

Subject: Segway Control Term Project – Part 1
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Introduction

By evaluating the governing dynamics of the Segway, the transfer function of outputs theta [$\Theta(t)$] and velocity [xVelocity(t)] were found. The written maple program to evaluate these functions is displayed in the appendix. The values of the variables that were used displayed below. **Note that the transfer function for theta based on voltage has a negative term in front of it so the PID gains can be positive.**

```

% Variable Initialization
% theta; % ~ Platform deviation from vertical
% x; % ~ Horizontal displacement of Segway vehicle
% V; % ~ Motor voltage
% T; % ~ Motor torque
% d; % ~ Horizontal displacement of rider load/mass (m)

R = 0.5; % C % Wheel radius (m)
L = 0.4; % C % Distance from wheel centerline to center of mass (m)
l = 1; % C % Distance from wheel centerline to rider load/mass (m)
M = 27; % C % System (vehicle) mass (~120 lbs)
m = 33.72; % C % Rider (load) mass(N) (~150 lbs)
k_t = 0.75; % C % Torque constant (N*m/A)
k_bemf = 0.5; % C % Back emf constant (V*sec)
r_a = 1.4; % C % Armature resistance
c1 = 0.01; % C % Small rotational damping with appropriate units
c2 = 0.01; % C % Small linear damping with appropriate units
g = 9.81; % C % Gravitational acceleration (m*s^2)
J = M*L^2+m*l^2; % C % Moment of inertia related to all the mass rotating
around the wheels (kg*m^2) (calculated)
alpha = k_t/r_a; % C % Motor constant (calculated)
beta = k_t*k_bemf/(R*r_a); % C % Motor constant (calculated)

```

Figure 1 – Variables, constants, and Equations

Task 1

```

invlaplace(G_xVelocity,s,t);
0.008710801394 e-0.008792102207 t

simplify(invlaplace(G_thetaV,s,t));
-4.647029483 10-8 e-0.008792102207 t
+ 4.647029483 10-8 e-0.00006510416667 t cosh(3.390588998 t)
+ 0.002039375548 e-0.00006510416667 t sinh(3.390588998 t)

simplify(invlaplace(G_thetaD,s,t));
-2.561778353 e-0.00006510416667 t sinh(3.390588998 t)

```

Figure 2 – Laplace Representations of $\Theta(t)$ and $v(t)$

Task 2

The Routh table analysis of the open looped system shows that velocity will be stable, but Θ will be unstable. However, the closed loop systems show that the velocity will be stable and the stability of Θ will be dependent of both the disturbance (d) and voltage (v).

```
> RouthTable(denom(G_xVelocity), s);

$$\begin{bmatrix} \frac{861}{20} & s \\ \frac{757}{2000} & 1 \end{bmatrix}$$

RouthTable(denom(G_xVelocity), s, 'stablecondition' = true);

$$true$$

RouthTable(denom(G_theta), s);

$$\begin{bmatrix} \frac{82656}{25} & -\frac{7601768243}{200000} s^3 \\ \frac{294993}{10000} & -\frac{6683553}{20000} s^2 \\ -\frac{3636154692811}{6555400000} & 0 & s \\ -\frac{6683553}{20000} & 0 & 1 \end{bmatrix}$$

RouthTable(denom(G_theta), s, 'stablecondition' = true);

$$false$$

```

Figure 3 - Open Loop Routh Table Analysis for Velocity

```
> RouthTable(denom(tf_theta), s);

$$\begin{aligned} & [[330624000000000000000000, -380088412200000000000000 \\ & + 228616071400000000 V - 28717794000000000000 d, s^3], \\ & [294993000000000000, -334177650000000000000000 \\ & + 2027678571000000 V - 25248978000000000000 d, s^2], \\ & \left[ -\frac{181807736279400000000000}{32777} + \frac{4446897385220000000}{32777} V \right. \\ & \left. - \frac{137366781300000000000000}{32777} d, 0, s \right], \\ & [-33417765000000000000 + 2027678571000000 V \\ & - 25248978000000000000 d, 0, 1]] \end{aligned}$$

RouthTable(denom(tf_theta), s, 'stablecondition' = true);

$$0 < -909038681397000 + 222344869261 V - 686833906500000 d$$


$$\text{and } 0 < -371308500000 + 225297619 V - 280544200000 d$$

```

Figure 4 - Closed Loop Routh Table Analysis for Velocity and Θ

Task 3

The program that evaluated the closed loop transfer functions of the system with a proportional, intergral, and derivative (PID) controller was written in maple. The maple program that was created is displayed in the appendix. Arbitrary gain values were created, where the system will remain stable according to the Routh table analysis. The chosen PID gains were verified with the equations derived from the Routh table analysis.

```
> CE := sort(simplify(denom(tf) ));  
4628.736000 s4 + 32.00625000 s3 kD + 41.29902000 s3  
+ 0.2838750000 s2 kD + 32.00625000 s2 kP - 53212.37770 s2  
+ 0.2838750000 s kP + 32.00625000 s ki - 467.8487100 s  
+ 0.2838750000 ki  
  
simplify(RouthTable(CE,s,'stablecondition'=true));  
0 < ki and 0 < 2753268 + 2133750 kD and 0 < -2128495108  
+ 1280250 kP + 11355 kD  
- 
$$\frac{185149440 (-31189914 + 18925 k_P + 2133750 k_i)}{2753268 + 2133750 k_D} \text{ and } 0$$
  
< -31189914 + 18925 kP + 2133750 ki - (11355 (2753268  
+ 2133750 kD) ki) 
$$\left/ \left( -2128495108 + 1280250 k_P + 11355 k_D \right. \right.$$
  
- 
$$\left. \left. \frac{185149440 (-31189914 + 18925 k_P + 2133750 k_i)}{2753268 + 2133750 k_D} \right) \right)$$
  
  
# Arbitrary controller values were selected to satisfy the stable  
condition requirements as required by the Routh Table analysis.  
kP := 1800;  
ki := 20;  
kD := 80;  
1800  
20  
80  
simplify(RouthTable(CE,s,'stablecondition'=true));  
true
```

Figure 5 - Closed Loop Routh Table Analysis with a PID Controller

Task 4

By implementing the PID controller created in task 3 a stable voltage input for the velocity transfer function was created. The input voltage is the response to stabilize the Segway and counteract the disturbance.

Based on the Routh table analysis in task 2, the open loop velocity system is expected to be stable without additional controls. This is determined to be true by the simulation performed with an open loop velocity system.

