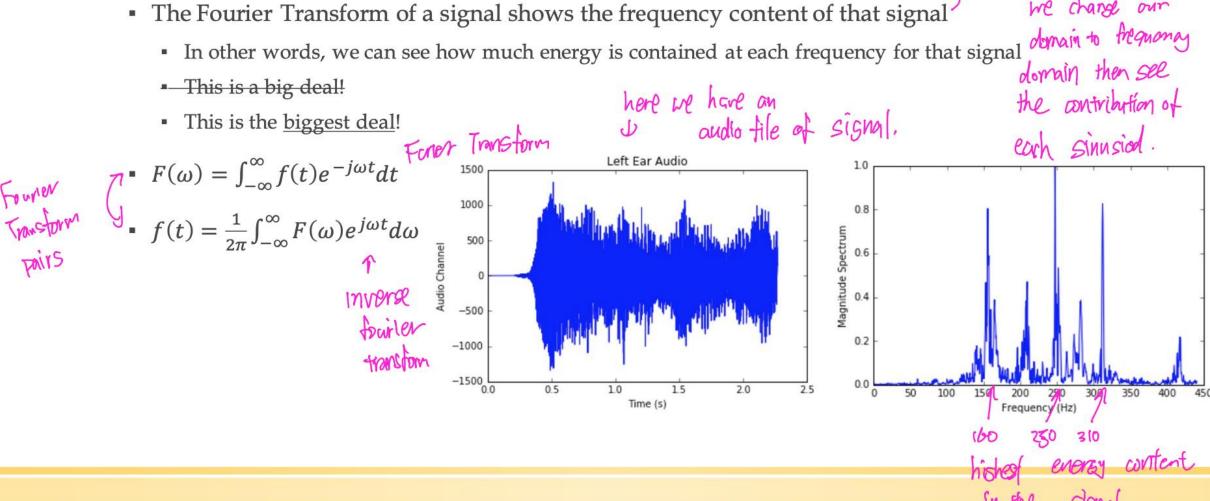
### **HKN ECE 210 Exam 3 Review Session**

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## **Topics**

- Fourier Transform
- Signal Energy and Bandwidth
- LTI System Response with Fourier Transform
- Modulation, AM, Coherent Demodulation
- Impulse Response and Convolution
- Sampling and Analog Reconstruction
- LTIC and BIBO Stability

### Fourier Transform



What it does is that It de composes a sismal

into many sinusoids de

## **Important Signals for Fourier Transform**

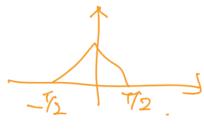
$$rect\left(\frac{t}{T}\right) = \begin{cases} 1, for |t| < \frac{T}{2} \\ 0, for |t| > \frac{T}{2} \end{cases}$$

$$|total | text | text | (0, for t < 0) | (1, for t > 0) | (1, for t >$$

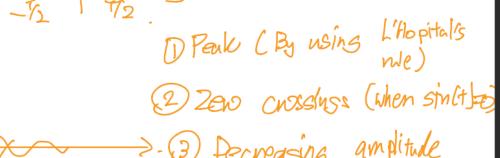
$$u(t) = \begin{cases} 0, & \text{for } t < 0 \\ 1, & \text{for } t > 0 \end{cases}$$

AUD they have conesponding Fourier Trensform pairs given in formula sheet

$$\Delta\left(\frac{t}{T}\right) = \begin{cases} 1 + \frac{2t}{T}, for \frac{-T}{2} < t < 0 \\ 1 - \frac{2t}{T}, for 0 < t < \frac{T}{2} \end{cases}$$



• 
$$sinc(t) = \begin{cases} \frac{\sin(t)}{t}, for \ t \neq 0 \\ 1, for \ t = 0 \end{cases}$$



## **Fourier Transform Tips**

- Convolution in the time domain is multiplication in the frequency domain
  - $f(t) * g(t) \leftrightarrow F(\omega)G(\omega)$
- Conversely, multiplication in the time domain is convolution in the frequency domain
  - $f(t)g(t) \stackrel{\mathcal{F}}{\leftrightarrow} \frac{1}{2\pi} F(\omega) * G(\omega)$
- Scaling your signal can force properties to appear; typically time delay
   Ex:  $e^{-2(t+1)}u(t-1) \rightarrow e^{-4}e^{-2(t-1)}u(t-1)$  terms that has to he in the same form before  $e^{-2(t+1)}u(t-1) \rightarrow e^{-4}e^{-2(t-1)}u(t-1)$
- The properties really do matter! Take the time to acquaint yourself with them.
- Remember that the Fourier Transform is linear, so you can express a spectrum as the sum of easier spectra
  - Ex: Staircase function
- Magnitude Spectrum is even symmetric, Phase Spectrum is odd symmetric for real valued signals due to Hermitian Property

