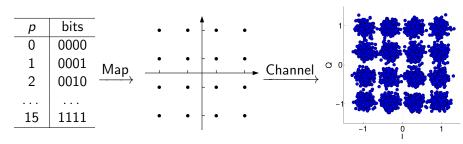
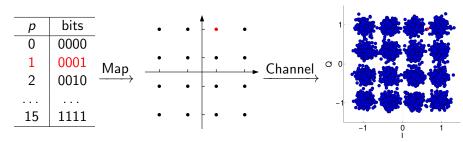
Modulation Mapping



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Single Transmission: Gray-mapping

Strategy (Gray-mapping)

Neighboring constellation points (horizontally or vertically) differ only by 1 bit, so as to minimize the Bit Error Rate (BER).

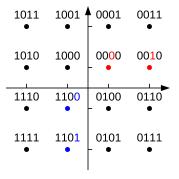


Figure: Gray-mapping for 16-QAM, 3GPP TS 25.213.

HARQ with Constellation Rearrangement (CoRe)

Hybrid Automatic Repeat reQuest (HARQ)

- Same piece of information is retransmitted again and again, and combined at the receiver until it is decoded successfully or expiration.
- An error control scheme widely used in modern wireless systems such as HSPA, WiMAX, LTE, etc.

Constellation Rearrangement (CoRe)

- ► For each round of retransmission, different modulation mappings are used (explained next).
- Exploit the Modulation Diversity (MoDiv).

An Example of CoRe

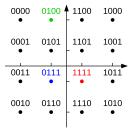
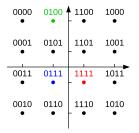


Figure: Original transmission.

▶ Original transmission: 0111 is easily confused with 1111, but well distinguished from 0100.

An Example of CoRe



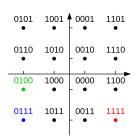


Figure: Original transmission.

Figure: First retransmission.

- ▶ Original transmission: 0111 is easily confused with 1111, but well distinguished from 0100.
- ► First retransmission: 0111 should now be mapped far away from 1111, but can be close to 0100.

General Design of MoDiv Through CoRe

Challenges

- 1. More than 1 retransmissions?
- 2. More general wireless channel models?
- 3. Larger constellations (e.g. 64-QAM)?

We formulated 2 different MoDiv design problems into Quadratic Assignment Problems (QAPs) and demonstrate the performance gain over existing CoRe schemes.

Two-Way Relay Channel (TWRC) with Analog Network Coding (ANC)

▶ System components: 2 sources (S_1, S_2) communicate with each other with the help of 1 relay (R).







Figure: TWRC-ANC channel.

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 - Multiple-Access Channel (MAC) phase: the 2 sources transmit to the relay simultaneously.
 - Broadcast Channel (BC) phase: the relay amplify and broadcast the signal received during the MAC phase back to the 2 sources



Figure: TWRC-ANC channel.

HARQ-Chase Combining (CC) Protocol

- Q: size of the constellation.
- M: maximum number of retransmissions.
- $\psi_m[p]$, $m = 0, \dots, M$, $p = 0, \dots, Q 1$: constellation mapping function between "label" p to a constellation point for the m-th retransission.

Due to symmetry of the channel, consider the transmission from S_1 to S_2 only. The received signal during the m-th retransmission of label p is:

$$y_2^{(m)} = \alpha^{(m)} g_2^{(m)} (h_1^{(m)} \psi_m[p] + h_2^{(\tilde{m})} \psi_{\tilde{m}}[\tilde{p}] + n_R^{(m)}) + n_2^{(m)},$$

HARQ-Chase Combining (CC) Protocol

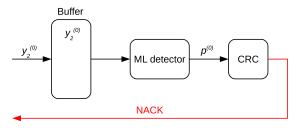
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 (after SIC)

HARQ-Chase Combining (CC) Protocol (Continued)

The receiver combines all the received symbols across all retransmissions so far until decoding is determined successful.

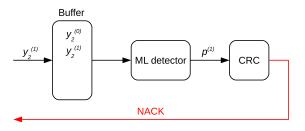


Maximum Likelihood (ML) detector

$$p^* = \arg\min_{p} \sum_{k=0}^{m} \frac{|y_2^{(k)} - \alpha^{(k)} g_2^{(k)} h_1^{(k)} \psi_k[p]|^2}{\sigma_2^2 + (\alpha^{(k)})^2 \sigma_R^2 |g_2^{(k)}|^2}.$$

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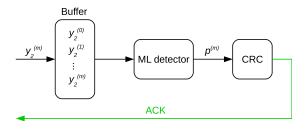


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MoDiv Design: Criterion

Bit Error Rate (BER) upperbound after m-th retransmission

$$P_{BER}^{(m)} = \sum_{p=0}^{Q-1} \sum_{q=0}^{Q-1} \frac{D[p,q]}{Q \log_2 Q} P_{PEP}^{(m)}(q|p),$$

- ▶ D[p, q]: hamming distance between the bit representation of label p and q.
- ▶ $P_{PEP}^{(m)}(q|p)$: pairwise error probability (PEP), the probability that when label p is transmitted, but the receiver decides q is more likely than p after m-th retransmission.