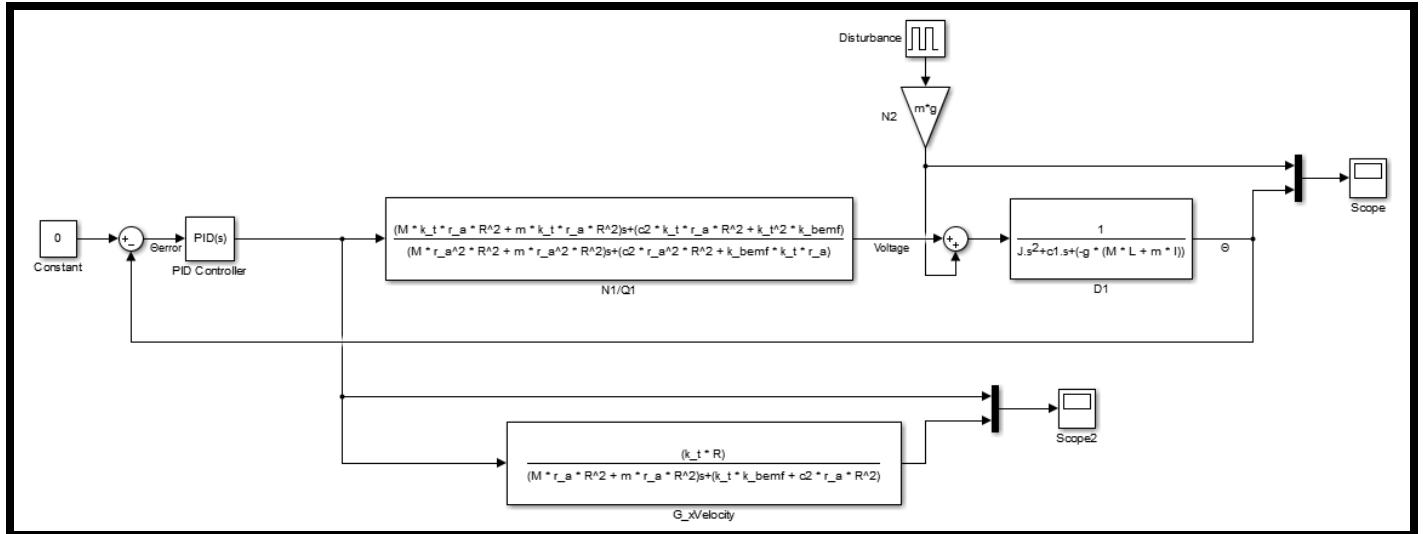


Figure 6 - Open Loop Velocity Stability Analysis

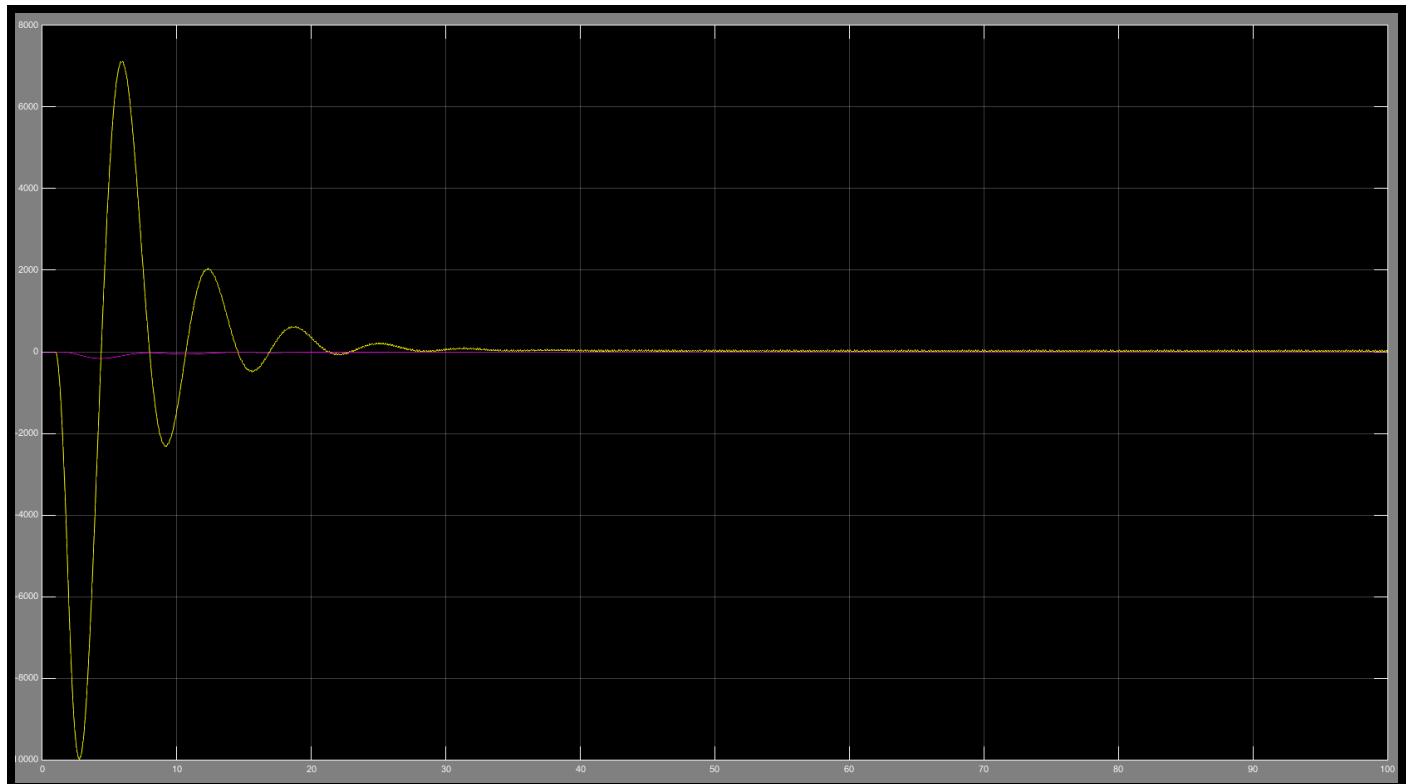


Figure 7 - Projected Open Loop Velocity Output

Task 5

The root locus analysis was performed through multiple series to refine the kp, ki, and kd gains. 2 out of the 3 gains are set to predetermined values, and the optimal value for the last gain will be determined using root locus analysis in MATLAB.

> # Arbitrary controller values were selected to satisfy the stable condition requirements as required by the Routh Table analysis.

$$k_P;$$
$$k_i := 20;$$
$$k_D := 80;$$

$$k_P$$

$$20$$

$$80$$

CE

$$4628.736000 s^4 + 2601.799020 s^3 - 53189.66770 s^2 + 32.00625000 s^2 k_P + 0.2838750000 s k_P + 172.2762900 s + 5.677500000$$

Figure 10 – CE when $ki=20$ and $kd=80$

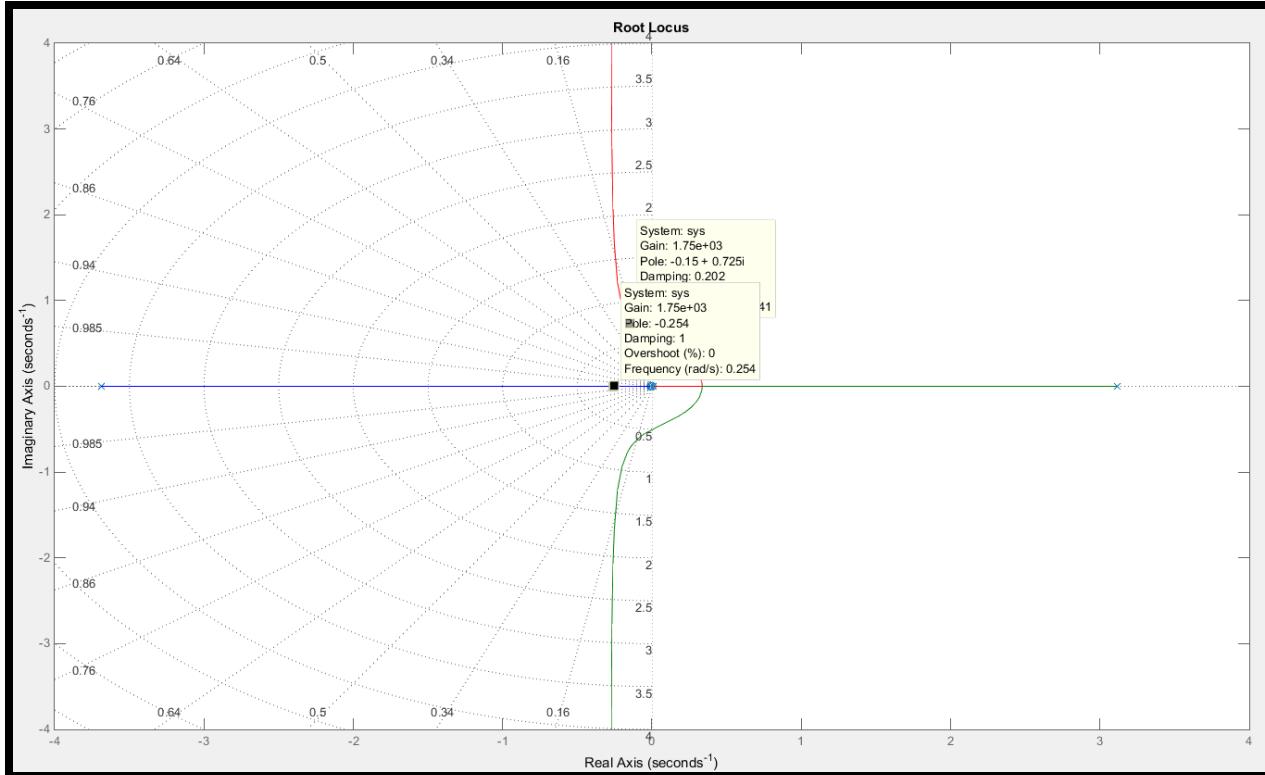


Figure 11 – Root Locus Analysis when $ki=20$ and $kd=80$

It seems that 1750 is the most appropriate kp gain when ki is set to 20 and kd is 80.

