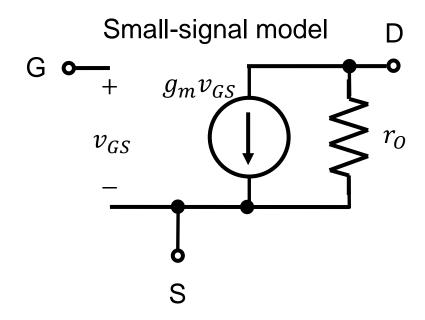
# Lecture15: Impedance

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#### Review of previous lecture

Small-signal model



# Example 6.14 (Razavi)

- Assume that  $\mu_n C_{ox} = 100 \, \mu \text{A V}^{-2}$  and  $\frac{W}{L} = 10$ .
  - When the drain bias current is 0.5 mA, the gate overdrive voltage,  $V_{GS} V_{TH}$ , is 1 V.
  - Then,

$$g_m = 1 \, \mathrm{S}$$

- Channel-length modulation
  - With the channel-length modulation coefficient,  $\lambda$ ,

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})^2 (1 + \lambda V_{DS})$$

- Then,

$$r_o = 20 \text{ k}\Omega$$

# Simple math

- Following relations are useful.
  - Sine and cosine functions can be expanded with  $e^{+j\omega t}$  and  $e^{-j\omega t}$ .

$$\sin \omega t = -\frac{j}{2}e^{+j\omega t} + \frac{j}{2}e^{-j\omega t}$$
$$\cos \omega t = \frac{1}{2}e^{+j\omega t} + \frac{1}{2}e^{-j\omega t}$$

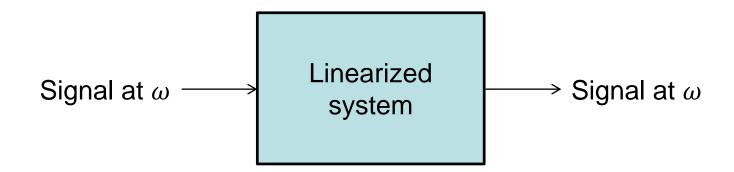
- Therefore, for a function of  $f(t) = f_s \sin \omega t + f_c \cos \omega t$ , the expansion is

$$f(t) = \left(-j\frac{f_s}{2} + \frac{f_c}{2}\right)e^{+j\omega t} + \left(+j\frac{f_s}{2} + \frac{f_c}{2}\right)e^{-j\omega t}$$

- A single complex number,  $-j\frac{f_s}{2} + \frac{f_c}{2}$ , is enough to represent f(t).

### Linearized system

- Our circuit is nonlinear in general.
- However, we have <u>linearized</u> it.
  - When the input signal has an angluar frequency,  $\omega$ , the output signal has the same one.
  - It is sufficient to consider the input-output relation at  $\omega$ .



#### **Impedance**

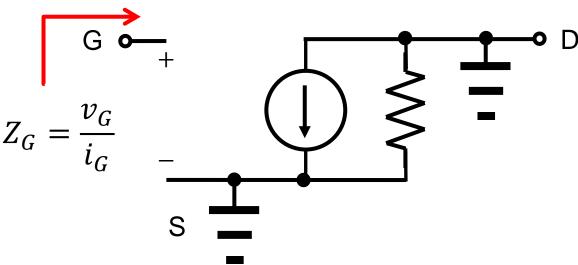
- Resistance, V(t) = R I(t)
  - It is assumed that V(t) and I(t) are in the same phase.
- Impedance,  $V(\omega) = Z(\omega)I(\omega)$ 
  - Consider  $V(t) = V_0 \sin \omega t$  and  $I(t) = I_0 \cos \omega t$ . (Different phases)
  - We introduce a phasor voltage,  $V(\omega)$ , and a phasor current,  $I(\omega)$ .
  - The relation between V(t) and  $V(\omega)$  is  $V(t) = Re[V(\omega)e^{j\omega t}]$ .
  - When  $V(t) = V_0 \sin \omega t$ , the phasor voltage is  $V(\omega) = -jV_0$ .
  - When  $I(t) = I_0 \cos \omega t$ , the phasor voltage is  $I(\omega) = I_0$ .
  - In this example,  $Z(\omega) = -j\frac{V_0}{I_0}$ . A purely imaginary number.

#### **Multi-terminal devices**

- When the number of terminals is 3,
  - We can define  $9 (= 3 \times 3)$  different impedances.
- Termination condition is important.
  - Depending on the termination condition, the impedance can be heavily changed.
  - In many cases, it is obvious from the problem.

### Impedances of MOSFET

- "Looking into the <u>TERMINAL</u>," we see the impedance of the <u>TERMINAL</u>.
  - Example) Looking into the gate. The source and drain are acgrounded.

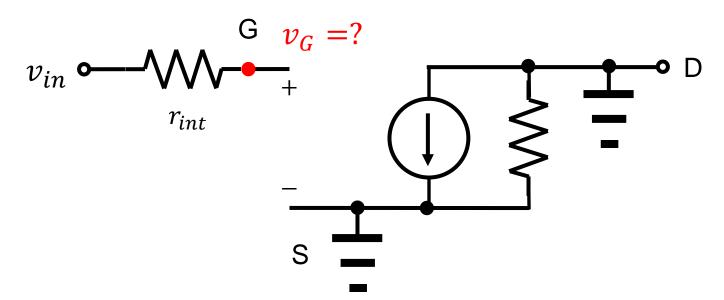


Similar for other terminals

#### Input impedance

- Consider a input signal with a finite internal resistance.
  - Usually, the internal resistance is small, but not zero.
  - The actual small-signal voltage applied to the gate terminal is given by

$$v_G(\omega) = v_{in} \frac{Z_G(\omega)}{r_{int} + Z_G(\omega)}$$



GIST Lecture on May 11, 2020 (Internal use only)