

# Ve 216: Introduction to Signals and Systems

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February 27, 2018

Based on Lecture Notes by Prof. Jeffrey A. Fessler

# Outline

- 1 1. Signals & Systems (Fundamentals)
  - Overview
  - Signal and System Definition
  - Classification of Signals
  - Signal Notation
  - Transformations of CT signals
  - Signal Characteristics
  - Exponential signals
  - Singularity functions (1.4)
  - Continuous-time systems
  - Summary

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# Overview

## Signals

- definition
- classes
- notation
- transformations  
(operations)
- important signals  
(Skip: 1.3.2, 1.3.3, 1.4.1)

## Systems

- definition
- block diagrams
- system interconnection
- classes
- **linearity, time-invariance**

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Goal: eventually system design; must first learn to analyze!

# Overview

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# Signal Definition

## Definition

A **signal** is any “physical” quantity that varies with time or space (or any other independent variable or variables).

Often when we discuss signals we refer to mathematical representation of the physical quantity.

## Example

An approaching **ambulance siren** produces a time-varying change in acoustic pressure that our ears perceive as sound.

$$s(t) = (1 + t) \sin\left(2\pi[1000t + 10t^2 + 300 \sin(2\pi t/2)]\right)$$

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# Example: Ambulance Siren

$$s(t) = (1 + t) \sin\left(2\pi[1000t + 10t^2 + 300 \sin(2\pi t/2)]\right)$$

- The  $(1 + t)$  amplitude term represents increasing loudness as the ambulance approaches.
- The  $1000t$  term represents the 1kHz siren oscillation.
- The  $10t^2$  term represents increasing pitch due to the Doppler effect as the ambulance approaches.
- The  $300 \sin(2\pi 2t)$  term represents the eeh-ooh-eeh-ooh periodic variation in pitch.



# System Definition

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A **system** is a physical “device” that performs an operation on a signal.

## Example

The **human ear** converts **acoustic signals** into **electrical nerve synapses** (another signal) that are processed by the brain. The input and output signals are different physical quantities.

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# Signal processing

One of the main roles of electrical engineers is to **design and analyze systems** that take some **input signal** and produce some related (but almost always different) **output signal**.

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# Signals and Systems

## Example

For an **audio amplifier**, ideally the output signal is “simply” an amplified version of the input signal. (On paper it is easy:

$$s_{\text{out}}(t) = as_{\text{in}}(t).$$

But implementing this in analog hardware with **minimal distortion** is **nontrivial**.)

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This course will emphasize **continuous-time** or **analog** signals, and briefly introduce **discrete-time** or **digital** signals at the end.

(Portions of Chapters 1-10)

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# A signal is a function

Mathematically, a **signal** is a **function** of one or more independent variables.

## Question

*What is a function?*

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*It is a rule that assigns to each value in the range set at least one of the values in the domain set.*

# Classification of Signals: Dimensionality (1)

One way to classify signals is by the dimension of the **domain of the function**, i.e. how many arguments the function has.

## Definition

A **one-dimensional** signal is a function of a single variable, e.g. time.

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# Classification of Signals: Dimensionality (3)

## Example

A sequence of BW TV pictures  $I(x, y, t)$  is a scalar valued function of two spatial coordinates  $x$  and  $y$  and time  $t$ , so it is a **3D signal**.

We will focus on **one-dimensional** signals in this course, generally considering the independent variable to be **time  $t$** .

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Another way to classify signals is by the dimension of the range of the function, *i.e.*, the space of values the function can take.

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A **scalar** or **single-channel** signal is a function of a real-valued scalar or complex-valued scalar.

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A **multichannel** signal is a function of a real vector or complex vector.

# Classification of Signals: Dimensionality (5)

## Example

A **color** TV picture can be described by a red, blue and green signal, whereas a **BW** TV picture is scalar valued.

We will focus on **scalar** signals in this course, both real and complex.

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A **continuous-time signal** or **analog signal** is a function defined for all times  $t \in (-\infty, \infty)$ , or at least over some continuous interval  $(a, b)$ .

## Example

$$x(t) = e^{-t^2}, \quad -\infty < t < \infty. \quad (\text{Picture})$$

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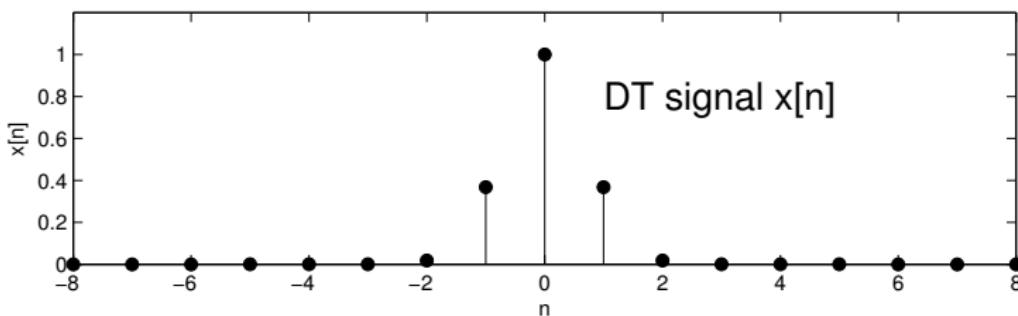
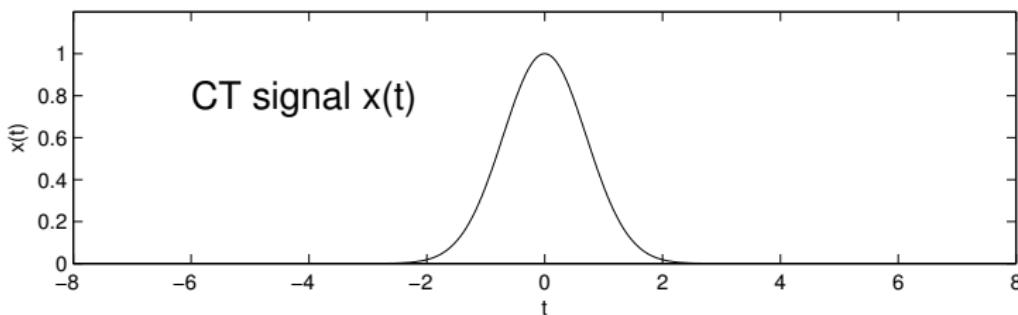
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# Continuous-time signals vs. discrete-time signals



# Classification of Signals: time characteristics

Classify signals by **time characteristics**

- ① **Continuous-time signals or analog signals**
- ② **Discrete-time signals**

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Discrete-time signals arise from

- **Sampling** a continuous signal at discrete time instants
- **Accumulating** a quantity over a period of time

## Example

When counting number of heart attacks per month,  $n$  would index the month, and  $x[n]$  would be the number.

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Voltage between 0 and 5 volts.

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- ① **Deterministic signals** can be described by an explicit mathematical representation.
- ② **Random signals** evolve over time in an unpredictable manner.

## Example

“Hiss” or “noise” in an audio system.

We will focus on **deterministic signals**, although reducing noise (eliminating a random component) is often a goal in designing signal processing systems.

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# Classification: Our focus

We will focus on

- single-channel, one-dimensional, continuous-valued, continuous-time signals.
- $x(t)$  is a scalar valued function of a real independent variable.
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$$x : \mathbb{R} \rightarrow \mathbb{R} \text{ or } x : \mathbb{R} \rightarrow \mathbb{C}$$

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# Signal notation (mathematical representation)(1)

① Graphically (**Picture**):

② Braces or piecewise notation:  $x(t) = \begin{cases} e^{-t}, & t > 0 \\ e^t, & t \leq 0. \end{cases}$

③ Formula:  $x(t) = e^{-|t|}$ .

④ In terms of other functions:  $x(t) = s(t) + s(-t)$  where

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⑤ Fourier representation:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{2}{1+\omega^2} e^{j\omega t} d\omega$$

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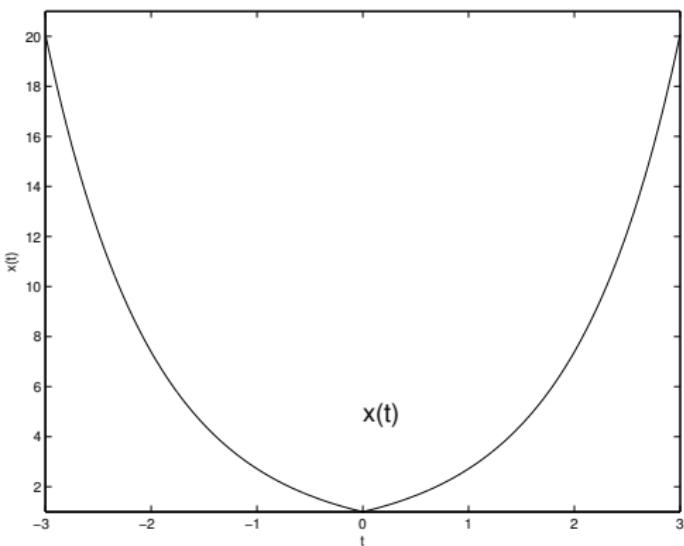
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# Eventual goal

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# Transformations of CT signals

- Time transformations
  - Folding/reflection/time-reversal
  - Time-scaling
  - Time-shifting/time-delay
  - General time transformation
- Amplitude transformations
  - Amplitude reversal
  - Amplitude scaling
  - Amplitude shifting
- More signal operations
  - Differentiator
  - Integrator
- Operations with two signals

# Change of variables

If  $x(t) = e^{-(t-2)}$  then  $y(t) = x\left(\frac{t-1}{3}\right)$  is another function;

$$y(t) = e^{-[(t-1)/3-2]} = e^{-\left(\frac{t-7}{3}\right)}.$$

In calculus, this type of transformation is called a **change of variables**.

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# Time transformations

Here we give some new names to such transformations to reflect the **physical meaning** of the mathematics.

## Example

$$x(t) = \begin{cases} e^{-(t-2)}, & t \geq 2, \\ 0, & \text{otherwise.} \end{cases} \quad (\text{Picture}).$$

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- Backwards play a movie/audio tape.
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# Folding/reflection/time-reversal: $y(t) = x(-t)$ (2)

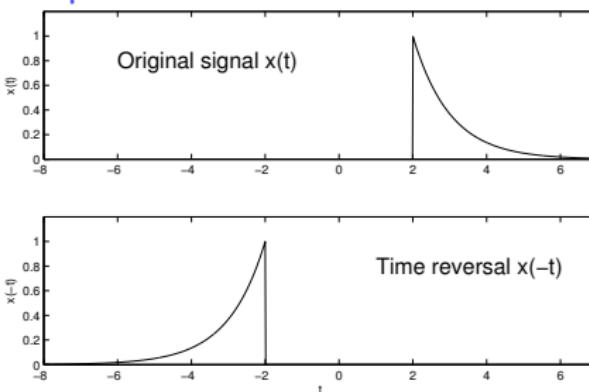
## Mathematical method:

- Replace all  $t$ 's with  $-t$ ,
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$$\begin{aligned} y(t) &= x(-t) \\ &= \begin{cases} e^{-(t+2)}, & -t \geq 2, \\ 0, & \text{otherwise} \end{cases} \\ &= \begin{cases} e^{t+2}, & t \leq -2, \\ 0, & \text{otherwise} \end{cases} \end{aligned}$$

Note that  $y(t)$  becomes a “mirror image” of  $x(t)$  around  $t = 0$ .

## Graphical method:



# Folding/reflection/time-reversal: $y(t) = x(-t)$ (2)

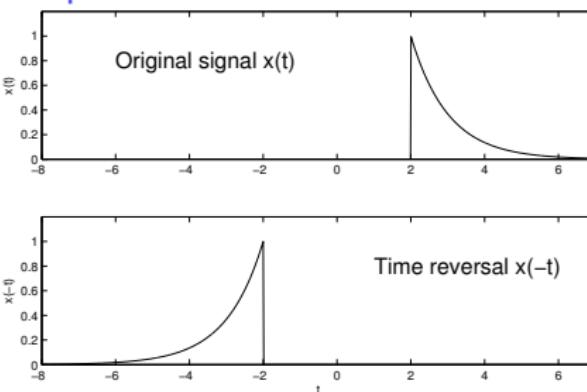
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Note that  $y(t)$  becomes a “mirror image” of  $x(t)$  around  $t = 0$ .

## Graphical method:



# Time-scaling: $y(t) = x(at)$ , $a > 1$

## Time-scaling

$$y(t) = x(at)$$

$a > 1$  will shrink or compress the signal

### Example

Playing a recording at 3 times the normal speed.

### Example

Find  $y(t) = x(3t)$  for  $x(t)$  above.

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$$y(t) = x(3t) = \begin{cases} e^{-(3t-2)}, & 3t \geq 2 \\ 0, & \text{otherwise} \end{cases} = \begin{cases} e^{-3(t-2/3)}, & t \geq 2/3 \\ 0, & \text{otherwise} \end{cases}$$

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slow-motion part of a movie

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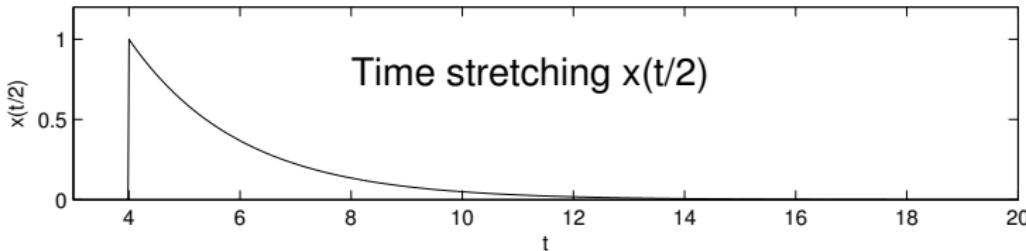
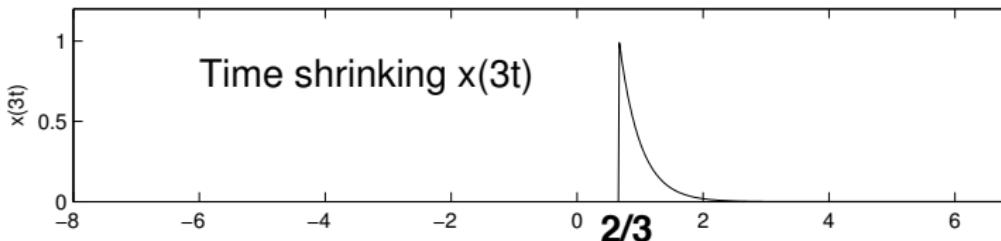
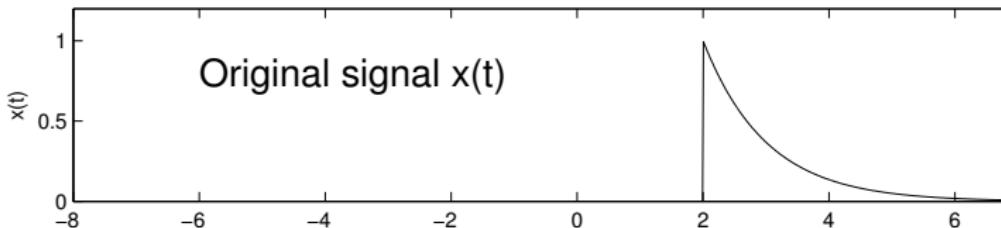
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Find  $y(t) = x(t/2)$  for  $x(t)$  above.

$$y(t) = x(t/2) = \begin{cases} e^{-(t/2-2)}, & t/2 \geq 2 \\ 0, & \text{otherwise} \end{cases} = \begin{cases} e^{-(t-4)/2}, & t \geq 4 \\ 0, & \text{otherwise} \end{cases}$$

# Time-scaling: $y(t) = x(at)$



# Time shifting: $y(t) = x(t - t_0)$

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- $t_0$  can be positive (delayed signal) or negative (advanced signal).
- Physical systems can only delay, not advance, in time.

## Example

“Park distance control” propagation delay

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Find  $y(t) = x(t - 1)$  for  $x(t)$  above.

