

# **ECE4634**

## **Introduction to Digital Communications**

### **Fall 2007**

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Lecture #13: Introduction to  
Digital Bandpass Modulation



Analog and Digital Communications



# Overview

- Modulation is the process for transferring information using EM waves
- Baseband systems – information signal modulates a baseband pulse stream
  - Up until this point we have considered this modulation
- Bandpass systems – information modulates a sinusoid
  - Necessary to allow frequency division multiplexing/multiple access
- We now consider a sinusoidal carrier signal
  - Analog – AM, FM (3614)
  - Digital – PSK, FSK, ASK (in the next 2-3 weeks)
- What to read – Section 7.1

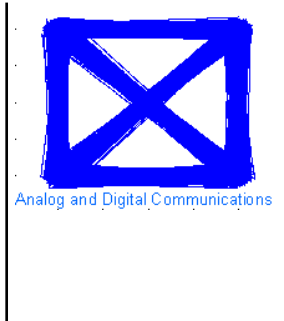
# Lecture Objective

- The objective of today's lecture is to introduce digital sinusoidal carrier modulation

# Modulation

- Modulation is defined as a process where an *information-bearing signal* causes changes to some characteristic of a carrier signal.
- The carrier signal can be either a pulse stream (a series of pulses in time) as in baseband communications or a sinusoid as in bandpass communications
- We first examined pulse modulation but will now consider sinusoidal carrier modulation

# Types of Sinusoidal Modulation



- In general, time-varying modulation of a sinusoid can be written as

$$s(t) = A(t) \cos(2\pi f_c t + \theta(t))$$

- $f_c$  – the nominal carrier frequency
  - $A(t)$  – time varying amplitude
  - $\theta(t)$  – time varying angle
- The information-bearing signal can be used to modulate (change) the amplitude or the angle of the sinusoid
  - The receiver can then recover the original information by examining the amplitude or angle of the received sinusoid

# Types of Sinusoidal Modulation



- Amplitude Modulation
  - The amplitude of the carrier is varied according to the message signal
  - Let  $F_{AM}(m(t))$  be the function or mapping of the message to the amplitude:

$$s(t) = F_{AM}(m(t)) \cos(2\pi f_c t)$$

- Angle Modulation
  - The angle of the carrier is varied according to the message of the signal
  - Frequency modulation – message directly affects the carrier frequency

$$s(t) = A_c \cos\left(2\pi \left[f_c + F_{FM}(m(t))\right] t\right)$$

- Phase modulation – message directly affects the carrier phase

$$s(t) = A_c \cos\left(2\pi f_c t + F_{PM}(m(t))\right)$$

# Analog vs. Digital Modulation



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- Whether the modulation scheme is analog or digital depends on the message signal
  - If the message takes on a continuum of values, we have analog modulation
  - If the message (not the waveform but the information) takes on a discrete number of values, we have digital modulation
- Types of analog modulation
  - Amplitude modulation (AM) – broadcast radio
  - Phase modulation (PM) - not widely used
  - Frequency modulation (FM) – broadcast radio, TV, original cell phones

# Amplitude Modulation

- Recall the signal for Large Carrier Amplitude Modulation or simply AM

$$s(t) = A_c [1 + k_a m(t)] \cos(2\pi f_c t)$$

$s(t)$  = transmit signal,  $m(t)$  is the analog message signal,  $k_a$  is a sensitivity constant and  $f_c$  is the carrier frequency (assumed to be much greater than the bandwidth of the signal)

