

Chapter 5: Reduction of Multiple Subsystems

ELG 3155 : Introduction to Control Systems

❑ Objective:

- Reduce a block diagram of multiple subsystems.
- Represent in signal-flow graph a system consisting of multiple subsystems

Outline

5 Reduction of Multiple Subsystems.

5.1 Introduction.

5.2 Block Diagrams.

5.3 Analysis and design of feedback systems.

5.4 Signal-Flow Graphs

5.5 Mason's Rule.

5.6 Signal-Flow Graphs of State Equations.

1. Introduction

- Finding the Transfer function of a system represented by a block diagram is not always straightforward.
- For example, find the transfer function of the system in Figure.
- Reducing (i.e., simplifying) the block diagram is one way to derive the transfer function of such a system.

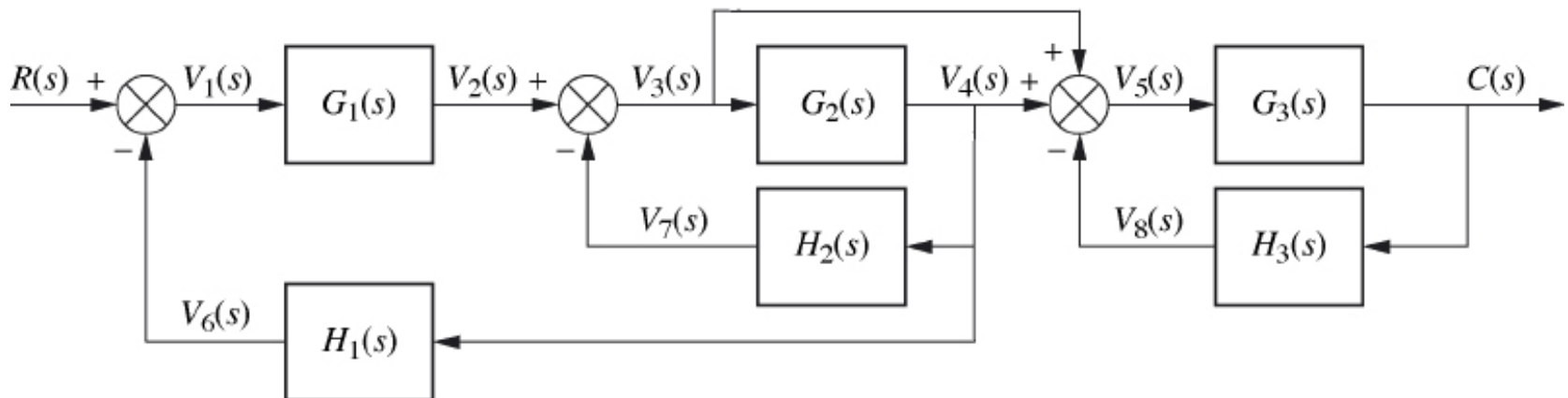
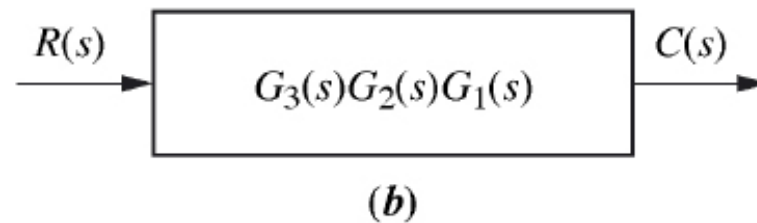
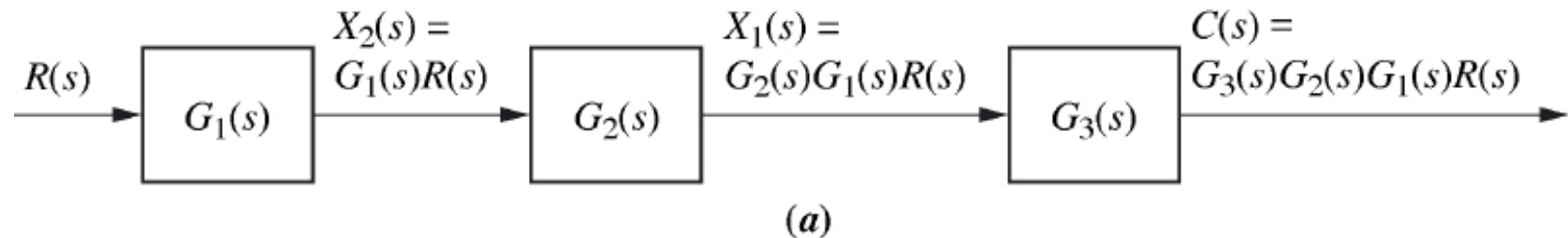


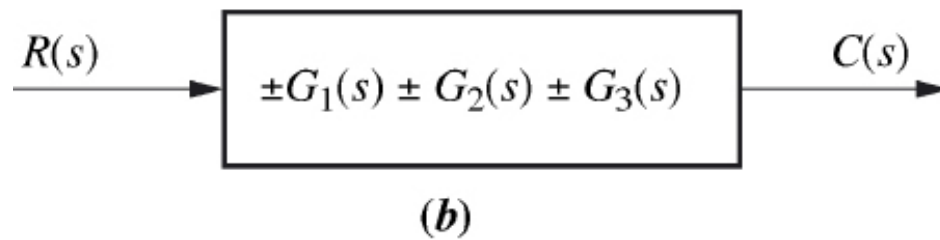
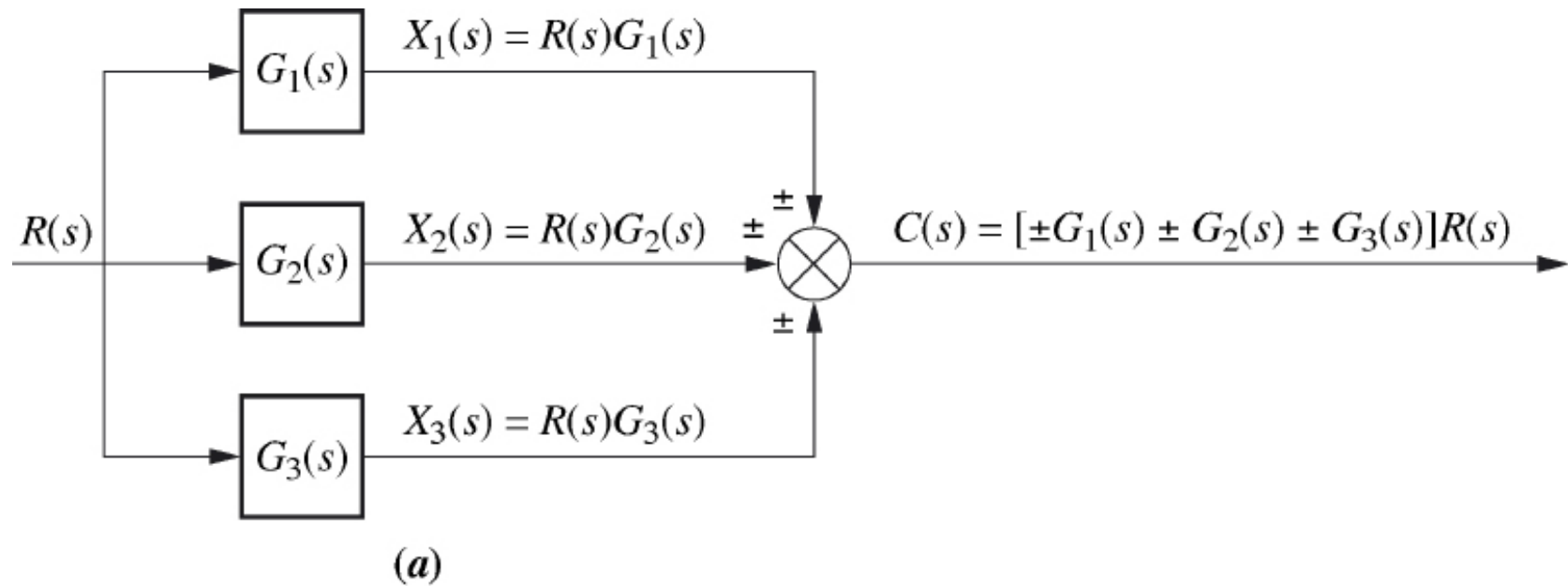
Figure 1: (Fig. 5.11 of [Nise, 2015])

