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Abstract

Keywords:

1. Problem Formulation

Given Q adjacent narrow-band emitters and L space separated base-stations which intercept the transmitted signals. Every base-station is equipped with an antenna array consisting of M elements. To avoid the angle ambiguity, the carrier frequency is small than the inverse of the propagation time over the array aperture. Denote the qth emitters position by the vector of coordinates $p_q \in \mathbb{R}^{D \times 1}$ (In general, D = 2 for plane geometry or D = 3 for solid geometry). The complex envelopes of the signals observed by the jth base- station array are given by

$$\boldsymbol{y}_{j}(t) = \sum_{q=1}^{Q} \alpha_{q,j} \boldsymbol{a}_{j}(\boldsymbol{p}_{q}) s_{q}(t - \tau_{j}(\boldsymbol{p}_{q}) - \mathring{t}_{q}) + \boldsymbol{w}_{j}(t), \quad 0 \leq t \leq T$$
 (1)

where $y_j(t)$ is the jth received signal, α_q, j is an unknown complex scalar representing the channel attenuation between the qth emitter and the jth base-station, $a_j(p_q)$ is the jth array response to the signal transmitted from position

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 p_q , and $s_q(t - \tau_j(p_q) - \mathring{t}_q)$ is the copy of the qth transmitted signal waveform, which is transmitted at time \mathring{t}_q and is received by the jth base-station after being delayed by $\tau_j(p_q)$. The vector $\boldsymbol{w}_j(t)$ represents noise and interference observed by the jth array. The received signal can be partitioned into L sections and the length of each section satisfies $T/L \gg \max \tau_j(p_q)$ to avoid the delay ambiguity. The emitters are assumed to be stationary during the total observation time T. Each section can be Fourier transformed and the result of this process is given by the following equation:

$$\mathbf{y}_{j}(n,l) = \sum_{q=1}^{Q} \alpha_{q,j} \mathbf{a}_{j}(\mathbf{p}_{q}) e^{-j2\pi f_{n}(\tau_{j}(\mathbf{p}_{q}) + \hat{t}_{q})} s_{q}(n,l) + \mathbf{w}_{j}(n,l)$$

$$j = 1, ..., N_{r}; n = 1, ..., N; l = 1, ..., L$$
(2)

where $\mathbf{y}_{j}(n,l)$ is the Fourier coefficient of the lth section of the jth received signal corresponding to frequency f_{n} , $s_{q}(n,l)$ is the nth Fourier coefficient of the lth section of the qth transmitted signal waveform, and $\mathbf{w}_{j}(n,l)$ represents the nth Fourier coefficient of the lth section of the noise waveform.

For compact representation, we define the following vectors:

$$\gamma_{j,q,n,l}(\boldsymbol{p}_q, \mathring{t}_q | s_q(n,l)) \triangleq \boldsymbol{a}_j(\boldsymbol{p}_q) e^{-j2\pi f_n(\tau_j(\boldsymbol{p}_q) + \mathring{t}_q)} s_q(n,l)$$
(3)

Here we assume that the spectrum of transmitted signals, like the synchronization and training sequences in cellular systems, are known to location system but the transmitted time \mathring{t}_q is unknown (which is assumed known in [1]). We observe that all information about the emitters position is embedded in the vector $\gamma_{j,q,n,l}$. This leads to the following representation of (2):

$$\mathbf{y}_{j}(n,l) = \sum_{q=1}^{Q} \alpha_{q,j} \boldsymbol{\gamma}_{j,q,n,l} + \mathbf{w}_{j}(n,l)$$
(4)

In matrix notation, (4) becomes

$$\mathbf{y}_{j}(n,l) = \mathbf{\Gamma}_{j,n,l}(\mathbf{P}, \dot{\mathbf{t}})\boldsymbol{\alpha}_{j} + \mathbf{w}_{j}(n,l)$$
 (5)

$$\Gamma_{j,n,l}(\boldsymbol{P}, \boldsymbol{\dot{t}}) \triangleq [\boldsymbol{\gamma}_{j,q,n,l}, ..., \boldsymbol{\gamma}_{j,q,n,l}] \in \mathbb{C}^{M \times Q}$$

$$\boldsymbol{\alpha}_{j} \triangleq [\alpha_{1,j}, ..., \alpha_{Q,j}]^{T} \in \mathbb{C}^{Q \times 1}$$
(6)

