

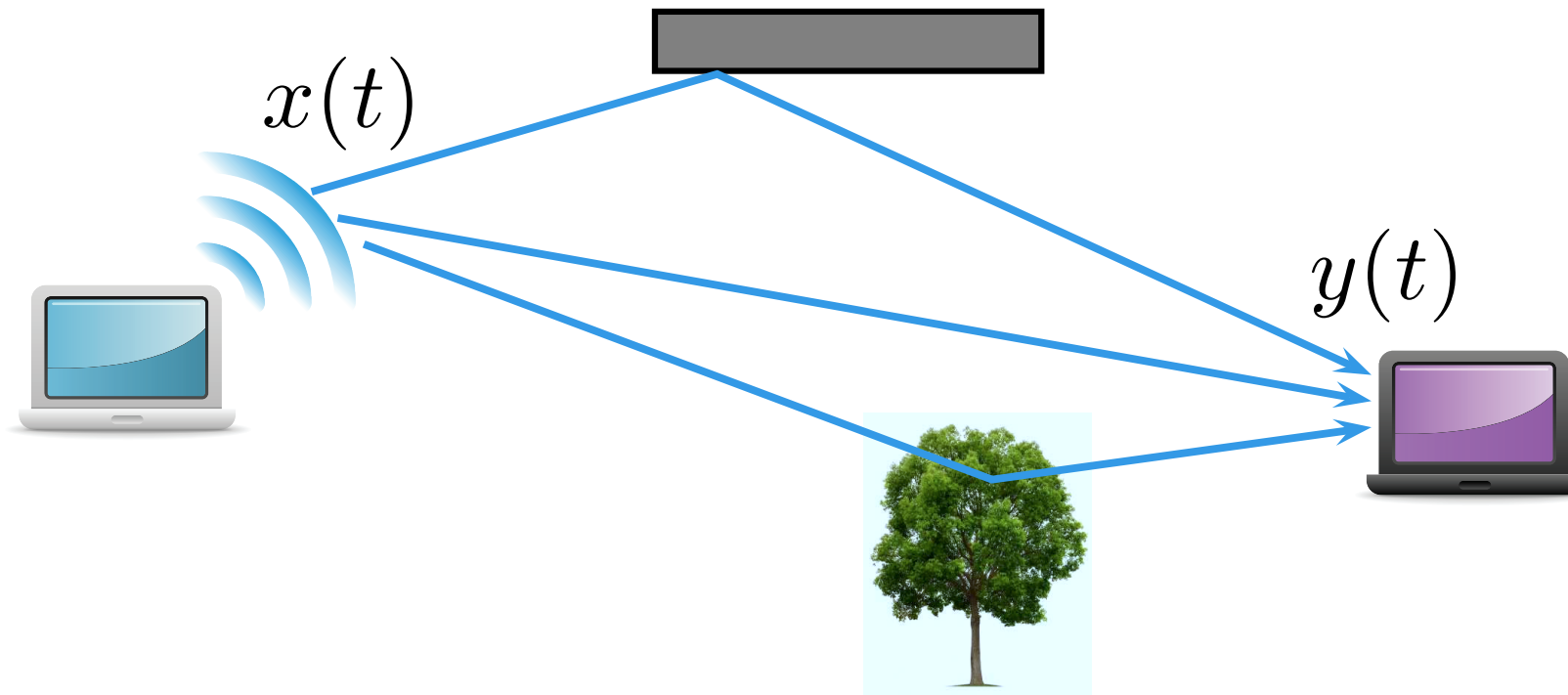
Lecture 7: Detection under Fading

Prof. Bobak Nazer 9/23/14

- Recall the continuous-time channel model:

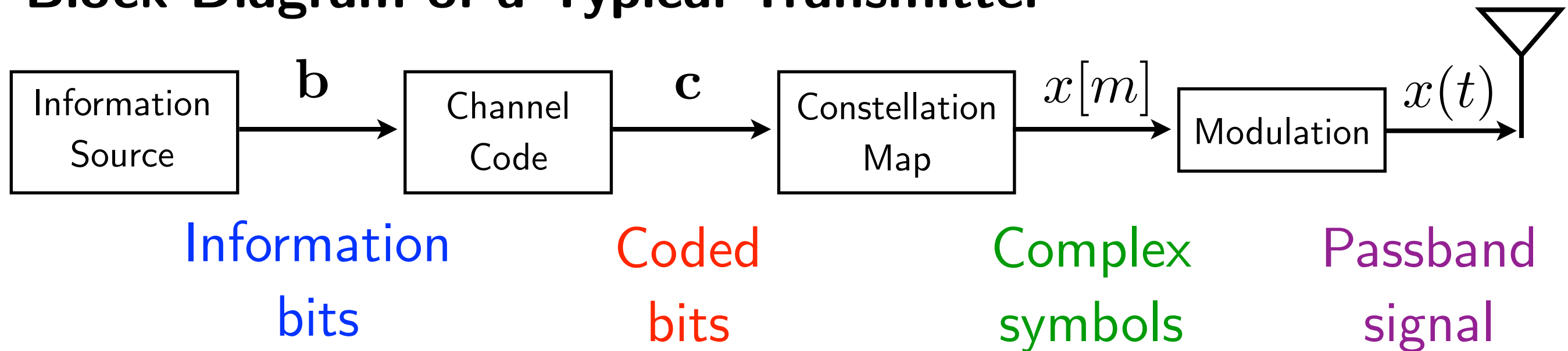
$$y(t) = \int_{-\infty}^{\infty} h(\tau, t) x(t - \tau) d\tau + w(t)$$

$$h(\tau, t) = \sum_i a_i(t) \delta(\tau - \tau_i(t)) \quad \text{Channel impulse response}$$

