

# **Digital Integrated Circuit**

## **EC4202-01**

### **HW#3**

**20194044**

**ChangJoo Park (박창주)**



Digital Integrated Circuit  
Homework assignment #3

20194044 Park ChangJoo (박창주)

(Ex 2-1)

$$\left\{ \begin{array}{l} h = \mu C_{ox} \frac{V}{L} = (350 \frac{C_{ox}}{\mu s}) \left[ \frac{(3.9)(8.854 \times 10^{-14} (\text{F/cm}^2))}{100 \cdot 10^{-10} (\text{cm})} \right] \\ = (120.86 \mu) (4/2) = 241.72 \mu \text{A/V}^2 \\ V_{th} = 0.7 \text{V} \end{array} \right.$$

$$I_{DS} = \begin{cases} 0 & V_{GS} < V_{th} \\ k(V_{GS} - V_{th} - \frac{V_D}{2}) V_{DS}, & V_{GS} > V_{th} \& V_{DS} < V_{DSAT} = V_{GS} - V_{th} \\ \frac{1}{2} k (V_{GS} - V_{th})^2 & , V_{GS} > V_{th} \& V_{DS} > V_{DSAT} = V_{GS} - V_{th} \end{cases}$$

\* ex.2.1 for SPICE3F5

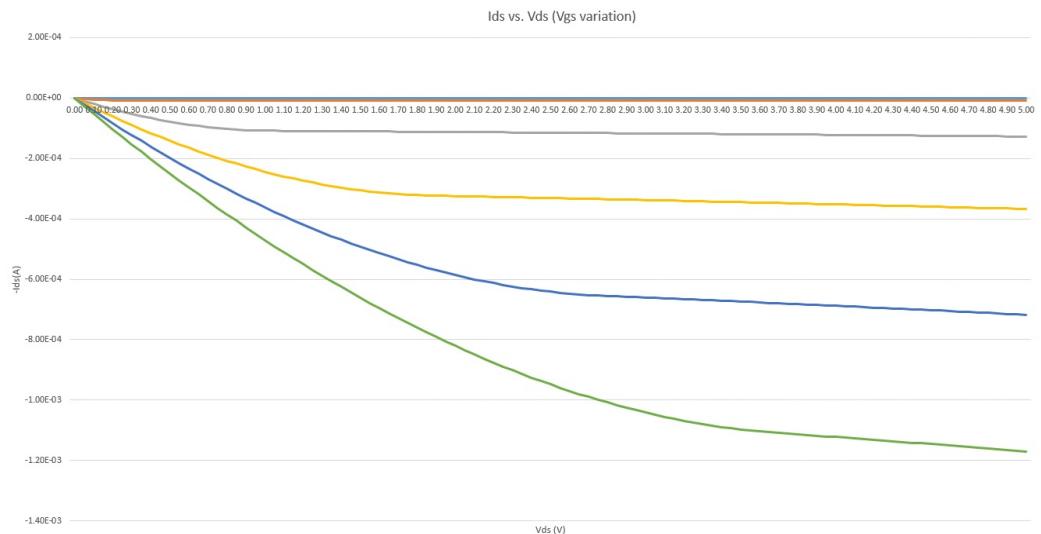
```
.MODEL NMOS NMOS LEVEL=2 LD=0.6U TOX=100.0E-10
+ NSUB=5.36726E+15 VT0=0.7 KP=1.208571E-04 GAMMA=0.543
+ PHI=0.6 U0=350 UEXP=0.157282 UCRT=31443.8
+ DELTA=2.39824 VMAX=55260.9 XJ=0.25U LAMBDA=0.0367072
+ NFS=1E+12 NEFF=1.001 NSS=1E+11 TPG=1.0 RSH=70.00
+ CGDO=4.3E-10 CGSO=4.3E-10 CJ=0.0003 MJ=0.6585
+ CJSW=8.0E-10 MJSW=0.2402 PB=0.58
```

```
Vgs g 0 0
Vds d 0 0
M1 d g 0 0 NMOS W=2.4 L=1.2

.dc Vds 0 5.0 0.05 Vgs 0 5.0 1.0
.print dc V(g) I(Vds)
.end
```

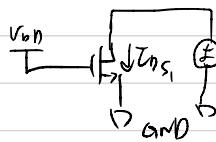
(1)  $I_{DS}$  for  $V_{GS}=1V$  is around  $\mu\text{A}$  unit.  
Although it is turned on, cannot be easily distinguished from  $V_{GS}=0\text{V}$  one.