P, \dot{t} are the sets of emitters' positions and transmitted times respectively. Collecting and catenating all the N coefficients of the lth section of the jth received signal,

$$\mathbf{y}_{j}(l) = \mathbf{\Gamma}_{j,l}(\mathbf{P}, \dot{\mathbf{t}})\alpha_{j} + \mathbf{w}_{j}(l) \tag{7}$$

where

$$\mathbf{y}_{j}(l) \triangleq [\mathbf{y}_{j}^{T}(1, l), ..., \mathbf{y}_{j}^{T}(N, l)]^{T} \in \mathbb{C}^{NM \times 1}$$

$$\mathbf{\Gamma}_{j, l}(\mathbf{P}, \mathring{\mathbf{t}}) \triangleq [\mathbf{\Gamma}_{j, 1, l}^{T}(\mathbf{P}, \mathring{\mathbf{t}}), ..., \mathbf{\Gamma}_{j, N, l}^{T}(\mathbf{P}, \mathring{\mathbf{t}})]^{T} \in \mathbb{C}^{NM \times Q}$$

$$\mathbf{w}_{j}(l) \triangleq [\mathbf{w}_{j}^{T}(1, l), ..., \mathbf{w}_{j}^{T}(N, l)]^{T} \in \mathbb{C}^{NM \times 1}$$
(8)

The problem that we address now is stated as follow: Given the measurements $\{y_j(l)\}_{j,l=1}^{N_r,L}$, how to efficiently and accurately estimate the locations of the emitters P.

2. the Concentrated Likelihood Function

In this subsection, we will derive the Concentrated Likelihood Function (CLF) that depends on the parameters of interest P and the auxiliary parameters $\mathring{\boldsymbol{t}}$. The transmitted times are conducive to acquiring the location information from the propagation delay $\tau_j(p)$ since the transmitted times are the same among base-stations. In fact, since $\boldsymbol{w}_j(l)$ are independent among different sections of different received signals and $\boldsymbol{w}_j(l) \sim N(\boldsymbol{0}, \sigma^2 \boldsymbol{I}_{NM})$, it can be shown that the Log Likelihood Function (LLF) after dropping the constant terms is given by,

$$\mathcal{L}(\boldsymbol{P}, \mathring{\boldsymbol{t}}, \boldsymbol{\alpha}) \triangleq -\sum_{j=1}^{N_r} \sum_{l=1}^{L} \|\boldsymbol{y}_j(l) - \boldsymbol{\Gamma}_{j,l}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) \boldsymbol{\alpha}_j\|^2$$
(9)

where P, \dot{t}, α stand for any candidate values. Now, define the matrix

$$\mathbf{\Gamma}_{i}(\mathbf{P}, \mathbf{\dot{t}}) \triangleq [\mathbf{\Gamma}_{i,1}^{T}, ..., \mathbf{\Gamma}_{i,L}^{T}]^{T} \in \mathbb{C}^{LNM \times Q}$$
(10)

it can be shown that (11) is equivalent to

$$\mathcal{L}(\boldsymbol{P}, \boldsymbol{\dot{t}}, \boldsymbol{\alpha}) = -\sum_{j=1}^{N_r} \|\boldsymbol{y}_j - \boldsymbol{\Gamma}_j(\boldsymbol{P}, \boldsymbol{\dot{t}})\boldsymbol{\alpha}_j\|^2$$
(11)

where $\boldsymbol{y}_j \triangleq [\boldsymbol{y}_j^T(1),...,\boldsymbol{y}_j^T(L)]^T$. Maximizing (11) jointly with respect to all parameters is extremely challenging. Yet, significant computational savings follow from the observation that for any given $\boldsymbol{P}, \mathring{\boldsymbol{t}}$, the problem of finding the optimal $\{\boldsymbol{\alpha}_j\}_{j=1}^{N_r}$ becomes a linear least squares (LS) problem [2] whose solution is given by

$$\hat{\boldsymbol{\alpha}}_j = \boldsymbol{\Gamma}_j^{\dagger}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) \boldsymbol{y}_j \tag{12}$$

where Γ_j^{\dagger} is the Moore-Penrose pseudo-inverse of Γ_j given by $(\Gamma_j^H \Gamma_j)^{-1} \Gamma_j^H$. Note here that Γ_j has full column rank and, therefore, $(\Gamma_j^H \Gamma_j)^{-1}$ always exists. Now, by substituting $\{\hat{\alpha}_j\}_{j=1}^{N_r}$ for α back in (11) and resorting to some straightforward algebraic manipulations, we obtain the so-called *concentrated* likelihood function,

$$\mathcal{L}_c(\boldsymbol{P}, \boldsymbol{\dot{t}}) = \sum_{j=1}^{N_r} \boldsymbol{y}_j^H \boldsymbol{\Gamma}_j (\boldsymbol{\Gamma}_j^H \boldsymbol{\Gamma}_j)^{-1} \boldsymbol{\Gamma}_j^H \boldsymbol{y}_j$$
 (13)

