







Session 1: Signal, Audio, Speech encoding





INTRODUCTION



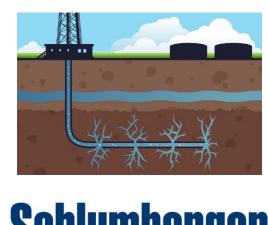
















COURSE STRUCTURE



Quick Summary

Audio processing for Al

- Signal, audio, speech encoding (4h)
- Deep learning for audio processing (4h)

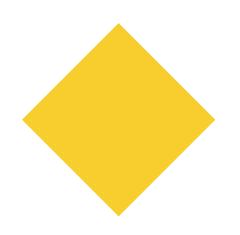
Automata for language modelling

- HMM for speech processing (4h)
- Automata and transducer (4h)

Towards speaking with an Al-bot

- Speech synthesis (4h)
- Automatic speech recognition (4h)
- Speaker and emotion recognition (4h)

SESSION 1: SIGNAL, AUDIO, SPEECH ENCODING



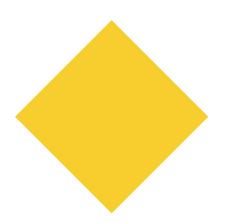
Quick Summary

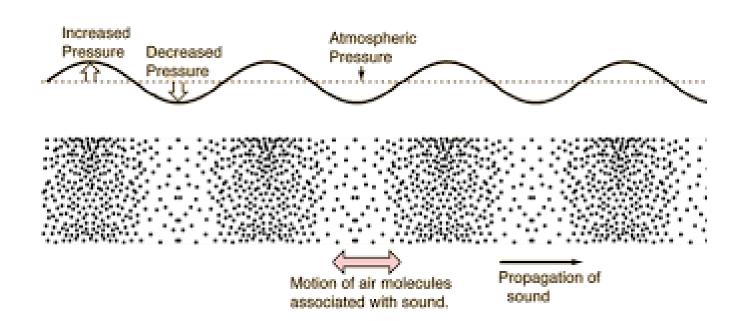
- 1. The physics of sound
- 2. Signal representation and purpose
- 3. Signal characteristics
- 4. Signal models
- 5. Classical ML approaches

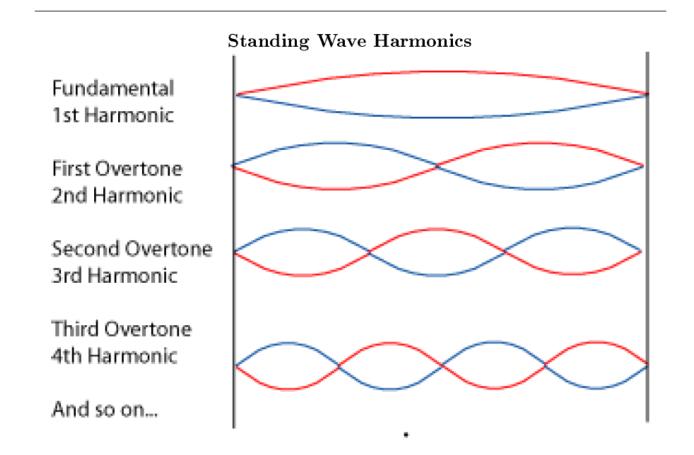
SIGNAL, AUDIO, SPEECH ENCODING.

