LITEON

Low Cost Current Mode PWM Controller

GENERAL DESCRIPTION

OB2262 is a highly integrated current mode PWM control IC optimized for high performance, low standby power and cost effective offline flyback converter applications in sub 30W range.

PWM switching frequency at normal operation is externally programmable and trimmed to tight range. At no load or light load condition, the IC operates in extended 'burst mode' to minimize switching loss. Lower standby power and higher conversion efficiency is thus achieved.

VDD low startup current and low operating current contribute to a reliable power on startup design with OB2262. A large value resistor could thus be used in the startup circuit to minimize the standby power.

The internal slope compensation improves system large signal stability and reduces the possible sub-harmonic oscillation at high PWM duty cycle output. Leading-edge blanking on current sense(CS) input removes the signal glitch due to snubber circuit diode reverse recovery and thus greatly reduces the external component count and system cost in the design.

OB2262 offers complete protection coverage with automatic self-recovery feature including Cycle-by-Cycle current limiting (OCP), over load protection (OLP), VDD over voltage clamp and under voltage lockout (UVLO). The Gate-drive output is clamped to maximum 16V to protect the power MOSFET.

Excellent EMI performance is achieved with On-Bright proprietary frequency shuffling technique together with soft switching control at the totem pole gate drive output.

The tone energy at below 20KHZ is minimized in the design and audio noise is eliminated during operation.

OB2262 is offered in SOT23-6, SOP-8 and DIP-8 packages.

FEATURES

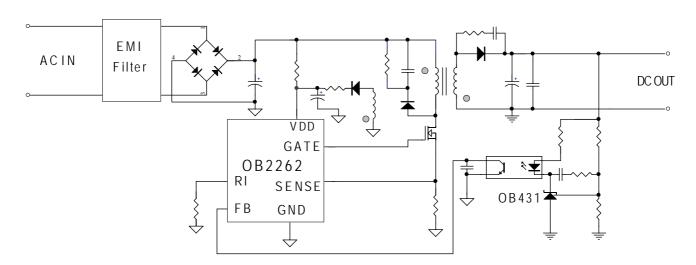
- Optimized for sub 30W Applications
- On-Bright Proprietary Frequency Shuffling Technology for Improved EMI Performance.
- Extended Burst Mode Control For Improved Efficiency and Minimum Standby Power Design
- Audio Noise Free Operation
- External Programmable PWM Switching Frequency
- Internal Synchronized Slope Compensation
- Low VDD Startup Current and Low Operating Current (2mA)
- Leading Edge Blanking on Current Sense Input
- Good Protection Coverage With Auto Self-Recovery
- VDD Over Voltage Clamp and Under Voltage Lockout with Hysteresis (UVLO)
- o Gate Output Maximum Voltage Clamp (16V)
- Line Compensated Cycle-by-Cycle Overcurrent Threshold Setting For Constant Over Power Protection (OPP) Over Universal Input Voltage Range.
- o Overload Protection (OLP).

APPLICATIONS

Offline AC/DC flyback converter for

- Battery Charger
- Power Adaptor
- Set-Top Box Power Supplies
- Open-frame SMPS

TYPICAL APPLICATION

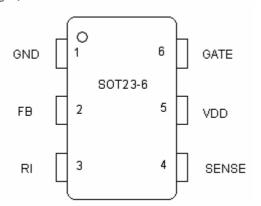


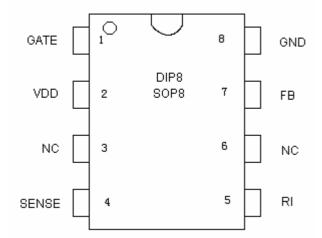
Low Cost Current Mode PWM Controller

GENERAL INFORMATION

Pin Configuration

The OB2262 is offered in SOT23-6, DIP8 and SOP8 packages, shown as below.





