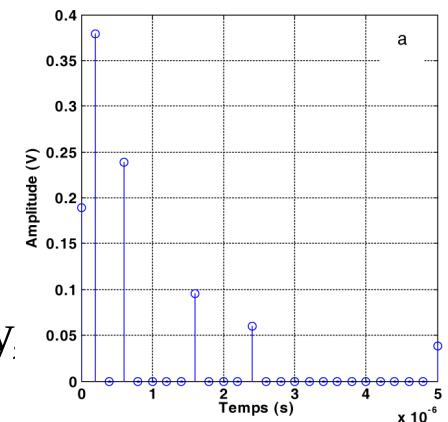
- Mobile radio channel = linear system:



- Channel Impulse Response (CIR) :

$$h(\tau, t) = \sum_{n=0}^N \alpha_n(t) e^{-j\phi_n(t)} \delta(\tau - \tau_n(t))$$

- ★ With N = number of paths, $\tau_n(t) = r_n(t)/c = n_{th}$ path delay
 ϕ_n = n_{th} path phase shift (Doppler) and $\alpha_n(t)$ n_{th} path amplitude.



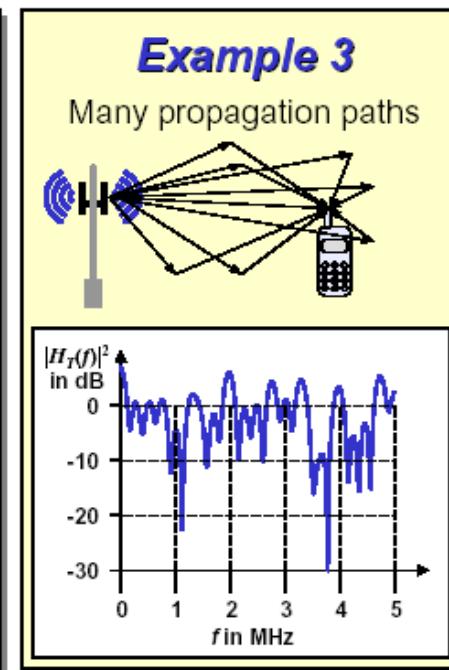
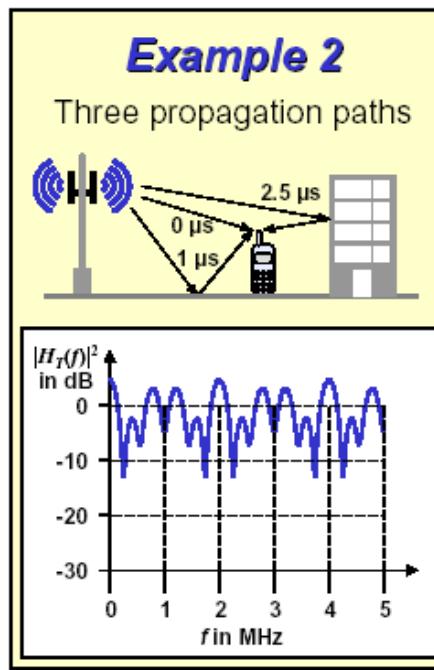
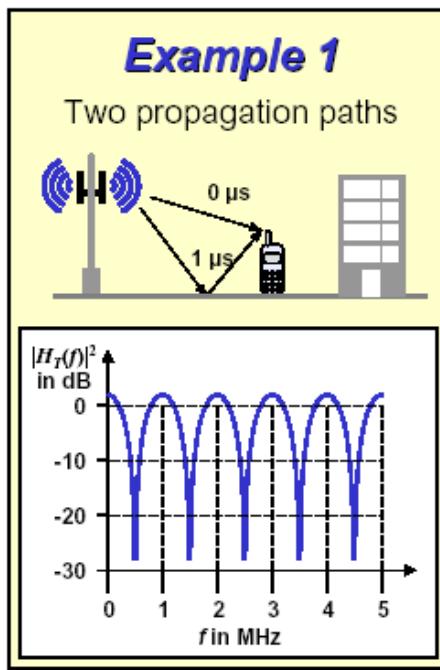
2. Modeling the mobile radio channel

- Two parameters can vary: τ and t

- ❖ $h(t, \tau)$ does not depend on t :

$$h(\tau, t) = h(\tau) = \sum_{n=0}^N \alpha_n e^{-j\varphi_n} \delta(\tau - \tau_n)$$

→ Time invariant channel.



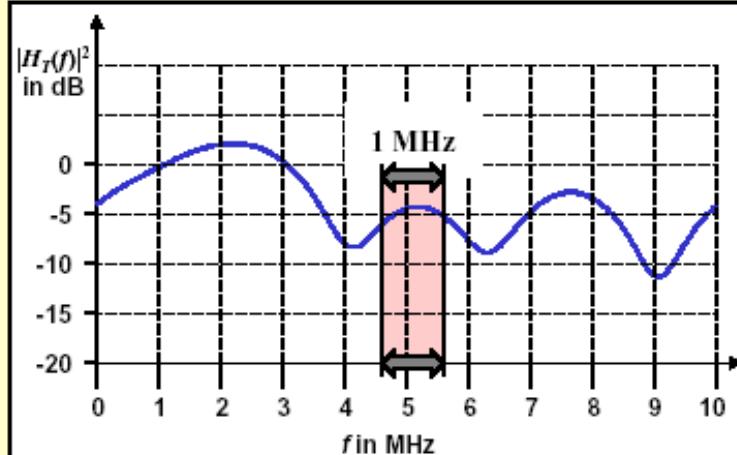
The signals coming from different paths interfere either constructively or destructively
→ FREQUENCY SELECTIVITY.

2. Modeling the mobile radio channel

- Influence of the delays duration on the channel transfer function:

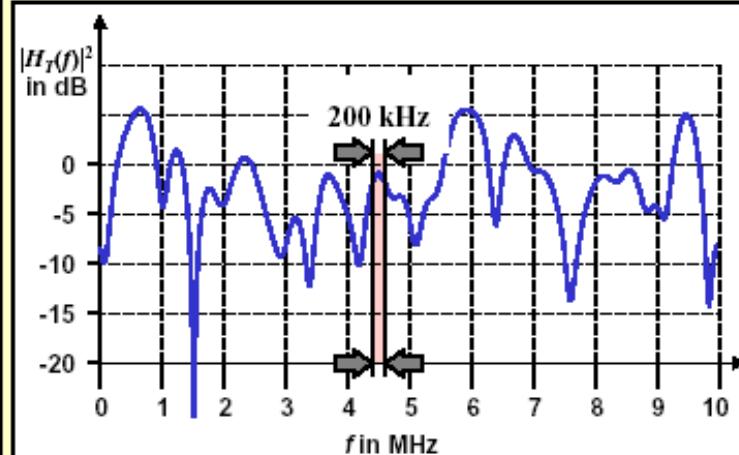
Small maximum path delay

$$\tau_{\max} = 1 \mu\text{s} \Rightarrow \frac{1}{\tau_{\max}} = 1 \text{ MHz}$$