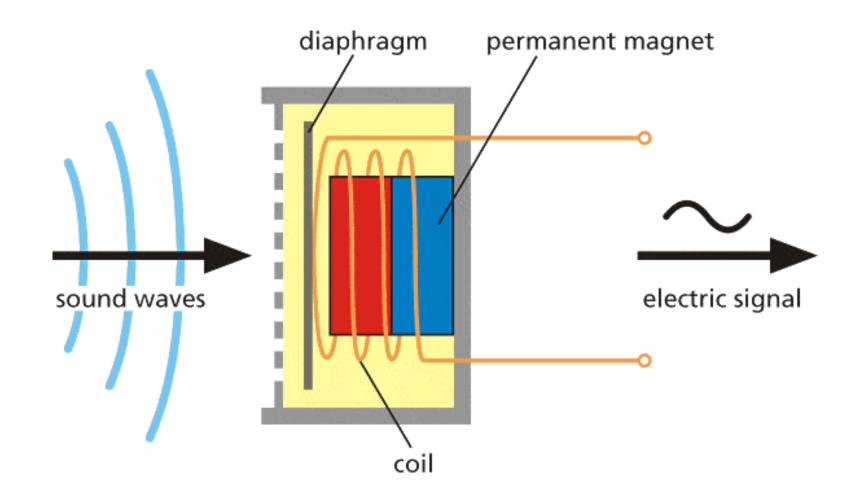
The physics of sound

THE PHYSICS OF SOUND



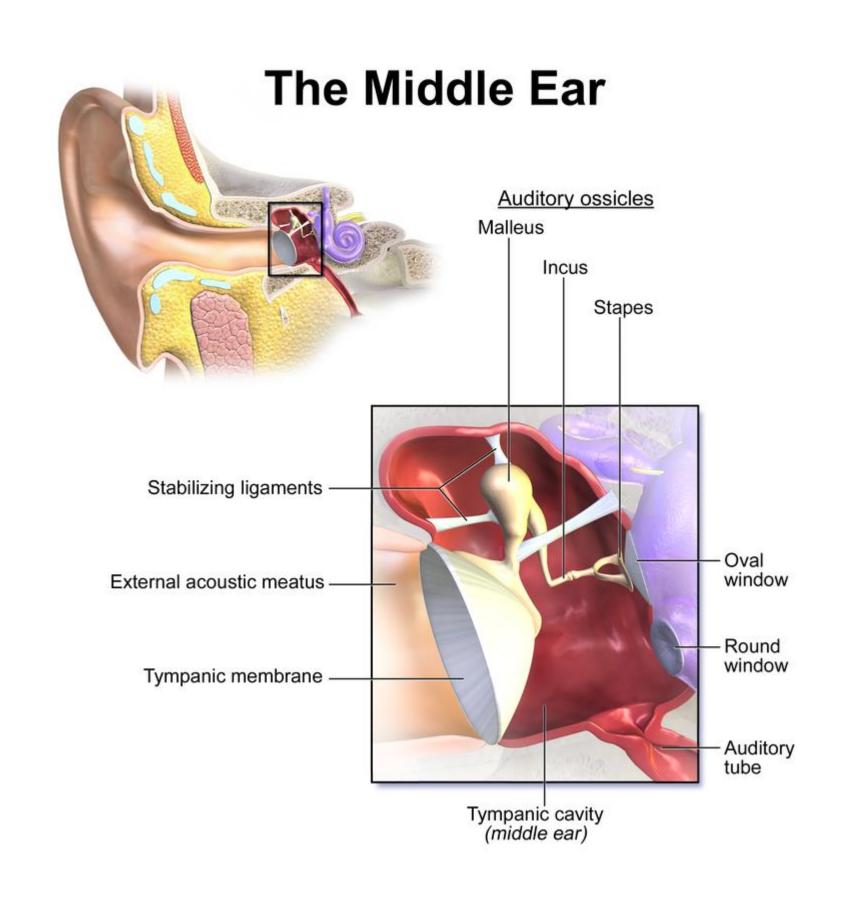


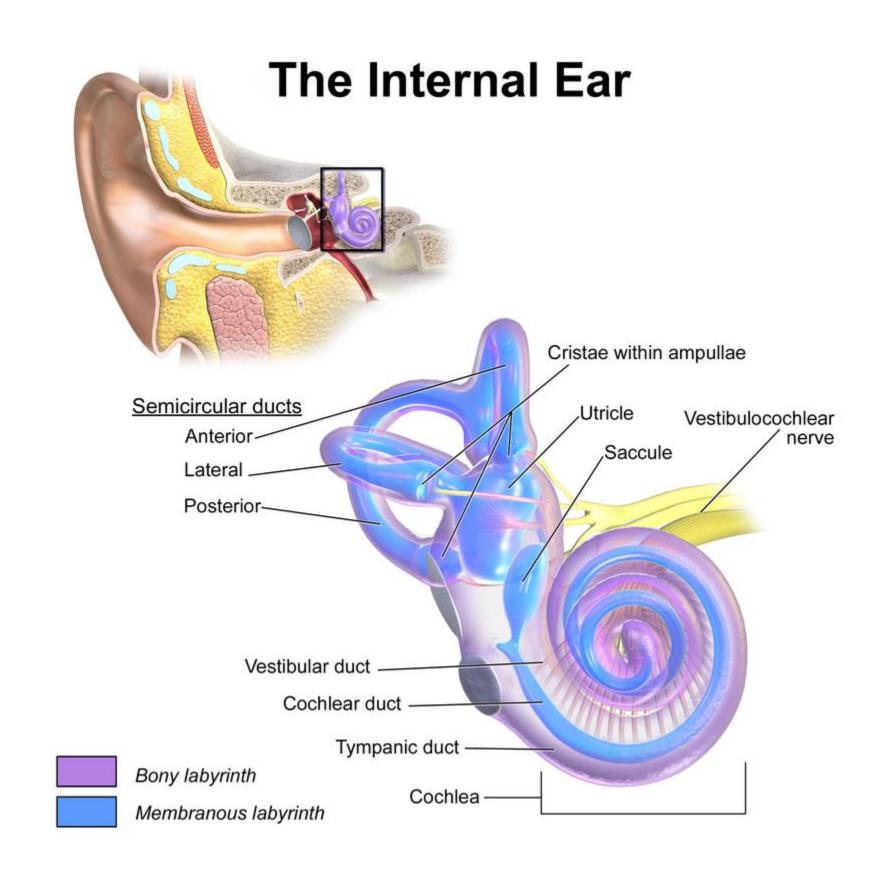




THE PHYSICS OF SOUND







SIGNAL, AUDIO, SPEECH ENCODING.

Signal representation & purpose



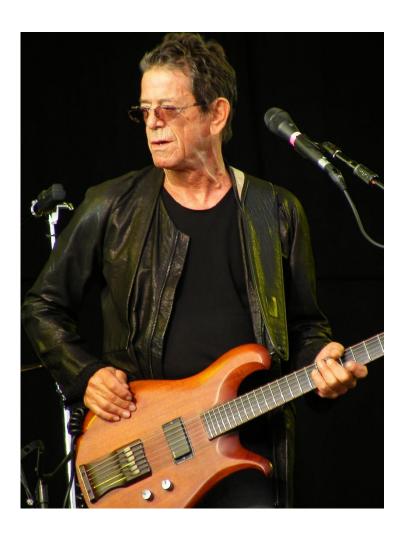
Different types of audio content

Noise



Multisource polyphonic with unstructure sound sources

Music



Multi timbre polyphonic with structured sound sources

Speech



Monophonic with structured sound source



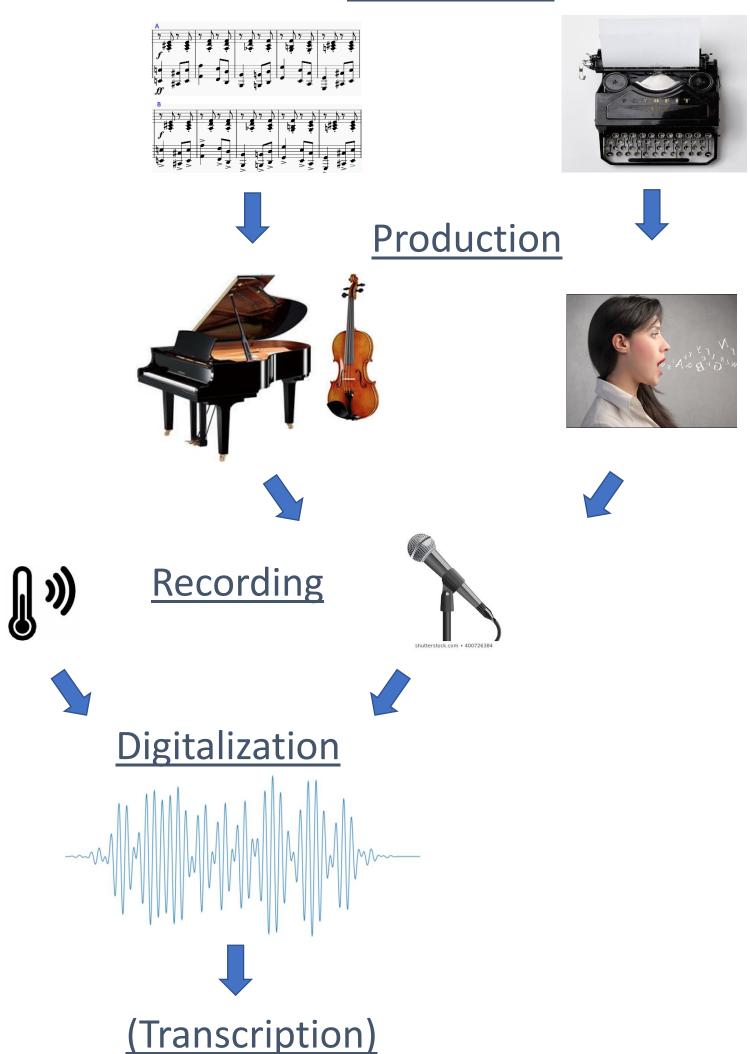






Signal & Audio symbolic representations

Codification



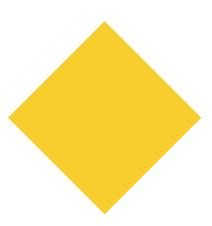
Vertical and horizontal organization

Frequency measurement:

- Sources
- Timbres
- Pitch polyphony
- Chord
- Tonality

Time measurment:

- Structure
- Bars
- Rythm



Applications of AI in signal processing

Musical Information Retrieval (ISMIR)

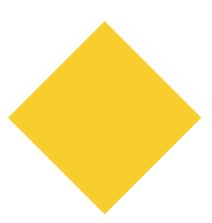
- Automatic Music Transcription
- Auto-tagging
- Music Recommendation
- Source separation
- Audio/Music generation
- Style transfer
- Fingerprinting

Spoken Content Retrieval

- Trigger word detection
- Segmentation
- Speech-to-text (ASR)
- •Text-to-speech
- Emotion recognition
- Speaker recognition, diarization
- Denoising
- Active Noise Control

Other signal processing applications

•Time series prediction (finance, insurance, healthcare,...)



Fourier series

$$f(t) = \sum_{\substack{n=0 \\ +\infty \\ +\infty}}^{+\infty} k_n \sin(2\pi n f_0 t + \phi_n)$$

$$= \sum_{\substack{n=0 \\ +\infty \\ +\infty}}^{+\infty} A_n \cos(2\pi n f_0 t) + B_n \cos(2\pi n f_0 t)$$