>

Arbitrary controller values were selected to satisfy the stable condition requirements as required by the Routh Table analysis.

$$k_P := 1750;$$

$$k_i;$$

$$k_D := 80;$$

CE

$$4628.736000 s^4 + 2601.799020 s^3 + 2821.26980 s^2 + 28.9325400 s \\ + 32.00625000 s k_i + 0.2838750000 k_i$$

Figure 12 – CE when $k_p=1750$ and $k_d=80$

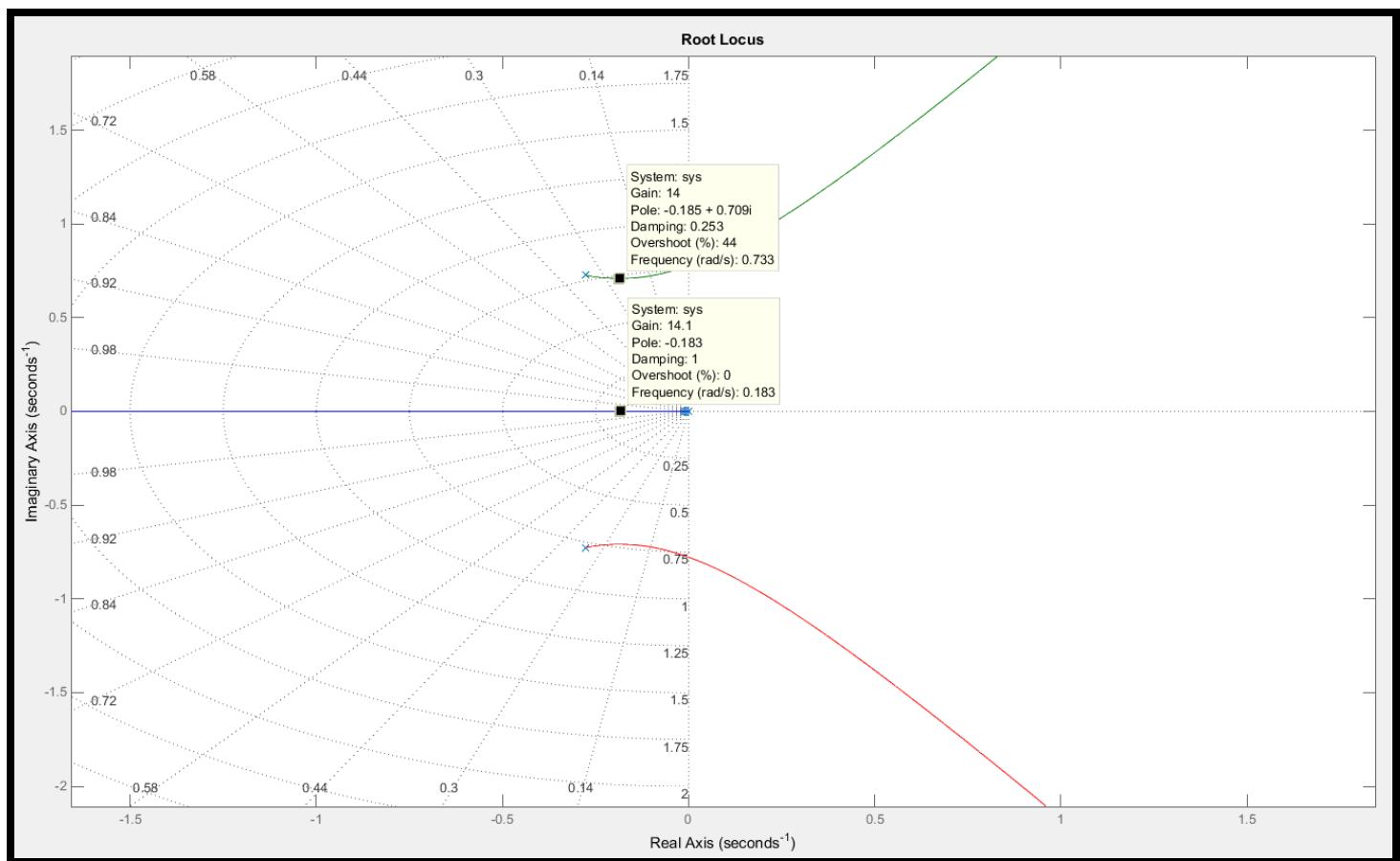


Figure 13 – Root Locus Analysis when $k_p=1750$ and $k_d=80$

It seems that 14 is the most appropriate k_i gain when k_p is set to 1750 and k_d is 80.

>

Arbitrary controller values were selected to satisfy the stable condition requirements as required by the Routh Table analysis.

$$k_P := 1750;$$

$$k_i := 14;$$

$$k_D;$$

$$1750$$

$$14$$

$$k_D$$

CE

$$\begin{aligned} & 4628.736000 s^4 + 32.00625000 s^3 k_D + 41.29902000 s^3 \\ & + 0.2838750000 s^2 k_D + 2798.55980 s^2 + 477.0200400 s \\ & + 3.974250000 \end{aligned}$$

Figure 14 – CE when $k_p=1750$ and $k_i=14$

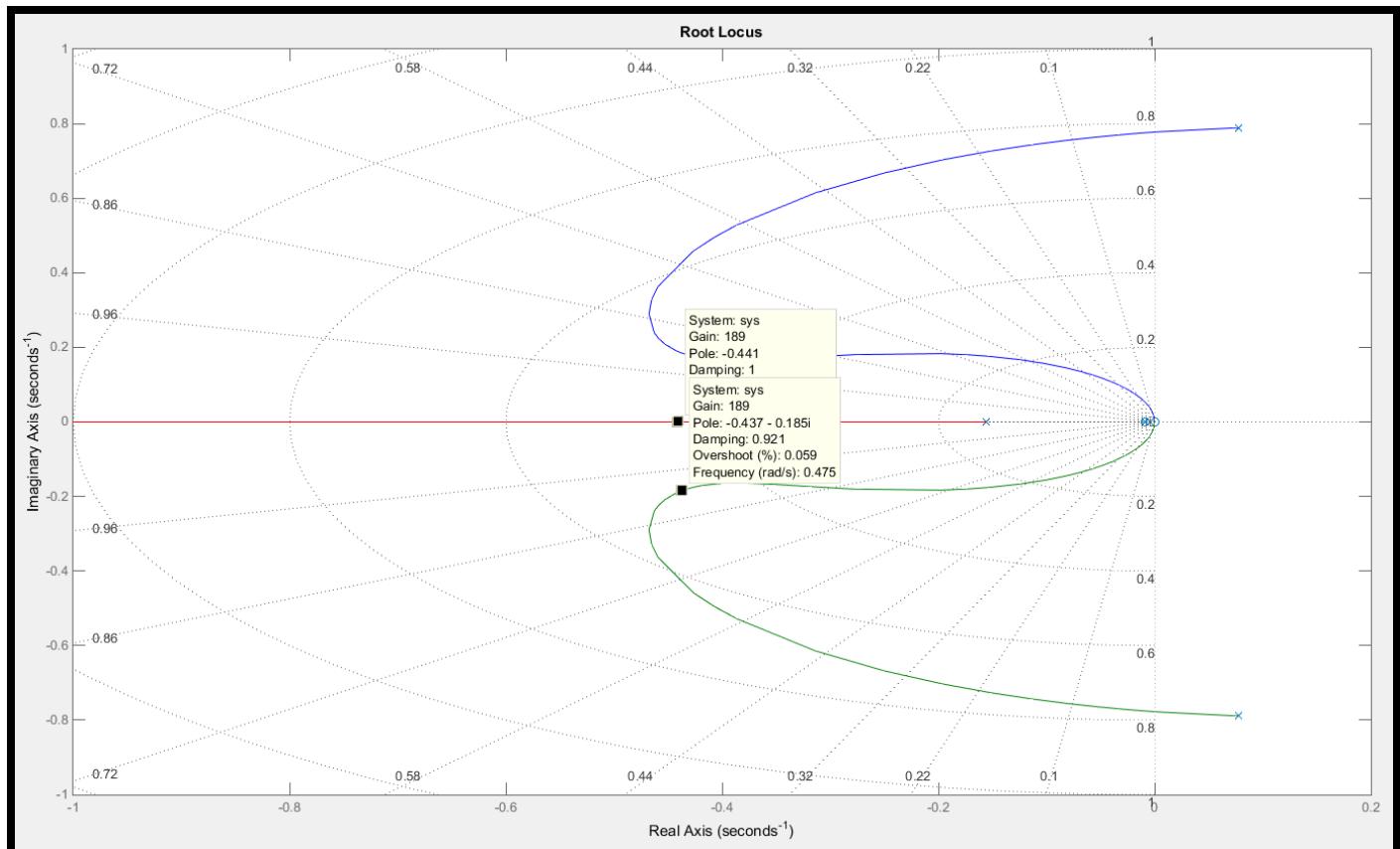


Figure 15 – Root Locus Analysis when $k_p=1750$ and $k_i=14$

It seems that 68.7 is the most appropriate kd gain when kp is set to 1750 and ki is 14.

