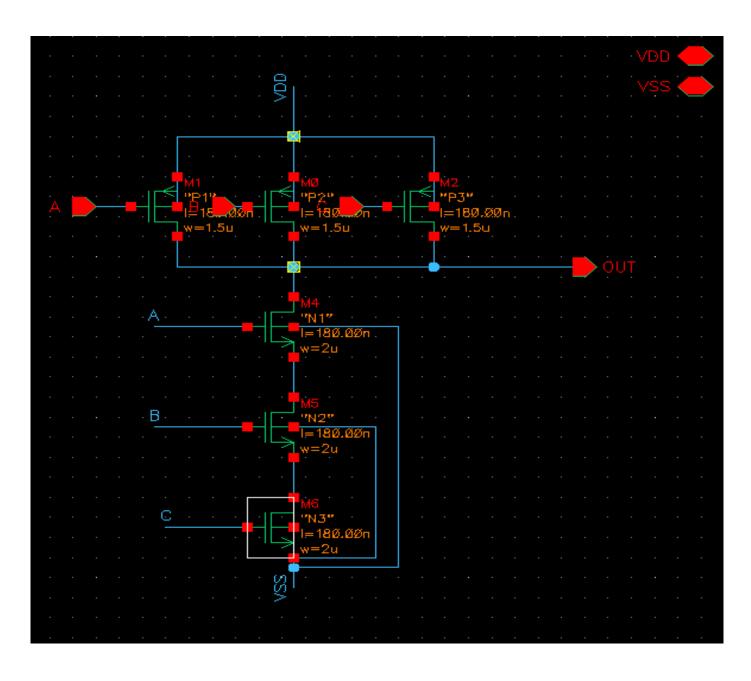
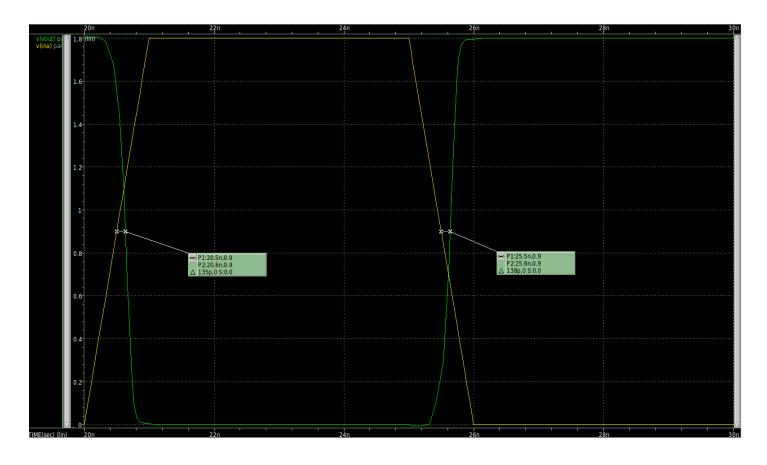
# 108061217 鍾永桓

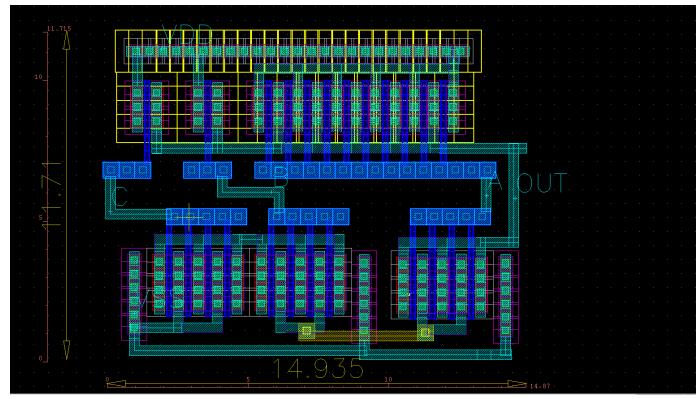
Part I (Combinational Circuit) (50%)

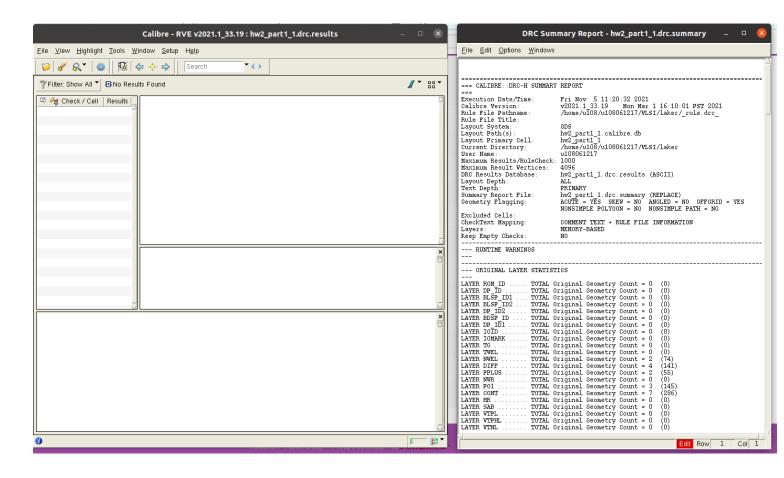
1. Please design a NAND3 shown in Fig. 1 with minimum length (0.18 $\mu$ m) of NMOS and PMOS. Under the condition TT25°C, the logic circuit has the same output rise delay (tpLH) and fall delay (tpHL) as the inverter in HW1 PartI with the following two input patterns. (4%)

