

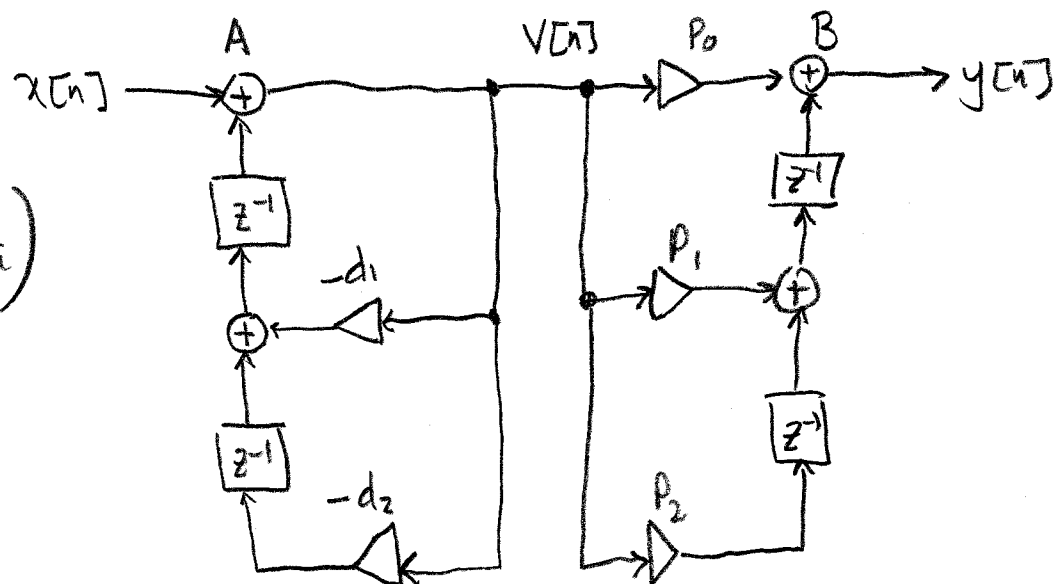
ECE 4213/5213

DSP

HW 4 SOLUTION

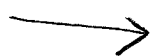
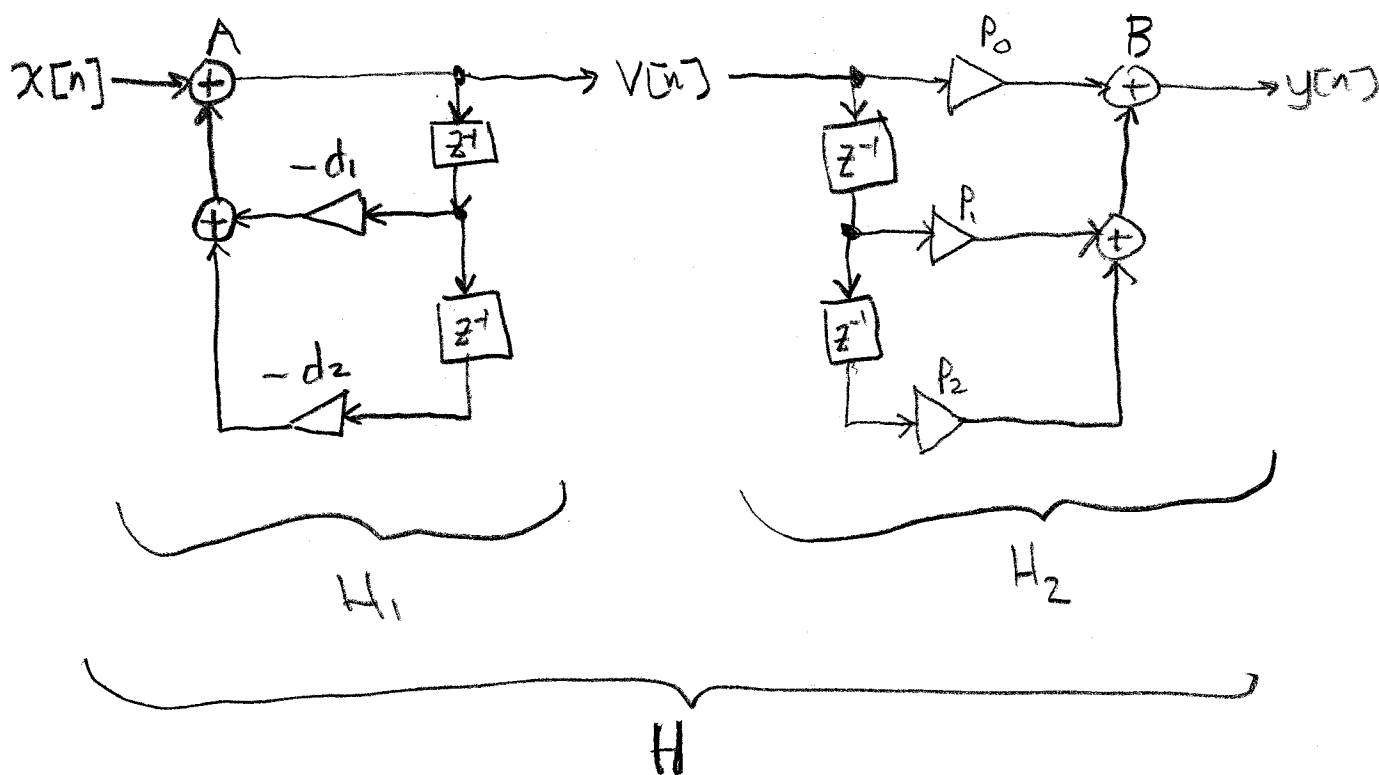
HAVLICEK

2.4a)



2.4a)-1

Move the delays and break into two systems;



- for system H_1 , the input is $x[n]$ and the output is $v[n]$.

2.4a)-z

At node A, we have:

$$v[n] = x[n] - d_1 v[n-1] - d_2 v[n-2]$$

$$v[n] + d_1 v[n-1] + d_2 v[n-2] = x[n]$$

z-transform: $V(z) + d_1 z^{-1} V(z) + d_2 z^{-2} V(z) = X(z)$

$$V(z) [1 + d_1 z^{-1} + d_2 z^{-2}] = X(z)$$

$$H_1(z) = \frac{V(z)}{X(z)} = \frac{1}{1 + d_1 z^{-1} + d_2 z^{-2}}$$

- for system H_2 , the input is $v[n]$ and the output is $y[n]$.