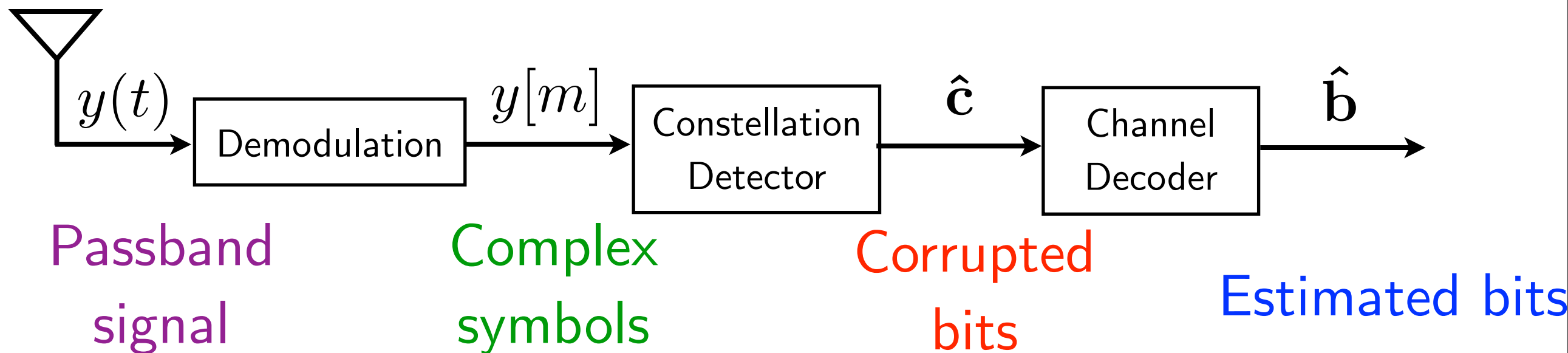


Typical Transmitter and Receiver

Block Diagram of a Typical Transmitter



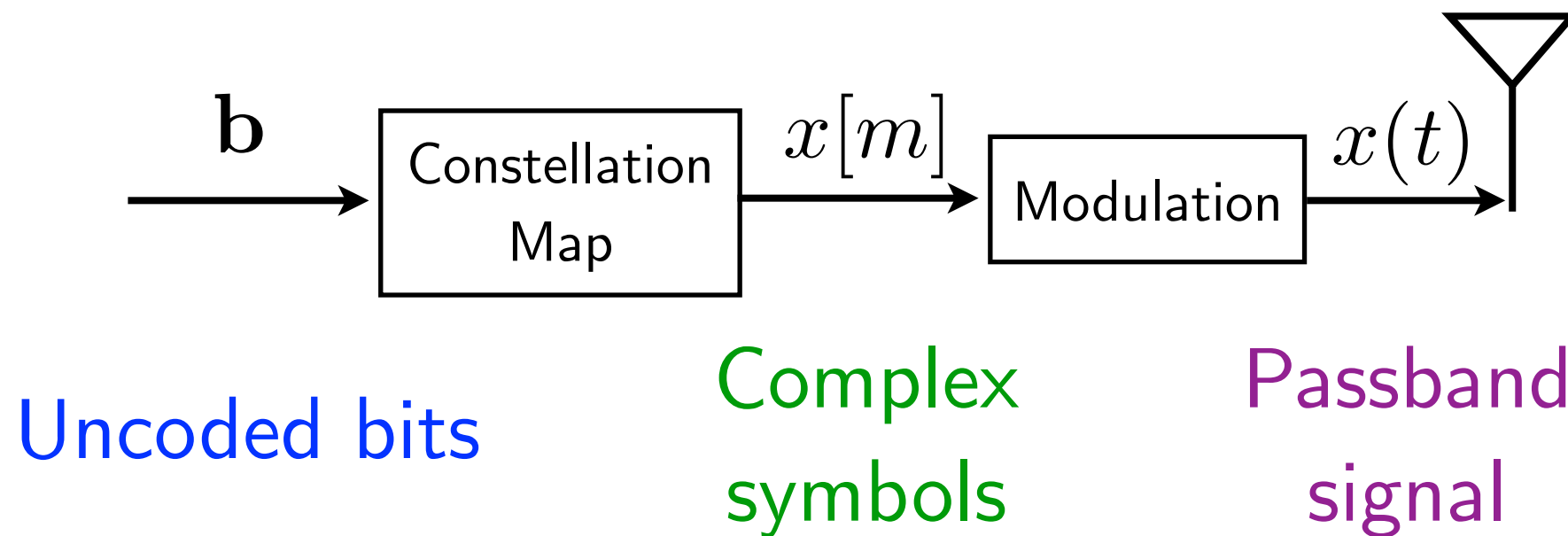
Block Diagram of a (Hard Decision) Receiver