# Signal Energy and Bandwidth

Parseval's Theorem

• Energy =  $W = \int_{-\infty}^{\infty} |f(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |F(\omega)|^2 d\omega$ 

Energy signals can be either low-pass or band-pass signals

· Why not high-pass? Not bard limited => & every

Bandwidth for Low-pass Signals

■ 3dB BW → fr W>O.

$$|F(\Omega)|^2 = \frac{1}{2}$$

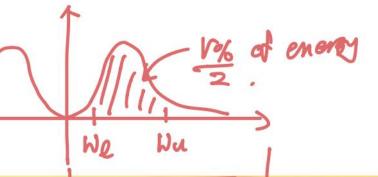
• r% BW

Find borndwidth so that energy contained within it is 1% at the entire energy  $\frac{1}{2\pi} \int_{-\Omega}^{\Omega} |F(\omega)|^2 d\omega = rW$ 

Bandwidth for Band-pass signals

r% BW

r% BW 
$$\frac{1}{2\pi} \int_{\omega_l}^{\omega_u} |F(\omega)|^2 d\omega = \frac{rW}{2}, \quad \Omega = \omega_u - \omega_l$$



### **Problem 1 FA19**

(a) The signal f(t) has Fourier transform  $F(\omega)$ . Determine the inverse Fourier transform of the following three signals. Leave your answers in terms of f(t).

iii. (6 pts) 
$$F_3(\omega) = F(200)e^{-j600}\delta(\omega - 200) - F(-200)e^{j600}\delta(\omega + 200)$$
a iii) When you see doth function not to another function  $f(\omega)$  samples properly

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) \qquad F(\omega) e^{-j600} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) \qquad F(\omega) e^{-j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega - 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega + 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega + 2\omega p) - F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

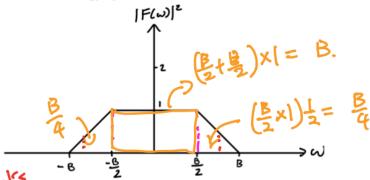
$$F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p)$$

$$F(\omega) e^{-j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega} f(\omega + 2\omega p) \qquad e^{j6\omega}$$

#### **Problem 2 FA20**

(c) Given the energy spectrum below:

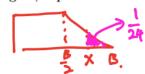


Parson i. (4 pts) Determine the signal energy, W, in terms of B.

eqn in frequency 
$$W = \frac{1}{2\pi} \int_{-\infty}^{\infty} |F(W)|^2 d\Omega = \frac{1}{2\pi} \left[ Bt \ 2 \cdot \frac{B}{4} \right]$$

$$= \frac{3}{4\pi} B J.$$

ii. (4 pts) What is the bandwidth,  $\Omega$ , needed to capture 11/12 of the energy in this signal, expressed in terms of B?



Energy of small 
$$b = Area of \Delta \cdot 2 \cdot \frac{1}{2\pi}$$

$$\frac{1}{12} \cdot \frac{3}{4\pi} B = \text{Area of } \triangle \cdot \pi$$

Energy of small 
$$\Delta = \text{Area of } \Delta \cdot 2 \cdot \frac{1}{2\pi}$$
.

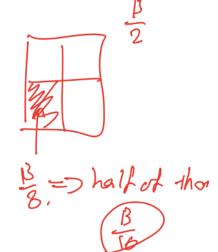
$$\frac{1}{12} \cdot \frac{3}{4\pi} B = \text{Area of } \Delta \cdot \pi$$

$$\frac{1}{12} \cdot \frac{3}{4\pi} B = \text{Area of } \Delta \cdot \pi$$

$$9 = \frac{2}{16} \times \frac{1}{4} \times \frac{1}{2}$$
Area of  $\Delta = \frac{1}{16} = \frac{1}{2} \left( \frac{13}{4} \times \frac{1}{2} \right)$ 

Area of 
$$\Delta = \frac{1}{16} = \frac{1}{2} \left( \frac{13}{4} \times \frac{1}{2} \right)$$

$$\Omega = \frac{3B}{4}$$



# LTI System Response using Fourier Transform

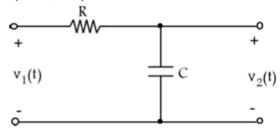
• Given the following LTI system:

$$f(t) \to H(\omega) \to y(t)$$

- $Y(\omega) = F(\omega)H(\omega)$
- Computationally speaking, and in future courses, we prefer to use Fourier Transforms instead of convolution to evaluate an LTI system
- Why?
  - So much faster: O(nlogn) vs.  $O(n^2)$