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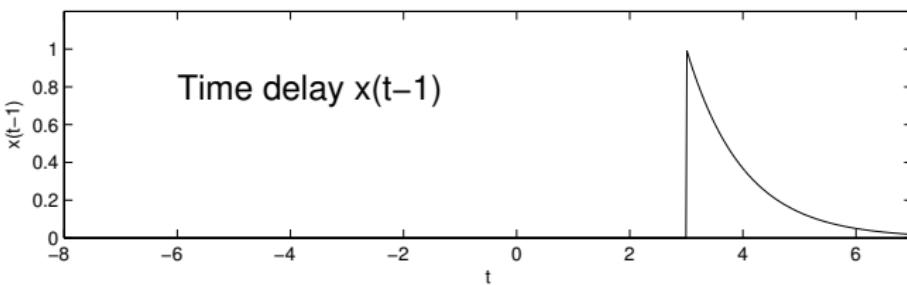
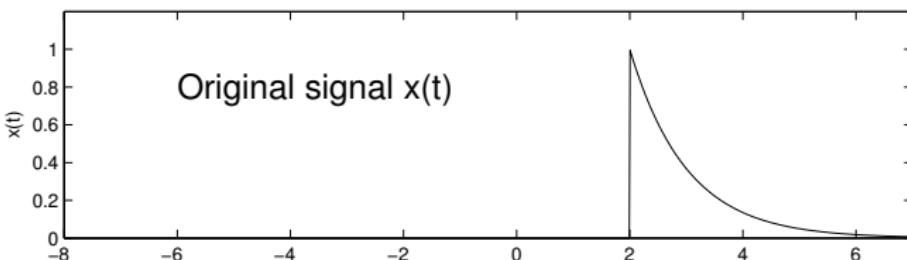
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$$y(t) = x(t-1) = \begin{cases} e^{-(t-1-2)}, & t-1 \geq 2 \\ 0, & \text{otherwise} \end{cases} = \begin{cases} e^{-(t-3)}, & t \geq 3 \\ 0, & \text{otherwise} \end{cases}$$

# Time shifting: $y(t) = x(t - t_0)$ (Cont.)



Note that  $x(t-1)$  delays the signal, which means it shifts to the right (starts later in time).

# General time transformation

General time transformation involves all three of the above time transformations (time reversal, time scaling, and time shifting).

Two distinct (but related) forms:

$$y(t) = x(at - b) = x\left(\frac{t - t_0}{w}\right)$$

where  $t_0 = b/a$  and  $w = 1/a$  or  
equivalently  $a = 1/w$  and  $b = t_0/w$ .

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mathematical recipe:

- ① Replace **all** occurrences of  $t$  in the definition of  $x(t)$  with  $at - b$  or with  $\frac{t-t_0}{w}$ .
- ② Manipulate algebraically to simplify.

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$$x(t) = \begin{cases} e^{-(t-2)}, & t \geq 2, \\ 0, & \text{otherwise.} \end{cases}$$

$$\begin{aligned} y(t) &= x(-t/2 + 5) \\ &= \begin{cases} e^{-(-t/2+5)-2}, & -t/2 + 5 \geq 2 \\ 0, & \text{otherwise} \end{cases} \\ &= \begin{cases} e^{(t-6)/2}, & t \leq 6 \\ 0, & \text{otherwise.} \end{cases} \end{aligned}$$

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## Question

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*How to perform a general time transformation graphically?  
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*The answer is: it depends on which of the two above forms you use to write the transformation!*

# Form 1: $y(t) = x(at - b)$

$$\boxed{\text{Form 1 : } y(t) = x(at - b)}$$

- Introduce an intermediate signal  $d(t) = x(t - b)$ .
- Clearly  $d(t)$  is just a **delayed** version of  $x(t)$  by amount  $b$ .
- But  $y(t) = d(at)$ , which is just a **scaled** version of the signal  $d(t)$ .

To find  $y(t) = x(at - b)$  graphically we must

- ➊ time-delay the signal  $x(t)$  by  $b$ .
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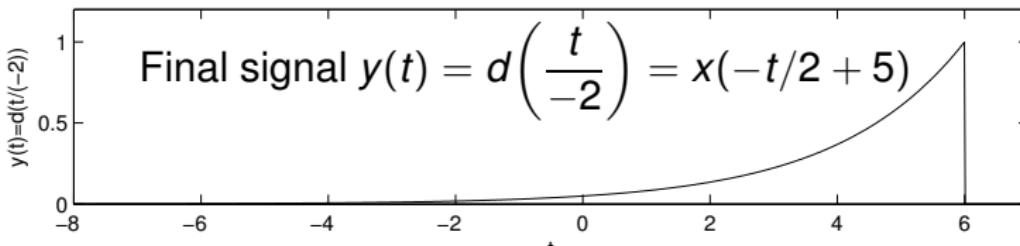
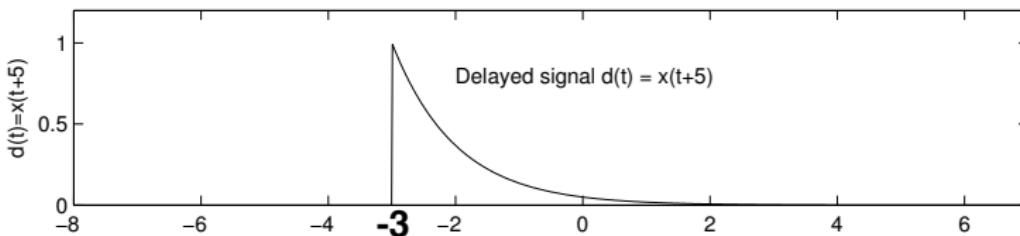
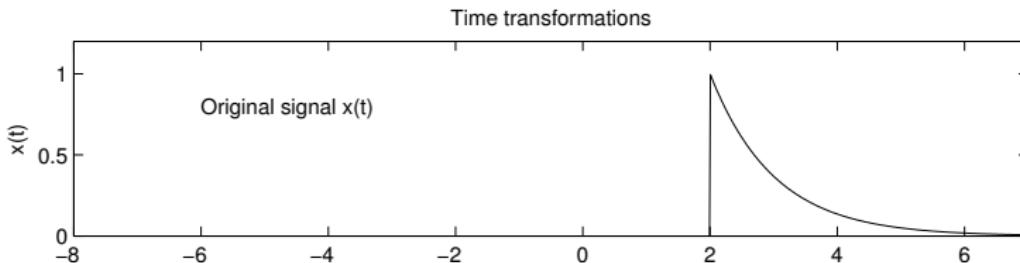
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**Form 2:**  $y(t) = x\left(\frac{t-t_0}{w}\right)$

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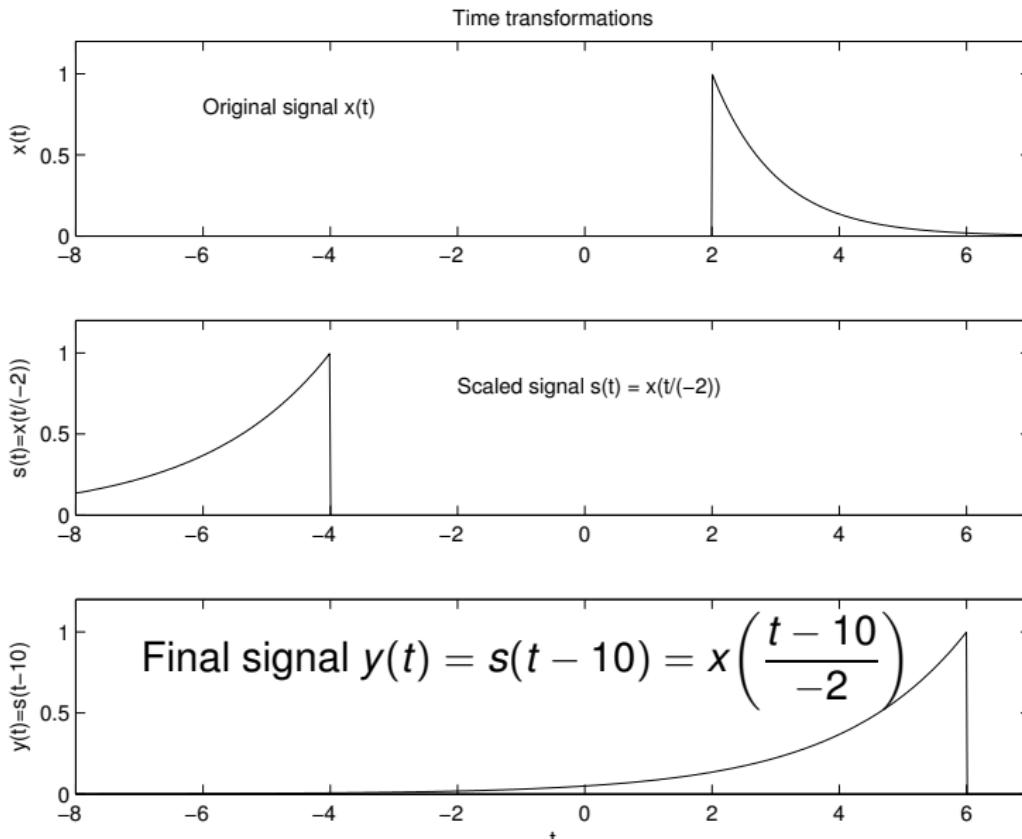
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# Amplitude transformations

- ① **amplitude reversal**  $y(t) = -x(t)$
- ② **amplitude scaling**  $y(t) = ax(t)$
- ③ **amplitude shifting**  $y(t) = x(t) + b$

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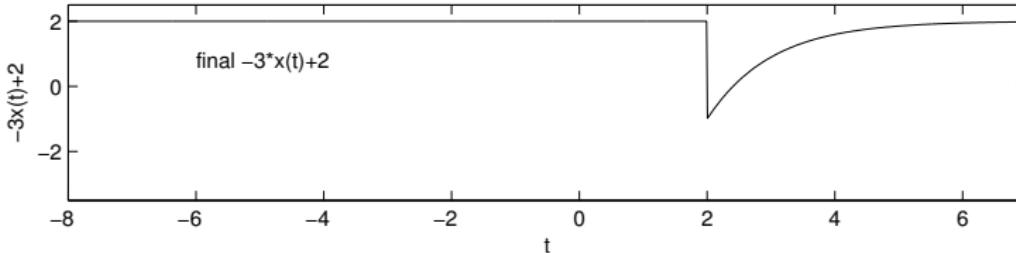
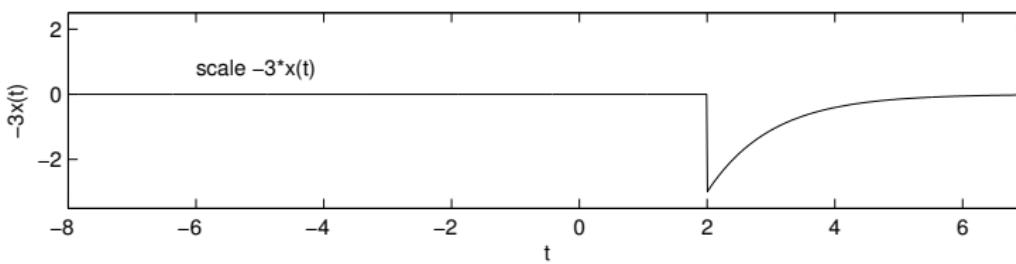
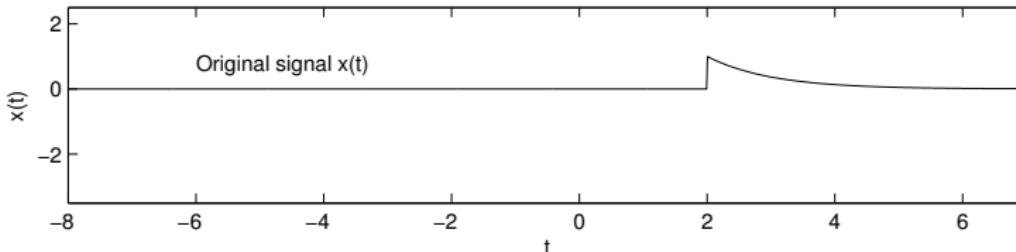
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$$y(t) = -3x(t) + 2 = \begin{cases} -3e^{-(t-2)} + 2, & t \geq 2 \\ -3 \cdot 0 + 2, & \text{o.w.} \end{cases}$$

# Amplitude transformations

*Graphically: scale amplitude by -3 and then shift up by 2.*



# Outline

1

## 1. Signals & Systems (Fundamentals)

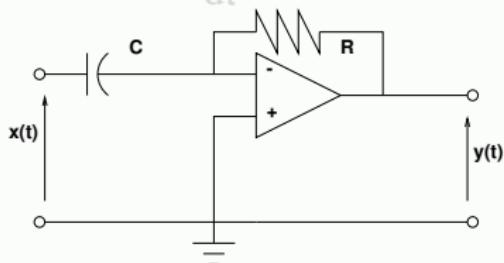
- Overview
- Signal and System Definition
- Classification of Signals
- Signal Notation
- **Transformations of CT signals**
- Signal Characteristics
- Exponential signals
- Singularity functions (1.4)
- Continuous-time systems
- Summary

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### Example

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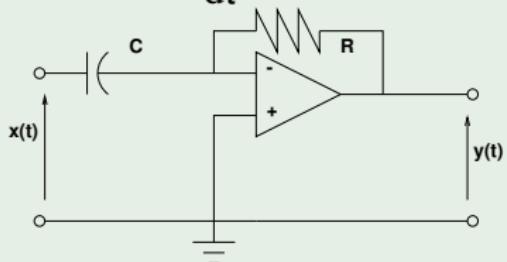
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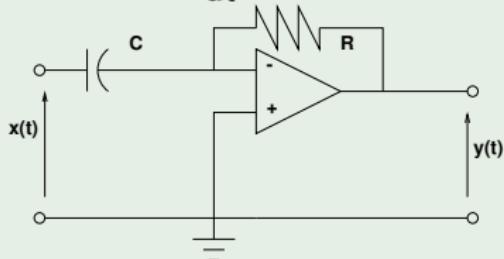
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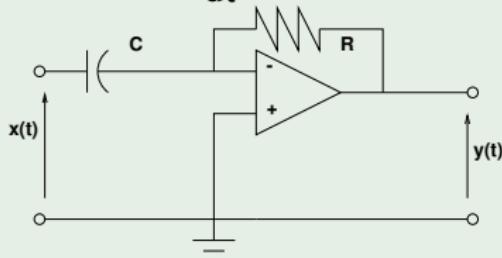
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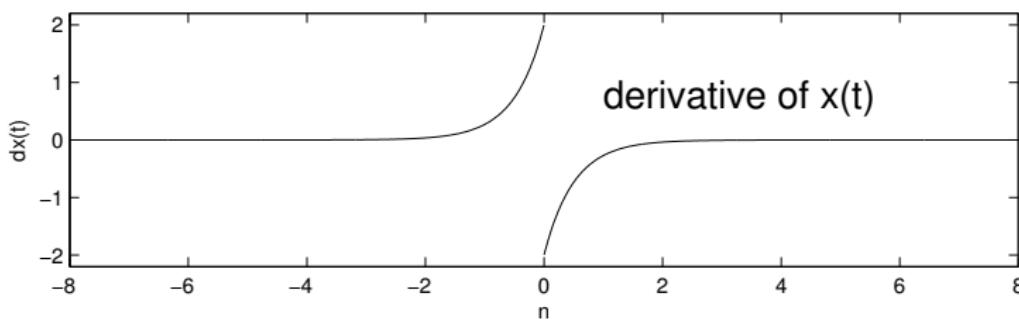
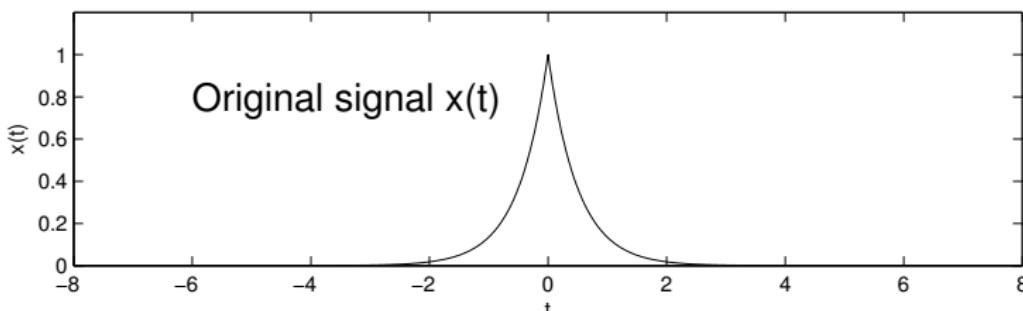
Find the differentiated signal of  $x(t) = e^{-2|t|}$ .

$$\begin{aligned} y(t) &= \frac{d}{dt}x(t) \\ &= \begin{cases} -2e^{-2t}, & t > 0 \\ 2e^{2t}, & t < 0 \\ ?, & t = 0. \end{cases} \end{aligned}$$