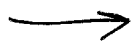
At node B, we have:

$$y[n] = p_0 v[n] + p_1 v[n-1] + p_2 v[n-2]$$

z-transform: $Y(z) = p_0 V(z) + p_1 z^{-1} V(z) + p_2 z^{-2} V(z)$

$$= [p_0 + p_1 z^{-1} + p_2 z^{-2}] V(z)$$

$$H_2(z) = \frac{Y(z)}{V(z)} = p_0 + p_1 z^{-1} + p_2 z^{-2}$$



The overall system H is a series connection of H_1 and H_2 , so

2.4a)-3

$$H(z) = H_1(z)H_2(z) = \frac{p_0 + p_1 z^{-1} + p_2 z^{-2}}{1 + d_1 z^{-1} + d_2 z^{-2}} = \frac{Y(z)}{X(z)}$$

Cross multiply:

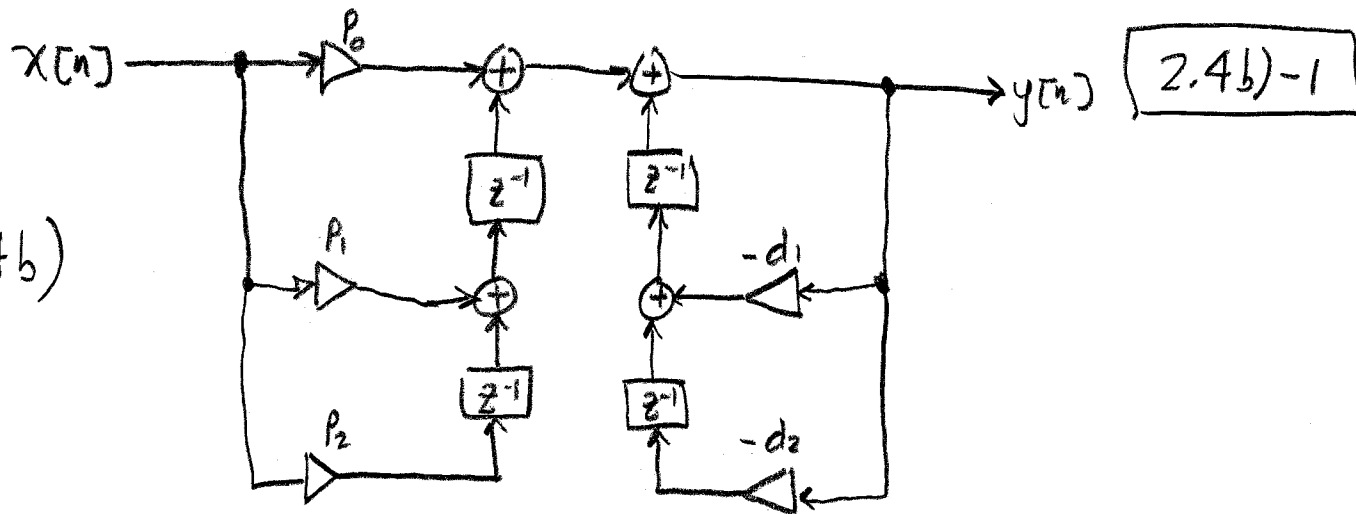
$$Y(z) [1 + d_1 z^{-1} + d_2 z^{-2}] = X(z) [p_0 + p_1 z^{-1} + p_2 z^{-2}]$$

$$Y(z) + d_1 z^{-1} Y(z) + d_2 z^{-2} Y(z) = p_0 X(z) + p_1 z^{-1} X(z) + p_2 z^{-2} X(z)$$

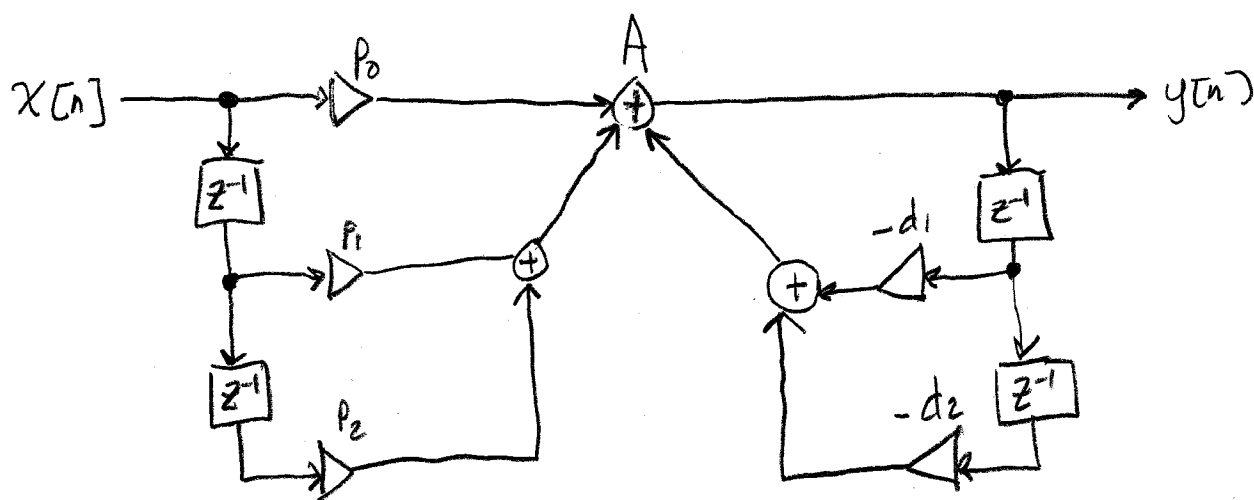
inverse z transform:

$$y[n] + d_1 y[n-1] + d_2 y[n-2] = p_0 x[n] + p_1 x[n-1] + p_2 x[n-2]$$

2.4b)



Commute the delays and multipliers and combine the two adders in the middle:

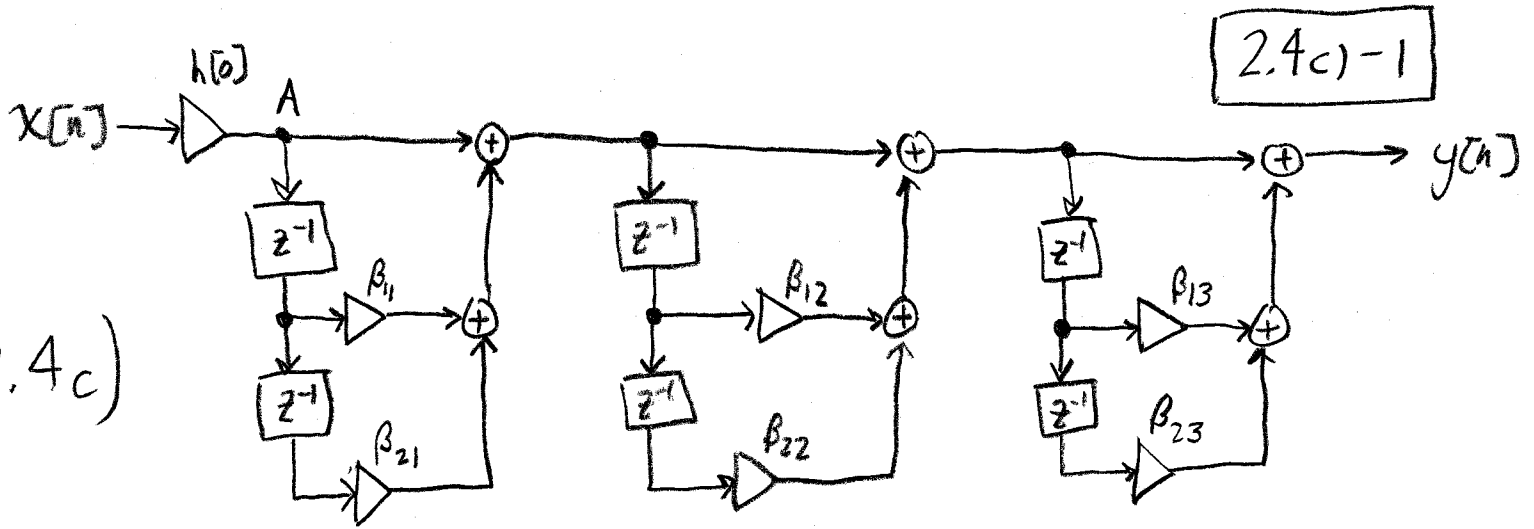


at node A, we have:

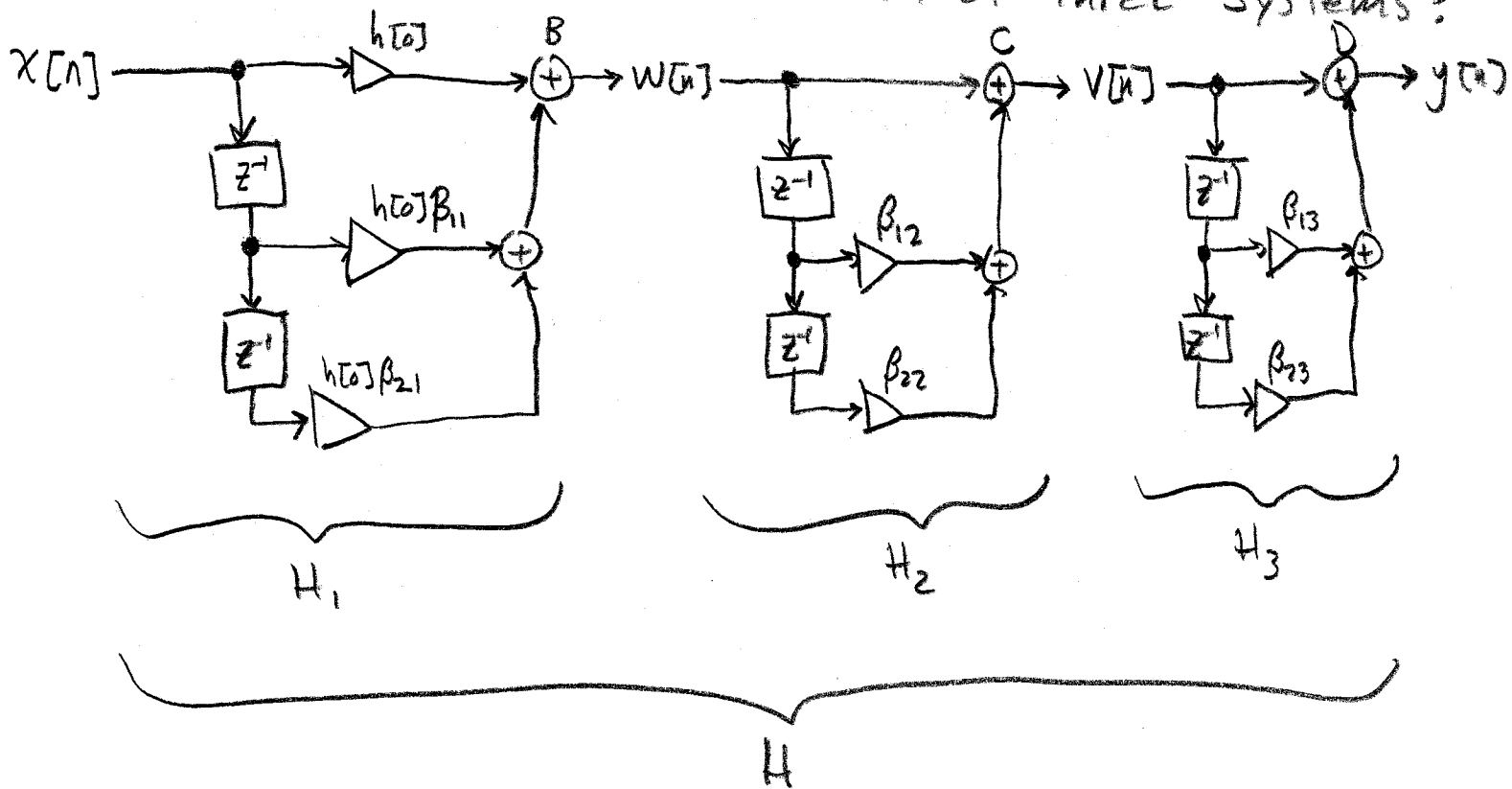
$$y[n] = p_0 x[n] + p_1 x[n-1] + p_2 x[n-2] - d_1 y[n-1] - d_2 y[n-2]$$

$$y[n] + d_1 y[n-1] + d_2 y[n-2] = p_0 x[n] + p_1 x[n-1] + p_2 x[n-2]$$

2.4c)



- Push the multiplication by $h[0]$ through node A and consider as a series connection of three systems:



For system H_1 , the input is $x[n]$ and the output is $w[n]$.
At node B, we have

$$w[n] = h[0]x[n] + h[0]\beta_{11}x[n-1] + h[0]\beta_{21}x[n-2]$$



Apply z-transform:

2.4c) - 2

$$W(z) = h[0] X(z) + h[0] \beta_{11} z^{-1} X(z) + h[0] \beta_{21} z^{-2} X(z)$$

$$W(z) = [h[0] + h[0] \beta_{11} z^{-1} + h[0] \beta_{21} z^{-2}] X(z)$$

$$H_1(z) = \frac{W(z)}{X(z)} = h[0] + h[0] \beta_{11} z^{-1} + h[0] \beta_{21} z^{-2}$$

For system H_2 , the input is $w[n]$ and the output is $v[n]$.