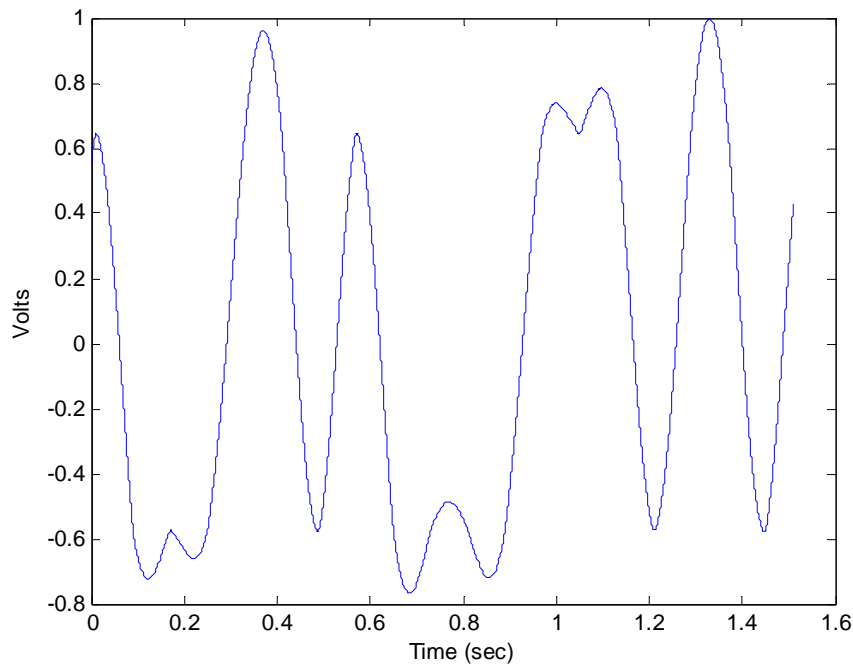
- For Double-Sideband Suppressed Carrier (DSBSC) AM