*The derivative of  $x(t)$  is undefined at  $t = 0$ .*



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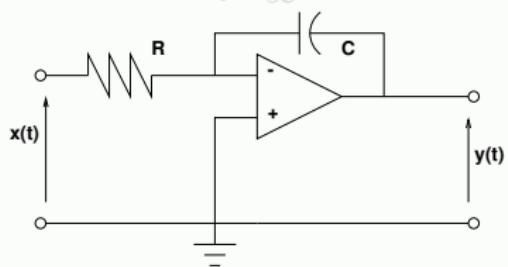


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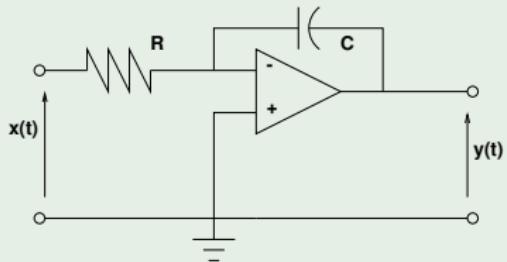
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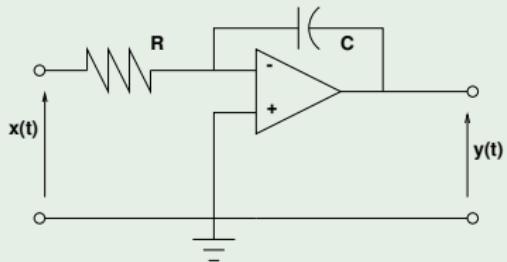
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# Integrator: example

## Solution

- 1 rewrite  $x(\cdot)$  in terms of  $\tau$

$$x(\tau) = \begin{cases} e^{-\tau}, & \tau > 0 \\ e^{\tau}, & \tau \leq 0. \end{cases}$$

- 2 For  $t \leq 0$ :

$$y(t) = \int_{-\infty}^t e^{\tau} d\tau = e^t$$

- 3 For  $t \geq 0$ :

$$\begin{aligned} y(t) &= \int_{-\infty}^0 e^{\tau} d\tau + \int_0^t e^{-\tau} d\tau \\ &= e^0 + (-e^{-\tau}) \Big|_0^t = 1 + (1 - e^{-t}) \\ &= 2 - e^{-t}. \end{aligned}$$

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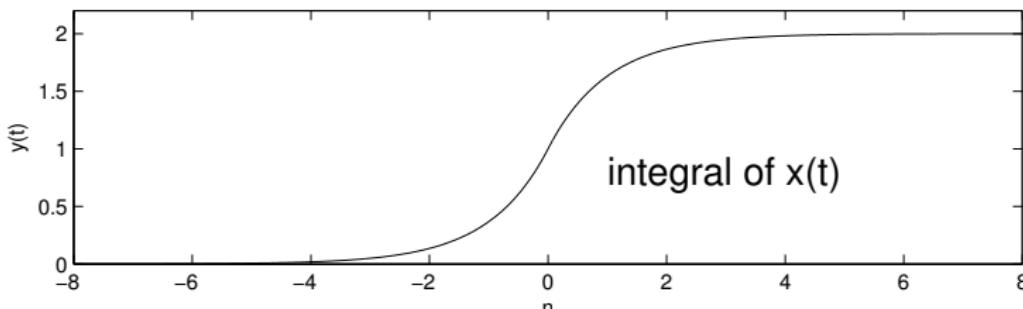
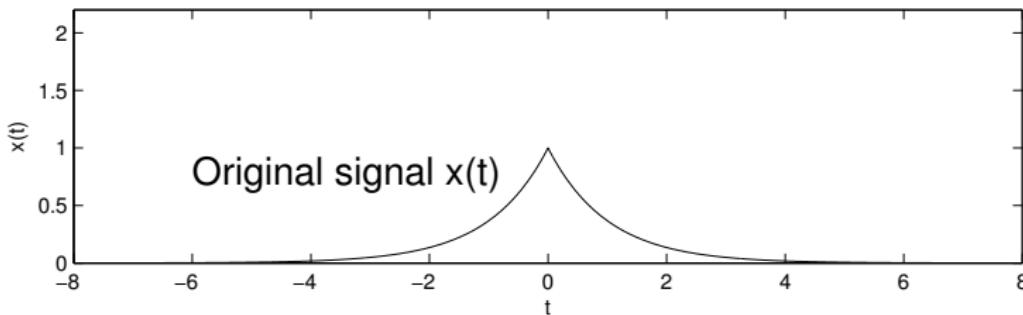
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# Integrator: example (Cont.)

$$y(t) = \int_{-\infty}^t x(\tau) d\tau = \begin{cases} e^t, & t \leq 0 \\ 2 - e^{-t}, & t > 0. \end{cases}$$



# Integrator vs. Integration in calculus

## Question

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*What is the **distinction** between **simple integration** of the kind learned in **calculus** (computing area under a curve) and the **integrator system** described here.*

- When you calculate “area under a curve” you compute a **single number** from the curve, e.g.  $\int_0^1 t^2 dt = 1/3$ .
- **Integrator system** in signals and systems transforms one function  $x(t)$  into another **function**  $y(t)$ .  
*It is a “running integral” whose upper limit is  $t$ .*



# Outline

## 1. Signals & Systems (Fundamentals)

- Overview
- Signal and System Definition
- Classification of Signals
- Signal Notation
- **Transformations of CT signals**
  - Time transformations
  - Amplitude transformations
  - More signal operations
  - **Operations with two signals**
- Signal Characteristics
  - Periodic/aperiodic signals
  - Even and odd signals
  - Energy and power signals
- Exponential signals
- Singularity functions (1.4)
  - Unit step signal
  - Rect(angle) function
  - Unit impulse function  $\delta(t)$ (1.4.2, 2.5)
- Continuous-time systems

# Operations with two signals

## Operations with two signals

- ① sum of two signals  $y(t) = x_1(t) + x_2(t)$
- ② product of two signals  $y(t) = x_1(t)x_2(t)$ .

Add or multiply two signals at every time point.

## Example

If  $x_1(t) = e^{-(t-3)^2}$ ,  $x_2(t) = \sin(13t)$ , then

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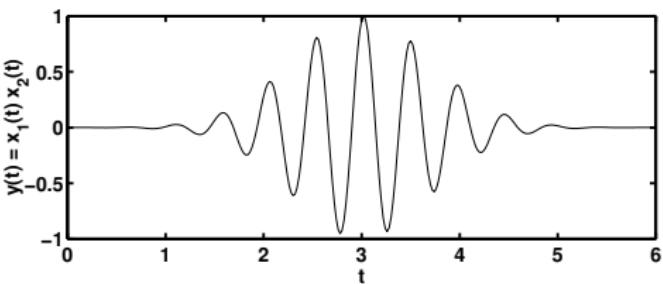
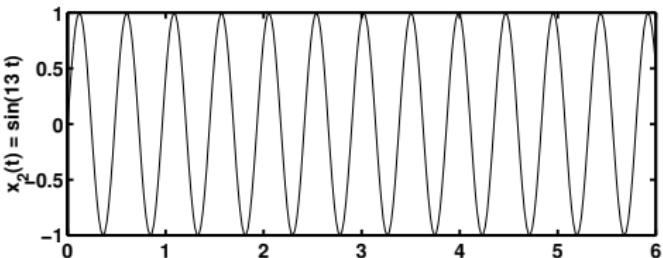
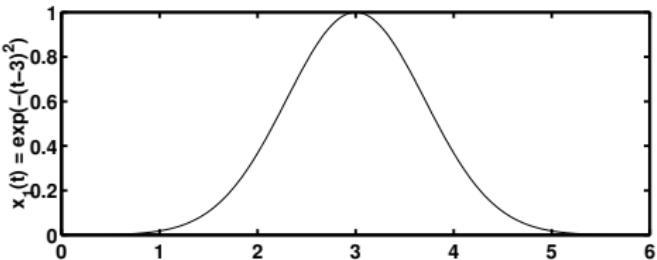
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# amplitude modulation



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# Periodic signals

Why study **periodic signals**?

- important for analysis
- solution to ideal LC electrical circuits
- periodic physical phenomena: frictionless pendulums, earth rotation, heart rhythms, etc.

Definition

$x(t)$  is **periodic** with a **period**  $T > 0$  iff

$$x(t + T) = x(t) \quad \forall t \tag{1}$$

e,p

Definition

If no such  $T > 0$  exists,  $x(t)$  is called **aperiodic**

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The **fundamental period**  $T_0$  of a signal is the smallest value of  $T$  satisfying (1).

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A signal that is periodic with period  $T > 0$ , is also periodic with period  $nT$  for any integer  $n \neq 0$ , i.e.

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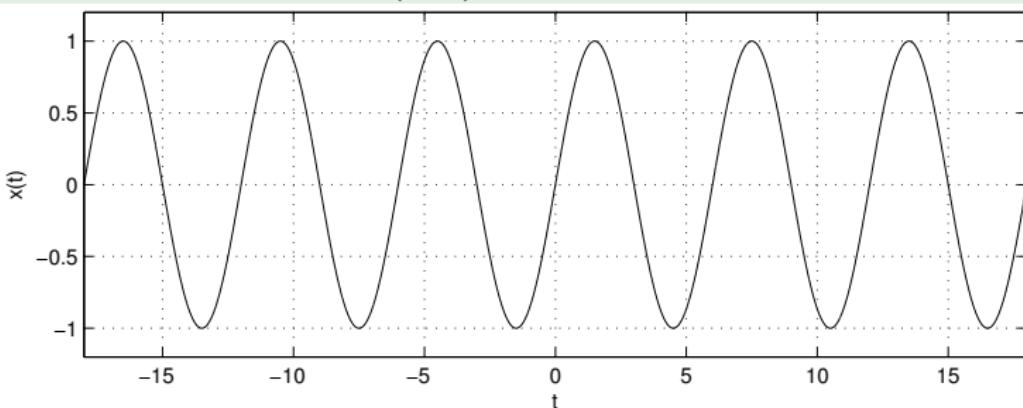
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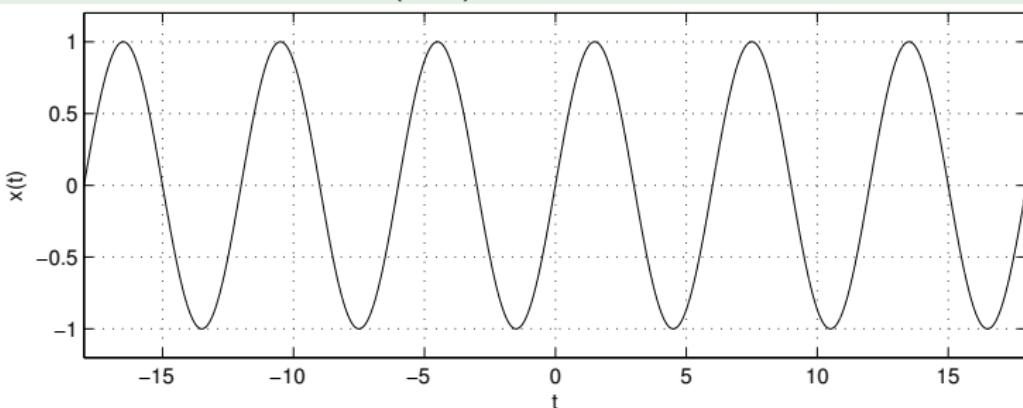
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$$T_0 = 6 = \frac{2\pi}{\pi/3}.$$

# Sums of two periodic signals

## Question

Suppose  $x_1(t)$  is periodic with period  $T_1$  and  $x_2(t)$  is periodic with period  $T_2$  and  $x(t) = x_1(t) + x_2(t)$ .

- Is  $x(t)$  periodic?
- If so, what is a period  $T$  of  $x(t)$ ?

# Sums of two periodic signals: solution (1)

## Solution

- ① *Easy case:* if  $T_1 = T_2$  then,  $x(t)$  is periodic, and  $T = T_1 = T_2$ .
- ② *General case.* We know  $x_1(t) = x_1(t + T_1)$  and  $x_2(t) = x_2(t + T_2)$  and  $x(t) = x_1(t) + x_2(t)$ .  
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## Sums of two periodic signals: solution (2)

Suppose there is a value of  $T > 0$  that satisfies  $T = n_1 T_1$  and  $T = n_2 T_2$ , for some nonzero integers  $n_1$  and  $n_2$ . Then

$$\begin{aligned}x(t + T) &= x_1(t + T) + x_2(t + T) = x_1(t + n_1 T_1) + x_2(t + n_2 T_2) \\&= x_1(t) + x_2(t) = x(t),\end{aligned}$$

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## Sums of two periodic signals: solution (3)

The conditions  $T = n_1 T_1$  and  $T = n_2 T_2$  are equivalent to requiring that

$$n_1 T_1 = n_2 T_2 \text{ so } T_1/T_2 = n_2/n_1,$$

which means that  $T_1/T_2$  is rational, a ratio of two integers.

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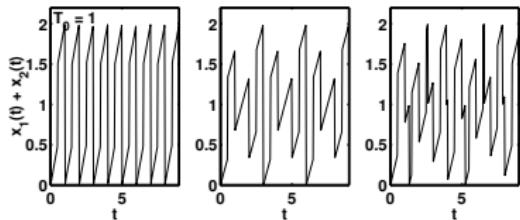
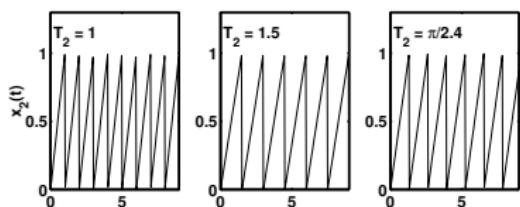
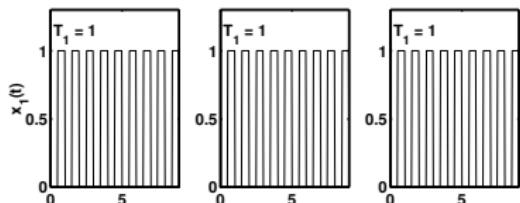
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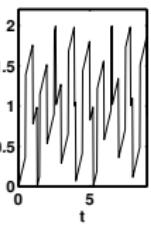
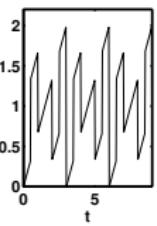
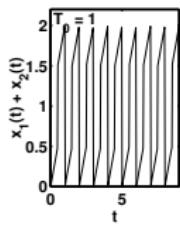
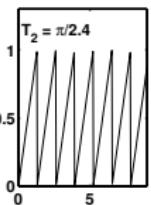
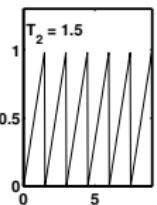
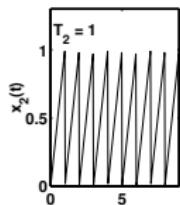
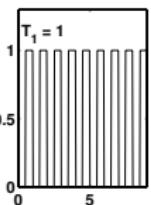
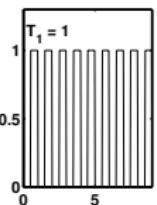
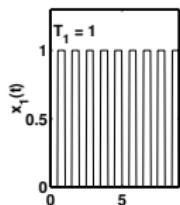
# Sums of two periodic signals: example

Are the signals in the third row periodic?



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Yes. Yes. No.

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*The least common multiple of the fundamental periods of the two signals is a period of the sum. Is it the fundamental period?*

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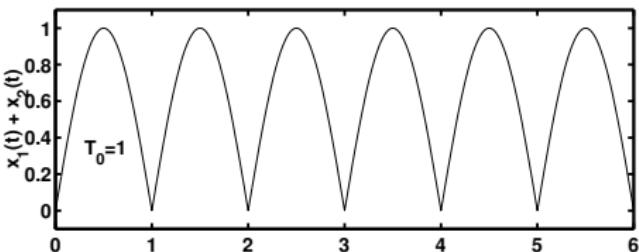
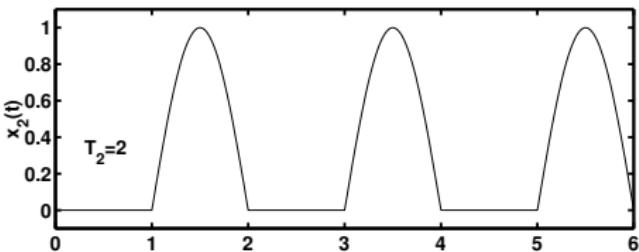
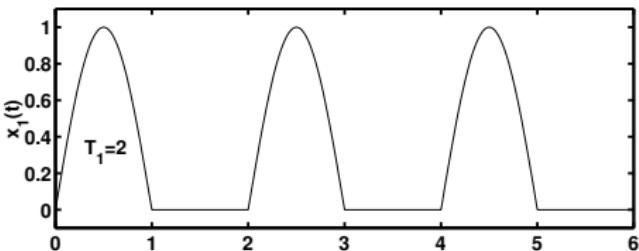
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*To find a least common multiple, you find the smallest values of  $n_1$  and  $n_2$  such that  $n_1 T_1 = n_2 T_2$ .*

# Sums of two periodic signals: fundamental period



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# Even and odd symmetry

## Definition

$x(t)$  has **even symmetry** iff  $x(-t) = x(t) \forall t$

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Note that if  $x(t)$  has **odd symmetry**, then  $x(0) = -x(0)$  so  $x(0) = 0$ .

