

1. Consider a PN junction at equilibrium. By using the depletion approximation, derive the depletion width.

2. When a forward bias,  $V_{app}$ , is applied to a PN junction, in the neutral p-type region, we must solve the following equation:

$$\frac{d^2}{dx^2} n = \frac{n - n_{p0}}{L_n^2}$$

In the above equation,  $n$  is the electron density,  $n_{p0}$  is the minority electron density at equilibrium, and  $L_n$  is the electron diffusion length. The neutral p-type region starts at  $x = 0$  and ends at  $x = \infty$ . At these boundaries, the boundary values of  $n$  are

$$n(x = 0) = n_{p0} \exp \frac{V_{app}}{k_B T / q},$$

$$n(x = \infty) = n_{p0}.$$

Calculate the solution,  $n(x)$ .

3. Repeat the same problem. However, in this problem, we assume that the neutral p-type region has a finite thickness of  $W_p$ . The boundary values of  $n$  are

$$n(x = 0) = n_{p0} \exp \frac{V_{app}}{k_B T / q},$$

$$n(x = W_p) = n_{p0}.$$

Calculate the solution,  $n(x)$ .

4. Consider a long-channel MOSFET operated in the saturation mode. In this case, we can assume that

$$I_d = \mu_{eff} C_{ox} \frac{W}{L} \frac{(V_{gs} - V_t)^2}{2m},$$

where quantities have the same meanings studied in our lectures. Let us assume that

$$V_{gs}(t) = 1.0 + V_t + 0.5 \cos 2\pi f t,$$

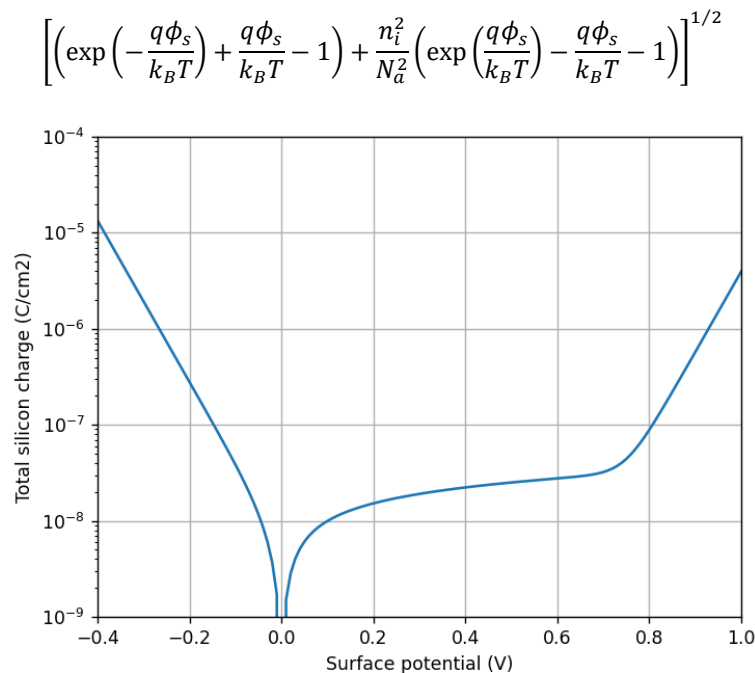
where  $f$  is the signal frequency. By calculating  $I_d$  explicitly, show that the output current has a signal with a frequency of  $2f$ .

5 (Counted as two problems). In our lectures, we have derived a general relation for the semiconductor charge,  $Q_s$ . It was given as

$$Q_s = \pm \sqrt{2\epsilon_{si} k_B T N_a} \left[ \left( \exp\left(-\frac{q\phi_s}{k_B T}\right) + \frac{q\phi_s}{k_B T} - 1 \right) + \frac{n_i^2}{N_a^2} \left( \exp\left(\frac{q\phi_s}{k_B T}\right) - \frac{q\phi_s}{k_B T} - 1 \right) \right]^{1/2}.$$

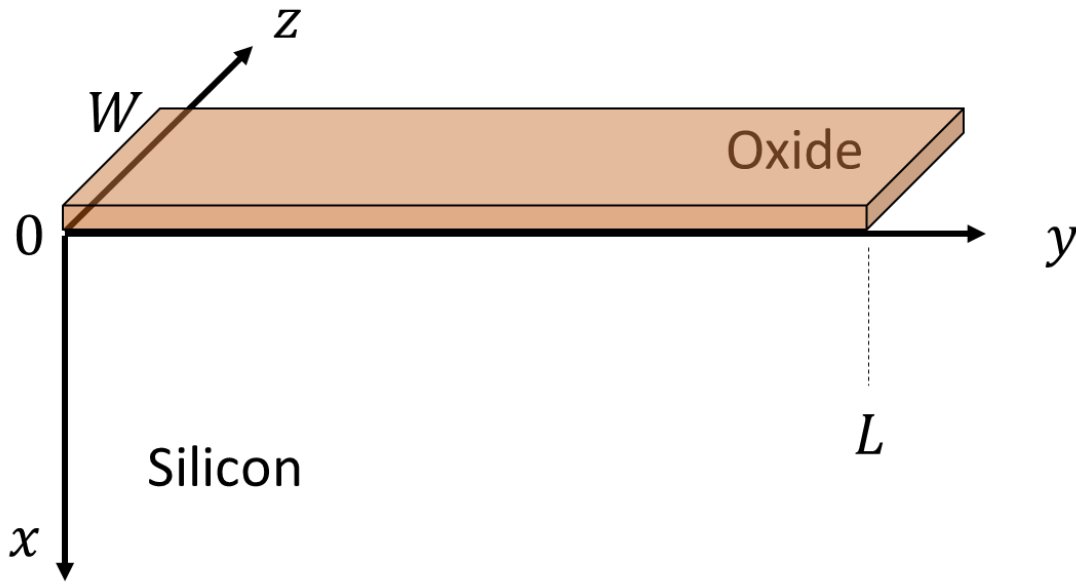
It is noted that the surface potential,  $\phi_s$ , is referenced to the p-type substrate. Derive this relation.