$$s(t) = A_c k_a m(t) \cos(2\pi f_c t)$$

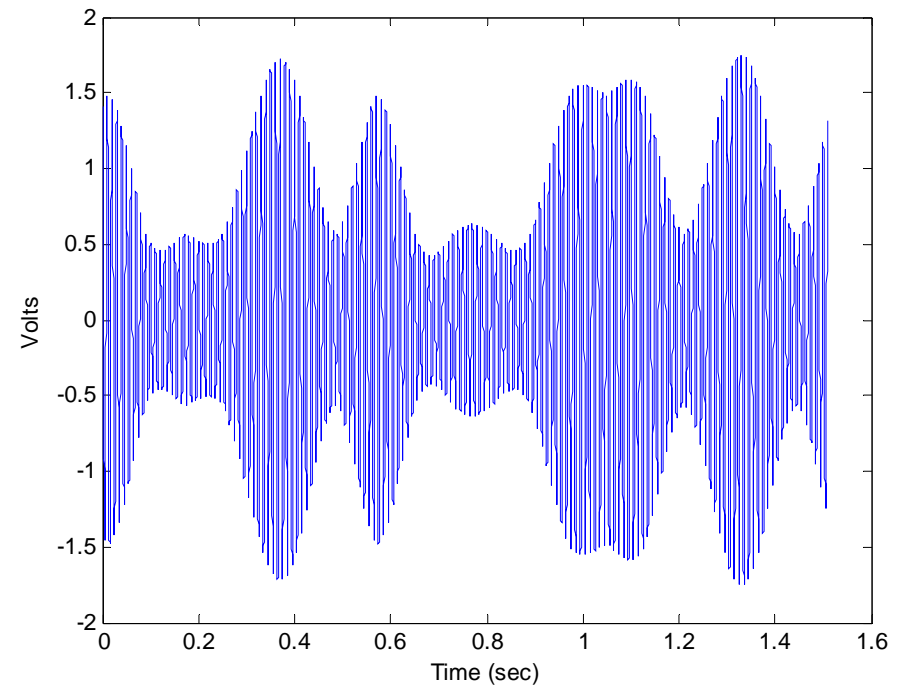


# Example – $k_a = 0.75$ , $\max\{m(t)\} = 1$

## Message signal

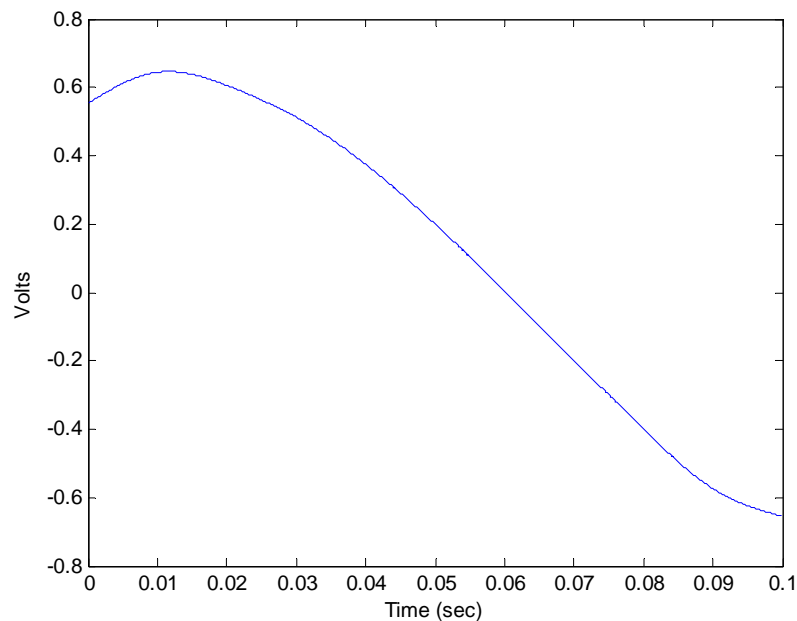


## Modulated carrier

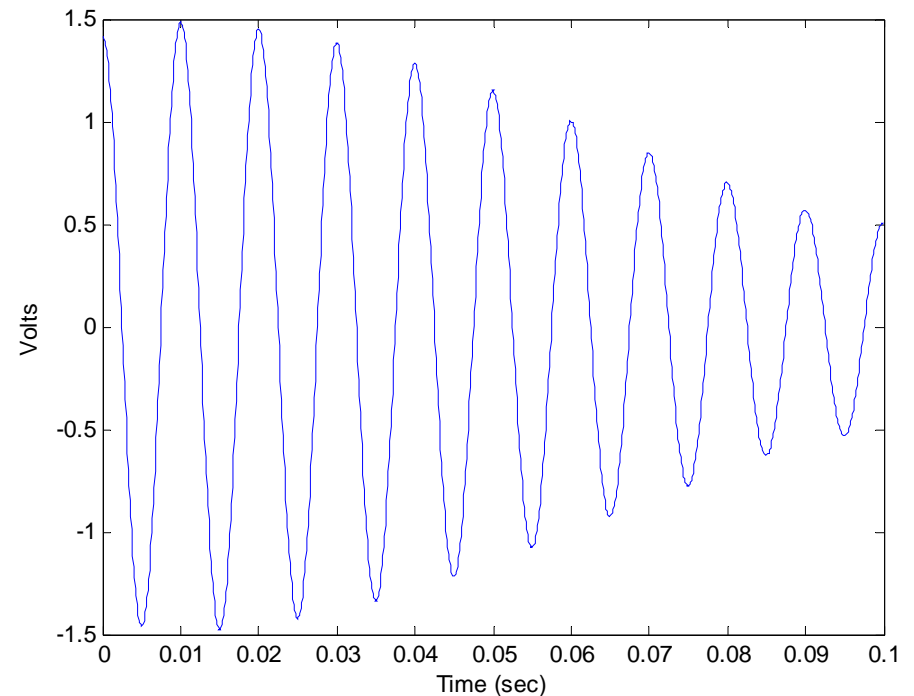


## Example – $k_a = 0.75$ , $\max\{m(t)\} = 1$

- Close-up  
