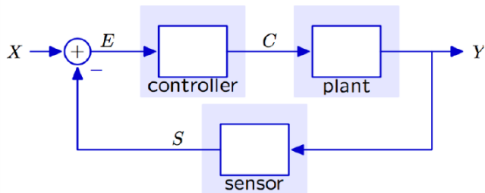


# VE216 Lecture 10

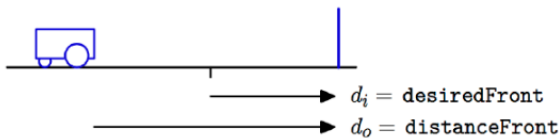
## Feedback and Control

### Structure of a Control Problem



- plant: the system to be controlled
- sensor: measures the output of plant
- controller: specify a command C to plant based on difference between X and S

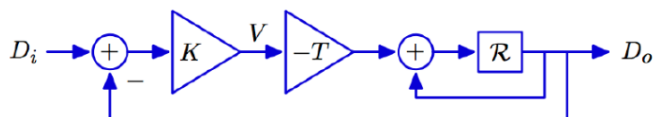
### Analysis of Wall Finder System



Controller:  $v[n] = K(d_i[n] - d_s[n])$

Sensor with no delay:  $d_s[n] = d_o[n]$

Locomotion:  $d_o[n] = d_o[n-1] - Tv[n-1]$

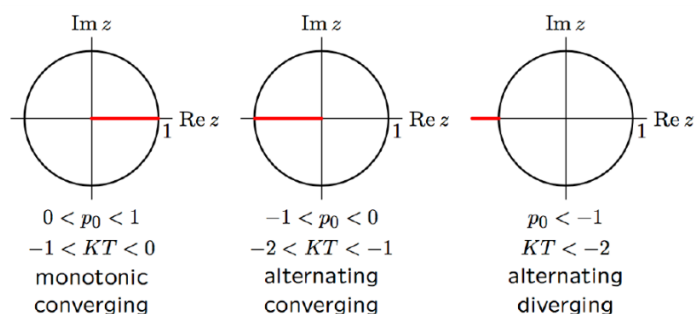


So we get  $\frac{D_o}{D_i} = \frac{-KTR}{1-(1+KT)R}$

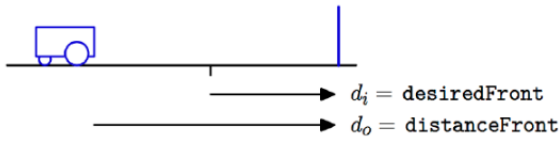
The single pole is  $z = 1 + KT$

### Pole Analysis

$$\frac{D_o}{D_i} = \frac{-KTR}{1-(1+KT)R} = \frac{(1-p_0)R}{1-p_0R} \text{ and } p_0 = 1 + KT$$



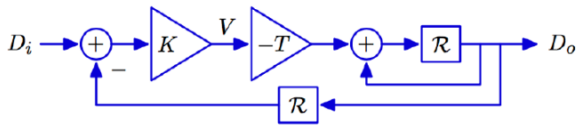
# Analysis of Wall Finder System: Adding Sensor Delay



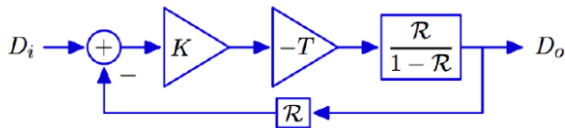
Controller:  $v[n] = K(d_i[n] - d_s[n])$

Sensor with no delay:  $d_s[n] = d_o[n - 1]$

Locomotion:  $d_0[n] = d_o[n - 1] - Tv[n - 1]$

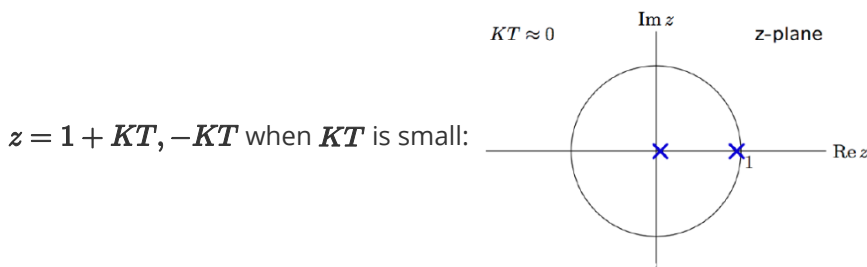


Or we get



$$\frac{D_o}{D_i} = \frac{-KTR}{1-R-KTR^2} \text{ with the poles } z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT}$$

## Poles Analysis



- pole near 0 fast response
- pole near 1 slow response
- slow response, slow mode dominates the response.

When  $KT$  become negative,  $z = \frac{1}{2}$  with  $KT = -\frac{1}{4}$

- The system is stable.
- Persistent responses (from pole  $> 1$ ) decay.

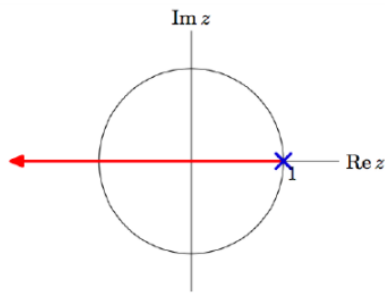
When  $KT < -\frac{1}{4}$ , then the poles become complex.

- Oscillation.
- $KT = -1$  the period of oscillation is 6 since  $p_0 = e^{\pm j\pi/3}$ .

# Destabilizing Effect of Delay

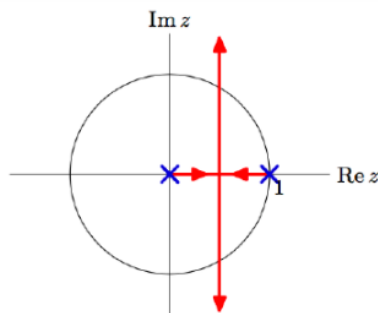
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- no delay:



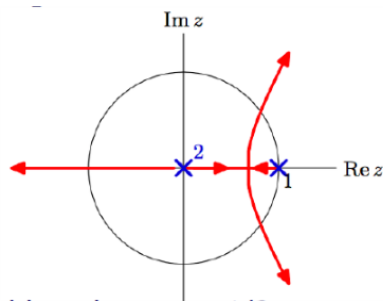
$z = 0$  is the fastest response with 0 delay.

- one delay:



$z = \frac{1}{2}$  is the fastest response with 1 delay.

- more delay:



$z = 0.682$  is the fastest response with 2 delay. Even slower.

## Summary

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- Stability of feedback system is determined by dominant pole.
- Delay tend to decrease the stability of the system.