

# EE102

## Lecture 8

# EE102 Announcements

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- **Third Homework due this Friday**
- **Midterm is 10/30**

**Slide Credits:** This lecture adapted from Prof. Jonathan Kao (UCLA) and recursive credits apply to Prof. Cabric (UCLA), Prof. Yu (Carnegie Mellon), Prof. Pauly (Stanford), and Prof. Fragouli (UCLA)

# Complex Numbers

## CYU: Midterm Review

Find the real and imaginary parts of the following signal:

$$x(t) = (1 - \sqrt{3}j)(\underline{e^{j(t+2)}})$$

$$e^{j(t+2)} = \cos(t+2) + j\sin(t+2)$$

$$\begin{aligned}\text{Ans. } & (1 - \sqrt{3}j) [\cos(t+2) + j\sin(t+2)] \\ &= \cos(t+2) + j\sin(t+2) - \sqrt{3}j\cos(t+2) + \sqrt{3}\sin(t+2) \\ &= \cos(t+2) + \sqrt{3}\sin(t+2) + \sin(t+2)j - \sqrt{3}\cos(t+2)j \\ & \quad \text{Re: } \cos(t+2) + \sqrt{3}\sin(t+2) \quad \text{Im: } \sin(t+2) - \sqrt{3}\cos(t+2)\end{aligned}$$

# CYU: Midterm Review

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# System Properties

## CYU: Midterm Review

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Is the following system linear? Time-invariant? Stable? Causal?

$$y(t) = \pi + tx(t+3)$$

Linearity: No. Violates Homogeneity  $S[0] \neq 0$

Time-Invariance: No. Direct Inspection of  $t$  term.

Stable: No. [E.g. Counter-e.g. if  $x=1$ , at  $t=\infty \rightarrow y$  is  $\infty$ .]

Causal: No [E.g.  $y(t)$  depends on  $x(3)$ ]

# CYU: Midterm Review

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# Convolution

## CYU: Convolution with Dirac Delta

Compute the convolution of the two signals below:

$$f(t) = \delta(t + 5) + 2\delta(t) \text{ and } g(t) = e^t$$

$$\begin{aligned} y(t) &= f(t) \star g(t) \\ &= [\delta(t+5) + 2\delta(t)] \star e^t \\ &= \delta(t+5) \star e^t + 2\delta(t) \star e^t \\ &= e^{t+5} + 2e^t \end{aligned}$$

# CYU: Convolution with Dirac Delta

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# Signal Properties

## Periodicity and Fundamental Period

Consider the following signals and determine whether they are periodic or not. If periodic, identify the fundamental period.

i.  $x(t) = \sin\left(\frac{4}{5}t + \frac{\pi}{3}\right)$

$$x(t+T_0) = \sin\left(\frac{4}{5}t + \frac{\pi}{3} + \frac{\frac{4}{5}T_0}{2\pi}\right)$$

ii.  $x(t) = \boxed{e^{-t}} \sin\left(\frac{4}{5}t + \frac{\pi}{3}\right)$

$$\frac{4}{5}T_0 = 2\pi$$

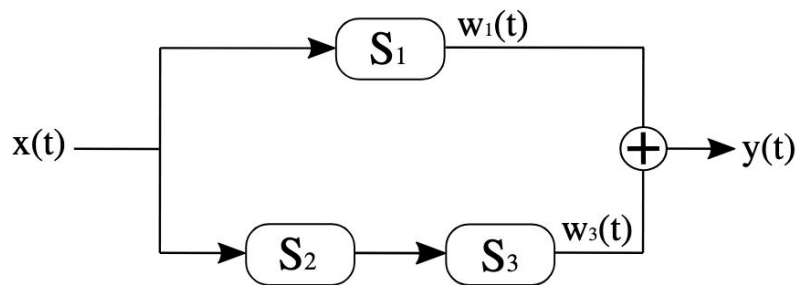
$$T_0 = \frac{10\pi}{4} = \boxed{\frac{5\pi}{2}}$$

Not periodic,  
By Direct Inspection, we observe  
the  $e^{-t}$  term.