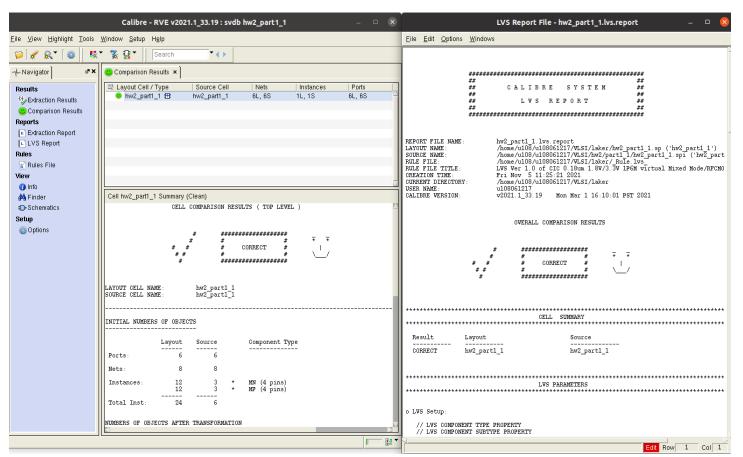
```
Use pattern (A,B,C)=(1,1,1)→(0,1,1) to find the rise delay (tpLH). Use pattern (A,B,C)=(0,1,1)→(1,1,1) to find the fall delay (tpHL). NMOS(A):2u/0.18u,m=4; PMOS(A):1.5u/0.18u,m=10; NMOS(B):2u/0.18u,m=4; PMOS(B):1.5u/0.18u,m=1; NMOS(C):2u/0.18u,m=4; PMOS(C):1.5u/0.18u,m=1; tpLH = 134.9ps(和 hw1 part1 (135.3ps) 的誤差:-0.2%) tpHL= 137.8ps(和 hw1 part1(130.4ps)的誤差: 5.7%)
```











2. Use the NAND3 designed above. Now, change one input from 0V to VDD at once and fix other inputs (see Table 1). Run simulation in different conditions and find output rise delay (tpLH) and fall delay (tpHL). Fill your results in Table 1. (0.5% for each gird, 6% in total)

Input A	Input B	Input C	Conclusion	tpLH(ps)	tpHL(ps)
pulse	1	1	SS 125°C	248.5	311.4
			SF 25°C	61.65	297.8
			<u>∓</u> ∓ 25°C	137.8	134.9
			FS 25°C	85.97	164.7
			FF -40°C	125.2	99.29
1	pulse	1	SS 125°C	1295	279.6
			SF 25°C	599.6	236.9
			<u>TT</u> 25°C	658.2	71.78
			FS 25°C	620.6	98.88
			FF -40°C	568.8	30.77
1	1	pulse	SS 125℃	1528	276.6
			SF 25°C	686.8	224.8
			<u>TT</u> 25°C	714.3	62.51
			FS 25°C	703.9	92.5
			FF -40°C	597.2	23.17

3. Comment what you have observed from Table 1 and give some discussion. (5%) (As long as your discussions are reasonable, you can get all points)

由 table1 的結果可以看出,無論是用A,B,C作為 pulse,在 SS 125℃的情況下所得到的 tpLH 和 tpHL 都是最大的,而 FF 所得到的 tpLH 和 tpHL 都是最小的,所以可知在 SS 下 delay 最大,FF 下 delay 最小,此外由這個結果可以看出在相同條件下,以 C 作為 pulse 時 tpLH 會最大,以 A 作為 pulse 則是 tpHL 最大,那是因為由於 A,B,C 作為 input pulse 到 out 的路徑不同,而且本題中 A,B,C 所接到的 gate 的 mos 尺寸差異也很大,所以 delay 也不相等。

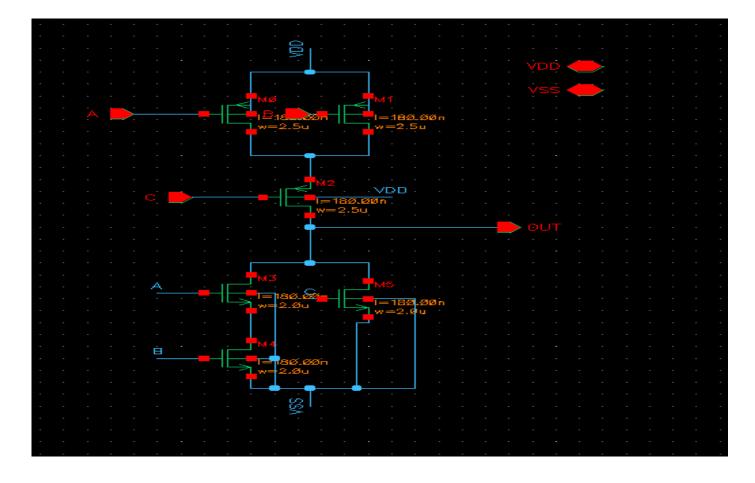
4. Please design a logic circuit shown in Fig. 2 with minimum length (0.18 $\mu$ m) of NMOS and PMOS. Under the condition TT25°C, the logic circuit has the same output rise delay (tpLH) and fall delay (tpHL) as the inverter in HW1 PartI with the following two input patterns. (4%)

