Science / Engineering / Signals and Systems (2nd Edition)

Exercise 16

Chapter 3, Page 253





Signals and Systems

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(a) Given, $x_1[n] = (-1)^n$, it is clear that $x_1[n]$ is a periodic signal with period N=2. Therefore, from **Section 3.6.1**, if x[n] is a periodic signal with period N, then

$$x[n] = \sum_{k=< N>} a_k e^{jk(2\pi/N)n}$$

and k=0,1,....N-1, which implies,

$$egin{array}{lll} x_1[n] &=& & (-1)^n \ &=& \displaystyle\sum_{k=<2>} a_k e^{jk(2\pi/2)n} & (k=0,1) \ &=& a_0(1) + a_1 e^{j\pi n} \ x_1[n] &=& a_0 + a_1(-1)^n & (\because e^{j\pi} = -1) \end{array}$$

Clearly, from eq(2) and eq(3), $a_0=0$ and $a_1=1$. Now form **Section 3.8**, if x[n] is a periodic signal with period N and is passed through a LTI system with frequency response $H(e^{j\omega})$, then the output y[n] is given by,

$$egin{align} y[n] &= \sum_{k=< N>} a_k H(e^{j\omega}) e^{jk(2\pi/N)n} \ y[n] &= \sum_{k=< N>} a_k H(e^{j2\pi/N}) e^{jk(2\pi/N)n} \ \end{array}$$

Therefore, substitute the results of eq(3) in eq(5),

$$egin{array}{ll} y_1[n] = & 0 + (1) H(e^{j2\pi/2}) e^{j(1)(2\pi/2)n} \ y_1[n] = & H(e^{j\pi}) e^{j\pi n} \end{array}$$

From eq(7), it is clear that $\omega=\pi$, and from the frequency response of $H(e^{j\omega})$ which is shown in **Figure P3.16**, $H(e^{j\omega})=0$, Therefore,

$$y_1[n]=0$$

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 ${f (b)}$ Given, $x_2[n]=1+\sin\left(rac{3\pi}{4}n+rac{\pi}{4}
ight)$, Therefore,

$$egin{aligned} x_2[n+N] &=& 1+\sinig(rac{3\pi}{4}(n+N)+rac{\pi}{4}ig) \ x_2[n] &=& 1+\sinig(rac{3\pi}{4}n+rac{3\pi}{4}N+rac{\pi}{4}ig) \end{aligned}$$

Therefore, $\frac{3\pi}{4}N$ should be a multiple of 2π , i.e., $\frac{3\pi}{4}N=2\pi m$. Therefore, N=16 is the least possible value of N at a value of m=3. Now $x_2[n]$ can be written as,

$$egin{array}{lll} x_2[n] &=& e^{j(2\pi/16)(0)n} + rac{1}{2j}ig(e^{j(2\pi/16)(3)n}e^{j\pi/4} - e^{-j(2\pi/16)(3)n}e^{-j\pi/4}ig) \ x_2[n] &=& e^{j(2\pi/16)(0)n} - rac{j}{2}ig(e^{j(2\pi/16)(3)n}e^{j\pi/4} - e^{-j(2\pi/16)(3)n}e^{-j\pi/4}ig) \ x_2[n] &=& e^{j(2\pi/16)(0)n} - rac{j}{2}ig(e^{j(2\pi/16)(3)n}e^{j\pi/4} - e^{j(2\pi/16)(13)n}e^{-j\pi/4}ig) \end{array}$$

In the above equation k=-3 is changed to k=13 by using the properties of exponential function. Now from **Section 3.6.1**, the non-zero Fourier series coefficients of $x_2[n]$ which are in the range $0 \le k \le 15$ are,

$$a_0=1,\quad a_3=-rac{je^{j\pi/4}}{2},\quad a_{13}=rac{je^{-j\pi/4}}{2}$$

Now form **Section 3.8**,

$$egin{align} y_2[n] &= \sum_{k=0}^{15} a_k H(e^{j2\pi k/16}) e^{jk(2\pi/16)n} \ y_2[n] &= a_0 H(0) + a_3 H(e^{j3\pi/8}) e^{j3\pi n/8} + a_{13} H(e^{j13\pi/8}) e^{j13\pi n/8} \ y_2[n] &= 0 - rac{j e^{j\pi/4}}{2} H(e^{j3\pi/8}) e^{j3\pi n/8} + rac{j e^{-j\pi/4}}{2} H(e^{j13\pi/8}) e^{j13\pi n/8} \ \end{pmatrix}$$