2. Block Diagram Reduction

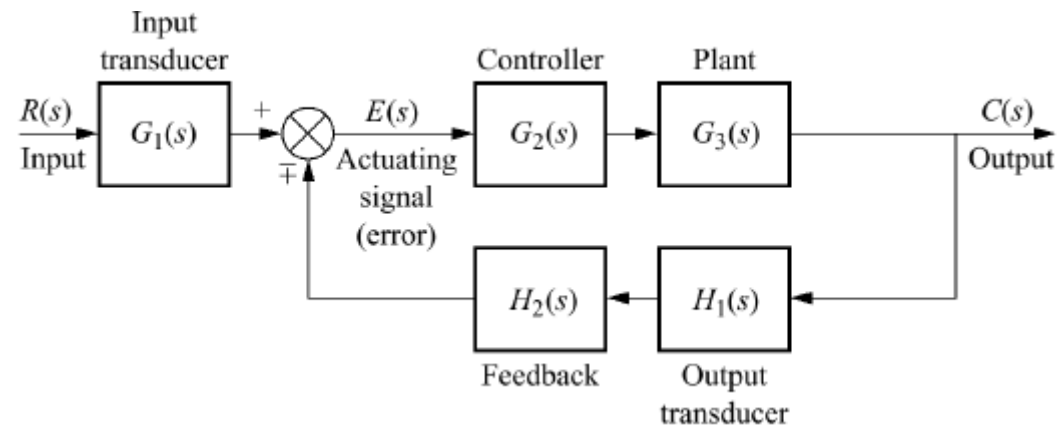
1. Cascade Form



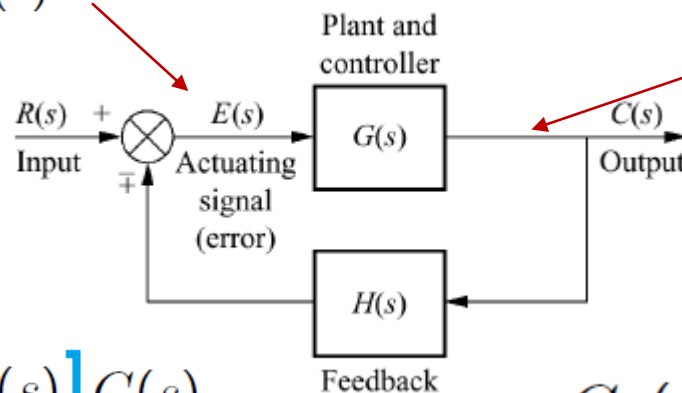
2. Parallel form



3. Feedback Form



$$E(s) = R(s) \mp C(s)H(s)$$

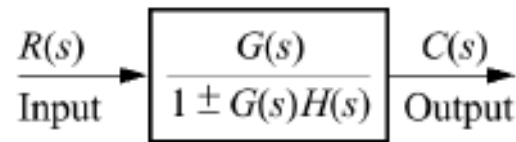


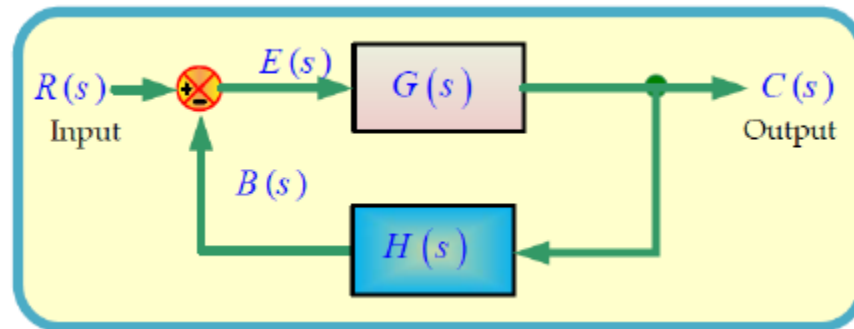
$$C(s) = E(s)G(s)$$

$$C(s) = [R(s) \mp C(s)H(s)] G(s)$$

$$G_e(s) = \frac{C(s)}{R(s)} = \frac{G(s)}{1 \pm G(s)H(s)}$$

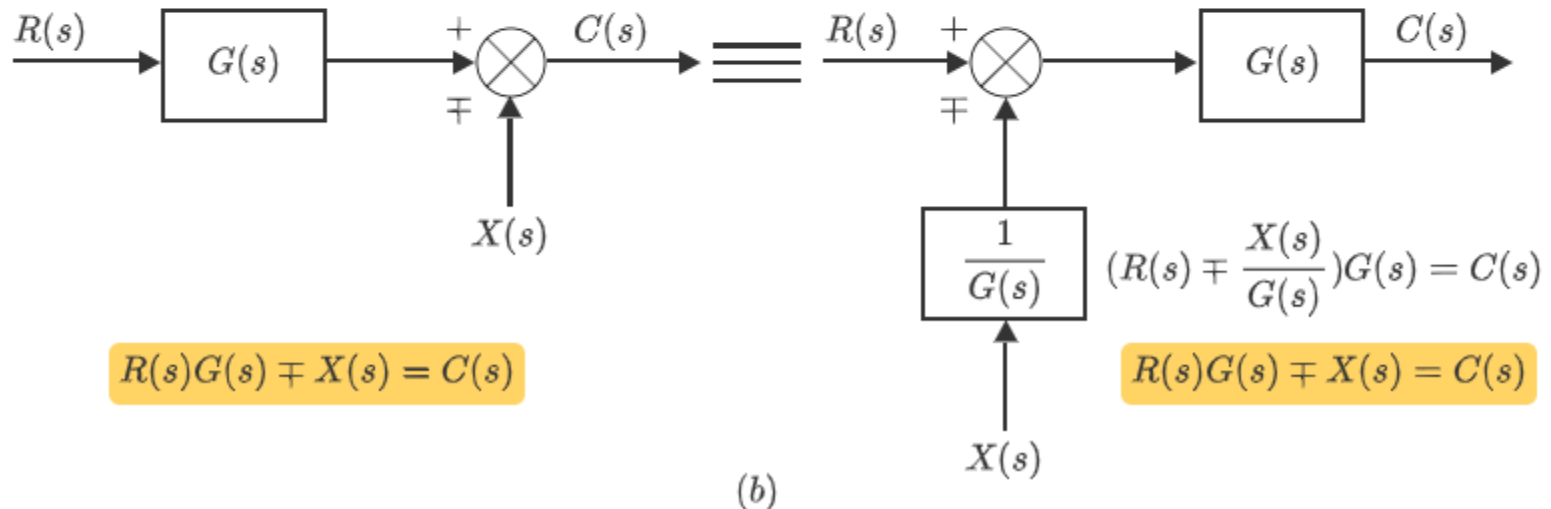
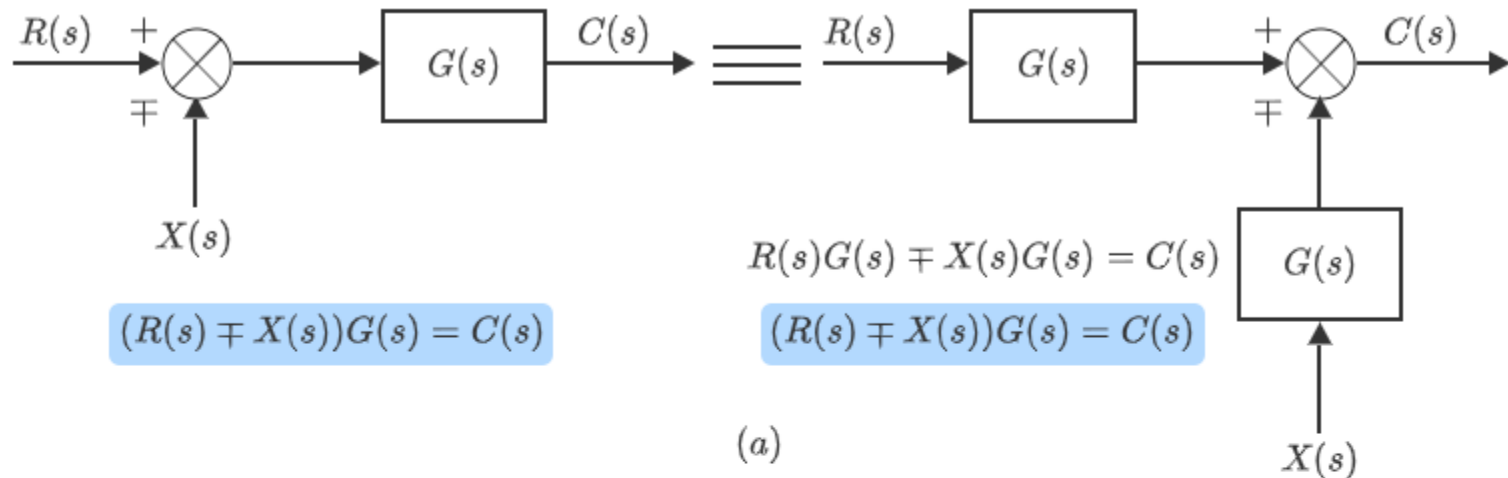
We call $G(s)H(s)$ the **open loop transfer function** or **loop gain**.

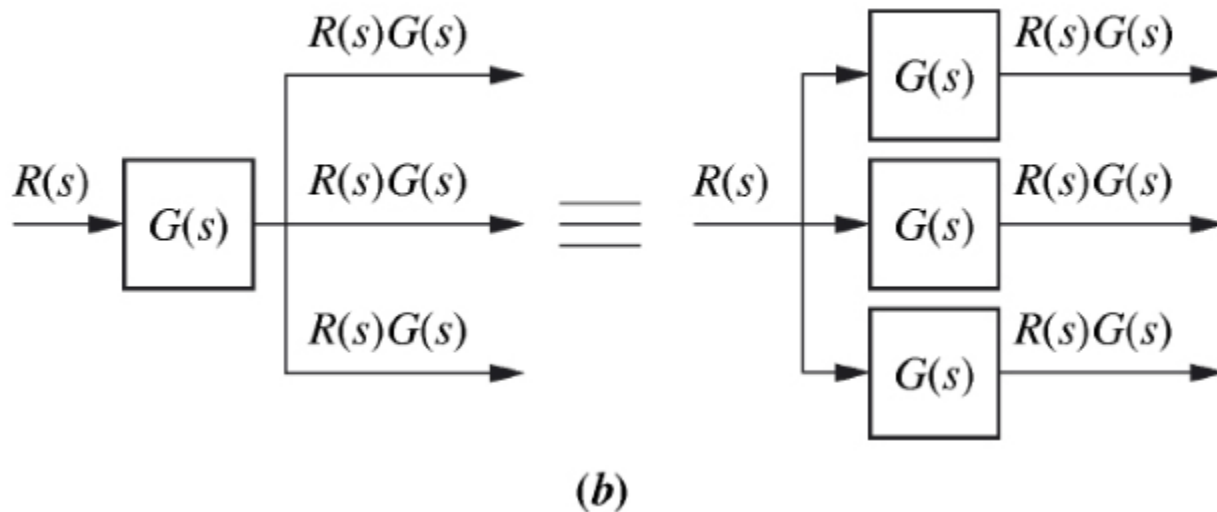
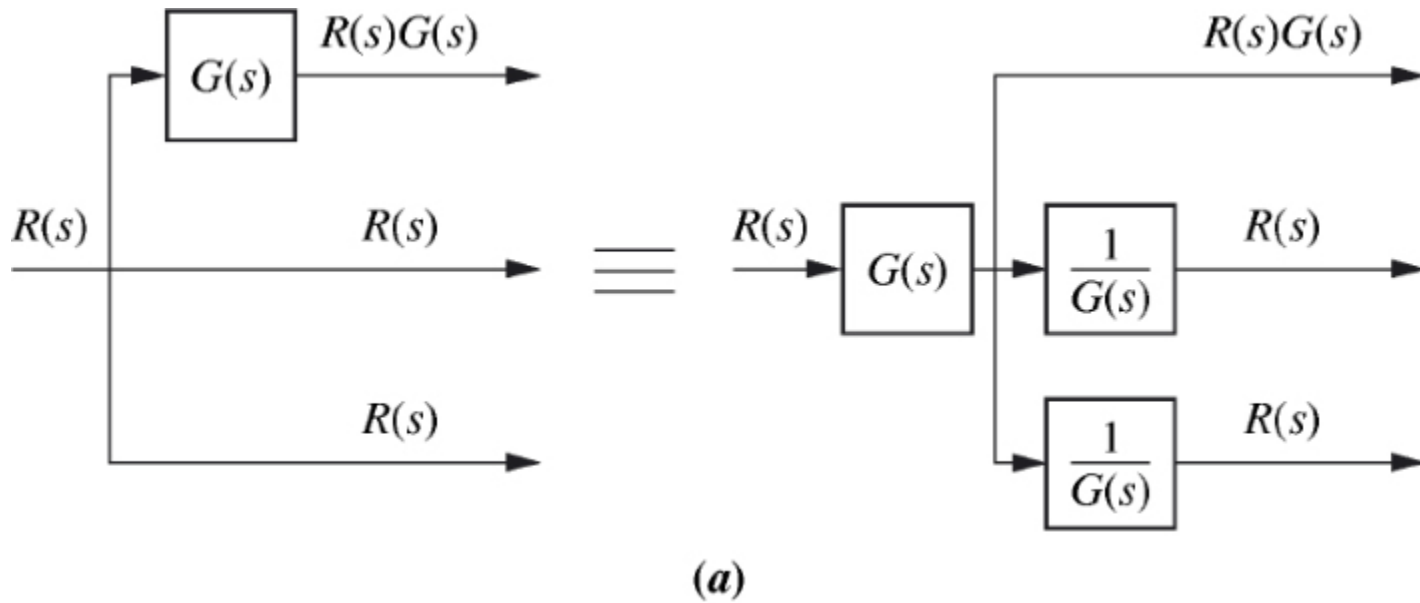