At node C, we have

$$v[n] = w[n] + \beta_{12} w[n-1] + \beta_{22} w[n-2]$$

z-transform: $V(z) = W(z) + \beta_{12} z^{-1} W(z) + \beta_{22} z^{-2} W(z)$

$$V(z) = [1 + \beta_{12} z^{-1} + \beta_{22} z^{-2}] W(z)$$

$$H_2(z) = \frac{V(z)}{W(z)} = 1 + \beta_{12} z^{-1} + \beta_{22} z^{-2}$$

For system H_3 , the input is $v[n]$ and the output is $y[n]$.

- The diagram for H_3 is the same as the one for H_2 .

Using the above result for H_2 , we have

$$H_3(z) = \frac{Y(z)}{V(z)} = 1 + \beta_{13} z^{-1} + \beta_{23} z^{-2}$$

The overall system H is the series connection of H_1 , H_2 , and H_3 . So, we have



$$H(z) = H_1(z) H_2(z) H_3(z)$$

$$\boxed{2.4c)-3}$$

$$= \left[h[0] + h[0]\beta_{11}z^{-1} + h[0]\beta_{21}z^{-2} \right] \left[1 + \beta_{12}z^{-1} + \beta_{22}z^{-2} \right] \left[1 + \beta_{13}z^{-1} + \beta_{23}z^{-2} \right]$$

$$= \left[h[0] + h[0]\beta_{12}z^{-1} + h[0]\beta_{22}z^{-2} + h[0]\beta_{11}z^{-1} + h[0]\beta_{11}\beta_{12}z^{-2} + h[0]\beta_{11}\beta_{22}z^{-3} \right. \\ \left. + h[0]\beta_{21}z^{-2} + h[0]\beta_{21}\beta_{12}z^{-3} + h[0]\beta_{21}\beta_{22}z^{-4} \right] \\ \times \left[1 + \beta_{13}z^{-1} + \beta_{23}z^{-2} \right]$$

$$= \left\{ h[0] + (h[0]\beta_{12} + h[0]\beta_{11})z^{-1} + (h[0]\beta_{22} + h[0]\beta_{11}\beta_{12} + h[0]\beta_{21})z^{-2} \right. \\ \left. + (h[0]\beta_{11}\beta_{22} + h[0]\beta_{21}\beta_{12})z^{-3} + h[0]\beta_{21}\beta_{22}z^{-4} \right\} \left[1 + \beta_{13}z^{-1} + \beta_{23}z^{-2} \right]$$

$$= h[0] + (h[0]\beta_{12} + h[0]\beta_{11})z^{-1} + (h[0]\beta_{22} + h[0]\beta_{11}\beta_{12} + h[0]\beta_{21})z^{-2} \\ + (h[0]\beta_{11}\beta_{22} + h[0]\beta_{21}\beta_{12})z^{-3} + h[0]\beta_{21}\beta_{22}z^{-4} \\ + h[0]\beta_{13}z^{-1} + (h[0]\beta_{12}\beta_{13} + h[0]\beta_{11}\beta_{13})z^{-2} \\ + (h[0]\beta_{22}\beta_{13} + h[0]\beta_{11}\beta_{12}\beta_{13} + h[0]\beta_{21}\beta_{13})z^{-3} \\ + (h[0]\beta_{11}\beta_{22}\beta_{13} + h[0]\beta_{21}\beta_{12}\beta_{13})z^{-4} + h[0]\beta_{21}\beta_{22}\beta_{13}z^{-5} \\ + h[0]\beta_{23}z^{-2} + (h[0]\beta_{12}\beta_{23} + h[0]\beta_{11}\beta_{23})z^{-3} \\ + (h[0]\beta_{22}\beta_{23} + h[0]\beta_{11}\beta_{12}\beta_{23} + h[0]\beta_{21}\beta_{23})z^{-4} \\ + (h[0]\beta_{11}\beta_{22}\beta_{23} + h[0]\beta_{21}\beta_{12}\beta_{23})z^{-5} + h[0]\beta_{21}\beta_{22}\beta_{23}z^{-6}$$

→

$$\dots H(z) = h[0] + (h[0]\beta_{12} + h[0]\beta_{11} + h[0]\beta_{13}) z^{-1}$$

$$\boxed{2.4c) - 4}$$

$$+ (h[0]\beta_{22} + h[0]\beta_{11}\beta_{12} + h[0]\beta_{21} + h[0]\beta_{12}\beta_{13} + h[0]\beta_{11}\beta_{13} + h[0]\beta_{23}) z^{-2}$$

$$+ (h[0]\beta_{11}\beta_{22} + h[0]\beta_{21}\beta_{12} + h[0]\beta_{22}\beta_{13} + h[0]\beta_{11}\beta_{12}\beta_{13} + h[0]\beta_{21}\beta_{13} \\ + h[0]\beta_{12}\beta_{23} + h[0]\beta_{11}\beta_{23}) z^{-3}$$

$$+ (h[0]\beta_{21}\beta_{22} + h[0]\beta_{11}\beta_{22}\beta_{13} + h[0]\beta_{21}\beta_{12}\beta_{13} + h[0]\beta_{22}\beta_{23} + h[0]\beta_{11}\beta_{12}\beta_{23} \\ + h[0]\beta_{21}\beta_{23}) z^{-4}$$

$$+ (h[0]\beta_{21}\beta_{22}\beta_{13} + h[0]\beta_{11}\beta_{22}\beta_{23} + h[0]\beta_{21}\beta_{12}\beta_{23}) z^{-5}$$