The joint Maximum Likelihood (ML) estimates of \boldsymbol{P} and \boldsymbol{t} are then obtained as the solution to the following reduced-dimension optimization problem

$$[\hat{\boldsymbol{P}}, \hat{\boldsymbol{t}}] = \arg \max_{\boldsymbol{P}, \hat{\boldsymbol{t}}} \mathcal{L}_c(\boldsymbol{P}, \hat{\boldsymbol{t}})$$
 (14)

The reduced-dimension optimization problem (14) still has Q(D+1) unknown parameters (including the D-dimensional real-valued location vectors $\{p_q\}_{q=1}^Q$ and Q transmitted times $\{\mathring{t}_q\}_{q=1}^Q$) remain to estimate. The ML estimation will therefore requires a complex multidimensional search over the parameter space.

To solve the multidimensional optimization problem (14), the approach proposed in [1] is to decouple the problem into several *D*-dimensional optimization

problems under the assumption that the number of sections $L \to \infty$ or L is sufficiently large in practice. Considering the length of section has low limit $T/L \gg \max \tau_j(\boldsymbol{p}_q)$, the number of sections L is proportional to the total observation time T. The observation time T is, however, limited in some practical reasons, like the request for real-time processing or the assumption that emitters are stationary during the observation. It is, therefore, difficult to be satisfied that L is sufficiently large in some practical cases, which means the decouple approach proposed in [1] may be failure, especially when the emitters are adjacent to one another.

s 3. Global Maximization of The CLF

As done previously within other estimation problems (see [3, 4, 5, 6, 7]), we resort to the optimization framework, which is based on the Pincus' theorem and the powerful Importance Sampling (IS) method, to efficiently and accurately estimate the adjacent emitters' positions using limited observation time. First of all, we introduce the theorem of Pincus [8] as follows.

Theorem 1. Let $F(x_1,...,x_n) = F(\boldsymbol{x})$ be a continuous function on the closure of a bounded domain S in n-dimensional Euclidean space of \mathbb{R}^n . Assume that F attains a global maximum at exactly on point $\hat{\boldsymbol{x}} = [\hat{x}_1,...,\hat{x}_n]^T$ of S. Then for i = 1,...,n,

$$\hat{x}_i = \lim_{\rho \to \infty} \frac{\int \cdots \int x_i e^{\rho F(\mathbf{x})} dx_1 \cdots dx_n}{\int \cdots \int e^{\rho F(\mathbf{x})} dx_1 \cdots dx_n}$$
(15)

Applying this general result to our multidimensional optimization problem (14) with $\mathbf{x} \triangleq [\mathbf{P}, \mathbf{t}]$ and $F(\mathbf{x}) \triangleq \mathcal{L}_c(\mathbf{P}, \mathbf{t})$ leads to the following expressions for the required ML estimators,

$$\hat{\boldsymbol{p}}_{q}(d) = \int \cdots \int \boldsymbol{p}_{q}(d) \mathcal{F}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) d\boldsymbol{p}_{1}(1) \cdots d\mathring{t}_{Q}$$
(16)

$$\hat{t}_q^{(0)} = \int \cdots \int \mathring{t}_q \mathcal{F}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) d\boldsymbol{p}_1(1) \cdots d\mathring{t}_Q$$

$$q = 1, ..., Q; d = 1, ..., D$$

$$(17)$$

$$\mathcal{F}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) \triangleq \lim_{\rho \to \infty} \frac{e^{\rho \mathcal{L}_c(\boldsymbol{P}, \mathring{\boldsymbol{t}})}}{\int \cdots \int e^{\rho \mathcal{L}_c(\boldsymbol{P}, \mathring{\boldsymbol{t}})} d\boldsymbol{p}_1(1) \cdots d\mathring{t}_Q}$$
(18)

It seems that the multidimensional grid search we attempt to avoid is transformed into several multidimensional integrations (16) and (17) bearing the same computational load. By closely analyzing the properties of function (18), however, it turns out that $\mathcal{F}(\boldsymbol{P}, \dot{\boldsymbol{t}})$ can be seen as a Probability Density Function (pdf) since it is nonnegative and integrates to one. As a pdf, furthermore, $\mathcal{F}(\boldsymbol{P}, \dot{\boldsymbol{t}})$ indicates the probability of the candidate values being true. The estimators in (16) and (17) can be regarded as statistical expectations, i.e., we have,

$$\hat{\boldsymbol{p}}_{q}(d) \triangleq \mathbb{E}_{\boldsymbol{P}, \hat{\boldsymbol{t}}} \{ \boldsymbol{p}_{q}(d) \} \quad \hat{\boldsymbol{t}}_{q} \triangleq \mathbb{E}_{\boldsymbol{P}, \hat{\boldsymbol{t}}} \{ \hat{\boldsymbol{t}}_{q} \}$$
(19)

