

McMillan-McGee: Characterizing Resistance and Inductance on a Copper Bus Bar

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Abstract

Voltage spikes occur when current is abruptly interrupted and can damage inadequately protected equipment. To install protective equipment, one must find the resistance and inductance of the object carrying the current. Using Maxwell's Equations one can relate the resistance and inductance to the intensity of the magnetic field induced by the current. The authors suggest a Helmholtz equation using discrete methods for computation to model the magnetic field intensity.

1 Introduction

The engineering firm McMillan-McGee came to *Math^{industry}* with a thermodynamics problem involving electromagnetic fields that required solving a certain non-homogenous boundary value problem involving Maxwell's equations. McMillan-McGee developed a high frequency inverter. This high frequency alternating current induces an electromagnetic field and if this current is abruptly interrupted then a voltage spike will occur that puts the equipment at risk of being damaged. To fix this it is necessary to install a suitable bus bar system that can absorb energy caused by switching transients from semiconductor devices. Hence, both the resistance and inductance of the DC bus bar that supplies the current must be characterized.

Previous work has been done on this subject by Norman McLachlan¹ using an ellipse to approximate a rectangular cross-section of a bus bar. In his work he developed formulas to find current density, power loss, and high frequency resistance. Unfortunately, McLachlan's¹ work is in Gaussian units instead of Meters, Kilogram, Seconds, Coulombs units, also known as MKSC, which is undesirable for an engineer. Finally, McLachlan¹ draws the conclusion that the surface distribution of current density is identical to that of a bar holding an electric charge. Hence, the total current flowing axially on the bus bar corresponds to the total surface charge.

In this report, models will be developed to characterize resistance and inductance along a bus bar

in MKSC units. The models will be developed in a way such that it is trivial to change the situation dependent constants such as the wave number or length of the bus bar. Two different models will be developed, one in elliptical coordinates and the other in rectangular coordinates.

2 Theory

The bus bar being made of copper allows itself to be a good conductor which the magnetic field barely penetrates. As such, the problem can be reduced from three dimensions to two dimensions, only observing the surface layer. The component of the magnetic field normal to the bar's surface tends to decay exponentially. Thus, the magnetic field is roughly tangential to the bus bar and it follows that on the surface the vector magnetic potential is constant and satisfies the same conditions as the electrostatic potential. Using Maxwell's Equation's we can relate the intensity of the electric field to the resistance and inductance. Letting $\mathcal{D} = \epsilon_0 \mathcal{E}$, $\mathcal{H} = \frac{\mathcal{B}}{\mu_0}$, where \mathcal{E} is the vector electric field, \mathcal{H} is the vector magnetic field, ρ is the charge density scalar field, ϵ_0 and μ_0 are scalar values that vary depending on the units you are working in, and J is the scalar current density field, then for Maxwell's equations we have³,

$$\begin{aligned} \frac{\partial \mathcal{D}}{\partial t} &= \nabla \times \mathcal{H} - J, & (\text{Ampère's Law}) \\ \frac{\partial \mathcal{B}}{\partial t} &= -\nabla \times \mathcal{E}, & (\text{Faraday's Law}) \\ \nabla \cdot \mathcal{D} &= \rho, & (\text{Gauss' Law}) \\ \nabla \cdot \mathcal{H} &= 0. & (\text{Coulomb's Law}) \end{aligned} \tag{1}$$

Given that we only need to concern ourselves with the surface layer of the bus bar we can look at Maxwell's Equations in two dimensions. Let $(0, 0, \mathcal{E})$ and $({}_1\mathcal{H}, {}_2\mathcal{H}, 0)$ denote the electric field and the magnetic fields, respectively. Then,

$$\mathcal{E}_y = -{}_1\mathcal{H}_t, \tag{2}$$

$$\mathcal{E}_x = {}_2\mathcal{H}_t, \tag{3}$$

$$\mathcal{E}_t = {}_2\mathcal{H}_x - {}_1\mathcal{H}_y. \tag{4}$$

In this scenario, our electromagnetic field is time harmonic. Letting i represent the complex number $\sqrt{-1}$ and k the wave number, then we have,