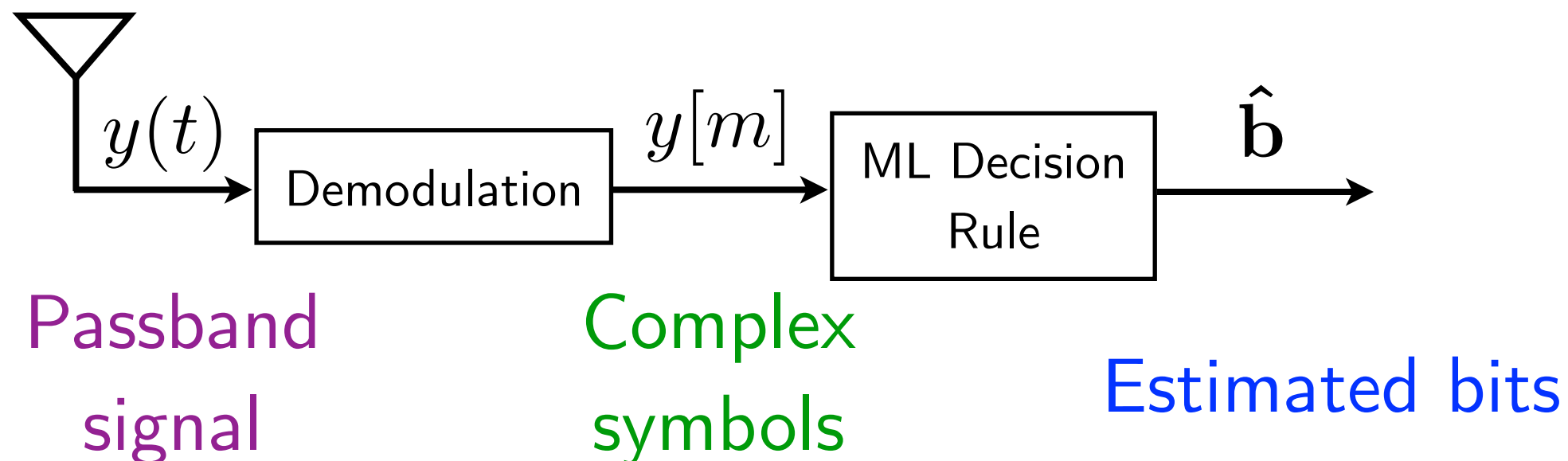
Uncoded Bits and Optimal Detection

- For now, we will ignore the possibility of channel coding:

Transmitter Block Diagram (without channel coding)



Receiver Block Diagram (without channel coding)

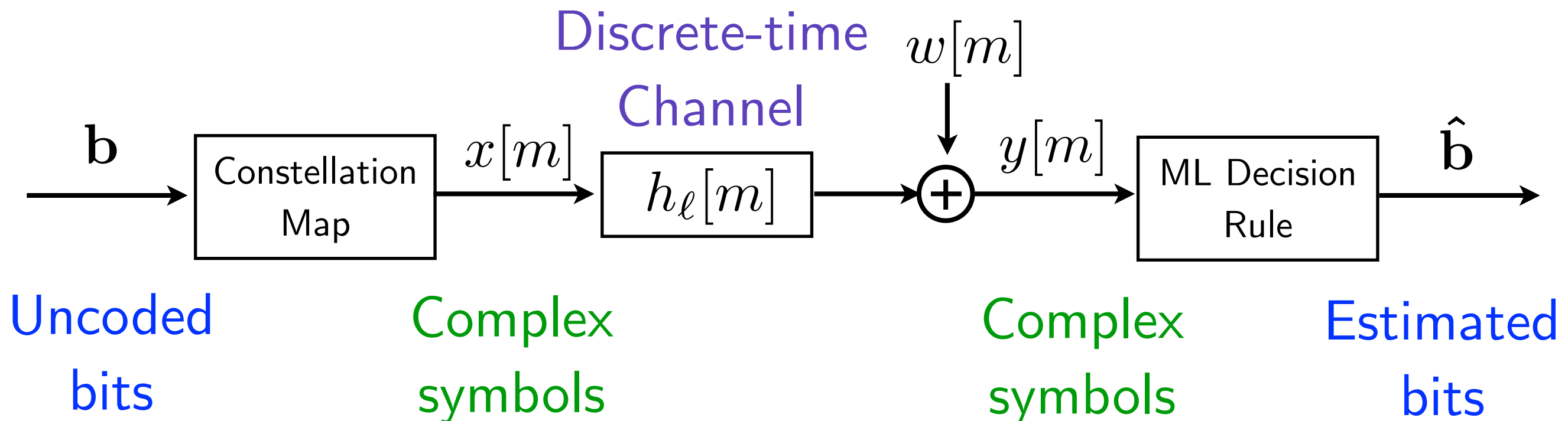


Discrete-Time Equivalent Block Diagram

- Easier to evaluate the performance in discrete time.
- Recall that the **discrete-time equivalent channel** is:

$$y[m] = \sum_{\ell} h_{\ell}[m] x[m - \ell] + w[m]$$

Discrete-Time Equivalent Transmitter and Receiver



Flat Fading

- To keep things simple, we will start by considering flat fading channels:

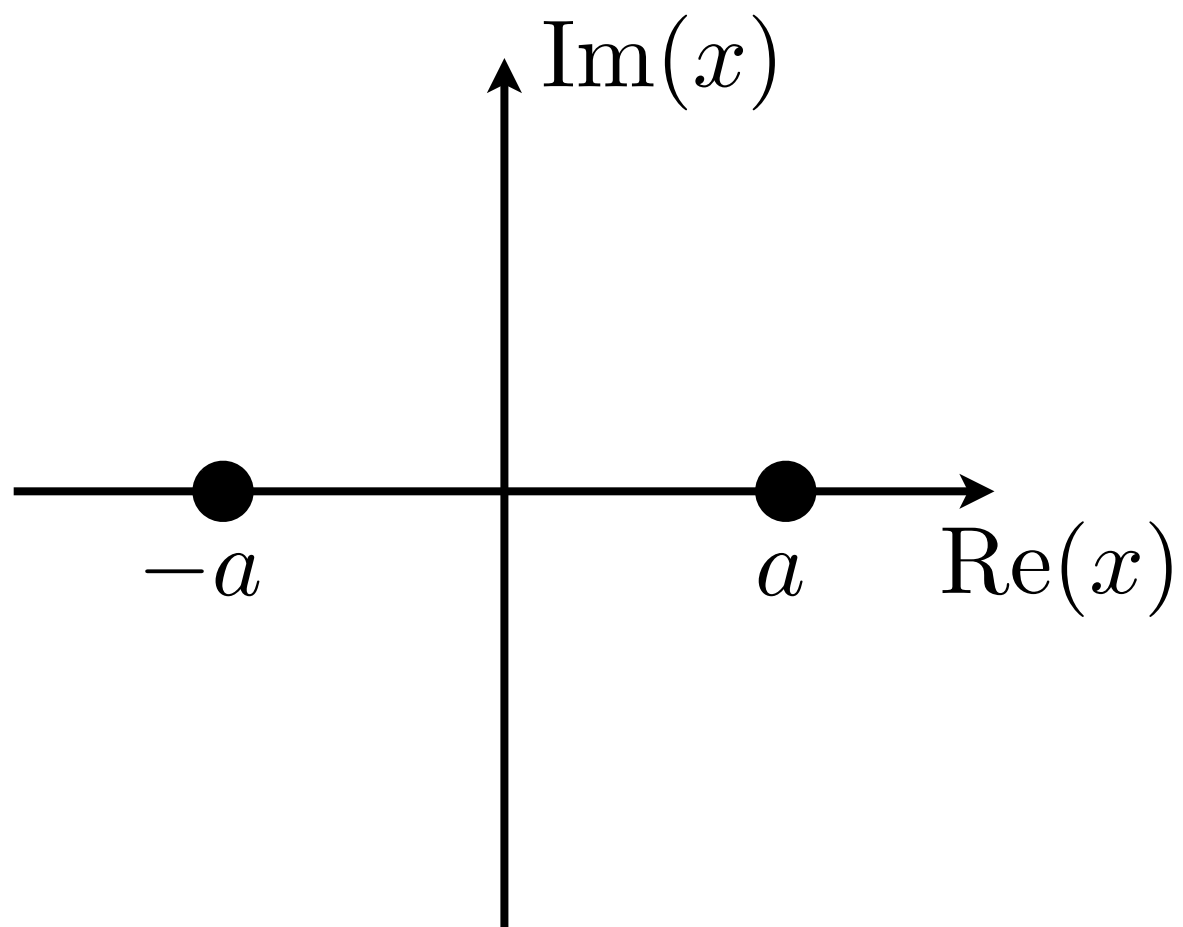
$$y[m] = h[m]x[m] + w[m]$$

- Note that we have dropped the tap index ℓ to simplify the notation.
- We will assume that the noise is i.i.d. across time and is distributed as $w[m] \sim \mathcal{CN}(0, N_0)$
- Let us start with the simplest possible setting: We have only one bit to communicate from the transmitter to the receiver.

AWGN Channels and BPSK

- For the sake of comparison, let us consider the case where there is **no fading**: $y[m] = x[m] + w[m]$
- To send **one bit**, we can use a BPSK constellation:

$$x[0] \in \{+a, -a\}$$



- **ML Decision Rule:**

$$\hat{x}[0] = \begin{cases} a & y[0] > 0 , \\ -a & y[0] \leq 0 . \end{cases}$$

AWGN Channels and BPSK

- The probability of error is $\mathbb{P}(\hat{x}[0] \neq x[0]) = Q\left(\frac{a}{\sqrt{N_0/2}}\right)$
- It is often useful to express the probability of error (and other quantities) in terms of the signal-to-noise ratio (SNR).
- $\text{SNR} = \frac{\text{Average Received Signal Energy (per Complex Sample)}}{\text{Average Noise Energy (per Complex Sample)}}$
- For this simple BPSK scenario,
 $\text{Average Received Signal Energy} = \mathbb{E}[|x[0]|^2] = a^2$
 $\text{Average Noise Energy} = \mathbb{E}[|w[0]|^2] = N_0$

$\text{SNR} = \frac{a^2}{N_0}$
- The probability of error is $\mathbb{P}(\hat{x}[0] \neq x[0]) = Q(\sqrt{2\text{SNR}})$

AWGN Channels and BPSK

- The probability of error is $\mathbb{P}(\hat{x}[0] \neq x[0]) = Q(\sqrt{2\text{SNR}})$
- Let's develop some intuition about how fast the probability of error decays with SNR.

- Upper and lower bounds on Q-function:

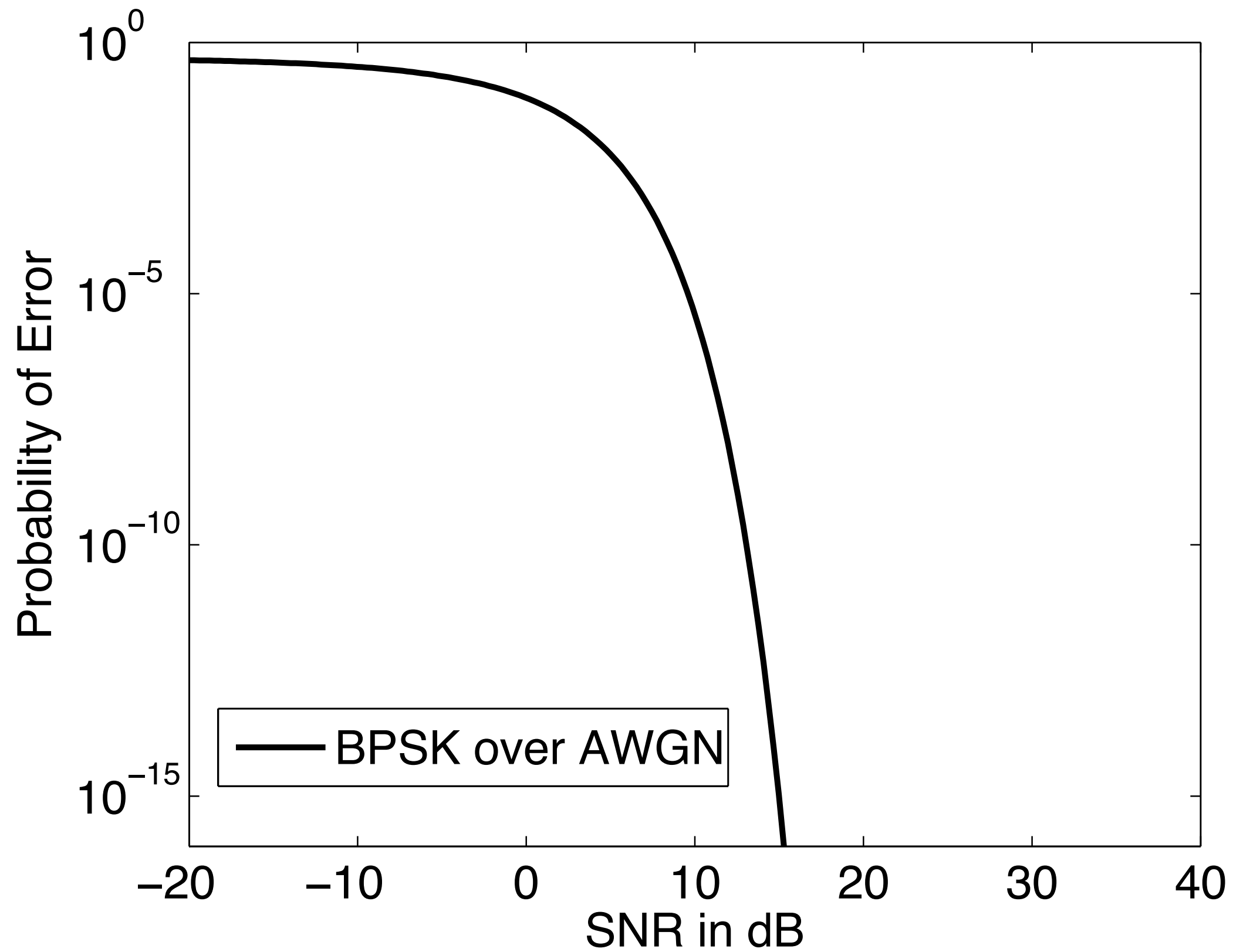
$$Q(v) < \exp\left(-\frac{v^2}{2}\right) \quad \text{for } v > 0$$
$$Q(v) > \frac{1}{v\sqrt{2\pi}}\left(1 - \frac{1}{v^2}\right) \exp\left(-\frac{v^2}{2}\right) \quad \text{for } v > 1$$