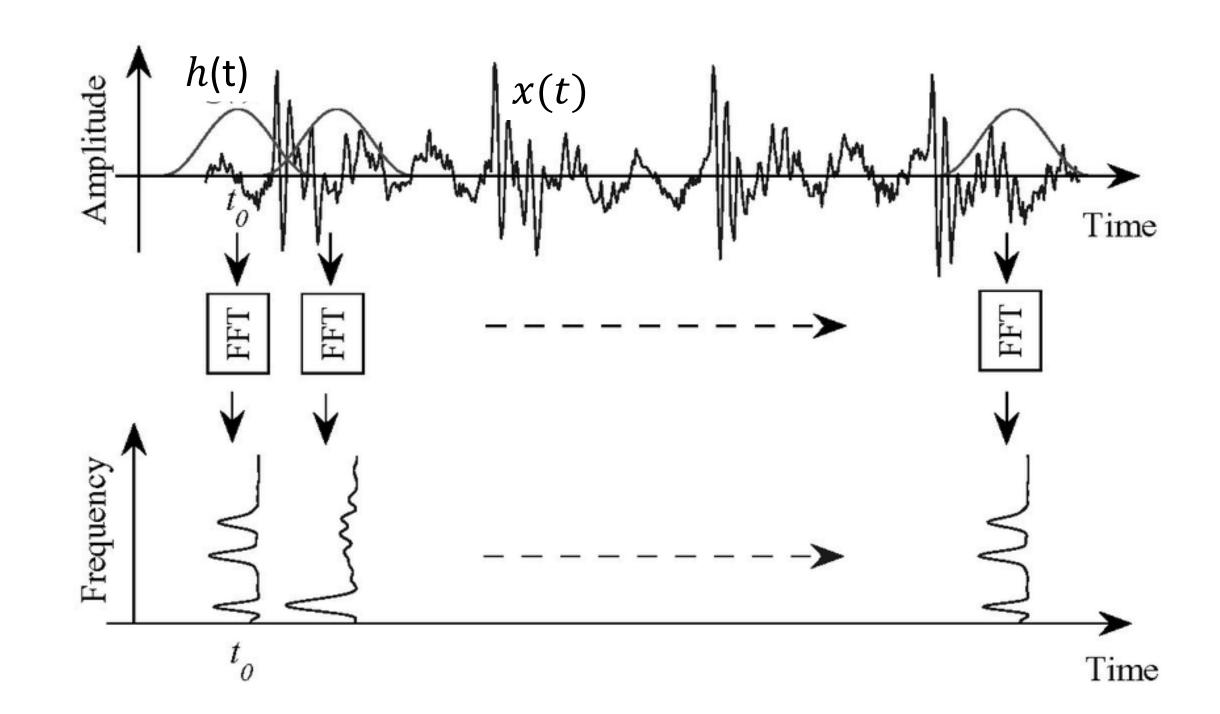
$$= \sum_{n=0}^{+\infty} c_n e_0^{j2\pi f t}$$

Discrete Fourier Transform

$$X(k) = \sum_{m=0}^{N-1} x(m) e_0^{j2\pi \frac{k}{N}m}$$

Short-Time-Fourier-Transform (STFT)

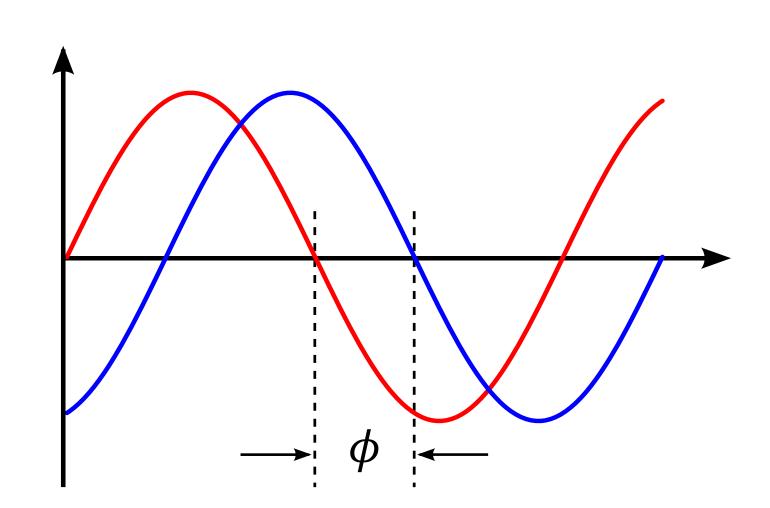
$$X(k,n) = \sum_{m=0}^{N-1} x(m) h(n-m) e_0^{j2\pi \frac{k}{N}m}$$



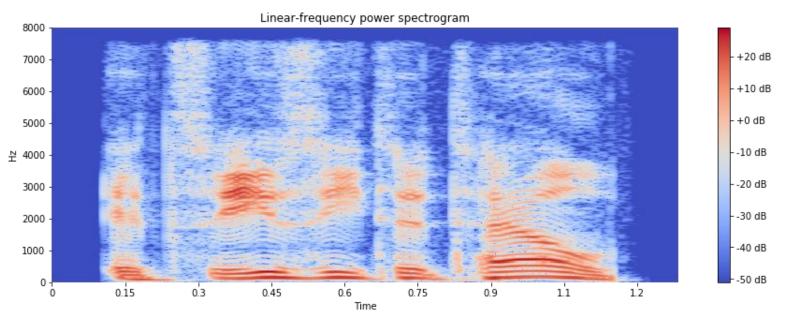
h: windowing $k_n, A_n, B_n, c_n, x(m), h(m) \in \mathbb{R}$

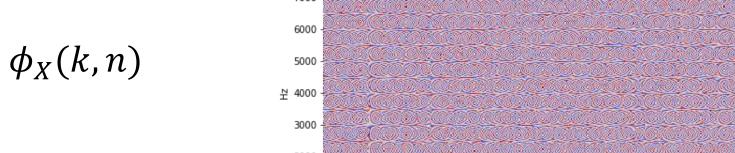


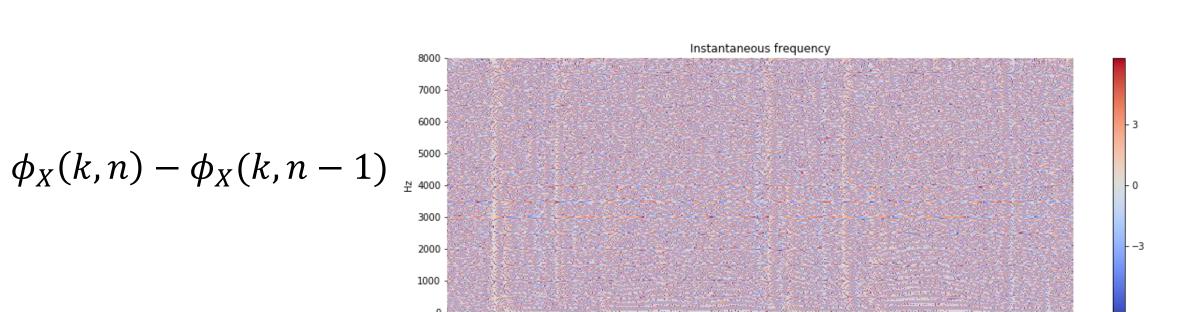
- $X(k) = |X(k)|e^{j\phi_X(k)}$
- Amplitude |X(k)|
- Phase $\phi_X(k)$, temporal localization of information (Phase is necessary to reconstruct temporal signal)
- Instantaneous frequency