Another series of root locus analysis was performed, refining the gains for kp, ki, and kd even further. The final values of the PID gains are:

$$kp = 1750;$$

$$ki = 14;$$

$$kd = 189$$

Figure 16 – Root Locus Analysis Optimized PID Gains

Task 6

(i) Impulse Disturbance

The system settles and the value of theta returns to zero after a few seconds. There is only one visible oscillation in the response, and this is likely due to inertia. The voltage demanded by the system is similar to the inverse of Θ .

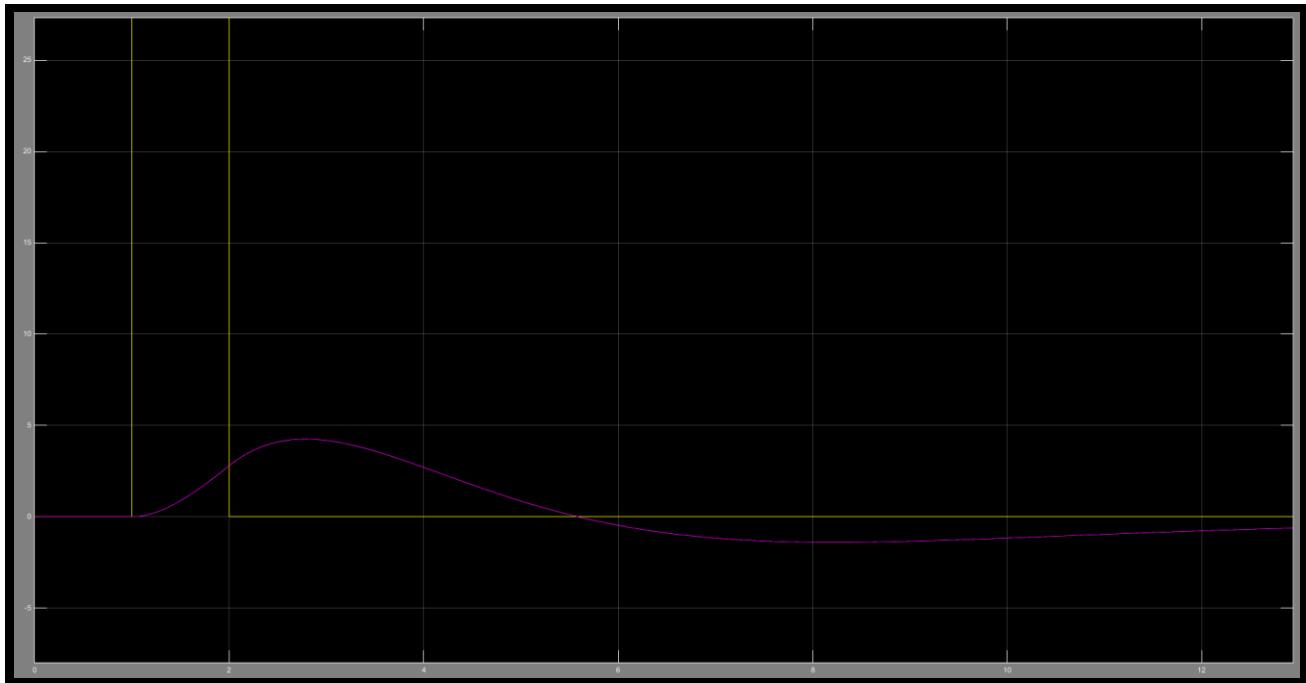


Figure 17 – Scope of θ after Impulse Disturbance

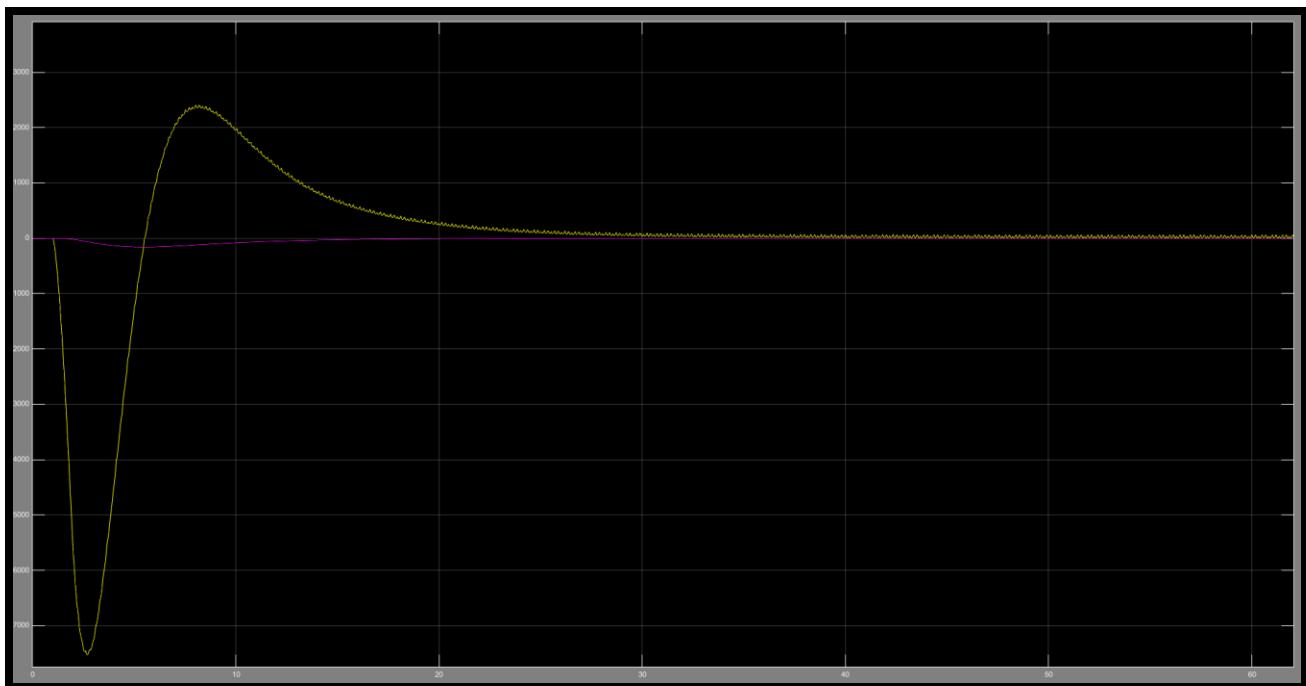


Figure 18 – Scope of velocity after Impulse Disturbance

(ii) Constant Step Disturbance

The system settles, but the value of Θ remains at a constant value that is not equal to zero. There seems to be a constant voltage demand by the system, and the velocity will eventually converge to a steady speed. The segway will be tilted at all time, and be moving either forward or backwards.