$$\frac{C(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$

■ Moving Blocks to create Familiar Forms





Definitions

In reference to Fig. 2, Feedforward gain = $G_1 G_2 \cdots G_n(s)$

Feedback gain = $H_1 H_2 \cdots H_m(s)$

Open-loop transfer function (OLTF) = feedforward \times feedback

$$= G_1 G_2 \cdots G_n H_1 H_2 \cdots H_m$$

Closed-loop transfer function (CLTF) = System's transfer function

$$= \text{feedforward} / (1 + \text{OLTF})$$

$$= \frac{G_1 G_2 \cdots G_n}{1 + G_1 G_2 \cdots G_n H_1 H_2 \cdots H_m}$$

For a positive feedback, CLTF = feedforward / $(1 - \text{OLTF})$

$$= \frac{G_1 G_2 \cdots G_n}{1 - G_1 G_2 \cdots G_n H_1 H_2 \cdots H_m}$$

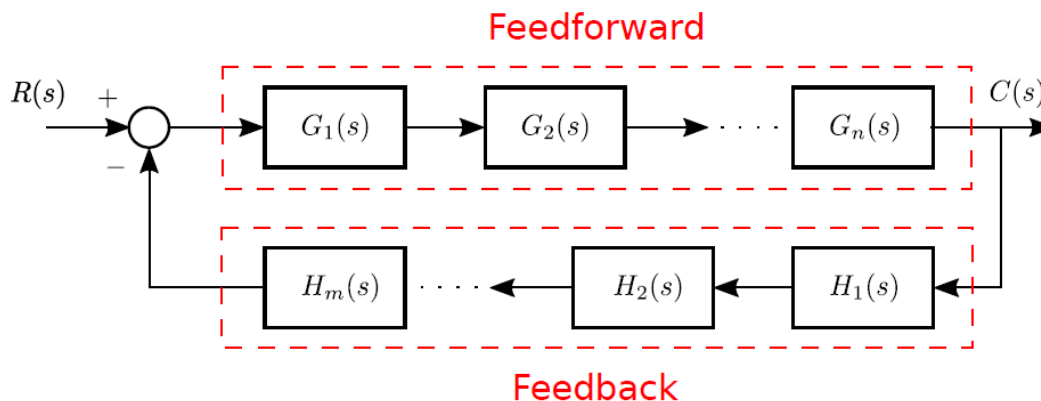
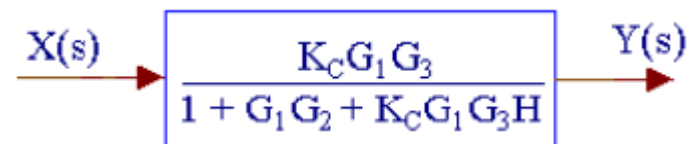
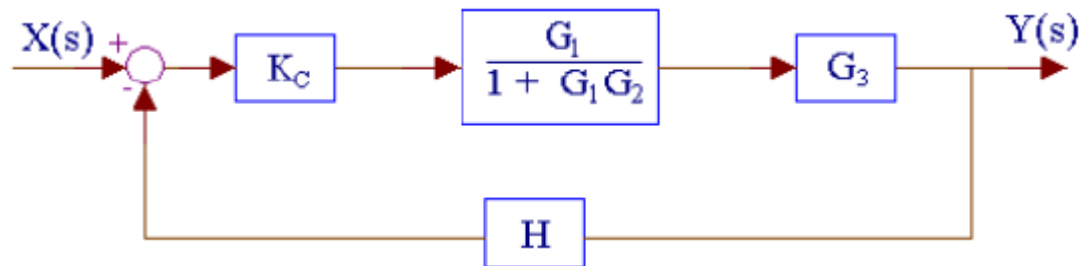
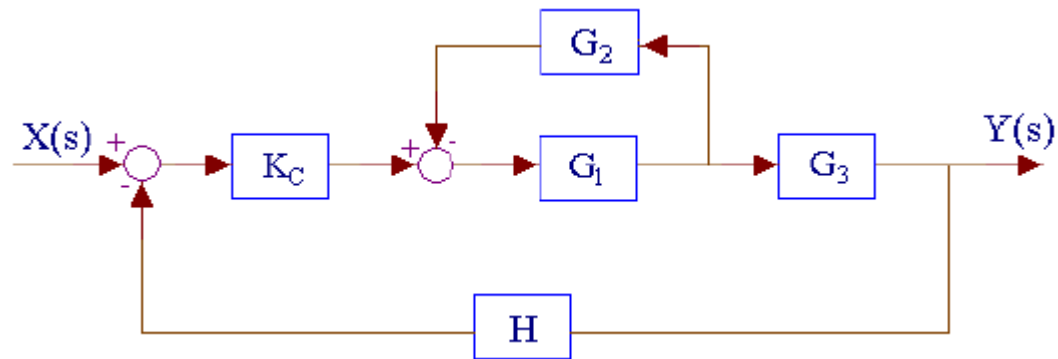


Figure 2: Block diagram of a system with a negative feedback

Example 1



Example 5.2: Find the transfer function of the system in Fig

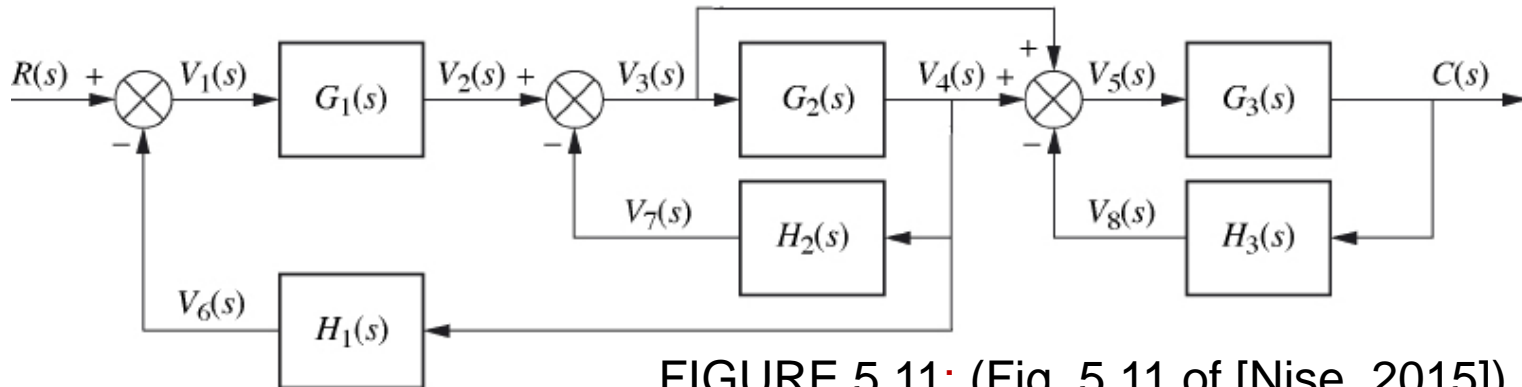
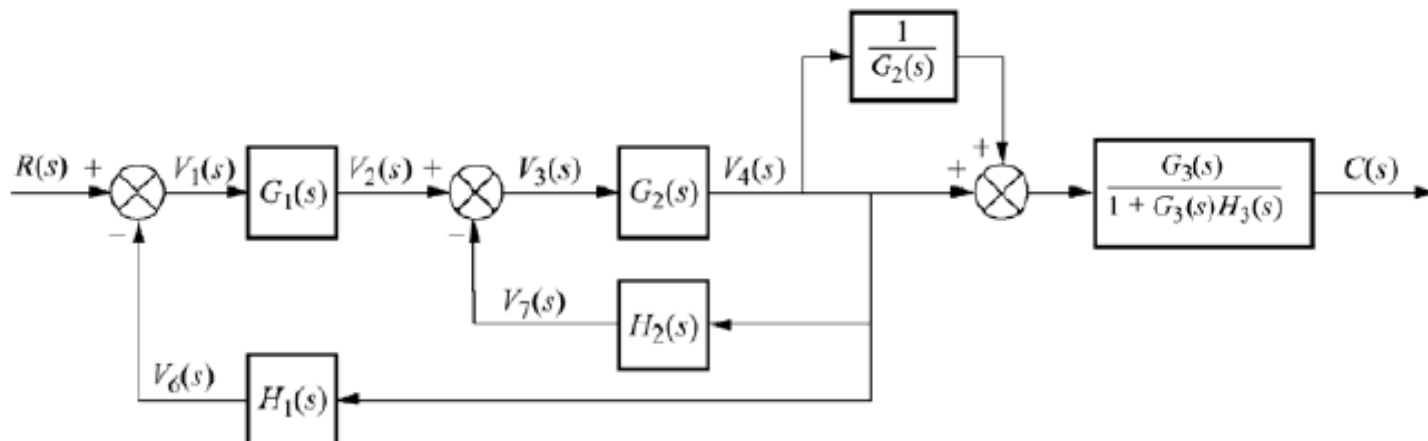
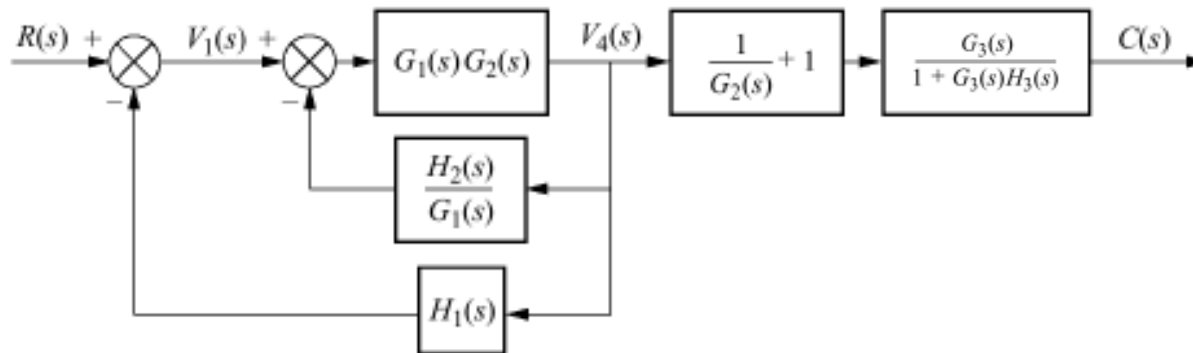


FIGURE 5.11: (Fig. 5.11 of [Nise, 2015])

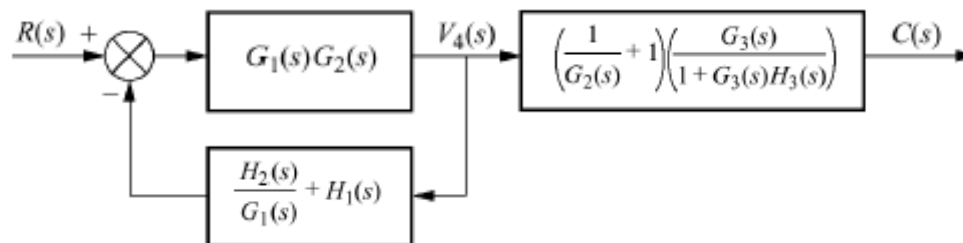
1. Move G_2 to left of pickoff point creating parallel form.
2. Reduce feedback system (G_3, H_3).