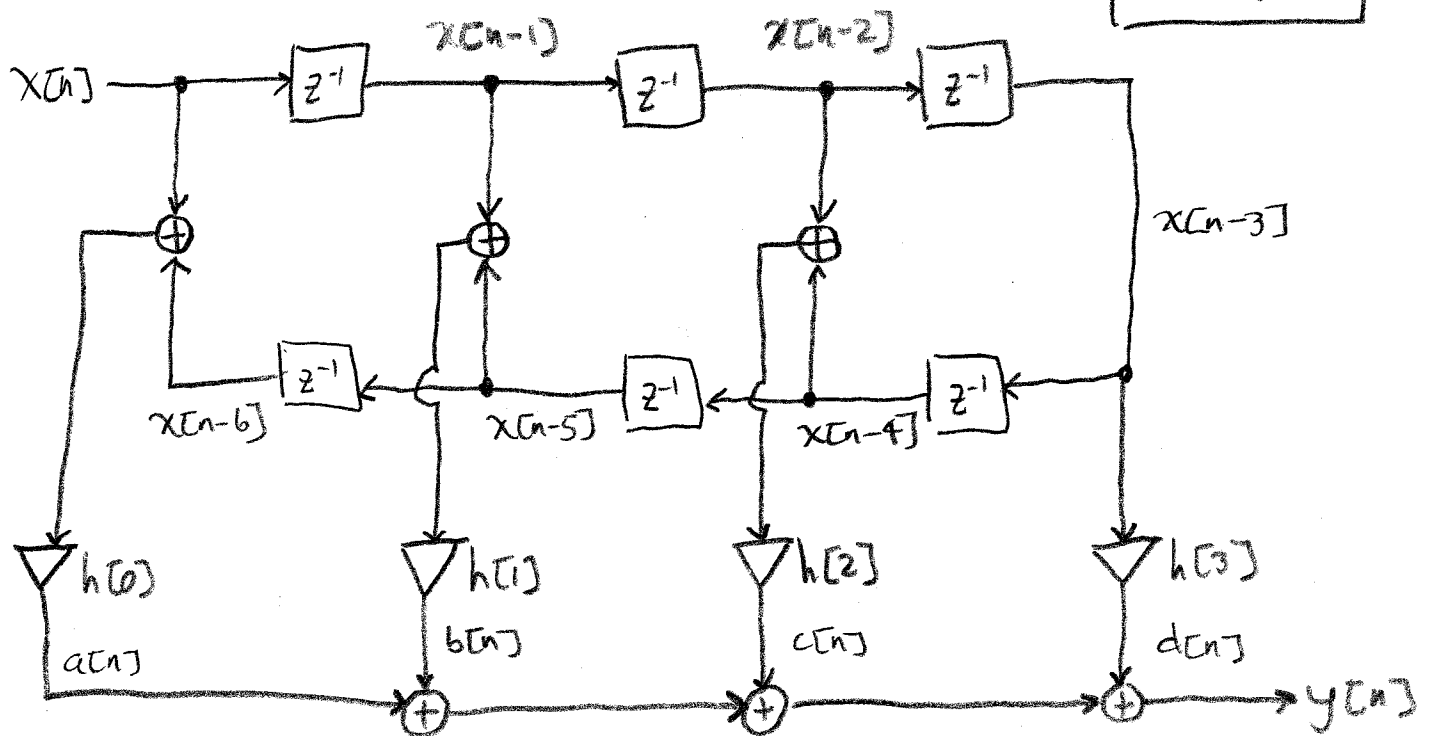
$$+ h[0]\beta_{21}\beta_{22}\beta_{23} z^{-6}$$

Note that $H(z) = \frac{Y(z)}{X(z)}$. Cross multiplying with the above expression for $H(z)$ and taking inverse z -transform, we have:

$$y[n] = h[0]x[n] + h[0](\beta_{12} + \beta_{11} + \beta_{13})x[n-1] \\ + h[0](\beta_{22} + \beta_{11}\beta_{12} + \beta_{21} + \beta_{12}\beta_{13} + \beta_{11}\beta_{13} + \beta_{23})x[n-2] \\ + h[0](\beta_{11}\beta_{22} + \beta_{21}\beta_{12} + \beta_{22}\beta_{13} + \beta_{11}\beta_{12}\beta_{13} + \beta_{21}\beta_{13} + \beta_{12}\beta_{23} + \beta_{11}\beta_{23})x[n-3] \\ + h[0](\beta_{21}\beta_{22} + \beta_{11}\beta_{22}\beta_{13} + \beta_{21}\beta_{12}\beta_{13} + \beta_{22}\beta_{23} + \beta_{11}\beta_{12}\beta_{23} + \beta_{21}\beta_{23})x[n-4] \\ + h[0](\beta_{21}\beta_{22}\beta_{13} + \beta_{11}\beta_{22}\beta_{23} + \beta_{21}\beta_{12}\beta_{23})x[n-5] \\ + h[0]\beta_{21}\beta_{22}\beta_{23}x[n-6]$$

2.4d)

2.4d)-1



$$a[n] = h[0](x[n] + x[n-6])$$

$$b[n] = h[1](x[n-1] + x[n-5])$$

$$c[n] = h[2](x[n-2] + x[n-4])$$

$$d[n] = h[3]x[n-3]$$

$$y[n] = a[n] + b[n] + c[n] + d[n]$$

$$y[n] = h[0](x[n] + x[n-6]) + h[1](x[n-1] + x[n-5]) + h[2](x[n-2] + x[n-4]) + h[3]x[n-3]$$

2.21a)

2.21a)-1

n	< -2	-2	-1	0	1	2	> 2
$x_1[n]$	0	$-1+j3$	$2-j7$	$4-j5$	$3+j5$	$-2-j$	0
$x_1^*[-n]$	0	$-2+j$	$3-j5$	$4+j5$	$2+j7$	$-1-j3$	0

$$x_{1cs}[n] = \frac{1}{2}(x_1[n] + x_1^*[-n])$$

$$= \frac{1}{2}\{-3+j4 \quad 5-j12 \quad 8 \quad 5+j12 \quad -3-j4\}, -2 \leq n \leq 2$$

$$= \left\{-\frac{3}{2}+j2 \quad \frac{5}{2}-j6 \quad 4 \quad \frac{5}{2}+j6 \quad -\frac{3}{2}-j2\right\}, -2 \leq n \leq 2$$

$$x_{1ca}[n] = \frac{1}{2}(x_1[n] - x_1^*[-n])$$

$$= \frac{1}{2}\{1+j2 \quad -1-j2 \quad -j10 \quad 1-j2 \quad -1+j2\}, -2 \leq n \leq 2$$

$$= \left\{\frac{1}{2}+j \quad -\frac{1}{2}-j \quad -j5 \quad \frac{1}{2}-j \quad -\frac{1}{2}+j\right\}, -2 \leq n \leq 2$$

$$2.21b) \quad x_2[n] = e^{j2\pi n/5} + e^{j\pi n/3}$$

$$2.21b) - 1$$

$$x_2^*[n] = e^{-j2\pi n/5} + e^{-j\pi n/3}$$

$$x_2^*[-n] = e^{j2\pi n/5} + e^{j\pi n/3}$$

$$\begin{aligned} x_{2cs}[n] &= \frac{1}{2} (x_2[n] + x_2^*[-n]) \\ &= \frac{1}{2} (e^{j2\pi n/5} + e^{j\pi n/3} + e^{j2\pi n/5} + e^{j\pi n/3}) \\ &= \frac{1}{2} (2e^{j2\pi n/5} + 2e^{j\pi n/3}) \\ &= e^{j2\pi n/5} + e^{j\pi n/3} \end{aligned}$$