Large maximum path delay

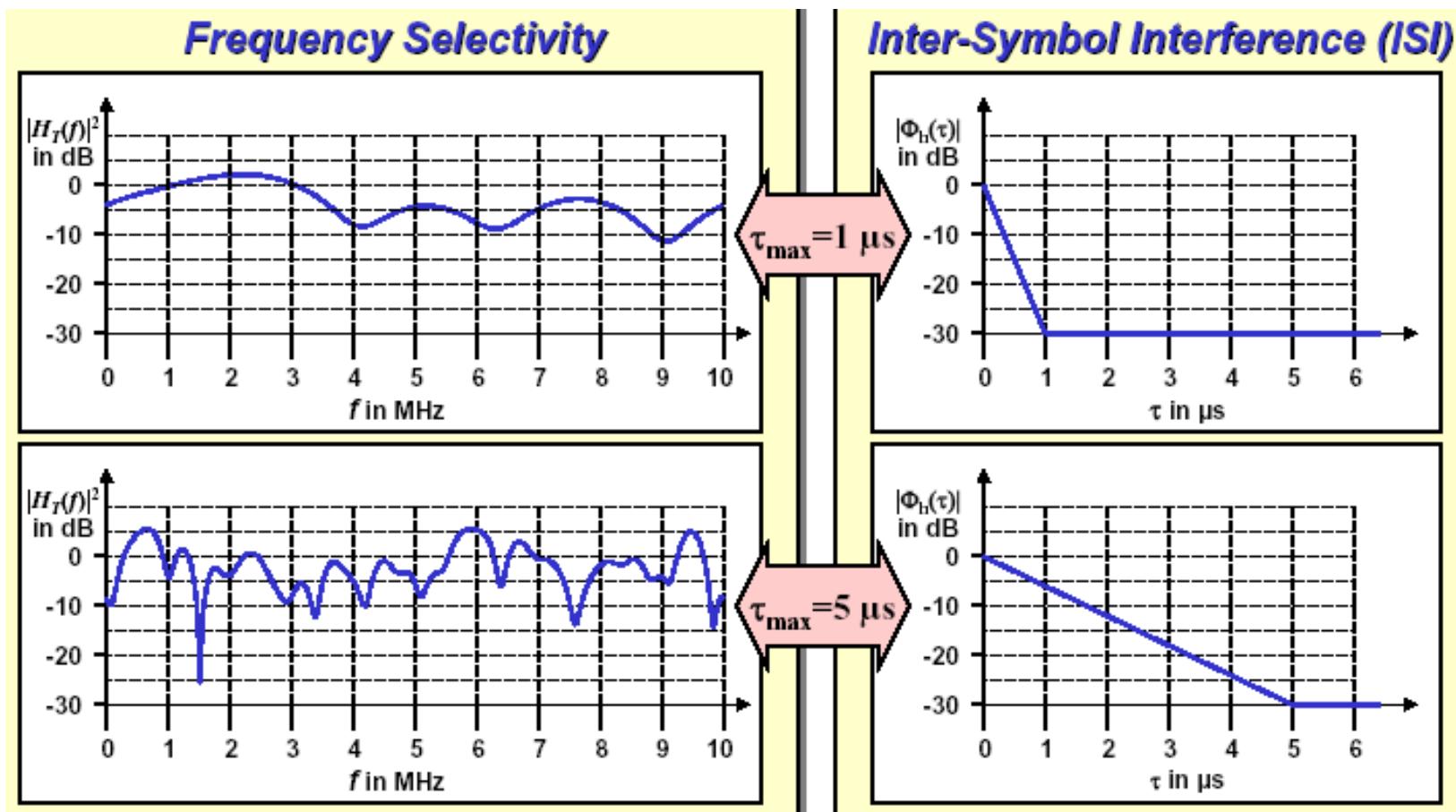
$$\tau_{\max} = 5 \mu\text{s} \Rightarrow \frac{1}{\tau_{\max}} = 200 \text{ kHz}$$



⇒ The larger τ_{\max} the more selective the channel.

2. Modeling the mobile radio channel

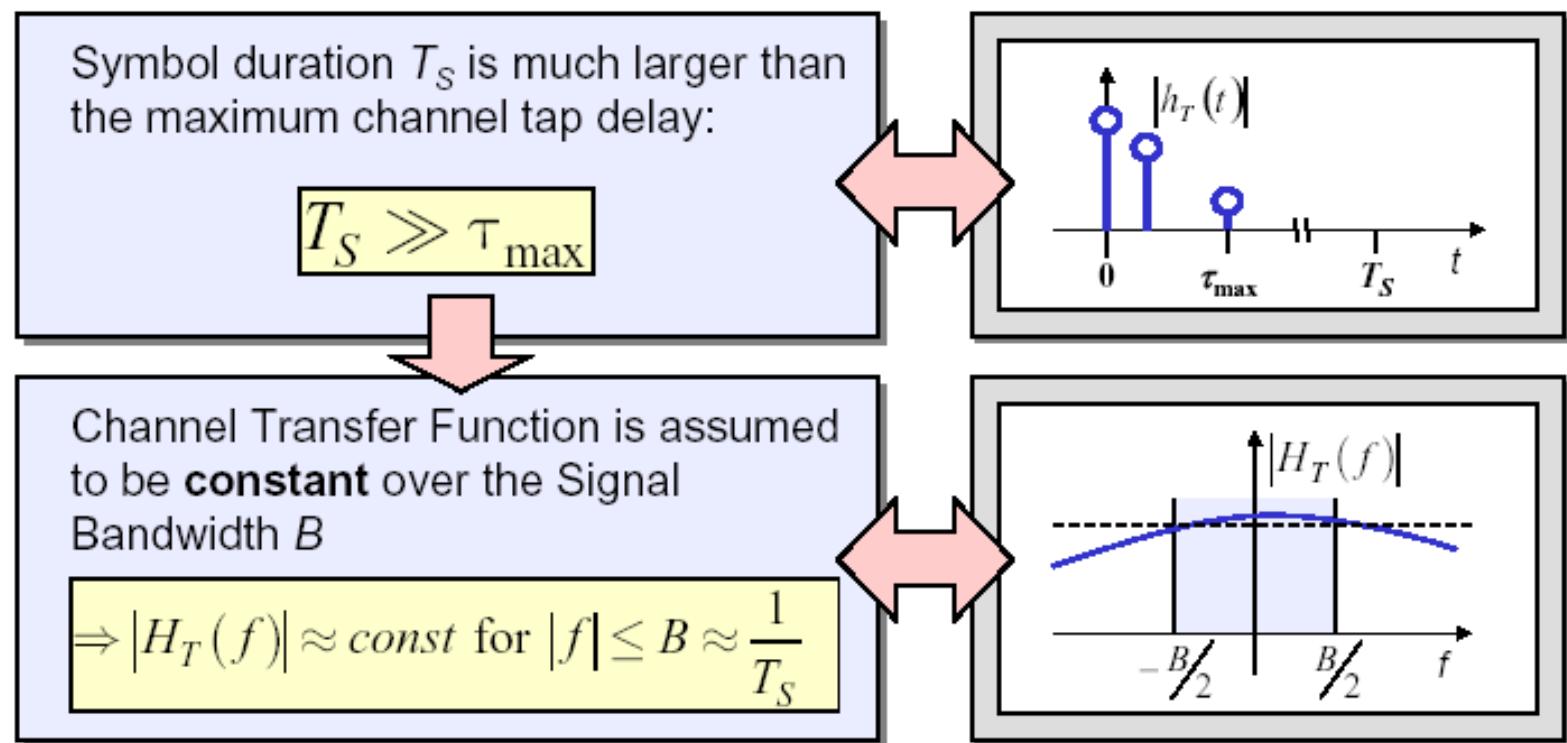
- Frequency selectivity = ISI:



→ The more frequency selectivity, the more ISI.

2. Modeling the mobile radio channel

- At this point, we can distinguish two types of channels:
 - ✦ The narrowband channel:



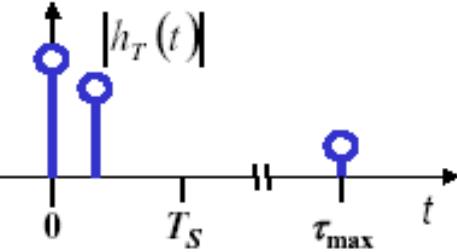
- ➔ Low frequency selectivity and therefore not much ISI.

