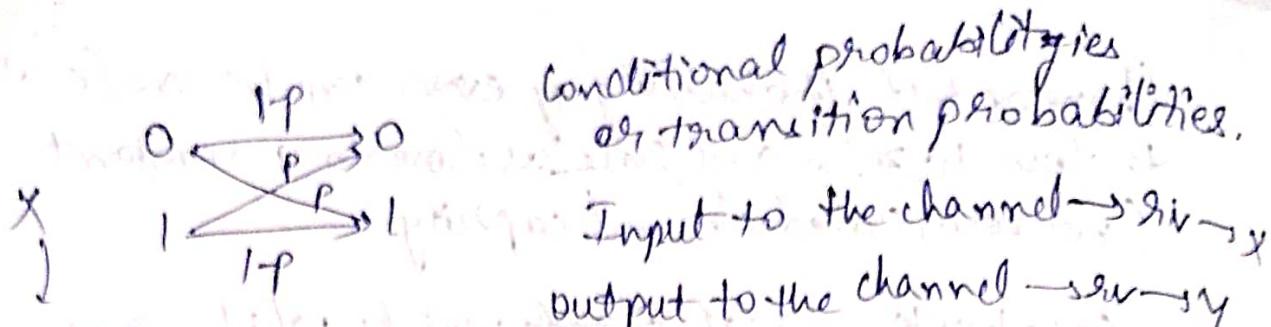


as n is increasing, such that $P_e \rightarrow 0$.

Channel Capacity.

Discrete Memoryless Channel

Examples \rightarrow Binary symmetric channel.



Input to the channel → binary
to the channel.

Input to the channel → binary
output to the channel → binary

$$P_{Y|X}(y|x)$$

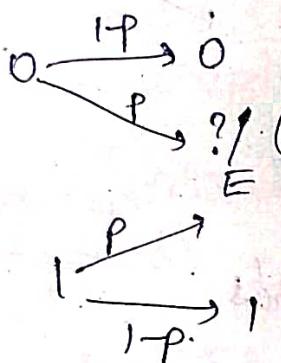
$P_{x,y}$ → joint
probability
is not known.

$$P_{Y|X}(0|0) = 1-p = P_{Y|X}(1|1)$$

$$P_{Y|X}(0|1) = p = P_{Y|X}(1|0)$$

If we want to find, $P_{Y|X}(y|x)$,
we need to know the value of $P_X(x)$.

Binary Erasure Channel



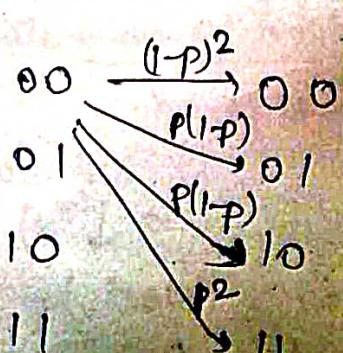
Output has been.

(output) $y \rightarrow$ erased
(we don't know what
the output is)

In this channel, there is no possibility of flipping.

E → something transmitted, but not received.
Signal

4-any symmetric channel



$$P_{Y|X}(00|00) = (1-p)^2$$

Discrete Memoryless Channel (DMC)

Input and output are both discrete random variables.

Memoryless channel

$$\hookrightarrow P_{Y^n/X^n}(y_1, y_2, \dots, y_n / x_1, x_2, \dots, x_n)$$

current output depends only on the current input.

$$= P_{Y_1/X_1 \dots X_n}(y_1/x_1, x_2, \dots, x_n) \times$$

for any general channel

$$P_{Y_2/Y_1, X_1 \dots X_n}(y_2/y_1, x_1, x_2, \dots, x_n) \times \dots$$

$$(\because P_{Y_1, Y_2}(y_1, y_2) = P_{Y_1}(y_1) \times P_{Y_2|Y_1}(y_2|y_1))$$

unconditional.

for conditional \rightarrow we have to condition it for every term.