### **Problem 4 FA17**

4. (25 pts) For the circuit below,  $R = 10 \mathrm{K}\Omega$  and  $C = 1 \mu F$ , determine:



- (a) The differential equation relating the input x (voltage  $v_1(t)$ ) to output y (voltage  $v_2(t)$ ).
- (b) The frequency response,  $H(\omega) =$ \_\_\_\_\_\_.
- (d) The response to  $v_1(t) = 5u(t)$  V.  $v_2(t) =$  \_\_\_\_\_\_

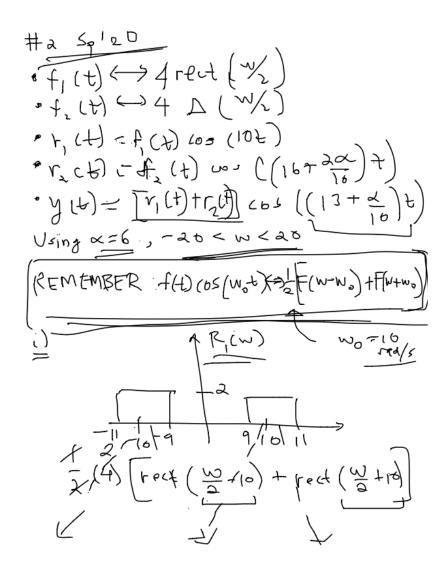
### Modulation, AM Radio, Coherent Demodulation

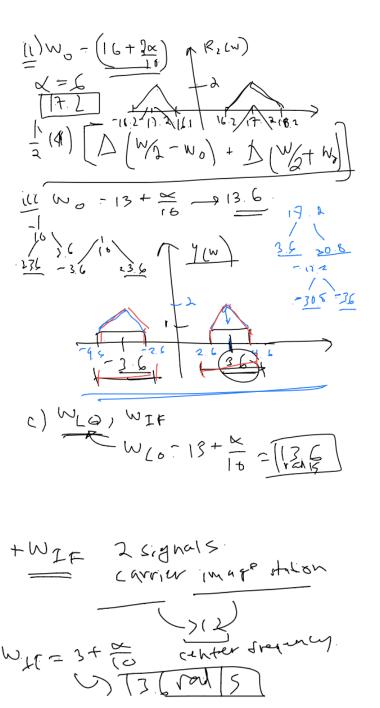
- Modulation Property
  - $f(t) \leftrightarrow F(\omega), f(t) \cos(\omega_o t) \leftrightarrow \frac{1}{2} [F(\omega \omega_o) + F(\omega + \omega_o)]$
- In general, for Amplitude Modulation in communications, we modulate with cosine in order to shift our frequency spectrum into different frequency bands
- Coherent Demodulation refers to the process of modulating a signal in order to make it band-pass, then modulating it back to the original low-pass baseband before low-pass filtering in order to recover the original signal
- Envelope detection is the process of Full-wave Rectification (absolute value), then low-pass filtering in order to extract the signal

#### Problem 2 SP20

- (b) [11 pts] Suppose  $f_1(t) \leftrightarrow 4\text{rect}(\frac{\omega}{2}), f_2(t) \leftrightarrow 4\Delta(\frac{\omega}{2}), r_1(t) = f_1(t)\cos(10t), r_2(t) = f_2(t)\cos((16+\frac{2\alpha}{10})t)$  and  $y(t) = [r_1(t)+r_2(t)]\cos((13+\frac{\alpha}{10})t)$ . Sketch the Fourier transforms of  $r_1(t), r_2(t), \text{ and } y(t) \text{ for } -20 < \omega < 20$ . Carefully label important points (both horizontal and vertical). sketches?
- (c) [6 pts] Suppose we want to use the equations in part (b) as a simplified model to illustrate the image station problem in AM communications, where  $r_1(t)$  and  $r_2(t)$  represent AM signals broadcast by two different stations. Identify the local oscillator frequency  $\omega_{LO}$  and the center frequency  $\omega_{IF}$  of the IF band used in the model. Explain your answers.