2. Modeling the mobile radio channel

★ The broadband channel:

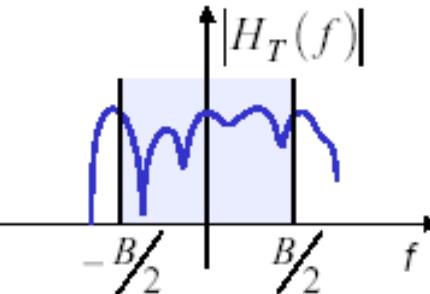
Symbol duration T_S is much smaller than the maximum channel tap delay:

$$T_S \ll \tau_{\max}$$



Channel Transfer Function **fluctuates** over the Signal Bandwidth B

$$\Rightarrow |H_T(f)| \neq \text{const} \text{ for } |f| \leq B \approx \frac{1}{T_S}$$



→ Frequency selectivity, high ISI.

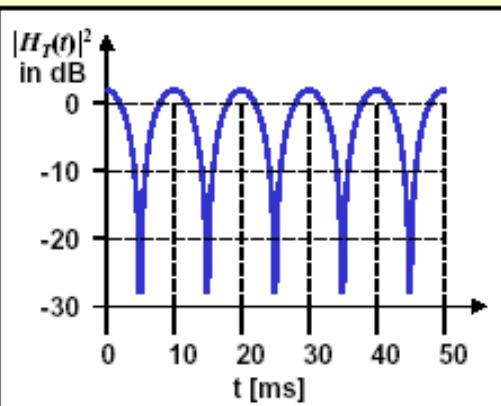
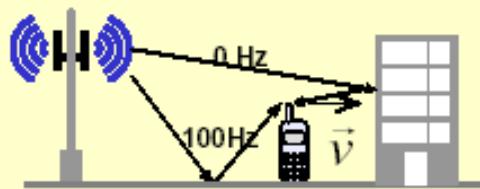
2. Modeling the mobile radio channel

- ❖ Let's consider t (E and/or R are moving):

$$h(\tau, t) = \sum_{n=0}^N \alpha_n(t) e^{-j\phi_n(t)} \delta(\tau - \tau_n(t))$$

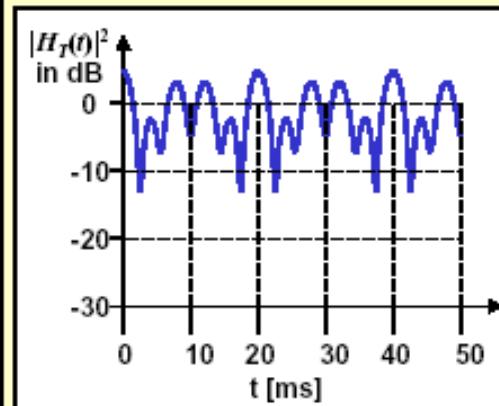
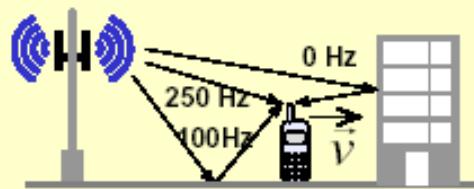
Example 1

Two propagation paths



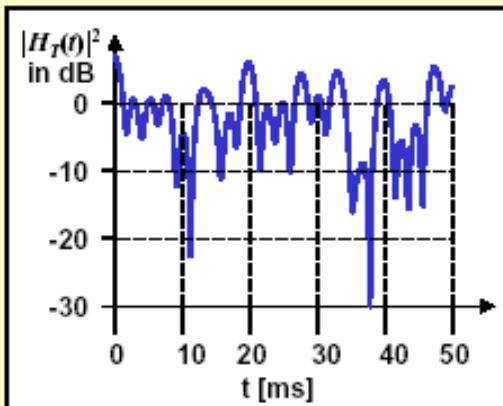
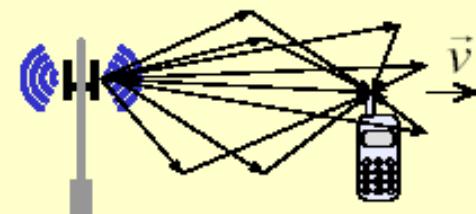
Example 2

Three propagation paths



Example 3

Many propagation paths



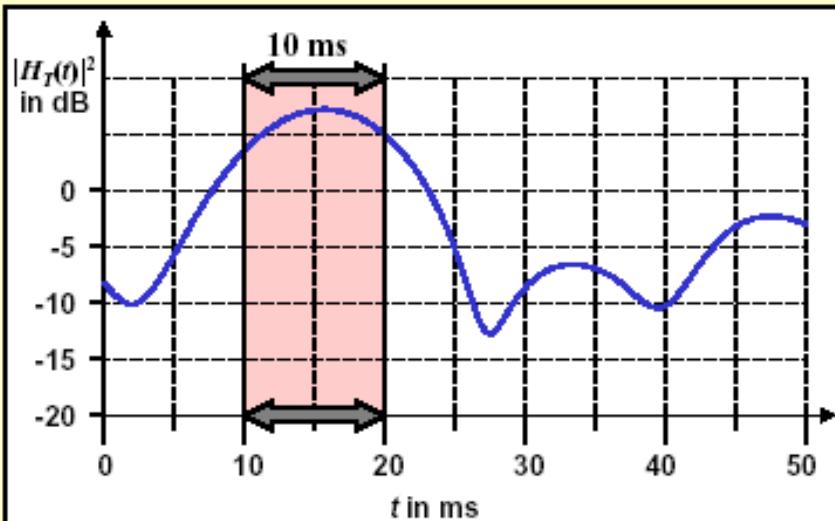
2. Modeling the mobile radio channel

- Influence of the max Doppler frequency:

time-variant narrowband channel

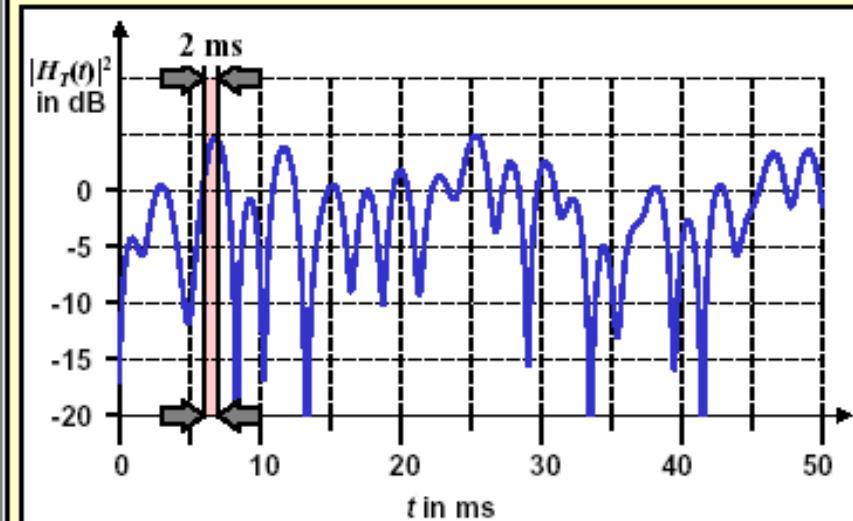
Example 1

$$f_{D,\max} = 100 \text{ Hz} \Rightarrow \frac{1}{f_{D,\max}} = 10 \text{ ms}$$



Example 2

$$f_{D,\max} = 500 \text{ Hz} \Rightarrow \frac{1}{f_{D,\max}} = 2 \text{ ms}$$



Time-variance increases with higher Doppler frequency (=higher mobility)

2. Modeling the mobile radio channel

➤ Summary:

	narrowband channel	broadband channel
time-invariant	<p>Time-invariant narrowband channel</p> <ul style="list-style-type: none">• Frequency non-selective• no Inter Symbol Interference occurs• CTF is a constant: $\Rightarrow H_T(f, t) = \text{const}$	<p>Time-invariant broadband channel</p> <ul style="list-style-type: none">• Frequency-selective• Inter Symbol Interference occurs• CTF is constant over time, but frequency-selective $\Rightarrow H_T(f, t) = H_T(f)$
time-variant	<p>Time-variant narrowband channel</p> <ul style="list-style-type: none">• Frequency non-selective• no Inter Symbol Interference occurs• CTF is constant over the considered bandwidth, but time-variant $\Rightarrow H_T(f, t) = H_T(t)$	<p>Time-variant broadband channel</p> <ul style="list-style-type: none">• Frequency-selective• Inter Symbol Interference occurs• CTF is time-variant and frequency-selective $\Rightarrow H_T(f, t) = H_T(f, t)$