If channel is memoryless,

$$P_{Y/X_1 \dots X_n}(y_1/x_1, x_2, \dots, x_n) = P_{Y_1/X_1}(y_1, x_1)$$

$$P_{Y_2/Y_1, X_1, X_2 \dots X_n}(y_2/y_1, x_1, x_2, \dots, x_n) = P_{Y_2/X_2}(y_2, x_2)$$

$$Y^n = (y_1, y_2, \dots, y_n)$$

$$X^n = (x_1, x_2, \dots, x_n)$$

$$P_{Y^n/X^n} = (y^n/x^n) = \prod_{i=1}^n P_{Y_i/X_i}(y_i/x_i)$$

Channel Coding Theorem

For every $R \leq C$, where C is the capacity of the channel, there exists a sequence of codes (n, R) such that

C_n has length n , codewords of $|C_n| = 2^{[nR]}$,

such that $P_e(C_n) \rightarrow 0$ as $n \rightarrow \infty$.

If $R = 0.5$, $C > 0.5$

$n = 2, 4, 5$

$|C_1| = 2, |C_2| = 4, |C_3| = 8, \dots$

Second part of channel coding theorem says that,
if $R \geq C$, no matter what sequence of codes we take,
 $P_e \rightarrow 0$.

capacity of DMC

$$C = \max_{P_X} I(X; Y)$$

the channel is specified by only conditional probability.

calculating capacity of BSC

$$\max_{P_X} I(X; Y)$$

$$= \max_{P_X} I(X; Y)$$

$$= \max_{P_X} H(Y) - H(Y|X)$$

$$H(Y|X=0) = P_X(0) H(Y|X=0)$$

$$+ P_X(1) H(Y|X=1)$$

$$H(Y|X=0) = -(1-p) \log(1-p) - p \log p = h_2(p)$$

(Binary entropy function)

$$H(Y|X=1) = h_2(p)$$

$$H(Y|X) = P_X(0) h_2(p) + P_X(1) h_2(p)$$

$$= \alpha h_2(p) + (1-\alpha) h_2(p)$$

$$H(Y/x) = h_2(p)$$

$$\max_{\alpha} (H(Y) - h_2(p))$$

not dependent on α

so we want
to maximize it

$$H(Y) \leq \log_2 M$$

where M is the number of values which Y is taking.

$$\therefore H(Y) \leq 1$$

$$H(Y) = 1 \text{ when } P_Y(0) = y_2, P_Y(1) = y_2.$$

$$\begin{aligned} P_Y(0) = y_2 &= P_X(0) P_{Y/X}(0|0) + P_X(1) P_{Y/X}(0|1) \\ &= \alpha(1-p) + (1-\alpha)p = y_2 \end{aligned}$$

$$\begin{aligned} P_Y(1) = y_2 &= P_X(0) P_{Y/X}(1|0) + P_X(1) P_{Y/X}(1|1) \\ &= \alpha p + (1-\alpha)(1-p) = y_2 \end{aligned}$$

$$\Rightarrow \alpha - \alpha p + p - \alpha p = y_2$$

$$\alpha + p - 2\alpha p = y_2 \quad \text{--- (1)}$$

$$\alpha p + 1 - \alpha - p + \alpha p = y_2 \quad \text{--- (2)}$$

$$\Rightarrow \alpha + p - 2\alpha p = y_2$$

$$1 - \alpha - p + 2\alpha p = y_2$$

$$\text{so } \max_{\alpha} (H(Y) - H(Y/x)) = \max_{\alpha} 1 - h_2(p)$$

4 - any symmetric channel.