# Periodicity and Fundamental Period

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# System Response of LTI System



We know that Systems 1, 2 and 3 are all LTI systems, which are used in parts a through c.  $w_1(t)$  and  $w_3(t)$  are the outputs of Systems 1 and 3, respectively. Let  $h_1(t)$ ,  $h_2(t)$  and  $h_3(t)$  represent the impulse response for System 1, 2 and 3, respectively. For parts (a) through (c), we have prior knowledge of Subsystem 3

$$h_3(t) = \int_{-\infty}^{t-3} \delta(\tau - 4) d\tau.$$

For parts (a) through (c), we also have prior knowledge of the input/output mapping of the entire system

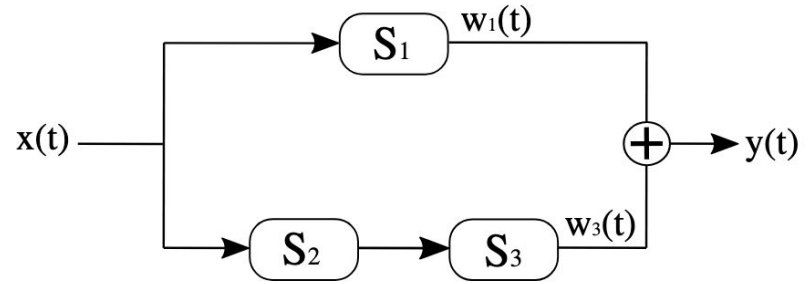
$$y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau + \int_{-\infty}^{t+1} x(\tau) d\tau.$$

- (a) (10 points) What is the impulse response of the entire system (i.e.  $S_{eq}$ )?
- (b) (10 points) Find  $h_1(t)$  and  $h_2(t)$  that satisfies the input and output relationship that is given. It might be useful to determine the values of  $w_1(t)$  and  $w_3(t)$  first.
- (c) (6 points) Using the solution from parts a and b, is System 1 Causal/Stable? Is the other subsystem (Cascaded Systems 2 and 3) Causal/Stable? Finally, is  $S_{eq}$  Causal/Stable?

Sifting + Integration

# System Response of LTI System

$$y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau + \int_{-\infty}^{t+1} x(\tau) d\tau.$$



(a) What is the impulse response of the entire system?

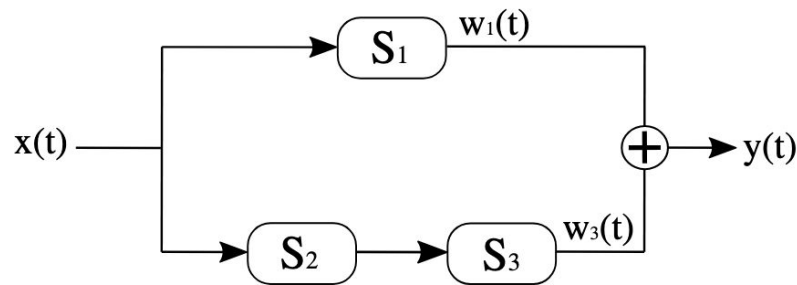
$$\begin{aligned} h(t) &= \int_{-\infty}^t e^{-3t} e^{3\tau} \delta(\tau) d\tau + \int_{-\infty}^{t+1} \delta(\tau) d\tau \\ &= e^{-3t} \int_{-\infty}^t e^{3\tau} \delta(\tau) d\tau + \int_{-\infty}^{t+1} \delta(\tau) d\tau \\ &= e^{-3t} u(t) + u(t+1) \end{aligned}$$

# System Response of LTI System

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# System Response of LTI System

$$y(t) = \int_{-\infty}^t e^{-3(t-\tau)} x(\tau) d\tau + \int_{-\infty}^{t+1} x(\tau) d\tau.$$