 $10\log_{10}|X(k,n)|^2$







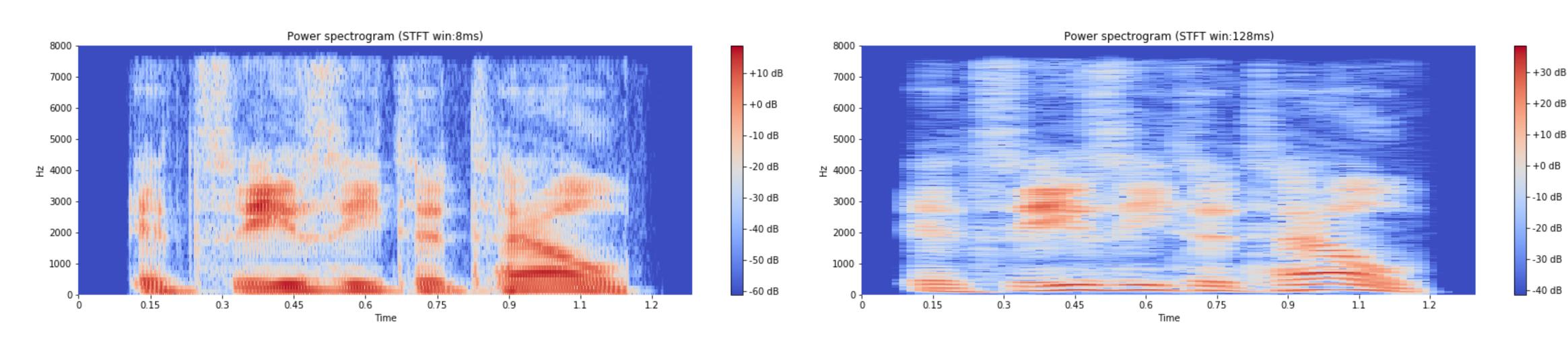


n: time windows index





Remark on trade-off temporal/frequential resolution



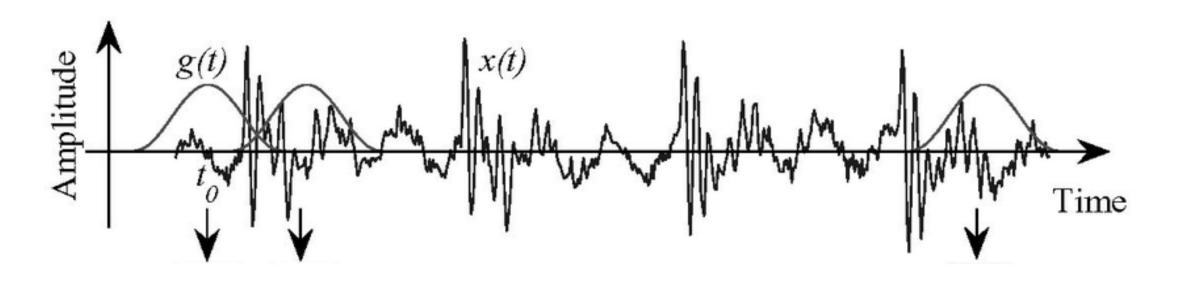
Remark on sound perception

log perception of sound amplitude
 → magnitude spectrogram usually in dB

$$10\log_{10}|X(k,n)|^2$$

log perception of sound frequency
 → magnitude spectrogram often in mel scale

$$10 \log_{10} |X(k|_{mel}, n)|^2$$



with mel scale:
$$m = 2595 \log_{10} \left(1 + \frac{f}{700}\right)$$



Wavelet transform (WT)

(not used in practice, but concept is used to define CQT, see next slide)

Continuous wavelet basis definition

$$\forall t \in \mathbb{R}, \psi_{S,\tau}(t) = \frac{1}{\sqrt{S}} \psi\left(\frac{t-\tau}{S}\right)$$



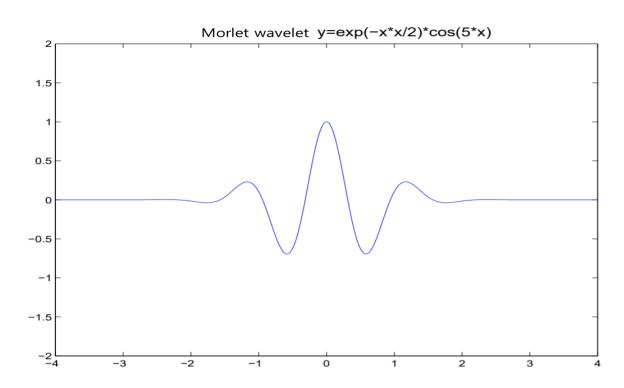


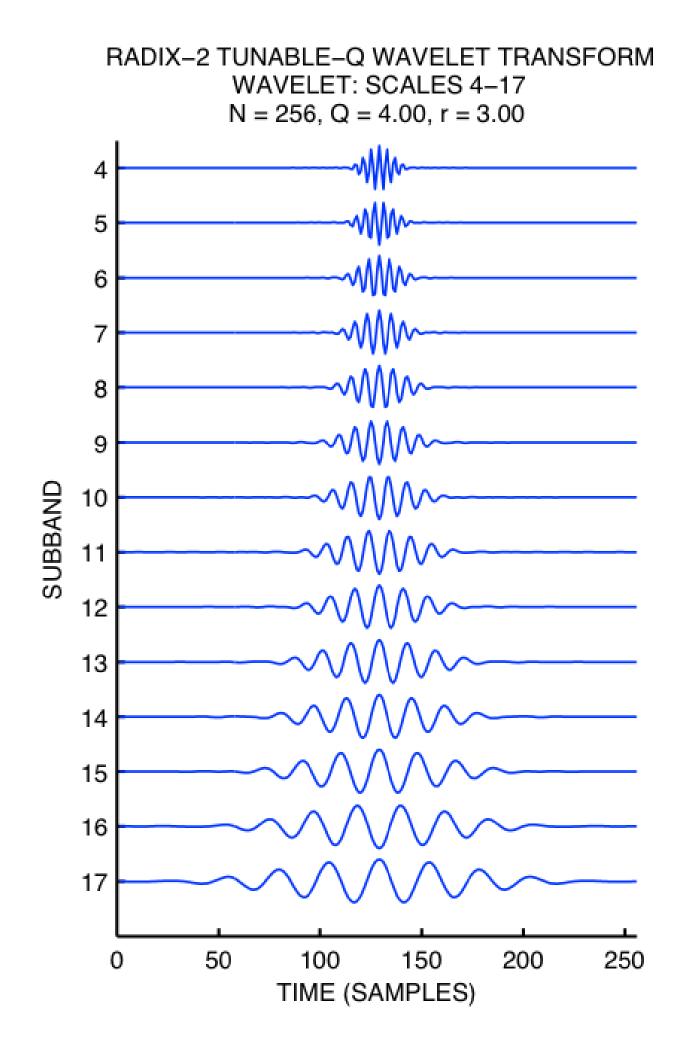
- discretization in wavelet (time-frequency) domain: au o n and s o k
- discretization in time domain: $t \rightarrow m$

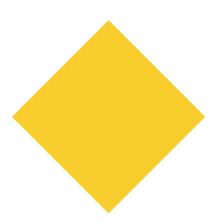
$$\psi_{k,n}[m] = s_0^{-\frac{\kappa}{2}} \psi(s_0^{-k}m - n\tau_0)$$
 with s_0 and τ_0 constants

Discrete wavelet transform definition

$$W_{\psi}[k,n] = rac{1}{\sqrt{N}} \sum_{m=0}^{N-1} x[m] \psi_{k,n}[m]$$
 with n time index and k frequency bin







Constant-Q transform (CQT)

(behavior of WT but no orthogonality CQ functions)

Frequency resolution changing with frequency considered

$$N_{[k]} = Q \frac{f_{S}}{f_{k}}$$

A pitch translation corresponds to a frequency translation on CQT

$$X_{[k,n]}^{CQ} = \sum_{m=n-\frac{N_k}{2}}^{n+\frac{N_k}{2}} \frac{h\left(\frac{n-j}{N_k} - \frac{1}{2}\right)}{N_k} e^{-i2\pi Q\left(\frac{n-j}{N_k} - \frac{1}{2}\right)}$$

h is a window (Hanning, Blackmann Harris, ...)

