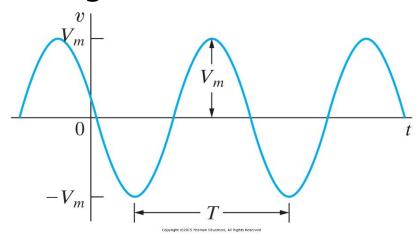
ELE 296 Basic Electric Circuits and Electronics

- Sinusoidal Steady State Analysis
 - Sinusoidal Source
 - Phasor Concept
 - •Kirchhoff Laws in the Frequency Domain
 - Source Transformation itFD
 - •Thévenin-Norton Equivalent Circuits itFD
 - Node Voltage Method itFD
 - Mesh Current Method itFD

Sinusoidal Voltage:



$$v(t) = V_m \cos(\omega t + \phi) V$$

• V_m : Amplitude

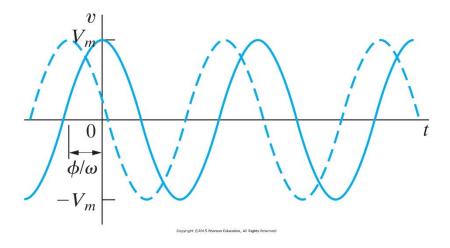
•*t* : time [s]

• ϕ : phase angle [deg]

• ω : angular frequency [rad/s] • $f = \omega / 2\pi$: frequency [Hz]

• T = 1/f: period [s]

A change in phase angle:



RMS (Root Mean Square) value:

$$V_{rms} = \sqrt{\frac{1}{T}} \int_{t_0}^{t_0+T} v^2(t) dt = \sqrt{\frac{1}{T}} \int_{t_0}^{t_0+T} V_m^2 \cos^2(\omega t + \phi) dt$$
$$= \frac{V_m}{\sqrt{2}}$$

Qn. 1: A sinusoidal voltage is given as

 $v = 300\cos(120\pi t + 30^{\circ})$

Define the period?
Define the frequency?
What is v for t=2.778ms?
What is the rms value of v?
Define v as a sine function.

• Qn. 1 (cont.): $v = 300\cos(120\pi t + 30^{\circ})$

$$\omega = 120\pi = \frac{2\pi}{T}$$
 $f = \frac{1}{T}$ $\Rightarrow T = 16.667 \,\text{ms}$ $\Rightarrow f = 60 \,\text{Hz}$

$$v = 300\cos(120\pi t + 30^{\circ}) = 300\cos(120\pi t \times \frac{180}{\pi} + 30^{\circ})$$

$$v = 300\cos(21600t + 30^{\circ})$$

$$\Rightarrow v|_{t=2.778ms} = 300\cos(60^{\circ} + 30^{\circ}) = 0V$$

• Qn. 1 (cont.): $v = 300\cos(120\pi t + 30^{\circ})$

$$V_{rms} = \frac{V_m}{\sqrt{2}}$$

$$\Rightarrow V_{rms} = \frac{300}{\sqrt{2}} = 212.13 \text{V}$$

$$\sin(x) = \cos(x - 90^{\circ})$$

$$\cos(x) = \sin(90^{\circ} - x)$$

$$\sin(-x) = -\sin(x)$$

$$\cos(-x) = \cos(x)$$

$$v = 300 \cos(120\pi t + 30^{\circ}) = 300 \cos(-120\pi t - 30^{\circ})$$

= $\sin(90^{\circ} - (-120\pi t - 30^{\circ}))$
 $\Rightarrow v = \sin(120\pi t + 120^{\circ})$

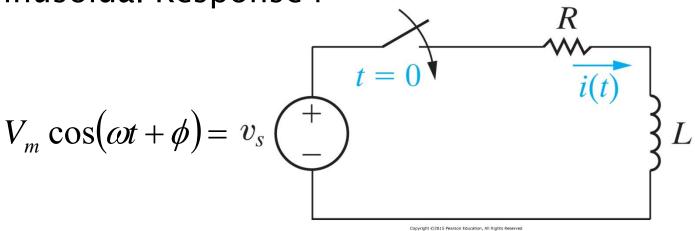
Sinusoidal Response:

$$V_m \cos(\omega t + \phi) = v_s + \sum_{i(t)} \frac{R}{i(t)}$$

$$KVL: -v_s + v_R + v_L = -V_m \cos(\omega t + \phi) + iR + L\frac{di}{dt} = 0$$

$$i = \frac{-V_m}{\sqrt{R^2 + \omega^2 L^2}} \cos(\phi - \theta) e^{-(R/L)t} + \frac{V_m}{\sqrt{R^2 + \omega^2 L^2}} \cos(\omega t + \phi - \theta), \quad \theta = \frac{\omega L}{R}$$

Sinusoidal Response :



Transient Response

Steady-State Response

$$i = \frac{-V_m}{\sqrt{R^2 + \omega^2 L^2}} \cos(\phi - \theta) e^{-(R/L)t} + \frac{V_m}{\sqrt{R^2 + \omega^2 L^2}} \cos(\omega t + \phi - \theta), \quad \theta = \frac{\omega L}{R}$$

- Steady–State Response:
 - •It is a sinusoidal function,
 - •The frequency of the response and the source are the same.
 - •The maximum amplitude of the response and the source are generally different.
 - •The phase of the response and the source are generally different.

$$v_s = V_m \cos(\omega t + \phi)$$
 $i_{ss} = \frac{V_m}{\sqrt{R^2 + \omega^2 L^2}} \cos(\omega t + \phi - \theta)$

Phasor:

- •It is a complex number that carries the amplitude and phase information.
- •Euler Equation:

$$e^{\pm j\theta} = \cos(\theta) \pm j\sin(\theta)$$

$$\cos \theta = \text{Re}\{e^{j\theta}\}$$

$$\sin \theta = \text{Im}\{e^{j\theta}\}$$

•Sinusoidal Source:

$$\begin{aligned} v &= V_m \cos(\omega t + \phi) \\ &= V_m \operatorname{Re} \left\{ e^{j(\omega t + \phi)} \right\} = V_m \operatorname{Re} \left\{ e^{j\omega t} e^{j\phi} \right\} \\ &= \operatorname{Re} \left\{ V_m e^{j\phi} e^{j\omega t} \right\} \end{aligned}$$

Phasor representation including the amplitude and phase!!!

