

STATISTICAL BEAMFORMING

Assignment 2 : EEL 6935 - SPRING.2018 Hector Lopez 2-17-2018

Description

Show that the maximum signal-to-interference-plus-noise ratio (max-SINR), minimum mean-square-error (MMSE), minimum-variance-distortionless-response (MVDR), and maximum likelihood (ML) with Gaussian disturbance beamformers are all equivalent; that is, they are scaled versions of each other.

Define Statistical Beam Forming Methods

The formulas for each of the beam former methods can be written as the set of equations below. Notice that the R matrix represents the autocorrelation of the receiving signal and the R_{I+N} represents the autocorrelation of the disturbances, interference and noise. The ρ is the power of the input signal and the S_{θ} is the array response vector of the antenna array elements.

$$w_{\max \text{ SINR}} = c R_{I+N}^{-1} \vec{S}_{\theta}$$

$$w_{\text{MMSE}} = \rho R^{-1} \vec{S}_{\theta} \quad ; \quad \rho = E\{|h(k)|^2\}$$

$$w_{\text{MVDR}} = \frac{1}{\vec{S}_{\theta}^H R^{-1} \vec{S}_{\theta}} \cdot R^{-1} \vec{S}_{\theta}$$

$$w_{\text{ML}} = \frac{R_{I+N}^{-1} \cdot \vec{S}_{\theta}}{\vec{S}_{\theta}^H \cdot R_{I+N}^{-1} \cdot \vec{S}_{\theta}}$$

Understanding the Auto-Correlation Matrix R

Breaking down the R matrix allows us to see the components of the received signal in terms of the desired input equaling the input component.

$$\begin{aligned}
R &= E \{ \vec{r}(t) \cdot \vec{r}(t)^H \} \\
&= E \left\{ \underbrace{\left(\frac{m(t)}{d(t)} e^{j2\pi f_c t} \cdot \vec{s}_0 + \vec{i}(t) + \vec{n}(t) \right)}_{\vec{r}(t)} \cdot \left(m(t) e^{j2\pi f_c t} \cdot \vec{s}_0 + \vec{i}(t) + \vec{n}(t) \right)^H \right\} \\
&= E \left\{ \left(\frac{m(t)}{d(t)} \cdot \vec{s}_0 + \vec{i}(t) + \vec{n}(t) \right) \cdot \left(d(t) \cdot \vec{s}_0 + \vec{i}(t) + \vec{n}(t) \right)^H \right\} \\
&= E \left\{ \vec{d}(t) \cdot \vec{s}_0 \cdot \vec{d}(t)^H \vec{s}_0^H + \vec{i}(t) \cdot \vec{i}(t)^H + \vec{n}(t) \cdot \vec{n}(t)^H \right\} \\
&\quad * \vec{r}(t) \text{ not correlated w/ } \vec{n}(t) \text{ \& } \vec{i}(t), \text{ \& } \vec{r}(t) \text{ \& } \vec{i}(t) \text{ are not correlated} \\
&= E \left\{ \vec{d}(t) \cdot \vec{s}_0 \cdot \vec{d}(t)^H \vec{s}_0^H \right\} + E \left\{ \vec{i}(t) \cdot \vec{i}(t)^H \right\} + E \left\{ \vec{n}(t) \cdot \vec{n}(t)^H \right\} \\
R &= \quad \quad \quad + R_I + R_N
\end{aligned}$$

We can group the auto-correlation of the interference and noise into the following solution for R.

$$R = E \{ \vec{d}(t) \cdot \vec{s}_0 \cdot \vec{d}(t)^H \vec{s}_0^H \} + R_{I+N}$$

We can now take the inverse of both sides and get a similar method that allows us to identify a scalar K between the R and R_{I+N} . With this relationship we can then try to find scalars between the beamformer formulas.

$$\begin{aligned}
R^{-1} &= [E \{ \quad \quad \quad \} + R_{I+N}]^{-1} \\
R^{-1} &= [E \{ \quad \quad \quad \}]^{-1} + R_{I+N}^{-1} \\
R^{-1} &= \underbrace{\left[\left[\quad \right] \cdot R_{I+N} + 1 \right]}_K \cdot R_{I+N}^{-1}
\end{aligned}$$

$$R^{-1} = \left[\left[E \{ d(t) d(t)^H \} \underbrace{s_{\theta}^* s_{\theta}}_{(M^{-1})??} \right]^{-1} \cdot R_{I+N} + 1 \right] \cdot R_{I+N}^{-1}$$

$$\left[\underbrace{\left[E \{ d(t) \cdot d(t)^H \} \right]^{-1}}_{\text{Autocorrelation of desired signal}} \cdot \underline{M} \cdot R_{I+N} + 1 \right] \cdot R_{I+N}^{-1}$$

$$(1) \quad R^{-1} = \underbrace{\left[R_d^{-1} \cdot M \cdot R_{I+N} + 1 \right]}_{\text{'K'}} \cdot R_{I+N}^{-1}$$

Solving for Scalars

The combination of the S_{θ} and S_{θ} Hermitian is similar the $W_{\max \text{SNR}}$ solution so here we can see that the combination of the array response vectors can be equivalent to the size of the antenna array, or the number of antenna elements M . Notice that the K value in the equation above is used in the following derivations.