--10 dB

--20 dB

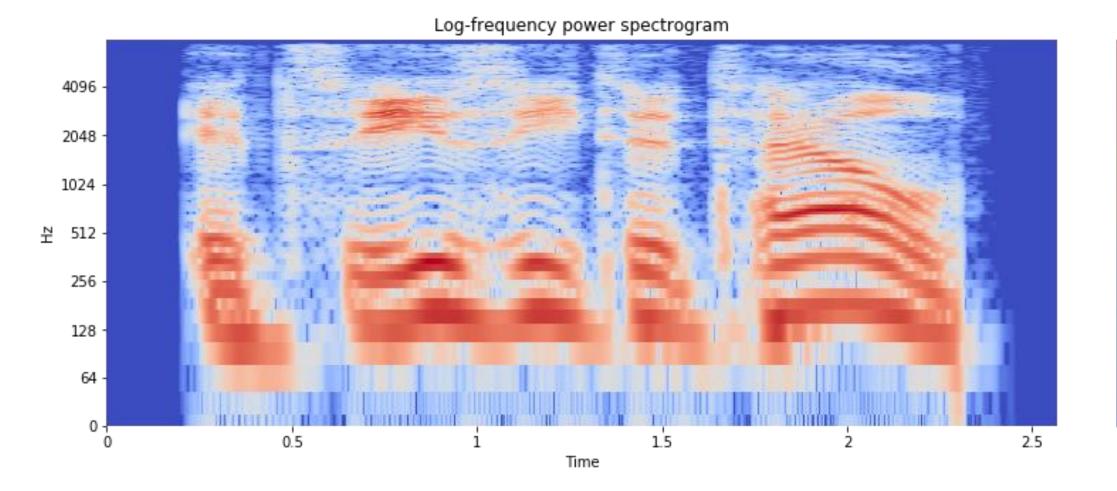
--30 dB

--40 dB

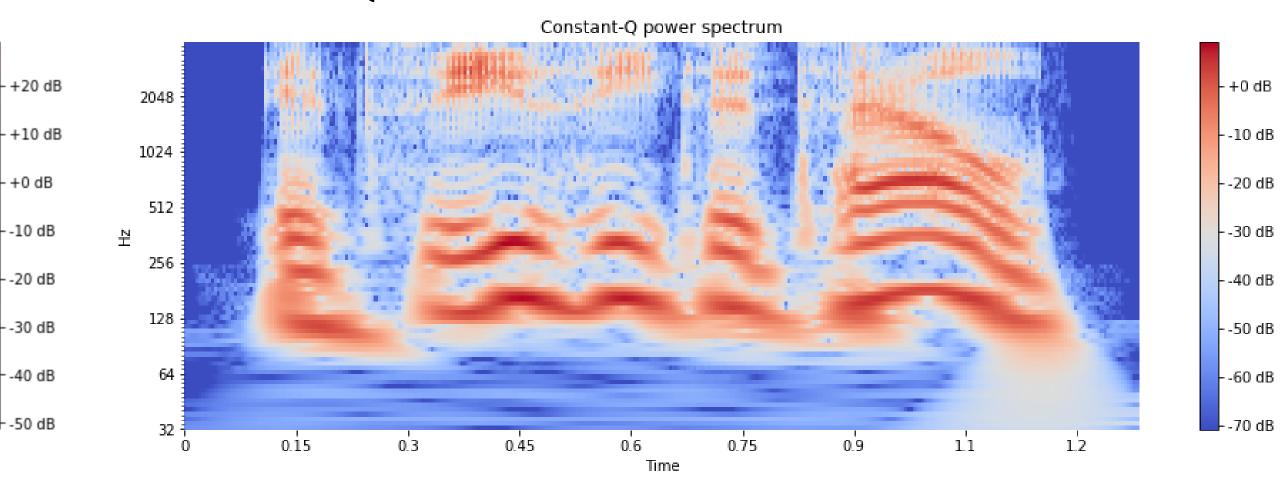


FT vs CQT





Constant-Q transform



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Signal characteristics

CHARACTERISTICS



Digital waveform characteristics

- Sample rate
- Bits depth
- level dBfs
- mean RMS (dBfs)
- Compression(dBA, dBC, SPL)

Artifacts

- noise (SNR)
- amplitude distortion (signal balance) target curve
- phase distortion (unprecise onset, spatialization)
- non-linearities (clipping for instance) THD
- delay (real-time, digitalization or processing)

Sources of artifacts

- Sound or signal capture,
- sound or signal production
- sound propagation in air
- signal propagation in wires,
- processing

SIGNAL, AUDIO, SPEECH ENCODING. Signal models

SIGNAL MODELS



Sine wave harmonics and noise model:

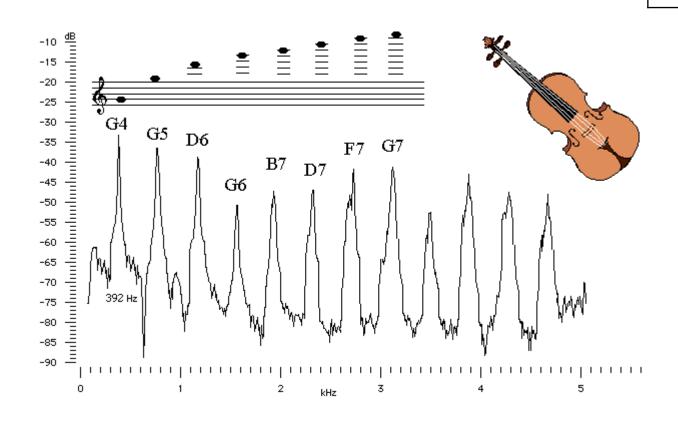
$$x(t) = \sum_{h=1}^{H(t)} A_h(t) \cos(\phi_h(t)) + b(t)$$

Assumption that the audio signals follows this model.

The spectral enveloppe $\{A_h\}$ defines the timbre of the instrument.

Example of voice or instruments harmonic spectrum (spectral envelop)

-10 dB -15 -20 -25 -30 -35 -35	
-25	6 •
-30 =	₹G4 D6
-35 =	G5
-40 =	
-45 🗏	∬
-50 =	D7 G7
-50 -55 -60 -65 -70 -75 -80 -80 -80	/\
-60 - ≣	- / \
-65	
-70 🖥	392 Hz 392 Hz
-75 =	" " " " " " " " " " " " " " " " " " "
=	water when prover your young proving p
-85	i i i i Mista Alada
-90 <u>=</u>	



Frequency	Order	Name	Standing wave representation
1 × f = 440 Hz	n = 1	1st harmonic (fundamental tone)	lu(t,x) displacement U(t,x) displacement
2 × f = 880 Hz	n = 2	2nd harmonic (1 st overtone)	first harmonic or fundamental frequency pipe side view
3 × f = 1320 Hz	n = 3	3rd harmonic (2 nd overtone)	first harmonic or fundamental frequency
4 × f = 1760 Hz	n = 4	4th harmonic (3 rd overtone)	U(1,2) displacement

SIGNAL MODELS



Source/filter model

Periodical pulse convolved with a resonant filter

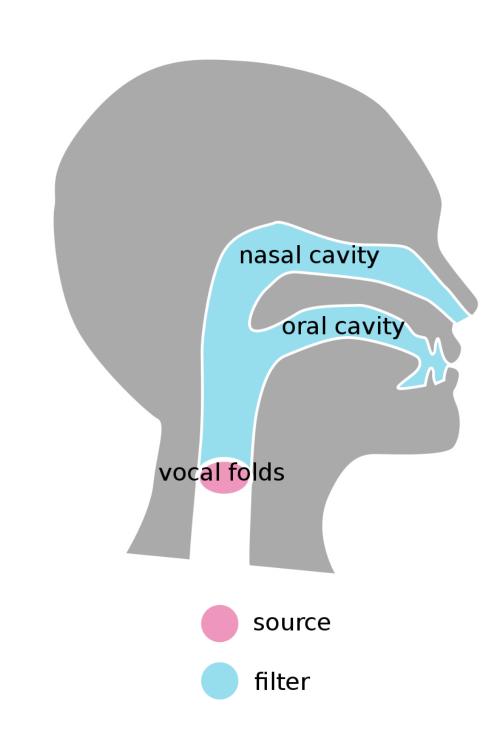
$$x(t) = e(t) * h(t)$$
$$X(f) = E(f) \cdot H(f)$$