- Phasor (cont.):
 - •Phasor Transformation:

$$\mathbf{V} = V_m e^{j\phi} = V_m \angle \phi = V_m \cos \phi + jV_m \sin \phi$$
$$= \mathbf{P} \{V_m \cos(\omega t + \phi)\}$$

- •Phasor transformation transforms the function from time domain to complex number domain.
- Also called Frequency Domain.
- •Inverse Phasor Transform:

$$\mathbf{P}^{-1} \{ V_m e^{j\phi} \} = V_m \cos(\omega t + \phi)$$

$$\omega \text{ can not be known}$$

• Qn. 2: $y_1 = 20\cos(\omega t - 30^\circ)$, $y_2 = 40\cos(\omega t + 60^\circ)$ define $y_1 + y_2$ using phasor consept.

$$y_1 = 20\cos(\omega t - 30^\circ) \Rightarrow \mathbf{Y}_1 = 20\angle - 30^\circ$$
$$y_2 = 40\cos(\omega t + 60^\circ) \Rightarrow \mathbf{Y}_2 = 40\angle 60^\circ$$

$$\mathbf{Y}_{1} + \mathbf{Y}_{2} = 20\angle -30^{\circ} + 40\angle 60^{\circ} = 20e^{-j30^{\circ}} + 40e^{j60^{\circ}}$$

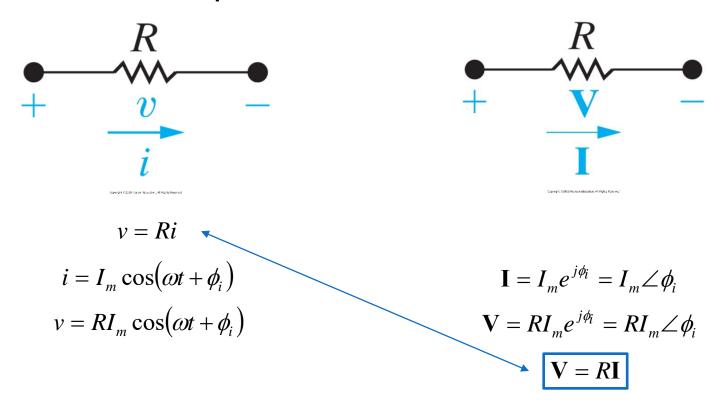
$$= (17.32 - j10) + (20 + j34.64)$$

$$= 37.32 + j24.64$$

$$= 44.72\angle 33.43^{\circ}$$

$$y_1 + y_2 = \mathbf{P}^{-1} \left\{ 44.72 \angle 33.43^\circ \right\} = \text{Re} \left\{ 44.72 e^{j33.43^\circ} e^{j\omega t} \right\}$$
$$= 44.72 \cos(\omega t + 33.43^\circ)$$

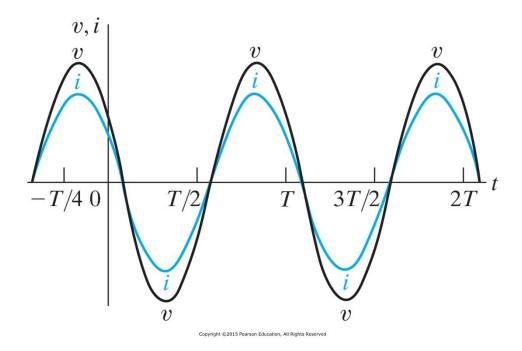
V-I Relationship for a Resistor:



• The voltage and current are called in-phase since the resistor does not cause a phase difference.

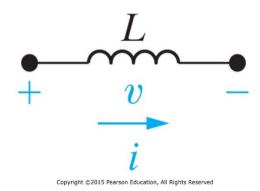
V-I Relationship for a Resistor (cont.):

$$V = RI$$



The resistor does not cause a phase difference.

V-I Relationship for an Inductor:



$$i = I_m \cos(\omega t + \phi_i)$$

$$v = L \frac{di}{dt} = -\omega L I_m \sin(\omega t + \phi_i)$$

$$= -\omega L I_m \cos(\omega t + \phi_i - 90^\circ)$$

$$j\omega L$$

$$+ V$$

$$I$$

$$V = j\omega LI$$

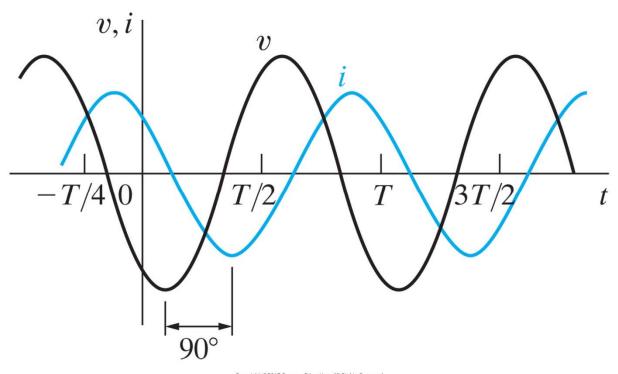
$$I = I_m e^{j\phi_i} = I_m \angle \phi_i$$

$$V = -\omega LI_m e^{j(\phi_i - 90^\circ)} = -\omega LI_m e^{j\phi_i} e^{-j90^\circ}$$

$$= j\omega LI_m e^{j\phi_i} = j\omega LI = \omega LI_m \angle (\phi_i + 90^\circ)$$

Voltage leads current by 90°.

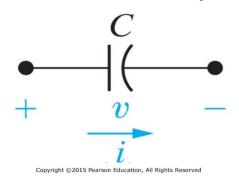
V-I Relationship for an Inductor (cont.):



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Voltage leads current by 90°.

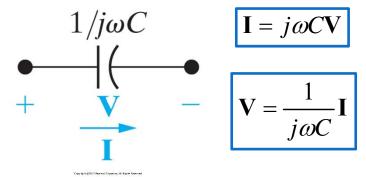
V-I Relationship for a Capacitor :



$$v = V_m \cos(\omega t + \phi_v)$$

$$i = C \frac{dv}{dt} = -\omega C V_m \sin(\omega t + \phi_v)$$

$$= -\omega C V_m \cos(\omega t + \phi_v - 90^\circ)$$



$$\mathbf{I} = -\omega C V_m e^{j(\phi_v - 90^\circ)} = -\omega C V_m e^{j\phi_v} e^{-j90^\circ}$$