Consequently, the multidimensional integrations are transformed into the statistical expectations, which can be approximated by the sample mean. We will then explain the method to generate R realizations, $\{P^{(r)}\}_{r=1}^{R}$ and $\{\mathring{\boldsymbol{t}}^{(r)}\}_{r=1}^{R}$, using the joint pdf $\mathcal{F}(\boldsymbol{P},\mathring{\boldsymbol{t}})$ and approximate the expectations in (19) using the following sample means,

$$\hat{\boldsymbol{p}}_{q}(d) = \frac{1}{R} \sum_{r=1}^{R} \boldsymbol{p}_{q}^{(r)}(d) \quad \hat{\hat{t}}_{q} = \frac{1}{R} \sum_{r=1}^{R} \hat{t}_{q}^{(r)}$$
(20)

Clearly, as the number of realizations R increases, the two sample mean in (20) approach the statistical expectations in (19). Unfortunately, the extremely non-linear property of the pdf $\mathcal{F}(\mathbf{P}, \mathbf{\dot{t}})$ makes it impractical to generate $\{\mathbf{P}^{(r)}\}_{r=1}^{R}$ and $\{\mathbf{\dot{t}}^{(r)}\}_{r=1}^{R}$ using itself. To sidestep this problem, we can resort to the Importance Sampling (IS) method [3, 4] and rewrite (16) and (17) in the following equivalent forms,

$$\hat{\boldsymbol{p}}_{q}(d) = \int \cdots \int \boldsymbol{p}_{q}(d) \frac{\mathcal{F}(\boldsymbol{P}, \mathring{\boldsymbol{t}})}{\mathcal{G}(\boldsymbol{P}, \mathring{\boldsymbol{t}})} \mathcal{G}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) d\boldsymbol{p}_{1}(1) \cdots d\mathring{t}_{Q}$$
(21)

$$\hat{\hat{t}}_q = \int \cdots \int \hat{t}_q \frac{\mathcal{F}(\boldsymbol{P}, \hat{\boldsymbol{t}})}{\mathcal{G}(\boldsymbol{P}, \hat{\boldsymbol{t}})} \mathcal{G}(\boldsymbol{P}, \hat{\boldsymbol{t}}) d\boldsymbol{p}_1(1) \cdots d\hat{t}_Q$$
 (22)

where $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ is another pdf called *importance function* (IF). By introducing the IF, the multidimensional integrations in (21) and (22) are interpreted as statistical expectations of two transformed random variates (RVs), i.e.,

$$\hat{\boldsymbol{p}}_{q}(d) \triangleq \mathbb{E}_{\boldsymbol{P}, \boldsymbol{\mathring{t}}} \{ \eta(\boldsymbol{P}, \boldsymbol{\mathring{t}}) \boldsymbol{p}_{q}(d) \} \quad \hat{\mathring{t}}_{q} \triangleq \mathbb{E}_{\boldsymbol{P}, \boldsymbol{\mathring{t}}} \{ \eta(\boldsymbol{P}, \boldsymbol{\mathring{t}}) \mathring{t}_{q} \}$$
(23)

where $\eta(\mathbf{P}, \mathbf{t})$ is called *importance weight* (IW) corresponding to the realizations of \mathbf{P}, \mathbf{t} and defined as,

$$\eta(\boldsymbol{P}, \dot{\boldsymbol{t}}) \triangleq \frac{\mathcal{F}(\boldsymbol{P}, \dot{\boldsymbol{t}})}{\mathcal{G}(\boldsymbol{P}, \dot{\boldsymbol{t}})}$$
(24)

If $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ is carefully designed, the expectations in (23) can be computed at any desired degree of accuracy (by increasing R) using the corresponding sample mean estimates. In fact, the choice of $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ is a tradeoff between the ease and efficiency of sampling. The appropriate choice of the IF will be discussed in the following section.