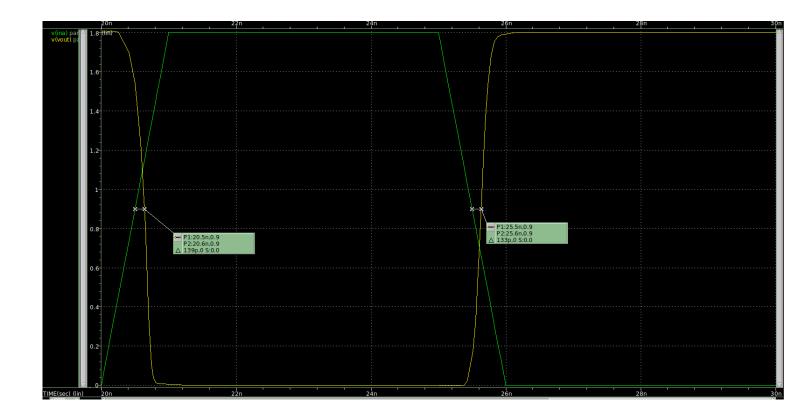
Note: Output loading Cload = 0.1pF (the same as HW1 Part I)

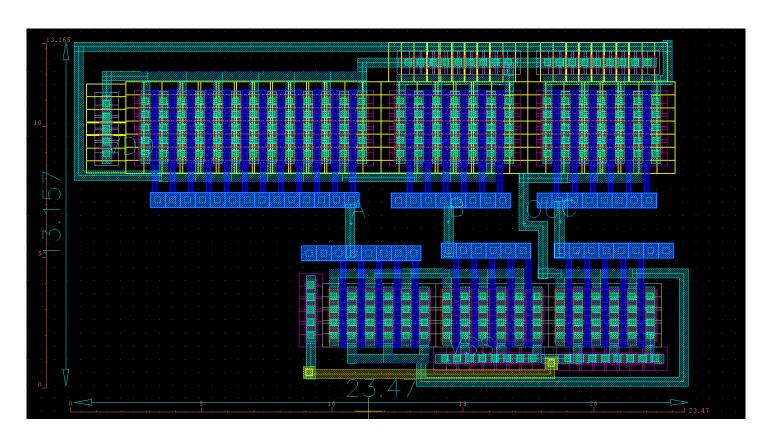
Use pattern (A, B, C) =  $(1, 1, 0) \rightarrow (0, 1, 0)$  to find the rise delay (tpLH).

Use pattern (A, B, C) =  $(0, 1, 0) \rightarrow (1, 1, 0)$  to find the fall delay (tpHL).

tpLH = 133.3ps(和 hw1 part1(135.3ps)的誤差:-1.5%) tpHL= 138.6ps(和 hw1 part1(130.4ps)的誤差: 6.3%) NMOS(A):2u/0.18u,m=5; PMOS(A):2.5u/0.18u,m=12; NMOS(B):2u/0.18u,m=5; PMOS(B):2.5u/0.18u,m=6; NMOS(C):2u/0.18u,m=5; PMOS(C):2.5u/0.18u,m=6;







5. Given the logic circuit in Fig. 2, please change one input from 0V to VDD at once and fix other inputs (see Table 2). Run simulation in different conditions and fined output rise delay (tpLH) and fall delay (tpHL). Please fill the simulation results in Table 2. (0.5% for each gird, 6% in total)

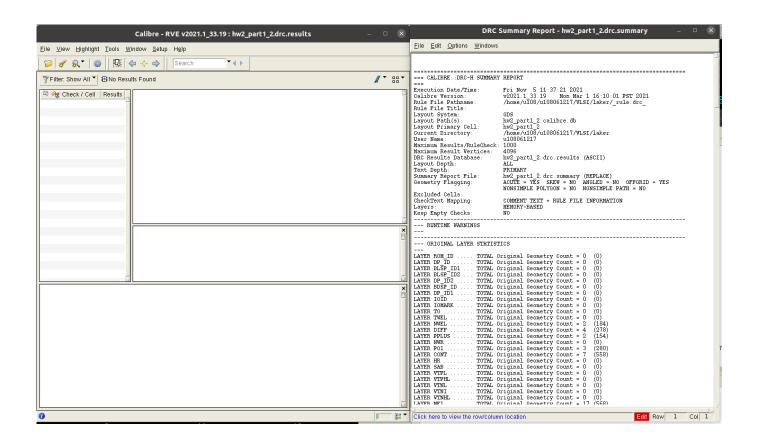
	Input A	Input B	Input C	Conclusion	tpLH(ps)	tpHL(ps)
Case1		1	0	SS 125°C	267.2	320.4
				SF 25°C	58.3	302.5
	pulse			TT 25°C	133.3	138.6
				FS 25°C	106	169.5
				FF -40°C	115.9	106.3
Case2		pulse	0	SS 125°C	454.1	279.3
				SF 25°C	171.5	256.8
	1			TT 25°C	239.2	104.2
				FS 25°C	212.5	130.6
				FF -40°C	210	70.79
Case3	0	1	pulse	SS 125°C	334.8	167.2
				SF 25°C	131.6	175.7
				TT 25°C	205.2	47.69
				FS 25°C	181.1	72.03
				FF -40°C	188.4	23.71