- Leads to **upper bound** on the probability of error:

$$\mathbb{P}(\hat{x}[0] \neq x[0]) = Q(\sqrt{2\text{SNR}}) < \exp(-\text{SNR})$$

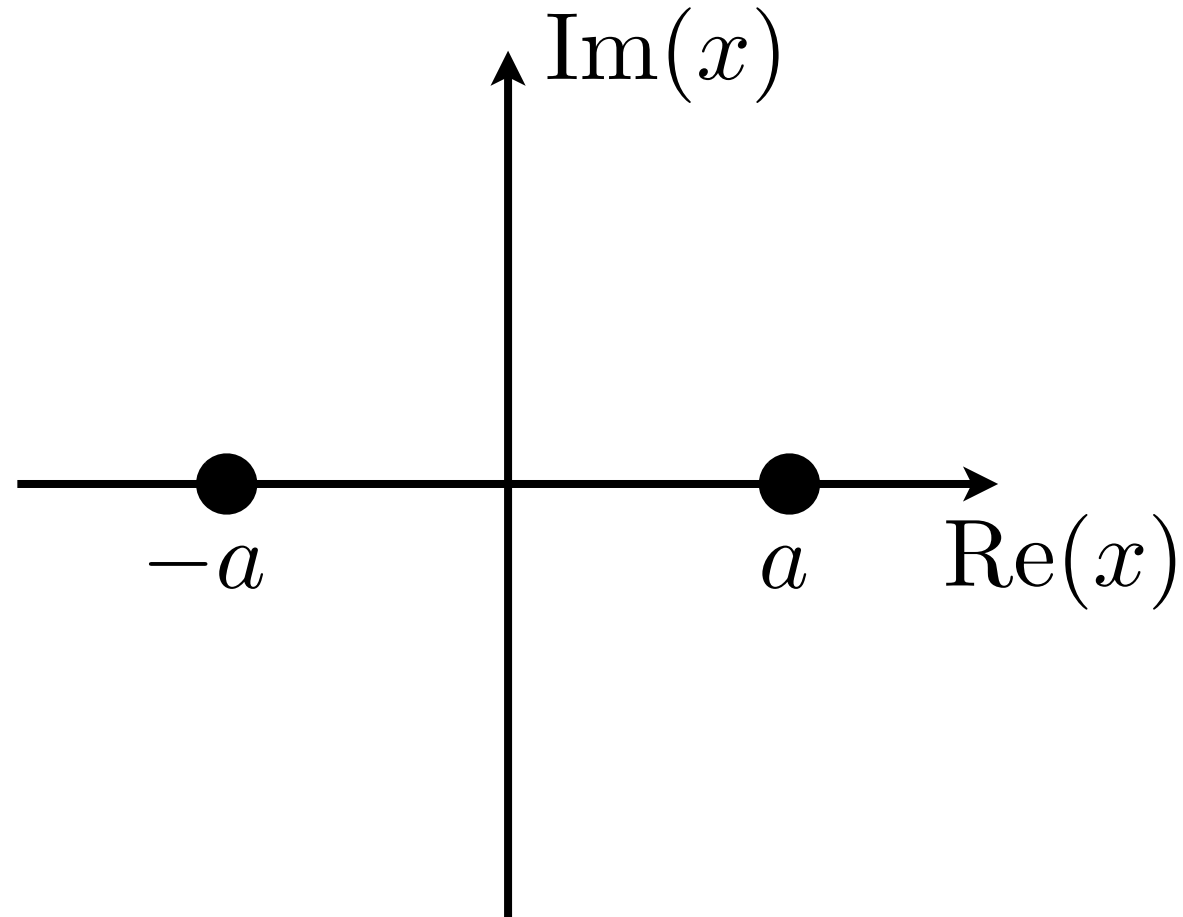
- This is great! Probability of error falls exponentially fast with the SNR.