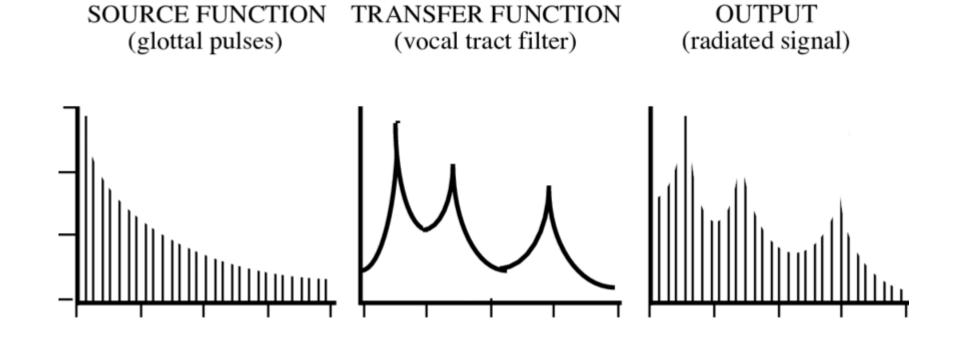
• Resonant filter is an AR filter:

$$H(z) = \frac{G}{1 + \sum_{k=1}^{K} a_k z^{-k}}$$

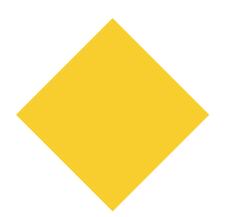
x(n) can be predicted with a linear combinaison of previous values

$$x(n) \approx \sum_{k=1}^{K} a_k x(n-k)$$





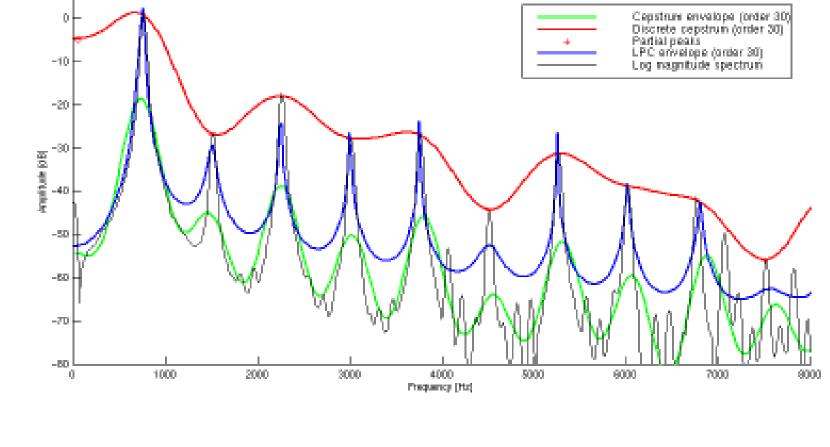
SIGNAL MODELS

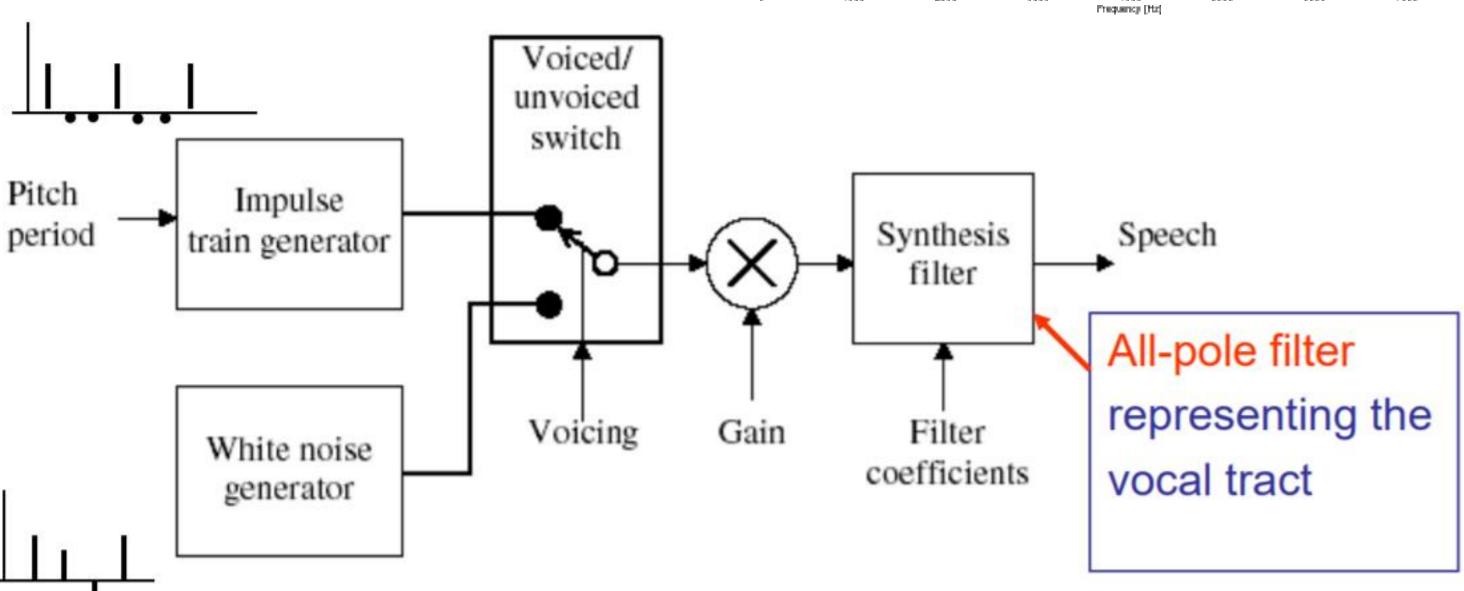


Classical speech production model: LPC

- Impulse train or white noise (voiced/unvoiced speech frame)
- Gain: energy level of frame
- AR synthesis filter
- Pitch period (pitch height)

$$y(n) = \sum_{i=1}^{p} a_i y(n-i) \pm Gx(n)$$





Variants: LPC-10, CELP, MELP, RELP, VSELP, ASELP, LD-CELP

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Classical ML approaches

CLASSICAL ML APPROACHES



Input Hand designed Mapping from features Output

Hand design features

Knowledge a priori for the design of audio features

MFCC

- Source-filter model $X(f) = E(f) \cdot H(f)$
- Cepstrum

$$cepstr(x)(\tau) = DCT^{-1}(\log(|X(f)|^2))$$

$$f(\tau) = DCT^{-1}(\log(|H(f)|^2)) + DCT^{-1}(\log(|E(f)|^2))$$

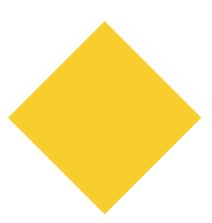
- Filter/source contribution separation
- Perception frequency scale: Mel

Chromas

Contribution of frequency height class

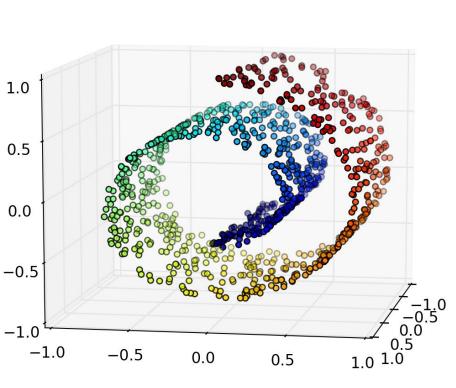
$$f(C) = X^{2}(f_{C_{1}}) + X^{2}(f_{C_{2}}) + \cdots$$