4. Appropriate Choice of Importance Function

The appropriate $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ must be designed as close as possible to $\mathcal{F}(\boldsymbol{P}, \boldsymbol{t})$ while allowing at the same time the easy generation of the required parameters realizations $\{\boldsymbol{P}^{(r)}\}_{r=1}^R$ and $\{\boldsymbol{t}^{(r)}\}_{r=1}^R$. In order to facilitate the process of generating the required R parameter realizations, we constrain IF's form separable in terms of the Q position-transmitted time pairs $\{(\boldsymbol{p}_q, \boldsymbol{t}_q)\}_{q=1}^Q$. In other words, our ultimate goal is to design $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ in a way that allows it to be factorized as follows,

$$\mathcal{G}(\boldsymbol{P}, \mathring{\boldsymbol{t}}) \triangleq \prod_{q=1}^{Q} g_q(\boldsymbol{p}_q, \mathring{t}_q)$$
 (25)

This allow us to interpret the pdf of Q(D+1)-dimensional random vector as the joint pdf of Q independent (D+1)-dimensional random vectors. Hence, we can independently generate Q groups of realizations for (D+1)-dimensional random vectors $\{(\boldsymbol{p}_q^{(r)}, \mathring{t}_q^{(r)})\}_{r,q=1}^{R,Q}$, instead of generating realizations for Q(D+1)-dimensional random vector $(\boldsymbol{P}, \mathring{\boldsymbol{t}})$ directly using $\mathcal{G}(\boldsymbol{P}, \mathring{\boldsymbol{t}})$. We start with the CLF (13), the separable form (25) requests that the CLF has summation form with respect to q. Note that the product $\boldsymbol{y}_{j}^{H}\boldsymbol{\Gamma}_{j}$ in CLF is a Q-dimensional vector. Diagonalizing the inverse matrix $(\boldsymbol{\Gamma}_{j}^{H}\boldsymbol{\Gamma}_{j})^{-1}$ is, therefore, the key step to the separable form (25). According to the definition of $\boldsymbol{\Gamma}_{j}$ in (10), $\boldsymbol{\Gamma}_{j,l}$ in (8), $\boldsymbol{\Gamma}_{j,n,l}$ in (6) and $\boldsymbol{\gamma}_{j,q,n,l}$ in (3), the (u,v)th element of the matrix product $\boldsymbol{\Gamma}_{j}^{H}\boldsymbol{\Gamma}_{j}$ is given by,

$$\begin{split} [\boldsymbol{\Gamma}_{j}^{H} \boldsymbol{\Gamma}_{j}]_{u,v} &= \sum_{l=1}^{L} \sum_{n=1}^{N} \boldsymbol{\gamma}_{j,q,n,l}^{H} \boldsymbol{\gamma}_{j,q,n,l} \\ &= \underbrace{\boldsymbol{a}_{j}^{H} (\boldsymbol{p}_{u}) \boldsymbol{a}_{j} (\boldsymbol{p}_{u})}_{r_{doa,j}^{(u,v)}} \underbrace{\sum_{n=1}^{N} e^{j2\pi f_{n}(\tau_{j}(\boldsymbol{p}_{u}) + \hat{t}_{u} - \tau_{j}(\boldsymbol{p}_{v}) - \hat{t}_{v})}}_{r_{toa,j}^{(u,v)}} \underbrace{\sum_{l=1}^{L} s_{u}^{*}(n,l) s_{v}(n,l)}_{r_{s}^{(u,v)}} \end{split}$$

where $r_{doa,j}^{(u,v)}$, $r_{toa,j}^{(u,v)}$, $r_s^{(u,v)}$ denote the (u,v)th non-normalized correlation coefficients in aspects of Direction Of Arrival (DOA) and Time Of Arrival (TOA) at the jth base-station and the transmitted signal, respectively. The matrix $\mathbf{\Gamma}_j^H \mathbf{\Gamma}_j$ is, therefore, a correlation matrix among all emitters. To avoid additional computational load, we directly assume, instead of strict diagonalizing, $\mathbf{\Gamma}_j^H \mathbf{\Gamma}_j$, as diagonal matrix, at the cost of losing the correlation information. The diagonal matrix, for $j=1,...,N_r$, is given by,

$$\mathbf{\Gamma}_{j}^{H}\mathbf{\Gamma}_{j} = MNL \cdot \mathbf{I} \tag{26}$$

where I is the identity matrix, the coefficient MNL is the diagonal element of $\Gamma_j^H \Gamma_j$.

substituting (26) into (13) and ignoring the constant coefficient, we get its separated form as follows,

$$\tilde{\mathcal{L}}_c(\boldsymbol{P}, \boldsymbol{\dot{t}}) = \sum_{j=1}^{N_r} \|\boldsymbol{\Gamma}_j^H \boldsymbol{y}_j\|^2$$
(27)

$$=\sum_{q=1}^{Q}\tilde{\ell}_{c}(\boldsymbol{p}_{q},\mathring{t}_{q})\tag{28}$$

$$\tilde{\ell}_c(\boldsymbol{p}_q, \mathring{t}_q) \triangleq \sum_{j,n,l=1}^{N_r, N, L} |\boldsymbol{\gamma}_{j,q,n,l}^H \boldsymbol{y}_j(n, l)|^2$$
(29)