6. The following figure shows  $|Q_s|$  as a function of  $\phi_s$ . We can clearly identify three different regions. For each of these three regions, identify the most important term in the square bracket.



7. Now, we can express  $Q_s$  as a function of  $\phi_s$ . However, the surface potential cannot be directly changed. Instead, by changing  $V_g$ , we can indirectly control the surface potential. Write down the relation between  $V_g$  and  $\phi_s$ . Of course,  $Q_s(\phi_s)$  should appear in your answer.

8. What is the gradual channel approximation? Explain it. (You may use the following coordinate system.)



9. We have assumed that the electron quasi-Fermi potential,  $V$ , is independent of the vertical direction,  $x$ . Justify this assumption.

10. When  $V$  is zero, we have the following results:

$$E_x^2(x, y) = \left(\frac{d\phi}{dx}\right)^2 = \frac{2k_B T N_a}{\epsilon_{si}} \left[ \left( \exp\left(-\frac{q\phi}{k_B T}\right) + \frac{q\phi}{k_B T} - 1 \right) + \frac{n_i^2}{N_a^2} \left( \left( \exp\left(\frac{q\phi}{k_B T}\right) - 1 \right) - \frac{q\phi}{k_B T} \right) \right].$$

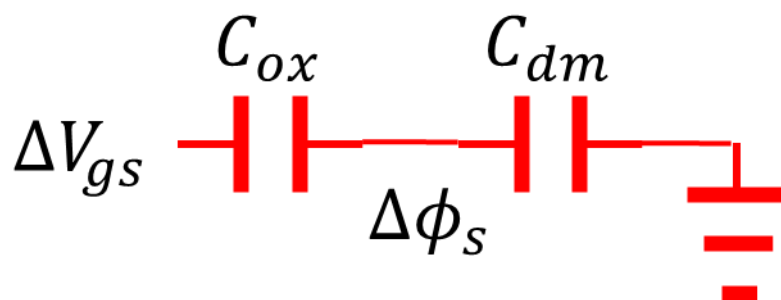
Now, at a certain  $y$ ,  $V(y)$  is non-zero. Modify the above equation to consider this non-zero  $V(y)$ .

11. After a long derivation procedure, we have found an expression for  $I_d$ :

$$I_d = \mu_{eff} \frac{W}{L} \left\{ C_{ox} \left( V_{gs} - V_{fb} + \frac{k_B T}{q} \right) \phi_s - \frac{1}{2} C_{ox} \phi_s^2 - \frac{2}{3} \sqrt{2\epsilon_{si} q N_a} \phi_s^{1.5} + \frac{k_B T}{q} \sqrt{2\epsilon_{si} q N_a} \phi_s \right\} \Big|_{\phi_{s,s}}^{\phi_{s,d}}$$

Now, assume that  $\phi_{s,s}$  and  $\phi_{s,d}$  are 1.006 V and 1.206 V, respectively.  $\mu_{eff} \frac{W}{L}$  is  $250 \text{ cm}^2 \text{ V}^{-1} \text{ sec}^{-1}$ . The vacuum permittivity is  $8.854 \times 10^{-12} \text{ F m}^{-1}$  and the relative permittivity of silicon is 11.7. The oxide is 10 nm-thick and  $V_{gs} - V_{fb}$  is 1.88 V. For the thermal voltage, use 25.85 mV. The substrate doping concentration is  $10^{17} \text{ cm}^{-3}$ . Of course,  $q$  is  $1.6 \times 10^{-19} \text{ C}$ . By using these numbers, calculate  $I_d$ .

12. In our lectures, we introduced a factor,  $m$ . Discuss its physical meaning. (You may use the following circuit diagram.)



13. In the subthreshold region, we have an approximate form for the inversion charge,

$$-Q_i \approx \sqrt{\frac{\epsilon_{si} q N_a k_B T}{2\phi_s}} \frac{n_i^2}{q N_a^2} \exp\left(\frac{q}{k_B T} (\phi_s - V)\right),$$

where  $\phi_s$  can be regarded as a constant. By using a relation,

$$I_d = \mu_{eff} \frac{W}{L} \int_0^{V_{ds}} [-Q_i(V)] dV,$$

derive an expression for  $I_d$ .

14. Draw a schematic diagram of a DRAM cell. Describe its READ operation. (The sense amplifier plays an important role.)