AWGN Channels and BPSK



Noncoherent Rayleigh Fading Channels and BPSK

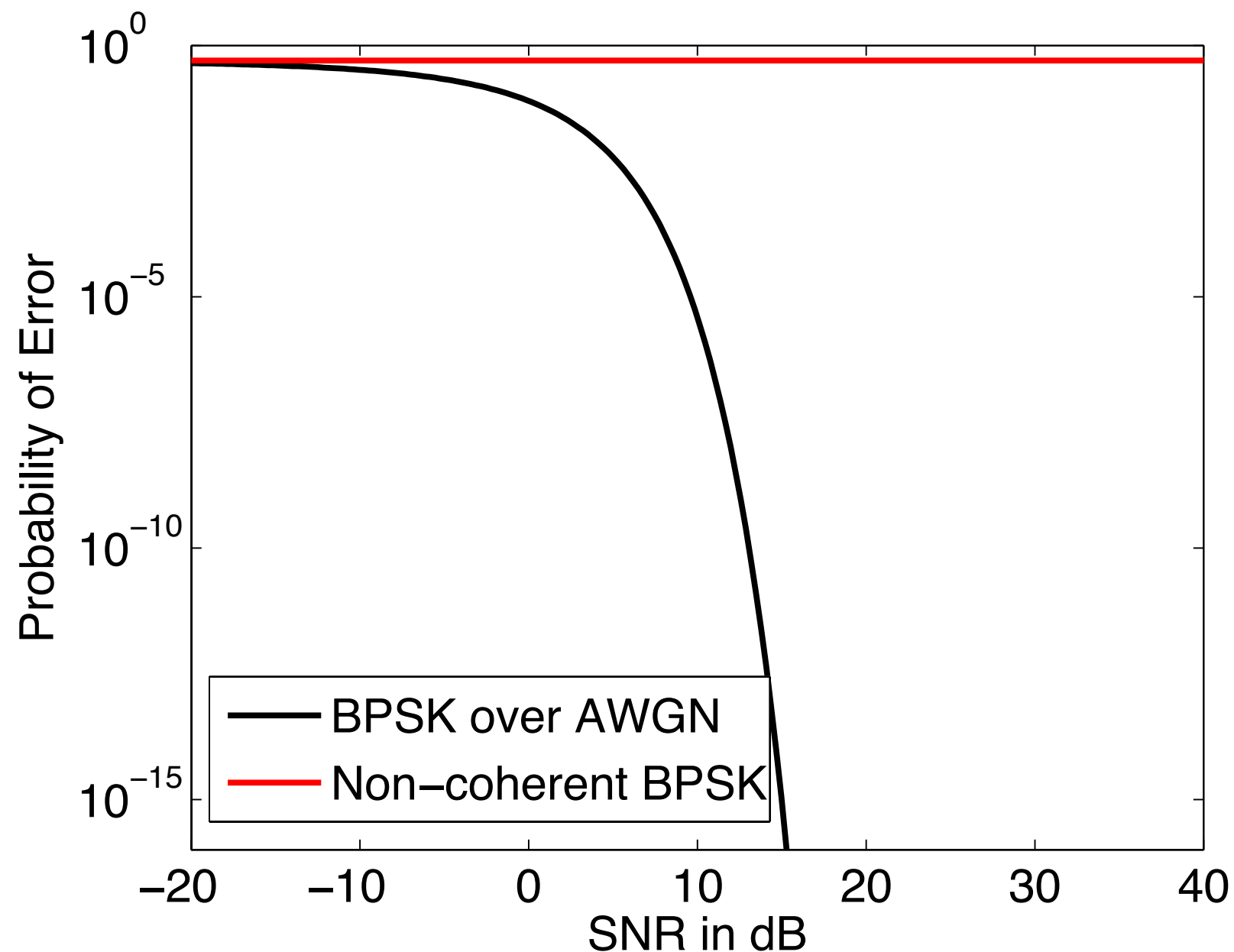
- Now that we have a solid understanding of how BPSK performs without fading, let's take a careful look at what happens under **Rayleigh fading**, $y[m] = h[m]x[m] + w[m]$
- By **Rayleigh fading**, we mean that $h[m]$ is i.i.d. across time and distributed as $h[m] \sim \mathcal{CN}(0, 1)$



- Let's assume that the channel is **non-coherent**, meaning that neither the transmitter nor the receiver know the realization of $h[m]$ (but they do know its statistics).

Noncoherent Rayleigh Fading Channels and BPSK

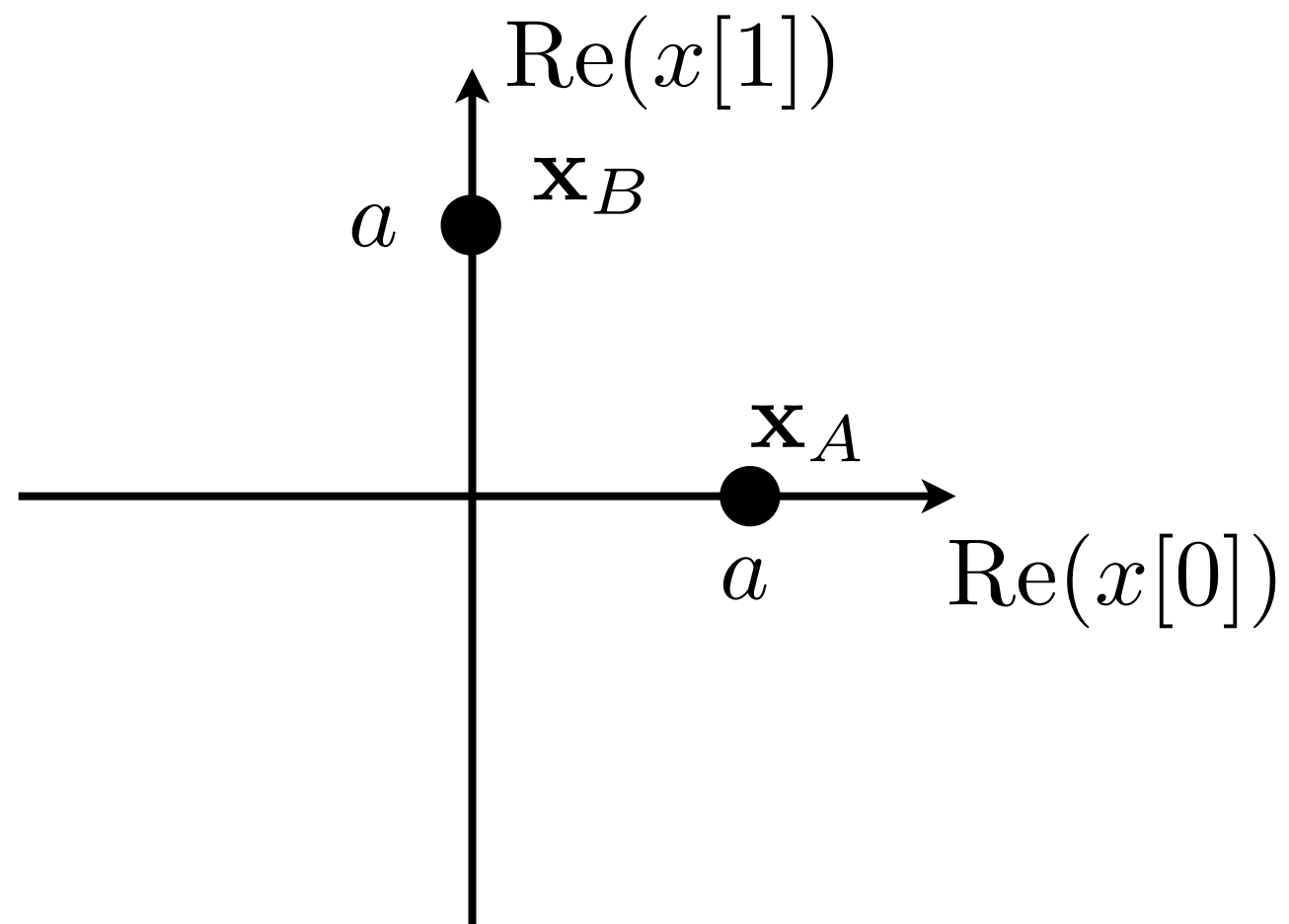
- ML Decision Rule: $\hat{x}[0] = a$
- Probability of Error: $\mathbb{P}(\hat{x}[0] \neq x[0]) = \frac{1}{2}$