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**No!** The property  $x(0) = 0$  is a **necessary** condition for  $x(t)$  to have odd symmetry, but it is not a **sufficient** condition.

# Even and odd components

We can decompose any signal into even and odd components:

$$x(t) = x_e(t) + x_o(t)$$

$$x_e(t) \triangleq \frac{1}{2} [x(t) + x(-t)], \quad x_o(t) \triangleq \frac{1}{2} [x(t) - x(-t)]$$

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Is the following  $x(t)$  even or odd?

$$x(t) = \begin{cases} 1, & -1 < t < 3 \\ 0, & \text{otherwise,} \end{cases}$$

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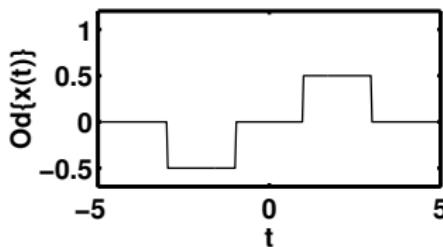
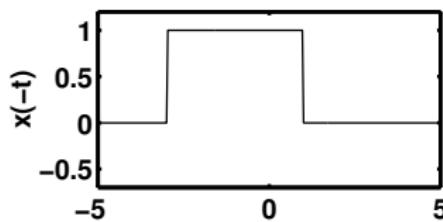
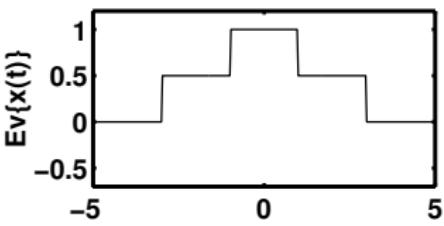
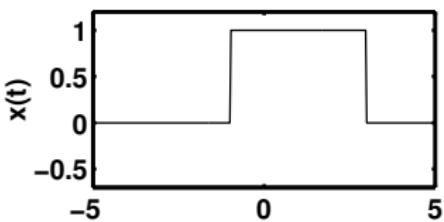
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*Neither.* Find even and odd components.

# Even and odd signals decomposition

$$x_e(t) = \begin{cases} 1/2, & -3 < t < -1 \\ 1, & -1 < t < 1 \\ 1/2, & 1 < t < 3 \\ 0, & \text{otherwise,} \end{cases}$$

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The **average value** of a signal is defined as

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The average value of an odd signal is zero.

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If  $E$  is infinite, then  $P$  can be either finite or infinite. If  $P$  is finite and nonzero, then  $x(t)$  is called a **power signal**.

Some signals are neither energy signals nor power signals, such as  $x(t) = t^2$ , for which  $E = \infty$  and  $P = \infty$ . Such signals are generally of little practical engineering importance.

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## Example

consider  $x(t) = 5 + a \cos t$  where  $0 < a < \infty$ .

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## Solution

The average value of this signal is:

$$\begin{aligned} A &\triangleq \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t) dt = \lim_{T \rightarrow \infty} \frac{1}{2T} (5t + a \sin t)|_{-T}^T \\ &= \lim_{T \rightarrow \infty} \frac{1}{2T} (10T + 2a \sin T) = \lim_{T \rightarrow \infty} \left( 5 - \frac{a \sin T}{2T} \right) = 5. \end{aligned}$$

# Energy and power signals: solution

## Solution

*The average power of this signal is (by a similar integral):*

$$\begin{aligned} P &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T (5 + a \cos t)^2 dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T 25 + 10a \cos t + a^2 \left(\frac{1}{2} + \frac{1}{2} \cos 2t\right) dt \\ &= \lim_{T \rightarrow \infty} \frac{25(2T) + 10a(2 \sin T) + \frac{a^2}{2}(2T + \sin(2T))}{2T} \\ &= 25 + a^2/2. \end{aligned}$$

# Energy and power signals: solution

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Since  $0 < P < \infty$ ,  $x(t)$  is a **power** signal.

Since  $P$  is nonzero and  $E$  is infinite,  $x(t)$  is **not an energy signal**.

# Outline

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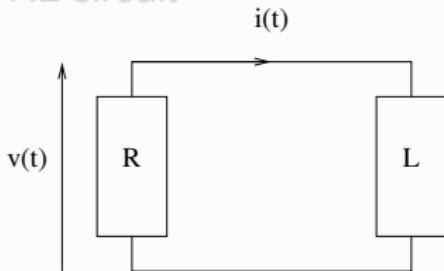
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- **Exponential signals**
- Singularity functions (1.4)
  - Unit step signal
  - Rect(angle) function
  - Unit impulse function  $\delta(t)$ (1.4.2, 2.5)
- Continuous-time systems

# Exponential signals (1)

Sinusoidal signals, exponential signals, and complex exponential signals are particularly important because they arise from the solutions of linear constant-coefficient differential equations.

## Example

an RL circuit



$$\frac{d}{dt}v(t) = \left(-\frac{R}{L}\right)v(t) = av(t)$$

where  $a = -R/L$ . Solution for  $t > 0$  is

$$v(t) = v(0)e^{at} = v(0)e^{-t/\tau}$$

where  $\tau = -1/a = L/R$  is called the time constant of the circuit. (**Picture**)

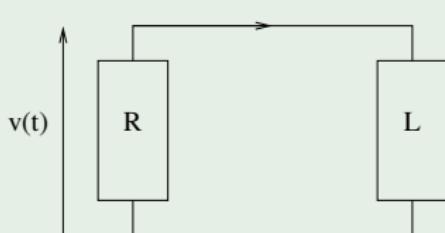
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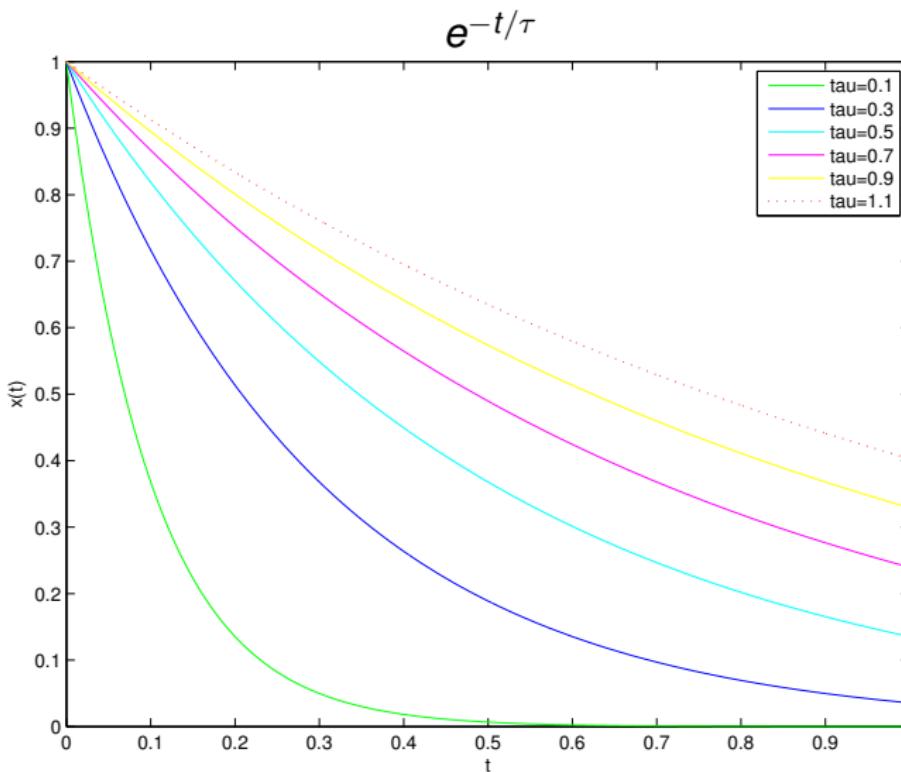
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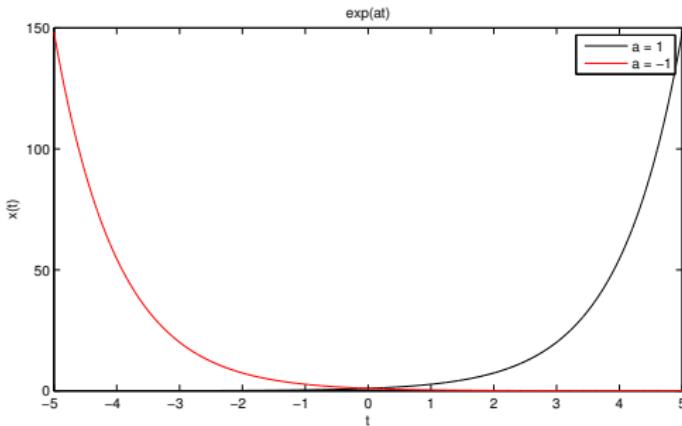
# Exponential signals (2)



# Exponential signals (3)

Signals of the form  $x(t) = ce^{at}$  are very important, for both real and complex  $c$  and  $a$ .

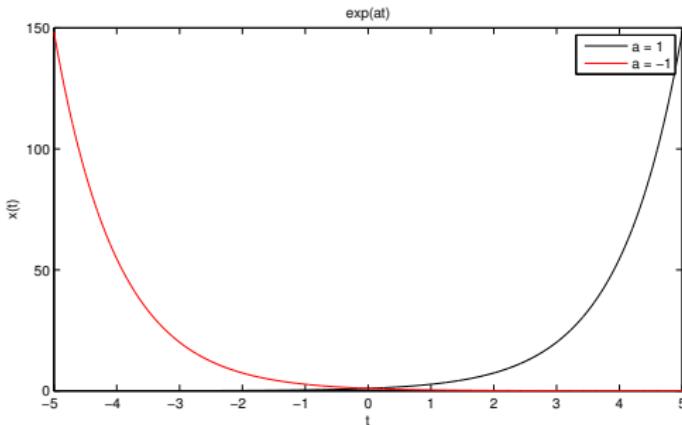
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- If  $a$  is purely imaginary, we get  $x(t) = ce^{j\omega_0 t}$ , a **complex exponential** signal.
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- If  $x(t) = e^{st}$  where  $s = a + j\omega_0$  and  $a < 0$ , then  $x(t) = e^{at}(\cos \omega_0 t + j \sin \omega_0 t)$  which is called a **damped sinusoid** signal. (See textbook p21.)

Euler's formula:

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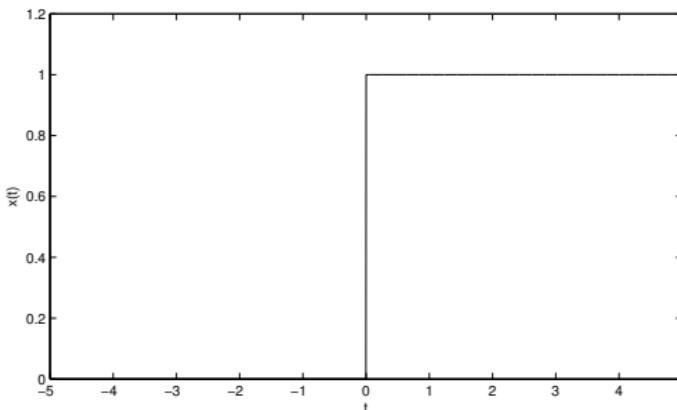
# Unit step function/Signal

## Definition

A **unit step function(signal)** is defined as

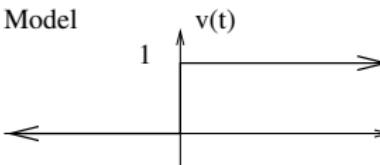
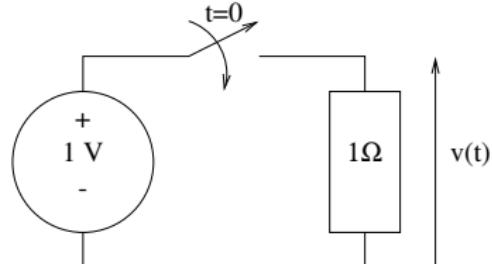
$$u(t) \triangleq \begin{cases} 1, & t > 0 \text{ or } t \geq 0 \\ 0, & t < 0 \end{cases}$$

The value at  $t = 0$  is arbitrary and unimportant! Reasonable choices are 0, 1, and  $\frac{1}{2}$ ; any will do.

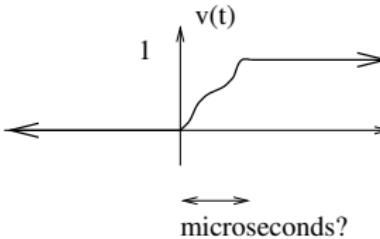


# Modeling a switch

The unit step is a useful model for a **switch**.



Reality (zoomed)

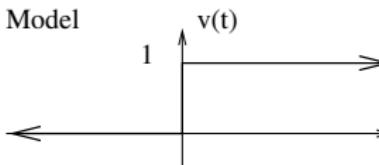
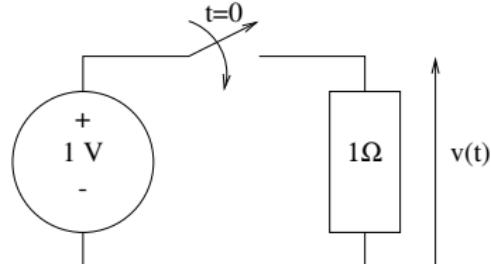


## Question

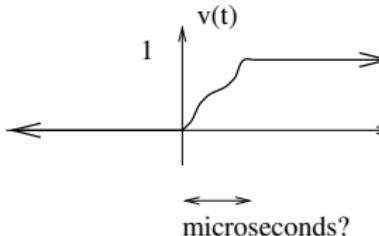
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## Question

- For a real switch in above is the voltage exactly a step function?
- Does the final voltage exactly equal 1 Volt?

# Modeling a switch (2)

## Solution

- *No, as the first atoms of the switch contacts begin to “touch”, there will be a small current flow. As more atoms touch, more current will flow. Eventually the contacts will “touch completely”. and the system will settle to steady state.*
- *No, because of internal resistance of source and resistances of switch itself and wires.*

## Modeling a switch (3)

- *But all of these effects are very small, so generally we can ignore them and the step function model  $v(t) = u(t)$  is very reasonable for most purposes.*
- *In some high speed applications the switching time is important, so a more accurate model is necessary.*
- *Any real system can have no discontinuities of the type exhibited by the unit step function. There is always a small transition.*
- *It is OK to ignore this transition as long as the transition time is small relative to other time constants in the system being studied.*

# Simplifying notation

The step function is useful for simplifying notation.

## Example

The following step function “switches on” the signal from a guitar string plucked at time  $t = 2$ .

$$x(t) = \begin{cases} e^{-t} \sin(5t), & t > 2 \\ 0, & \text{otherwise} \end{cases} = e^{-t} \sin(5t) u(t - 2).$$

The first notation is messy, the second way is neat, and hides the braces within the definition of  $u(t - 2)$ .