2. Modeling the mobile radio channel

➤ Mobile fading channels modeling

➤ Rayleigh channel:

- ★ The max delay duration << Ts (narrowband)
- ★ The received signal is a superposition of a large number of subpaths WITHOUT a LOS path
- ★ I (real) et Q (imag) received components follow a Gaussian distribution
- ★ In this case, we have:

$$z(t) = |r(t)| = \sqrt{r_I^2(t) + r_Q^2(t)}$$

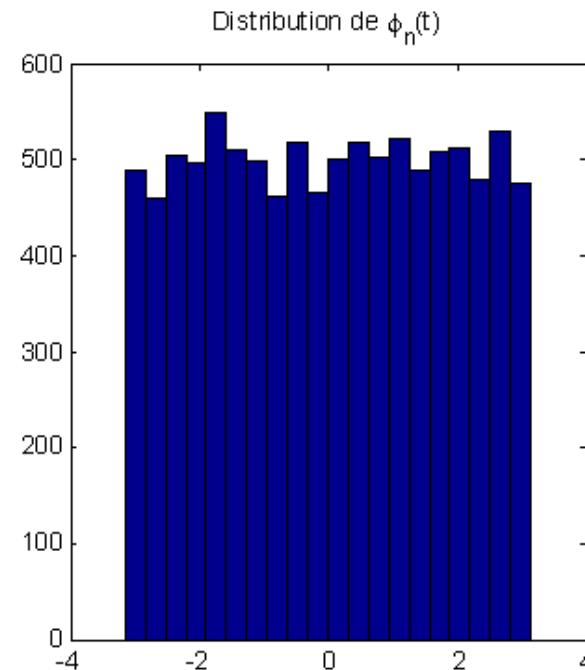
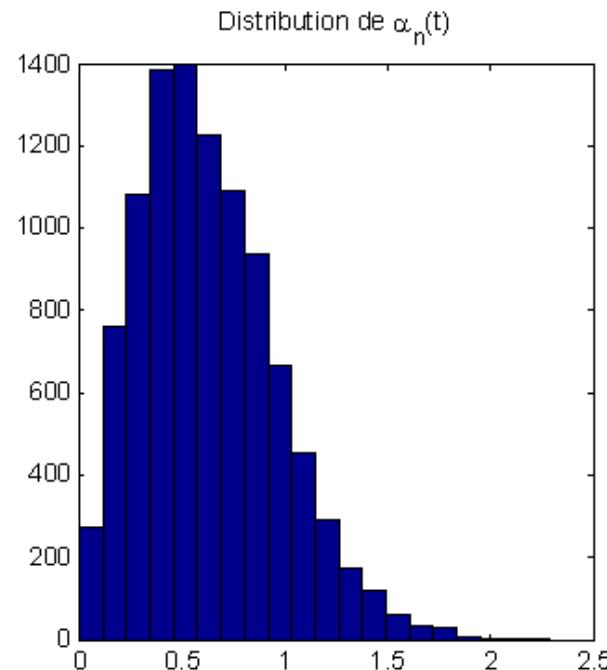
and $z(t)$ follows a Rayleigh distribution:

$$p_z(z) = \frac{2z}{Pr} \exp(-z^2/Pr) = \frac{z}{\sigma^2} \exp(-z^2/(2\sigma^2)), z \geq 0$$

2. Modeling the mobile radio channel

➤ Rayleigh channel (cont'd) :

- ★ $\phi(t)$ the phase of $r(t)$ follows a uniform distribution

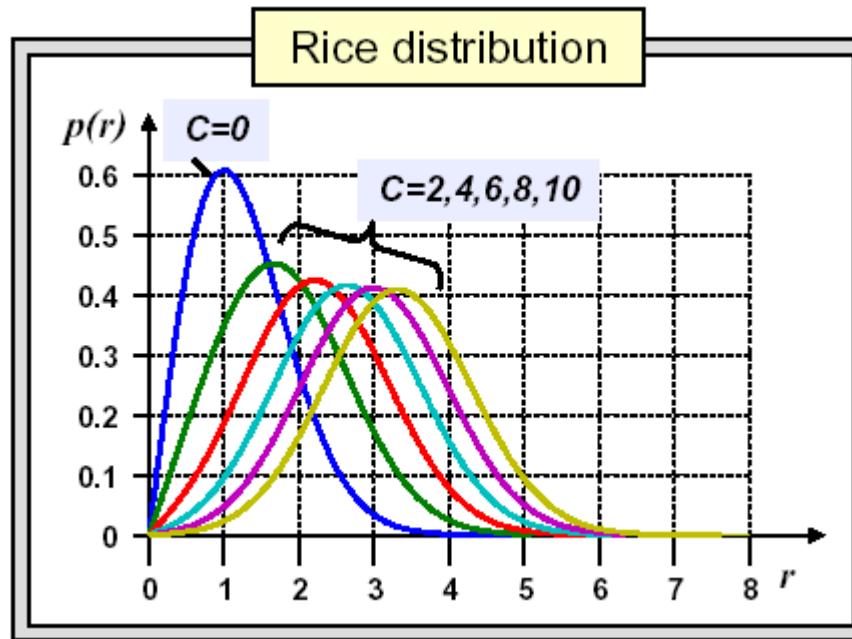


2. Modeling the mobile radio channel

➤ Rice channel:

- ⊕ The received signal is a superposition of many reflected paths and one LOS path
- ⊕ The Rice K factor (or C) is the power of the LOS path s^2 over the power of the NLOS paths:

$$K = \frac{s^2}{2\sigma^2}$$

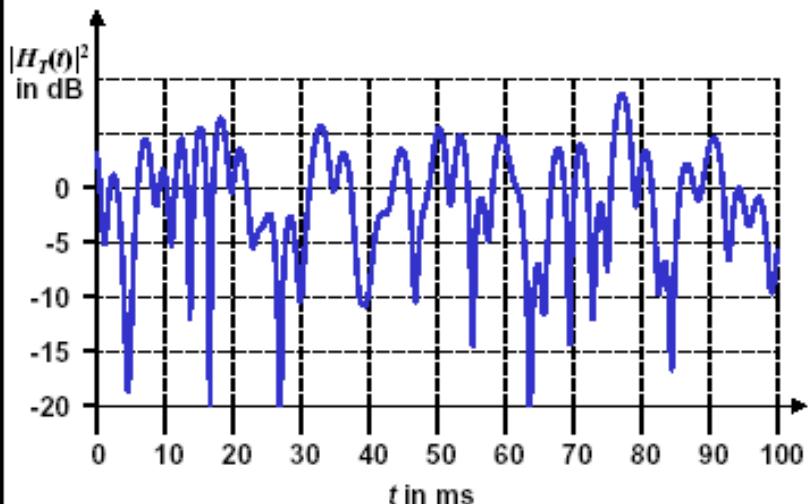


2. Modeling the mobile radio channel

- Rayleigh and Rice comparison:

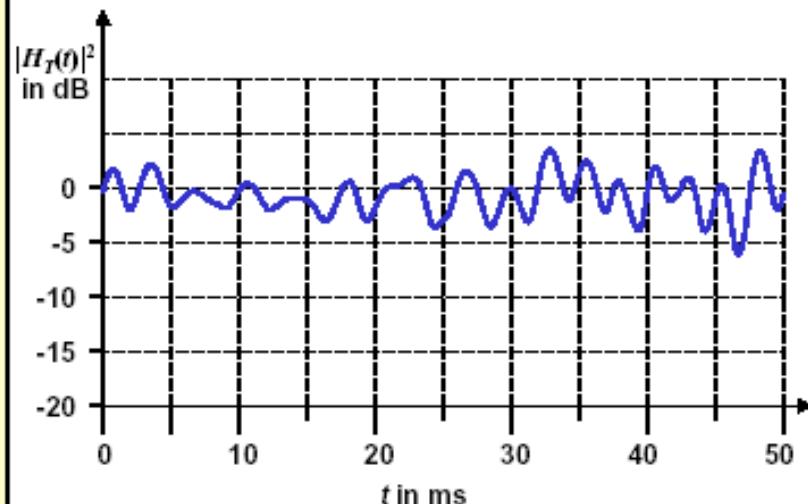
Rayleigh Channel

$$(f_{D,\max} = 200 \text{ Hz})$$



Rice Channel ($C=10 \text{ dB}$)

$$(f_{D,\max} = 200 \text{ Hz})$$



In comparison to a **Rayleigh channel**, the power fluctuations of a **Rice channel** are considerably smaller

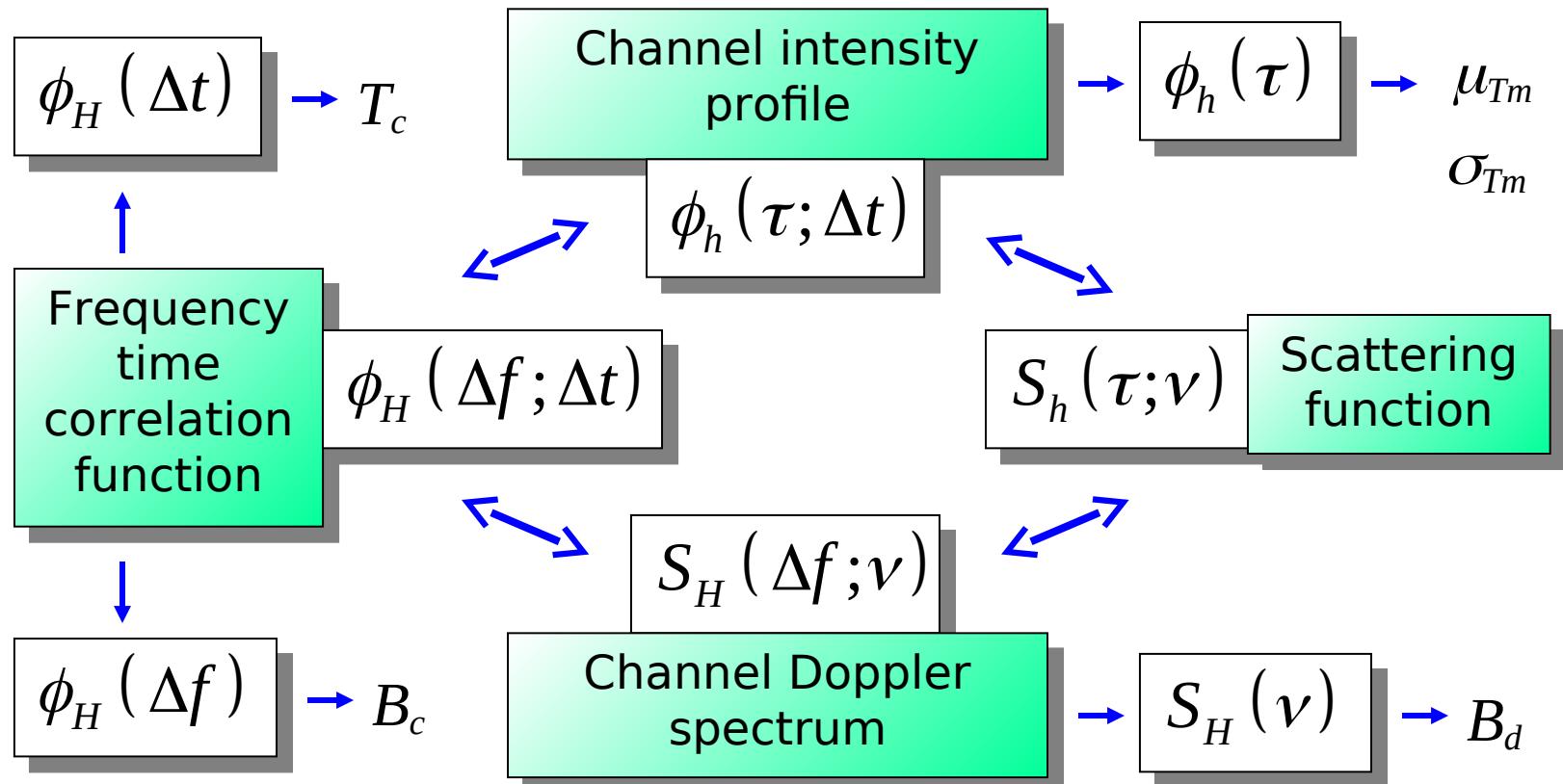
2. Modeling the mobile radio channel

- Statistical channel models and the WSSUS model:
- ◆ The CIR $h(\tau, t)$ is a stochastic process and is characterized by its **autocorrelation function**:
$$\varphi_h(\tau_1, \tau_2; t_1, t_2) = E\{h(\tau_1, t_1) \cdot h^*(\tau_2, t_2)\}$$
 - ◆ In the case of the WSSUS approximation, one suppose that:
 - The stochastic process is Wide Sense Stationnary (WSS), which means the autocorrelation function does not depend on t :
$$\varphi_h(\tau_1, \tau_2; \Delta t) = E\{h(\tau_1, t) \cdot h^*(\tau_2, t + \Delta t)\} \quad \text{with} \quad \Delta t = t_2 - t_1$$
 - The different paths are not correlated (Uncorrelated Scattering US):
$$\varphi_h(\tau_1, \tau_2; \Delta t) = 0 \quad \forall \tau_1 \neq \tau_2$$

$$\varphi_h(\tau; \Delta t) = E\{h(\tau, t) \cdot h^*(\tau, t + \Delta t)\}$$

2. Modeling the mobile radio channel

➤ WSSUS characterization:



2. Modeling the mobile radio channel

- The power delay profile:

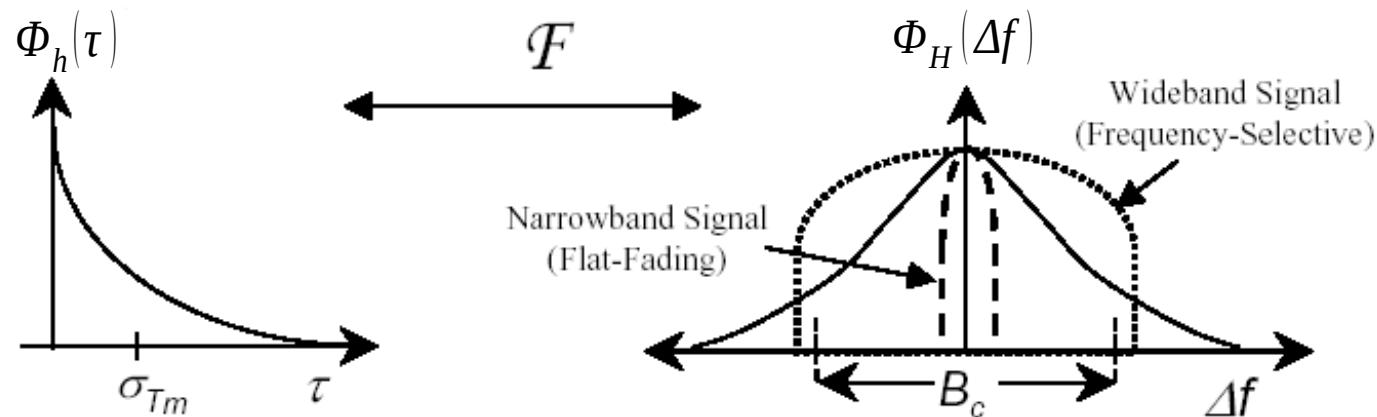
$$\Phi_h(\tau, 0) = \Phi_h(\tau)$$

- ✚ Represents the mean power associated to a path with a given multipath delay. Easily measured empirically.
- ✚ The mean delay spread and the RMS delay spread are defined by:

$$\mu_{T_m} = \frac{\int_0^\infty \tau \cdot \Phi_h(\tau) d\tau}{\int_0^\infty \Phi_h(\tau) d\tau} \quad \sigma_{T_m} = \sqrt{\frac{\int_0^\infty (\tau - \mu_{T_m})^2 \cdot \Phi_h(\tau) d\tau}{\int_0^\infty \Phi_h(\tau) d\tau}}$$

