

[illegible]

1

1

2

1

j

1

1

1

1

1



1

1

1

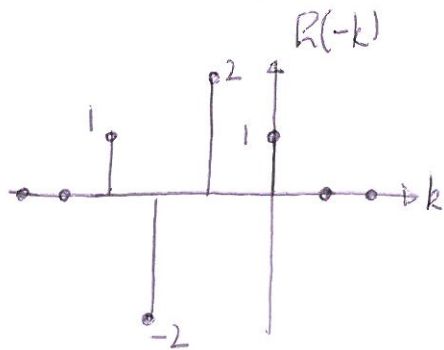
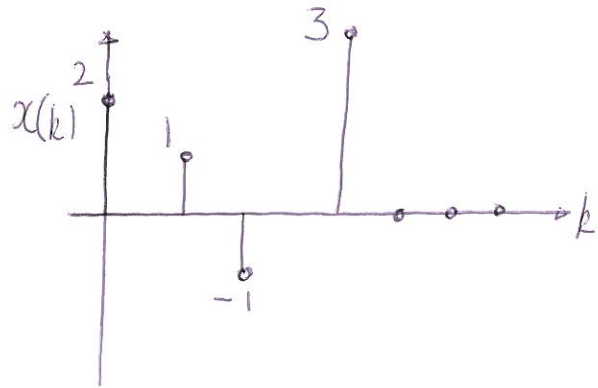
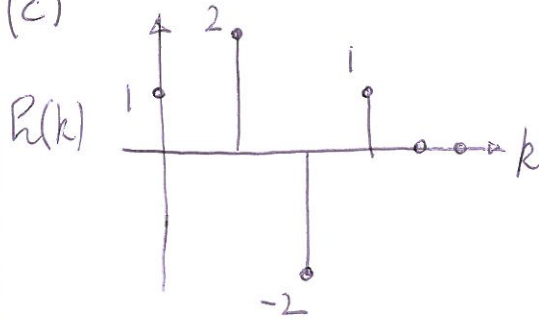
$$L_2 = \sqrt{(0.7071 - 0.2)^2 + (0.7071 + 0.2)^2} = 1.9734$$

$$L_3 = \sqrt{(0.7071 + 0.1)^2 + (0.7071 - 0.6)^2} = 0.8142$$

$$L_4 = \sqrt{(0.7071 + 0.1)^2 + (0.7071 + 0.6)^2} = 1.5362$$

$$\Rightarrow |H(e^{j\omega})| = \frac{(0.7071)(1.9734)}{(0.8142)(1.5362)} = 1.1158$$

(c)

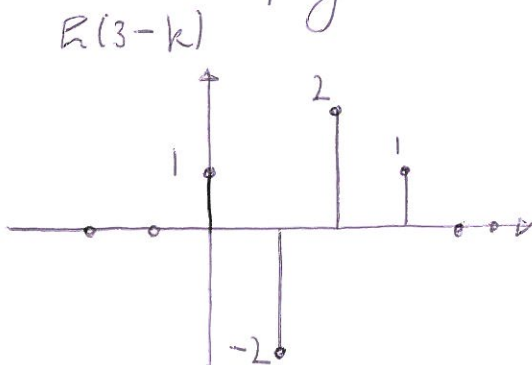


$$y(n) = \sum_{k=-\infty}^{\infty} x(k) R(n-k)$$

$$y(0) = 2(1) = 2$$

$$y(1) = 2(2) + 1(1) = 5$$

$$y(2) = 2(-2) + 1(2) - 1(1) = -3$$

Calculation of $y(3)$:

$$y(3) = \sum_{k=-\infty}^{\infty} x(k) R(3-k)$$

$$= \sum_{k=0}^3 x(k) R(3-k)$$

$$= 2(1) + 1(-2) - 1(2) + 3(1)$$

$$= 1$$

$$y(4) = 1(1) - 1(-2) + 3(2) = 9$$

$$y(5) = -1(1) + 3(-2) = -7$$

$$y(6) = 3(1) = 3$$

$$\therefore y(n) = \{2, 5, -3, 1, 9, -7, 3\} \text{ starting at } n=0$$

2. (a) $H(z) = \frac{0.1}{1-0.9z^{-1}} = \frac{1-a}{1-az^{-1}}$ where $a=0.9$

$$H(\theta) = \frac{1-a}{1-ae^{j\theta}} = \frac{1-a}{1-a\cos\theta + ja\sin\theta}$$