Noncoherent Rayleigh Fading and Orthogonal Signaling

- Since the channel is **non-coherent**, we cannot encode information in the phase.
- Consider the following **orthogonal** signaling scheme over two time slots.

$\mathbf{x} = \begin{pmatrix} x[0] \\ x[1] \end{pmatrix}$ is equally likely to be $\mathbf{x}_A = \begin{pmatrix} a \\ 0 \end{pmatrix}$ or $\mathbf{x}_B = \begin{pmatrix} 0 \\ a \end{pmatrix}$



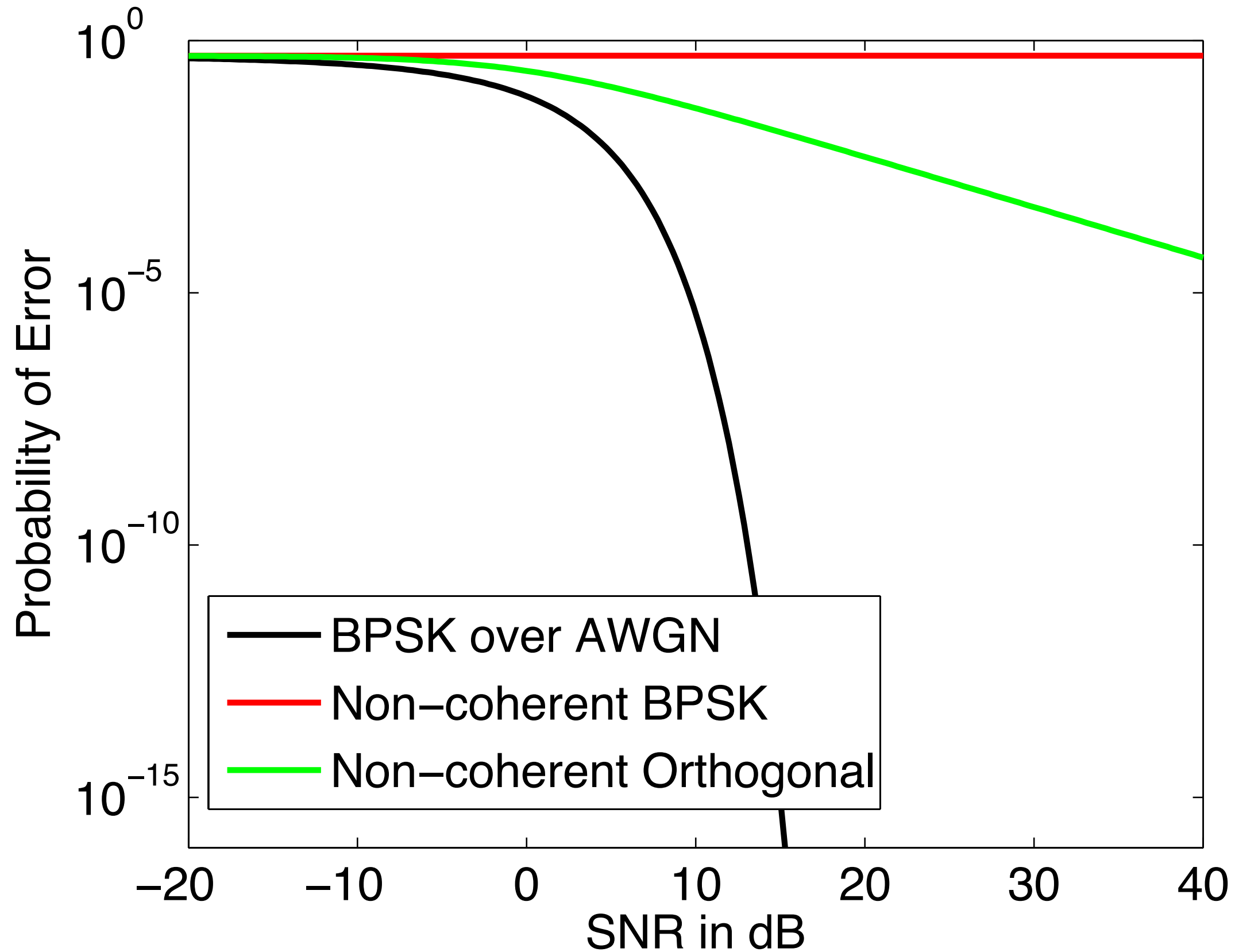
Noncoherent Rayleigh Fading and Orthogonal Signaling