$$H(Y/x) = h_2(1)$$

$$00 \xrightarrow{(1-p)^2} 00$$

$$01 \xrightarrow{p(1-p)} 01$$

$$10 \xrightarrow{p(1-p)} 10$$

$$11 \xrightarrow{p^2} 11$$

$$P(H(Y/x=00))$$

$$= H(Y/x=01) = H(Y/x=10)$$

$$= H(Y/x=11)$$

$$\leftarrow \max H(Y) - H(Y|X)$$

α

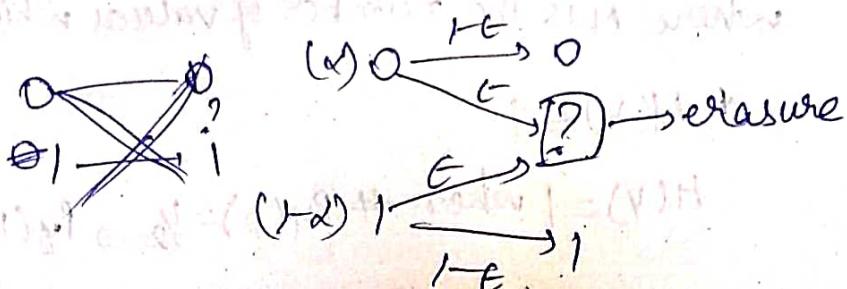
$$\Rightarrow H(Y) \leq 2 \Rightarrow \text{Ht}$$

$$= \max_{\alpha} \underline{\alpha}$$

Tut \rightarrow

$$1.) C = \max_{P_X} I(X; Y) = \max_{P_X} H(Y) - H(Y|X)$$

$$= 1 - E$$



$$C = \max I(X; Y)$$

&

$$= \max_{P_X} H(X) - H(X|Y) \neq \max H(X) -$$

&

$$\max_{P_X} H(X|Y)$$

$$= \max_{P_X} H(X) - E H(X|Y=?)$$

$$(1-e) H(X|Y \neq ?)$$

$$H(X|Y) = P(Y=?) H(X|Y=?)$$

$$Y=X \\ H(X|X) \geq 0$$

$$+ P(Y \neq ?) H(X|Y \neq ?)$$

$$P(Y=?) = P(Y=?|X=0) P(X=0) + P(Y=?|X=1) P(X=1)$$

$$= t_e(\alpha) + t_{(1-\alpha)} = e$$

$$H(X|Y=?) \Rightarrow P_{X \neq Y} = \frac{1}{2} \quad (\text{From Bayes's theorem})$$

$$= (P_X(0), P_X(1))$$

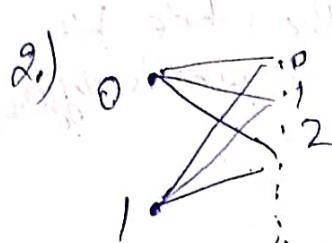
$$\Rightarrow = \max_{P_X} H(X) - e H(X) = 0$$

$$\therefore \max_{P_X} H(X)(1-e) = \boxed{1-e}$$

capacity \rightarrow max no. of bits that can be transferred reliably across the channel.

$N\epsilon$ bits are erased,

$N(1-\epsilon)$ bits are unerased.



$$|y| \leq \infty$$

K outputs

$$\Omega(y/x) = \Omega(g(y)/1),$$

where $g: y \rightarrow y$
is one-one

g is onto \Rightarrow bijective

$$C = \max_{P_x} I(x; y)$$

$$\Omega = \Omega_{Y/X},$$

$$= \max_{P_x} H(Y) - H(Y/x)$$

$$= \begin{cases} \Omega_{Y=0/x=0} & \dots \Omega_{Y=k/x=0} \\ \Omega_{Y=0/x=1} & \dots \Omega_{Y=k/x=1} \end{cases}$$

$$P(x=0)H(Y/x=0) + P(x=1)H(Y/x=1)$$

$$\Omega = \Omega \left[\begin{array}{c} 0 \ 1 \ 2 \ \dots \ k-1 \\ R_1 \\ R_2 \end{array} \right] \quad R_1 \text{ is a permutation}$$

of R_2

$$\cancel{P} \rightarrow P_{Y/x=0} = \left(\underbrace{\underline{\underline{P}}}_{K \text{ entries}} \right) \quad \begin{matrix} \rightarrow \text{bold font} \\ \text{Pf entry} \\ \text{used to represent} \\ \text{probability} \end{matrix}$$

$$P_{Y/x=1} = \left(\dots \right) \quad \begin{matrix} \rightarrow \text{permutation} \\ \text{measure} \end{matrix}$$