$$|H(\theta)|^2 = \frac{(1-a)^2}{(1-a\cos\theta)^2 + a^2\sin^2\theta}$$

$$= \frac{(1-a)^2}{1-2a\cos\theta + a^2}$$

At $\theta = \theta_c$, $|H(\theta)| = \frac{1}{\sqrt{2}} \Rightarrow |H(\theta)|^2 = \frac{1}{2}$

$$\Rightarrow \frac{(1-a)^2}{1-2a\cos\theta_c + a^2} = 0.5$$

$$\Rightarrow (1-a)^2 = 0.5 - a\cos\theta_c + 0.5a^2$$

$$\Rightarrow \cos\theta_c = \frac{1}{a} [0.5 - (1-a)^2 + 0.5a^2]$$

$$= \frac{1}{a} [0.5 - 1 + 2a - a^2 + 0.5a^2]$$

$$= \frac{1}{a} [-0.5 + 2a - 0.5a^2]$$

Substitute $a=0.9$ to obtain

$$\cos\theta_c = \frac{1}{0.9} [-0.5 + 2(0.9) - 0.5(0.9)^2]$$

$$= 0.9944$$

$$\Rightarrow \theta_c = 0.1059 \text{ rads}$$

$$= 2\pi \frac{f_c}{f_s}$$

$$\Rightarrow f_c = \frac{\theta_c f_s}{2\pi}$$

$$= 134.83 \text{ Hz}$$

(b) $h(n) = \{1, -1, 0, 3\}$ starting @ $n=0$

$$\Rightarrow H(z) = 1 - z^{-1} + 3z^{-3}$$

$x(n) = \{1, 3, -1, 2\}$ starting @ $n=2$

$$\Rightarrow X(z) = z^{-2} + 3z^{-3} - z^{-4} + 2z^{-5}$$

$$X(z)H(z):$$

$$\begin{array}{r} z^{-2} + 3z^{-3} - z^{-4} + 2z^{-5} \\ - z^{-3} - 3z^{-4} + z^{-5} - 2z^{-6} \\ \hline 3z^{-5} + 9z^{-6} - 3z^{-7} + 6z^{-8} \end{array}$$

$$Y(z) = z^{-2} + 2z^{-3} - 4z^{-4} + 6z^{-5} + 7z^{-6} - 3z^{-7} + 6z^{-8}$$

$$\therefore y(n) = \{1, 2, -4, 6, 7, -3, 6\}$$

starting @ $n=2$

$$(c) \quad H(z) = \frac{1 - 2z^{-1}}{1 + 0.3z^{-1} - 0.55z^{-2}}$$

$$|H(\theta)| = \frac{1 - 2e^{-j\theta}}{1 + 0.3e^{-j\theta} - 0.55e^{-j2\theta}}$$

$$= \frac{1 - 2\cos\theta + j2\sin\theta}{1 + 0.3\cos\theta - j0.3\sin\theta - 0.55\cos 2\theta + j0.55\sin 2\theta}$$

$$|H(\theta)| = \frac{\sqrt{(1 - 2\cos\theta)^2 + (2\sin\theta)^2}}{\sqrt{(1 + 0.3\cos\theta - 0.55\cos 2\theta)^2 + (-0.3\sin\theta + 0.55\sin 2\theta)^2}}$$

$$f = \frac{f_s}{4} \Rightarrow \theta = \frac{\pi}{2}$$

$$\begin{aligned} \Rightarrow |H(\theta)| &= \frac{\sqrt{(1 - 2(0))^2 + (2)^2}}{\sqrt{(1 + 0.3(0) - 0.55(-1))^2 + (-0.3(1) + 0.55(0))^2}} \\ &= \frac{\sqrt{1 + 4}}{\sqrt{2.4025 + 0.09}} \quad 1.107 + 0.191 = 1.298 \\ &= 1.4163 \end{aligned}$$

$$(d) \quad \theta_0 = 2\pi \frac{f_0}{f_s} = 2\pi \frac{2}{10} = 0.4\pi$$

$$r = 1 - \frac{\Delta f}{f_s} \pi = 1 - \frac{30}{10000} \pi = 0.9906$$

Coefficients:

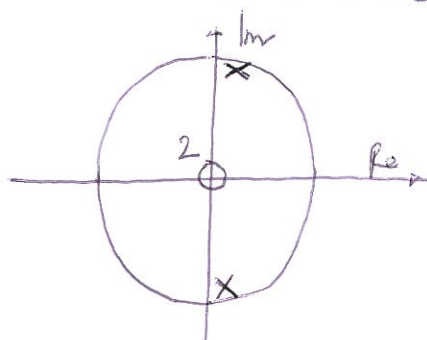
$$b_1 = -2r \cos \theta_0 = -2(0.9906) \cos(0.4\pi) = -0.6122$$

$$b_2 = r^2 = 0.9813$$

$$\text{DC gain of } 1 \Rightarrow \text{numerator} = 1 + b_1 + b_2 = 1.3691$$

$$\text{DC gain of } 0.8 \Rightarrow \text{numerator} = 0.8(1.3691) = 1.0953$$

$$\therefore H(z) = \frac{1.0953}{1 - 0.6122z^{-1} + 0.9813z^{-2}}$$



3. (a) Notch filter with two notches required (cascade)

(i) poles: $\theta_0 = 2\pi \frac{100}{1000} = 0.2\pi$

radius $r = 1 - \frac{10}{1000}\pi = 0.9686$

Denominator: $b_{11} = -2(0.9686)\cos(0.2\pi)$
 $= -1.5672$

$b_{12} = r^2$
 $= 0.9382$

zeros: $\theta_0 = 0.2\pi, r = 1$

Numerator: $a_{11} = -2\cos(0.2\pi)$
 $= -1.6180$

$a_{12} = 1$

(ii) poles: $\theta_0 = 2\pi \frac{150}{1000} = 0.3\pi$

radius $r = 0.9686$

Denominator: $b_{21} = -2(0.9686)\cos(0.3\pi)$
 $= -1.1387$

$b_{22} = 0.9382$

Numerator: $a_{21} = -2\cos(0.3\pi)$
 $= -1.1756$

$a_{22} = 1$

$\therefore H(z) = \frac{1 - 1.6180z^{-1} + z^{-2}}{1 - 1.5672z^{-1} + 0.9382z^{-2}} \cdot \frac{1 - 1.1756z^{-1} + z^{-2}}{1 - 1.1387z^{-1} + 0.9382z^{-2}}$

(b) FIR filter, $N = 512$

linear phase \Rightarrow coefficients are symmetric

\Rightarrow 256 "unique" coefficients

Each coefficient multiplies two digital filter memory values added together, in each sample period

\Rightarrow each y(n) calculation requires

$256 + 256 = 512 \text{ ADD}$
 256 MPY

10 seconds of data at $f_s = 16,000$ Hz requires:
 $(10 \times 16,000) \times 512 = 81,920,000$ ADD
 $(10 \times 16,000) \times 256 = 40,960,000$ MPY

FFT: each frame requires the following number of multiplies:

Windowing	N	512
512-point FFT	$2N \log_2(N)$	9,216
Multiply $H(e) \times X(e)$	$4N$	2,048
Inverse FFT	$2N \log_2(N)$	9,216
		<hr/>
		20,992 MPY/frame

10 seconds of data = 312.5 frames \Rightarrow 313

50% overlap \Rightarrow "effective" number of frames = 626

\Rightarrow Total MPY = $626 \times 20,992 = 13,140,992$

Saving = $1 - \frac{13,140,992}{40,960,000} = 67.92\%$

(c) Oscillator: $b_1 = 2 \cos \theta_0$

$b_2 = -1$

$\theta_0 = 2\pi \frac{2}{48} = \frac{\pi}{12} \Rightarrow b_1 = 2 \cos(\frac{\pi}{12}) = 1.9318$

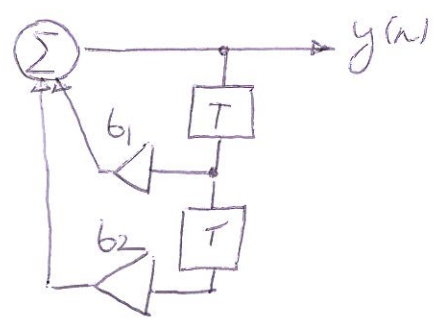
Initial conditions: we require a phase shift of $\frac{\pi}{4}$ which is $\frac{1}{8}$ of a period

Each cycle contains 24 samples

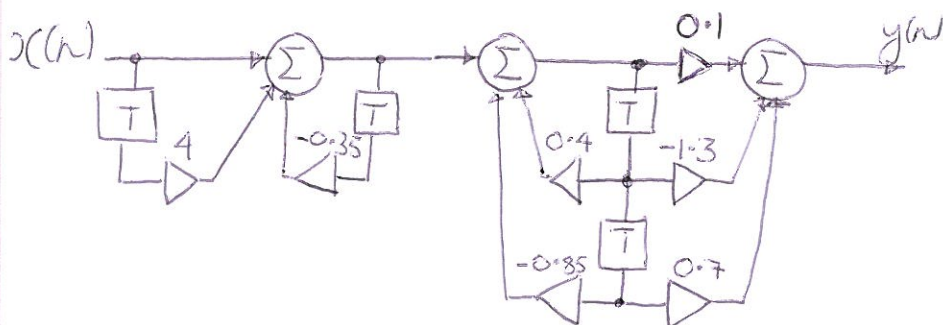
$\Rightarrow \frac{\pi}{4}$ corresponds to 3 samples