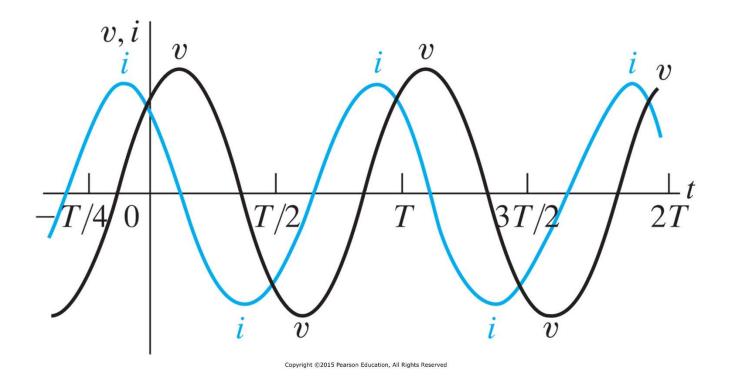
$$= j\omega C V_m e^{j\phi_i} = j\omega C \mathbf{V} = \omega C V_m \angle (\phi_v + 90^\circ)$$

$$\mathbf{V} = \frac{1}{i\omega C} \mathbf{I} = -\frac{j}{\omega C} \mathbf{I} = \frac{1}{\omega C} I_m \angle (\phi_i - 90^\circ)$$

 $\mathbf{V} = V_m e^{j\phi_V} = V_m \angle \phi_V$

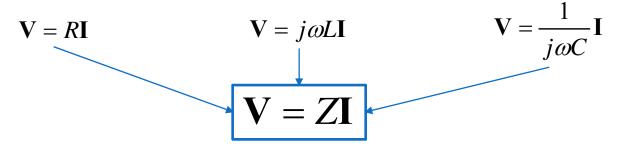
Current leads voltage by 90°.

V-I Relationship for a Capacitor (cont.):



Current leads voltage by 90°.

Impedance and Reactance:



- •Z is the impedance of the circuit element.
- •Unit for impedance is Ohm.
- •It is a complex number but NOT A PHASOR.
- •Impedance is defined in the frequency domain.
- •The imaginary part of the impedance is called Reactance.

 Qn. 3: The current on a 20mH inductor is given as 10cos(10000t+30°) mA.

What is the reactance of the inductor? What is the impedance of the inductor? What is the phasor voltage V on the inductor? Find the steady state voltage v(t) on the inductor.

• Qn. 3 (cont.):
$$i(t) = 10\cos(10000t + 30^\circ) \text{mA}$$

$$V = j\omega LI$$

$$\Rightarrow$$
 Reactance: $\omega L = 10000 \times 20 \times 10^{-3} = 200\Omega$

$$\Rightarrow$$
 Impedance: $Z = j\omega L = j10000 \times 20 \times 10^{-3} = j200\Omega$

$$\Rightarrow$$
 V = $j\omega L$ **I** = $j200 \times 10 \times 10^{-3} \angle 30^{\circ} = 2\angle 120^{\circ}$ V

$$\Rightarrow v(t) = 2\cos(10000t + 120^{\circ})V$$

 Qn. 4: The voltage on a 5µF capacitor is given as 30cos(4000t+25°) V.

What is the reactance of the capacitor? What is the impedance of the capacitor? What is the phasor current I on the capacitor? Find the steady state current i(t) on the capacitor.

• Qn. 4 (cont.):
$$v(t) = 30\cos(4000t + 25^{\circ})V$$

$$\mathbf{V} = -\frac{j}{\omega C}\mathbf{I} \Leftrightarrow \mathbf{I} = j\omega C\mathbf{V}$$

$$\Rightarrow$$
 Reactance: $-\frac{1}{\omega C} = -\frac{1}{4000 \times 5 \times 10^{-6}} = -50\Omega$

$$\Rightarrow$$
 Impedance: $Z = -\frac{j}{\omega C} = -\frac{j}{4000 \times 5 \times 10^{-6}} = -j50\Omega$

$$\Rightarrow$$
 I = $j\omega C$ **V** = $j0.02 \times 30 \angle 25^{\circ} = 0.6 \angle 115^{\circ}$ A

$$\Rightarrow i(t) = 0.6\cos(4000t + 115^{\circ})A$$

 Kirchoff Laws in Frequency Domain: Kirchoff Voltage Law

$$v_{1} + v_{2} + \dots + v_{n} = 0$$

$$V_{m_{1}} \cos(\omega t + \phi_{1}) + V_{m_{2}} \cos(\omega t + \phi_{2}) + \dots + V_{m_{n}} \cos(\omega t + \phi_{n}) = 0$$

$$\operatorname{Re} \{V_{m_{1}} e^{j\phi_{1}} e^{j\omega t}\} + \operatorname{Re} \{V_{m_{2}} e^{j\phi_{2}} e^{j\omega t}\} + \dots + \operatorname{Re} \{V_{m_{n}} e^{j\phi_{n}} e^{j\omega t}\} = 0$$

$$\operatorname{Re} \{V_{m_{1}} e^{j\phi_{1}} e^{j\omega t} + V_{m_{2}} e^{j\phi_{2}} e^{j\omega t} + \dots + V_{m_{n}} e^{j\phi_{n}} e^{j\omega t}\} = 0$$

$$\operatorname{Re} \{(V_{m_{1}} e^{j\phi_{1}} + V_{m_{2}} e^{j\phi_{2}} + \dots + V_{m_{n}} e^{j\phi_{n}}) e^{j\omega t}\} = 0$$

$$\operatorname{Re} \{(V_{1} + V_{2} + \dots + V_{n}) e^{j\omega t}\} = 0$$

$$V_{1} + V_{2} + \dots + V_{n} = 0$$

 Kirchoff Laws in Frequency Domain: Kirchoff Current Law

$$i_{1} + i_{2} + \dots + i_{n} = 0$$

$$I_{m_{1}} \cos(\omega t + \phi_{1}) + I_{m_{2}} \cos(\omega t + \phi_{2}) + \dots + I_{m_{n}} \cos(\omega t + \phi_{n}) = 0$$

$$\operatorname{Re}\left\{I_{m_{1}}e^{j\phi_{1}}e^{j\omega t}\right\} + \operatorname{Re}\left\{I_{m_{2}}e^{j\phi_{2}}e^{j\omega t}\right\} + \dots + \operatorname{Re}\left\{I_{m_{n}}e^{j\phi_{n}}e^{j\omega t}\right\} = 0$$