MoDiv Design: Criterion (Continued)

Is minimizing $P_{BER}^{(m)}$ over the mappings $\psi_1[\cdot], \dots, \psi_m[\cdot]$ directly a good idea?

- 1. No one knows how many retransmissions is needed in advance (value of m).
- 2. Jointly designing all *m* mappings is prohibitively complex.
- 3. $P_{PEP}^{(m)}(q|p)$ can only be evaluated numerically, very slow and could be inaccurate.

MoDiv Design: Modified Criterion

1. Successive optimization instead of joint optimization.

Joint:
$$\min_{\psi^{(k)}, k=0,\dots,m} P_{BER}^{(m)}, m=1,\dots,M$$

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2. A closed-form approximation to $P_{PEP}^{(m)}(q|p)$ that can be iteratively updated for growing m.

$$\begin{split} \tilde{P}_{PEP}^{(m)}(q|p) &= \tilde{P}_{PEP}^{(m-1)}(q|p)\tilde{E}_k[p,q] \\ \tilde{P}_{PEP}^{(-1)}(q|p) &= 1/2 \end{split}$$

Representation of CoRe

Representing $\psi_m[\cdot]$ with Q^2 binary variables:

$$\mathbf{x}_{pi}^{(m)} = \left\{ egin{array}{ll} 1 & ext{if } \psi_m[p] = \psi_0[i] \\ 0 & ext{otherwise.} \end{array} \right. \quad p,i = 0,\ldots,Q-1$$

 ψ_0 represents Gray-mapping for the original transmission (fixed).

Constraints: $\psi_m[\cdot]$ as a permutation of $0, \ldots, Q-1$

$$\sum_{p=0}^{Q-1} x_{pi} = 1$$

$$\sum_{i=0}^{Q-1} x_{pi} = 1$$

	i = 0	i = 1	i = 2	<i>i</i> = 3
p=0	0	1	0	0
p=1	0	0	1	0
p = 2	1	0	0	0
<i>p</i> = 3	0	0	0	1

A Successive KB-QAP Formulation

MoDiv design via successive KB-QAP

1. Set m=1. Initialize the "distance" matrix and the approximated PEP, for $i,j,p,q=0,\ldots,Q-1$:

$$d_{ij} = \tilde{E}_k[i,j], \ \tilde{P}_{PEP}^{(0)}(q|p) = d_{pq}/2$$

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3. Solve the *m*-th KB-QAP problem:

$$\min_{\{x_{p_i}^{(m)}\}} \sum_{p=0}^{Q-1} \sum_{i=0}^{Q-1} \sum_{q=0}^{Q-1} \sum_{j=0}^{Q-1} f_{pq}^{(m)} d_{ij} x_{pi}^{(m)} x_{qj}^{(m)}$$

A Successive KB-QAP Formulation (Continued)

MoDiv design via successive KB-QAP

4. Update PEP:

$$\tilde{P}_{PEP}^{(m)}(q|p) = \sum_{i=0}^{Q-1} \sum_{j=0}^{Q-1} \tilde{P}_{PEP}^{(m-1)}(q|p) d_{ij} \hat{x}_{pi}^{(m)} \hat{x}_{qj}^{(m)}$$

where $\hat{x}_{pi}^{(m)}$ is the solution from Step 3.

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5. Increase m by 1, return to Step 2 if $m \leq M$.

▶ 64-QAM constellation (Q = 64).

¹E. Taillard, "Robust taboo search for the quadratic assignment problem," Parallel Computing, vol.17, no.4, pp.443-455, 1991.

² "Enhanced HARQ Method with Signal Constellation Rearrangement," in 3rd Generation Partnership Project (3GPP), Technical Specification TSGR1#19(01)0237, Mar. 2001.

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Numerical Results: Uncoded BER

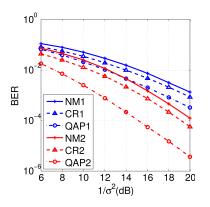


Figure : m = 1, 2.

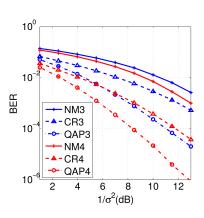
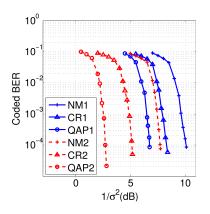


Figure: m = 3, 4.

Numerical Results: Coded BER

Add a Forward Error Correction (FEC) code so that the coded BER drop rapidly as the noise power is below a certain level. The result is termed "waterfall curve" which is commonly used to highlight the performance gain in dB.



10⁰ 10 H 10⁻² Soded I 10 10 -5 $1/\sigma^2(dB)$

Figure: m = 1, 2.

Figure : *m*= 3, 4.



Numerical Results: Average Throughput

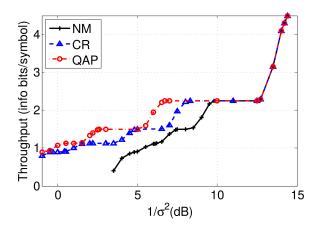


Figure: Throughput comparison.