$${}_1\mathcal{H}(x, y, t) = e^{ikt} {}_1h(x, y), \tag{5}$$

$${}_2\mathcal{H}(x, y, t) = e^{ikt} {}_2h(x, y), \tag{6}$$

$$\mathcal{E}(x, y, t) = e^{ikt} u(x, y). \tag{7}$$

After putting these two relationship mappings together, we see by equations (2), (5) and (7) that we have,

$$e^{ikt}u_y(x, y) = -ike^{ikt}{}_1h(x, y).$$

Thus ,

$$\frac{i}{k}u_y(x, y) = {}_1h(x, y).$$

Similarly by equations (3), (6) and (7),

$$e^{ikt}u_x(x, y) = ike^{ikt}{}_2h(x, y)$$

this implies,

$$\frac{-i}{k}u_x(x, y) = {}_2h(x, y)$$

Finally, by combining the previous results and equation (4),

$$ike^{ikt}u(x, y) = (\frac{-i}{k}u_{xx}(x, y) - \frac{i}{k}u_{yy}(x, y))e^{ikt}$$

give us,

$$-\Delta u(x, y) - k^2u(x, y) = 0.$$

Notice that from Maxwells equations we have recovered a two-dimensional Helmholtz equation. Given the propensity of the electrons to cluster at the endpoints of the bus bar, we will require non-homogenous Dirichlet boundary conditions. The k value, referred to as the wave number, is determined by ω , μ and σ the circular frequency, permeability in a vacuum, and electrical conductivity of copper, respectively, with relationship $k^2 = -i\omega\mu\sigma$.

3 Methodology

The computations for this project were all done on ‘MATLAB R2019a’. The boundary conditions were selected in a way to match the propensity of electrons to cluster at the endpoints of the copper bus bar. When observing the rectangular case, $(x, y) \in [0, a] \times [0, b]$, the Dirichlet boundary conditions imposed were,

$$u(0, y) = u(a, y) = \text{Amplitude} \times \cos^2(\frac{\pi x}{a}),$$

$$u(x, 0) = u(x, b) = \text{Amplitude} \times \cos^2(\frac{\pi y}{b}).$$

In the case of the ellipse we determined that its area should be equal to the area of the rectangle, to match the results of the rectangular case. Now, recall that an ellipse can be characterized as the points $(x, y) \ni \frac{x^2}{\gamma^2} + \frac{y^2}{\delta^2} = 1$. Hence, the boundary condition can be parameterized as $x(\theta) = \gamma\cos(\theta)$ and $y(\theta) = \delta\sin(\theta)$ where $\theta \in [0, 2\pi)$. The initial condition can then be given by,

$$u(x(\theta), y(\theta)) = \text{Amplitude} * \cos^2(\theta).$$

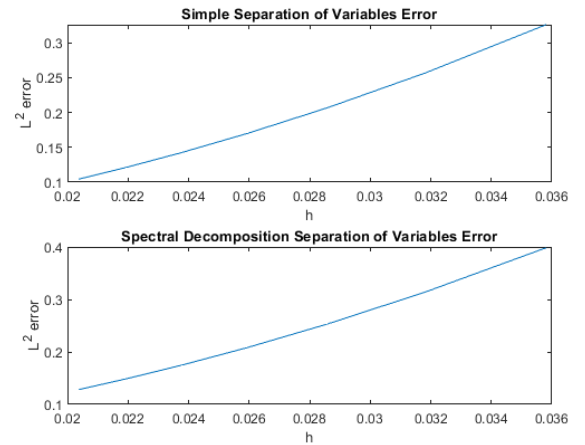
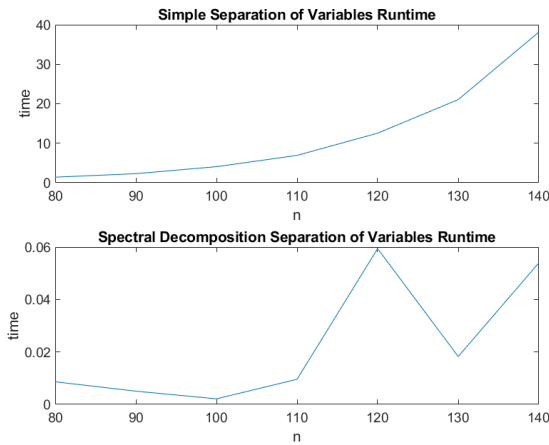
The code has been written in such a way that it is generalized for an arbitrary problem of a similar nature. For demonstration purposes, McMillan-McGee provided us with sample values to use that matched their particular problem. The constants used are as follows (where A is amplitude),