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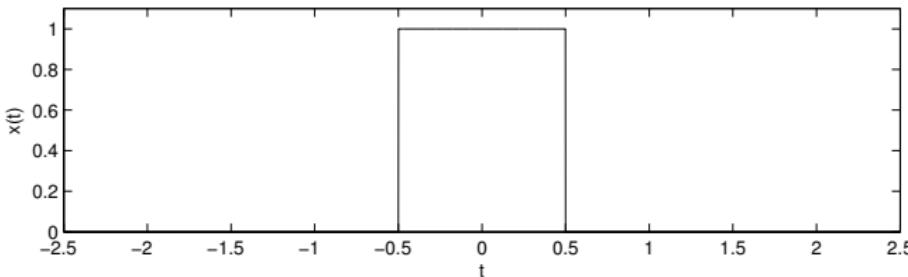
# Rect(angle) function

## Definition

A **rect(angle)** function is defined as

$$\text{rect}(t) \triangleq \begin{cases} 1, & -1/2 < t < 1/2 \\ 0, & \text{otherwise} \end{cases}$$

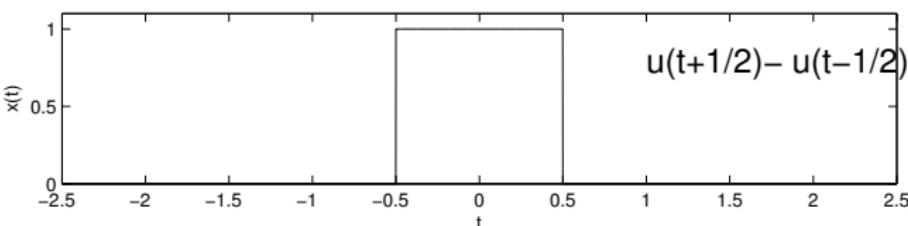
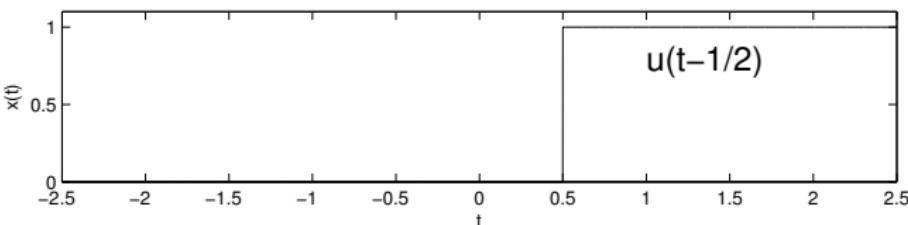
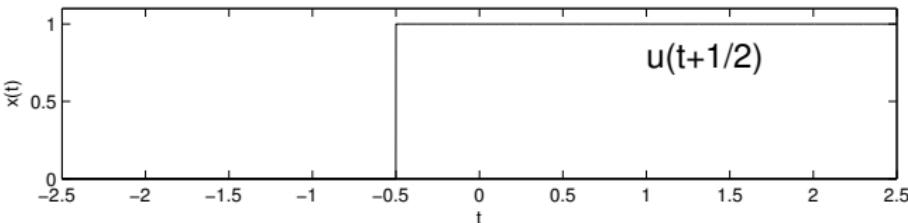
Centered at zero with unit width and unit height.



# Rect function and step functions (1)

Can be represented using step functions:

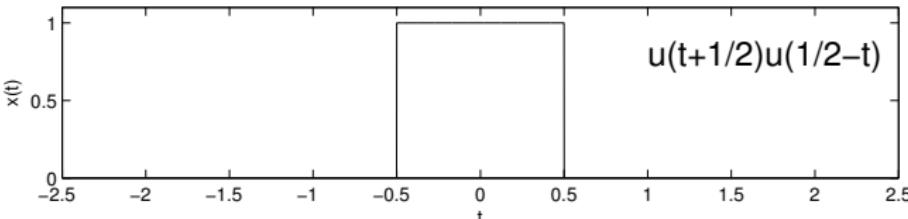
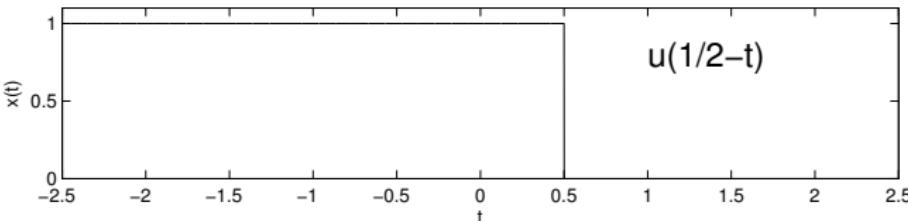
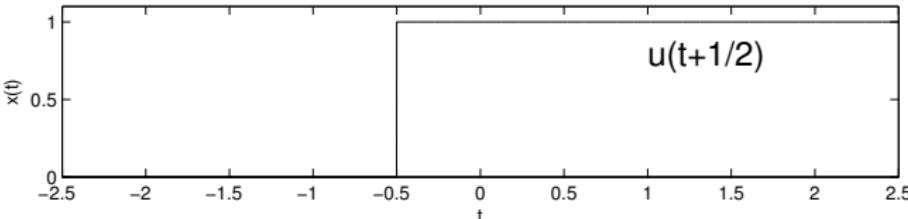
$$\text{rect}(t) = u(t + 1/2) - u(t - 1/2)$$



# Rect function and step functions (2)

Can be represented using step functions:

$$\text{rect}(t) = u(t + 1/2)u(1/2 - t)$$

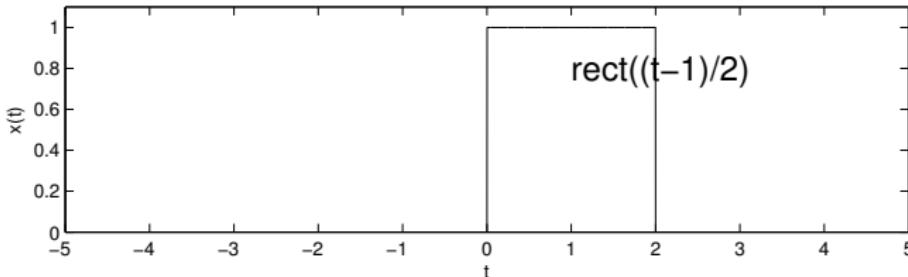


# Transformed rect functions

Time-scaled and time-shifted rect function

$$\begin{aligned}\text{rect}\left(\frac{t - t_0}{T}\right) &= \begin{cases} 1, & -1/2 < \frac{t-t_0}{T} < 1/2 \\ 0, & \text{otherwise} \end{cases} \\ &= \begin{cases} 1, & t_0 - T/2 < t < t_0 + T/2 \\ 0, & \text{otherwise.} \end{cases}\end{aligned}$$

Centered at  $t_0$  with width  $T$

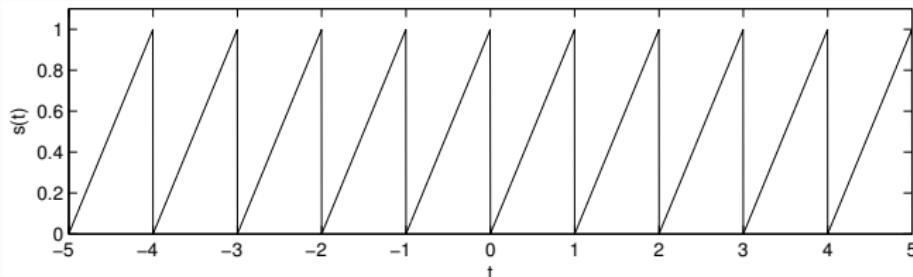


# Rect function: example

Useful for “switching on and off” other functions, or for “extracting” part of a signal, such as one period of a periodic signal.

## Example

Find mathematical expression for a sawtooth signal  $s(t)$ .

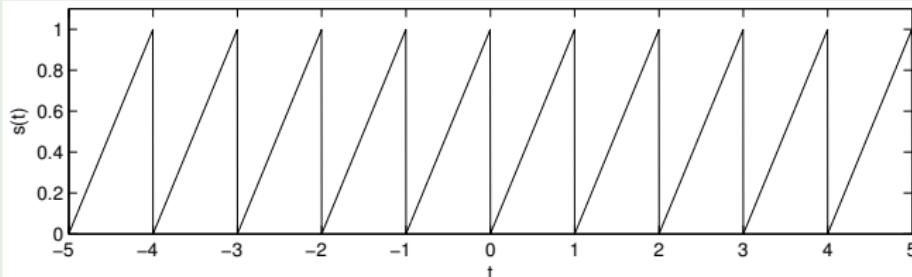


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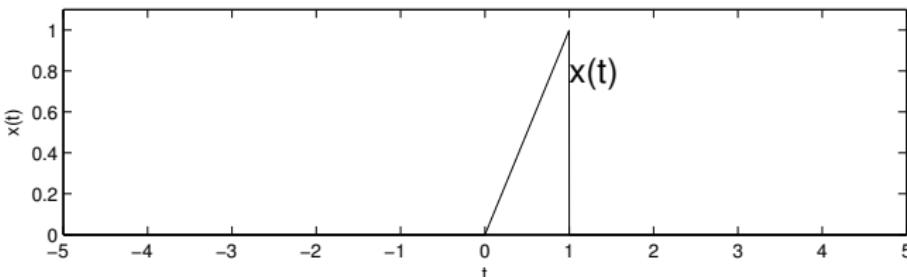
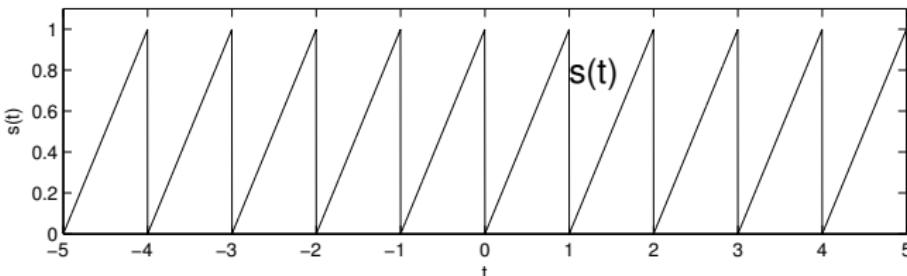
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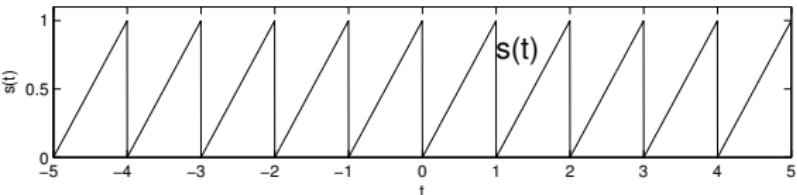
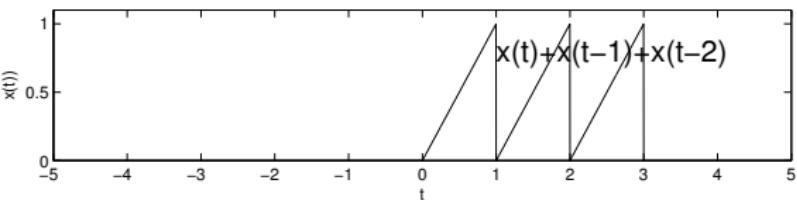
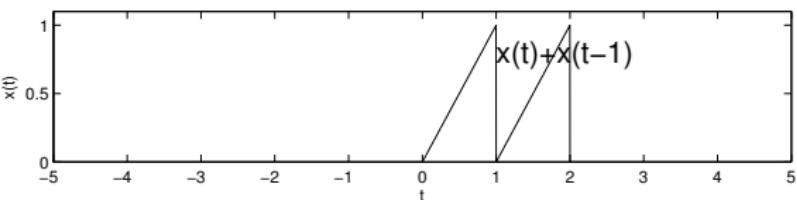
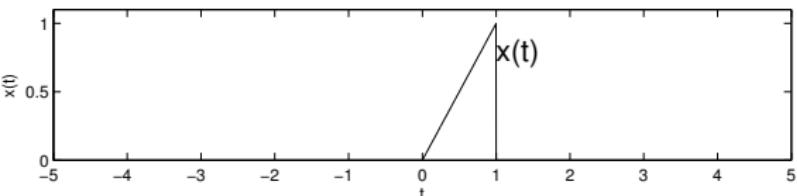
# Rect function: solution (1)

Let  $x(t)$  be the signal that is nonzero only over one period of  $s(t)$  (one “tooth”).

$$x(t) = \begin{cases} t, & 0 < t < 1 \\ 0, & \text{otherwise} \end{cases} = t \operatorname{rect}(t - 1/2).$$



# Rect function: solution (2)



## Rect function: solution (3)

$$s(t) = x(t) + x(t - 1) + x(t - 2) + \dots$$

$$s(t) = \sum_{k=-\infty}^{\infty} x(t - k) = \sum_{k=-\infty}^{\infty} (t - k) \operatorname{rect}(t - k - 1/2), \quad k \in \mathbb{Z}.$$

*Very convenient closed-form representation. This form will be particularly useful for analyzing periodic signals using the Fourier series.*

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- It is another **mathematical idealization** that cannot occur in nature (like the unit step function), but is nevertheless useful for modeling certain phenomena, just as the step function is a useful idealization of a switch.
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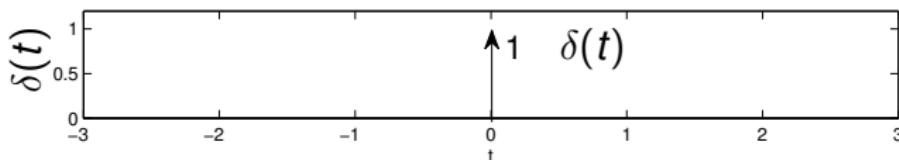
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Graphical representation using upward arrow, labeled with area (called **weight**). (see text p.34 for scaled impulse.)

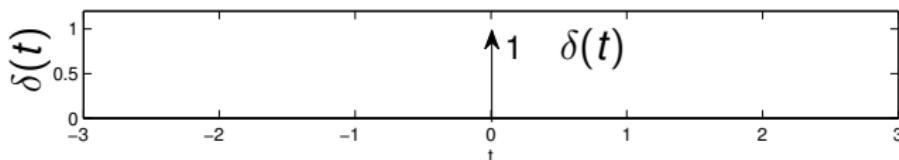
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# Minor Properties

## Property

- ① *unit area property*  $\int_{-\infty}^{\infty} \delta(t - t_0) dt = 1$  for any  $t_0$
- ② *scaling property*  $\delta(at + b) = \frac{1}{|a|}\delta(t + b/a)$  for  $a \neq 0$ .
- ③ *symmetry property*  $\delta(t) = \delta(-t)$
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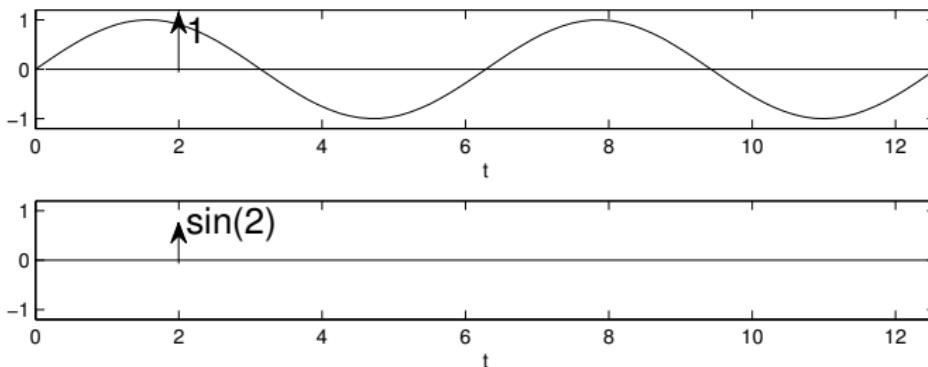
- ① *unit area property*  $\int_{-\infty}^{\infty} \delta(t - t_0) dt = 1$  for any  $t_0$
- ② *scaling property*  $\delta(at + b) = \frac{1}{|a|} \delta(t + b/a)$  for  $a \neq 0$ .
- ③ *symmetry property*  $\delta(t) = \delta(-t)$
- ④ *support property*  $\delta(t - t_0) = 0$  for  $t \neq t_0$
- ⑤ *relationships with unit step function:*  $\delta(t) = \frac{d}{dt} u(t),$   
 $u(t) = \int_{-\infty}^t \delta(\tau) d\tau$

# Major properties: Sampling property

## Property

**Sampling property** holds when  $x(t)$  is continuous at  $t_0$ :

$$x(t)\delta(t - t_0) = x(t_0)\delta(t - t_0).$$

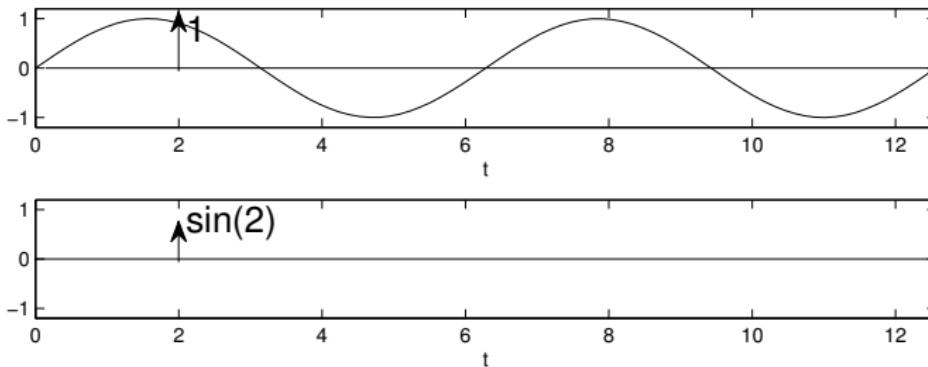


# Major properties: Sifting property

## Property

**Sifting property** holds when  $x(t)$  is continuous at  $t_0$ :

$$\int_{-\infty}^{\infty} x(t) \delta(t - t_0) dt = x(t_0).$$



# Algebraic property

## Property

*Algebraic property*

$$x\delta(t) = 0$$

# Scaling property

## Example

Show that  $\delta(at) = \frac{1}{|a|}\delta(t)$  for  $a \neq 0$ .