2. Modeling the mobile radio channel

- The coherence bandwidth B_c :



- ★ A commonly used approximation is:

$$B_c \approx 1/\sigma_{Tm}$$

2. Modeling the mobile radio channel

- Doppler spectrum and coherence time:
 - ★ Doppler effect can be characterized by taking the FT of the frequency-time correlation function $\Phi_H(\Delta f, \Delta t)$ relative to Δt .
 - ★ To characterize Doppler at a single frequency, we set $\Delta f = 0$. We then get the power Doppler spectrum (FT of an autocorrelation function):

$$S_H(v) = \int_{-\infty}^{+\infty} \Phi_H(\Delta t) e^{-j2\pi v \Delta t} d\Delta t$$

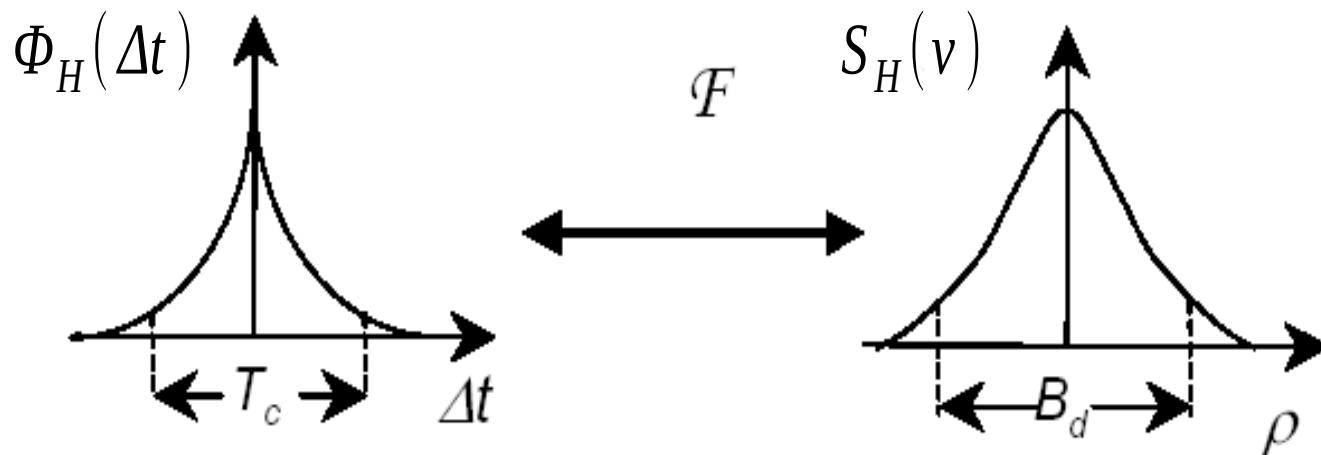
- ★ The maximal value v for which $S_H(v)$ is nonzero is called the **Doppler spread** and is denoted B_d .

2. Modeling the mobile radio channel

➤ Doppler spectrum and coherence time (cont'd):

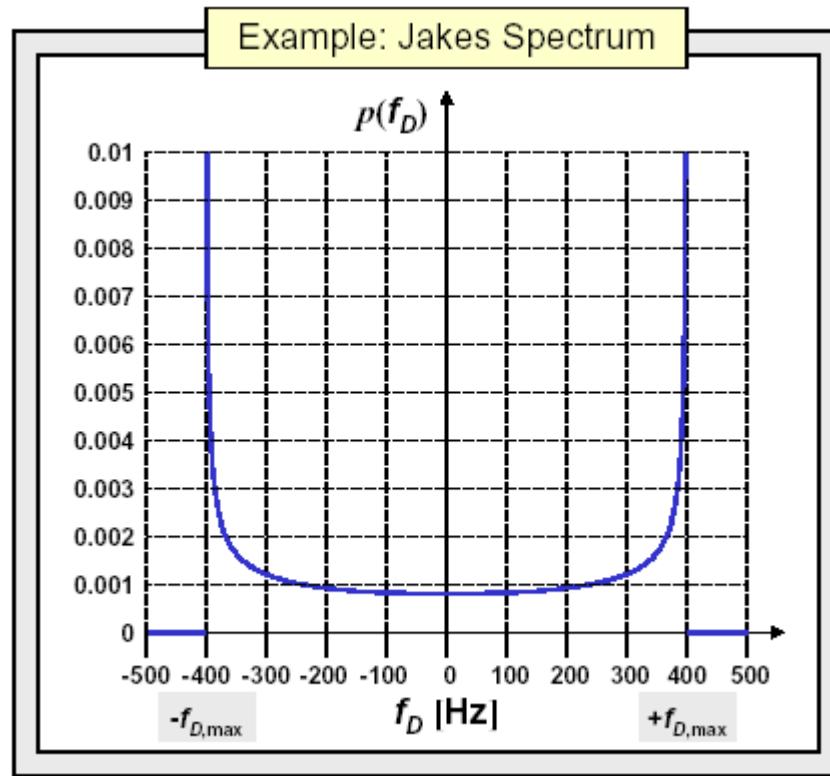
- ★ The time during which $\Phi_H(\Delta t)$ is nonzero, is called the channel ***coherence time*** T_c .
- ★ A commonly used approximation is:

$$B_d \approx 1/T_c$$



2. Modeling the mobile radio channel

- Doppler spectrum and coherence time (cont'd):
 - ★ A common Doppler spectrum shape, the Jakes spectrum:

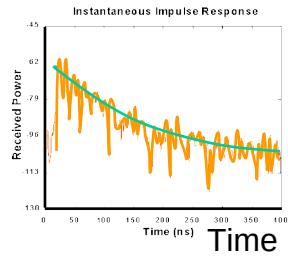


2. Modeling the mobile radio channel

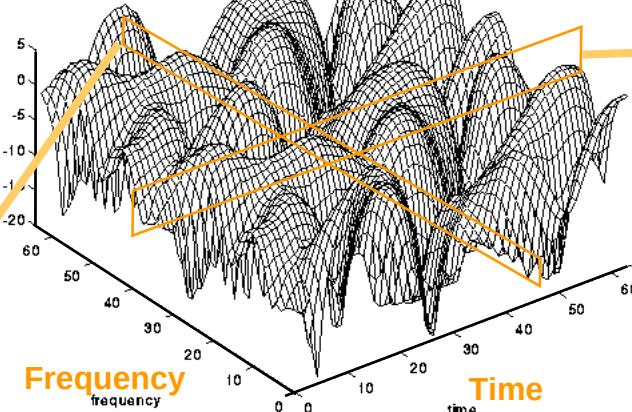
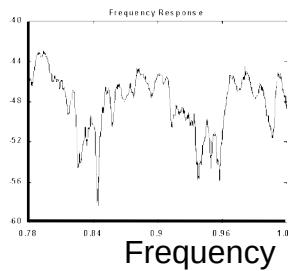
➤ In Summary:

- ★ The mobile radio channel is a *stochastic process*.
- ★ It can be analyzed and modeled using statistical tools.
- ★ It has to be analyzed in the **time and frequency** domains.