3. Reduce parallel form containing $\frac{1}{G_2(s)}$ and unity.
4. Push $G_1(s)$ to the right past summing junction. Creates parallel form (H_1 and $[\frac{1}{G_1}, H_2]$).
5. Combine serial forms (G_1, G_2) and $(\frac{1}{G_1}, H_2)$.



6. Collapse summing junctions, and combine parallel form.
7. Combine serial form on right.



8. Collapse feedback form.
9. Combine the two cascade blocks.

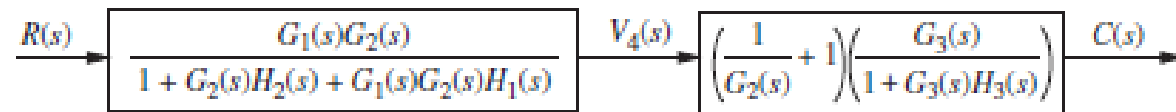
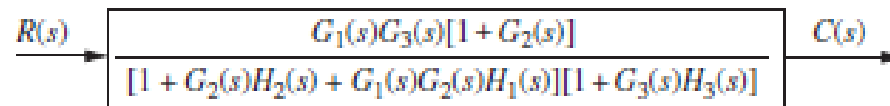


FIGURE 5.12 Steps in the block diagram reduction for Example 5.2



Exercise 5.1 Find the equivalent transfer function,.

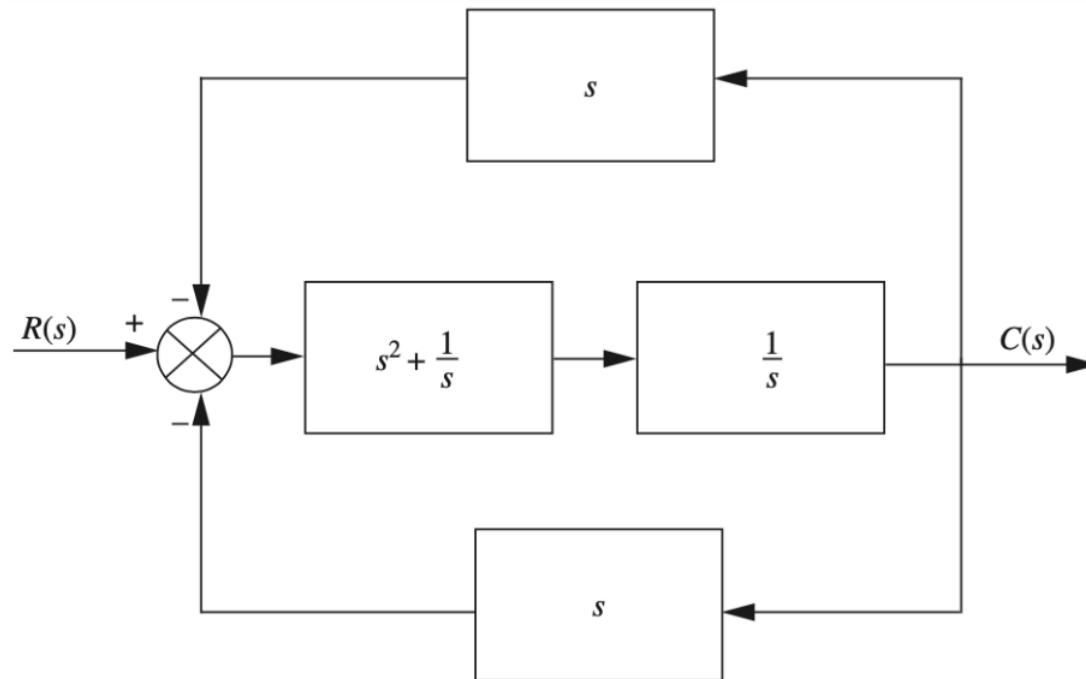


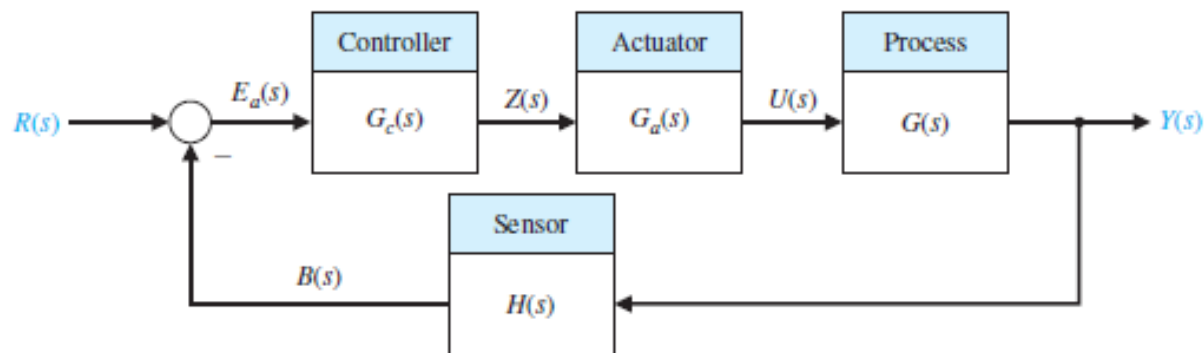
FIGURE 5.13 Block diagram for Skill-Assessment Exercise 5.1[Nise, 2015]

Answer

Combine the parallel blocks in the forward path. Then, push $\frac{1}{s}$ to the left past the pickoff point.

$$T(s) = \frac{s^3 + 1}{2s^4 + s^2 + 2s}$$

3. Analysis and Design of Feedback Systems



$$\frac{Y(s)}{R(s)} = \frac{G(s)G_a(s)G_c(s)}{1 + G(s)G_a(s)G_c(s)H(s)}$$

Example 5.3: It explains analysis and design of feedback systems that reduce to second-order systems. Percent overshoot, settling time, peak time, and rise time can then be found from the equivalent transfer function. ($a=5$)

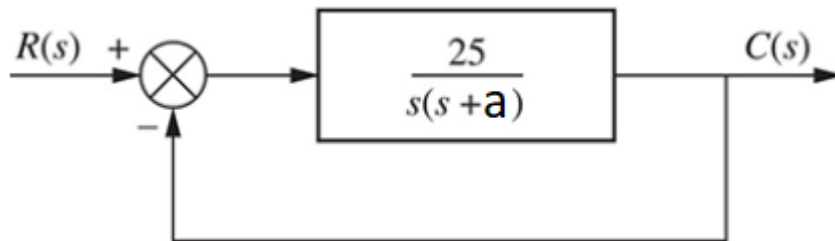


FIGURE 5.15 Feedback system for Example 5.3
[Nise, 2015]

$$T(s) = \frac{25}{s^2 + 5s + 25}$$

$$\omega_n = \sqrt{25} = 5$$

$$2\zeta\omega_n = 5 \Rightarrow \zeta = 0.5$$

$$T_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}} = 0.726 \text{ second}$$

$$\%OS = e^{-\zeta\pi/\sqrt{1-\zeta^2}} \times 100 = 16.303$$

$$T_s = \frac{4}{\zeta\omega_n} = 1.6 \text{ seconds}$$

```
clear; close all;
a=5; % change a to yield desired %OS
G=tf(25,poly([0,-a]));
T=feedback(G,1);
[numt, dent]=tfdata(T,'v');
wn=sqrt(dent(3)) % natural frequency
z=dent(2)/(2*wn) % zeta
Ts=4/(z*wn)
Tp=pi/(wn*sqrt(1-z^2))
percent OS=exp(-z*pi/sqrt(1-z^2))*100
Tr=(1.76*z^3-0.417*z^2 + 1.039*z + 1)/wn
step(T),grid
```

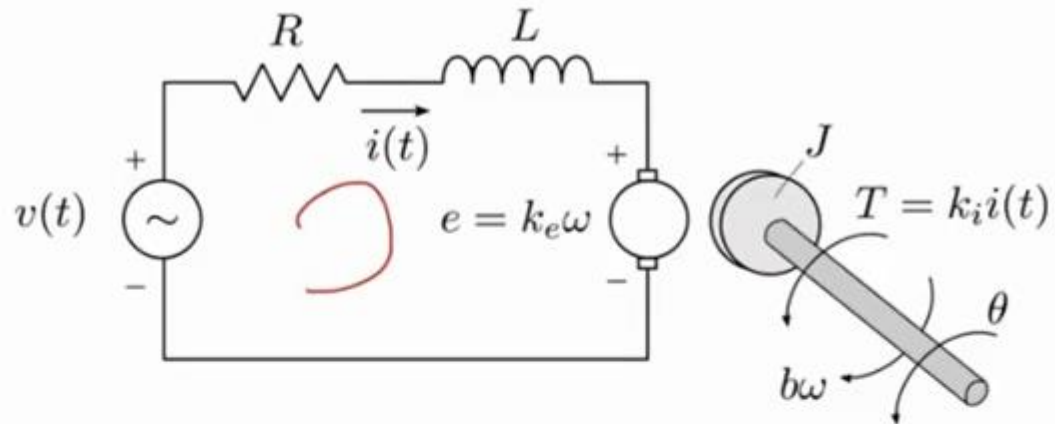
Block diagram of a DC motor

$$I(s) = \frac{V(s) - V_b(s)}{Ls + R} \quad \text{e}$$

$$\omega(s) = \frac{T(s) - T_d(s)}{Js + b}$$

$$T(s) = k_i I(s)$$

$$V_m(s) = k_m \omega(s)$$



$$\rightarrow V(s) = (R + Ls)I(s) + \omega(s)k_m \rightarrow I(s) = \frac{V(s) - \omega(s)k_m}{Ls + R}$$

$$\rightarrow T(s) = (Js + b)\omega(s) + T_d(s) \rightarrow \omega(s) = \frac{I(s)k_i - T_d(s)}{Js + b}$$

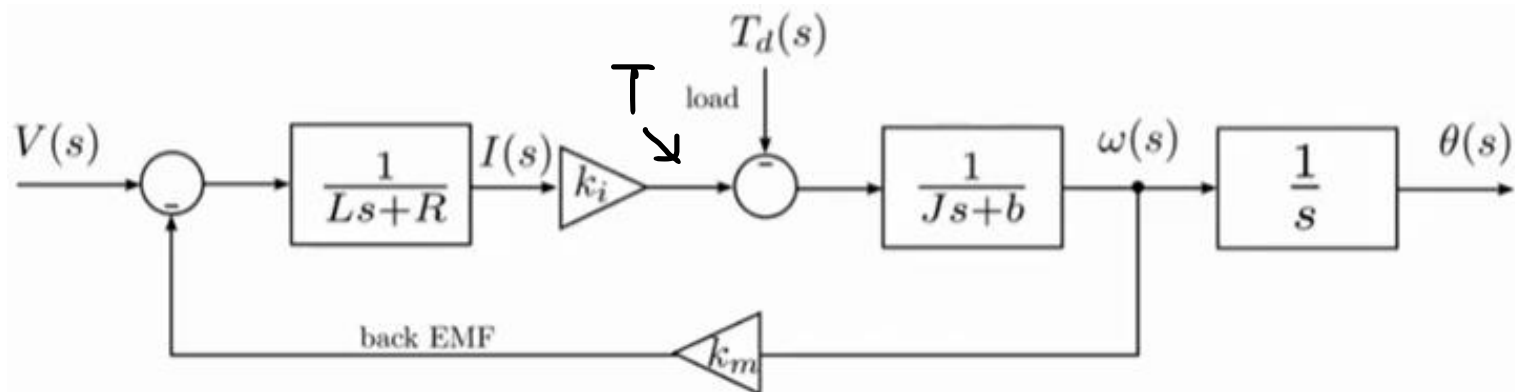
Under steady-state condition:

We have two equations. One describes the electrical and one describes the mechanical.

$$\omega = \frac{k_i V - RT_d}{k_i k_m + Rb},$$

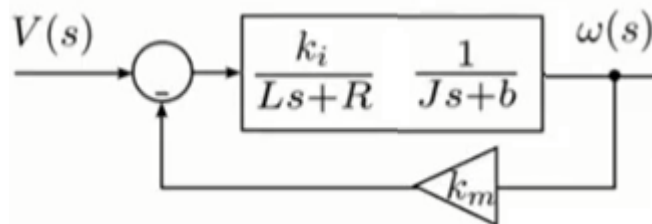
$$T = \frac{k_i (Vb + k_m T_d)}{k_m k_i + Rb}$$

Transfer Function



Speed to voltage transfer function for $T_d(s) = 0$

$$S(s) = \frac{\omega(s)}{V(s)}$$

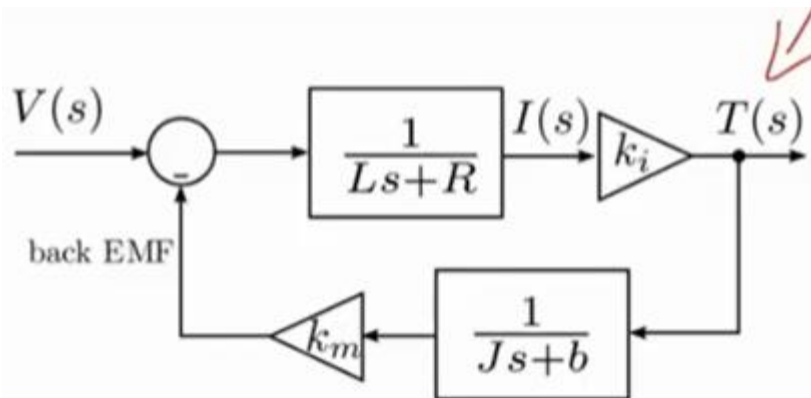


$$S(s) = \frac{\omega(s)}{V(s)} = \frac{k_i}{(Ls + R)(Js + b) + k_i k_m}$$

Position to voltage transfer function for $T_d(s) = 0$

$$P(s) = \frac{\theta(s)}{V(s)} = \frac{k_i}{s[(Ls + R)(Js + b) + k_i k_m]}$$

Transfer function



Torque constant	k_i	mNm/A
Speed constant	k_m	mV/(rad/s)
Resistance	R	Ω
Inductance	L	H
Rotor inertia	J	kg·m ²
Friction constant	b	Nm/(rad/s)

Torque to voltage transfer function for $T_d(s) = 0$.

$$M(s) = \frac{T(s)}{V(s)} = \frac{\frac{k_i}{Ls+R}}{1 + \frac{k_m}{Js+b} \frac{k_i}{Ls+R}}$$

$$M(s) = \frac{T(s)}{V(s)} = \frac{k_i(Js+b)}{(Js+b)(Ls+R) + k_i k_m} =$$

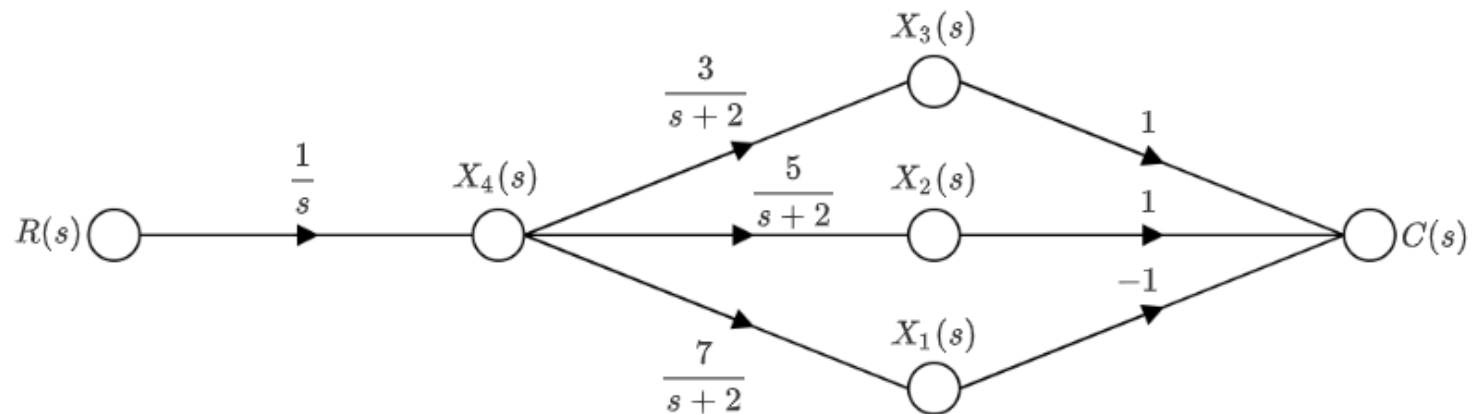
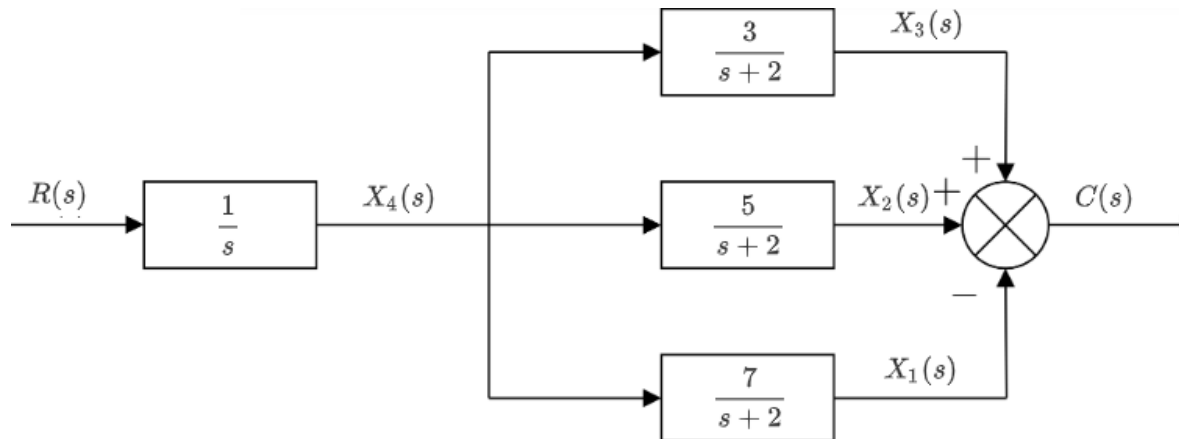
$S(s) (Js+b)$

4. Signal-Flow Graphs

- Extracting the transfer function from the system's block diagram is not a systematic procedure.
⇒ cannot be automated or programmed using a computer.
- The transfer function of a system with a complicated block diagram can be derived in a systematic fashion by transforming the block diagram into a **signal-flow graph** ; by applying Mason's formula.
- A block diagram can be transformed into a signal-flow graph by :
 1. transforming every signal (link) in the block diagram to a **node** in the signal-flow graph.
 2. Transforming each block (local TF) in the block diagram to a **directed link** in the signal-flow graph.

Definitions

- A **source** node is a node from which signals flow only away from it (has no incoming signal).
In a signal-flow graph, source nodes correspond to the system inputs and initial conditions.
- A **sink** node is a node that has only incoming signals (branches) and no departing signals.
- A **path** is a continuous connection of branches from one node to another following the signal flow direction.
- A **forward path** is path that connects a source node to a sink node without visiting any node more than once.
- Two paths are said to be **nontouching** if they have no node in common.
- A **path gain** is the product of the transfer functions of all the branching forming the path.
- A **loop** is a closed path in which no node is visited more than once (departure node = arrival node).



5.5 Mason's Rule

Definitions

- **Loop gain.** The product of branch gains found by traversing a path that starts at a node and ends at the same node

1. $G_2(s)H_1(s)$
2. $G_4(s)H_2(s)$
3. $G_4(s)G_5(s)H_3(s)$
4. $G_4(s)G_6(s)H_3(s)$

- **Forward-path gain.** The product of gains found by traversing a path from the input node to the output node of the signal-flow graph in the direction of signal flow.

1. $G_1(s)G_2(s)G_3(s)G_4(s)G_5(s)G_7(s)$
2. $G_1(s)G_2(s)G_3(s)G_4(s)G_6(s)G_7(s)$

- **Nontouching loops.** Loops that do not have any nodes in common.

$G_2(s)H_1(s)$ does not touch loops

$G_4(s)H_2(s)$, $G_4(s)G_5(s)H_3(s)$, and $G_4(s)G_6(s)H_3(s)$.

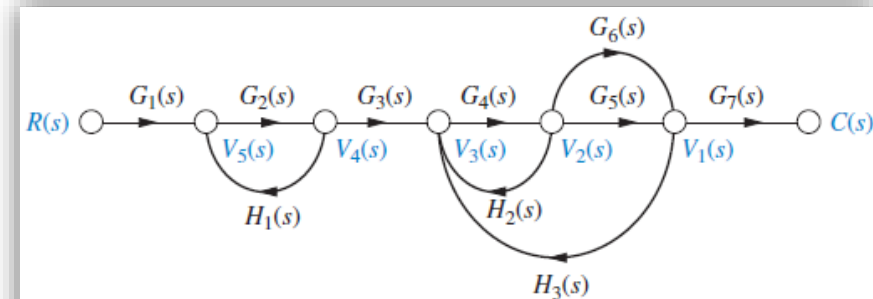


FIGURE 5.20 Signal-flow graph for demonstrating Mason's rule

- **Nontouching-loop gain.** The product of loop gains from nontouching loops taken two, three, four, or more at a time.

1. $[G_2(s)H_1(s)][G_4(s)H_2(s)]$
2. $[G_2(s)H_1(s)][G_4(s)G_5(s)H_3(s)]$
3. $[G_2(s)H_1(s)][G_4(s)G_6(s)H_3(s)]$

Mason's Rule

- The transfer function between a source $R(s)$ and a sink $C(s)$ in a signal-flow graph is formulated as :

$$G(s) = \frac{C(s)}{R(s)} = \frac{\sum_k T_k \Delta_k}{\Delta}$$

$k \equiv$ number of forward paths between $R(s)$ and $C(s)$

$T_k \equiv$ gain of forward path number k

$$\begin{aligned} \Delta = & 1 - (\sum \text{loop gains}) \\ & + (\sum \text{nontouching loop gains taken 2 at a time}) \\ & - (\sum \text{nontouching loop gains taken 3 at a time}) \\ & + \dots \end{aligned}$$

Δ_k is calculated in the same way as Δ but with dropping the loops that touch the forward path number k (those having at least one common node with this path).