 $\omega_{\mathbf{LO}}, \ \omega_{\mathbf{IF}} = ?$ 





## Impulse Response and Convolution

- Convolution
  - $x(t) * y(t) = \int_{-\infty}^{\infty} x(\tau)y(t-\tau)d\tau = \int_{-\infty}^{\infty} x(t-\tau)y(\tau)d\tau$
  - We "flip and shift" one signal and evaluate the integral of the product of the two signals at any value of t (our delay for the shift)
- Representing LTI Systems
  - y(t) = x(t) \* h(t), where h(t) is the **impulse response** of the system
- Impulse Response is the system output to a  $\delta(t)$  input
- Graphical convolution helps to visualize the process of flipping and shifting

## Helpful Properties for Convolution

- Derivative
  - $h(t) * f(t) = y(t) \to \frac{d}{dt}h(t) * f(t) = h(t) * \frac{d}{dt}f(t) = \frac{d}{dt}y(t)$
  - Use of Derivative property: Finding the impulse response from the unit-step response

• If 
$$y(t) = u(t) * h(t)$$
, then  $\frac{d}{dt}y(t) = \frac{d}{dt}u(t) * h(t) = \delta(t) * h(t) = h(t)$ 

- Start Point
  - If the two signals have start points at  $t_1$  and  $t_2$ , then the start point of their convolution will be at  $t_1 + t_2$
- End Point
  - Similarly for the end points, if the two signals have end points at  $t_1$  and  $t_2$ , then the end point of their convolution will be at  $t_1 + t_2$
- Width
  - From the above two properties, we can see that if the two signals have widths  $W_1$  and  $W_2$ , then the width of their convolution will be  $W_1 + W_2$

## The Impulse Function $\delta(t)$

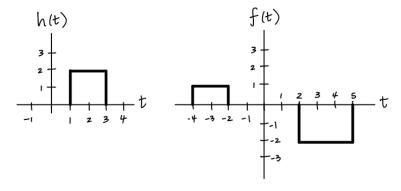
- The impulse function is the limit of  $\frac{1}{T}rect\left(\frac{t}{T}\right)as T \to 0$ 
  - Infinitesimal Width
  - Infinite Height
  - Of course, it integrates to 1.  $(0 * \infty = 1)$
- Sifting

• 
$$\int_{a}^{b} \delta(t - t_{o}) f(t) dt = \begin{cases} f(t_{0}), t_{0} \in [a, b] \\ 0, & else \end{cases}$$

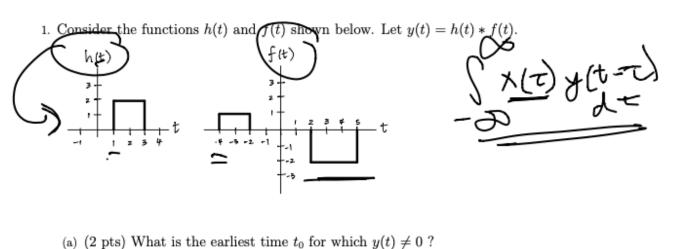
- Sampling
  - $f(t)\delta(t-t_o) = f(t_o)\delta(t-t_o)$
- Unit-step derivative
  - $\frac{du}{dt} = \delta(t)$

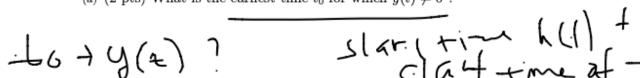
### Problem 1 FA20

1. Consider the functions h(t) and f(t) shown below. Let y(t) = h(t) \* f(t).

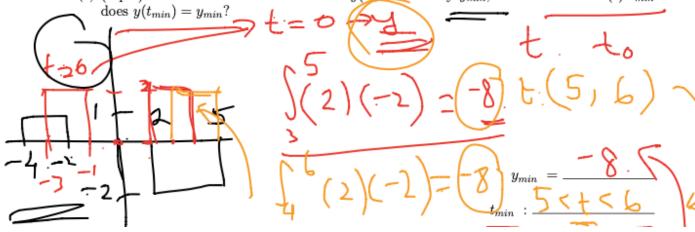


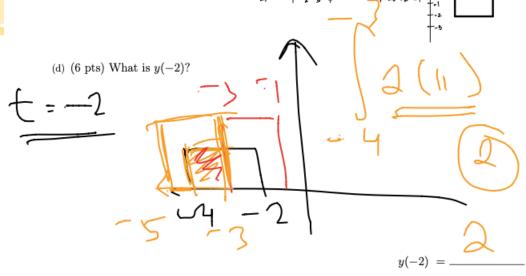
- (a) (2 pts) What is the earliest time  $t_0$  for which  $y(t) \neq 0$ ?
- (b) (4 pts) What is the minimum value of y(t), denoted by  $y_{min}$ , and at what time(s)  $t_{min}$  does  $y(t_{min}) = y_{min}$ ?
- (d) (6 pts) What is y(-2)?
- (e) (10 pts) Plot y(t) over the range -10 < t < 10. Be sure to label both the vertical and horizontal axes.