$$\text{So } H(Y/x=0) = H(Y/x=1)$$

$$C = \max_{P_x} H(Y) - H(Y/x)$$

$$H(Y/x=0)$$

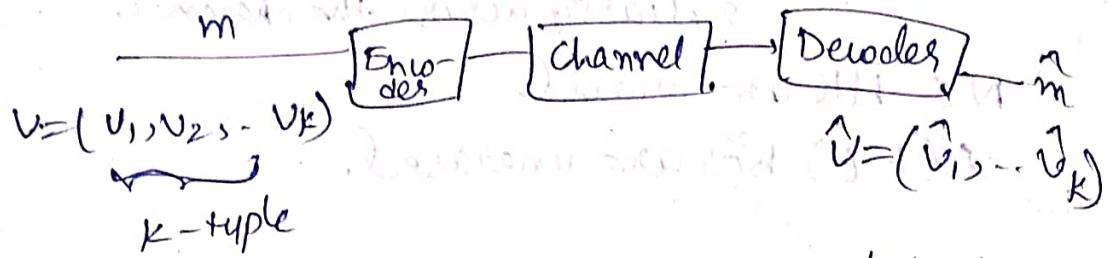
$$= H(\max_{P_x} H(Y)) - H(Y/x=0) = \log k - H(Y/x=0)$$

$$\cancel{P} \rightarrow P_x = \left(\frac{1}{2}, \frac{1}{2} \right)$$

gives rise to

$$P_Y = \left(\frac{1}{k}, \frac{1}{k}, \dots, \frac{1}{k} \right) - \left(\frac{1}{k}, \dots, \frac{1}{k} \right)$$

3.)

 K bits form a message

$$\{P_{\text{block}}\} \triangleq P(V \neq \hat{V})$$

(block error probability)

the whole bit

stream is different

Symbol error prob / Bit error prob

$$\{P_{\text{bit}} = P_{\text{symbol}}\} \triangleq \frac{1}{K} \sum_{j=1}^K P(\hat{v}_j \neq v_j)$$

$$\text{Show that: } P_{\text{block}} \geq P_{\text{symbol}} \geq \frac{P_{\text{block}}}{K}$$

 $\{V \neq \hat{V}\} \Rightarrow \{v_j \neq \hat{v}_j\}, \text{ for some } j.$

$$\begin{aligned} P_{\text{block}} &= P\left(\bigcup_j \{v_j \neq \hat{v}_j\}\right) \\ &\leq \sum_j P(\{v_j \neq \hat{v}_j\}) \quad \left| \begin{array}{l} \{v \neq \hat{v}\} = \\ \bigcup_j \{v_j \neq \hat{v}_j\} \end{array} \right. \\ &\quad \underbrace{\quad}_{K \text{ symbols.}} \end{aligned}$$

$$\Rightarrow P_{\text{symbol}} \geq \frac{P_{\text{block}}}{K}$$

$$\mathbb{I}\{\hat{v} \neq v\} \geq \mathbb{I}\{v_j \neq \hat{v}_j\}, \forall j$$

"Indicator" RV

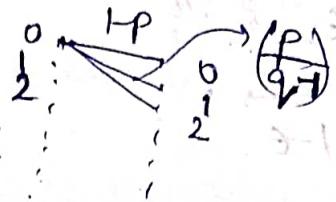
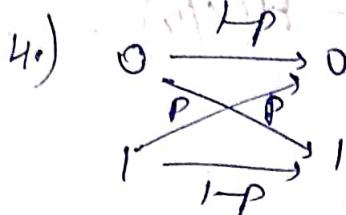
$$\begin{cases} 1 & \text{if } \hat{v} = v \\ 0 & \text{if } \hat{v} \neq v \end{cases}$$

$$\Rightarrow \mathbb{I}\{\hat{v} \neq v\} \geq \frac{1}{K} \sum_{j=1}^K \mathbb{I}\{v_j \neq \hat{v}_j\}$$

$$\mathbb{E}[\mathbb{I}(\text{event})] = P(\text{event})$$

$$\Rightarrow P(\hat{v} \neq v) \geq \frac{1}{k} \sum P(v_j \neq \hat{v}_j)$$

$$\Rightarrow P_{\text{block}} \geq P_{\text{symbol}}$$



BSC

q -ary symmetric channel

MAP: "maximum a posteriori"

decoder $\max_x P(x|y)$ (event which has happened before)