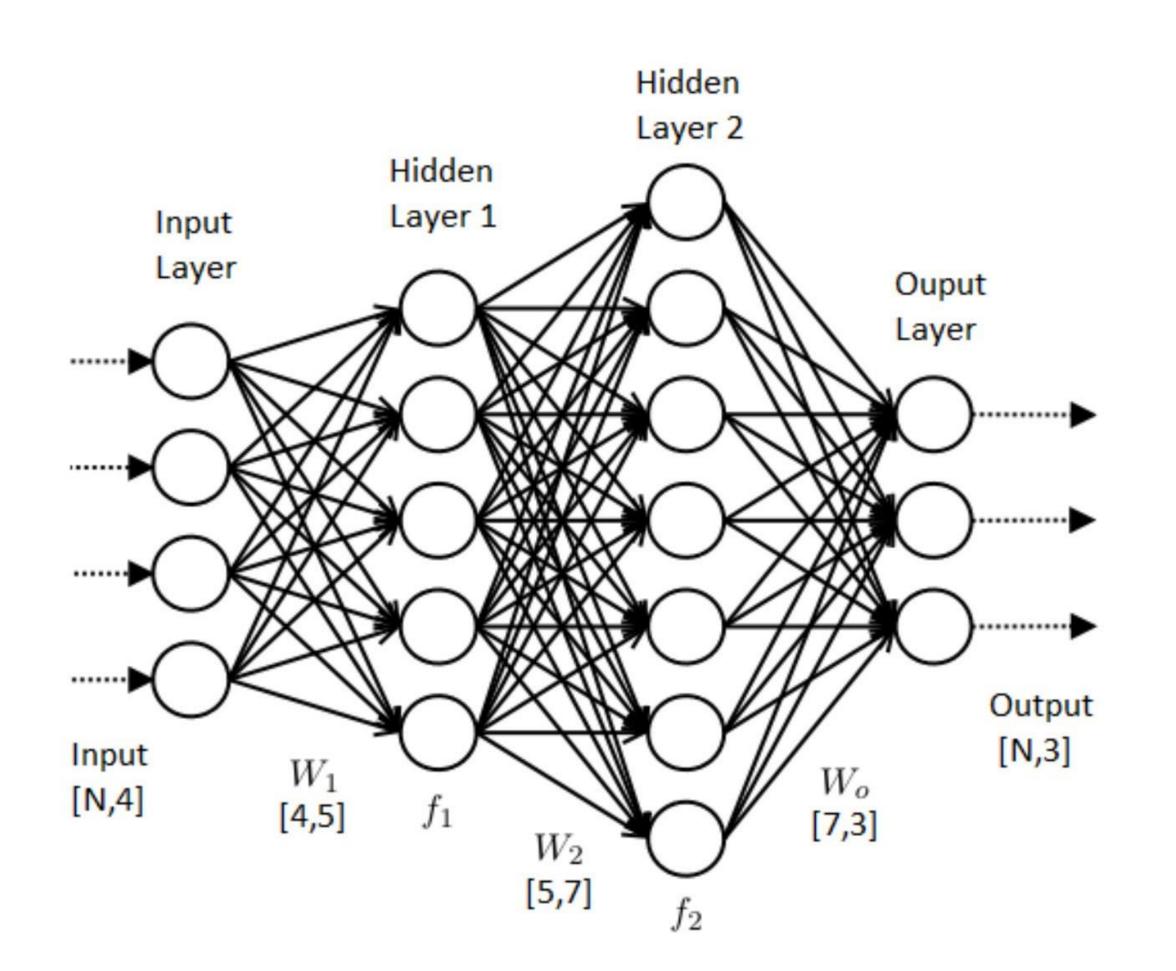
CLASSICAL ML APPROACHES



Mapping from features

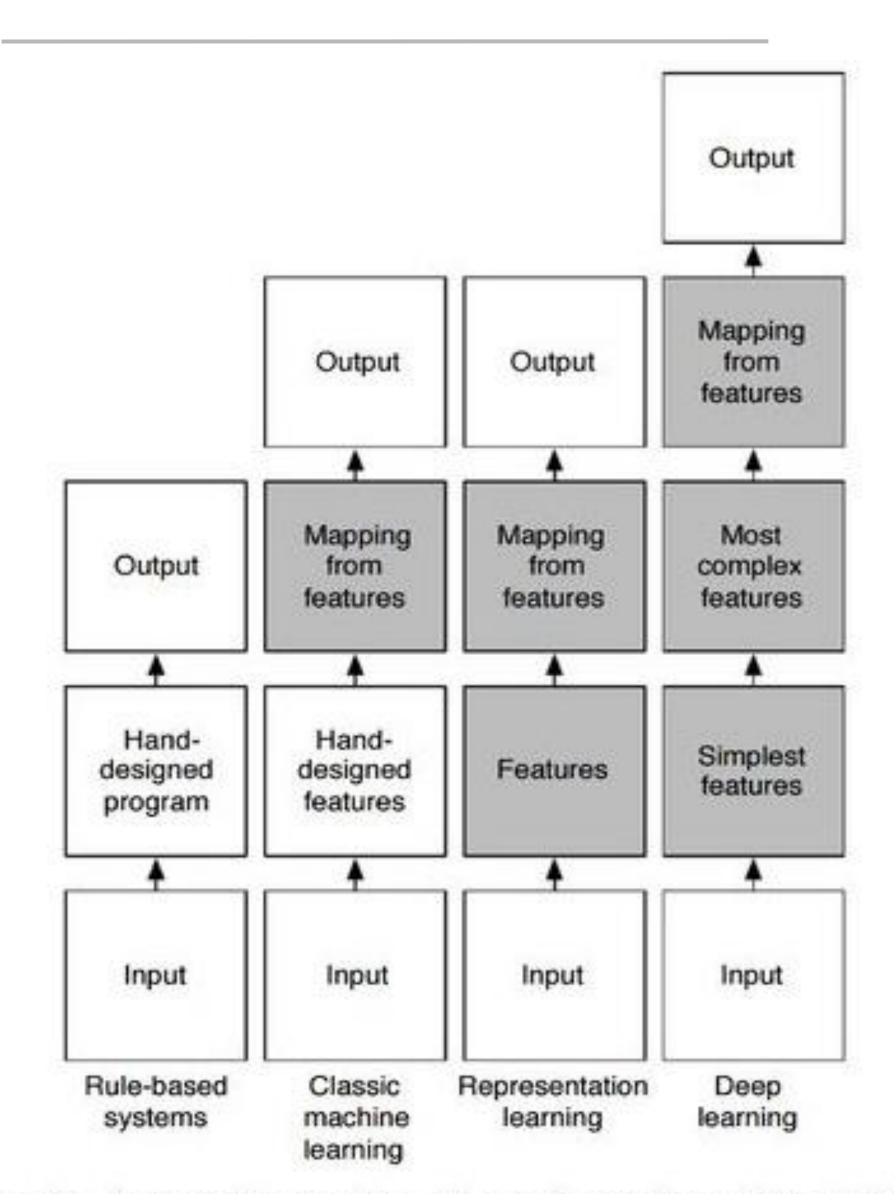
- Preactivation (linear)
- Activation (non-linear)
- Non-linear operator enabling a linear separation
- The Manifold Hypothesis states that real-world high-dimensional data lie on low-dimensional manifolds embedded within the high-dimensional space.





CLASSICAL ML APPROACHES





J. Humphrey, J. P. Bello and Y. Lecun. Moving beyond feature design: Deep architecures and automatic feature learning in music informatics. ISMIR 2012.

SIGNAL, AUDIO, SPEECH ENCODING.

Thank you for your attention.