## Property

Let  $g_1(t)$  and  $g_2(t)$  be generalized functions. Then the **equivalence property** states that  $g_1(t) = g_2(t)$  iff

$$\int_{-\infty}^{\infty} \phi(t)g_1(t) dt = \int_{-\infty}^{\infty} \phi(t)g_2(t) dt$$

for all suitably defined testing function  $\phi(t)$ .

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We prove this in two steps, for  $a > 0$  and then for  $a < 0$ .

- When  $a > 0$ ,

$$\begin{aligned} \int_{-\infty}^{\infty} \phi(t) \delta(at) dt &= \int_{-\infty}^{\infty} \phi(t) \delta(|a|t) dt \\ &= \int_{-\infty}^{\infty} \phi\left(\frac{1}{|a|}y\right) \delta(y) \frac{1}{|a|} dy \quad (y = |a|t \implies dy = |a|dt) \\ &= \frac{1}{|a|} \phi\left(\frac{1}{|a|}y\right) \Big|_{y=0} = \boxed{\frac{1}{|a|}\phi(0)} \quad (\text{sifting property}) \end{aligned}$$

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Thus, for any  $a$

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# Ideal and practical impulse function

- *Mathematically, one can use the above properties to solve all the problems in this course.*
- *But physically we would like more insight, so we consider the “ideal” unit impulse function as a type of limit of “practical” impulse functions, just like the unit step function is a limit of practical almost-step functions.*

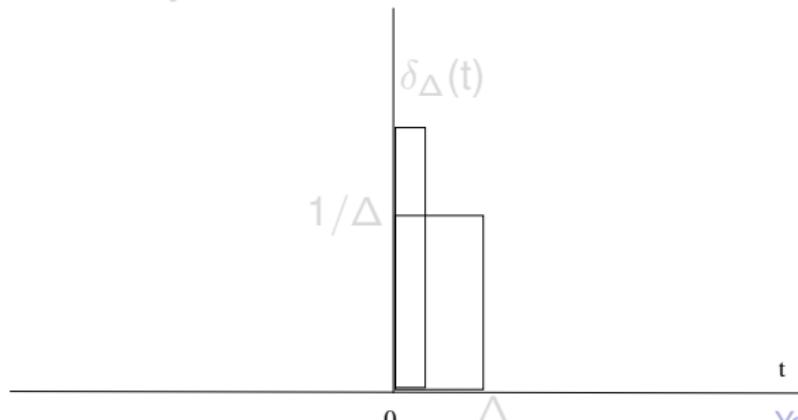
# Practical impulse function

## Definition

**Practical impulse function**, defined for any  $\Delta > 0$ :

$$\delta_\Delta(t) \triangleq \begin{cases} 1/\Delta, & 0 < t < \Delta \\ 0, & \text{otherwise.} \end{cases}$$

Note that area is **unity**, width approaches zero as  $\Delta \rightarrow 0$ ; height approaches infinity as  $\Delta \rightarrow 0$ .



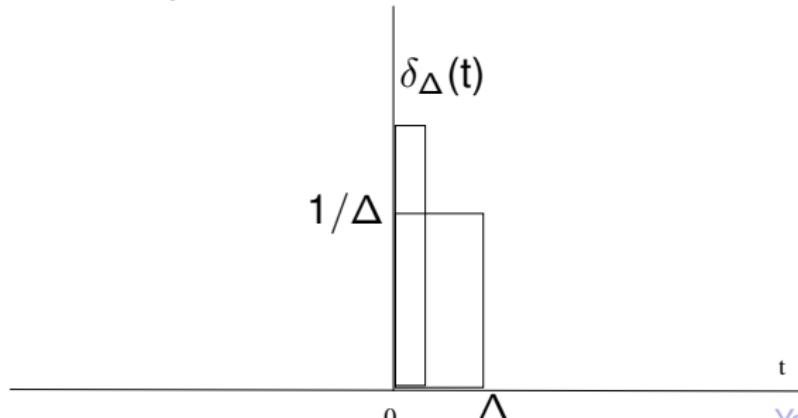
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# Practical impulse function: example

## Example

drumstick striking a drum (applied force vs time)

## Example

metal hammer tapping a pendulum (applied force vs time)

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It is tempting to try to write

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but the limit is **not well defined mathematically**. Nevertheless, this is the intuition.

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Instead we “define”  $\delta(t)$  in terms of its **properties**, making sure that the properties are consistent with the above “limit”. Such objects are called **generalized functions** in mathematics.

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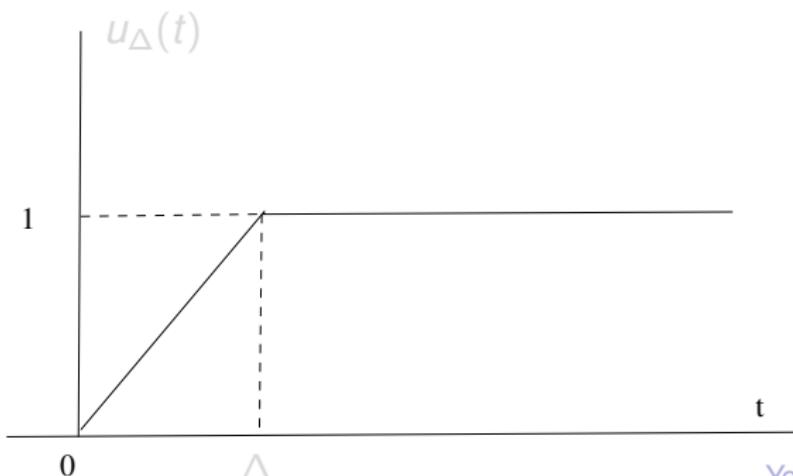
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# Relationship to unit step function (1)

Explanation of  $\delta(t) = \frac{d}{dt}u(t)$  using limiting step function.

Define a practical almost-step function as

$$u_{\Delta}(t) \triangleq \begin{cases} 0, & t \leq 0 \\ t/\Delta, & 0 < t < \Delta \\ 1, & t \geq \Delta \end{cases}$$



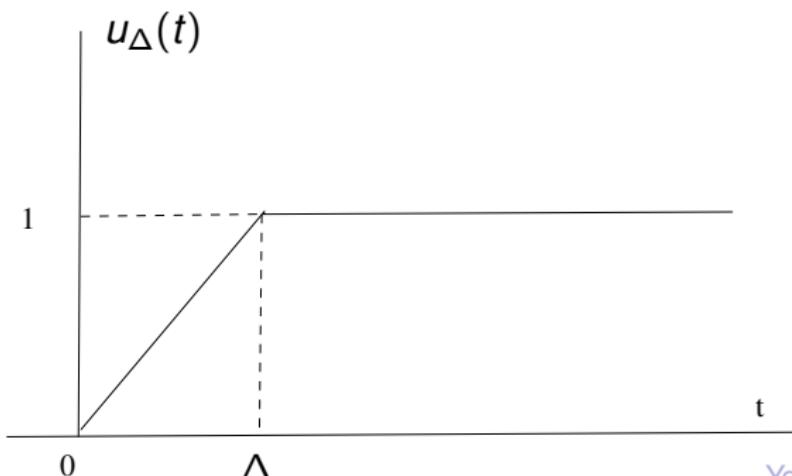
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## Relationship to unit step function (2)

$$\frac{d}{dt} u_{\Delta}(t) = \begin{cases} 1/\Delta, & 0 < t < \Delta \\ 0, & \text{otherwise} \end{cases} = \delta_{\Delta}(t)$$

Then since  $u(t) = \lim_{\Delta \rightarrow 0} u_{\Delta}(t)$ , by taking the limit of both sides we have that

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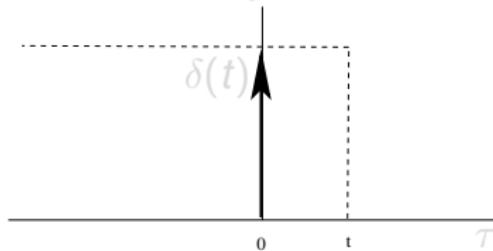
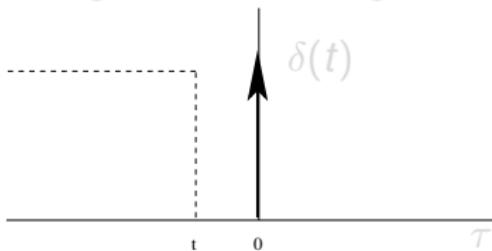
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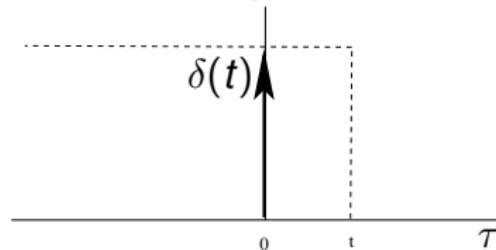
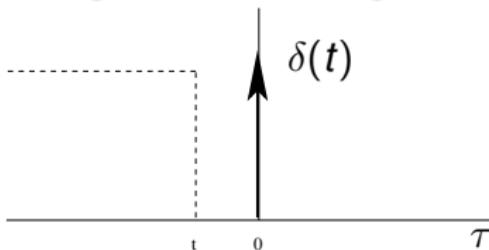
- For any  $t < 0$ , the range of integration over  $(-\infty, t)$  will not cover zero, so the integral is simply zero.
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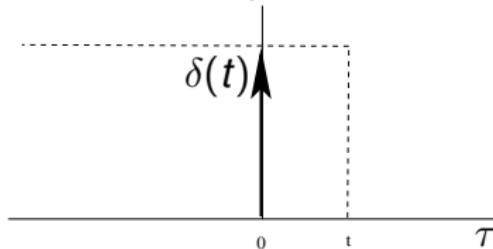
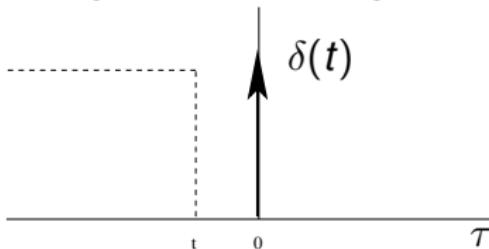
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# Example (1)

## Example

- By Newton's laws, velocity is the time-integral of acceleration.
- When a hammer taps a stationary pendulum, the pendulum (almost) instantaneously changes from being stationary to moving with some velocity that is related to how "hard" the hammer taps the pendulum.
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### Solution

$$x(t) = 2 \operatorname{rect}(t/2 - 3) = 2 \operatorname{rect}((t - 6)/2) = 2u(t - 5) - 2u(t - 7)$$

*turns on at  $t = 5$  and turns off at  $t = 7$ .*

$$y(t) = \frac{d}{dt}x(t) = 2\delta(t - 5) - 2\delta(t - 7)$$

## Example (3)

$$\begin{aligned}x(t) &= 2 \operatorname{rect}(t/2 - 3) = 2 \operatorname{rect}((t - 6)/2) \\&= 2u\left(\frac{t-6}{2} + \frac{1}{2}\right) - 2u\left(\frac{t-6}{2} - \frac{1}{2}\right) \\&= 2u\left(\frac{t-5}{2}\right) - 2u\left(\frac{t-7}{2}\right)\end{aligned}$$

directly plug in  $\operatorname{rect}(t) = u(t + 1/2) - u(t - 1/2)$ .

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# Time-scaled unit step function

## Question

$$u\left(\frac{t-t_0}{w}\right) = u(t - t_0) \text{ where } w > 0$$

# Time-scaled unit step function

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$u\left(\frac{t-t_0}{w}\right) = u(t - t_0)$  where  $w > 0$  Yes.

- $u\left(\frac{t-t_0}{w}\right)$  turns on at  $t_0$ .

$$\frac{t - t_0}{w} \geq 0 \implies t \geq t_0$$

- $u(t - t_0)$  turns on at  $t_0$ .

$$t - t_0 \geq 0 \implies t \geq t_0$$

# Outline

## 1. Signals & Systems (Fundamentals)

- Overview
- Signal and System Definition
- Classification of Signals
- Signal Notation
- Transformations of CT signals
  - Time transformations
  - Amplitude transformations
  - More signal operations
  - Operations with two signals
- Signal Characteristics
  - Periodic/aperiodic signals
  - Even and odd signals
  - Energy and power signals
- Exponential signals
- Singularity functions (1.4)
  - Unit step signal
  - Rect(angle) function
  - Unit impulse function  $\delta(t)$ (1.4.2, 2.5)
- Continuous-time systems

# Continuous-time systems

## Definition

A **continuous-time system** is a device or process that, according to some well-defined rule, transforms one CT signal called the **input signal** or **excitation** into another CT signal called the **output signal** or **response**.

The input signal  $x(t)$  is **transformed** by the system into a signal  $y(t)$ , which we express mathematically as

$$y(\cdot) = \mathcal{T}[x(\cdot)] \quad \text{or} \quad y(t) = \mathcal{T}[x(\cdot)](t) \quad \text{or} \quad x(\cdot) \xrightarrow{\mathcal{T}} y(\cdot).$$

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*The notation  $y(t) = \mathcal{T}[x(t)]$  is mathematically vague. Why?*

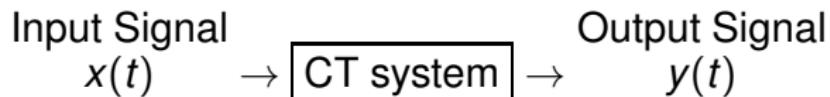
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*In general  $y(t)$  is a function of the entire signal  $x(\cdot)$  not just the single time point  $x(t)$  as illustrated by auto example below.*

# Diagram

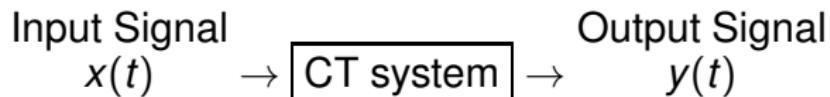


The arrows in this diagram are not necessarily wires! They represent whatever medium transports the signal from one part of the system to another part.

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At the systems level, we are less interested in the details of the implementation than in the mathematical relationships and system properties.

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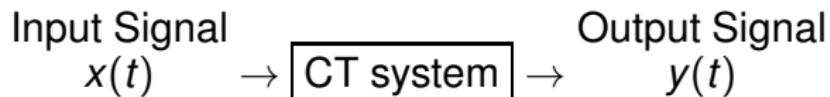


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# Example (1)

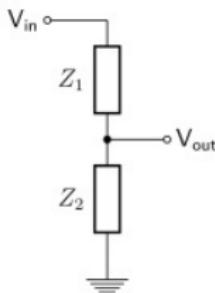
## Example

voice (acoustic pressure) → **microphone** → electrical current

## Example

voltage divider

([https://en.wikipedia.org/wiki/Voltage\\_divider](https://en.wikipedia.org/wiki/Voltage_divider))



For identical resistors, the output is  $y(t) = \frac{1}{2}x(t)$ . Called a **static** system.