Ordering Information

Part Number	Description
OB2262MP	SOT23-6, Pb-free
OB2262AP	DIP8, Pb-free
OB2262CP	SOP8, Pb-free

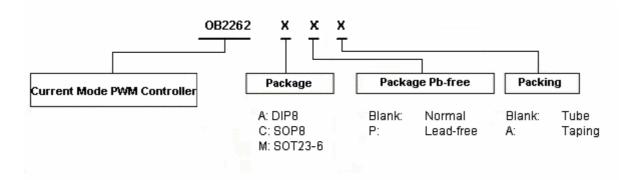
Package Dissipation Rating

Package	RθJA (°C/W)
DIP8	90
SOP8	150
SOT23-6	200

Absolute Maximum Ratings

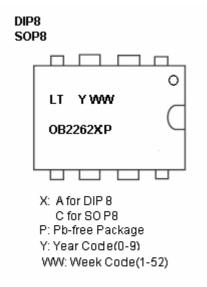
Parameter	Value
VDD Input Voltage	36 V
VDD Input DC Current	10 mA
V _{FB} Input Voltage	-0.3 to 7V
V _{SENSE} Input Voltage to Sense	-0.3 to 7V
Pin	
V _{RI} Input Voltage to RI Pin	-0.3 to 7V
Min/Max Operating Junction	-20 to 150 °C
Temperature T _J	
Min/Max Storage Temperature	-55 to 160 °C
T_{stg}	

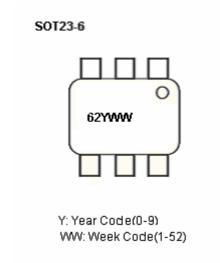
Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.





Marking Information

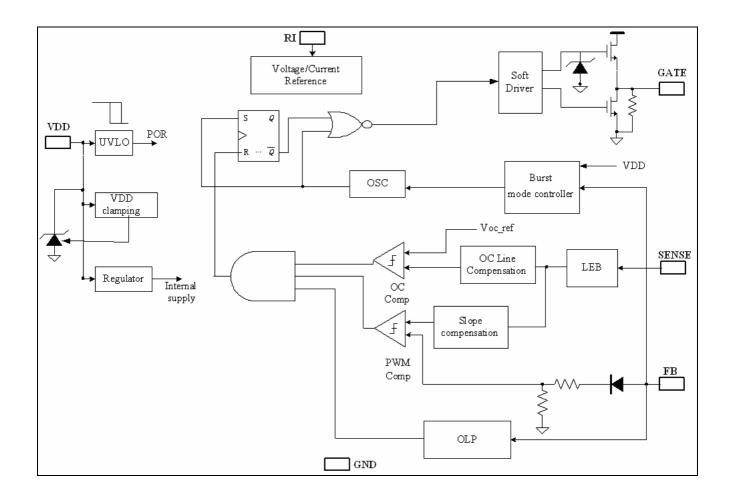




TERMINAL ASSIGNMENTS

Pin Name	I/O	Description
GND	P	Ground
FB	I	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and
		the current-sense signal at Pin 6.
RI	I	Internal Oscillator frequency setting pin. A resistor connected between RI and GND sets
		the PWM frequency.
SENSE	I	Current sense input pin. Connected to MOSFET current sensing resistor node.
VDD	P	Chip DC power supply pin.
GATE	О	Totem-pole gate drive output for the power MOSFET.

BLOCK DIAGRAM



Low Cost Current Mode PWM Controller

ESD INFORMATION

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
HBM	Human Body Model	MIL-STD		2		KV
MM	Machine Model	JEDEC-STD		250		V

ELECTRICAL CHARACTERISTICS

 $(T_A = 25^{\circ}C \text{ if not otherwise noted})$

$\frac{(1_A - 23 + C)11 \text{ not oth}}{\text{Symbol}}$	Parameter	Test Conditions	Min	Тур	Max	Unit
Supply Voltage (V	DD)		•		•	
I_VDD_Startup	VDD Start up	VDD=15V, RI=100K		4	20	uA
	Current	Measure Leakage current				
		into VDD				
I_VDD_Operation	Operation Current	VDD=16V, RI=100Kohm,		1.4		mA
		$V_{FB}=3V$				
UVLO(ON)	VDD Under		9.3	10.3	11.3	V
	Voltage Lockout					
	Enter					
UVLO(OFF)	VDD Under		15.5	16.5	17.5	V
	Voltage Lockout					
	Exit (Recovery)					
VDD_Clamp	VDD Zener Clamp	$I_{DD} = 10 \text{ mA}$		34		V
	Voltage					
Feedback Input Se						
V _{FB} Open	V _{FB} Open Loop			4.8		V
	Voltage					
I _{FB} Short	FB pin short circuit	Short FB pin to GND and		0.5		mA
_	current	measure current				
V _{TH} _PL	Power Limiting FB			3.7		V
	Threshold Voltage					
T _D _PL	Power limiting			45		mSec
	Debounce Time					
Z _{FB} _IN	Input Impedance			10		Kohm
Current Sense Inp	ut(Sense Pin)					
T_blanking	Leading edge			700		ns
	blanking time					
Z _{SENSE} _IN	Input Impedance			40		Kohm
T _D _OC	Over Current	From Over Current Occurs		150		nSec
	Detection and	till the Gatedrive output				
	Control Delay	start to turn off				
V _{TH} _OC	Internal Current	VAC = 90V	0.80	0.85	0.90	V
	Limiting Threshold					
	Voltage					
Oscillator						
Fosc	Normal Oscillation	RI = 100 Kohm	60	65	70	KHZ
	Frequency		<u>L</u>		<u> </u>	
	Frequency	VDD = 16V, RI =		5		%
	Temperature	100Kohm				
	Stability					
	Frequency Voltage	VDD = 16-20V, RI =		5		%
	Stability	100Kohm				
RI_range	Operating RI Range		50	100	150	Kohm
V RI open	RI open load			2		V