- (b) Find  $h_1(t)$  and  $h_2(t)$  that satisfies the input and output relationship that is given. It might be useful to determine the values of  $w_1(t)$  and  $w_3(t)$  first.

$$h(t) = e^{-3t} u(t) + \underbrace{u(t+1)}_{\substack{\text{TRY this being} \\ h_1}} + \underbrace{u(t+1)}_{\substack{\text{if } h_2 \text{ and} \\ h_3 \text{ giving} \\ \text{me this branch}}}$$

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Given  $h_3$ :  $h_3(t) = \int_{-\infty}^{t-3} \delta(\tau-4) d\tau$   
 $= u(t-7)$

$h_2$ 's action on this is a time shift by 8  
units.

$$\therefore h_2(t) = \delta(t+8)$$

# System Response of LTI System

- (c) Using the solution from parts a and b, is System 1 Causal/Stable? Is the other subsystem (Cascaded Systems 2 and 3) Causal/Stable? Finally, is  $S_{eq}$  Causal/Stable?

$$h_1 = e^{-3t} u(t) \Rightarrow \text{Causal, Stable}$$

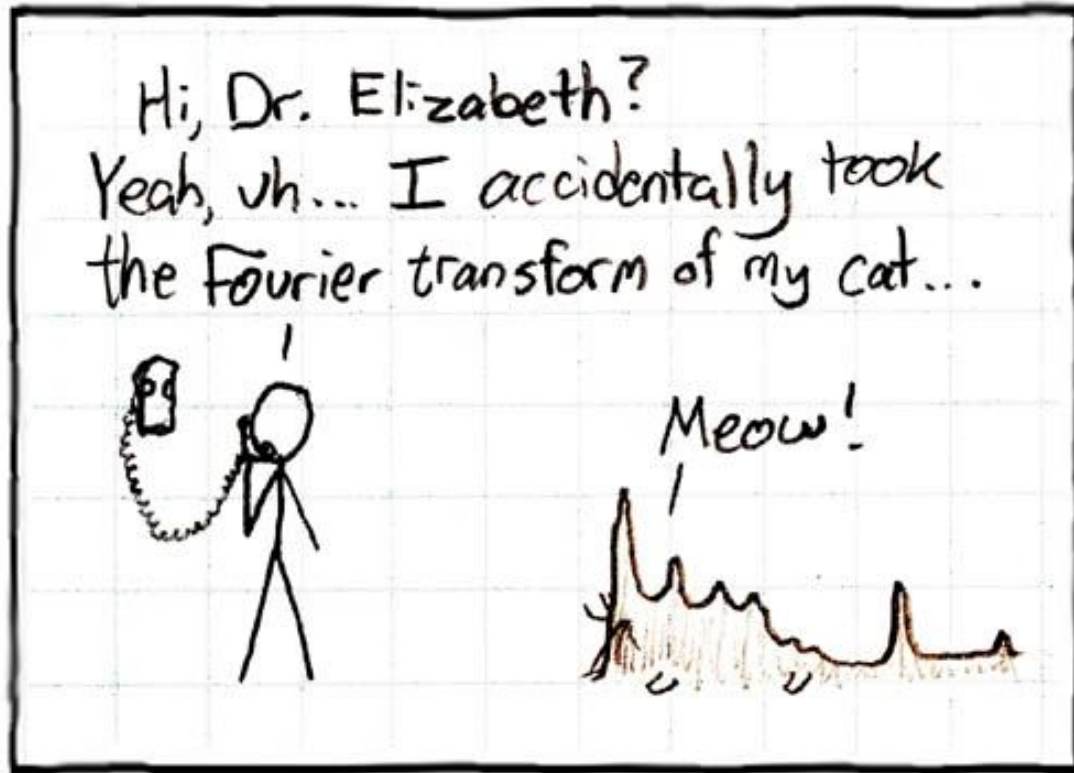
$$h_2 \text{ ~~and~~ } h_3 = u(t+1) \Rightarrow \text{Not Causal, Not Stable}$$