Message signal

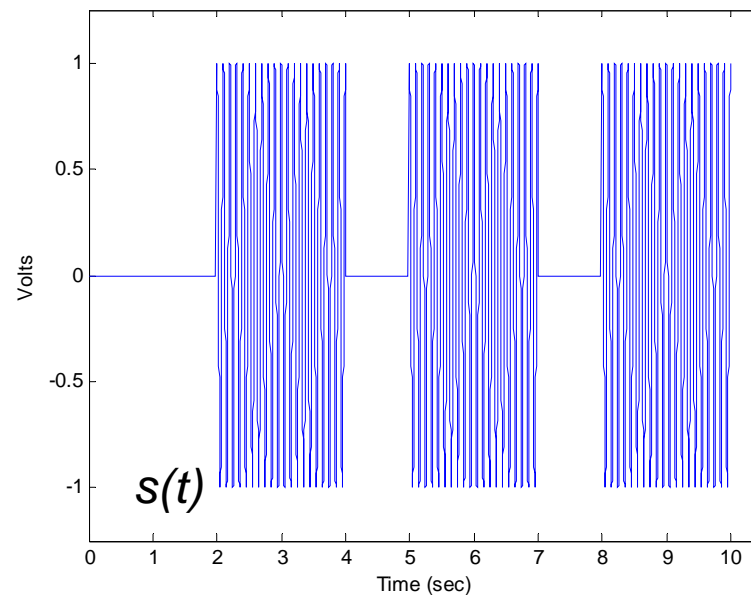
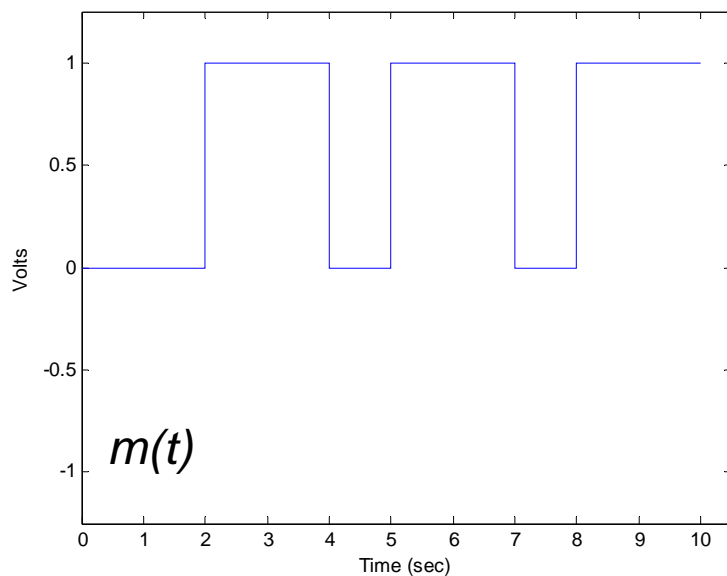


- Modulated carrier



# Digital Versions of AM

- If the message signal  $m(t)$  is digital, e.g., a unipolar NRZ line code, we have



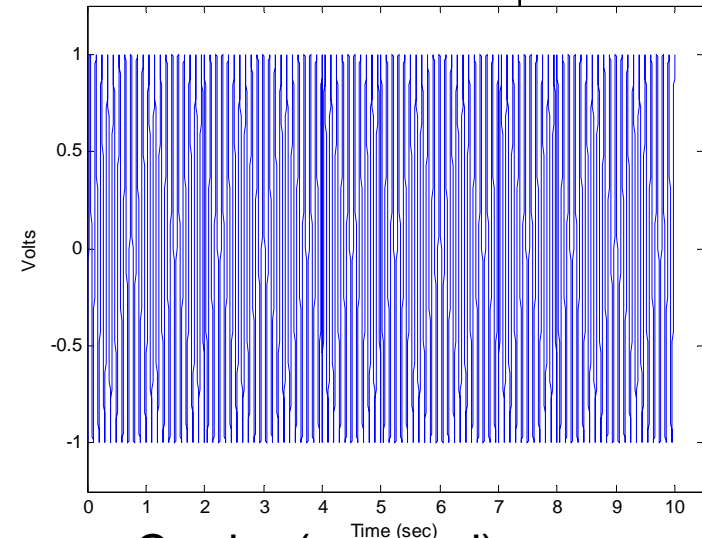
- This is simply Binary Amplitude Shift Keying (BASK)