ML → "maximum likelihood"

$\max_x P(y|x)$ n inputs → n outputs
 likelihood $(P \leq 1 - \frac{1}{q})$

$$\max_x P(y|x) = \max_x \prod_{i=1}^n P(y_i|x_i)$$

$$(1-p)^{n-d_H(x,y)} \left(\frac{p}{q-1} \right)^{d_H(x,y)}$$

$$d_H(x,y) = \#\{(i, x_i \neq y_i)\}$$

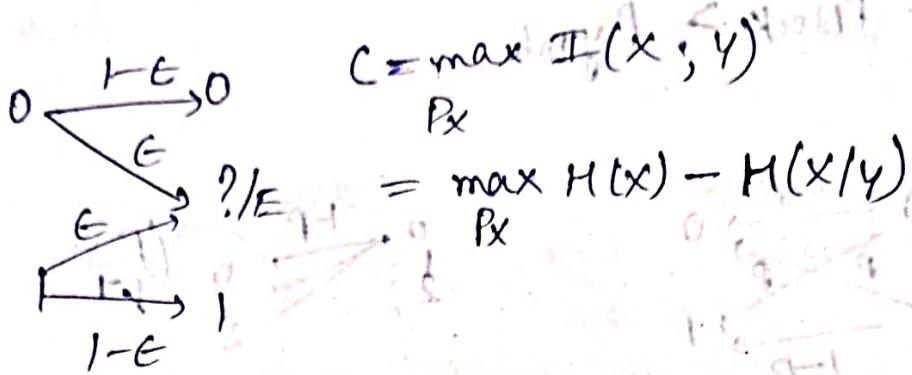
$$\max \text{Hamming distance} = (1-p)^n \left(\frac{p}{(1-p)(q-1)} \right)^{d_H(x,y)} \quad \text{minimize.}$$

$$\text{Given: } p \leq 1 - \frac{1}{q}$$

$$\Rightarrow \frac{p}{(1-p)(q-1)} \leq 1$$

ML decoder = min. decoder

Channel Capacity



$$H(X|Y=0) = 0 = H(X|Y=1)$$

$$P_X(0) = \alpha, P_X(1) = (1-\alpha).$$

$$H(X|Y=E) = \alpha H(X=0|Y=E) + (1-\alpha) H(X=1|Y=E)$$

$$= \alpha E + (1-\alpha) E = E$$

$$P_{X|Y}(x|y) = \frac{P_{XY}(x,y)}{P_Y(y)} = \frac{P_{Y|X}(y|x) \cdot P_X(x)}{P_Y(y)}$$

$$P_X(0|E) = P_X(1|E) = \frac{1}{2}$$

$$H(X|Y) = H(X|Y=E) \cdot P_Y(E) + H(X|Y \neq E) \cdot P_Y(Y \neq E)$$

$$\Rightarrow C = \max_{P_X} H(X) - H(X|Y)$$

$$= \max_{P_X} H(X) - E H(X) = \max_{P_X} (1-E) H(X)$$

$$= (1-E) \quad (\text{as } \max_{P_X} H(X) = 1)$$

Error Correcting Codes

Information theoretic Security (Cryptography)

(computationally bounded/unbounded adversary)

Security capacity, one-time pad.

\mathbb{Z}_p (prime fields)

codes \rightarrow Hamming code

Reed Solomon

Reed Muller

Binary Field (\mathbb{F}_2 , +, .)

\oplus	0	1
0	0	1
1	1	0

\cdot	0	1
0	0	0
1	0	1

(XOR)

Prime Field

$\mathbb{Z}_p = \{0, 1, 2, \dots, p-1\}$ all elements are prime numbers.

$$(a+b) \% \text{mod } p$$

$$(a \cdot b) \% \text{mod } p$$

There exists a multiplicative inverse for any element in \mathbb{Z}_p such that

$$a \cdot b = 1 \text{ mod } p$$

as p is prime, $\gcd(a, p) = 1$

$$\Rightarrow ax + py = 1$$

for any arbitrary integers - x, y

$$\Rightarrow ax \text{ mod } p = 1$$

$$(ax) \text{ mod } p = (a \text{ mod } p)(x \text{ mod } p) \text{ mod } p.$$

Vector Space = set of vectors.