$$h_{eq} = e^{-3t} u(t) + u(t+1) \Rightarrow \text{Not Causal, Not Stable}$$





# This Lecture Breaks a New Seal



Joseph Fourier

# Fourier was a Super Interesting Guy

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- Born, France.
- Son of a Tailor; Orphan at 10 years old.
- Aspires to be Newton, but at age 21 feels like a failure.
- Joins politics and goes to Jail
- Narrowly escapes guillotine in the French Revolution
- Confidante of Napoleon

French  
Revolution

# Intuition: Why Fourier Series?

Why Fourier Series?

Bottom Line of F.S.

Any periodic signal can be expressed as a sum of sinusoids

$$1) \quad \underbrace{x(t)}_{\substack{\Downarrow \\ \text{Some periodic} \\ \text{signal}}} = \underbrace{x_1(t)}_{\substack{\Downarrow \\ \text{Sinusoid}}} + \underbrace{x_2(t)}_{\substack{\Downarrow \\ \text{Sinusoid}}} + \dots$$

△ Goal: Simplify analysis of systems

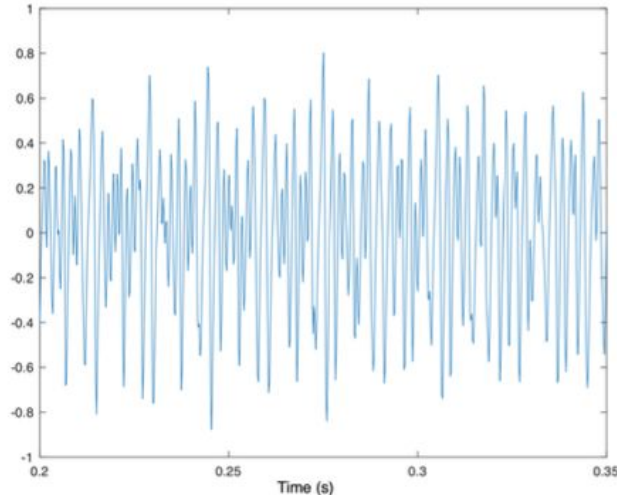
2) A different & Interesting way to think about data/info.  
Can reveal interesting and unexpected structure.

# Fourier Series

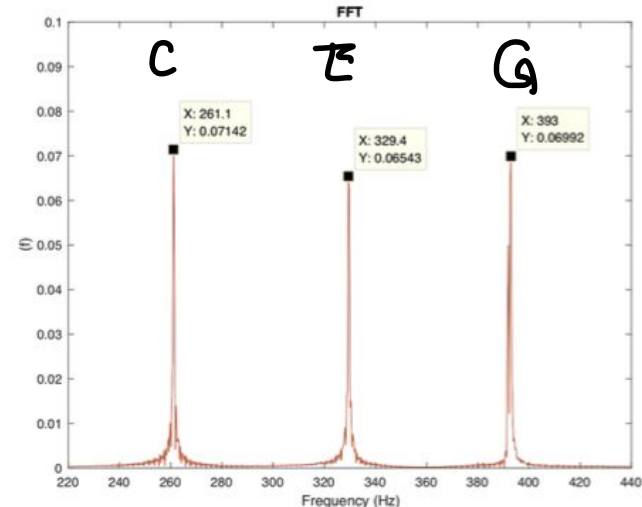
Extracts *frequency structure* from a signal.

Below: we have a C-major chord. It consists of three notes: C (261.1 Hz), E (329.4 Hz) and G (393 Hz). When we play these three notes at the same time, they create a waveform indicated by the blue line. It's hard to see structure here.

The Fourier series is able to write this as a sum of sinusoids. When we do, we find that there are only 3 frequencies, corresponding to our C-major chord.