Low Cost Current Mode PWM Controller

	voltage						
F_BM	Burst Mode Base	VDD = 16V, RI =		22		KHZ	
	Frequency	100Kohm					
Gate Drive Output	ţ						
VOL	Output Low Level	VDD = 16V, Io = 20 mA			1	V	
VOH	Output High Level	VDD = 16V, Io = 20 mA	12			V	
V_Clamp	Output Clamp			18		V	
	Voltage Level						
T_r	Output Rising Time	VDD = 16V, $CL = 1nf$		280		nSec	
T_f	Output Falling	VDD = 16V, $CL = 1nf$		85		nSec	
	Time						
Frequency Shufflin	Frequency Shuffling						
Δf_{OSC}	Frequency	RI=100K	-4		4	%	
	Modulation range						
	/Base frequency						
f_shuffling	Shuffling	RI=100K		65		HZ	
	Frequency						

II =ON

Low Cost Current Mode PWM Controller

OPERATION DESCRIPTION

The OB2262 is a highly integrated PWM controller IC optimized for offline flyback converter applications in sub 30W power range. The extended burst mode control greatly reduces the standby power consumption and helps the design easily international power conservation meet the requirements.

Startup Current and Start up Control

Startup current of OB2262 is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet provides reliable startup in application. For AC/DC adaptor with universal input range design, a 2 M Ω , 1/8 W startup resistor could be used together with a VDD capacitor to provide a fast startup and low power dissipation solution.

Operating Current

The Operating current of OB2262 is low at 1.4mA. Good efficiency is achieved with OB2262 low operating current together with extended burst mode control features.

Frequency shuffling for EMI improvement

frequency Shuffling/jittering (switching frequency modulation) is implemented in OB2262. The oscillation frequency is modulated with a random source so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore reduces system design challenge.

Extended Burst Mode Operation

At zero load or light load condition, majority of the power dissipation in a switching mode power supply is from switching loss on the MOSFET transistor, the core loss of the transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the number of switching events within a fixed period of time. Reducing switching events leads to the reduction on the power loss and thus conserves the energy. OB2262 self adjusts the switching mode according to the loading condition. At from no load to light/medium load condition, the FB input drops below burst mode threshold level. Device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on state.

Otherwise the gate drive remains at off state to minimize the switching loss and reduces the standby power consumption to the greatest extend. The frequency control also eliminates the audio noise at any loading conditions.

Oscillator Operation

A resistor connected between RI and GND sets the constant current source to charge/discharge the internal cap and thus the PWM oscillator frequency is determined. The relationship between RI and switching frequency follows the below equation within the specified RI in Kohm range at nominal loading operational condition.

 $F_{PWM} = 6500/RI (KHZ)$

Current Sensing and Leading Edge Blanking Cycle-by-Cycle current limiting is offered in OB2262 current mode PWM control. The switch current is detected by a sense resistor into the sense pin. An internal leading edge blanking circuit chops off the sense voltage spike at initial MOSFET on state due to Snubber diode reverse recovery so that the external RC filtering on sense input is no longer needed. The current limit comparator is disabled and cannot turn off the external MOSFET during the blanking period. The PWM duty cycle is determined by the current sense input voltage and

Internal Synchronized Slope Compensation

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

• Gate Drive

the FB input voltage.