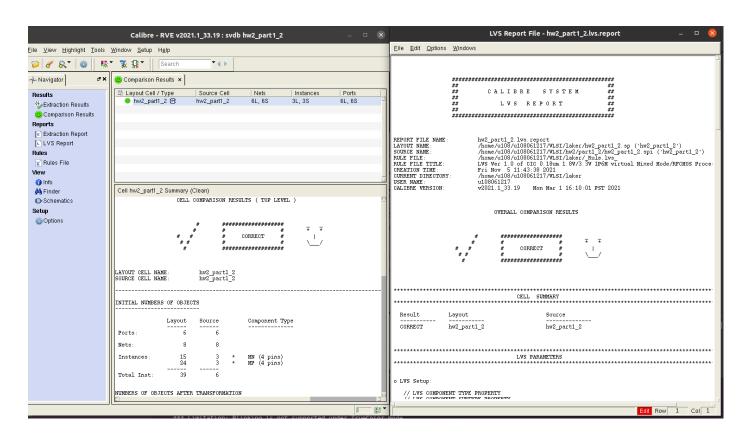
6. Comment what you have observed from Table 2 and give some discussions. (5%) (As long as your discussions are reasonable, you can get all points)

表中很明顯可以看出溫度和 corner 對於 delay 的影響,FF 時 delay 在大多情況下最小而 SS 則在大多情況下 delay 最大,此外不同 case 會影響導通的路徑,它們使用不同的 input pulse,而且開啟不同的路徑,如果電容充放電時路徑不導通可能會使其 delay 很大,由表中的結果可以看出 case2 的 tpLH 在所有情况下都是最大的,可能是由於其A為 1 使以 A 作為 gate input 的 pmos 不導通,所以在 high to low 時阻值很大,所以 case2 時 tpLH 會較大。

7. Complete the layout of Fig. 2. Show the figure of your layout and DRC/LVS report. (10%) and measure the area. (5%) Note: Area of the layout is the area of minimum single rectangle that can cover all circuits

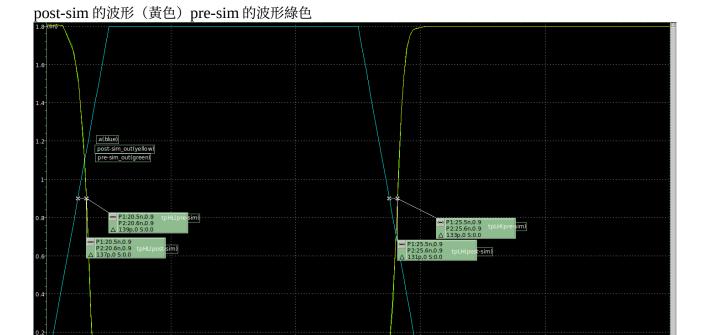
Area: 308.39639um^2





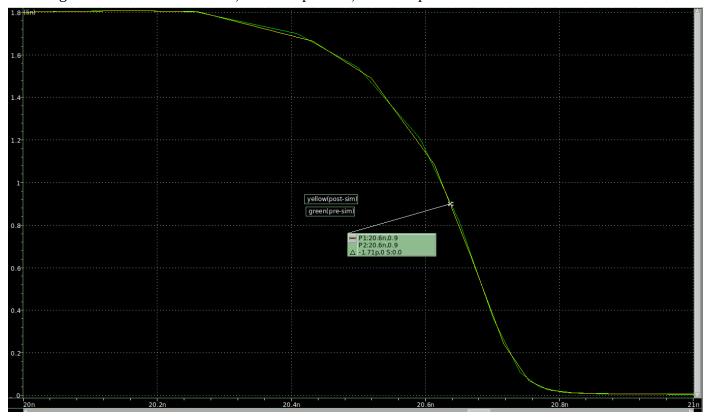
8. Run post-sim (with R-C-CC extraction), compare the result and waveform with pre-sim. (5%)

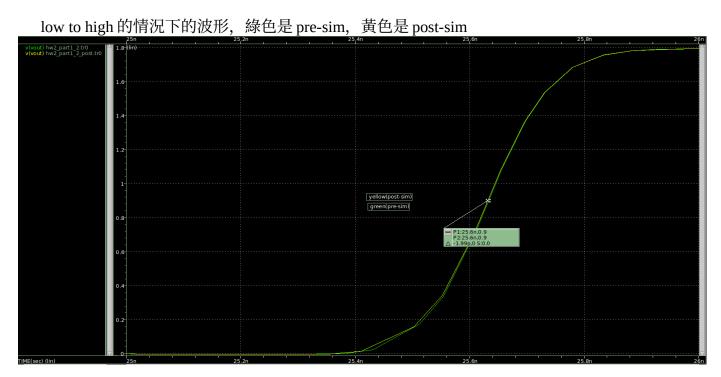
	tpLH(ps)	tpHL(ps)	
pre-sim	133.3	138.6	
post-sim	131.4	136.9	



post-sim 和 pre-sim 做比較(因不太明顯,所以放大分成兩張):

high to low 的情況下的波形,綠色是 pre-sim,黃色是 post-sim





由比較結果看出不論 tpLH 還是 tpHL 在 post-sim 之後延遲似乎沒有非常明顯的變化,甚至略下降,波形也幾乎相同,。會變化如此不明顯可能是因為寄生電容的大小相較於 C load 而言很小,在這樣的條件下就算考慮寄生電容和電阻,也無法看出太明顯的變化。

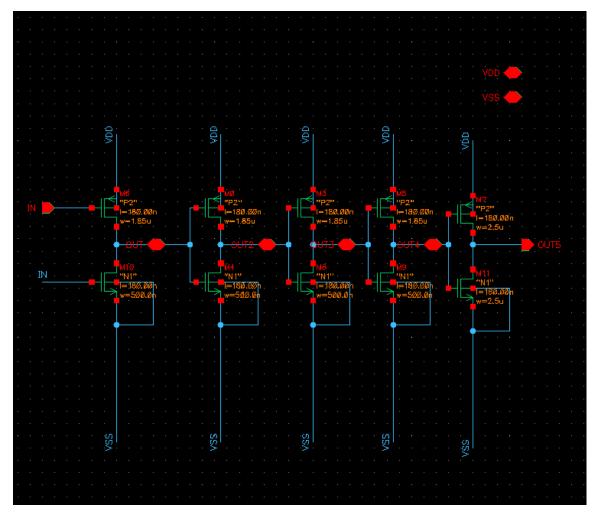
Part II (Inverter chain) (50%)