NOTE: in this case, $x_2[n] = x_{2cs}[n]$, since $x_2[n]$ was already conjugate symmetric to start with.

$$\begin{aligned} x_{2ca}[n] &= \frac{1}{2} (x_2[n] - x_2^*[-n]) \\ &= \frac{1}{2} (e^{j2\pi n/5} + e^{j\pi n/3} - e^{j2\pi n/5} - e^{j\pi n/3}) \\ &= \underline{\underline{0}} \end{aligned}$$

NOTE: in this case, $x_{2ca}[n]$ is zero because $x_2[n]$ was already conjugate symmetric to start with; e.g., the conjugate antisymmetric part is identically zero.

$$2.21c) \quad x_3[n] = j\cos(2\pi n/7) - \sin(2\pi n/4)$$

$$\boxed{2.21c)-1}$$

$$x_3^*[n] = -j\cos(2\pi n/7) - \sin(2\pi n/4)$$

$$\begin{aligned} x_3^*[-n] &= -j\cos(-2\pi n/7) - \sin(-2\pi n/4) \\ &= -j\cos(2\pi n/7) + \sin(2\pi n/4) \end{aligned}$$

$$\begin{aligned} x_{3cs}[n] &= \frac{1}{2} \{ x_3[n] + x_3^*[-n] \} \\ &= \frac{1}{2} \{ j\cos(2\pi n/7) - \sin(2\pi n/4) - j\cos(2\pi n/7) + \sin(2\pi n/4) \} \\ &= \underline{\underline{0}} \end{aligned}$$

Note: $x_3[n]$ was conjugate antisymmetric to start with, so the conjugate symmetric part is identically zero.

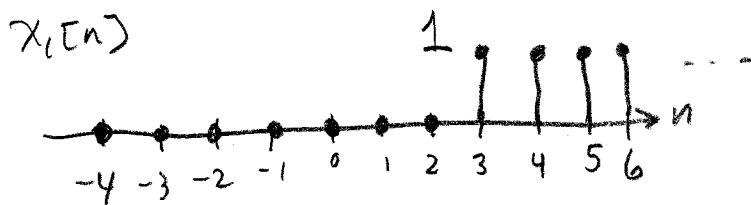
$$\begin{aligned} x_{3ca}[n] &= \frac{1}{2} \{ x_3[n] - x_3^*[-n] \} \\ &= \frac{1}{2} \{ j\cos(2\pi n/7) - \sin(2\pi n/4) + j\cos(2\pi n/7) - \sin(2\pi n/4) \} \\ &= \frac{1}{2} \{ 2j\cos(2\pi n/7) - 2\sin(2\pi n/4) \} \\ &= \underline{\underline{j\cos(2\pi n/7) - \sin(2\pi n/4)}} \end{aligned}$$

Note: $x_3[n]$ was conjugate antisymmetric to start with, so the conjugate antisymmetric part is just $x_3[n]$ itself.

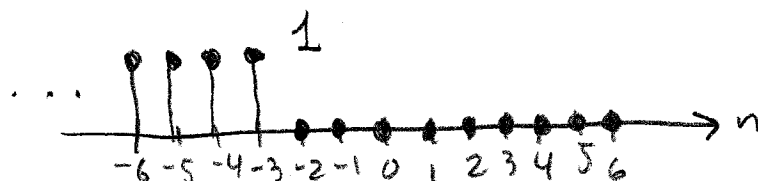
2.27a)

$$x_1[n] = u[n-3]$$

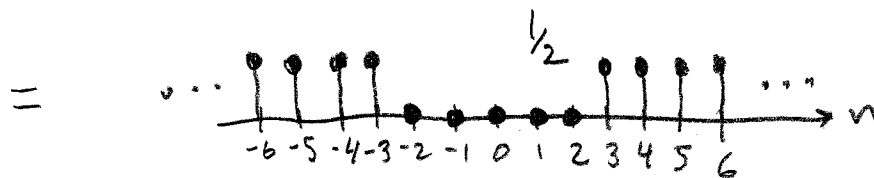
2.27a)-1



$$x_1[-n]$$

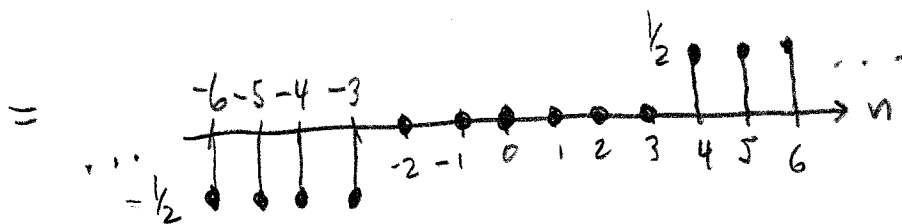


$$x_{1ev}[n] = \frac{1}{2} (x_1[n] + x_1[-n])$$



$$= \frac{1}{2} u[-n-3] + \frac{1}{2} u[n-3] = \frac{1}{2} u[|n|-3]$$

$$x_{1od}[n] = \frac{1}{2} (x_1[n] - x_1[-n])$$

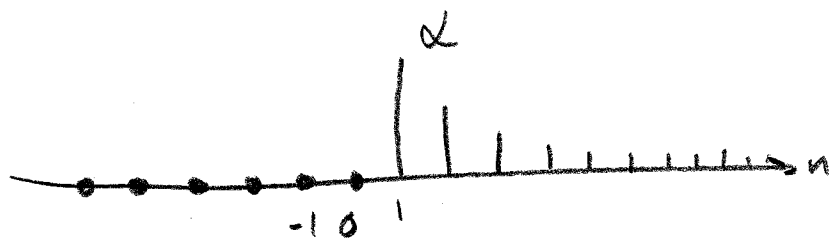


$$= -\frac{1}{2} u[-n-3] + \frac{1}{2} u[n-3] = \frac{1}{2} \operatorname{sgn}(n) u[|n|-3]$$