Replacing $\mathcal{L}_c(\boldsymbol{P}, \boldsymbol{t})$ and ρ in (18) with $\tilde{\mathcal{L}}_c(\boldsymbol{P}, \boldsymbol{t})$ and an appropriate ρ_2 , we get the separated form of $\mathcal{G}(\boldsymbol{P}, \boldsymbol{t})$ (25) and the qth component is given by,

$$g_q(\mathbf{p}_q, \mathring{t}_q) \triangleq \frac{e^{\rho_2 \tilde{\ell}_c(\mathbf{p}_q, \mathring{t}_q)}}{\int \int e^{\rho_2 \tilde{\ell}_c(\mathbf{p}_q, \mathring{t}_q)} d\mathbf{p}_q d\mathring{t}_q}$$
(30)

The generation of qth group of realizations $\{(\boldsymbol{p}_q^{(r)}, \mathring{t}_q^{(r)})\}_{r=1}^R$ follows, furthermore, the well-known general result from probability theory, which is that the joint pdf, $g_q(\boldsymbol{p}_q,\mathring{t}_q)$, can be factorized as the product of marginal and conditional pdfs,

$$g_q(\boldsymbol{p}_q, \mathring{t}_q) = g_q^{\text{mar}}(\boldsymbol{p}_q) g_q^{\text{con}}(\mathring{t}_q | \boldsymbol{p}_q)$$
(31)

The usual method to obtain the marginal pdf $g_q^{\rm mar}(\boldsymbol{p}_q)$ is integrating the joint pdf $g_q(\boldsymbol{p}_q,\mathring{t}_q)$ with respect to the transmitted time \mathring{t}_q . The analytical integration is, however, difficult since $g_q(\boldsymbol{p}_q,\mathring{t}_q)$ has very high non-linear about \mathring{t}_q . The discrete approximation is also involving D+1-dimensional grid search. We have another alternative method to construct $g_q^{\rm mar}(\boldsymbol{p}_q)$ directly and efficiently,

$$g_q^{\text{mar}}(\boldsymbol{p}_q) \triangleq \frac{e^{\rho_1 \ell(\boldsymbol{p}_q)}}{\int e^{\rho_1 \ell(\boldsymbol{p}_q)} d\boldsymbol{p}_q}$$
(32)

where $\ell(p_q)$ is the single (Q=1) Likelihood Function (SLF) of the position p_q , which is derived from (29) by eliminating the transmitted time \mathring{t}_q . We will, then, describe the derivation in details.

Our derivation starts with a vector definition as follows,

$$\boldsymbol{\gamma}_{q,j,l} \triangleq \boldsymbol{\Phi}_{q,j,l}(\boldsymbol{p}_q)\boldsymbol{b}(\mathring{t}_q) \tag{33}$$

$$\Phi_{q,j,l}(\boldsymbol{p}_q) \triangleq \Lambda_j(\boldsymbol{p}_q, s_q(l)) \otimes \boldsymbol{a}_j(\boldsymbol{p}_q)$$
(34)

$$\boldsymbol{b}(\mathring{t}_q) \triangleq [e^{-j2\pi f_1\mathring{t}_q}, ..., e^{-j2\pi f_N\mathring{t}_q}]^T$$
(35)

$$\mathbf{\Lambda}_{j}(\mathbf{p}_{q}, s_{q}(l)) \triangleq \begin{bmatrix}
s_{q}(1, l)e^{-j2\pi f_{1}\tau_{j}(\mathbf{p}_{q})} & \dots & 0 \\
\vdots & \ddots & \vdots \\
0 & \dots & s_{q}(N, l)e^{-j2\pi f_{N}\tau_{j}(\mathbf{p}_{q})}
\end{bmatrix}$$
(36)

substituting (33) and (34) into (29), we get

$$\tilde{\ell}_c(\boldsymbol{p}_q, \mathring{t}_q) = \sum_{i=1}^{N_r} \sum_{l=1}^{L} \boldsymbol{\gamma}_{j,q,l}^H \boldsymbol{y}_j(l) \boldsymbol{y}_j^H(l) \boldsymbol{\gamma}_{j,q,l}$$
(37)

$$= \boldsymbol{b}^{H}(\mathring{t}_{q}) \underbrace{\left(\sum_{j=1}^{N_{r}} \sum_{l=1}^{L} \boldsymbol{\Phi}_{q,j,l}^{H}(\boldsymbol{p}_{q}) \boldsymbol{y}_{j}(l) \boldsymbol{y}_{j}^{H}(l) \boldsymbol{\Phi}_{q,j,l}(\boldsymbol{p}_{q})\right)}_{\triangleq \boldsymbol{\Pi}(\boldsymbol{p}_{q})} \boldsymbol{b}(\mathring{t}_{q})$$
(38)