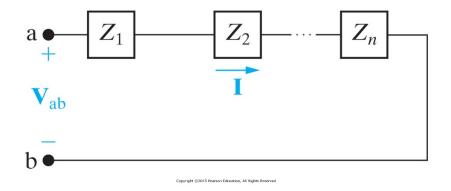
$$\operatorname{Re}\left\{I_{m_{1}}e^{j\phi_{1}}e^{j\omega t} + I_{m_{2}}e^{j\phi_{2}}e^{j\omega t} + \dots + I_{m_{n}}e^{j\phi_{n}}e^{j\omega t}\right\} = 0$$

$$\operatorname{Re}\left\{\left(I_{m_{1}}e^{j\phi_{1}} + I_{m_{2}}e^{j\phi_{2}} + \dots + I_{m_{n}}e^{j\phi_{n}}\right)e^{j\omega t}\right\} = 0$$

$$\operatorname{Re}\left\{\left(I_{1} + I_{2} + \dots + I_{n}\right)e^{j\omega t}\right\} = 0$$

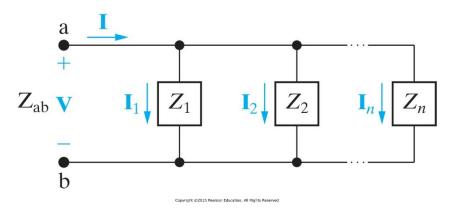
$$I_{1} + I_{2} + \dots + I_{n} = 0$$

Series and Parallel Impedances:



$$KVL : \mathbf{V}_{ab} = Z_1 \mathbf{I} + Z_2 \mathbf{I} + \dots + Z_n \mathbf{I}$$
$$= (Z_1 + Z_2 + \dots + Z_n) \mathbf{I}$$
$$Z_{ab} = \frac{\mathbf{V}_{ab}}{\mathbf{I}} = Z_1 + Z_2 + \dots + Z_n$$

Series and Parallel Impedances (cont.):

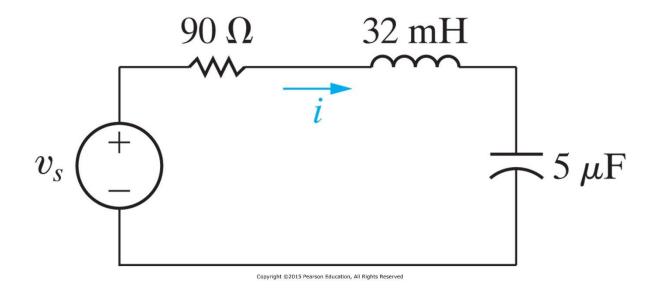


$$KCL : \mathbf{I} = \frac{\mathbf{V}}{Z_1} + \frac{\mathbf{V}}{Z_2} + \dots + \frac{\mathbf{V}}{Z_3}$$
$$= \left(\frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_3}\right) \mathbf{V}$$

$$\frac{1}{Z_{ab}} = \frac{\mathbf{I}}{\mathbf{V}} = \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_3}$$

• Qn. 5: $v_s = 750\cos(5000t + 30^\circ) \text{ V}$.

Draw the circuit in frequency domain. Find the steady state current i(t) using phasor concept.

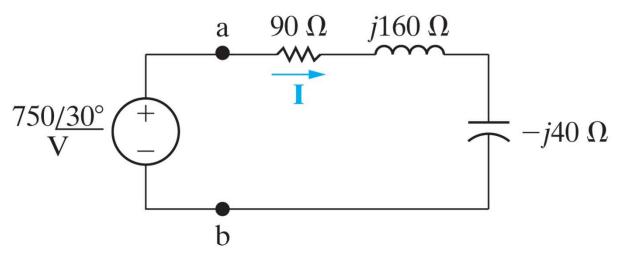


• Qn. 5 (cont.): $v_s(t) = 750\cos(5000t + 30^\circ)V \Rightarrow V = 750\angle 30^\circ V$

$$Z_R = R = 90\Omega$$

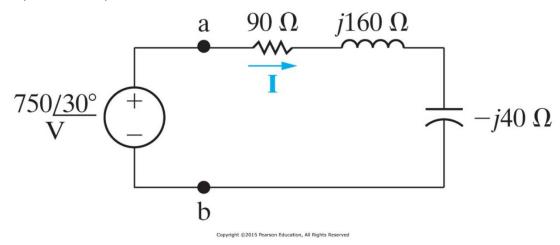
$$Z_L = j\omega L = j5000 \times 32 \times 10^{-3} = j160\Omega$$

$$Z_C = -\frac{j}{\omega C} = -\frac{j}{5000 \times 5 \times 10^{-6}} = -j40\Omega$$



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Qn. 5 (cont.):

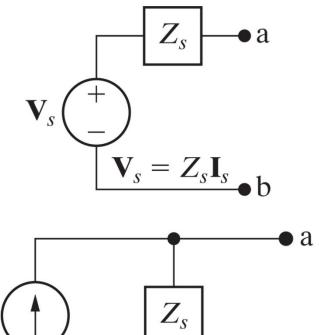


$$KVL: \mathbf{V} = Z_1 \mathbf{I} + Z_2 \mathbf{I} + Z_3 \mathbf{I} = (90 + j160 - j40)\mathbf{I}$$

 $\mathbf{V} = (90 + j120)\mathbf{I}$
 $750 \angle 30^\circ = 150 \angle 53.13^\circ \mathbf{I}$

$$\Rightarrow \mathbf{I} = \frac{750 \angle 30^{\circ}}{150 \angle 53.13^{\circ}} = 5 \angle -23.13^{\circ} \mathbf{A} \qquad \Rightarrow i(t) = 5\cos(5000t - 23.13^{\circ}) \mathbf{A}$$

Source Transformation:



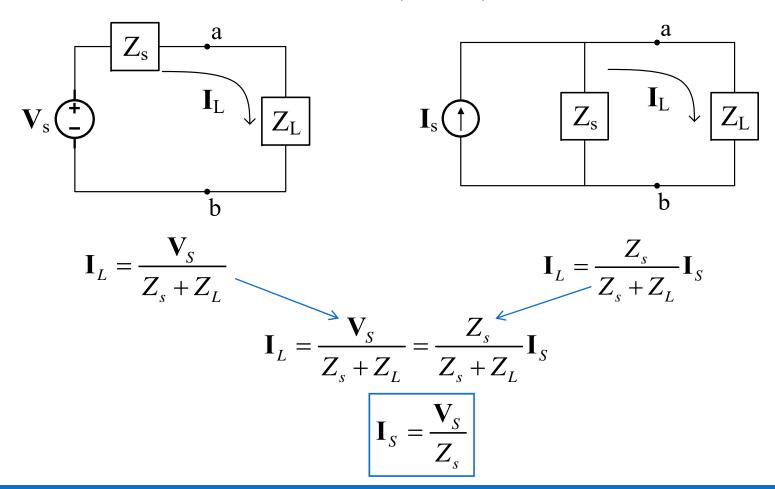
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The circuits on the left are equivalent in time domain.