$\Rightarrow y(n-1) = \cos(2\theta_0) = 0.8660$

$y(n-2) = \cos(\theta_0) = 0.9659$



4. (a) Cascade implementation:



(b) $H(s) = \frac{W_c}{s + W_c}$

Desired digital cut-off frequency

$$\theta_c = 2\pi \frac{2}{20} = 0.2\pi$$

Pre-warp: $W_c = \frac{2}{T} \tan\left(\frac{\theta_c}{2}\right)$

$$= 40,000 \tan(0.1\pi)$$

$$= 12,997 \text{ rad/s}$$

Bilinear Transform:

$$s = \frac{2}{T} \cdot \frac{1 - \bar{z}^{-1}}{1 + \bar{z}^{-1}}$$

$$H(z) = \frac{W_c}{\frac{2}{T} \frac{1 - \bar{z}^{-1}}{1 + \bar{z}^{-1}} + W_c} = \frac{W_c(1 + \bar{z}^{-1})}{(2f_s + W_c) + (W_c - 2f_s)\bar{z}^{-1}}$$

Substitute for W_c and f_s to obtain:

$$\begin{aligned} & \frac{12997(1 + \bar{z}^{-1})}{(40000 + 12997) + (12997 - 40000)\bar{z}^{-1}} \\ &= \frac{12997(1 + \bar{z}^{-1})}{52997 - 27003\bar{z}^{-1}} \\ &= \frac{0.2452(1 + \bar{z}^{-1})}{1 - 0.5095\bar{z}^{-1}} \end{aligned}$$

If pre-warping is not carried out:

$$\theta_d = 2 \tan^{-1}\left(\frac{W_c T}{2}\right)$$

$$= 2 \tan^{-1}\left(\frac{40000\pi}{40000}\right) = 0.6058$$

$$\theta_d = 2\pi \frac{f_d}{f_s}$$

$$\Rightarrow f_d = \frac{\theta_d f_s}{2\pi} = 19.379 \text{ Hz} (< 2 \text{ kHz})$$

$$(c) \quad H(s) = \frac{2}{(s+3)(s+8)}$$

$$= \frac{A}{s+3} + \frac{B}{s+8}$$

$$A = H(s)(s+3) \Big|_{s=-3} = \frac{2}{s+8} \Big|_{s=-3} = \frac{2}{5}$$

$$B = H(s)(s+8) \Big|_{s=-8} = \frac{2}{s+3} \Big|_{s=-8} = -\frac{2}{5}$$

$$\Rightarrow H(s) = \frac{0.4}{s+3} - \frac{0.4}{s+8}$$

Impulse Invariant Transformation:

$$\frac{K}{s+a} \rightarrow \frac{K}{1-e^{aT}z^{-1}}$$

$$\Rightarrow H(z) = \frac{0.4}{1-e^{-3T}z^{-1}} - \frac{0.4}{1-e^{-8T}z^{-1}}$$

$$= \frac{0.4(1-e^{-8T}z^{-1}) - 0.4(1-e^{-3T}z^{-1})}{(1-e^{-3T}z^{-1})(1-e^{-8T}z^{-1})}$$

$$= \frac{0.4(e^{-3T} - e^{-8T})z^{-1}}{1 - (e^{-3T} + e^{-8T})z^{-1} + e^{-11T}z^{-2}}$$

Choice of sampling frequency:

$$\text{Highest pole frequency} = 8 \text{ rad/s} = 1.2732 \text{ Hz}$$

$$\Rightarrow f_s = 10 \times 1.2732 = 12.732 \text{ Hz}$$

$$\Rightarrow T = 0.07855$$

$$\begin{aligned} \Rightarrow H(z) &= \frac{0.4(0.7902 - 0.5337)z^{-1}}{1 - (0.7902 + 0.5337)z^{-1} + 0.4217z^{-2}} \\ &= \frac{0.1026z^{-1}}{1 - 1.3239z^{-1} + 0.4217z^{-2}} \end{aligned}$$