# Example (1)

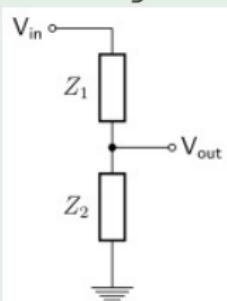
## Example

voice (acoustic pressure) → **microphone** → electrical current

## Example

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## Example (2)

### Example

accelerator pedal position → **engine/car** → car velocity

- Input signal induces a response of the system.
- $x(t) = \frac{1}{2} + \frac{1}{2}u(t - 3)$  (one pushes the gas pedal to the floor)
- $y(t) = 40 + 20(1 - e^{-t/\tau})$  (rise time or transient response)
- $y(t)$  is not solely a function of  $x(t)$  at time  $t$ , but also a function of previous input signal values and the present and past state of the system.
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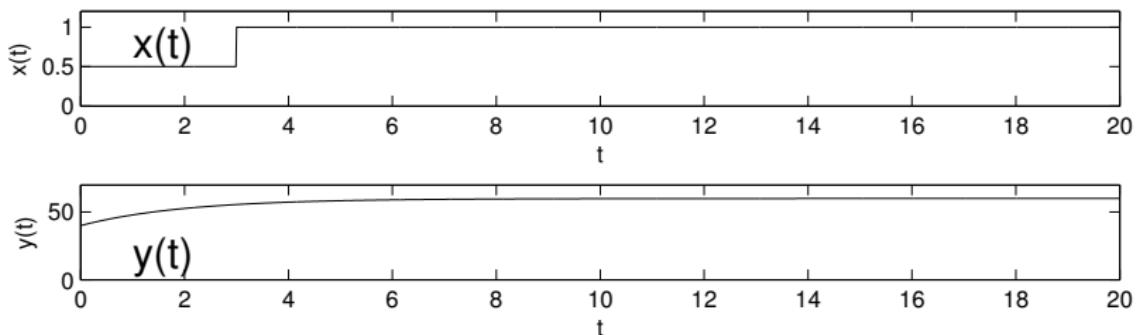
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# Example (3)

$$x(t) = \frac{1}{2} + \frac{1}{2}u(t - 3)$$

$$y(t) = 40 + 20(1 - e^{-t/\tau}), \quad \tau = 2$$



# Outline

## 1. Signals & Systems (Fundamentals)

- Overview
- Signal and System Definition
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- Transformations of CT signals
  - Time transformations
  - Amplitude transformations
  - More signal operations
  - Operations with two signals
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  - Periodic/aperiodic signals
  - Even and odd signals
  - Energy and power signals
- Exponential signals
- Singularity functions (1.4)
  - Unit step signal
  - Rect(angle) function
  - Unit impulse function  $\delta(t)$ (1.4.2, 2.5)
- Continuous-time systems

# Input-output description of systems (1)

Pictures/diagrams are a starting point, but for quantitative analysis every system must have a **input-output relationship**.

## Definition

**Input-output relationship** is a mathematical expression that precisely defines how the output signal is related to the input signal.

## Example

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# Moving average

## Example

**moving average filter**

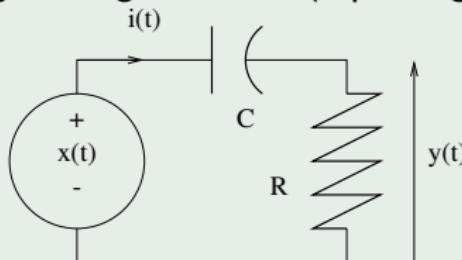
$$y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau.$$



# Input-output description of systems (2)

## Example

RC circuit driven by voltage source (input signal).



$$\frac{1}{CR}y(t) + \frac{d}{dt}y(t) = \frac{d}{dt}x(t).$$

- This input-output relation is not of the form  $y(t) = \text{some\_function}[x(t)]$ .
- When combined with an appropriate initial condition (such as 0 charge on the capacitor at time  $t = 0$ ) one can solve the diffeq to determine  $y(t)$  for any  $x(t)$  (**later lectures.**)

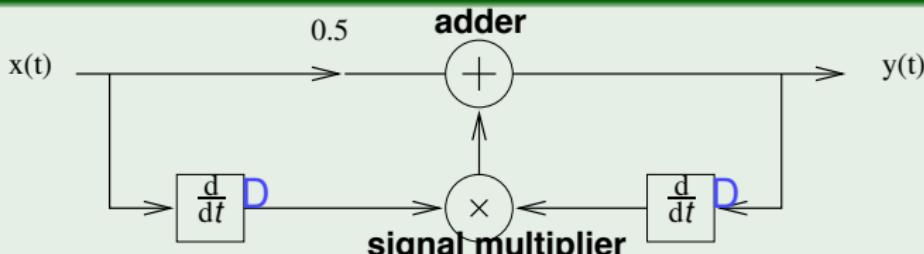
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# Block diagram representation of CT systems

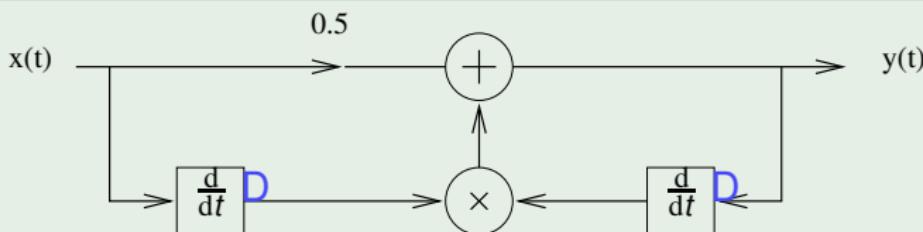
## Example



The lower-right part is called a **feedback connection**.

# Block diagram representation of CT systems

## Example



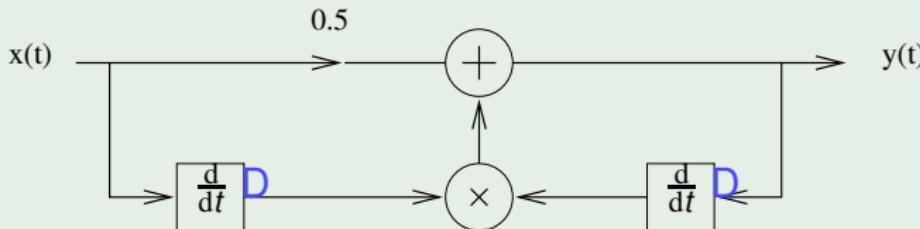
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## Basic elements:

- adder (see text Fig. 2.29 on P. 126)
- constant multiplier (amplifier) (see text Fig. 2.29 on P. 126)
- signal multiplier
- differentiator (see text Fig. 2.29 on P. 126)
- integrator (see text Fig. 2.31 & 2.32 on P. 127)

# Block diagram representation of CT systems

## Example



The lower-right part is called a **feedback connection**.

Input-output relationship defined by the diagram:

$$y(t) = 0.5x(t) + \left( \frac{d}{dt}x(t) \right) \cdot \left( \frac{d}{dt}y(t) \right) \cdot D.$$

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# Interconnection of systems

## ① Series connection

$$x(t) \rightarrow \boxed{\mathcal{T}_1} \rightarrow \boxed{\mathcal{T}_2} \rightarrow y(t)$$

Mathematically:  $y(t) = \mathcal{T}_2[\mathcal{T}_1[x(t)]]$ .

## ② Parallel connection

(See text Fig. 1.42(b) on P. 42 )

Mathematically:  $y(t) = \mathcal{T}_1[x(t)] + \mathcal{T}_2[x(t)]$ .

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$x(t) \rightarrow \boxed{\text{amplifier, gain}=5} \rightarrow \boxed{\text{differentiator}} \rightarrow y(t)$

$$y(t) = 5 \frac{d}{dt} x(t)$$

In this example **the order of interconnection is irrelevant**. We will learn soon that this is because both subsystems are **linear**.

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- **Continuous-time systems**
- Summary

# Classification of CT systems

Two general aspects to categorize:

- Amplitude properties
  - A-1 linearity (1.6.6)
  - A-2 stability (1.6.4)
  - A-3 invertibility (1.6.2)
- Time properties
  - T-1 causality (1.6.3)
  - T-2 memory (1.6.1)
  - T-3 time-invariance (1.6.5)

*Skill: Determining classifications of a given CT system*

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**Skill: Determining classifications of a given CT system**

# A-1 Linearity (1)

## Definition

A system  $\mathcal{T}$  is **linear** iff

$$\mathcal{T}[a_1x_1(t) + a_2x_2(t)] = a_1\mathcal{T}[x_1(t)] + a_2\mathcal{T}[x_2(t)]$$

for **any signals**  $x_1(t), x_2(t)$  and **any (even complex) constants**  $a_1$  and  $a_2$ . Otherwise the system is called **nonlinear**.

Response to a weighted sum of input signals is the weighted sum of the individual responses. (*Picture*)

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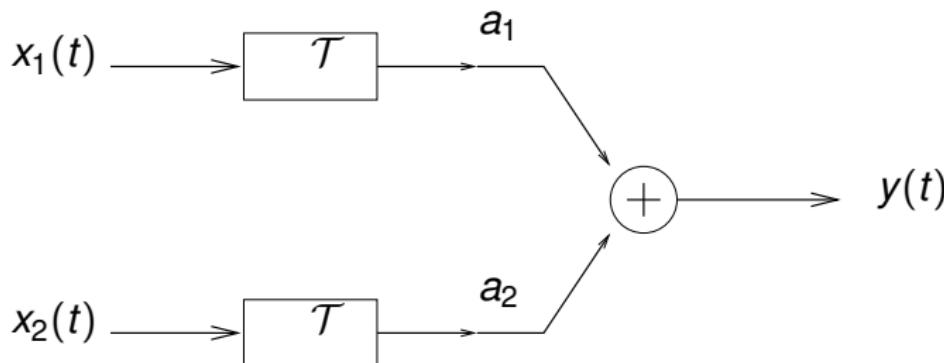
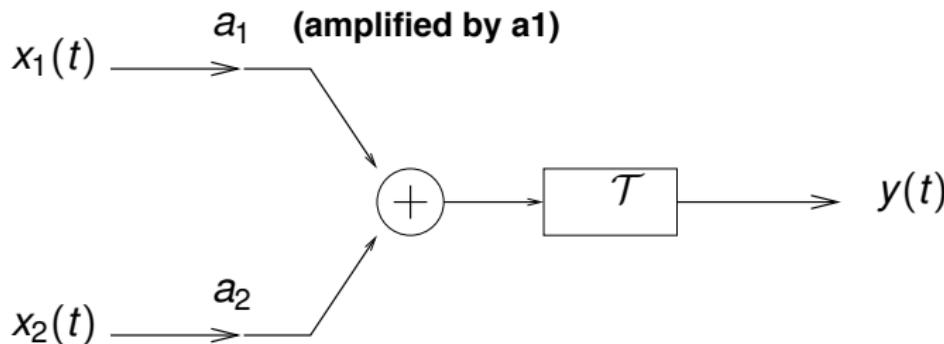
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Response to a weighted sum of input signals is the weighted sum of the individual responses. (**Picture**)

# A-1 Linearity (2)



**The order doesn't matter. The system  $T$  is linear.**

# A-1 Linearity (3)

## Question

*We will focus on linear systems. Why?*

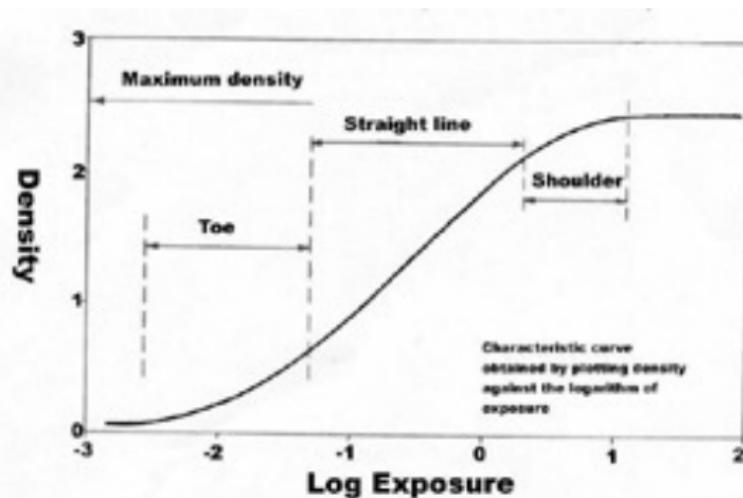
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## Question

*We will focus on linear systems. Why?*

- ① *The class of linear systems is easier to analyze.*
- ② *Often linearity is desirable - avoids distortions  
(Example: amplifiers, audio mixers (superposition!)).*
- ③ *Many nonlinear systems are approximately linear  
(Example: the characteristic curve of a photographic film  
relating the optical density of the film to the logarithm of the  
incident exposure. (Picture)).*

## A-1 Linearity (4)



Real systems are never perfectly linear, but often they are approximately linear over an appropriate operating range.

# Two important special cases of linearity property (1)

## Property

*scaling property or homogeneity property:*

$$\mathcal{T}[ax(t)] = a\mathcal{T}[x(t)]$$

Note that from  $a = 0$  we see that zero input signal implies zero output signal for a linear system.

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*additivity property:*

$$\mathcal{T}[x_1(t) + x_2(t)] = \mathcal{T}[x_1(t)] + \mathcal{T}[x_2(t)]$$

Using proof-by-induction, one can easily extend this property to the general superposition property

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*general superposition property*

$$\mathcal{T}\left[\sum_{k=1}^K x_k(t)\right] = \sum_{k=1}^K \mathcal{T}[x_k(t)].$$

In words: the response of a linear system to the sum of several signals is the sum of the response to each of the signals.

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- In general superposition need not hold for **infinite sums**; additional continuity assumptions are required.
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# Determining a system is linear or nonlinear

## Skill: Determining a system is linear or nonlinear.

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# Example (1)

## Example

Prove that the integrator is a linear system, where  
 $y(t) = \int_{-\infty}^t x(\tau) d\tau.$

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①  $y_1(t) = \int_{-\infty}^t x_1(\tau) d\tau$

②  $y_2(t) = \int_{-\infty}^t x_2(\tau) d\tau.$

③ If the input is  $x(t) = a_1x_1(t) + a_2x_2(t)$ , then the output is

$$\begin{aligned} y(t) &= \int_{-\infty}^t x(\tau) d\tau = \int_{-\infty}^t [a_1x_1(\tau) + a_2x_2(\tau)] d\tau \\ &= a_1 \int_{-\infty}^t x_1(\tau) d\tau + a_2 \int_{-\infty}^t x_2(\tau) d\tau = a_1y_1(t) + a_2y_2(t). \end{aligned}$$

④ Since this holds for all  $t$ , for all input signals  $x_1(t)$  and  $x_2(t)$ , and for any constants  $a_1$  and  $a_2$ , the integrator is linear.

## Example (2)

### Example

Determine whether linearity holds for  $y(t) = \int_{-\infty}^t x^3(\tau) d\tau$ .

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Determine whether linearity holds for  $y(t) = \int_{-\infty}^t x^3(\tau) d\tau$ .

*Direct method:*

①  $y_1(t) = \int_{-\infty}^t x_1^3(\tau) d\tau$

②  $y_2(t) = \int_{-\infty}^t x_2^3(\tau) d\tau$ .

③ If the input is  $x(t) = a_1x_1(t) + a_2x_2(t)$ , then the output is

$$\begin{aligned}y(t) &= \int_{-\infty}^t x(\tau) d\tau = \int_{-\infty}^t [a_1x_1(\tau) + a_2x_2(\tau)]^3 d\tau \\&\neq a_1y_1(t) + a_2y_2(t).\end{aligned}$$

## Example (2)

### Example

Determine whether linearity holds for  $y(t) = \int_{-\infty}^t x^3(\tau) d\tau$ .

*Counter-example:* 

- if  $x(t) = u(t)$ , then  $y(t) = tu(t)$ , which is called the **unit ramp signal**.
- But if  $x(t) = 2u(t)$ , then  $y(t) = 8tu(t)$ , so doubling the input did not double the output.
- Thus the scaling property is violated, so this system is nonlinear.

# Example (3)

## Example

Are the following systems linear?

- $y(t) = x^3(t)$
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Are the following systems linear?

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- To show that a signal is *nonlinear*, all that is needed is a counter-example to the scaling and additivity properties.
- The scaling property will usually (but not always!) suffice.

## Example (3)

### Example

Are the following systems linear?

- $y(t) = x^3(t)$
  - $y(t) = 2x(t) + 3$
- 
- Let  $x_1(t) = 1$ , a constant signal. Then  $y_1(t) = 1$ .
  - Now suppose the input is  $x(t) = 2x_1(t) = 2$ , then the output is  $y(t) = 2^3 = 8 \neq 2y_1(t) = 2$
  - So the system is *nonlinear*.

# Example (3)



## Example

Are the following systems linear?

- $y(t) = x^3(t)$
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*nonlinear, a zero input yields the output  $y(t) = 3$ , which is nonzero.*

# Example (4)

## Example

Is  $y(t) = \text{Real}[x(t)]$  linear?

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Is  $y(t) = \text{Real}[x(t)]$  linear?

For a system to be called linear, we require the *superposition property* to hold even for complex signals and scaling constants.

## Example (4)

### Example

Is  $y(t) = \text{Real}[x(t)]$  linear?

$y(t) = \text{Real}[x(t)]$  satisfies additive property, but not scaling property for complex  $a$ . Thus it is *not linear*.

## A-2 Stability (1)

### Definition

A system is **bounded-input bounded-output (BIBO) stable** iff every bounded input produces a bounded output.

If  $\exists M_x$  s.t.  $|x(t)| \leq M_x < \infty \forall t$ , then there must exist an  $M_y$  s.t.  
 $|y(t)| \leq M_y < \infty \forall t$ .