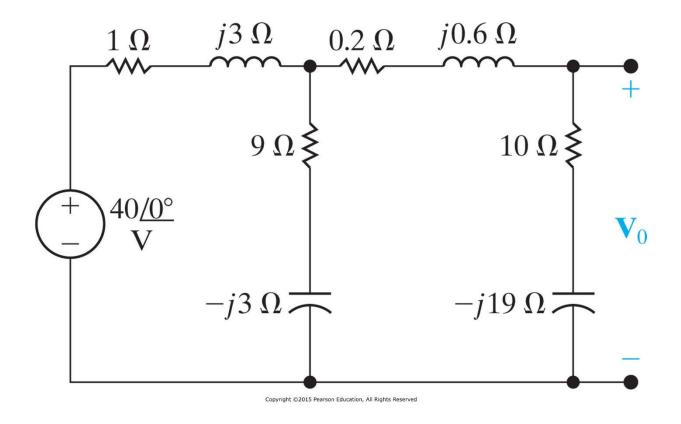
It can also be shown that they are equivalent in the frequency domain also.

The voltage and current in the terminals ab must be equal in order to say that these circuits are equivalent!!!

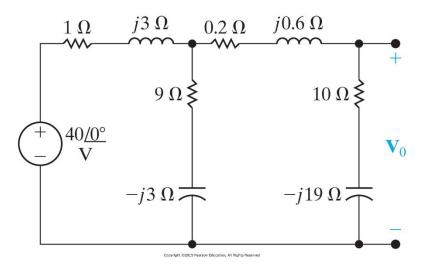
• Source Transformation (cont.):



• Qn. 6: Find V_0 using source transformation.

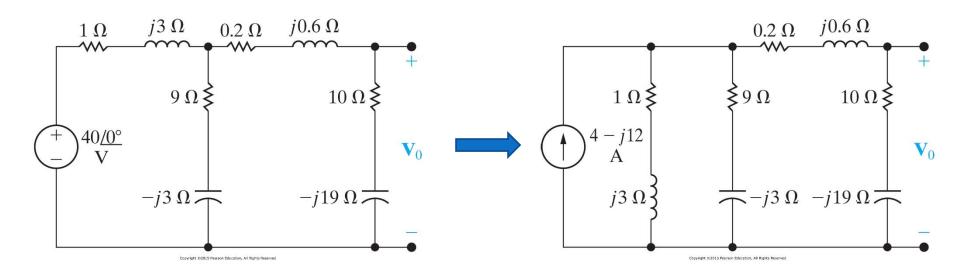


• Qn. 6 (cont.):



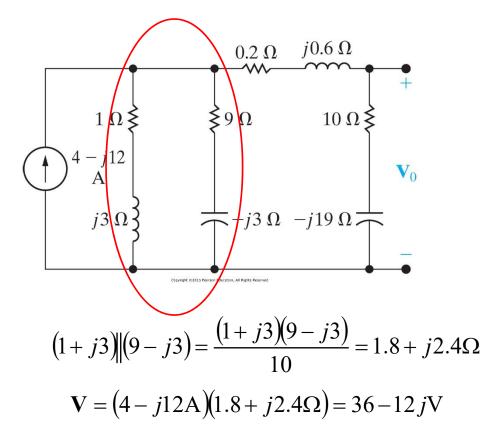
$$\mathbf{I} = \frac{40 \angle 0^{\circ}}{1+j3} = \frac{40}{(1+j3)(1-j3)} (1-j3) = 4-j12A$$

• Qn. 6 (cont.):

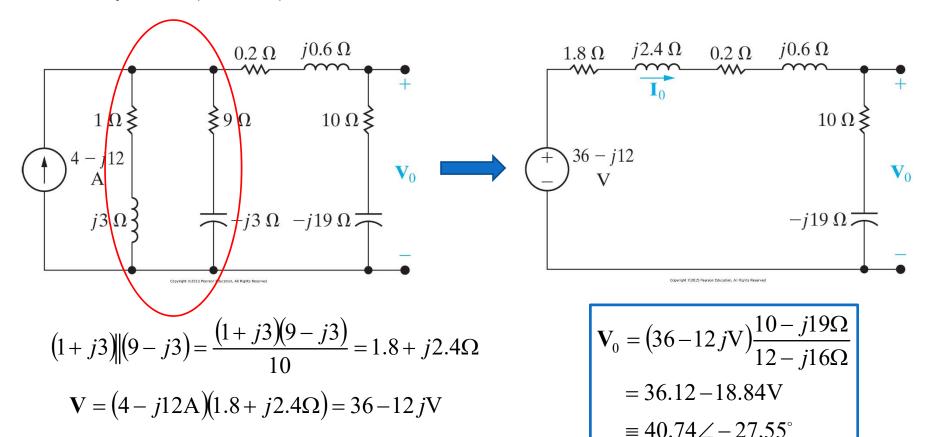


$$\mathbf{I} = \frac{40 \angle 0^{\circ}}{1+j3} = \frac{40}{(1+j3)(1-j3)} (1-j3) = 4-j12A$$

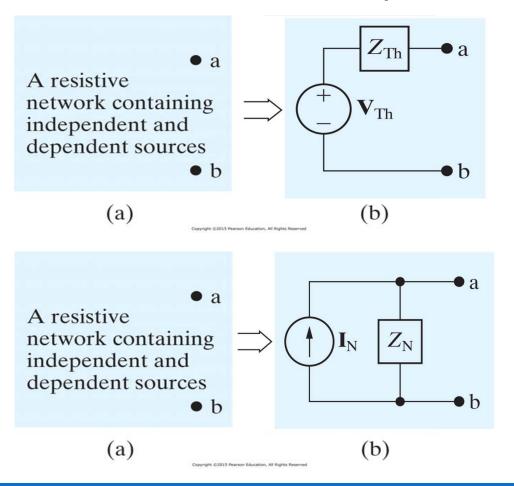
• Qn. 6 (cont.):



Qn. 6 (cont.):