Fourier



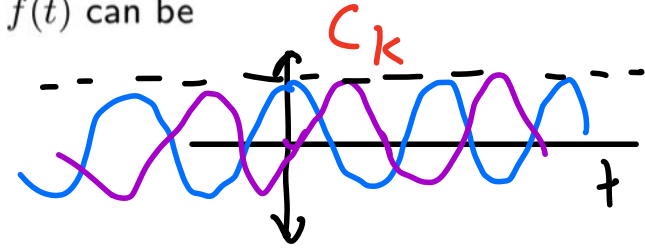
# Fourier Series - Bottom Line

$$e^{jk\omega_0 t} = \cos(k\omega_0 t) + j\sin(k\omega_0 t)$$

These are the main mathematical results of this lecture, written here for convenience.

If  $f(t)$  is a well-behaved periodic signal with period  $T_0$ , then  $f(t)$  can be written as a Fourier series

$$f(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$



1) Any Periodic Signal

where  $\omega_0 = \frac{2\pi}{T_0}$  and

2) Find the coefficients of each sinusoid.

$$c_k = \frac{1}{T_0} \int_{\tau}^{\tau+T_0} f(t) e^{-jk\omega_0 t} dt$$

for all integers  $k$ . The  $c_k$  are called the *Fourier coefficients* of  $f(t)$ .

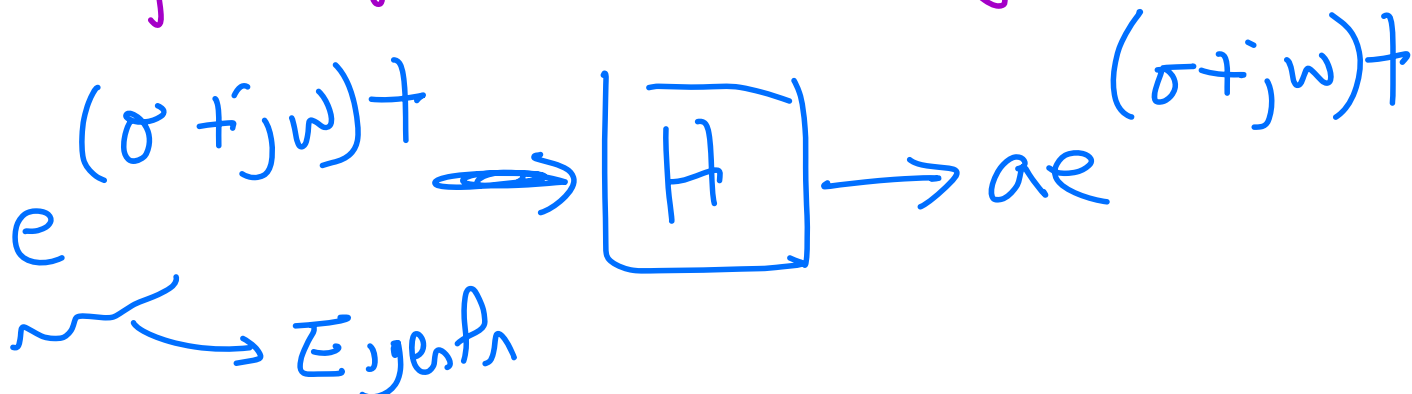
Here,  $f(t)$  is the *weighted average* of complex exponentials (which are simply complex sines and cosines).

# Eigenfunctions

*Eigenfunction*  $x(t)$  is an *eigenfunction* of a system if, when inputting  $x(t)$  to the system, the output is simply a scaled version of  $x(t)$ , i.e.,  $y(t) = ax(t)$  where  $a$  is a constant (called an eigenvalue). Note that  $a$  may be a complex constant.