Refenencences:

- Geoffroy Peeters, Telecom Paris Tech
- Ian Goodfellow, Deep learning, 2018

ANNEXE 1: FOURIER TRANSFORM



The Fourier transform (FT) is an integrable function $\hat{f} \colon \mathbb{R} \to \mathbb{C}$, defined as following (one among several convention):

$$F[\omega] = [\int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt]$$

with pulsation: $\omega = 2\pi \cdot f_{reg}$

The Fourier inverse function is the following:

$$f(t) = \int_{-\infty}^{\infty} F[\omega] e^{j\omega t} d\omega$$

<u>Usefull properties</u> (let be x, y real functions):

- Linearity $ax(t) + by(t) \overset{ ext{F.T}}{\longleftrightarrow} aX(\omega) + bY(\omega)$
- FT of an impulse function $\delta(\omega)=1$
- Time scaling $x(at) \frac{1}{|a|} X \frac{\omega}{a}$
- Time shifting / frequency shifting $x(t-t_0) \overset{\mathrm{F.T}}{\longleftrightarrow} e^{-j\omega t_0} X(\omega)$ and $e^{j\omega_0 t}. x(t) \overset{\mathrm{F.T}}{\longleftrightarrow} X(\omega-\omega_0)$
- Multiplication and convolution $x(t).y(t) \overset{\mathrm{F.T}}{\longleftrightarrow} X(\omega) * Y(\omega)$ and $x(t) * y(t) \overset{\mathrm{F.T}}{\longleftrightarrow} \frac{1}{2\pi} X(\omega).Y(\omega)$
- Vectorized representation of FT values $X(\omega) = |X(\omega)|e^{j\theta(\omega)}$ where $\theta(\omega) = argX(\omega)$ and $|X(\omega)|, \theta(\omega)$ are called magnitude and phase spectrum of $X(\omega)$.

ANNEXE 2: SIGNAL SAMPLING



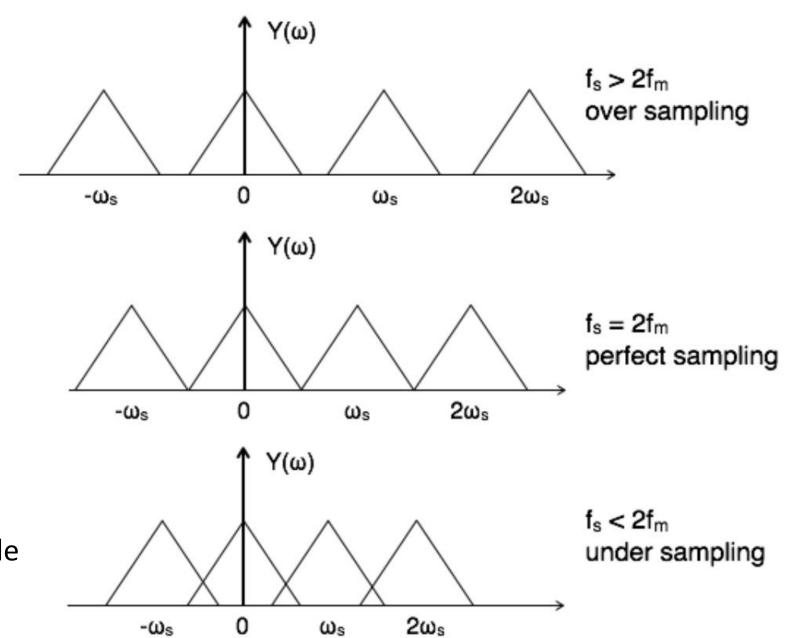
Sampling of a continuous signal into a discrete signal:

$$y(t) = x(t) imes ext{ impulse train}$$
 \mathbf{x} \mathbf

Nyquist theorem: A continuous time signal can be represented in its samples and can be recovered back when sampling frequency f_s is greater than or equal to the twice the highest frequency component of message signal. i. e.

$$f_s \geq 2 f_m$$

frequency overlap here is calle d aliasing



ANNEXE 3: CONVOLUTION & CORRELATION



Convolution of x by H (or H by x):

$$y(t) = x(t)*h(t)$$
 Input LTI SYSTEM Output
$$h(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$

$$y(t) = x(t)*h(t)$$

Property of Fourier Transform (see annexe 1)

$$x(t).\,y(t) \stackrel{ ext{F.T}}{\longleftrightarrow} X(\omega) * Y(\omega)$$

$$x(t)*y(t) \overset{ ext{F.T}}{\longleftrightarrow} rac{1}{2\pi} X(\omega).\,Y(\omega)$$

Correlation of 2 signals x_1 and x_2 :

$$\int_{-\infty}^{\infty} x_1(t) x_2(t-\tau) dt$$

ANNEXE 4: Z-TRANSFORM AND FILTERING



Definition of Z-Transform:

$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$$
 with $z = re^{j\omega}$

Possibility to create y signal that depends on x(n), x(n-1), x(n-2), ... and y(n-1), y(n-2), y(n-3), ... This operation of y creation is called x filtering. And the Z-transform helps to characterize this filter (linear and invariant in time):

$$\sum_{l=-N}^{N} a_{l} y[n-l] = \sum_{k=-M}^{M} b_{k} x[n-k]$$

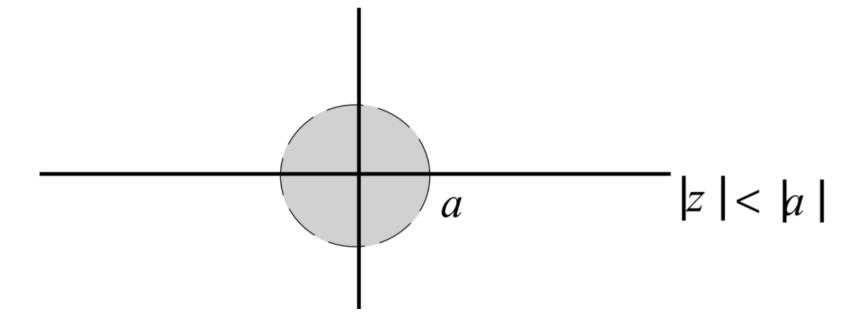
$$\sum_{l=-N}^{N} a_l z^{-l} Y(z) = \sum_{k=-M}^{M} b_k z^{-k} X(z)$$

$$\sum_{l=-N} a_l z^{-l} Y(z) = \sum_{k=-M} b_k z^{-k} X(z)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=-M}^{M} b_k z^{-k}}{\sum_{l=-N}^{N} a_l z^{-l}}$$

Specific properties of H:

- if H denominators (complex) zeros have a modu lus greater than 1, H is not stable.
- Resonances at pulsation close to argument of d enominators zeros



ANNEXE 5: WAVES INTERFERENCES & WAVEGUIDES

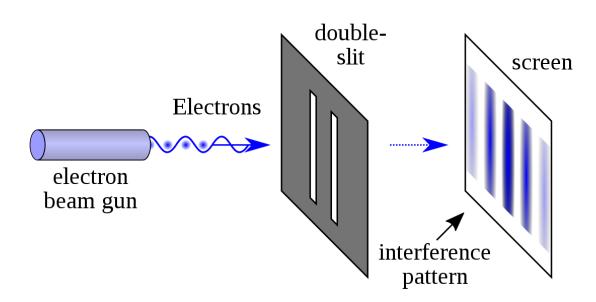
Sound wave

Propagation of a local oscillation of air pressure in a medium (air for instance).

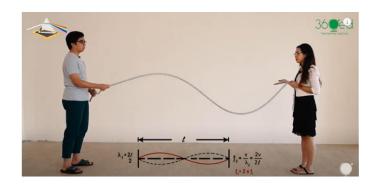
It takes the form of longitudinal or transverse (for solids only) wave.

Wave interferences

Example of light wave interference: Young experiment.



Harmonics and wave guides



If λ is the wavelength (depending on frequency f) and v is sound speed (340 m/s): $\lambda = \frac{v}{f}$

Frequency	Order	Name	Standing wave representation
$1 \times f = 440 \text{ Hz}$	<i>n</i> = 1	1st harmonic (fundamental tone)	[u(t,x) displacement] [first harmonic or fundamental frequency] [pipe side view]
2 × f = 880 Hz	n = 2	2nd harmonic (1 st overtone)	li(_x) displacement U(_x) displacement
3 × f = 1320 Hz	n = 3	3rd harmonic (2 nd overtone)	first harmonic or fundamental frequency)
4 × f = 1760 Hz	n = 4	4th harmonic (3 rd overtone)	U(1,x) displacement U(1,x) displacement