OB2262 Gate is connected to an external MOSFET gate for power switch control. Too weak the gate drive strength results in higher conduction and switch loss of MOSFET while too strong gate drive output compromises the EMI.

A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme. An internal 16V clamp is

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Low Cost Current Mode PWM Controller

added for MOSFET gate protection at higher than expected VDD input.

• Protection Controls

Good power supply system reliability is achieved with its rich protection features including Cycle-by-Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO).

With On-Bright Proprietary technology, the OCP is line voltage compensated to achieve constant

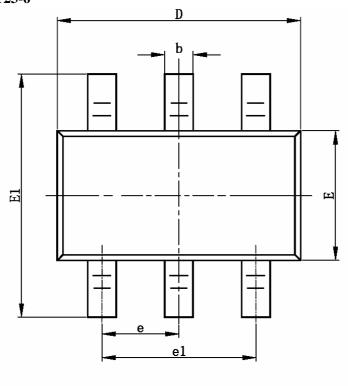
output power limit over the universal input voltage range.

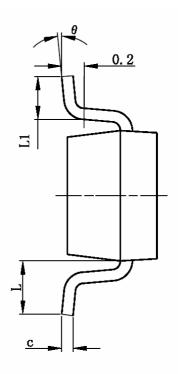
At overload condition when FB input voltage exceeds power limit threshold value for more than TD_PL, control circuit reacts to shut down the output power MOSFET. Device restarts when VDD voltage drops below UVLO limit.

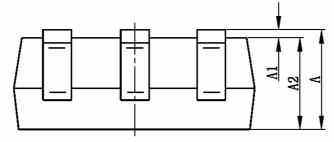
VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than threshold value. The power MOSFET is shut down when VDD drops below UVLO limit and device enters power on start-up sequence thereafter.

PACKAGE MECHANICAL DATA

SOT23-6

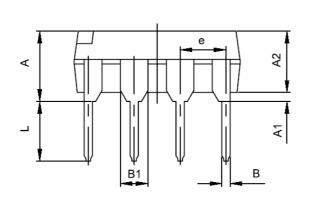


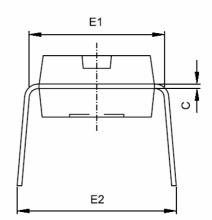


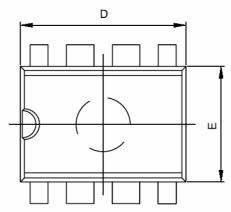


Symbol	Dimensions In Millimeters		Dimensions In Inches		
Syllibol	Min	Max	Min	Max	
Α	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.400	0.012	0.016	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
Е	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950	OTYP	0.037TYP		
e1	1.800	2.000	0.071	0.079	
L	0.700	OREF	0.028REF		
L1	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	

8-Pin Plastic DIP

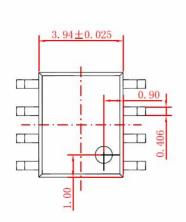


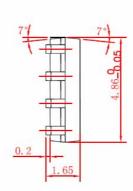


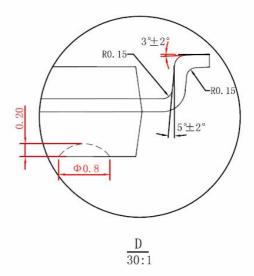


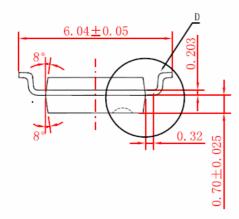
Complete	Dimensions In Millimeters		Dimensions In Inches		
Symbol	Min	Max	Min	Max	
Α	3.710	4.310	0.146	0.170	
A1	0.510		0.020		
A2	3.200	3.600	0.126	0.142	
В	0.360	0.560	0.014	0.022	
B1	1.524(TYP)		0.060(TYP)		
С	0.204	0.360	0.008	0.014	
D	9.000	9.400	0.354	0.370	
E	6.200	6.600	0.244	0.260	
E1	7.620(TYP)		0.300(TYP)		
е	2.54	-0(TYP)	0.100(TYP)		
L	3.000	3.600	0.118	0.142	
E2	8.200	9.400	0.323	0.370	

8-Pin Plastic SOP









Low Cost Current Mode PWM Controller

IMPORTANT NOTICE

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