Example 5.7

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Find the transfer function, $C(s)/R(s)$, for the signal-flow graph

Solution

First, identify the *forward-path gains*

$$G_1(s)G_2(s)G_3(s)G_4(s)G_5(s)$$

Second, identify the *loop gains*. There are four, as follows:

1. $G_2(s)H_1(s)$
2. $G_4(s)H_2(s)$
3. $G_7(s)H_4(s)$
4. $G_2(s)G_3(s)G_4(s)G_5(s)G_6(s)G_7(s)G_8(s)$

Third, identify the *nontouching loops taken two at a time*.

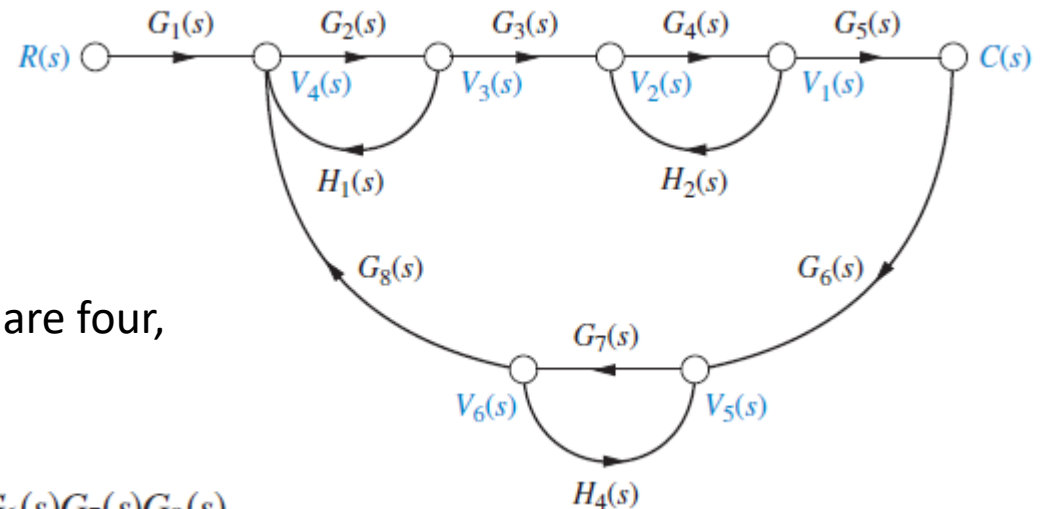
Loop 1 and loop 2: $G_2(s)H_1(s)G_4(s)H_2(s)$

Loop 1 and loop 3: $G_2(s)H_1(s)G_7(s)H_4(s)$

Loop 2 and loop 3: $G_4(s)H_2(s)G_7(s)H_4(s)$

Finally, the *nontouching loops taken three at a time* are as follows:

Loops 1, 2, and 3: $G_2(s)H_1(s)G_4(s)H_2(s)G_7(s)H_4(s)$



(Fig. 5.21 of [Nise, 2015])

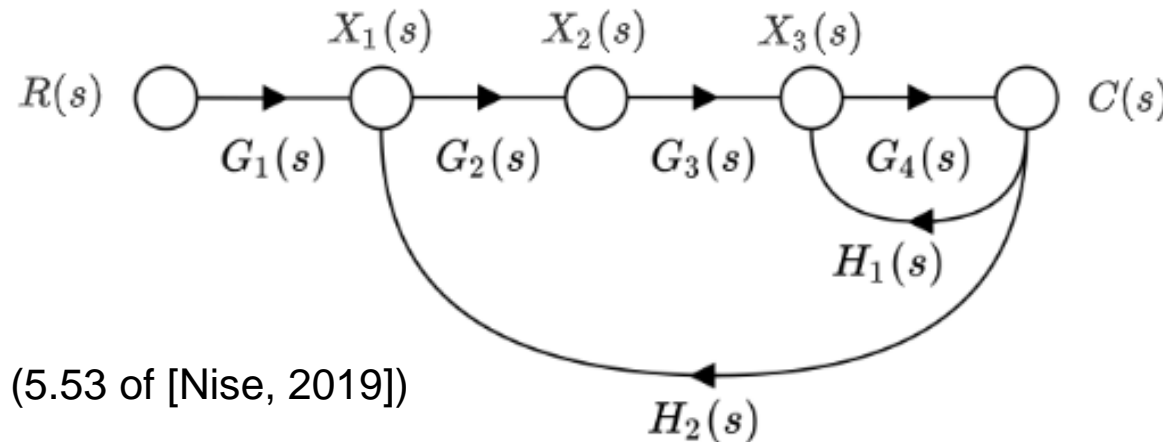
Now, we form Δ and Δ_k . Hence,

$$\begin{aligned}\Delta = 1 & - [G_2(s)H_1(s) + G_4(s)H_2(s) + G_7(s)H_4(s) \\ & \quad + G_2(s)G_3(s)G_4(s)G_5(s)G_6(s)G_7(s)G_8(s)] \\ & + [G_2(s)H_1(s)G_4(s)H_2(s) + G_2(s)H_1(s)G_7(s)H_4(s) \\ & \quad + G_4(s)H_2(s)G_7(s)H_4(s)] \\ & - [G_2(s)H_1(s)G_4(s)H_2(s)G_7(s)H_4(s)]\end{aligned}$$

We form Δ_k by eliminating from Δ the loop gains that touch the k th forward path:

$$\Delta_1 = 1 - G_7(s)H_4(s)$$

$$G(s) = \frac{T_1 \Delta_1}{\Delta} = \frac{[G_1(s)G_2(s)G_3(s)G_4(s)G_5(s)][1 - G_7(s)H_4(s)]}{\Delta}$$



(5.53 of [Nise, 2019])

Mason's Rule

$$G(s) = \frac{C(s)}{R(s)} = \frac{\sum_k T_k \Delta_k}{\Delta}$$

 $k = \text{forward paths} = 1$

$$T_k = T_1 = G_1(s)G_2(s)G_3(s)G_4(s)$$

$$\Delta = 1 - \Sigma(\text{loop gains}) + 0$$

$$= 1 - (G_2(s)G_3(s)G_4(s)H_2(s) + G_4(s)H_1(s))$$

Forward-path gain

$$G_1(s)G_2(s)G_3(s)G_4(s)$$

Loop gain 1

$$G_2(s)G_3(s)G_4(s)H_2(s)$$

Loop gain 2

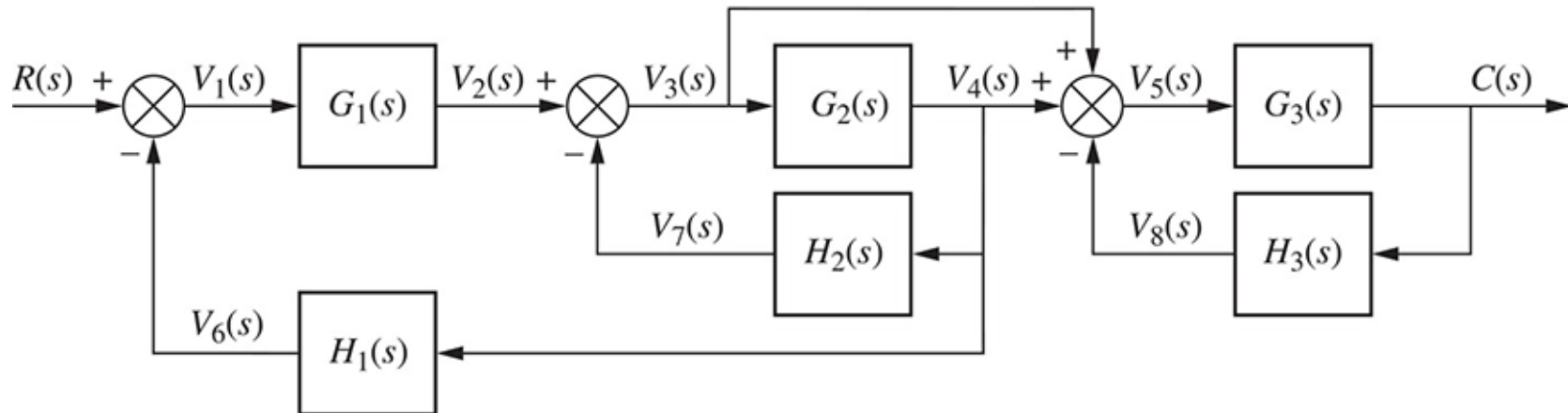
$$G_4(s)H_1(s)$$

 $\Delta_k = \Delta_1 = \text{eliminating from } \Delta \text{ the loop gains touching the forward path}$

$$= 1$$

$$G(s) = \frac{G_1(s)G_2(s)G_3(s)G_4(s)(1)}{1 - (G_2(s)G_3(s)G_4(s)H_2(s) + G_4(s)H_1(s))}$$

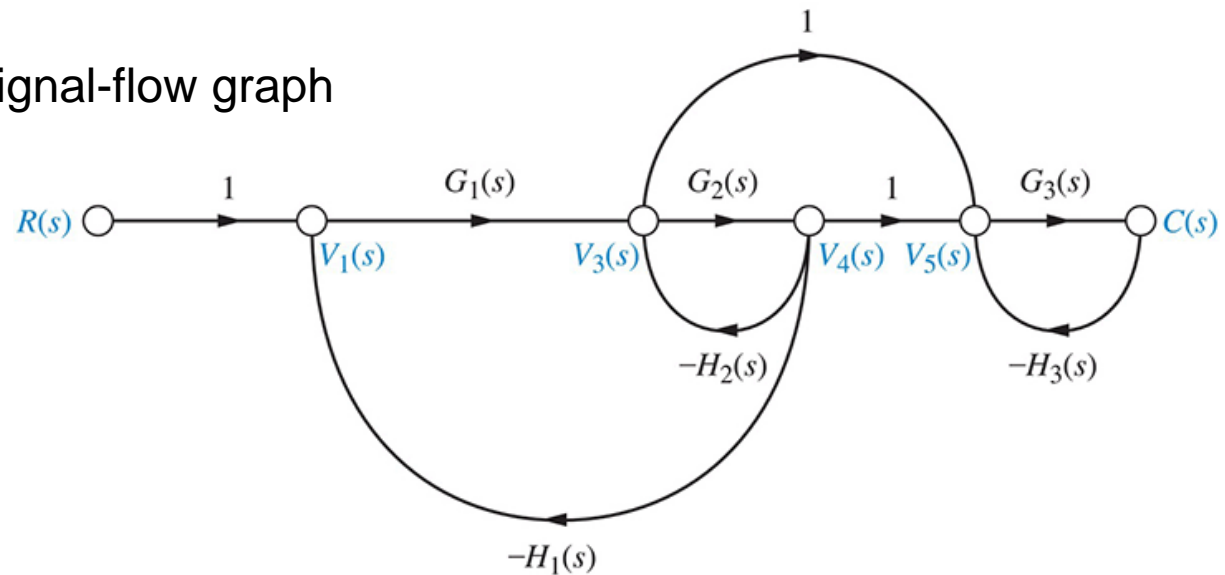
Example. Find the transfer function of the system using Mason's formula.