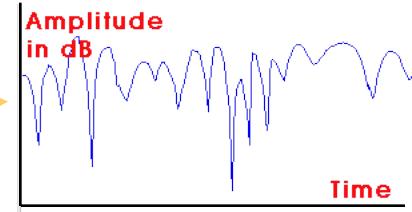
Delay spread



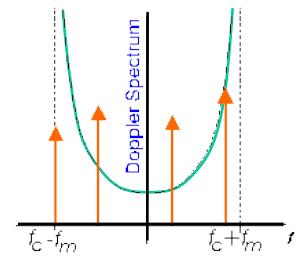
↔ FT



Doppler shift

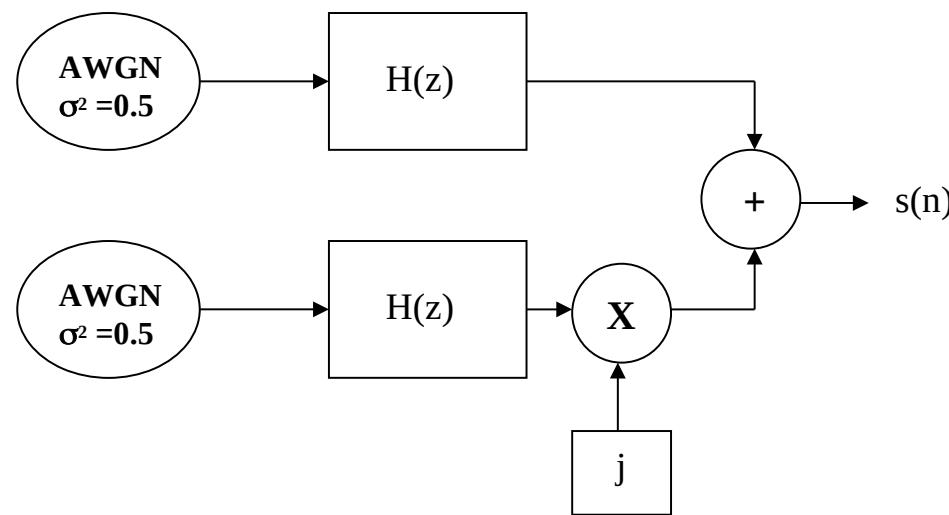


↔ FT



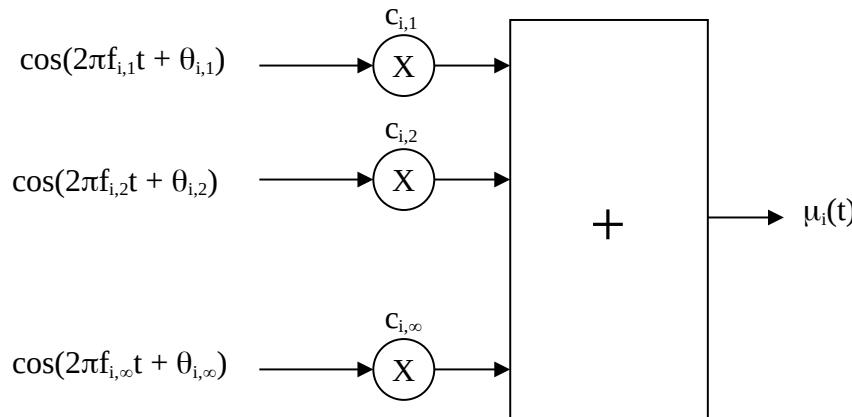
3. Simulation techniques

- Techniques for simulating mobile radio channels:
 - ★ To be able to design efficiently digital communication systems, it is important to have tools to simulate communication channels.
 - ★ There are two main techniques:
 - The filter method:



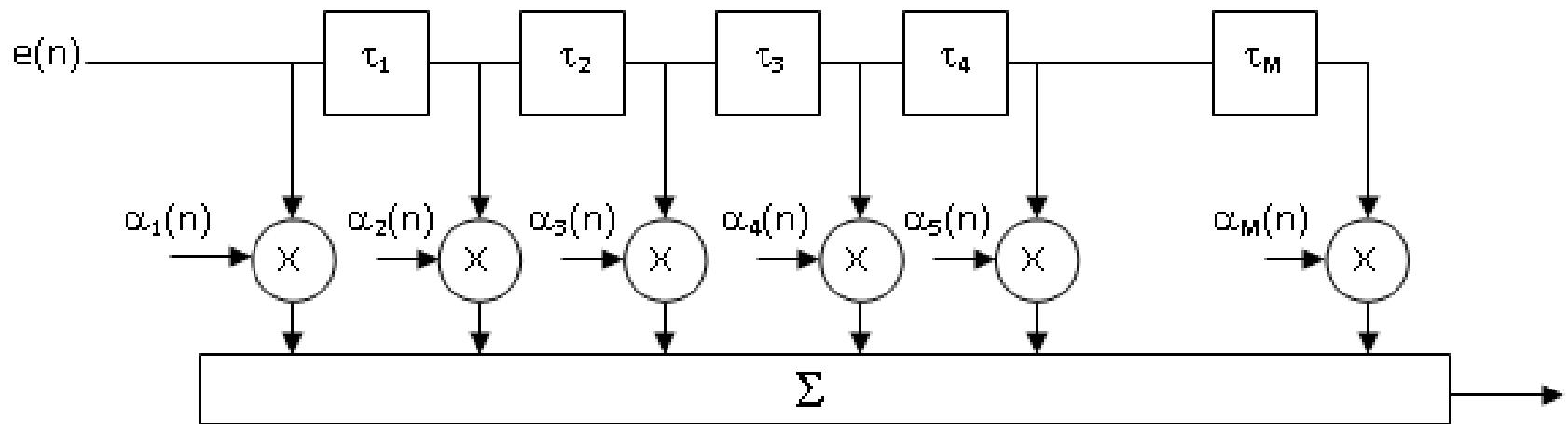
3. Simulation techniques

- Techniques for simulating mobile radio channels (cont'd):
 - ★ The Rice Sum of Sine (SoS) method:



3. Simulation techniques

- Techniques for simulating mobile radio channels (cont'd):
 - ★ More than one path leads to a tapped delay line structure:



3. Simulation techniques

- Normalized mobile radio channels models:
 - ★ COST207: an example of a WSSUS channel model:

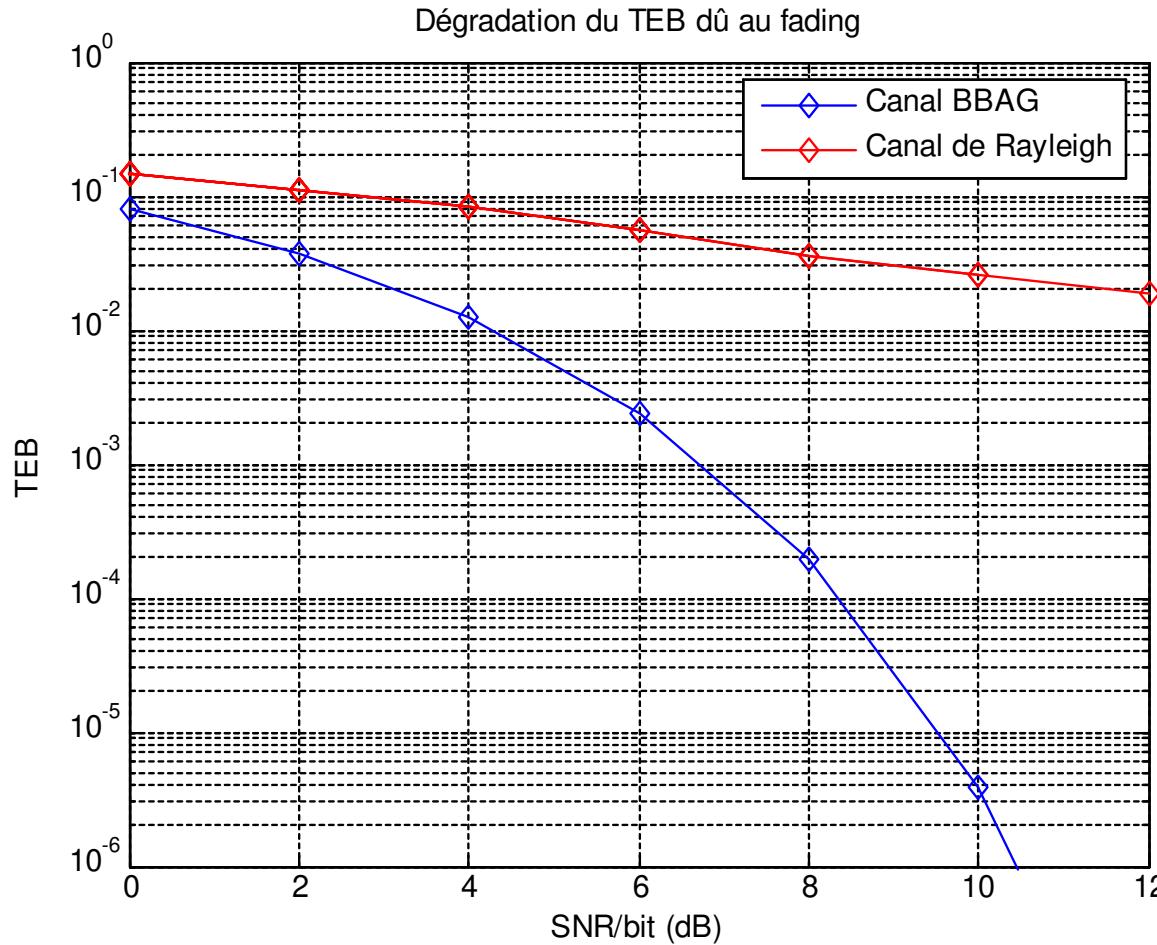
Path no. ℓ	Propagation delay τ'_ℓ	Path power (lin.)	(dB)	Category of the Doppler PSD	Delay spread $B_{\tau'\tau'}^{(2)}$
--------------------	--------------------------------------	-------------------------	------	-----------------------------------	---

(iii) Typical Urban: 6-path channel model (alternative)					
0	$0.0 \mu s$	0.5	-3	“Jakes”	
1	$0.2 \mu s$	1	0	“Jakes”	
2	$0.5 \mu s$	0.63	-2	“Jakes”	
3	$1.6 \mu s$	0.25	-6	“Gauss I”	
4	$2.3 \mu s$	0.16	-8	“Gauss II”	$1.0 \mu s$
5	$5.0 \mu s$	0.1	-10	“Gauss II”	

- ★ Modern signal generators integrate these models.

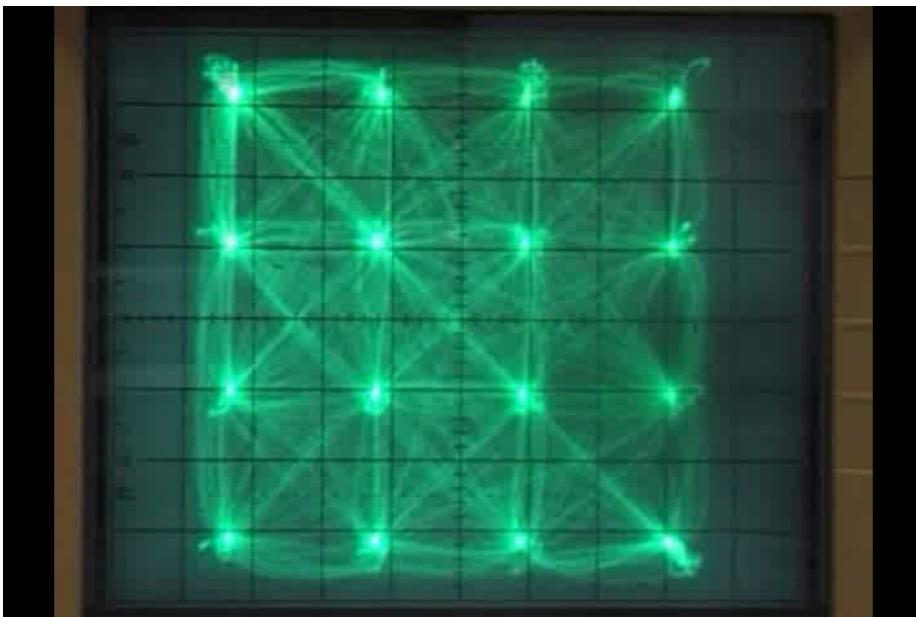
3. Simulation techniques

➤ BER degradation (Rayleigh channel):

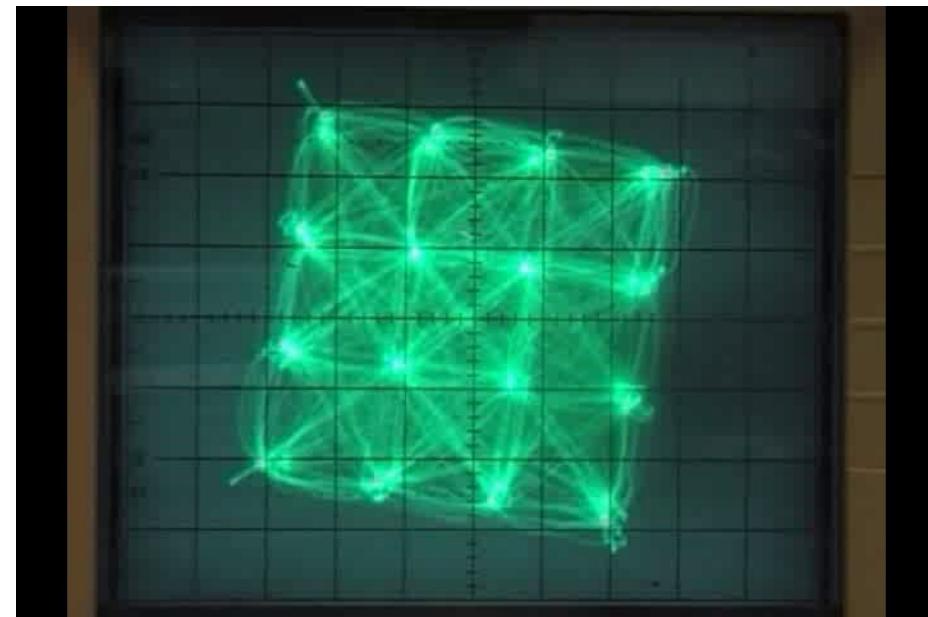


3. Simulation techniques

- Illustration of the BER degradation:



16QAM AWGN channel



16QAM Rayleigh channel

3. Simulation techniques

➤ Demonstration:

- ✦ Generation of Rayleigh channel samples using the filter method + BER curve (Octave).
- ✦ COST207TU channel model and the resolvable paths concept (Matlab).
- ✦ Digital communication system with Rayleigh channel (GNU Radio).
- ✦ Digital communication system with COST207TU frequency selective fading channel (GNU Radio).
⇒ Files available on my github :

https://github.com/hboeglen/Mobile_Fading_Channel_Tutorial_SDRA_2022

4. Conclusion

- The mobile fading channel is the real world!
- Worst case: time and frequency selectivity.
- The expected BER is very high
- ➔ Need for sophisticated digital communication techniques (e.g. OFDM, interleaving, spread spectrum etc.).

Bibliography

- Books:
 - Digital Communications: A Discrete-Time Approach, M. Rice, Prentice Hall, 2008.
 - Wireless Communications, A. Goldsmith, Cambridge University Press, 2005.
 - Principles of Mobile Communication, G. Stuber, Springer, 2011.
 - Mobile radio channels, M. Pätzold, Wiley, 2011.