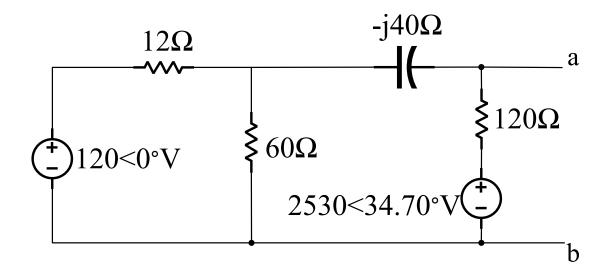


• Thévenin-Norton Equivalent Circuits:

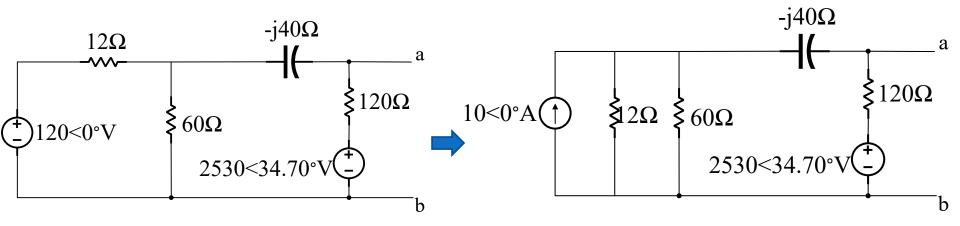


- •1st Step: Find open circuit voltage (V_{th}).
- •2nd Step: Find short circuit current (I_{sc}).
- •3rd Step: $R_{\rm th} = V_{\rm th} / I_{\rm sc}$
- •4th Step (Optional): Find Norton equivalent using source transformation.

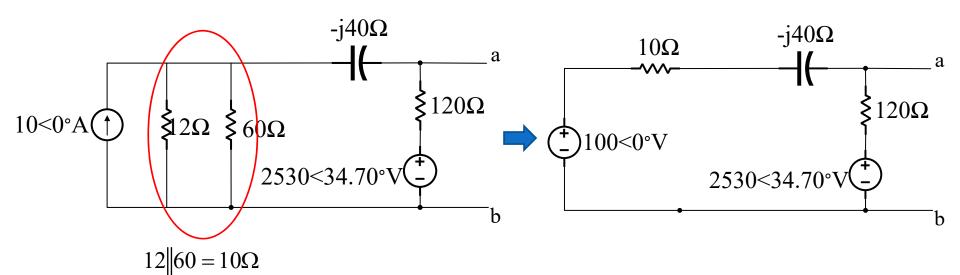
 Qn. 7: Find Thévenin equivalent w.r.t. terminals ab.



• Qn. 7 (cont.):



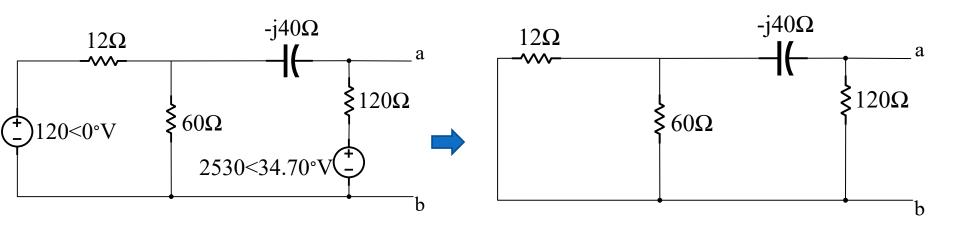
Qn. 7 (cont.):



$$\mathbf{I} = \frac{100 \angle 0^{\circ} - 2530 \angle 34.70^{\circ}}{130 - j40} = -10.8 - j14.4A$$

$$\mathbf{V}_{ab} = \mathbf{V}_{Th} = 2350 \angle 34.70^{\circ} + 120(-10.8 - j14.4)$$
$$= 835.22 \angle -20.17^{\circ}$$

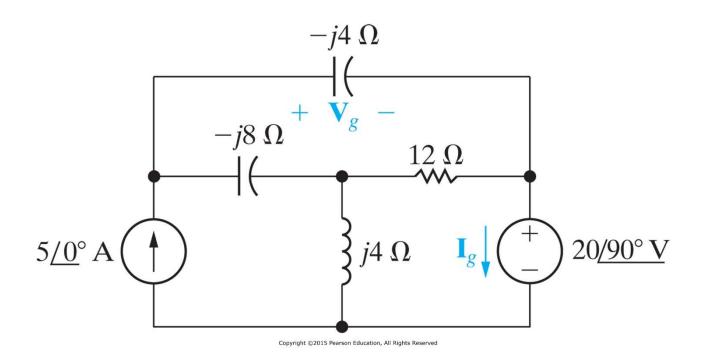
• Qn. 7 (cont.):



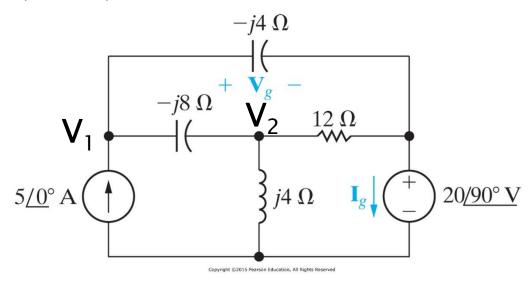
$$Z_{ab} = Z_{Th} = (12||60 + j40)||120 = 18.81 - j31.14\Omega$$

- Node Voltage Method:
 - 1. Define essential nodes
 - 2. Pick a reference node
 - 3. Give names to remaining essential nodes
 - 4. Solve KCL for each essential node
 - 5. Check the power balance

• Qn. 8: Find the phasor V_g using node voltage method.

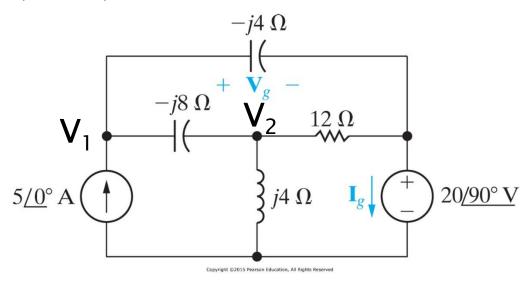


Qn. 8 (cont.):



$$-5\angle 0^{\circ} + \frac{\mathbf{V}_{1} - \mathbf{V}_{2}}{-j8} + \frac{\mathbf{V}_{1} - 20\angle 90^{\circ}}{-j4} = 0$$
$$\frac{\mathbf{V}_{2} - \mathbf{V}_{1}}{-j8} + \frac{\mathbf{V}_{2}}{j4} + \frac{\mathbf{V}_{2} - 20\angle 90^{\circ}}{12} = 0$$