Usually  $M_y$  will depend on  $M_x$ .

Otherwise the system is called **unstable**, and it is possible that a small input signal will make the output “blow up.”

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## A-2 Stability: example (1)

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Is the integrator system  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  BIBO stable?

## A-2 Stability: example (1)

### Example

Is the integrator system  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  BIBO stable?

Consider input signal  $x(t) = u(t)$ , which is bounded by  $M_x = 1$ .  
But  $y(t) = tu(t)$  blows up, so the integrator is an unstable system.

# Triangle inequality

The **triangle inequality** is sometimes useful for proving that a system is BIBO stable.

- $|a + b| \leq |a| + |b|$  (Easily proved by considering 4 cases where  $a$  and  $b$  are positive or negative.)
- $|\sum_n a_n| \leq \sum_n |a_n|$
- $\left| \int f(t) dt \right| \leq \int |f(t)| dt$

## A-2 Stability: example (2)

### Example

Is the moving average  $y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau$  for  $T > 0$  system BIBO stable?

## A-2 Stability: example (2)

### Example

Is the moving average  $y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau$  for  $T > 0$  system BIBO stable?

Suppose  $|x(t)| \leq M_x < \infty \forall t$ , so  $x(t)$  is a bounded input. Then by the triangle inequality:

$$|y(t)| \leq \frac{1}{T} \int_{t-T}^t |x(\tau)| d\tau \leq \frac{1}{T} \int_{t-T}^t M_x d\tau = M_x,$$

so the output signal is also bounded for a bounded input. Thus the moving average system is **BIBO stable**.

## A-2 Stability: example (3)

### Example

Is  $y(t) = x^5(t)$  BIBO stable?

## A-2 Stability: example (3)

### Example

Is  $y(t) = x^5(t)$  BIBO stable?

Suppose  $|x(t)| \leq M_x < \infty$ . Then  $|y(t)| = |x^5(t)| \leq M_x^5 < \infty$ . So this system is *BIBO stable*.

## A-3 Invertibility

### Definition

A system  $\mathcal{T}$  is called **invertible** iff each (possible) output signal is the response to only one input signal. Otherwise  $\mathcal{T}$  is not **invertible**.

### Property

If a system  $\mathcal{T}$  is invertible, then there exists a system  $\mathcal{T}^{-1}$  such that

$$x(t) \rightarrow [\mathcal{T}] \rightarrow y(t) \rightarrow [\mathcal{T}^{-1}] \rightarrow z(t) = x(t).$$

Mathematically:

$$\mathcal{T}^{-1}[\mathcal{T}[x(t)]] = x(t)$$

Design of  $\mathcal{T}^{-1}$  is important in many signal processing applications.

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## A-3 Invertibility: example (1)

### Example

encryption/decryption for secure communication. Needs to be invertible for no loss of information.

### Example

digital speedometer

velocity → speed sensor → voltage

→ mathematical inverse of sensor law → velocity display.

We display the velocity, not the voltage, so there should be a one-to-one relationship between the two.

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## A-3 Invertibility: example (2)

### Example

Is the full-wave rectifier:  $y(t) = |x(t)|$  invertible?

## A-3 Invertibility: example (2)

### Example

Is the full-wave rectifier:  $y(t) = |x(t)|$  invertible?

*The distinct input signals  $x(t) = \sin t$  and  $x(t) = -\sin t$  yield the same output signal. So **not invertible**.*

## A-3 Invertibility: example (3)

### Example

Is the exponential-law device:  $y(t) = e^{x(t)}$  invertible?

## A-3 Invertibility: example (3)

### Example

Is the exponential-law device:  $y(t) = e^{x(t)}$  invertible?

*Invertible* (for real input signals anyway). *Inverse system is  $x(t) = \log y(t)$ , a log-law device.*

## A-3 Invertibility: example (4)

### Example

Is the ideal amplifier:  $y(t) = 2x(t)$  invertible?

## A-3 Invertibility: example (4)

### Example

Is the ideal amplifier:  $y(t) = 2x(t)$  invertible?

*Invertible.* *Inverse system is*  $x(t) = \frac{1}{2}y(t)$ .

# T-1 Causal systems

## Definition

For a **causal** system, the output  $y(t)$  at any time  $t$  depends **only** on the “present” and (possibly) “past” inputs i.e. on  $x(t)$  and on various  $x(t_0)$  for  $t_0 \leq t$  only, but **not** on future inputs.  
Otherwise **noncausal** system.

Causality is necessary for **real-time implementation**. Noncausal systems arise primarily when  $t$  is some other variable than time, such as space.

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Is the integrator:  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  causal?

## Solution

*It only depends on values of  $x$  for  $\tau \leq t$ , so causal.*

# T-1 Causal systems: example (2)

## Example

Is the symmetric moving average:  $y(t) = \frac{1}{2T} \int_{t-T}^{t+T} x(\tau) d\tau$  causal? (Useful for image processing.)

# T-1 Causal systems: example (2)

## Example

Is the symmetric moving average:  $y(t) = \frac{1}{2T} \int_{t-T}^{t+T} x(\tau) d\tau$  causal? (Useful for image processing.)

## Solution

*noncausal. It depends on  $t + T \geq \tau > t$*

# T-2 Memory

## Definition

For a **static system** or **memoryless** system, the output  $y(t)$  depends only on the current input  $x(t)$ , not on previous or future values of the input signal.

Otherwise it is a **dynamic system** and must have memory.

## Example

- $y(t) = e^{x(t)} / \sqrt{|t+3|}$ .
- moving average  $y(t) = \frac{1}{T} \int_{-\infty}^T x(\tau) d\tau$

Dynamic systems are the interesting ones and will be our focus. (This time we take the more complicated choice!)

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- moving average  $y(t) = \frac{1}{T} \int_{-\infty}^T x(\tau) d\tau$  *dynamic*

**Dynamic systems** are the interesting ones and will be **our focus**. (This time we take the more complicated choice!)

# Memory vs. causality

## Question

- *Is a memoryless system necessarily causal?*
- *Is a dynamic system necessarily noncausal?*

# Memory vs. causality

## Question

- *Is a memoryless system necessarily causal?* Yes
- *Is a dynamic system necessarily noncausal?* No. Dynamic systems can be causal or noncausal.

# Memory vs. causality

## Question

- *Is a memoryless system necessarily causal?*
- *Is a dynamic system necessarily noncausal?*

*Memory is often associated with stored energy, such as the charge on a capacitor, or the kinetic energy of a moving object.*

# T-3 Time-invariance (1)

Systems whose input-output behavior does not change with time are called **time-invariant** will be our focus.

- “Easier” to analyze.
- Time-invariance is a desired property of many systems.

We will focus primarily, but not exclusively, on time-invariant systems.

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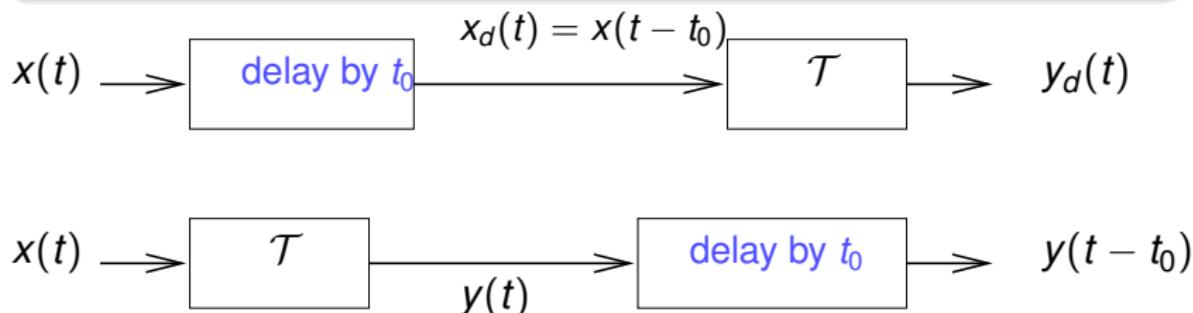
## T-3 Time-invariance (2)

### Definition

A system  $\mathcal{T}$  is called **time invariant** or **shift invariant** iff

$$x(t) \xrightarrow{\mathcal{T}} y(t) \text{ implies that } x(t - t_0) \xrightarrow{\mathcal{T}} y(t - t_0)$$

for **every** input signal  $x(t)$  and time shift  $t_0$ . Otherwise the system is called **time variant** or **shift variant**.



# Recipe for showing time-invariance

## Recipe for showing time-invariance

- ① Determine output signal  $y(t)$  due to a generic input signal  $x(t)$ .
- ② Determine the **delayed output** signal  $y(t - t_0)$ , by **replacing  $t$  with  $t - t_0$**  in  $y(t)$  expression.
- ③ Determine output signal  $y_d(t)$  due to a **delayed input** signal  $x_d(t) = x(t - t_0)$ .
- ④ If  $y_d(t) = y(t - t_0)$ , then system is time-invariant.

# Time-invariance: example (1)

## Example

Is the symmetric moving average filter

$$y(t) = \frac{1}{3}[x(t - 1) + x(t) + x(t + 1)]$$

# Time-invariance: example (1)

## Example

Is the symmetric moving average filter

$y(t) = \frac{1}{3}[x(t-1) + x(t) + x(t+1)]$  time-invariant? Yes

- ① Output due to  $x(t)$  is  $y(t) = \frac{1}{3}[x(t-1) + x(t) + x(t+1)]$
- ② Delayed output is

$$y(t - t_0) = \frac{1}{3}[x(t - t_0 - 1) + x(t - t_0) + x(t - t_0 + 1)]$$

- ③ Output due to delayed input  $x_d(t) = x(t - t_0)$  is

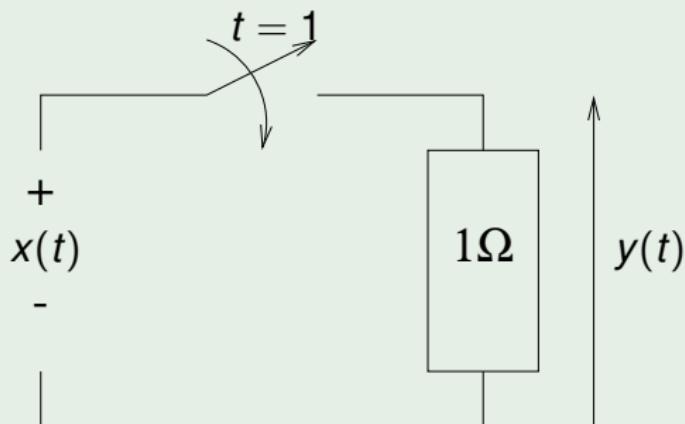
$$\begin{aligned} y_d(t) &= \frac{1}{3}[x_d(t-1) + x_d(t) + x_d(t+1)] \\ &= \frac{1}{3}[x(t-t_0-1) + x(t-t_0) + x(t-t_0+1)] \end{aligned}$$

- ④  $y_d(t) = y(t - t_0)$

# Time-invariance: example (2)

## Example

A switch that closes at  $t = 1$ .



- 1 How to represent input-output relationship mathematically?
- 2 Is it Time invariant? If no, find a counter-example.

# Time-invariance: solution (2)

## Solution

1  $y(t) = u(t - 1)x(t).$

2 No. Time-varying gain  $u(t - 1).$

3 A counter-example.

- If  $x(t) = \delta(t)$ , then  $y(t) = 0.$

- But if  $x_d(t) = \delta(t - 2)$ , then

$$y_d(t) = \delta(t - 2)u(t - 1) = \delta(t - 2) \neq y(t - 2).$$

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# Time-invariance: solution (3)

## Example

A switch that closes at  $t = 1$ :  $y(t) = u(t - 1)x(t)$ .

For certain  $t_0$ ,  $y_d(t) = y(t - t_0)$ .

if  $x_d(t) = \delta(t + 2)$ , then  $y_d(t) = \delta(t + 2)u(t - 1) = 0 = y(t + 2)$ .

---

A time-varying gain results in a time-varying system while systems with constant gains are time-invariant (e.g.,  $y(t) = 2x(t)$ ).

# Time-invariance: solution (3)

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# Time-invariance: example (3)

## Example

Is the modulator  $y(t) = \cos(\pi t) x(t)$  time-invariant?

# Time-invariance: example (3)

## Example

Is the modulator  $y(t) = \cos(\pi t) x(t)$  time-invariant?

**No.** Gain changes with time.

# Time-invariance: example (3)

## Example

Is the modulator  $y(t) = \cos(\pi t) x(t)$  time-invariant?

No. *How do we show lack of a property? Find a counter-example.*

## Time-invariance: example (3)

### Example

Is the modulator  $y(t) = \cos(\pi t) x(t)$  time-invariant?

No. How do we show lack of a property? Find a counter-example. A simple counter-example is all that is needed, and usually impulse functions will suffice .

# Time-invariance: example (3)

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Is the modulator  $y(t) = \cos(\pi t) x(t)$  time-invariant?

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$$\text{If } x(t) = \delta(t) \text{ then } y(t) = \delta(t)$$

$$\text{If } x_d(t) = \delta(t-1) \text{ then } y_d(t) = -\delta(t-1) \neq \delta(t-1) = y(t-1)$$

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If  $x_d(t) = \delta(t-1)$  then  $y_d(t) = -\delta(t-1) \neq \delta(t-1) = y(t-1)$

The modulator is a useful system! (AM radio)

# Time-invariance: example (4)

## Example

Is the amplified time reversal  $y(t) = 3x(-t)$  time-invariant?

# Time-invariance: example (4)

## Example

Is the amplified time reversal  $y(t) = 3x(-t)$  time-invariant?

## Solution

No. Any time shift in the input will be time-reversed.

$$\begin{aligned} x(t) = \delta(t) \xrightarrow{\tau} y(t) &= 3x(-t) = 3\delta(-t) \\ &= 3\delta(t) \text{ (by symmetry property of unit impulse,)} \end{aligned}$$

$$\begin{aligned} x_d(t) = \delta(t - 1) \xrightarrow{\tau} y_d(t) &= 3x_d(-t) = 3x_d(\textcolor{green}{z}) \quad (\text{where } z = -t) \\ &= 3\delta(\textcolor{green}{z} - 1) = 3\delta(-t - 1) = 3\delta(-(t + 1)) \\ &= 3\delta(t + 1) \neq 3\delta(t - 1) = y(t - 1), \end{aligned}$$

$y_d(t) \neq y(t - 1)$  for  $x_d(t) = \delta(t - 1)$ , so system is time varying.

# Time-invariance: example (4)

## Example

Is the amplified time reversal  $y(t) = 3x(-t)$  time-invariant?

*Any time shift in the input will be time-reversed. Similarly, time-scaled (compressed and expanded) systems are time-variant.* (text Example 1.16)

# Time-invariance: example (5)

## Example

Is the integrator  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  time-invariant?

# Time-invariance: example (5)

## Example

Is the integrator  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  time-invariant?

Yes.

①  $y(t - t_0) = \int_{-\infty}^{t-t_0} x(\tau) d\tau$

②  $x_d(t) = x(t - t_0)$

$$\begin{aligned}y_d(t) &= \int_{-\infty}^t x_d(\tau) d\tau = \int_{-\infty}^t x(\tau - t_0) d\tau \\&= \int_{-\infty}^{t-t_0} x(\tau') d\tau' \quad (\tau' = \tau - t_0) \\&= y(t - t_0)\end{aligned}$$

# Outline

## 1. Signals & Systems (Fundamentals)

- Overview
- Signal and System Definition
- Classification of Signals
- Signal Notation
- Transformations of CT signals
  - Time transformations
  - Amplitude transformations
  - More signal operations
  - Operations with two signals
- Signal Characteristics
  - Periodic/aperiodic signals
  - Even and odd signals
  - Energy and power signals
- Exponential signals
- Singularity functions (1.4)
  - Unit step signal
  - Rect(angle) function
  - Unit impulse function  $\delta(t)$ (1.4.2, 2.5)
- Continuous-time systems

# Summary (1)

- signal notation
- signal transformations
  - time transformations
  - amplitude transformations
  - differentiator / integrator systems
  - two-signal operations
- signal classes
  - even/odd signals
  - energy/power signals
  - periodic/aperiodic signals
  - exponential signals

# Summary (2)

- singularity functions
  - unit step / rect signals
  - unit impulse function
    - impulse function properties (sifting, sampling, scaling)
- CT systems
- block diagrams
- system classes
  - amplitude properties: linearity, stability, invertibility
  - time properties: causality, memory, time-invariance

# Key concepts/skills to study

## Key concepts/skills to study

- time transformations
- braces/plots to rects/steps
- running integral operation
- properties of  $\delta(t)$
- identifying signal properties
- identifying (all six) system properties