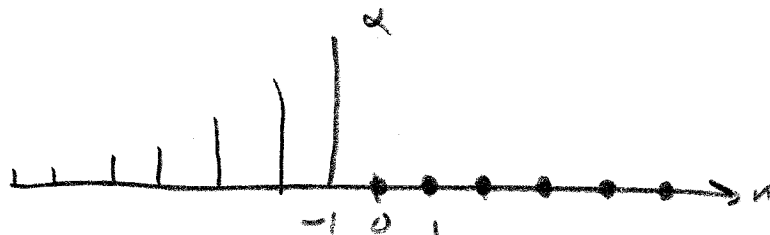
2.27b)

$$x_2[n] = \alpha^n u[n-1]$$

2.27b)-1

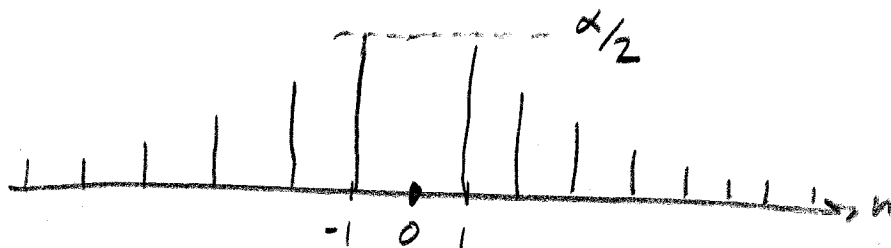


$$x_2[-n] = \alpha^{-n} u[-n-1]$$



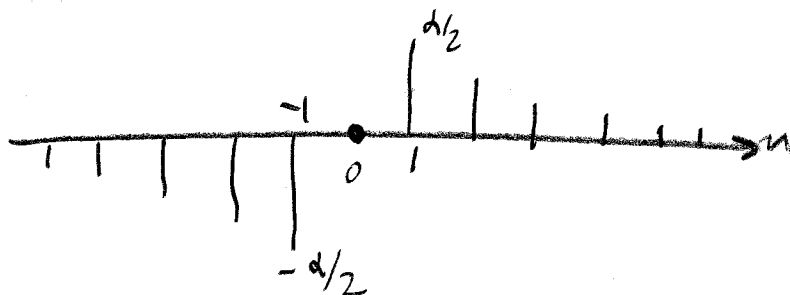
$$x_{2ev}[n] = \frac{1}{2} (x_2[n] + x_2[-n]) = \frac{1}{2} \alpha^n u[n-1] + \frac{1}{2} \alpha^{-n} u[-n-1]$$

$$= \frac{1}{2} \alpha^{|n|} u[|n|-1]$$



$$x_{2od}[n] = \frac{1}{2} (x_2[n] - x_2[-n]) = \frac{1}{2} \alpha^n u[n-1] - \frac{1}{2} \alpha^{-n} u[-n-1]$$

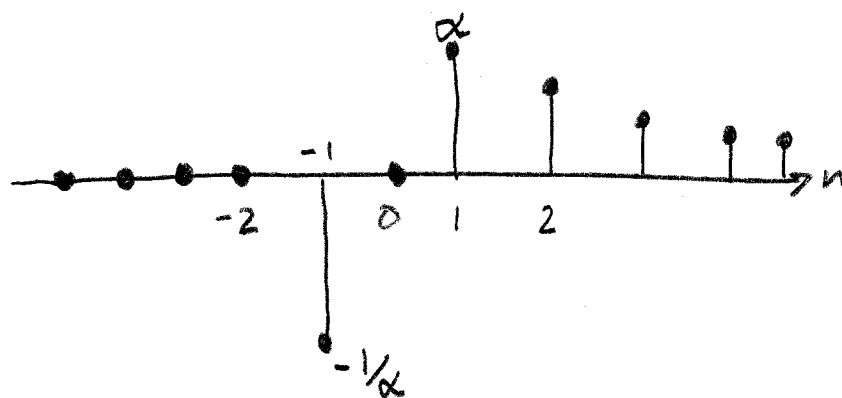
$$= \frac{1}{2} \operatorname{sgn}(n) \alpha^{|n|} u[|n|-1]$$



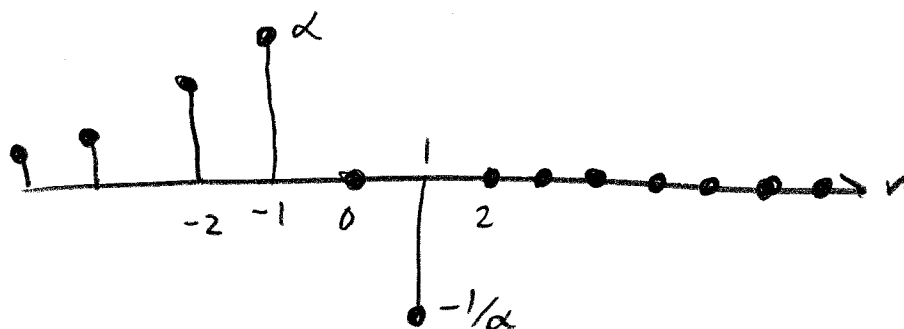
2.27c)

$$x_3[n] = n\alpha^n u[n+1]$$

2.27c)-1



$$x_3[-n] = -n\alpha^{-n} u[-n+1]$$



From these graphs, we have

$$x_{3ev}[n] = \frac{1}{2}(x_3[n] + x_3[-n]) = \begin{cases} -\frac{n}{2}\alpha^{-n}, & n \leq -2 \\ \frac{\alpha}{2} - \frac{1}{2\alpha}, & n = -1 \\ 0, & n = 0 \\ \frac{\alpha}{2} - \frac{1}{2\alpha}, & n = 1 \\ \frac{n}{2}\alpha^n, & n \geq 2 \end{cases}$$

$$= \frac{|n|}{2} \alpha^{|n|} u[|n|-2] + \frac{\alpha^2 - 1}{2\alpha} \delta[|n|-1]$$

→

$$x_{3od}[n] = \frac{1}{2} (x_3[n] - x_3[-n])$$

2.27c)-2

$$= \begin{cases} \frac{n}{2} \alpha^n, & n \leq -2 \\ -\frac{\alpha}{2} - \frac{1}{2\alpha}, & n = -1 \\ 0, & n = 0 \\ \frac{\alpha}{2} + \frac{1}{2\alpha}, & n = 1 \\ \frac{n}{2} \alpha^n, & n \geq 2 \end{cases}$$

$$= \frac{n}{2} \alpha^{|n|} u[|n|-2] + \operatorname{sgn}(n) \frac{\alpha^2 + 1}{2\alpha} \delta[|n|-1]$$