Qn. 8 (cont.):



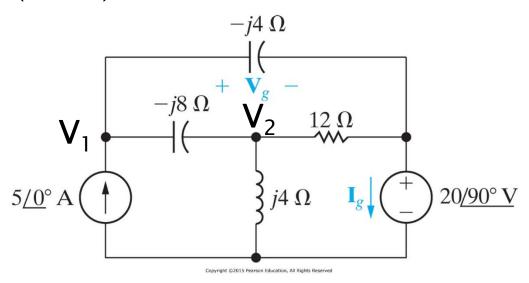
$$-5\angle 0^{\circ} + \frac{\mathbf{V}_{1} - \mathbf{V}_{2}}{-j8} + \frac{\mathbf{V}_{1} - 20\angle 90^{\circ}}{-j4} = 0$$

$$\frac{\mathbf{V}_{2} - \mathbf{V}_{1}}{-j8} + \frac{\mathbf{V}_{2}}{j4} + \frac{\mathbf{V}_{2} - 20\angle 90^{\circ}}{12} = 0$$

$$\Rightarrow \mathbf{V}_{1} = -2.67 + j1.33V$$

$$\mathbf{V}_{2} = -8 + j4V$$

Qn. 8 (cont.):



$$-5 \angle 0^{\circ} + \frac{\mathbf{V}_{1} - \mathbf{V}_{2}}{-j8} + \frac{\mathbf{V}_{1} - 20 \angle 90^{\circ}}{-j4} = 0$$

$$\frac{\mathbf{V}_{2} - \mathbf{V}_{1}}{-j8} + \frac{\mathbf{V}_{2}}{j4} + \frac{\mathbf{V}_{2} - 20 \angle 90^{\circ}}{12} = 0$$

$$\Rightarrow \mathbf{V}_{1} = -2.67 + j1.33V$$

$$\mathbf{V}_{2} = -8 + j4V$$

$$\Rightarrow \frac{\mathbf{V}_1 = -2.67 + j1.33 \text{V}}{\mathbf{V}_2 = -8 + j4 \text{V}}$$

$$\mathbf{V}_{g} = \mathbf{V}_{1} - 20 \angle 90^{\circ}$$

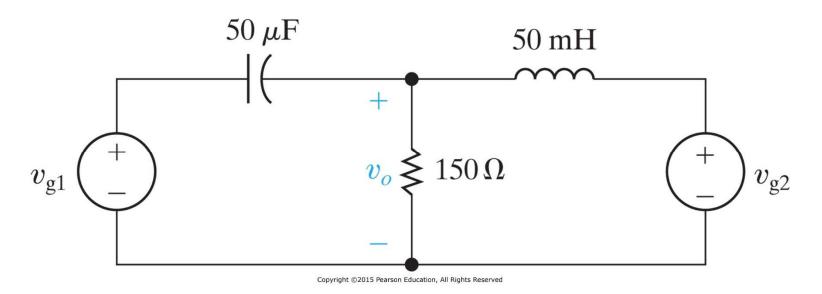
$$= -2.67 - j18.67 \text{V}$$

$$= 18.86 \angle -98.14^{\circ} \text{V}$$

why -98.14 but not 81.86 degrees?

- Mesh Current Method:
 - 1. Define meshes
 - 2. Define currents to each mesh (son't forget to define the directions)
 - 3. Solve KVL equations for each mesh
 - 4. Check the power balance

• Qn. 9: Find the steady state value $v_o(t)$ using mesh current method.

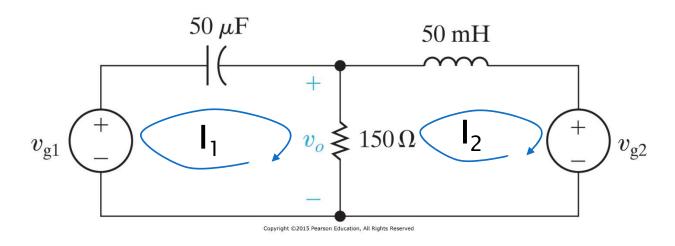


$$v_{g1}(t) = 25\sin(400t + 143.13^{\circ})V$$

 $v_{g2}(t) = 18.03\cos(400t + 33.69^{\circ})V$

• Qn. 9 (cont.):

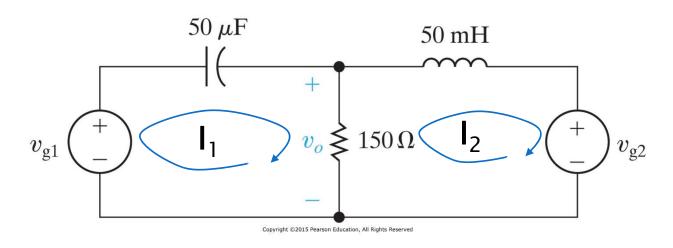
$$v_{g1}(t) = 25\sin(400t + 143.13^{\circ})V$$
 $V_{g1} = 25\angle 53.13^{\circ}V$
 $v_{g2}(t) = 18.03\cos(400t + 33.69^{\circ})V$ $V_{g2} = 18.03\angle 33.69^{\circ}V$



$$-25 \angle 53.13^{\circ} + (-j50)\mathbf{I}_{1} + 150(\mathbf{I}_{1} - \mathbf{I}_{2}) = 0$$
$$18.03 \angle 33.69^{\circ} + 150(\mathbf{I}_{2} - \mathbf{I}_{1}) + j20\mathbf{I}_{2} = 0$$

Qn. 9 (cont.):

$$v_{g1}(t) = 25\sin(400t + 143.13^{\circ})V$$
 $V_{g1} = 25\angle 53.13^{\circ}V$
 $v_{g2}(t) = 18.03\cos(400t + 33.69^{\circ})V$ $V_{g2} = 18.03\angle 33.69^{\circ}V$



$$-25\angle 53.13^{\circ} + (-j50)\mathbf{I}_{1} + 150(\mathbf{I}_{1} - \mathbf{I}_{2}) = 0$$