1. Complete the layout of your inverter chain from HW1 Part II. Show the figure of your layout and DRC/LVS report. (15%) and measure the area. (2.5%)

(Output loading Cload = 10pF)

Bonus: FoM = delay \* area (delay will get from post-sim)

FOM:2.8\*10^5(um^2\*ps)



本題中的 inverter chain 有做如下調整:

hw1:

inverter1:

pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1

inverter2:

pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1

inverter3:

 $pmos(1.85u/0.18u)\;, m=3; nmos(0.5u/0.18u)\;, m=3$ 

inverter4:

 $pmos(1.85u/0.18u)\;, m=5; nmos(0.5u/0.18u)\;, m=5$ 

inverter5:

```
pmos(11.4u/0.18u),m=7;nmos(4.2u/0.18u),m=7
```

hw2:

inverter1:

pmos(1.85u/0.18u),m=1;nmos(0.5u/0.18u),m=1

inverter2:

pmos(1.85u/0.18u),m=1;nmos(0.5u/0.18u),m=1

inverter3:

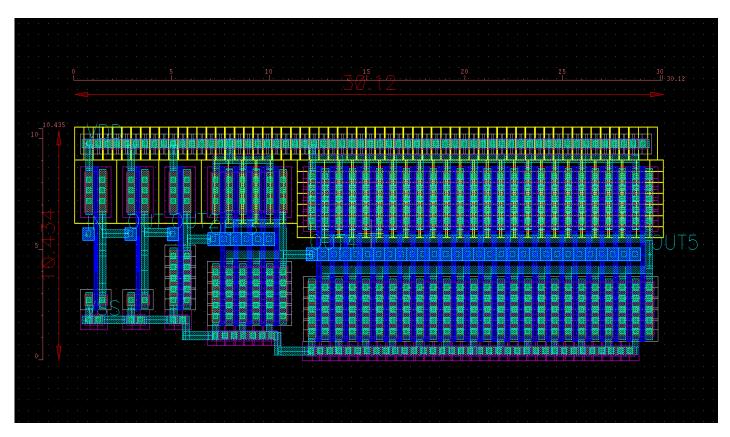
pmos(1.85u/0.18u),m=1;nmos(2.5u/0.18u),m=1

inverter4:

pmos(1.85u/0.18u),m=5;nmos(2.5u/0.18u),m=5

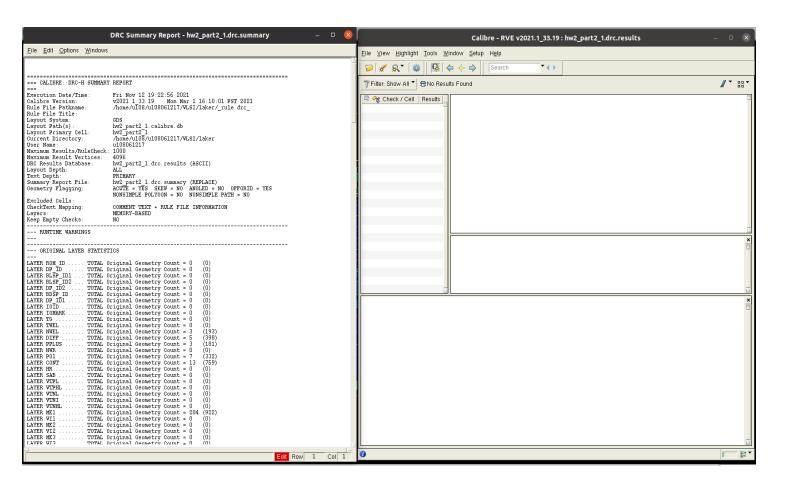
inverter5:

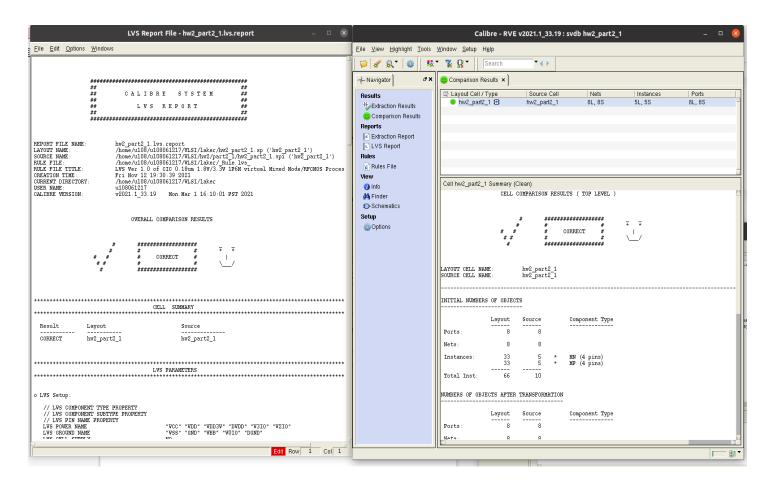
pmos(2.5u/0.18u), m=25; nmos(2.5u/0.18u), m=25



## area:

area = 10.434\*30.12 = 314.27208(um^2)