~~\*\*\*~~ Complex Exponentials are eigenfunctions of LTI systems



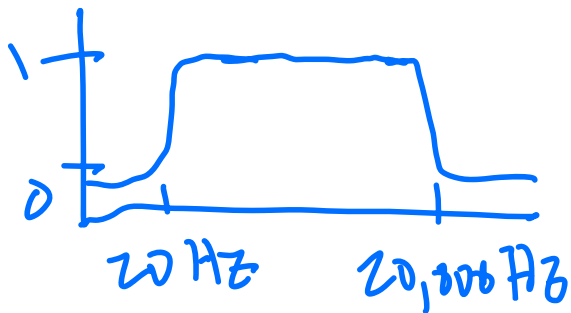
$$y(t) = \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau$$

# Eigenfunctions of LTI Systems

Consider an LTI system with impulse response  $h(t)$ . If the input is a complex exponential, i.e.,  $e^{st}$  where  $s = \sigma + j\omega$ , then

$$e^{st} \rightarrow \boxed{h(\tau)} \rightarrow H(s) e^{st}$$

Human Ear  
 $H(s)$



$$y(t) = \int_{-\infty}^{\infty} h(\tau) e^{s(t-\tau)} d\tau$$

Algebra  $\swarrow$

$$= \int_{-\infty}^{\infty} h(\tau) e^{st} e^{-s\tau} d\tau$$

Pop Out  $\swarrow$

$$= e^{st} \int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau$$

Constant  $\swarrow$

"Transfer Fn"

$$H(s) \triangleq \int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau$$

$$= e^{st} H(s)$$



# Eigenfunctions of LTI Systems (Intuition)

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This shows that the complex exponential is an eigenfunction of an LTI system.

- If I input a complex exponential into the LTI system, I get the same complex exponential out scaled by  $H(s)$ .

From here, we can see how Fourier series might help:

- First, I decompose my signal,  $x(t)$ , into the sum of complex exponentials.
- After this, I put each complex exponential into my LTI system. Since the system is LTI, each complex exponential comes out scaled by  $H(s)$ .
- Then, I can add my scaled complex exponentials at the output to get the system output.
- Conveniently, since the output is a sum of (scaled) complex exponentials, it is also a Fourier series.

Let's formalize this intuition we've stated here and get to the math.

# Eigenfunctions of LTI Systems

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# Eigenfunctions of LTI Systems

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For LTI systems, a simple way to analyze them is to:

- Decompose  $x(t)$ , the input, into its Fourier series, i.e., a sum of complex exponentials. This represents its decomposition into “fundamental” components, which are sinusoids at different frequencies  $k\omega_0$ .
- Because the complex exponential is an eigenfunction of an LTI system, if I pass in a complex exponential into my system, I get the same complex exponential at the output, scaled by  $\hat{H}(jk\omega_0)$ .
- Since LTI systems are distributive, if I pass in a sum of these complex exponentials into my system, I get back an output,  $y(t)$ , that is a sum of scaled complex exponentials (where the scale term is \_\_\_\_\_).

# (Simple) Fourier Series Example

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## Motivation for Fourier series

With this motivation, we now need to know how to actually calculate Fourier series, i.e., how do I find the  $c_k$  so that

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

And further, when is this possible? The rest of this lecture will cover this.

# (Simple) Fourier Series Example

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## Fourier series of a cosine (cont.)

Let's start simple. Consider the sinusoid:

$$f(t) = A \cos(\omega_0 t + \theta)$$

# (Simple) Fourier Series Example

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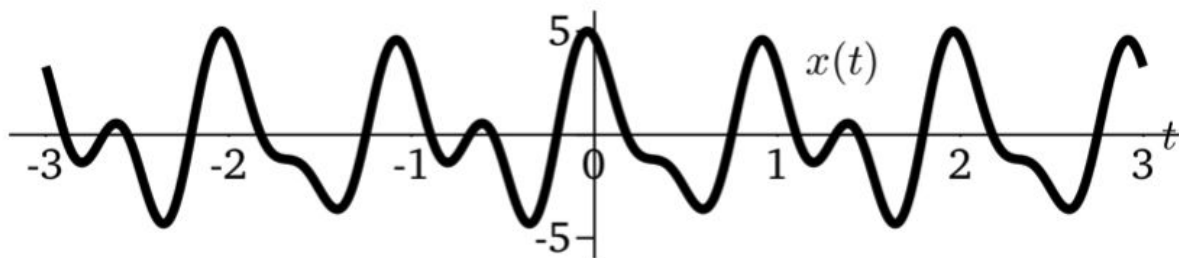
## Using spectrum to find structure

For the cosine example, it doesn't look like we made our lives easier by representing it as a spectrum. But for any more complex signal, it can.

Consider the signal

$$x(t) = 3 \cos(2\pi t) + \cos(3\pi t - \pi/4) + 2 \cos(4\pi t + \pi/3)$$

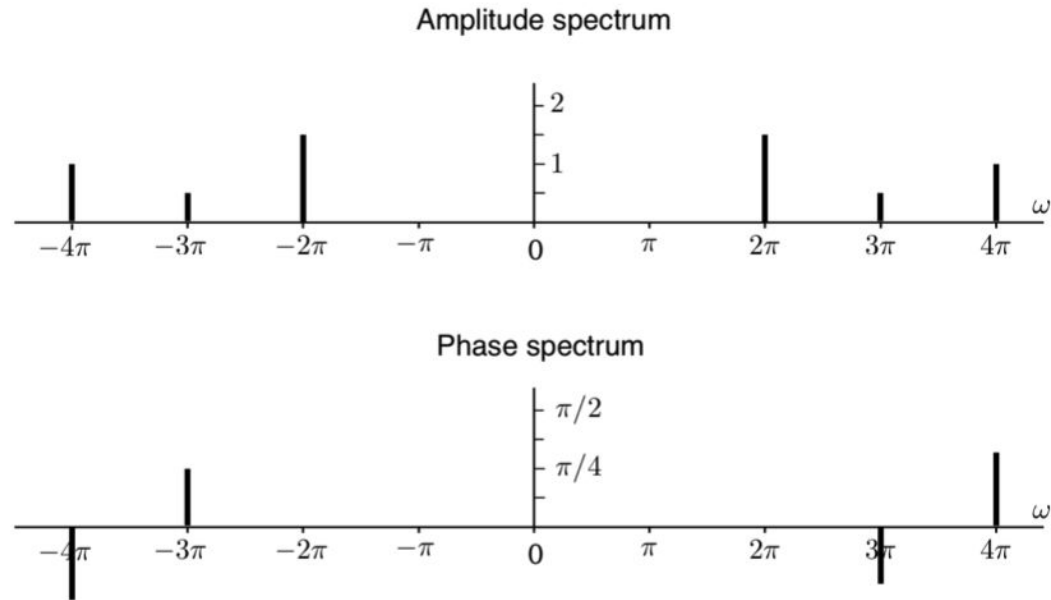
This signal is very simple. However, if I gave you a plot of its time domain representation, it'd be hard to recover exactly what  $x(t)$  is.



# (Simple) Fourier Series Example

## Using spectrum to find structure (cont.)

However, if we plotted the spectrum of this signal, it would look like the following:



From the spectrum, we can read off that this signal is composed of sinusoids at three different frequencies ( $2\pi$ ,  $3\pi$ , and  $4\pi$ ) with amplitudes given by the amplitude spectrum and phases given by the phase spectrum.

# Deriving the Fourier Series Coefficients

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## Deriving Fourier series

How do we find the  $c_k$ ?

Our derivation is as follows:

- First, we *assume* that the signal  $f(t)$  can be written as a sum of complex exponentials that are scaled by coefficients  $c_k$ .
- Given this assumption, we find if there are  $c_k$  such that we can represent  $f(t)$  in this way.