The right hand of the second equation is a non-normalized Rayleigh Quotient with respect to \mathring{t}_q . Consequently, maximizing $\tilde{\ell}_c(\boldsymbol{p}_q,\mathring{t}_q)$ with respect to \mathring{t}_q , we get,

$$\ell(\boldsymbol{p}_q) \triangleq \lambda_{\max}(\boldsymbol{\Pi}(\boldsymbol{p}_q)) \tag{39}$$

Substituting (39) into (32), we get the marginal pdf of the qth emitter's position,

$$g_q^{\text{mar}}(\boldsymbol{p}_q) = \frac{e^{\rho_1 \lambda_{\text{max}}(\boldsymbol{\Pi}(\boldsymbol{p}_q))}}{\int e^{\rho_1 \lambda_{\text{max}}(\boldsymbol{\Pi}(\boldsymbol{p}_q))} d\boldsymbol{p}_q} \tag{40}$$

Given position realization $p_q^{(r)}$ generated by $g_q^{\max}(p_q)$, the conditional pdf is defined as,

$$g_q^{\text{con}}(\mathring{t}_q|\boldsymbol{p}_q^{(r)}) = \frac{g_q(\boldsymbol{p}_q^{(r)},\mathring{t}_q)}{\int_{\mathring{t}} g_q(\boldsymbol{p}_q^{(r)},\mathring{t}_q) d\mathring{t}}$$
(41)

Substituting (30) into (41), we have,

$$g_q^{\text{con}}(\mathring{t}_q|\boldsymbol{p}_q^{(r)}) = \frac{e^{\rho_2 \tilde{\ell}_c(\boldsymbol{p}_q^{(r)},\mathring{t}_q)}}{\int_{\mathring{t}} e^{\rho_2 \tilde{\ell}_c(\boldsymbol{p}_q^{(r)},\mathring{t}_q)} d\mathring{t}}$$
(42)

Note that we use $\int_{\hat{t}} e^{\rho_2 \tilde{\ell}_c(\boldsymbol{p}_q^{(r)}, \hat{t}_q)} d\mathring{t}$, instead of $g_q^{\text{mar}}(\boldsymbol{p}_q^{(r)})$, to normalize the conditional pdf (41) correctly.

The process of generating the required realizations $\{(\boldsymbol{p}_q^{(r)}, \mathring{t}_q^{(r)})\}_{r,q=1}^{R,Q}$ are as follows: for qth group of realizations, generate $\{\boldsymbol{p}_q^{(r)}\}_{r=1}^R$ using the marginal pdf $g_q^{\text{mar}}(\boldsymbol{p}_q)$ (40) and then use the conditional pdf $g_q^{\text{con}}(\mathring{t}_q|\boldsymbol{p}_q^{(r)})$ (42) to generate $\{\mathring{t}_q\}_{r=1}^R$.

5. Generations Of The Required Realizations

We recall the following lemma [9] that will be used to generate the required realizations:

Lemma 1. Let $X \in \mathcal{X}$ be any RV with pdf $f_X(x)$ and the Cumulative Distribution Function (CDF) $F_X(x)$ and denote the inverse CDF as $F_X^{-1}(\cdot): [0,1] \to \mathcal{X}, u \to x$ s.t. $F_X(x) = u$. Then, for any uniform RV, $U \in [0,1]$, the RV $\tilde{X} = F_X^{-1}(U)$ is distributed according to $f_X(\cdot)$.

According to the generating process described before, firstly, we should use the marginal pdf $g_q^{\text{mar}}(\boldsymbol{p}_q)$ (40) to generate $\{\boldsymbol{p}_q^{(r)}\}_{r=1}^R$. $g_q^{\text{mar}}(\boldsymbol{p}_q)$ is, however, unsuitable to be used along with the result of Lemma 1 since \boldsymbol{p}_q is a D-dimensional vector (defining $\boldsymbol{p}_q = (x_q, y_q)$ for D=2) instead of a scalar. To acquire the D=2 scalar pdfs, consequently, $g_q^{\text{mar}}(\boldsymbol{p}_q)$ should be factorized as follows,

$$g_q^{\text{mar}}(x_q, y_q) \triangleq g_x^{\text{mar}}(x_q) g_{y|x}^{\text{con}}(y_q|x_q)$$
(43)

where

$$g_x^{\text{mar}}(x_q) \triangleq \int g_q^{\text{mar}}(x_q, y_q) dy_q$$
 (44)

$$g_{y|x}^{\text{con}}(y_q|x_q) \triangleq \frac{g_q^{\text{mar}}(x_q, y_q)}{g_x^{\text{mar}}(x_q)} \tag{45}$$