Here we define the constant that would set the $W_{\max \text{SNR}}$ equal to the W_{ML} :

$$W_{\max \text{SNR}} = \underline{C} \cdot R_{I+N}^{-1} S_{\theta} \quad \& \quad W_{\text{ML}} = \frac{1}{\vec{S}_{\theta}^H \cdot R_{I+N}^{-1} \cdot \vec{S}_{\theta}} \cdot R_{I+N}^{-1} \vec{S}_{\theta}$$

$$W_{\max \text{SNR}} \stackrel{?}{=} W_{\text{ML}}$$

$$C \cdot R_{I+N}^{-1} \cdot \vec{S}_{\theta} = \frac{1}{\vec{S}_{\theta}^H \cdot R_{I+N}^{-1} \cdot \vec{S}_{\theta}} \cdot R_{I+N}^{-1} S_{\theta}$$

$$\text{if } \left| C = \frac{1}{\vec{S}_{\theta}^H \cdot R_{I+N}^{-1} \cdot \vec{S}_{\theta}} \right| \text{ then, } W_{\max \text{SNR}} \equiv W_{\text{ML}}$$

Here we try to find the link between the same $W_{\max \text{SNR}}$ and W_{MVDR} . The use of the K value we defined earlier in equation (1) is needed.

$$W_{\max \text{ SINR}} \stackrel{?}{=} W_{\text{MMSE}}$$

$$W_{\text{MMSE}} = \frac{1}{\vec{s}_0^H R^{-1} \vec{s}_0} \cdot R^{-1} \vec{s}_0 \xrightarrow{\text{given "k" (.)}} \frac{1}{\vec{s}_0^H K R_{I+N}^{-1} \vec{s}_0} \cdot K \cdot R_{I+N}^{-1} \vec{s}_0$$

$$C \cdot [R_{I+N}^{-1} \cdot \vec{s}_0]^{-1} = \frac{1}{\vec{s}_0^H \cdot K \cdot R_{I+N}^{-1} \vec{s}_0} \cdot K \cdot [R_{I+N}^{-1} \vec{s}_0] \cdot [\quad]^{-1}$$

$$\text{if } \left[C = \frac{1}{\vec{s}_0^H \cdot K \cdot R_{I+N}^{-1} \vec{s}_0} \cdot K \cdot I_M \right] \text{ then } W_{\max \text{ SINR}} = W_{\text{MMSE}}$$

Here we try to find the relationship between the $W_{\max \text{ SINR}}$ and the W_{MMSE} . The K is used again by substituting the R^{-1} / K matrix into the SINR formula. We then see the relationship between the two methods.

$$W_{\max \text{ SINR}} \stackrel{?}{=} W_{\text{MMSE}}$$

$$C \cdot R_{I+N}^{-1} \vec{s}_0 = \frac{1}{K} R^{-1} \vec{s}_0$$

$$\frac{1}{K} \cdot R^{-1} \cdot \vec{s}_0 = R^{-1} \vec{s}_0$$

$$\text{if } \left[C = \frac{1}{K} \right] \text{ then } W_{\max \text{ SINR}} = W_{\text{MMSE}}$$

$$C_s = \left(\frac{1}{R_d^{-1} \cdot m \cdot R_{I+N}^{-1}} \right) \vec{s}$$

Lets use the constants that we found and try to solve for each equation below.

Given C_1, C_2, C_3 ; $W_{\max \text{ SINR}} = W_{ML} \neq$
 $" " = W_{MVDZ}$
 $" " = W_{MSE}$

So

scalar

W_{ML}
 W_{MVDZ}
 W_{MSE}
 $W_{\max \text{ SINR}}$

$$\left[\begin{array}{l} C_1 \cdot W_{\max \text{ SINR}} = W_{ML} \\ C_2 \cdot W_{\max \text{ SINR}} = W_{MVDZ} \\ C_3 \cdot W_{\max \text{ SINR}} = W_{MSE} \end{array} \right.$$

$$\left[\begin{array}{l} \frac{C_2}{C_1} \cdot W_{ML} = W_{MVDZ} \\ \frac{C_3}{C_1} \cdot W_{ML} = W_{MSE} \\ \frac{C_3}{C_2} \cdot W_{MVDZ} = W_{MSE} \\ \frac{C_3}{C_2} \cdot W_{MVDZ} = W_{ML} \end{array} \right.$$

Derivation of eq.

$$W_{\max \text{ SINR}} = \frac{W_{ML}}{C_1}$$

$$W_{\max \text{ SINR}} = \frac{W_{MVDZ}}{C_2}$$

$$\frac{W_{ML}}{C_1} = \frac{W_{MVDZ}}{C_2}$$

$$\frac{C_2}{C_1} W_{ML} = W_{MVDZ}$$

Conclusion

By finding the constants that relate between the $W_{\max \text{ SINR}}$ and the other methods allows us to see that each beam former method can be written as a scalar of the other.