$$(V, +, \mathbb{F}_2)$$

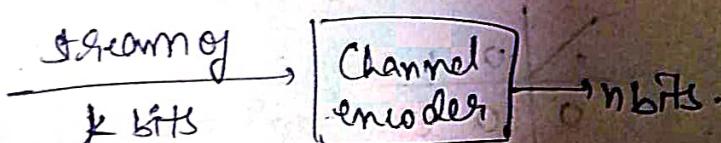
Two operations → vector addition
scalar multiplication

($\mathbb{F}_2^n, +, \mathbb{F}_2$) → Example of vector space over \mathbb{F}_2 .

$$u+v = \begin{pmatrix} u_1, u_2, \dots, u_n \\ v_1, v_2, \dots, v_n \end{pmatrix} \quad \begin{array}{l} \text{component wise} \\ \text{addition.} \end{array}$$

$$\alpha v = (\alpha v_1, \alpha v_2, \dots, \alpha v_n)$$

1.) Linear Block Codes



Eg. if we have 100 bits,

$k=4$, then

we divide 100 into 25 parts of 4 bits each and if $n=7$, then each of the 4 bits is mapped to 2 bits.

Channel encoder

= matrix multiplication

Linear Block Code over

\mathbb{F}_2 is a subspace of vector spa

$$\mathbb{F}_2^n : \text{code}$$

k -dimensional subspace of \mathbb{F}_2^n

Parameters of Linear Block Code

1.) length of the code, n .

2.) Dimension of the code k .

3.) Rate of the code = k/n .

4.) Minimum distance of the code d .

$$d_{min} = \min_{\substack{n \neq y \\ x, y \in C}} d_H(x, y)$$

$x, y \in C$

$d_H(x, y)$ = no. of positions in which x and y differ

e.g. $\begin{array}{cccccc} 1 & 1 & 0 & 1 & 1 & 0 & 0 & 1 & 0 \\ \hline 1 & 1 & 1 & 0 & 0 & 1 & 0 & 1 & 1 \end{array} \rightarrow 9$

Hamming distance = 5

Generator matrix of a code

(denoted by G_r , $(k \times n)$ matrix)

$G_r \rightarrow k \times n$.

It is a matrix in which k rows are k basic elements of code C .
(basis of subspace)

2) Repetition Code

$$C = \{00\ldots 0, 11\ldots 1\}$$

length of code = n .

$k=1$ (no. of basic elements in $C=1$)

$$\text{Rate} = \frac{1}{n}$$

Min distance = n

$$G_r = [1 \ 1 \ \dots \ 1]_{1 \times n}$$

Simple parity check code

all vectors will have even no. of ones

$$C = \{ (m_1, m_2, \dots, m_{n-1}, \sum_{i=1}^{n-1} m_i) \}$$

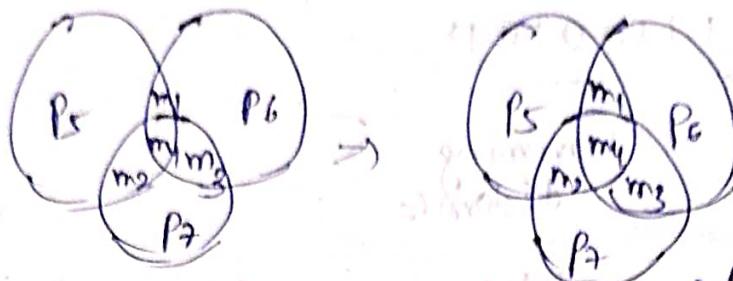
length = n ($m_1 \oplus m_2 \oplus \dots \oplus m_{n-1}$)

Dimension = $n-1$

Minimum distance = 2. (if one of m_i changes then m_n also changes
(parity)
so 2 bits are changed)

$$G_r = \left(\begin{array}{cccc|c} 1 & 0 & 0 & \dots & 0 & 1 \\ 0 & 1 & 0 & & & 1 \\ \vdots & & \vdots & & & \vdots \\ 0 & 0 & 0 & \dots & 0 & 1 \end{array} \right)$$