(2) Although  $I_{DS}$  for NMOS is positive, SPICE3F5 returns it as negative quantity.



Ex 2.2

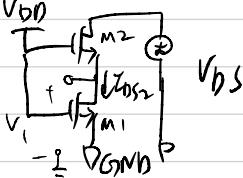
(a)



$$I_{DS,1} = \frac{1}{2} k \left( V_{GS,1} - V_{th} - \frac{V_{DS}}{2} \right) V_{DS}$$

$$= \frac{1}{2} k \left( V_{DD} - V_{th} - \frac{V_{DS}}{2} \right) V_{DS}$$

(b)



$$\begin{cases} V_{GS,1} = V_{DD} - V_1 \\ V_{DS,1} = V_{DS} - V_1 \end{cases} \quad \begin{cases} V_{GS,2} = V_{DD} \\ V_{DS,2} = V_{DS} - V_1 \end{cases}$$

$$I_{DS,2} = I_{DS,1} = k \left( V_{GS,1} - V_{th} - \frac{1}{2} V_{DS,1} \right) V_{DS,1}$$

$$= k \left( V_{DD} - V_{th} - \frac{V_1}{2} \right) V_1$$

$$= I_{DS,1} = k \left( V_{GS,2} - V_{th} - \frac{1}{2} V_{DS,2} \right) V_{DS,2}$$

$$= k \left( V_{DD} - V_{th} - \frac{V_1}{2} - \frac{V_{DS}-V_1}{2} \right) (V_{DS}-V_1)$$

$$\left( V_{DD} - V_{th} - \frac{V_1}{2} \right) V_1 = \left( V_{DD} - V_{th} - \frac{V_1}{2} - \frac{V_{DS}}{2} \right) (V_{DS}-V_1)$$

$$-\frac{V_1^2}{2} + (V_{DD}-V_{th})V_1 = \frac{V_1^2}{2} - V_1 \left( V_{DD} - V_{th} - \frac{V_{DS}}{2} + \frac{V_{DS}}{2} \right) + (V_{DD}V_{DS} - V_{th}V_{DS} - \frac{V_{DS}^2}{2})$$

$$V_1^2 - 2V_1(V_{DD}-V_{th}) + V_{DS}(V_{DD}-V_{th}-\frac{V_{DS}}{2}) = 0$$

$$V_1 = \frac{2(V_{DD}-V_{th}) \pm \sqrt{4(V_{DD}-V_{th})^2 - 4(V_{DD}-V_{th}-\frac{V_{DS}}{2})V_{DS}}}{2}$$

$$= V_{DD}-V_{th} - \sqrt{(V_{DD}-V_{th})^2 - (V_{DD}-V_{th}-\frac{V_{DS}}{2})V_{DS}} \quad (\because \text{triode! } V_1 < V_{DD}-V_{th})$$

$$= A - B \quad \left\{ \begin{array}{l} A = V_{DD}-V_{th} \\ B = \sqrt{(V_{DD}-V_{th})^2 - (V_{DD}-V_{th}-\frac{V_{DS}}{2})V_{DS}} \end{array} \right.$$

$$(i) I_{DS,1} = \frac{1}{2} k \left( V_{DD} - V_{th} - \frac{V_{DS}}{2} \right) V_{DS}$$

$$(ii) I_{DS,2} = k \left( V_{DD} - V_{th} - \frac{V_1}{2} \right) V_1$$

$$= k \left( A - \frac{V_1}{2} \right) V_1$$

$$= k \left( A - \frac{A-B}{2} \right) (A-B)$$

$$= \frac{k}{2} (A+B)(A-B) = \frac{k}{2} (A^2 - B^2)$$

$$= \frac{k}{2} \left[ (V_{DD}-V_{th})^2 - (V_{DD}-V_{th})^2 + \left( V_{DD}-V_{th}-\frac{V_{DS}}{2} \right) V_{DS} \right]$$