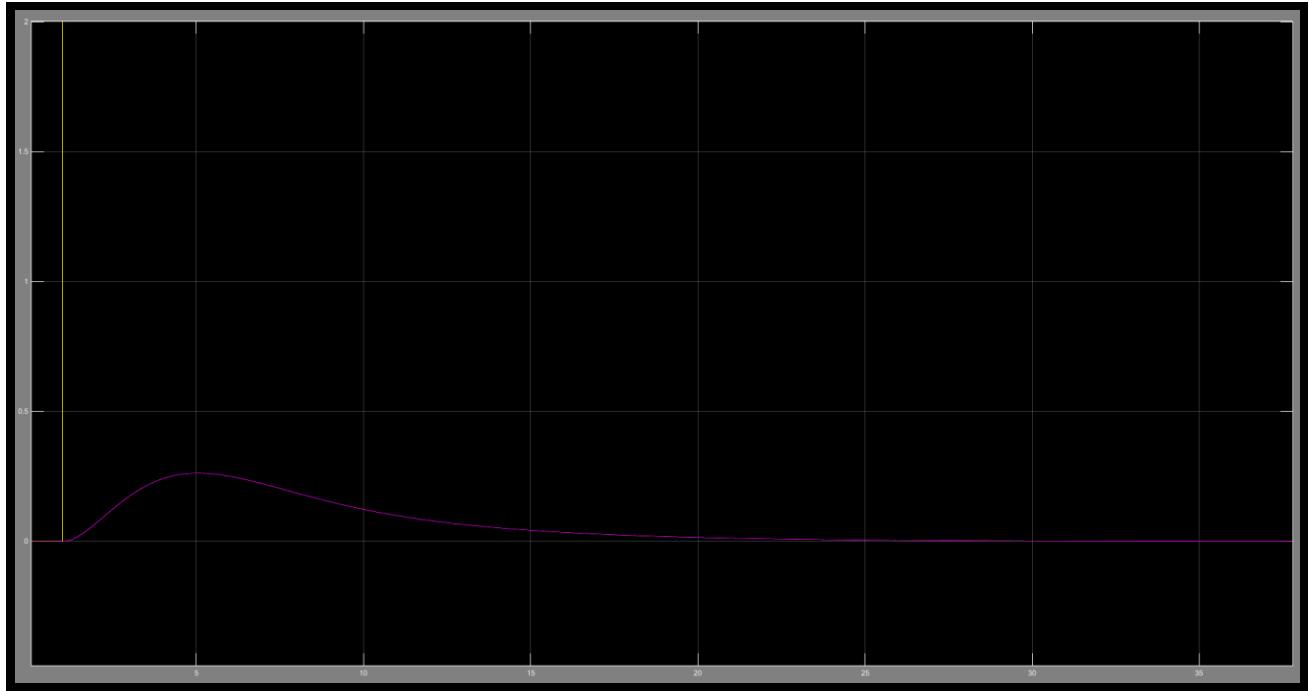


Figure 19 – Scope of Θ after Constant Disturbance

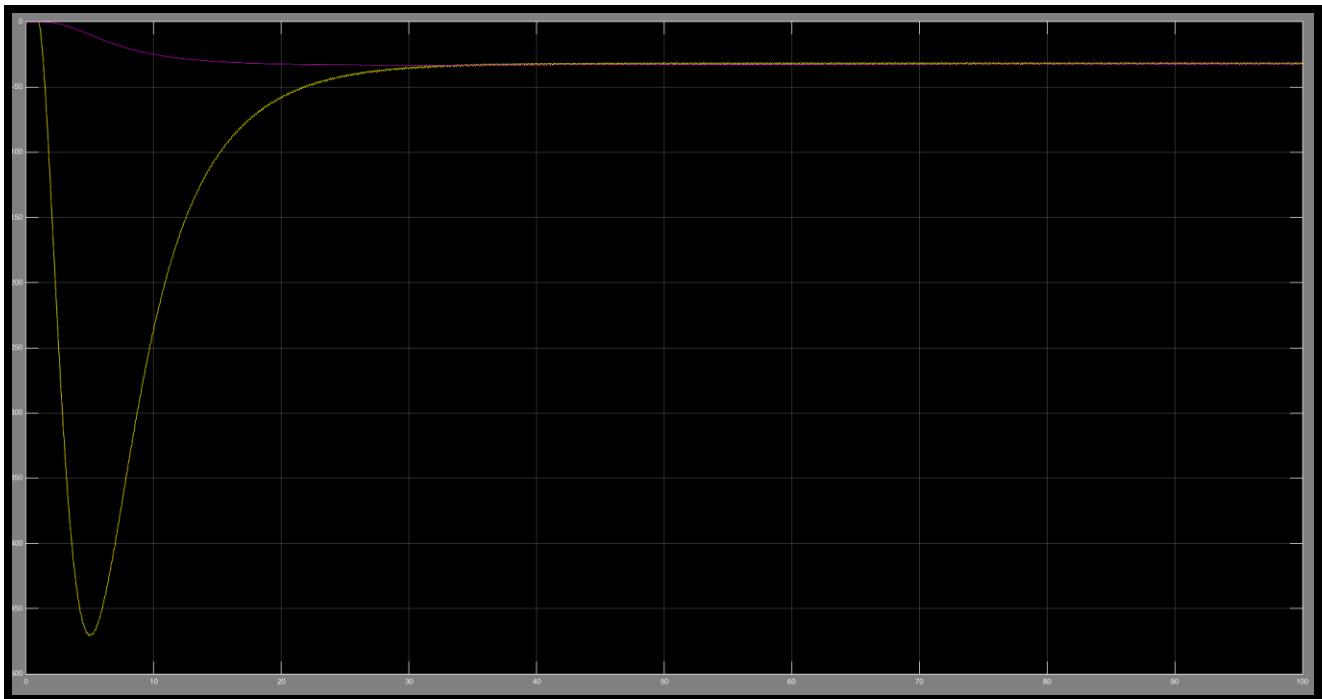


Figure 20 – Scope of velocity after Constant Disturbance

Task 7

It seems that with a saturated voltage, the system take slightly longer to settle. This is insignificant on the scope, but may alter performance and user experience greatly.