3.) Hamming Code (Code of 3 symbols)



$$C = \{ (m_1, m_2, m_3, m_4, m_1 + m_2 + m_3, m_2 + m_3 + m_4, m_1 + m_3 + m_4, m_1 + m_2 + m_3 + m_4) \}$$

$m_i \in \mathbb{F}_2$

Length = 7

Dimension = 4

$$G = \begin{pmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 1 \end{pmatrix} \rightarrow \begin{matrix} m_1 \\ m_2 \\ m_3 \\ m_4 \end{matrix} \rightarrow \begin{matrix} m_1 + m_2 + m_3 \\ m_2 + m_3 + m_4 \\ m_1 + m_3 + m_4 \\ m_1 + m_2 + m_3 + m_4 \end{matrix}$$

Minimum distance = 3. (atmost 3).

G is a matrix (which always contains a zero vector because it is a subspace, so maximum of minimum distance = 0.)

\Rightarrow Single error correcting code and also a double error detecting code (Hamming Code)

If a code has minimum distance d_{min} , then

No. of errors you can detect = $d_{min} - 1$

No. of errors you can correct = $\left\lfloor \frac{d_{min} - 1}{2} \right\rfloor$

d_{\min} is the minimum value such that of any $n-d_{\min}+1$ columns of G matrix have rank $= k$.

Theorem:

Linear Block Codes

- subspace of \mathbb{F}_2^n (or \mathbb{F}_q^n)
- length n , dimension k , minimum distance d_{\min}
- generator matrix G .
- 1.) Repetition
- 2.) Simple parity check.
- 3.) $(7,4)$ Hamming code.

→ Generator matrix of Hamming Code

$$G = \begin{pmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 1 \end{pmatrix} \quad \begin{aligned} v_1 + v_2 + v_3 + v_4 &= 0 \\ \text{so } n-d_{\min}+1 &\neq 4 \\ n-d_{\min}+1 &= 5 \end{aligned}$$

Applying Rank = $k=4$ (dimension)

Theorem: d_{\min} is the minimum the integers such that any $n-d_{\min}+1$ columns of generator matrix have rank k .

$$7 - d_{\min} + 1 \Rightarrow d_{\min} = 3$$

Two things to show:

- 1) Any $n-d_{\min}+1$ columns of G have rank $= k$.
- 2) There exists a set of $n-d_{\min}$ columns which do not have rank k .

Hamming weight → no. of non-zero elements.

Eg. 1110011

Hamming weight = 5.

Proof:

- 1.) if rank $< K$, then for another basis for the same code,
the row-reduced form of the matrix will
have all zeroes in the last row. (3 marks - Major Points)

Knows: $\begin{bmatrix} n-d_{\min}+1 \\ \vdots \\ 00 \dots 0 \end{bmatrix}$ no. of columns $= d_{\min}$ (non-zero elements)
which is a contradiction of minimum distance d_{\min} .
In d_{\min} columns can be $(d_{\min}+1)$ columns. $\leq d_{\min}+1$

so maximum rank possible $= d_{\min}+1$
which is a contradiction as $d_{\min}+1$ cannot be
the minimum distance, because by definition,
 d_{\min} is the minimum distance.

$$2.) \begin{bmatrix} n-d_{\min}+1 \\ \vdots \\ 00 \dots 0 \end{bmatrix} \quad \begin{pmatrix} 01100001 \\ 10100100 \\ 11001000 \\ 11110000 \end{pmatrix} = 10$$

(row reduced) of $\begin{pmatrix} 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \\ 1 & 0 & 1 & 0 & 0 & 1 & 0 & 0 \\ 1 & 1 & 0 & 0 & 1 & 0 & 0 & 0 \\ 1 & 1 & 1 & 1 & 0 & 0 & 0 & 0 \end{pmatrix}$
 $n-d_{\min}$ columns. d_{\min} (non-zero elements)

now if the d_{\min} is the minimum distance,
then Hamming weight $= d_{\min}$ has non-zero
elements. If these then the row-reduced matrix

obtained from the $n-d_{\min}$ columns gives $(0, \dots, 0)$ at
the last row, because if it does not form $\begin{pmatrix} 0 \\ \vdots \\ 0 \end{pmatrix}$, then
there is a non-zero element, which means that
 $d_{\min}+1 = \text{Hamming weight}$ which contradicts the
fact that $d_{\min} = \text{Hamming weight}$ and hence,
the rank of $n-d_{\min}$ columns is always less than K .