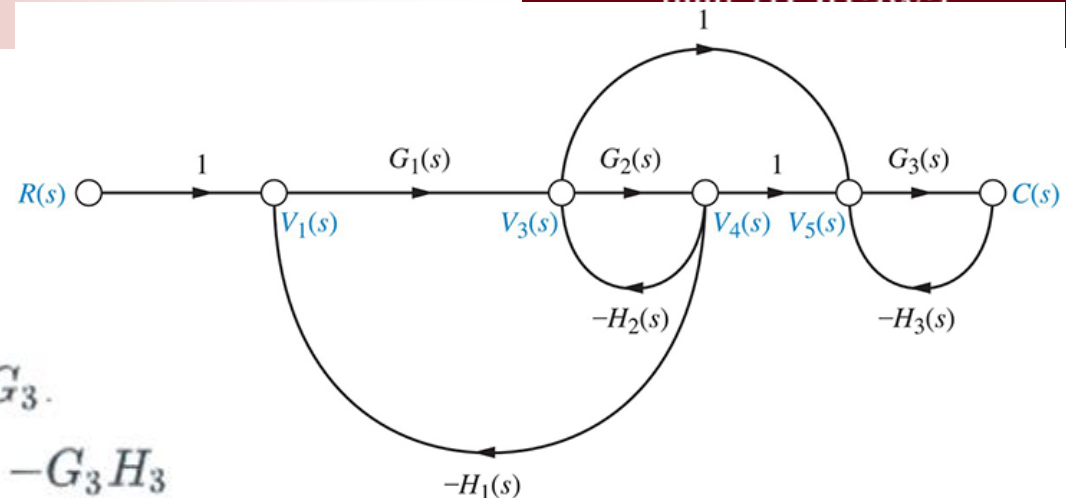


(Fig. 5.11 of [Nise, 2015])

Solution.

First convert it to signal-flow graph





Forward-path gains are $G_1 G_2 G_3$ and $G_1 G_3$.

Loop gains are $-G_1 G_2 H_1$, $-G_2 H_2$, and $-G_3 H_3$

Nontouching loops are $[-G_1 G_2 H_1][-G_3 H_3] = G_1 G_2 G_3 H_1 H_3$ and $[-G_2 H_2][-G_3 H_3] = G_2 G_3 H_2 H_3$.

$\Delta = 1 - \Sigma$ loop gains + Σ nontouching-loop gains taken two at a time $- \Sigma$ nontouching-loop gains taken three at a time + Σ nontouching-loop gains taken four at a time $- \dots$

$$\Delta = 1 + G_1 G_2 H_1 + G_2 H_2 + G_3 H_3 + G_1 G_2 G_3 H_1 H_3 + G_2 G_3 H_2 H_3$$

Finally, $\Delta_1 = 1$ and $\Delta_2 = 1$.

Substituting these values into

$$T(s) = \frac{C(s)}{R(s)} = \frac{\sum T_k \Delta_k}{\Delta}$$

$$T(s) = \frac{G_1(s)G_3(s) [1 + G_2(s)]}{[1 + G_2(s)H_2(s) + G_1(s)G_2(s)H_1(s)] [1 + G_3(s)H_3(s)]}$$

Exercise 2

Find the transfer function of the system in Fig. 6 using Mason's formula.

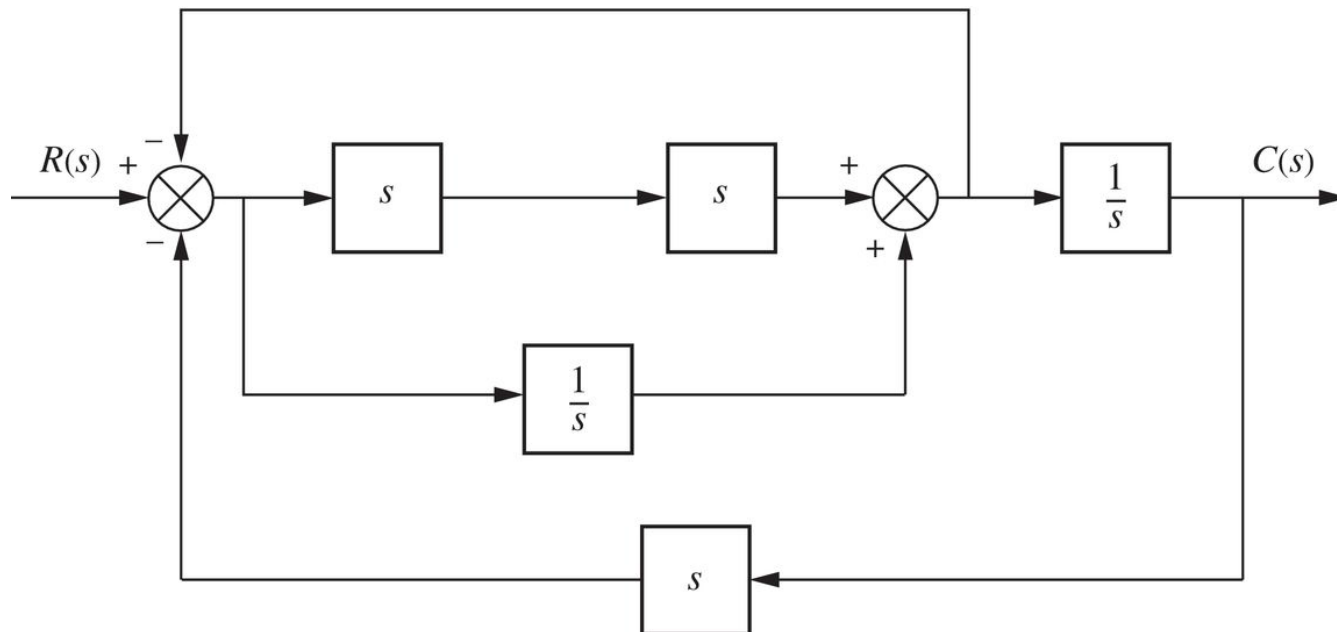


Figure 6: (Fig. 5.13 of [Nise, 2015])

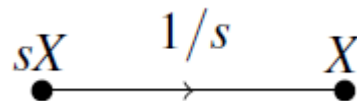
Answer
$$T(s) = \frac{s^3 + 1}{2s^4 + s^2 + 2s}$$

6. Signal-Flow Graphs of State Equations

Question

Knowing the state space representation of a system, how to draw its signal-flow graph ?

1. Create a node for each state variable.
2. To the left of each state variable node, create a node that corresponds to its time-derivative.
3. Connect each pair of these nodes with links of the following form :



4. Create nodes corresponding to the system's inputs and outputs.
5. Use the state and output equations to interconnect the nodes of the graph.

Example 5

An LTI system is represented by the following state-space model :

$$\dot{x}_1 = 2x_1 - 5x_2 + 3x_3 + 2r$$

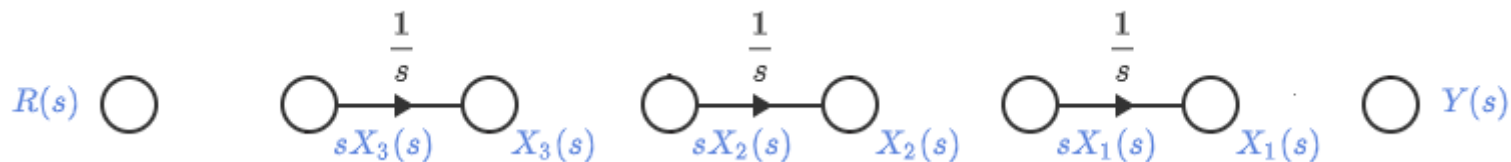
$$\dot{x}_2 = -6x_1 - 2x_2 + 2x_3 + 5r$$

$$\dot{x}_3 = x_1 - 3x_2 - 4x_3 + 7r$$

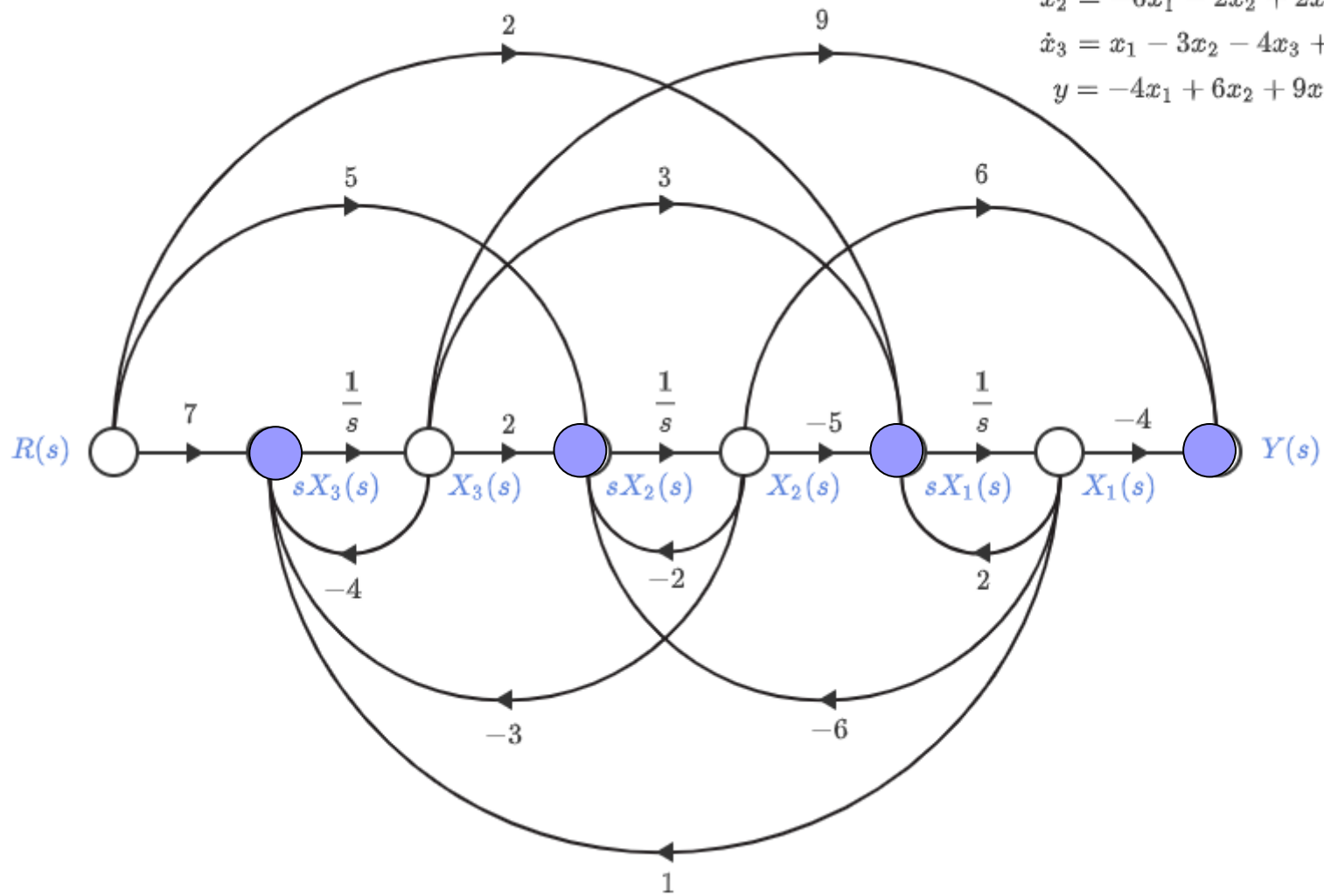
$$y = -4x_1 + 6x_2 + 9x_3$$

Find the system's transfer function using Mason's formula.