From eq(3), it is clear that $\omega_1=3\pi/8$ and $\omega_2=13\pi/8$, and from the frequency response of $H(e^{j\omega})$ which is shown in **Figure P3.16**, $H(e^{j\omega_1})=1=H(e^{j\omega_2})$, Therefore,

$$egin{array}{ll} y_2[n] = & -rac{je^{j\pi/4}}{2}e^{j3\pi n/8} + rac{je^{-j\pi/4}}{2}e^{j13\pi n/8} \ y_2[n] = & -rac{je^{j\pi/4}}{2}e^{j3\pi n/8} + rac{je^{-j\pi/4}}{2}e^{-j3\pi n/8} \ & -rac{3\pi}{2}e^{-j3\pi n/8} \end{array}$$

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$$y_2[n] = \sin\left(\frac{sn}{8}n + \frac{n}{4}\right)$$

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$$(\mathbf{c})$$
 Given, $x_3[n] = \sum_{k=-\infty}^{\infty} \left(rac{1}{2}
ight)^{n-4k} u[n-4k]$, Therefore,

$$egin{align} x_3[n] = & \left[\left(rac{1}{2}
ight)^n u[n]
ight] * \sum_{k=-\infty}^\infty \delta[n-4k] \ x_3[n] = & d[n] * b[n] \ \end{cases}$$

where, $d[n]=\binom{1}{2}^nu[n]$ and $b[n]=\sum_{k=-\infty}^\infty\delta[n-4k]$. therefore, the output of the filter $y_3[n]$ can be obtained by first passing the signal b[n] through the filter with the filter response $H(e^{j\omega})$, and them convolving the result with d[n].

From the signal b[n], it is clear that the period of the signal is N=4 Now from ${\bf Section~3.6.1}$, the non-zero Fourier series coefficients of $x_3[n]$ which are in the range $0 \le k \le 3$ are,

$$egin{align} a_k = & rac{1}{4} \sum_{n=0}^3 b[n] e^{-jk(2\pi/4)n} \ a_k = & rac{1}{4} \sum_{n=0}^3 \sum_{k=-\infty}^\infty \delta[n-4k] e^{-jk(2\pi/4)n} \ a_k = & rac{1}{4} \sum_{k=-\infty}^\infty \sum_{n=0}^3 \delta[n-4k] e^{-jk(2\pi/4)n} \ a_k = & rac{1}{4}, & ext{for all } k & (\because \delta[n] = 0 & orall n
otag = 0. \end{align}$$

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Now form **Section 3.8**,

$$egin{align} r[n] &= \sum_{k=0}^{3} a_k H(e^{j2\pi k/4}) e^{jk(2\pi/4)n} \ r[n] &= rac{1}{4} igl[H(e^{j0}) e^{j0} + H(e^{j2\pi/4}) e^{j(2\pi/4)n} + \ H(e^{j4\pi/4}) e^{j2(2\pi/4)n} + H(e^{j6\pi/4}) e^{j3(2\pi/4)n} igr] \ r[n] &= rac{1}{4} igl[H(e^{j0}) e^{j0} + H(e^{j\pi/2}) e^{j(\pi/2)n} + \ H(e^{j\pi}) e^{j(\pi)n} + H(e^{j3\pi/2}) e^{j3(\pi/2)n} igr] \end{aligned}$$

From the above equation, it is clear that $\omega_1=0$, $\omega_2=\pi/2$, $\omega_3=\pi$ and $\omega_4=3\pi/2$ and from the frequency response of $H(e^{j\omega})$ which is shown in **Figure P3.16**, $H(e^{j\omega_1})=H(e^{j\omega_2})=H(e^{j\omega_3})=H(e^{j\omega_4})=0$, which implies,

$$r[n] = 0$$

Therefore, the final output,

$$egin{array}{lll} y_3[n] &=& d[n]*(b[n]*h[n]) \ y_3[n] &=& d[n]*r[n] \ y_3[n] &=& d[n]*0 \ y_3[n] &=& 0 \end{array}$$

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(a)
$$y_1[n]=0$$

(b) $y_2[n]=\sin\left(\frac{3\pi}{8}n+\frac{\pi}{4}\right)$
(c) $y_3[n]=0$

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