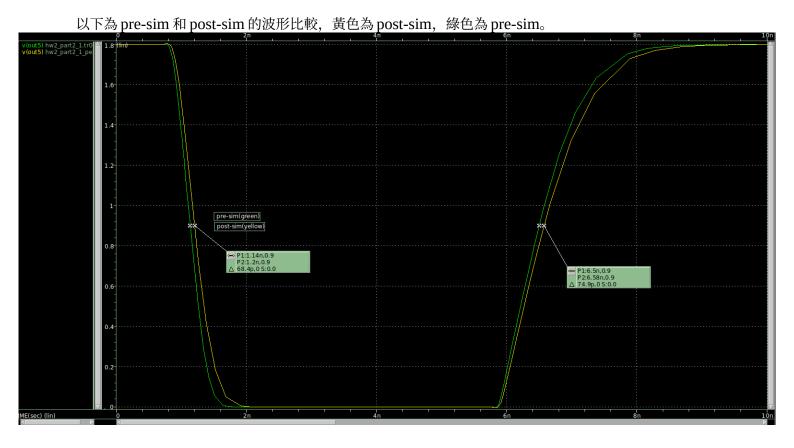
2. Run post-sim (with R-C-CC extraction), compare the result and waveform with pre-sim from HW1 Part II. (5%)

tpHL: from 635.5ps(pre-sim) to 703.9ps(post-sim) tpLH: from 1003ps(pre-sim) to 1078ps(post-sim)

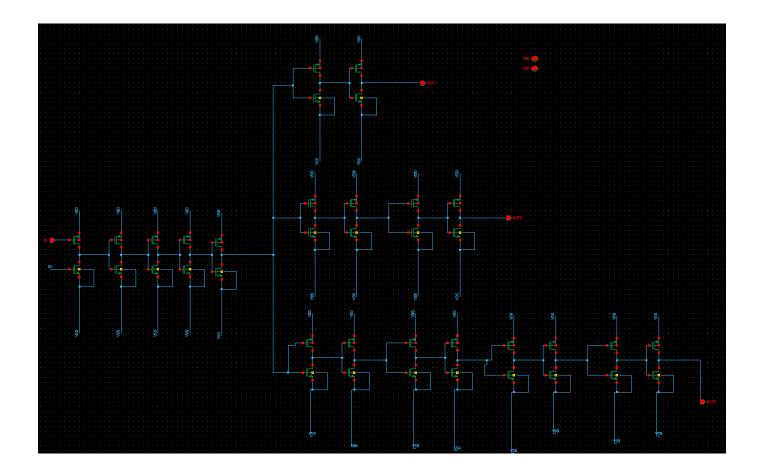
propagation delay: from 819.2ps(pre-sim) to 890.9ps(post-sim)

delay\_time 變化(post-sim - pre-sim):71.7ps

可以看出 post-sim 相較於 pre-sim delay 明顯較大,原因應是因為 pre-sim 所模擬的情況並未考量到寄生電阻和電容的影響,而在 inverter chain 當中,有許的個 inverter 串接在一起,所以會有較大的寄生電阻與電容,post-sim 考量到了這些寄生電阻語電容,這些寄生的電阻電容使 delay 更大。



3. Use different stage as branch to measure the propagation delay. The first half is your inverter chain in HW1 Part II and the last half is three kinds of inverter chain with different stage. (5%) (Output loading Cload = 10pF) Note: The size of first inverter in last half (where the red arrow points) of long inverter chain is fixed: (W/L)n = (0.5um/0.18um), (W/L)p = (1.85um/0.18um).



#### last half:

```
inverter chain1:
```

inverter1

pmos(1.85u/0.18u),m=5;nmos(0.5u/0.18u),m=5

inverter2:

pmos(2.5u/0.18u), m=25; nmos(2.5u/0.18u), m=25

### inverter chain2:

inverter1:

pmos(1.85u/0.18u),m=1;nmos(0.5u/0.18u),m=1

inverter2:

pmos(1.85u/0.18u),m=1;nmos(0.5u/0.18u),m=1

inverter3:

pmos(1.85u/0.18u),m=5;nmos(2.5u/0.18u),m=5

inverter4:

pmos(2.5u/0.18u),m=25;nmos(2.5u/0.18u),m=25

#### inverter chain3:

inverter1:

pmos(1.85u/0.18u),m=1;nmos(0.5u/0.18u),m=1

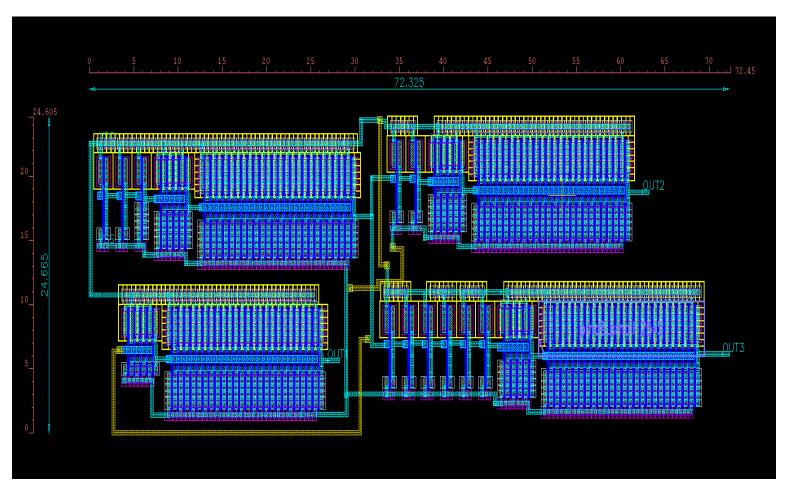
inverter2:

 $pmos(1.85u/0.18u)\;, m=1; nmos(0.5u/0.18u)\;, m=1$ 

inverter3:

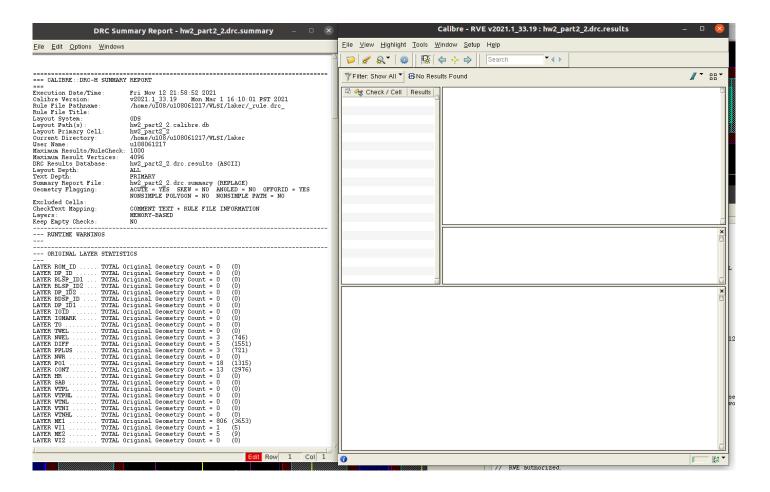
```
pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1 inverter4:
pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1 inverter5:
pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1 inverter6:
pmos(1.85u/0.18u) ,m=1;nmos(0.5u/0.18u) ,m=1 inverter7:
pmos(1.85u/0.18u) ,m=5;nmos(2.5u/0.18u) ,m=5 inverter8:
pmos(2.5u/0.18u) ,m=25;nmos(2.5u/0.18u) ,m=25
```

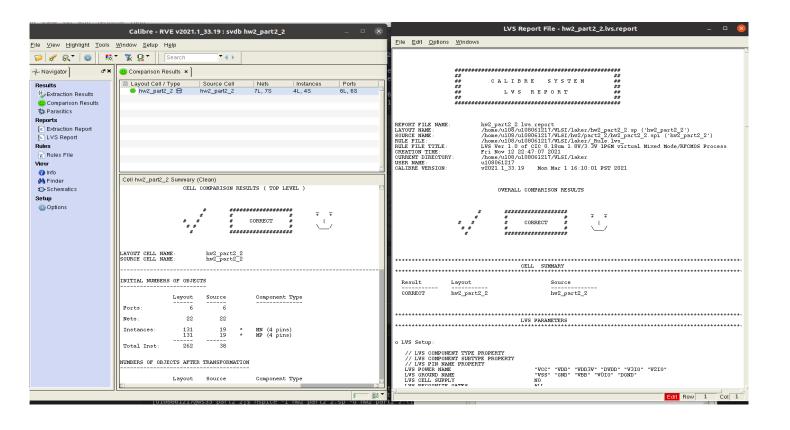
4. Complete the layout of Fig. 3. Show the figure of your layout and DRC/LVS report. (15%) and measure the area. (2.5%)



### area:

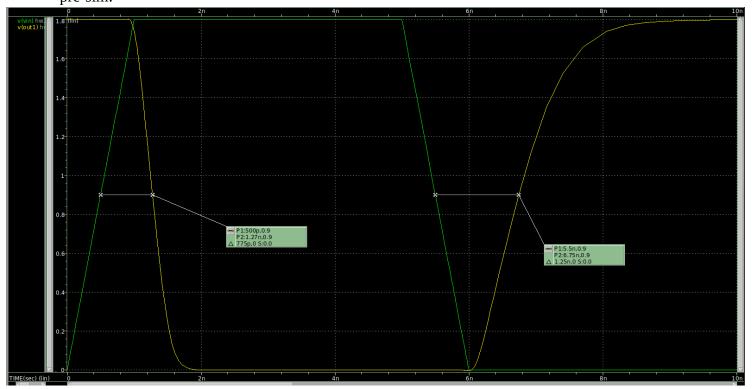
area = 24.665\*72.325 = 1783.896125(um^2)



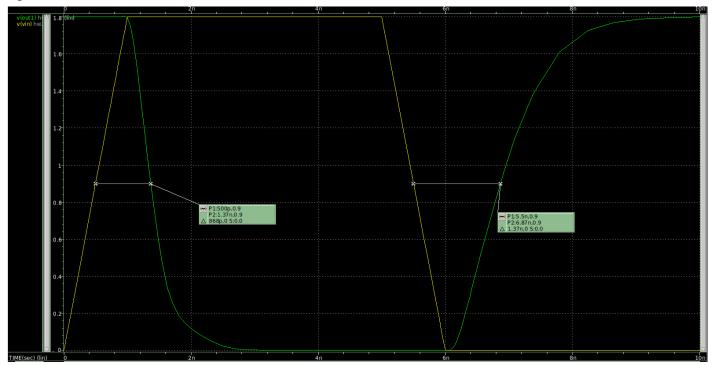


5. Run post-sim (with R-C-CC extraction), compare the delay of three delay chain. (5%) 比較了三個 stage 的結果,可以看出不管後面接哪種,最後的 delay 都會比原來大,而尤其是後半段接上 8 個 inverter 的情況 delay 最大,而且其 post-sim 的 delay 比起 pre-sim 變化也最多,可能是由於後半段接 8 個時 inverter chain 太長了,太長的 inverter chain 會導致 delay 變大,而且太長的 inverter chain 受到寄生電阻和電容的影響特別的明顯,所以才會在 post-sim 後 delay 變化特別大。