$$= \frac{k}{2} \left( V_{DD} - V_{th} - \frac{V_{DS}}{2} \right) V_{DS} = I_{DS,2}$$

$$\boxed{\therefore I_{DS,1} = I_{DS,2}}$$

Ex 2.3  
 When  $V_{sb} > 0$ , body effect is applied, thus,  $V_{th}$  increases.  
 Increased  $V_{th}$  will decrease drain current in top MOSFET  
 since drain current is proportional to " $V_{ds} = V_{gs} - V_{th}$ ".

$$\therefore I_{DS1} > I_{DS2}$$

Ex 2.4  $C_g = C_{ox}WL = \epsilon_{permittivity} \cdot W$

$$\epsilon_{permittivity} = C_{ox} \cdot L = \frac{\epsilon_{ox}}{t_{ox}} \cdot L \quad \left\{ \begin{array}{l} t_{ox} = 16 \text{ Å} = 16 \times 10^{-8} \text{ cm} \\ L = 90 \text{ nm} = 9 \times 10^{-6} \text{ cm} \end{array} \right.$$

$$= \frac{(3.9)(8.854 \times 10^{-12})}{16 \times 10^{-8}} \cdot (9 \times 10^{-6}) = 19.42 \text{ pF/cm}$$

$$= \frac{19.42 \text{ pF}}{\text{cm}} \cdot \frac{1 \text{ cm}}{10^4 \mu\text{m}} = 1.942 \text{ fF}/\mu\text{m}$$

∴  $\epsilon_{permittivity} = 1.942 \text{ fF}/\mu\text{m}$

Ex 2.5  $\left\{ \begin{array}{l} C_J = 0.42 \text{ fF}/\mu\text{m}^2 \\ M_J = 0.44 \\ C_J S_W = 0.38 \text{ fF}/\mu\text{m} \\ M_J S_W = 0.12 \\ \gamma_0 = 0.98 \end{array} \right.$

$$\left\{ \begin{array}{l} S_{min, diff capacitance} = 4 \times 5 \lambda \\ = 1.2 \times 15 = 1.8 \mu\text{m}^2 \\ P_{min, diff capacitance} = 1.2 \times 2 + 1.5 \times 2 = 5.4 \mu\text{m} \end{array} \right.$$

$$(i) C_{db}(0V) = S_{min, diff capacitance} \cdot C_J + P_{min, diff capacitance} \cdot C_J S_W \\ = (1.8)(0.42) + (5.4)(0.33) = 2.538 \text{ fF}$$

$$(ii) C_{db}(5V) = S_{min, diff capacitance} \cdot C_J \cdot \left(1 + \frac{V_{SB}}{\gamma_0}\right)^{-M_J} \\ + P_{min, diff capacitance} \cdot C_J S_W \left(1 + \frac{V_{SB}}{\gamma_0}\right)^{-M_S_W} \\ = (1.8)(0.42) \left(1 + \frac{5}{0.98}\right)^{-0.44} + (5.4)(0.33) \left(1 + \frac{5}{0.98}\right)^{-0.12} \\ \approx 1.775 \text{ fF}$$

$\left\{ \begin{array}{l} C_{db}(0V) = 2.538 \text{ fF} \\ C_{db}(5V) \approx 1.775 \text{ fF} \end{array} \right.$

\* channel potential versus channel position.

$$\left\{ \begin{array}{l} V_g = 1.4V \\ V_{th} = 0.4V \\ \lambda_c = 180 \text{ nm} \end{array} \right.$$

$$\left\{ \begin{array}{l} V(x) = V_g - V_{th} - \sqrt{(V_g - V_{th})^2 - \frac{2I_0}{\mu_n C_ox W} x} \\ I_b = \frac{1}{2} \mu_n C_ox \left( \frac{W}{L} \right) \left[ (V_g - V_{th}) V_{DS} - \frac{1}{2} V_{DS}^2 \right] \quad (V_{DS} < V_g - V_{th}) \\ I_D = \frac{1}{2} \mu_n C_ox \left( \frac{W}{L} \right) (V_{DS} - V_{th})^2 \quad (V_{DS} \geq V_g - V_{th}) \end{array} \right.$$