# Example – DSB-SC with Digital Message

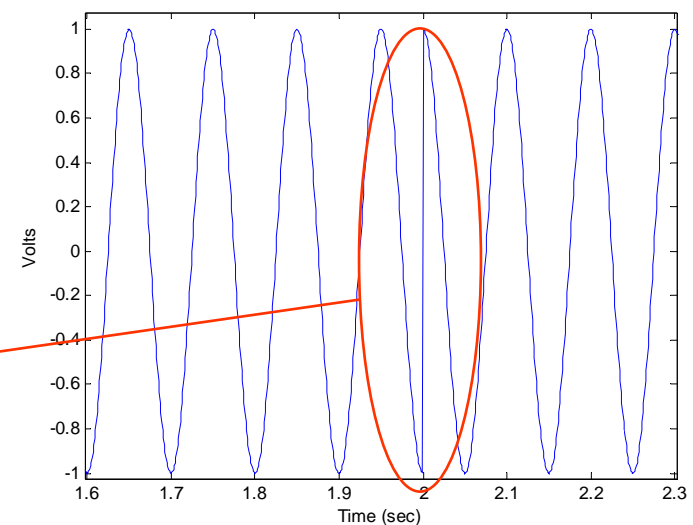


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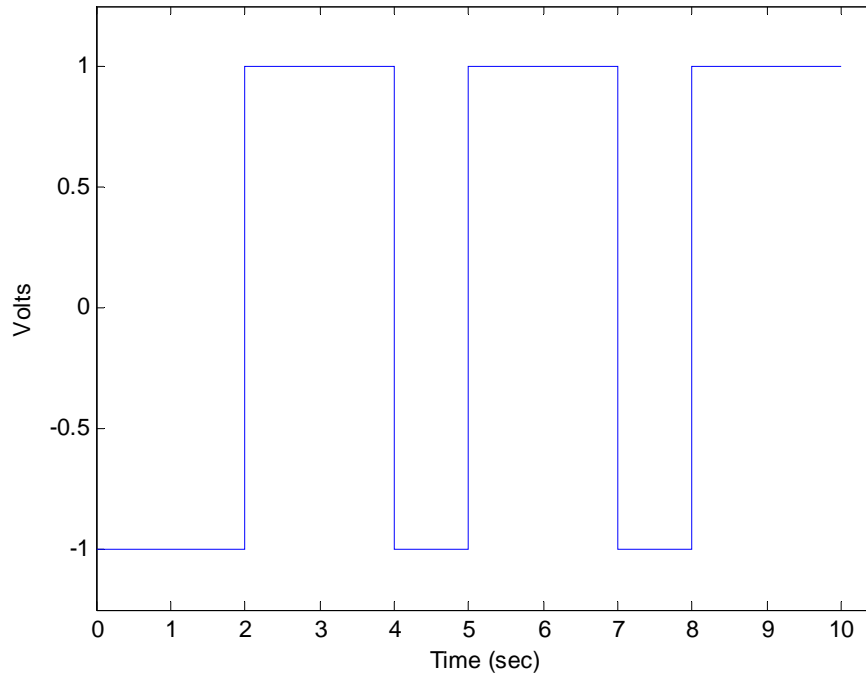
Carrier



Carrier (zoomed)



Message



Note phase change.  
This is simply BPSK

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# Digital Versions of AM

- If  $m(t)$  is a unipolar NRZ line code and we use DSB-SC AM we end up with the digital scheme known as Binary Amplitude Shift Keying (BASK)
  - Could also get BASK with Polar NRZ line code and regular large carrier AM with 50% modulation
- If  $m(t)$  is a polar NRZ line code and we use DSB-SC AM we end up with the digital scheme known as Binary Phase Shift Keying (BPSK)
  - Note that “negative” amplitudes are really phase changes

# Bandpass Assumption

- A bandpass signal is one whose frequency content is concentrated about some center frequency  $f_c$ .
- Additionally, for bandpass signals, if  $W$  is the bandwidth of the signal, we assume that

$$f_c \gg W$$

- This means that the message signal changes much more slowly than the carrier
- No overlap between positive and negative frequencies
- Complex baseband notation can be applied (to be discussed next class)



# Notation

- Symbol – signal transmitted over one symbol interval (equal to a bit interval in binary modulation)
  - Example: linear modulation (where symbol is line code times the carrier)

$$s(t) = A_c b(t) \cos(2\pi f_c t + \phi_c) \quad 0 \leq t \leq T_b$$

$b(t)$  represent  
the data bits

- Often we typically define the amplitude of the carrier  $A_c$ , in terms of the bit duration  $T_b$  such that it has unit energy during one bit/symbol time

$$A_c = \sqrt{\frac{2}{T_b}}$$

- For linear modulation this results in

$$\begin{aligned} s(t) &= \sqrt{\frac{2}{T_b}} b(t) \cos(2\pi f_c t + \phi_c) \quad 0 \leq t \leq T_b \\ &= b(t) c(t) \end{aligned}$$