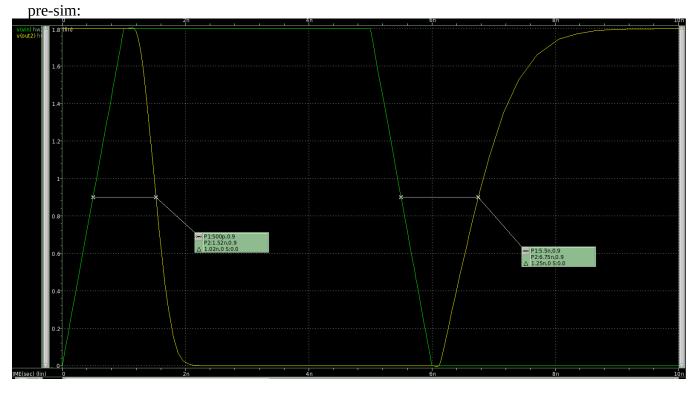
inverter chain with 2 stage(last half) pre-sim:1011ps; post-sim:1118ps;delay\_time 變化(pre-sim-post-sim): 107ps(tpHL:747.7ps to 868ps;tpLH:1247ps to 1368ps) pre-sim:



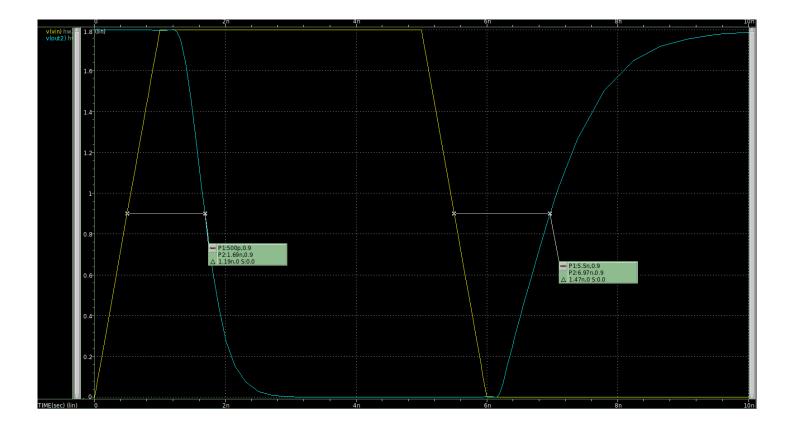
## post-sim:



inverter chain with 4 stage(last half) pre-sim:1138ps ; post-sim:1330ps;delay\_time 變化(pre-sim - post-sim): 192ps(tpHL:1021ps to 1191ps;tpLH:1254ps to 1469ps)

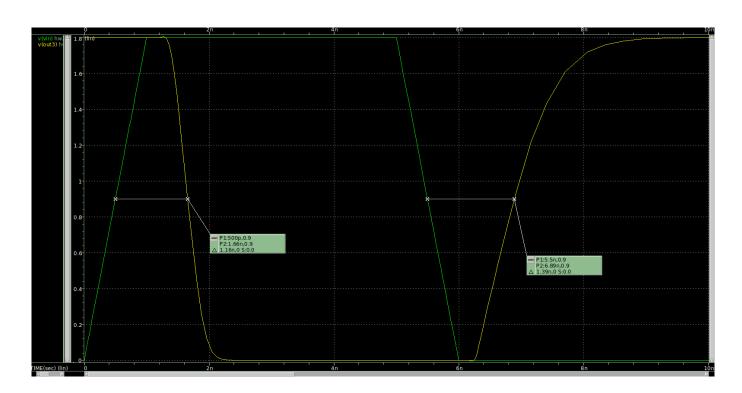


post-sim:

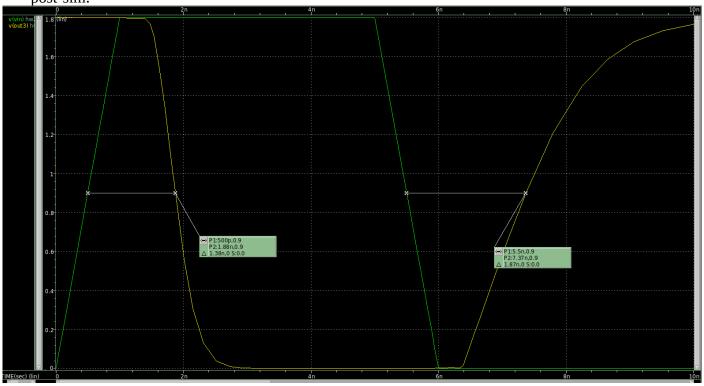


inverter chain with 8 stage(last half) pre-sim:1274ps; post-sim:1623ps;delay\_time 變化(pre-sim-post-sim): 349ps(tpHL:1157ps to 1375ps;tpLH:1391ps to 1870ps)

pre-sim:



post-sim:



# What I have learned in this HW:

在 hw2 中我學到的主要是如何使用 laker 和 composer,並且還學到關於 layout 的規定和格式,此外也觀察到 post-sim 的結果會和如何不同,還有如何實現用 nmos 和 pmos 來表示 logic circuit,和觀察不同長度的 inverter chain 間的差異。