$$\begin{aligned}
 a &= 8m, \\
 b &= 1m, \\
 \gamma &= 4m \\
 \delta &= \frac{2}{\pi}m, \\
 \mu &= 4 * \pi 10^{-7} \frac{N}{A^2}, \\
 \sigma &= 58.7 \cdot 10^6 \frac{S}{m}, \\
 \omega &= 2\pi 10^6 s^{-1}, \\
 A &= 10m.
 \end{aligned}$$

With those conditions and constants, we apply the usual second-order finite difference scheme to discretize the Helmholtz equation on the rectangular domain. We solve the resultant linear system using discrete separation of variables. This method has a time complexity of $O(n^3)$ as opposed to Gaussian Elimination which has a cumbersome $O(n^6)$ time complexity², where $n = \frac{1}{h}$ and h is the size of the grid in each direction. In the elliptic case we applied continuous piecewise linear finite elements to solve the problem. To ensure the integrity of the data from our algorithm we tested it on an exact solution where the electric intensity was $u(x, y) = x^2 - y^2$ the wave number $k = 1$ and the forcing term $F = y^2 - x^2$. Testing both the Gaussian Elimination and discrete separation of variables we tracked both the runtime and error of each algorithm,

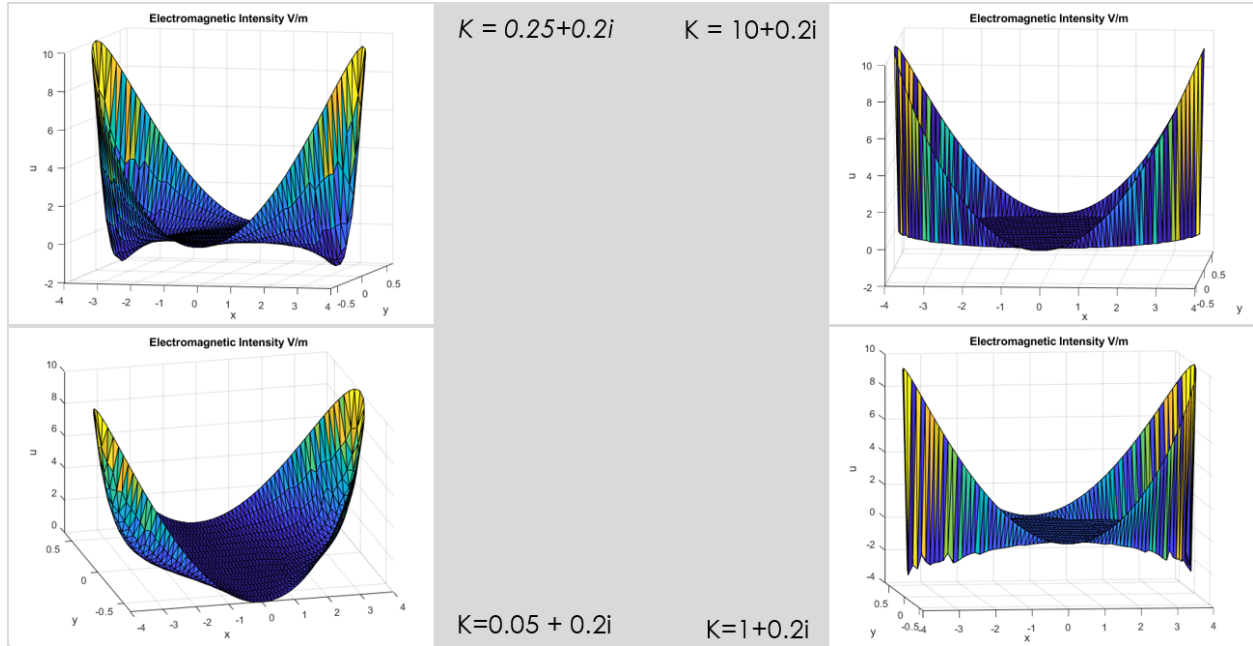
Runtime

Error



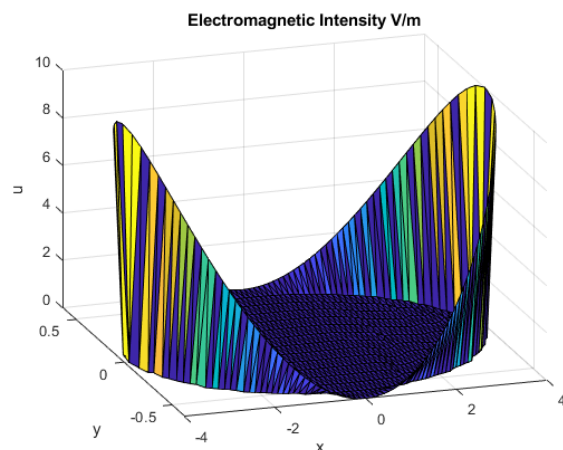
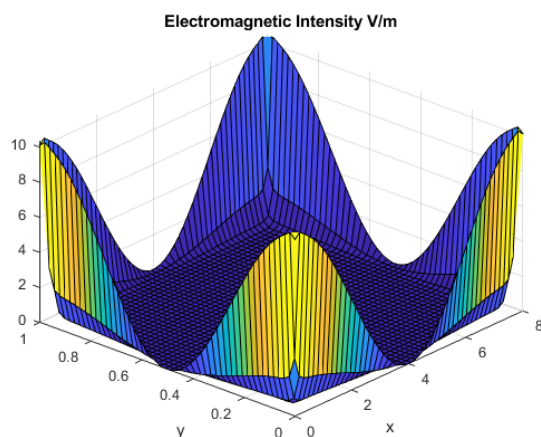
From the results we decided to go with the discrete separation of variables approach.

Additionally, we did some test runs for different values of the wave number k , to study the behaviour of our model.



4 Results

Using the equations and parameters described in the previous sections, we arrived at the following results for our two regions of interest i.e. a rectangular bus bar and an elliptical one. Firstly, we concluded that the electromagnetic intensity is greater at the end points, and decays in the midsections of the copper bus bar as shown in the figures below. This implied that our models adequately represent a real world scenario of the problem according to our discussion with our industry mentor at McGillan-McGee. Additionally, we would like to mention that from the two chosen regions used for our models, the rectangular bus bar is more applicable as it is most resembles the real world situation; since this is most common shape of bus bars. The elliptical model results will allow our industry partner at McGillan-McGee to continue investigating the work done by McLachlan¹ on the subject, which was the original problem that was presented to us. It is important to remark that our models are generalized so it is trivial to substitute the parameters for alternative conditions.



5 Conclusions

Having created the rectangular model that accurately depicts the scenario of current flowing through a copper bus bar, Maxwells Equations can be used to characterize resistance and inductance. Knowing the resistance and inductance will allow McGillan-McGee to alter the copper bus bar in such a way that voltage spikes will not run back through it and ruin the attached machinery. The elliptical model we created will allow McGillan-McGee to continue following along McLachlan's¹ work and potentially pursue a different avenue.

6 Acknowledgements

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7 References

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