# Notation (cont.)

- Proof:

$$\begin{aligned} E_c &= \int_0^{T_b} c^2(t) dt \\ &= \int_0^{T_b} \left( \sqrt{\frac{2}{T_b}} \cos(2\pi f_c t + \phi) \right)^2 dt \\ &= \frac{2}{T_b} \int_0^{T_b} \cos^2(2\pi f_c t + \phi) dt \\ &= \frac{2}{T_b} \frac{1}{2} \int_0^{T_b} dt + \frac{2}{T_b} \frac{1}{2} \int_0^{T_b} \cos(4\pi f_c t + \phi) dt \\ &\approx 1 \end{aligned}$$

$$\begin{aligned} f_c &\gg B \propto 1/T_b \\ \int_0^{T_b} \cos(4\pi f_c t + \phi) dt &\approx 0 \\ \text{"Bandpass Assumption"} \end{aligned}$$





# Energy per Bit

- The performance of various modulation schemes is compared based on the bit error rate performance for a received energy per bit,  $E_b$
- Since the carrier has unit energy,  $E_b$  is determined by the data waveform

$$\begin{aligned} E_b &= \int_0^{T_b} s^2(t) dt \\ &= \int_0^{T_b} \left( \sqrt{\frac{2}{T_b}} b(t) \cos(2\pi f_c t + \phi) \right)^2 dt \\ &= \frac{2}{T_b} \int_0^{T_b} b^2(t) \cos^2(2\pi f_c t + \phi) dt \\ &= \frac{2}{T_b} \frac{1}{2} \int_0^{T_b} b^2(t) dt + \frac{2}{T_b} \frac{1}{2} \int_0^{T_b} b^2(t) \cos(4\pi f_c t + \phi) dt \\ &\approx \frac{1}{T_b} \int_0^{T_b} b^2(t) dt \end{aligned}$$

Since  $f_c \gg W$ ,  $b(t)$  remains constant over one cycle of the carrier

$$\int_0^{T_b} b^2(t) \cos(4\pi f_c t + \phi) dt \approx 0$$

“Bandpass Assumption”<sup>17</sup>

# Types of Binary Digital Modulation



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$$\sqrt{\frac{2}{T_b}} \underbrace{A_d(t)}_{\text{data modulation}} \cos \left( 2\pi f_c t + \underbrace{\theta_d(t)}_{\text{data modulation}} + \phi_c \right)$$

- Binary Amplitude Shift Keying (BASK)

$$\sqrt{\frac{2}{T_b}} \underbrace{b(t)}_{\text{data}} \cos(2\pi f_c t + \phi_c)$$

- Binary Phase Shift Keying (BPSK)

$$\sqrt{\frac{2}{T_b}} \cos \left( 2\pi f_c t + \underbrace{b(t)}_{\text{data}} \pi + \phi_c \right)$$

- Binary Frequency Shift Keying (BFSK)

$$\sqrt{\frac{2}{T_b}} \cos \left( 2\pi \left( f_c + \Delta f \underbrace{b(t)}_{\text{data}} \right) t + \phi_c \right)$$

# Coherent Demodulation

- At the receiver we “demodulate” the signal (i.e., retrieve the data) by mixing the received signal to baseband for further processing
- This requires a local replica of the carrier wave
- If the phase of our local carrier wave is made to be equal to the incoming wave, our receiver is *coherent*.
  - Requires phase tracking circuitry
- If the phase is not the same, our receiver is *non-coherent*
  - Less complex

Received signal

$$\sqrt{\frac{2}{T_b}} \underbrace{A_d(t)}_{\text{data modulation}} \cos \left( 2\pi f_c t + \underbrace{\theta_d(t)}_{\text{data modulation}} + \phi_c \right)$$

Local carrier

$$\sqrt{\frac{2}{T_b}} \cos(2\pi f_c t + \phi_r)$$

For a coherent receiver

$$\phi_r = \phi_c$$

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# In-class drill



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# Summary

- Today we have introduced digital sinusoidal modulation
- Digital schemes are similar to old analog modulation schemes with the analog message replaced with a digital message signal
- We will assume bandpass signals where the bandwidth is much lower than the carrier frequency
- In the coming weeks we will study various modulation schemes, receiver structures and their bit error rate performance