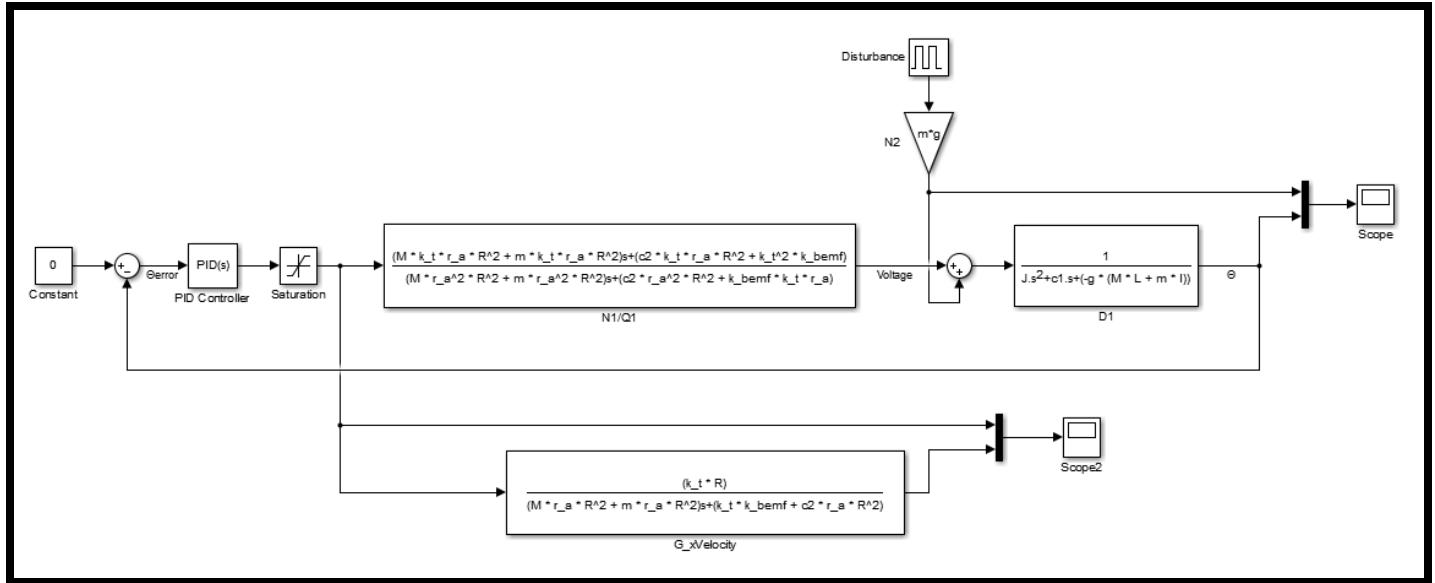


Figure 21 – Simulink Including a Saturation Block Limited Voltage

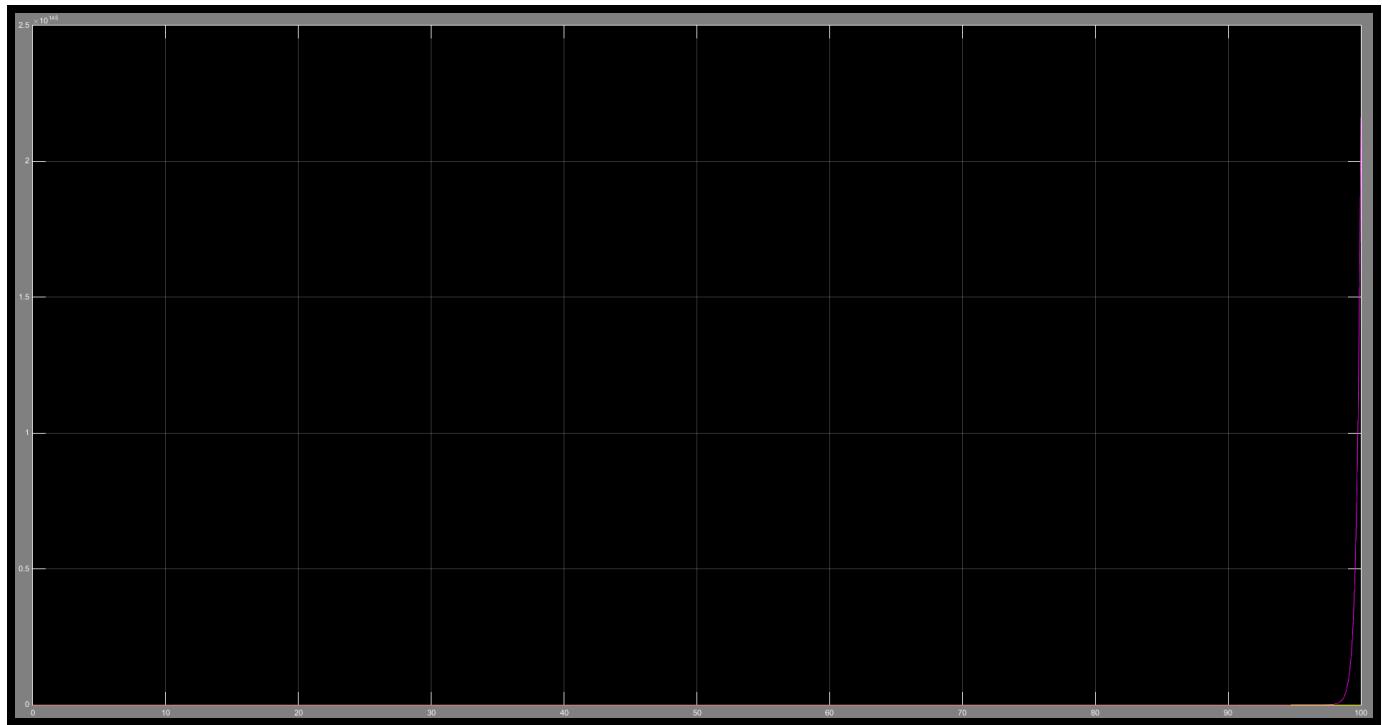


Figure 22 – Scope of θ after Impulse Disturbance with Voltage Saturation

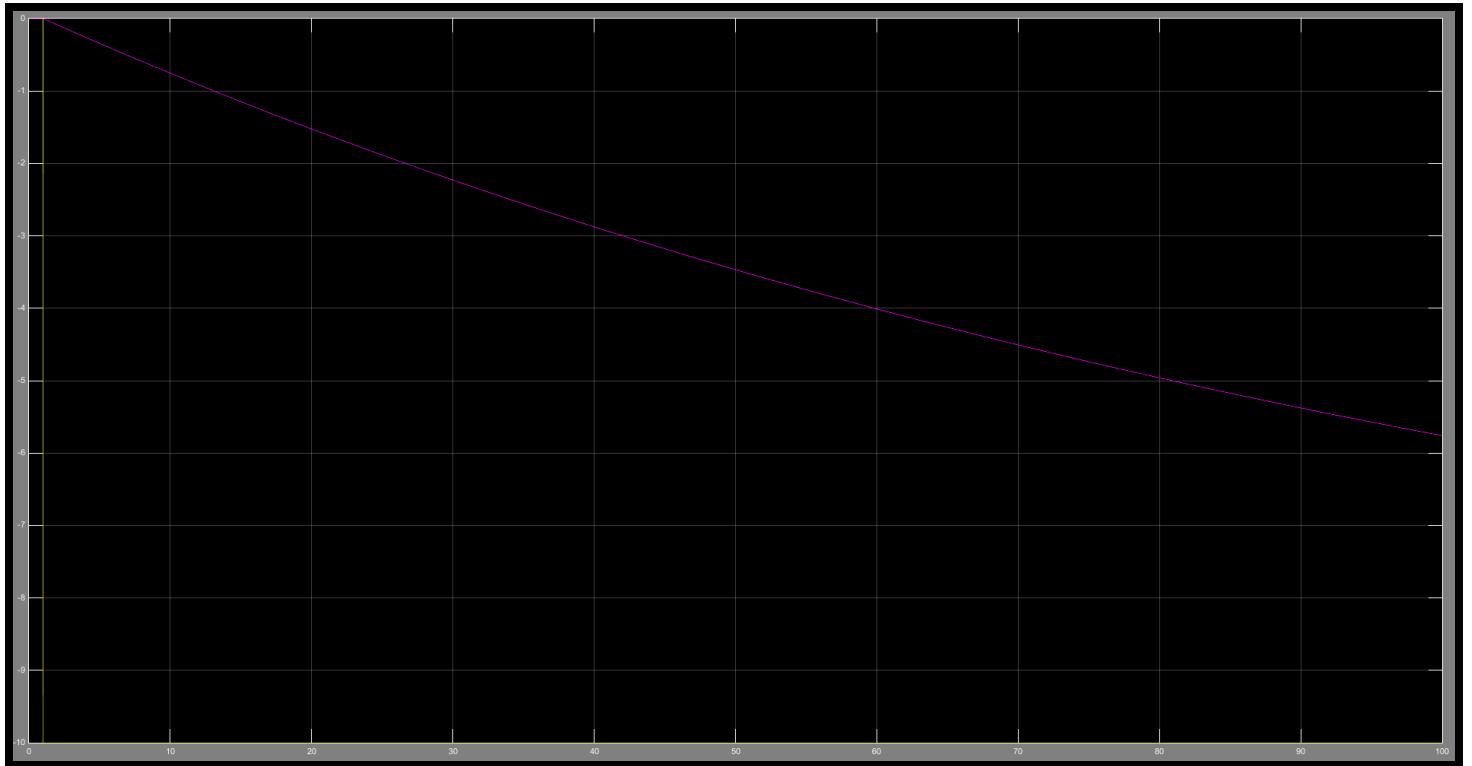


Figure 23 – Scope of Velocity after Impulse Disturbance with Voltage Saturation

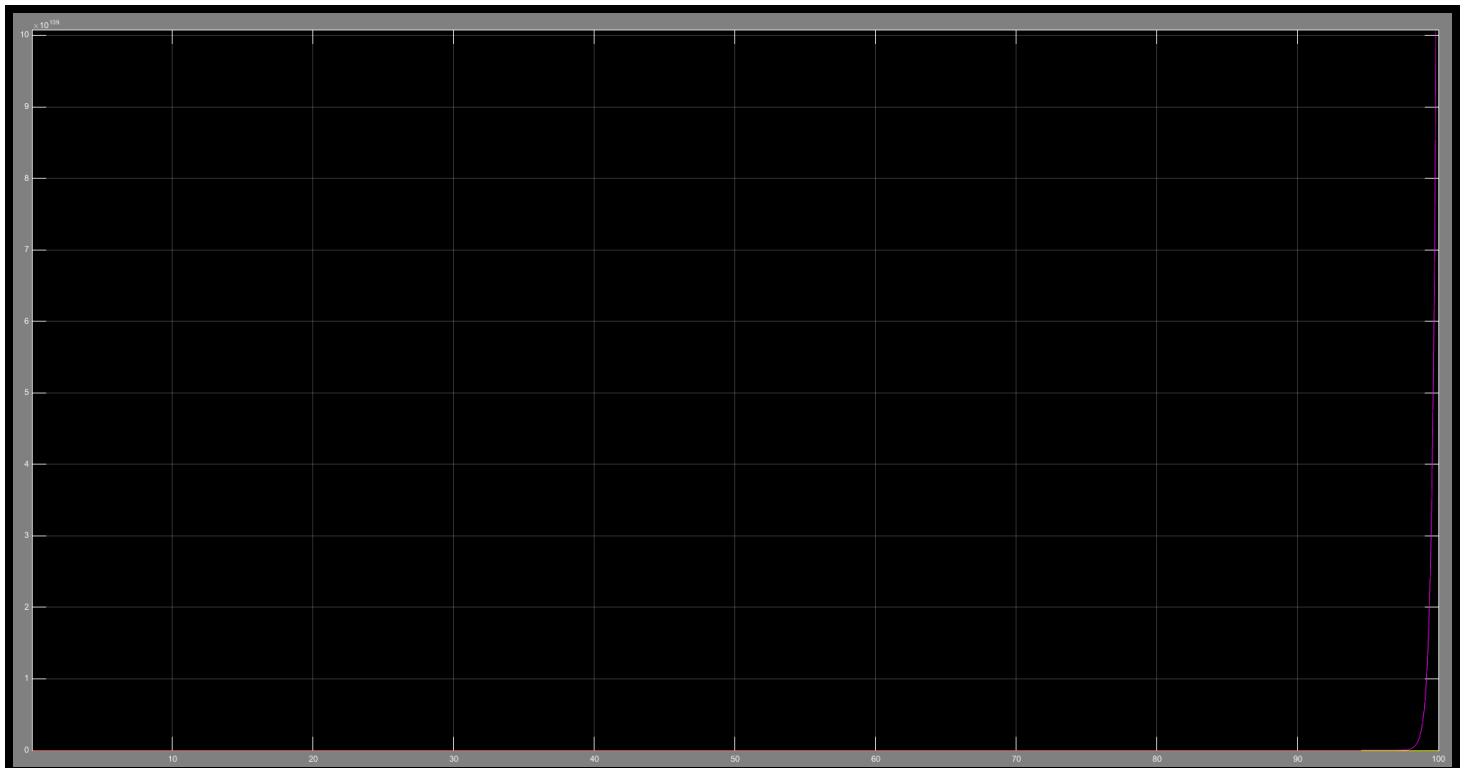


Figure 24 – Scope of Θ after Step Disturbance with Voltage Saturation

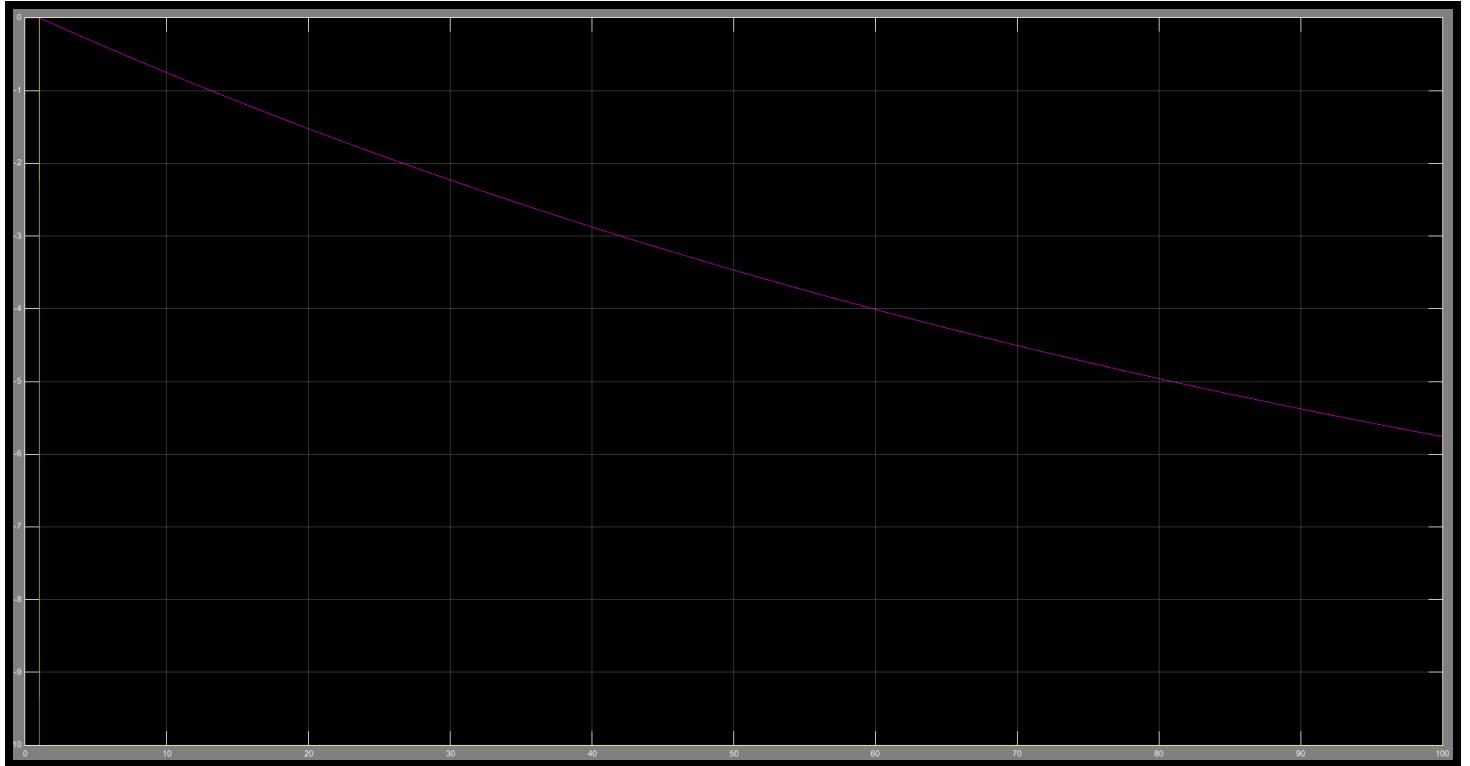


Figure 25– Scope of Velocity after Step Disturbance with Voltage Saturation

It seems that the segway cannot handle large disturbances due to the fact the voltage is saturated and can no longer deliver the power to restabilize the system, since the voltage is capped out at $\pm 10V$.

Appendix

(i) Calculate open loop tranfer functions

```
> restart; # Include Librarieswith(inttrans); with(DynamicSystems);