- ML Decision Rule: $\hat{\mathbf{x}} = \begin{cases} \mathbf{x}_A & |y[0]|^2 \geq |y[1]|^2 \\ \mathbf{x}_B & |y[0]|^2 < |y[1]|^2 \end{cases},$

$$\begin{aligned} \text{Average Received Signal Energy} &= \mathbb{E}[|x[0]|^2] = \frac{a^2}{2} \\ \text{Average Noise Energy} &= \mathbb{E}[|w[0]|^2] = N_0 \end{aligned} \quad \boxed{\text{SNR} = \frac{a^2}{2N_0}}$$

- The probability of error is $\mathbb{P}(\hat{\mathbf{x}} \neq \mathbf{x}) = \frac{1}{2 + \frac{a^2}{N_0}} = \frac{1}{2(1 + \text{SNR})}$
- Overall, the error decays like $1/\text{SNR}$ rather than $\exp(-\text{SNR})$!
- Maybe this is because we have assumed the channel is non-coherent...

Noncoherent Rayleigh Fading and Orthogonal Signaling

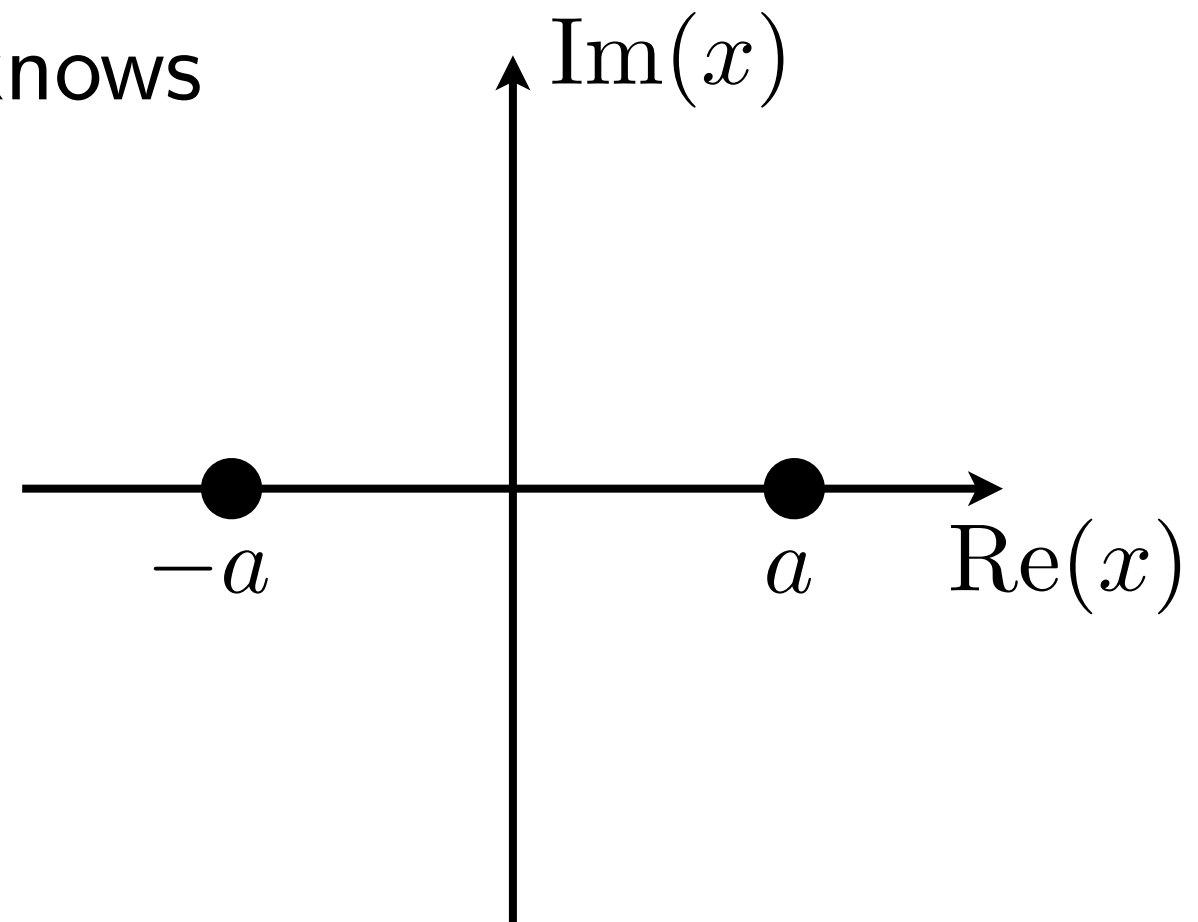


Coherent Rayleigh Fading Channels and BPSK

- Let's go back to BPSK and now assume that we have a coherent Rayleigh fading channel

$$y[m] = h[m]x[m] + w[m]$$

- By coherent, we mean that the receiver knows $h[m]$ perfectly while the transmitter does not (but it still knows the channel statistics).



Coherent Rayleigh Fading Channels and BPSK

- ML Decision Rule:
$$\hat{x}[0] = \begin{cases} a & \frac{h^*[0]}{|h[0]|} y[0] \geq 0 \\ -a & \frac{h^*[0]}{|h[0]|} y[0] < 0 \end{cases}$$

Average Received Signal Energy $= \mathbb{E}[|x[0]|^2] = a^2$

Average Noise Energy $= \mathbb{E}[|w[0]|^2] = N_0$

$$\text{SNR} = \frac{a^2}{N_0}$$

- The probability of error is

$$\mathbb{P}(\hat{x}[0] \neq x[0]) = \frac{1}{2} \left(1 - \sqrt{\frac{\text{SNR}}{1 + \text{SNR}}} \right) \approx \frac{1}{4\text{SNR}}$$

- The error still decays like $1/\text{SNR}$ rather than $\exp(-\text{SNR})$!
- So whether or not we know the channel at the receiver, the error scaling is fundamentally different with fading.

Coherent Rayleigh Fading Channels and BPSK