$$(i) V(x) = V_g - V_{th} - \sqrt{(V_g - V_{th})^2 - \frac{x}{L} \left[ (V_{DS} - V_{th}) - \frac{1}{2} V_{DS} \right] V_{DS}}$$

$$= 1 - \sqrt{1 - \frac{x}{180} \left[ 1 - \frac{1}{2} V_{DS} \right] V_{DS}} \quad (V_{DS} < V_g - V_{th})$$

$$(ii) V(x) = V_g - V_{th} - \sqrt{(V_g - V_{th})^2 - \frac{x}{L} (V_{DS} - V_{th})^2} = 1 - \sqrt{1 - \frac{x}{180} (V_{DS} - V_{th})} \quad (V_{DS} > V_g - V_{th})$$

```

clear all;
close all;

figure(1);

for Vds=0:0.1:1.4
    table(:,1)=1:1:180;

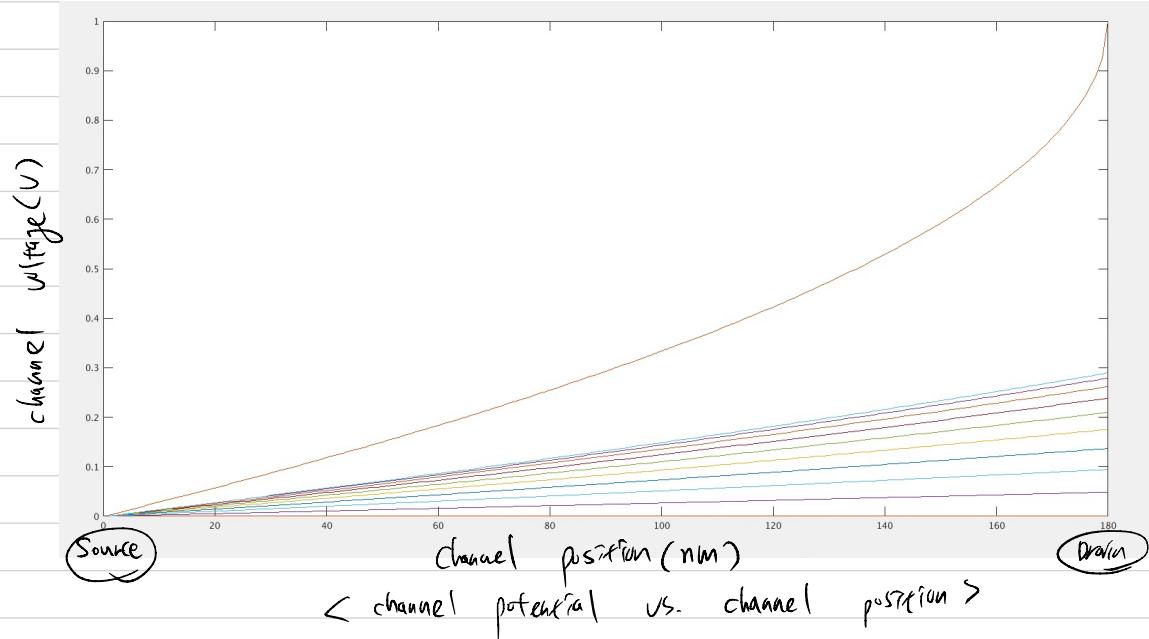
    for x=1:1:180
        if(Vds >= 1.0)
            table(x,2)=1-sqrt(1-x/180);
        else
            table(x,2)=1-sqrt(1-(x/180)*(1-Vds/2)*Vds);
        end
    end

    plot(table);
    hold on;

end
ylim(1);

```

<matlab code for plotting graph>



- (1) As channel position value ( $x$ ) increases to drain node, the channel voltage increases.
- (2) As  $V_{DS}$  increases, channel potential increases for triode region.
- (3) Without channel length modulation effect, channel potential is all same for different  $V_{DS}$  value. The potential value itself is higher than the ones in triode regions.