[addtable,fourier,fouriercos,fouriersin,hankel,hilbert,invfourier,
invhilbert,invlaplace,invmellin,laplace,mellin,savetable]

[AlgEquation,BodePlot,CharacteristicPolynomial,Chirp,
Coefficients,ControllabilityMatrix,Controllable,DiffEquation,
DiscretePlot,EquilibriumPoint,FrequencyResponse,GainMargin,
Grammians,ImpulseResponse,ImpulseResponsePlot,IsSystem,
Linearize,MagnitudePlot,NewSystem,NicholsPlot,NyquistPlot,
ObservabilityMatrix,Observable,PhaseMargin,PhasePlot,
PrintSystem,Ramp,ResponsePlot,RootContourPlot,
RootLocusPlot,RouthTable,SSModelReduction,
SSTransformation,Simulate,Sinc,Sine,Square,StateSpace,Step,
StepProperties,System,SystemConnect,SystemOptions,
ToDiscrete,TransferFunction,Triangle,Verify,ZeroPoleGain,
ZeroPolePlot]
```

summation of all moments

$$\text{sumEqn}_{\theta} := J \cdot \theta \cdot s^2 = M \cdot g \cdot L \cdot \theta + m \cdot g \cdot l \cdot \theta - T - c_1 \cdot \theta \cdot s + m \cdot g \cdot d;$$