1. For the state equations above, nodes are drawn for state variables x_1, x_2 and x_3 in the frequency domain. A node $sX_k(s)$ for each derivative \dot{x}_k is inserted to the left of the state variable. Finally, Input $R(s)$ and output $Y(s)$ nodes are added.



$$\begin{aligned}\dot{x}_1 &= 2x_1 - 5x_2 + 3x_3 + 2r \\ \dot{x}_2 &= -6x_1 - 2x_2 + 2x_3 + 5r \\ \dot{x}_3 &= x_1 - 3x_2 - 4x_3 + 7r \\ y &= -4x_1 + 6x_2 + 9x_3\end{aligned}$$



Exercise 3

Draw the signal-flow graph corresponding to the following state space model :

$$\dot{x} = \begin{bmatrix} -2 & 1 & 0 \\ 0 & -3 & 1 \\ -3 & -4 & -5 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} r$$
$$y = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix} x$$

7. Alternative Representations in State Space

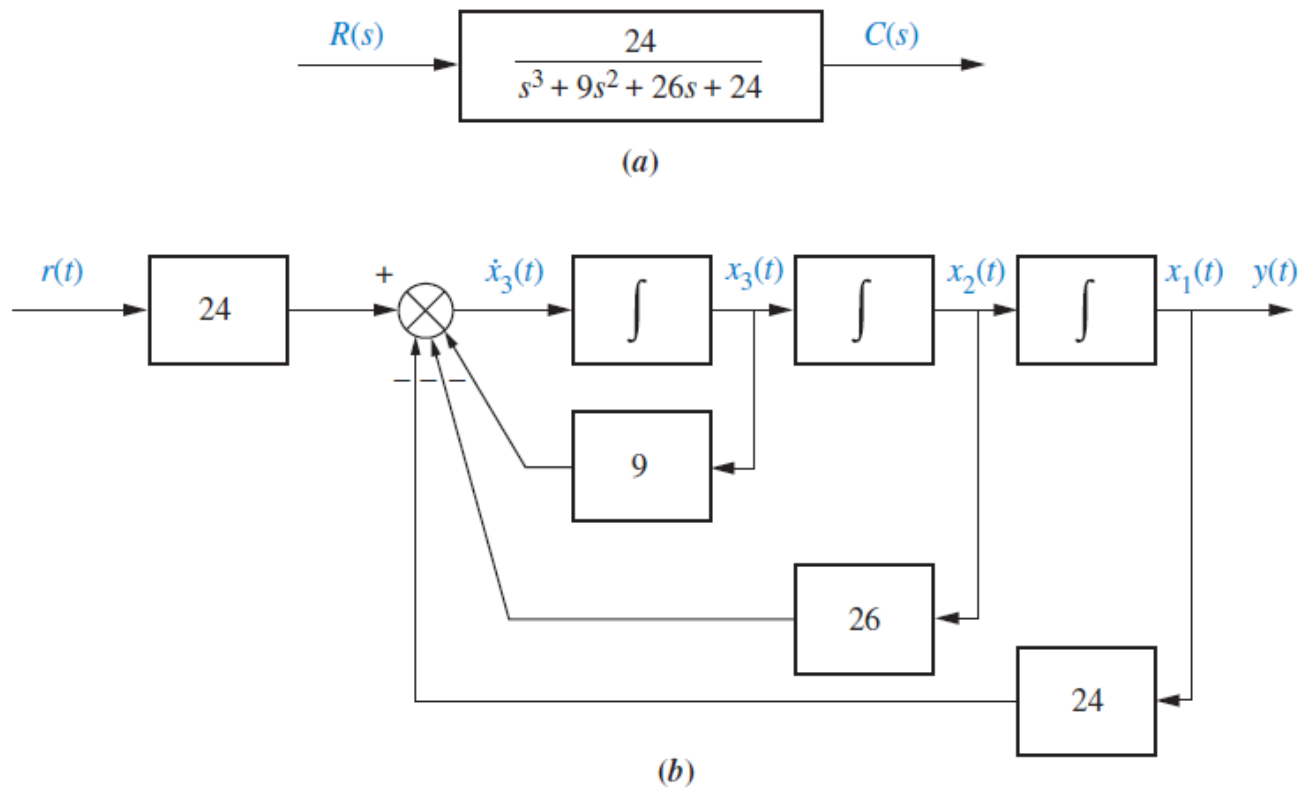


Figure 5: (Fig. 3.10 of [Nise, 2015])

Problem 5.25 (modified)

An LTI system has the following transfer function :

$$\frac{C(s)}{R(s)} = \frac{20}{s(s-2)(s+5)(s+8)}$$

Express the system in a state-space representation (phase-variable form) and draw its corresponding signal-flow graph.

Exercise 4

- In Chapter 2, the transfer function $G(s) = V_L(s)/V(s)$ of the circuit in Fig. was derived straight from the system's dynamic equations using (i) mesh analysis, and (ii) nodal analysis techniques. It was found to be :

$$G(s) = \frac{V_L(s)}{V(s)} = \frac{(Ls + R)^2}{L^2s^2 + 5RLs + 2R^2} \quad L = 1 \text{ H}, R = 1 \Omega$$

- Repeat the exercise by : deriving a block diagram for the system → Signal flow graph → Mason's formula.

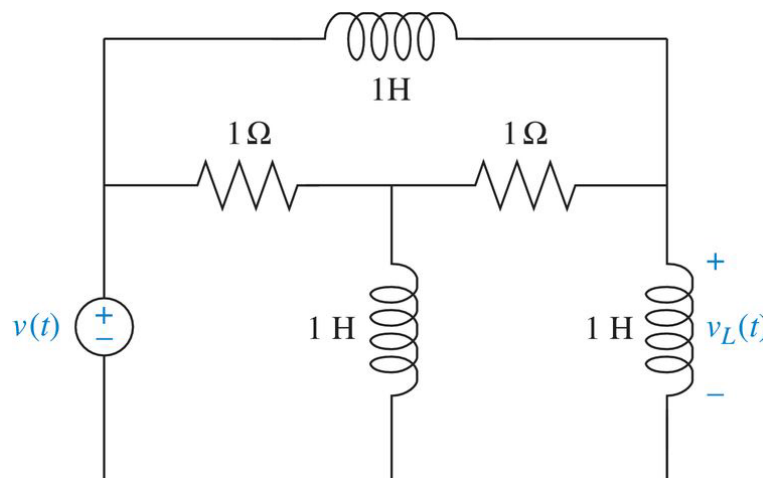


Figure: Circuit of Skill-Assessment Exercise 2.6 [Nise, 2015].

Summary of the modeling techniques

Fig. shows a summary of the modeling techniques covered so far.

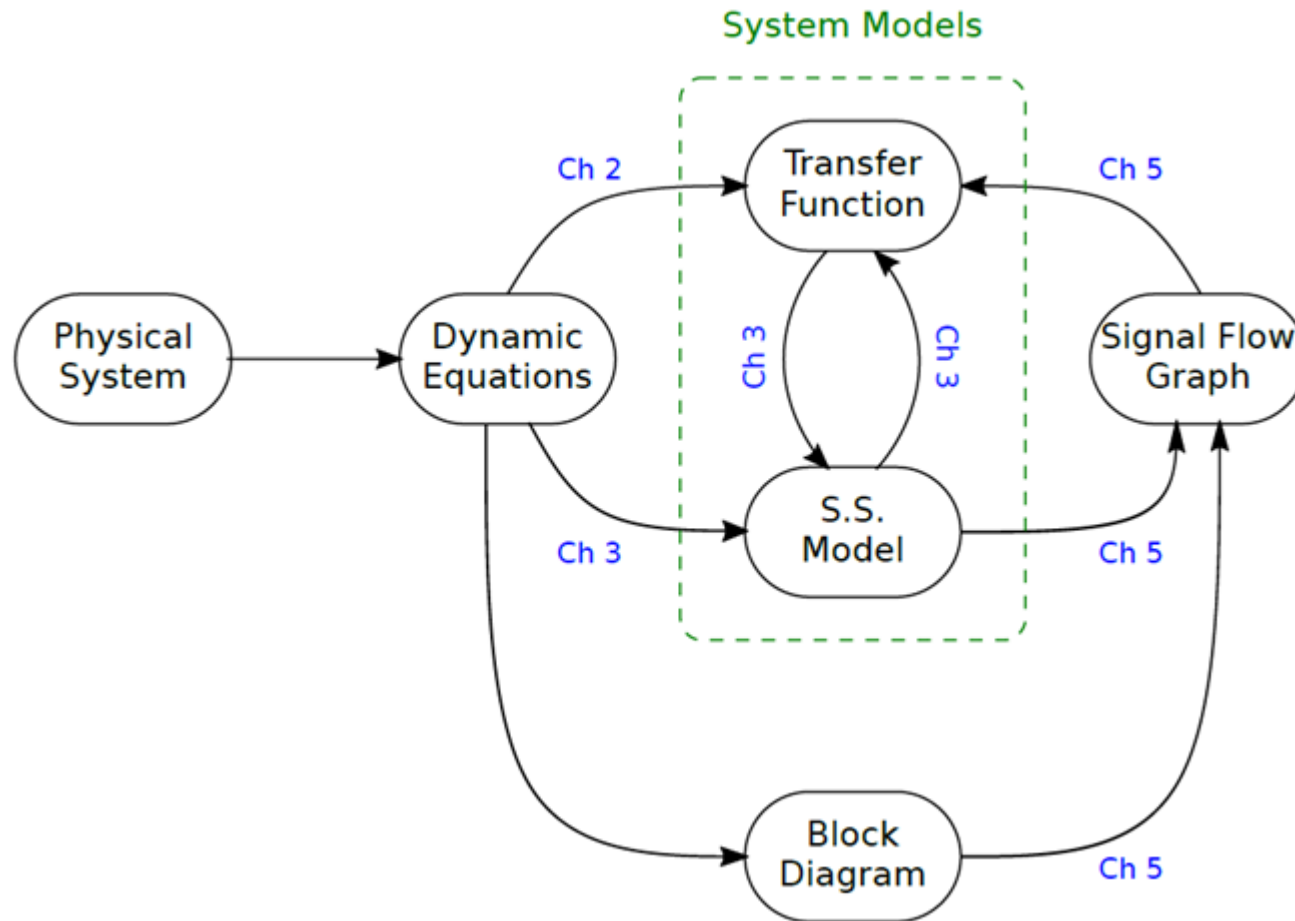


Figure : Summary of the modeling techniques

References



Nise, N. S. (2015).
Control Systems Engineering.
Wiley, 7 edition.