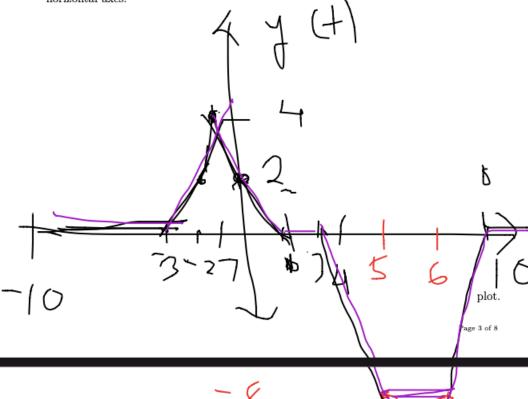


(b) (4 pts) What is the minimum value of y(t), denoted by  $y_{min}$ , and at what time(s)  $t_{min}$ 





(e) (10 pts) Plot y(t) over the range -10 < t < 10. Be sure to label both the vertical and horizontal axes.



### Problem 3 FA20

(b) [5 pts] Find the impulse response h(t) of the LTI system having the following unit-step response:

$$y(t) = t^2 u(t)$$

## Sampling and Analog Reconstruction

- If we have an original analog signal f(t)
- Our digital samples of the signal are obtained through sampling property as:
  - f[n] = f(nT) where T is our sampling period; this is Analog to Digital (A/D) conversion
  - This results in infinitely many copies of the original signal's Fourier Transform spaced by  $\frac{2\pi}{T}$
- We must make sure to satisfy Nyquist Criterion:
  - $T < \frac{1}{2B}$  or  $f_S > 2B$ , B = Bandwidth (in Hz)
- Following A/D conversion, we perform D/A conversion, then low-pass filter our signal in order to obtain our original signal according to the following relation

$$f(t) = \sum_{n} f_{n} sinc\left(\frac{\pi}{T}(t - nT)\right)$$

• For a more complete explanation, take ECE 310!

# Sampling and Analog Reconstruction (cont'd)

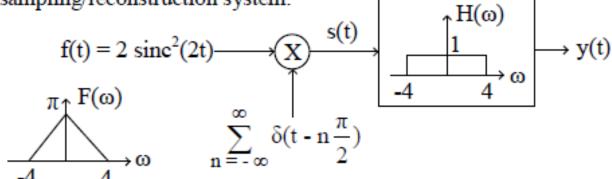
• The complete mathematical relationship between the continuous frequency spectrum and the discrete time (sampled) spectrum in analog frequency

$$F_{S}(\omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} F\left(\omega + \frac{2\pi k}{T}\right)$$

The nuances of this representation will be explored and clarified in ECE 310!

### Problem 4 SP14

) Consider this sampling/reconstruction system.



- Circle the correct answer and explain it below.
  - f(t) is: UNDERSAMPLED / OVERSAMPLED / SAMPLED AT NYQUIST RATE

Explanation:

- Sketch S(ω) and Y(ω) on the axes below.
- iii) Determine y(t)

## **BIBO Stability**

- A system is BIBO stable if for any bounded input, we obtain a bounded output
- Two ways to check BIBO stability
- By the definition:
  - If  $|f(t)| \le \alpha < \infty$ , then  $|y(t)| \le \beta < \infty$ ,  $\forall t$
- By Absolute Integrability
  - $\int_{-\infty}^{\infty} |h(t)| dt < \infty$

#### **LTIC**

- LTIC stands for Linearity Time-Invariance and Causality
- Linearity
  - Satisfy Homogeneity and Additivity
  - Can be summarized by Superposition
    - If  $x_1(t) \to y_1(t)$  and  $x_2(t) \to y_2(t)$ , then  $ax_1(t) + bx_2(t) \to ay_1(t) + by_2(t)$
- Shift Invariance
  - If  $x(t) \rightarrow y(t)$  then  $x(t t_o) \rightarrow y(t t_o) \forall t_o$  and x(t)
- Causality
  - Output cannot depend on future input values
  - h[n] = 0 for n < 0.

### Problem 4 SP 20

[12 pts] A system has the input-output relation  $y(t) = \text{rect}(t-\alpha-1)*f^2(t+1)$ . Determine whether or not the system is BIBO stable and/or causal. Explain your reasoning.

BIBO, causal?

Note: alpha is a positive integer