$$J \theta s^2 = M g L \theta + m g l \theta - T - c_1 \theta s + m g d$$

summation of all forces in the x-direction

$$\text{sumEqn}_{velocity} := (M + m) \cdot x \cdot s^2 = \frac{T}{R} - c_2 \cdot x \cdot s;$$

$$(M + m) x s^2 = \frac{T}{R} - c_2 x s$$

Predefined Equations

$$\beta := \frac{k_t \cdot k_{bemf}}{R \cdot r_a};$$

$$\frac{k_t k_{bemf}}{R r_a}$$

$$\alpha := \frac{k_t}{r_a};$$

$$\frac{k_t}{r_a}$$

$$T := \alpha \cdot V - \beta \cdot x \cdot s;$$

$$\frac{k_t V}{r_a} - \frac{k_t k_{bemf} x s}{R r_a}$$

This section solves for the transfer function of the velocity

$$G_xPosition := \text{solve}(\text{sumEqn}_{velocity}, x);$$

$$\begin{aligned}
& \frac{k_t V R}{s(k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m)} \\
G_{xVelocity} &:= \frac{G_{xPosition} \cdot s}{V}; \\
& \frac{k_t R}{k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m} \\
\# This section solves for the transfer function of theta \\
x &:= G_{xVelocity} \cdot V; \\
& \frac{k_t R V}{k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m} \\
G_\theta &:= -solve(sumEqn_\theta, \theta); \\
& \left(-mg dr_a k_t k_{bemf} - mg dr_a^2 c_2 R^2 - mg dr_a^2 s R^2 M - m^2 g dr_a^2 s R^2 + \right. \\
& \quad k_t^2 V k_{bemf} + k_t V c_2 R^2 r_a + k_t V s R^2 r_a M + k_t V s R^2 r_a m - \\
& \quad \left. k_t^2 k_{bemf} V s \right) / \left(r_a (k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m) \left(-MgL - mgl + c_1 s + Js^2 \right) \right) \\
\# This simplifies the TF of theeta into two components (V & d) \\
G_{\theta V} &:= coeff(G_\theta, V); \\
& \left(k_t^2 k_{bemf} + k_t c_2 R^2 r_a + k_t s R^2 r_a M + k_t s R^2 r_a m - k_t^2 k_{bemf} s \right) / \\
& \quad \left(r_a (k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m) \left(-MgL - mgl \right. \right. \\
& \quad \left. \left. + c_1 s + Js^2 \right) \right) \\
G_{\theta d} &:= simplify(coeff(G_\theta, d)); \\
& -\frac{mg}{-MgL - mgl + c_1 s + Js^2}
\end{aligned}$$

(ii) Calculate closed loop transfer functions

> # This breaks down the transfer functions, separating the disturbance from the model equation

$$N2 := \text{numer}(G_{\theta d});$$

$$-mg$$

$$D1 := \text{denom}(G_{\theta d});$$

$$-MgL - mg l + c_1 s + Js^2$$

$$N1 := \text{numer}(G_{\theta V});$$

$$k_t(k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m - k_t k_{bemf} s)$$

$$Q1 := \text{denom}(G_{\theta V} \cdot D1);$$

$$r_a(k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m)$$

This creates a Closed Loop transfer function with PID Controller for theta

$$C := k_P + \frac{k_i}{s} + k_D \cdot s;$$

$$k_P + \frac{k_i}{s} + k_D s$$

$$\theta = \text{simplify} \left(\frac{1}{1 + \frac{C \cdot NI}{QI \cdot DI}} \cdot d \cdot m \cdot g + \frac{1}{1 + \frac{C \cdot NI}{QI \cdot DI}} \theta_{ref} \right)$$

$$\begin{aligned} \theta = & \left(s r_a (k_t k_{bemf} + c_2 R^2 r_a + s R^2 r_a M + s R^2 r_a m) (-MgL - mg l \right. \\ & \left. + c_1 s + Js^2) (mgd + \theta_{ref}) \right) / (k_t k_P s^2 R^2 r_a M + k_t k_P s^2 R^2 r_a m \\ & + k_t k_i s R^2 r_a M + k_t k_i s R^2 r_a m + k_t k_D s^2 c_2 R^2 r_a + k_t k_D s^3 R^2 r_a M \\ & + k_t k_D s^3 R^2 r_a m - s^2 r_a^2 R^2 M^2 g L - s^2 r_a^2 R^2 m^2 g L \\ & + k_t k_P s c_2 R^2 r_a - s r_a k_t k_{bemf} M g L - s r_a k_t k_{bemf} m g L - s \\ & r_a^2 c_2 R^2 M g L - s r_a^2 c_2 R^2 m g L - s^2 r_a^2 R^2 M m g L - s^2 r_a^2 R^2 m M g L \\ & + k_P s k_t^2 k_{bemf} - k_P s^2 k_t^2 k_{bemf} - k_t k_t^2 k_{bemf} s + k_D s^2 k_t^2 k_{bemf} \\ & - k_D s^3 k_t^2 k_{bemf} + s^2 r_a^2 c_2 R^2 c_1 + s^3 r_a^2 c_2 R^2 J + s^3 r_a^2 R^2 M c_1 + s^4 \\ & r_a^2 R^2 M J + s^3 r_a^2 R^2 m c_1 + s^4 r_a^2 R^2 m J + k_t k_i c_2 R^2 r_a \\ & \left. + s^2 r_a k_t k_{bemf} c_1 + s^3 r_a k_t k_{bemf} J + k_i k_t^2 k_{bemf} \right) \end{aligned}$$

$$tf := \text{simplify} \left(\frac{C \cdot NI + d \cdot N2 \cdot QI}{DI \cdot QI + C \cdot NI} \right)$$

$$\begin{aligned}
& - \left(-k_P s k_t^2 k_{bemf} - k_t k_P s c_2 R^2 r_a - k_t k_P s^2 R^2 r_a M - k_t k_P s^2 R^2 r_a m \right. \\
& + k_P s^2 k_t^2 k_{bemf} - k_i k_t^2 k_{bemf} - k_t k_i c_2 R^2 r_a - k_t k_i s R^2 r_a M \\
& - k_t k_i s R^2 r_a m + k_i k_t^2 k_{bemf} s - k_D s^2 k_t^2 k_{bemf} - k_t k_D s^2 c_2 R^2 r_a \\
& - k_t k_D s^3 R^2 r_a M - k_t k_D s^3 R^2 r_a m + k_D s^3 k_t^2 k_{bemf} \\
& + d m g r_a s k_t k_{bemf} + d m g r_a^2 s c_2 R^2 + d m g r_a^2 s^2 R^2 M + d m^2 g \\
& r_a^2 s^2 R^2 \Big) / \left(k_t k_P s^2 R^2 r_a M + k_t k_P s^2 R^2 r_a m + k_t k_i s R^2 r_a M \right. \\
& + k_t k_i s R^2 r_a m + k_t k_D s^2 c_2 R^2 r_a + k_t k_D s^3 R^2 r_a M \\
& + k_t k_D s^3 R^2 r_a m - s^2 r_a^2 R^2 M^2 g L - s^2 r_a^2 R^2 m^2 g l \\
& + k_t k_P s c_2 R^2 r_a - s r_a k_t k_{bemf} M g L - s r_a k_t k_{bemf} m g l - s \\
& r_a^2 c_2 R^2 M g L - s r_a^2 c_2 R^2 m g l - s^2 r_a^2 R^2 M m g l - s^2 r_a^2 R^2 m M g L \\
& + k_P s k_t^2 k_{bemf} - k_P s^2 k_t^2 k_{bemf} - k_i k_t^2 k_{bemf} s + k_D s^2 k_t^2 k_{bemf} \\
& - k_D s^3 k_t^2 k_{bemf} + s^2 r_a^2 c_2 R^2 c_1 + s^3 r_a^2 c_2 R^2 J + s^3 r_a^2 R^2 M c_1 + s^4 \\
& r_a^2 R^2 M J + s^3 r_a^2 R^2 m c_1 + s^4 r_a^2 R^2 m J + k_t k_i c_2 R^2 r_a \\
& \left. + s^2 r_a k_t k_{bemf} c_1 + s^3 r_a k_t k_{bemf} J + k_i k_t^2 k_{bemf} \right)
\end{aligned}$$