In conclusion, the approach of generating the qth group of realizations $\{(\boldsymbol{p}_q^{(r)}, \mathring{t}_q^{(r)})\}_{r=1}^R$ can be stated as follows: generate R realizations $\{u_x^{(r)}\}_{r=1}^R \sim U[0,1]$ and obtain $x_q^{(r)} = G_x^{-1}(u_x^{(r)})$ where $G_x(\cdot)$ is the CDF associated to $g_x^{\max}(x_q)$ in (44); For r=1,...,R, then, generate two realizations $u_y^{(r)}, u_{\widetilde{t}}^{(r)} \sim U[0,1]$ and obtain

 $y_q^{(r)} = G_{y|x_q^{(r)}}^{-1}(u_y^{(r)})$ and $\mathring{t}_q^{(r)} = G_{\mathring{t}|\boldsymbol{p}_q^{(r)}}^{(-1)}(u_{\mathring{t}}^{(r)})$ where $G_{y|x}(\cdot)$ and $G_{\mathring{t}|\boldsymbol{p}}(\cdot)$ are the CDF associated to $g_{y|x}^{\text{con}}(y_q|x_q)$ in (44) and $g_q^{\text{con}}(\mathring{t}_q|\boldsymbol{p}_q^{(r)})$ in (41) respectively;

6. Implementation Details

In this section, we give all the necessary details for an efficient implementation of our proposed IS-based ML DPD algorithm. To narrow the range of sampling, we would like to generate the realizations around the true parameters. The area of interest, where emitters may be put in, is confined within $x \in [x_{\min}, x_{\max}]$ and $y \in [y_{\min}, y_{\max}]$ and the range of transmitted time is $\mathring{t} \in [0, \mathring{t}_{\max}]$, where \mathring{t}_{\max} can be freely chosen as high as desired. The starting point of the algorithm is, therefore, the evaluation of the SLF (39) at multiple grid points $\{(x^{(i)}, y^{(j)})\}_{i,j=1}^{I,J}$ with relatively large step Δx and Δy . Then, by approximating integrals with discrete sums, we evaluate the qth marginal pdf with respect to $\mathbf{p} = (x, y)$ in (32), $g_q^{\max}(x, y)$, at every grid point as follows,

$$g_q^{\max}(x^{(i)}, y^{(j)}) \simeq \frac{e^{\rho_1 \ell(x^{(i)}, y^{(j)})}}{\sum_{i=1}^{I} \sum_{j=1}^{J} e^{\rho_1 \ell(x^{(i)}, y^{(j)})} \Delta x \Delta y}$$
(46)

The estimation of qth emitter position can be initialized as the grid point at which the maximum peak of $g_q^{\text{mar}}(x^{(i)}, y^{(j)})$ is. The initialization can be written as,

$$[\mathring{x}_q, \mathring{y}_q] = \arg\max_{(x^{(i)}, y^{(j)})} g_q^{\max}(x^{(i)}, y^{(j)}), \quad q = 1, ..., Q$$

$$(47)$$

The distance between the ML estimator $\hat{\boldsymbol{p}}_q$ and the initial estimator $\mathring{\boldsymbol{p}}_q$, however, is non-negligible since the decouple processing in (27) lose the correlation information among emitters. To force $\{(\boldsymbol{p}_q^{(r)})\}_{r=1}^R$ to be generated in the vicinity of $\hat{\boldsymbol{p}}_q$, we fix the qth local area

$$D_{\mathring{x}_q} \triangleq [\mathring{x}_q - \Delta_x, \mathring{x}_q + \Delta_x], \quad D_{\mathring{y}_q} \triangleq [\mathring{y}_q - \Delta_y, \mathring{y}_q + \Delta_y]$$
 (48)

The distribution of marginal probability $g_x^{\text{mar}}(x_q)$ in local x-interval $D_{\mathring{x}_q}$ (or conditional probability $g_{y|x}^{\text{con}}(y_q|x_q)$ in local y-interval $D_{\mathring{y}_q}$) is governed by the design parameter ρ_1 . The range parameter Δ_x (or Δ_y) should, therefore, match with ρ_1 , in order to generate the required realizations efficiently.

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