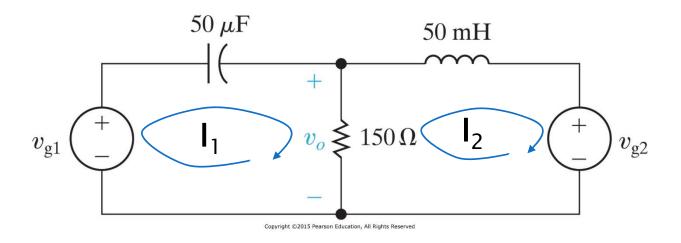
$$18.03\angle 33.69^{\circ} + 150(\mathbf{I}_{2} - \mathbf{I}_{1}) + j20\mathbf{I}_{2} = 0$$

$$\Rightarrow \mathbf{I}_{1} = -0.4A$$

$$\mathbf{I}_{2} = -0.5A$$

• Qn. 9 (cont.):

$$v_{g1}(t) = 25\sin(400t + 143.13^{\circ})V$$
 $V_{g1} = 25\angle 53.13^{\circ}V$
 $v_{g2}(t) = 18.03\cos(400t + 33.69^{\circ})V$ $V_{g2} = 18.03\angle 33.69^{\circ}V$



$$\frac{-25\angle 53.13^{\circ} + (-j50)\mathbf{I}_{1} + 150(\mathbf{I}_{1} - \mathbf{I}_{2}) = 0}{18.03\angle 33.69^{\circ} + 150(\mathbf{I}_{2} - \mathbf{I}_{1}) + j20\mathbf{I}_{2} = 0} \Rightarrow \frac{\mathbf{I}_{1}}{\mathbf{I}_{2}}$$

$$\mathbf{V}_{o} = 150(\mathbf{I}_{1} - \mathbf{I}_{2}) = 15V$$

 $v_{o}(t) = 15\cos(400t)V$

- We can use each method that we learned in time domain in the frequency domain:
 - Kirchoff Laws.
 - Source Transformation.
 - Thévenin-Norton Equivalents.
 - Node Voltage and Mesh Current Methods.

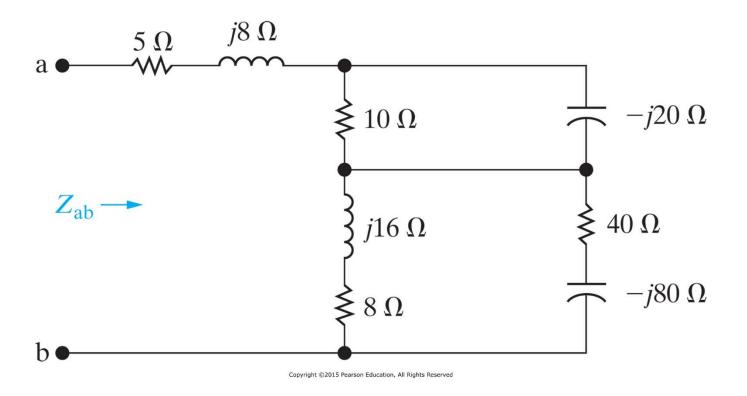
Moreover, the methods below are also applicable. You are responsible to solve questions for the following methods in the frequency domain.

- Superposition.
- Current and Voltage Divisions.

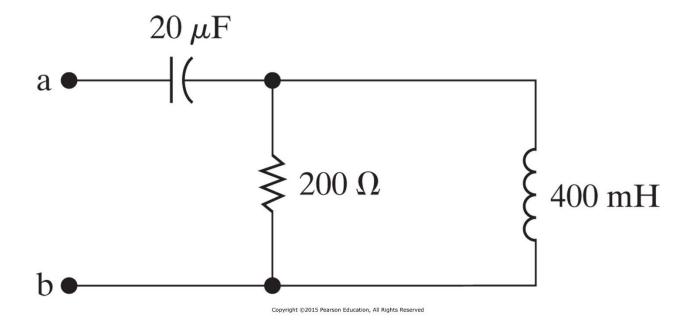
Course Content

- Circuit Components and Variables, Ohm's Law and Kirchoff's Laws
- Circuit Analysis Techniques I Node Voltage & Mesh Current Methods
- Circuit Analysis Techniques II Thevenin and Norton Equivalent Circuits
- First Order RL and RC Circuits
- Phasor Concept and AC Analysis of Circuits
- Introduction to Semiconductors
- Midterm
- BJT DC Analysis
- BJT AC Analysis
- FET Analysis
- Operational Amplifier
- Logic Gates
- Make-up Exams
- Selected Circuits

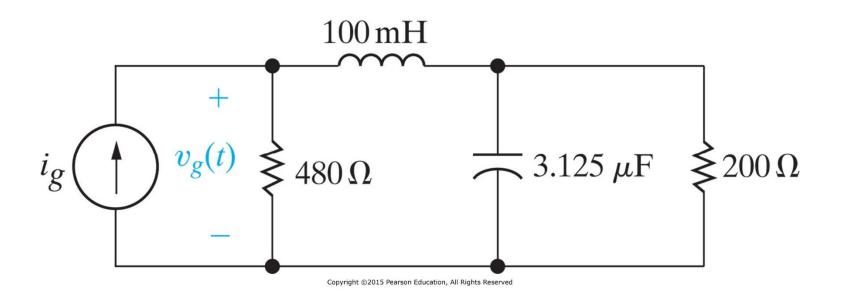
• Qn. 10: Find the equivalent impedance w.r.t. terminals ab.



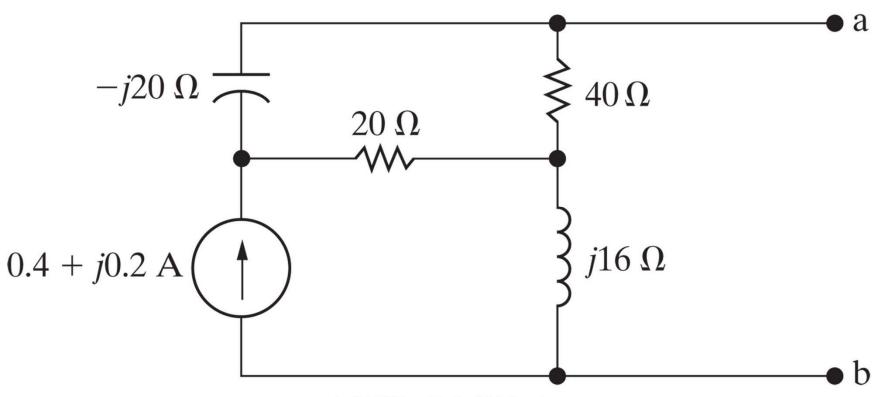
• Qn. 11: What should be the frequency for the given circuit to be fully resistive?



 Qn. 12: What should be the frequency for the voltage v_g and the current i_g to be in-phase?



 Qn. 13: Find the Thevenin equivalent w.r.t. terminals ab.



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 Qn. 14: Find the shown currents using mesh current method.