Extended Hamming Code (parity bit)
 $(m_1, m_2, m_3, m_4, p_5, p_6, p_7, \sum m_i + \sum p_i)$

length = 8, dimension = 4, min dist = 3.

Min distance = 4.

(if any one of m_i is changed, then 2 p_i will change
 so, in $\sum m_i + \sum p_i$, 3 bits will change, and hence
 this bit will also change and hence 4 bits change
 In total, so the min dist = 4.)

If weight($m_1, \dots, \sum m_i + p_i$) ≥ 4
 then parity bit will not change

If weight($m_1, \dots, \sum m_i + p_i$) = 3
 then parity bit also changes and hence weight becomes 4.

Read Solomon Codes

$\mathbb{F}_p \rightarrow p$ is prime

$\mathbb{F}_p^n \rightarrow$ is vector space

length = n , dimension = k , d_{min} = mindist.

$C_{k \times n} \rightarrow$ Generator matrix.

$(m_1, m_2, \dots, m_{k-1}), m_i \in \mathbb{F}_p$

Message polynomial.

$$f(x) = \sum_{i=0}^{k-1} m_i x^i$$

Let $\alpha_1, \dots, \alpha_n$ be n distinct elements in \mathbb{F}_p .

Codeword $\rightarrow C \rightarrow \mathbb{F}_p^n$

$$C = (f(\alpha_1), f(\alpha_2), \dots, f(\alpha_n))$$

(1) $C = m$ (or) $C^H = \text{matrix of } 2 \times K$ (or)

Evaluating the message polynomial of n distinct points in the field.

$$C = [c_1 \ c_2 \ \dots \ c_n] = [m_1 \ m_2 \ \dots \ m_k] C_S$$

$$\{c_1 \ c_2 \ \dots \ c_n\} = [m_1 \ m_2 \ \dots \ m_k] C_S$$

of rank K (or) m_i matrix of size $1 \times K$ (or) m_i

$$= [m_1 \ m_2 \ \dots \ m_k] \begin{bmatrix} \alpha_1 & \alpha_2 & \dots & \alpha_m \\ \alpha_1^2 & \alpha_2^2 & \dots & \alpha_m^2 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha_1^{k-1} & \alpha_2^{k-1} & \dots & \alpha_m^{k-1} \end{bmatrix}$$

Now the pfrequring matrix

of rank k

$$c_1 = f(\alpha_1), \ c_2 = f(\alpha_2), \ \dots \ c_n = f(\alpha_n)$$

rank d_{min} of f

$$d_{min} = n - k + 1$$

Theorem: d_{min} is the min value such that any $n - d_{min} + 1$ columns of C have rank K .

$$d_{min} = n - k + 1$$

$$\Rightarrow [n - d_{min} + 1 = k]$$

Any $(k-1)$ degree polynomial can have atmost $(k-1)$ roots

$$(f(\alpha_1) \ f(\alpha_2) \ \dots \ f(\alpha_n))$$

So $(k-1)$ zeroes are possible, so the weight can be $k-1$ (maximum).

So the maximum no. of $(k-1)$ zeroes

$$\text{Min distance} \geq n - k + 1$$

$(k-1)$ zeroes, $(n - k + 1)$ non-zeroes. \rightarrow maximum weight

$$\chi(n) = \prod_{i=0}^{k-1} (n - \alpha_i)$$

$\alpha_1, \alpha_2, \dots, \alpha_n$

so for $(k+1)$ columns, it becomes zero as $(n - \alpha_i) = 0$
as α_i is a root, the rest $(n-k+1)$ columns
produce non-zero answers.