2.27d)

$$x_4[n] = \alpha^{|n|}$$

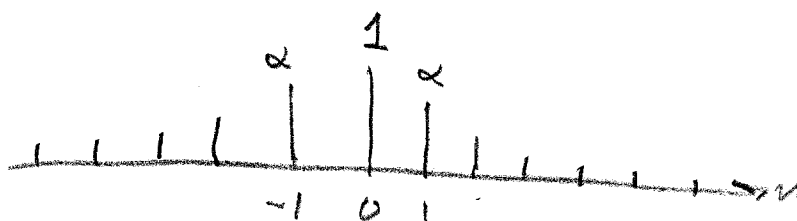
2.27d)-1

Since $x_4[n]$ is even, it follows immediately that

$$x_{4ev}[n] = x_4[n] = \alpha^{|n|} \quad \text{and} \quad x_{4od}[n] = 0.$$

But let's work it out and show this:

$$x_4[n] = \alpha^{|n|}$$



$$x_4[-n] = \alpha^{|-n|} = \alpha^{|n|} = x_4[n]$$

$$x_{4ev}[n] = \frac{1}{2} (x_4[n] + x_4[-n]) = \frac{1}{2} (x_4[n] + x_4[n]) = x_4[n]$$

$$x_{4od}[n] = \frac{1}{2} (x_4[n] - x_4[-n]) = \frac{1}{2} (x_4[n] - x_4[n]) = 0.$$

$$2.39a) \quad \tilde{x}_a[n] = e^{j\pi n/4}$$

$$\omega_0 = \frac{\pi}{4}$$

$$\frac{\omega_0}{2\pi} = \frac{\pi/4}{2\pi} = \frac{1}{8}$$

$$\text{Fundamental Period} = \underline{\underline{8}}$$

One period of $\tilde{x}_a[n]$ looks like one period of $\tilde{x}_a(t)$.

$$2.39b) \quad \tilde{x}_b[n] = \cos(0.6\pi n + 0.3\pi) = \cos\left(\frac{6\pi}{10}n + \frac{3\pi}{10}\right)$$

$$\omega_0 = \frac{6\pi}{10} = \frac{3\pi}{5}$$

$$\frac{\omega_0}{2\pi} = \frac{\frac{3\pi}{5}}{2\pi} = \frac{3}{10}$$

$$\text{Fundamental Period} = \underline{\underline{10}}$$

One period of $\tilde{x}_b[n]$ looks like three periods of $\tilde{x}_b(t)$.

$$2.39c) \quad \tilde{x}_c[n] = \operatorname{Re}\{e^{j\pi n/8}\} + \operatorname{Im}\{e^{j\pi n/5}\} \quad \boxed{2.39d-1}$$

$$= \cos\left(\frac{\pi}{8}n\right) + \sin\left(\frac{\pi}{5}n\right)$$

For the cosine, $\omega_0 = \frac{\pi}{8}$ and $\frac{\omega_0}{2\pi} = \frac{\pi}{16\pi} = \frac{1}{16}$

For the sine, $\omega_0 = \frac{\pi}{5}$ and $\frac{\omega_0}{2\pi} = \frac{\pi}{10\pi} = \frac{1}{10}$

So the cosine has fundamental period 16 and the sine has fundamental period 10.

The fundamental period of $\tilde{x}_c[n]$ is the lowest common multiple of these.

$$N_0 = \operatorname{lcm}(16, 10) = \operatorname{lcm}(2^4, 2 \cdot 5)$$

↑ ↑ we need four 2's and a 5.

$$= 2 \cdot 2 \cdot 2 \cdot 2 \cdot 5 = \underline{\underline{80}}$$

$$2.39d) \tilde{x}_d[n] = 6 \sin\left(\frac{15\pi}{100}n\right) - \cos\left(\frac{12\pi}{100}n + \frac{\pi}{10}\right)$$

For the sine, $\omega_0 = \frac{15\pi}{100} = \frac{3\pi}{20}$ and $\frac{\omega_0}{2\pi} = \frac{3\pi}{40\pi} = \frac{3}{40}$.

So the fundamental period of the sine is 40.

For the cosine, $\omega_0 = \frac{12\pi}{100} = \frac{3\pi}{25}$ and $\frac{\omega_0}{2\pi} = \frac{3\pi}{50\pi} = \frac{3}{50}$.

So the fundamental period of the cosine is 50.

Overall, the fundamental period of $\tilde{x}_d[n]$ is

$$N_0 = \text{lcm}(40, 50) = \text{lcm}(2 \cdot 2 \cdot 2 \cdot 5, 2 \cdot 5 \cdot 5) \\ = 2 \cdot 2 \cdot 2 \cdot 5 \cdot 5 = \underline{\underline{200}}$$

$$2.39e) \tilde{x}_e[n] = \underbrace{\sin\left(\frac{\pi}{10}n + \frac{75\pi}{100}\right)}_{\omega_0 = \frac{\pi}{10}} - 3 \underbrace{\cos\left(\frac{8\pi}{10}n + \frac{2\pi}{10}\right)}_{\omega_0 = \frac{8\pi}{10} = \frac{4\pi}{5}} + \underbrace{\cos\left(\frac{13\pi}{10}n\right)}_{\omega_0 = \frac{13\pi}{10}}$$

$$\omega_0 = \frac{\pi}{10} \\ \frac{\omega_0}{2\pi} = \frac{\pi}{20\pi} = \frac{1}{20}$$

$$N_0 = 20$$

$$\omega_0 = \frac{8\pi}{10} = \frac{4\pi}{5}$$

$$\frac{\omega_0}{2\pi} = \frac{4\pi}{10\pi} = \frac{4}{10} = \frac{2}{5}$$

$$N_0 = 5$$

$$\omega_0 = \frac{13\pi}{10}$$

$$\frac{\omega_0}{2\pi} = \frac{13\pi}{20\pi}$$

$$= \frac{13}{20}$$

$$N_0 = 20$$

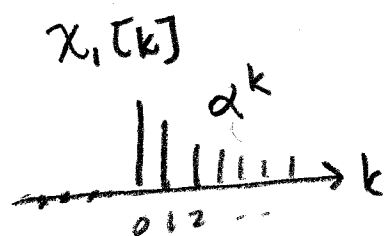
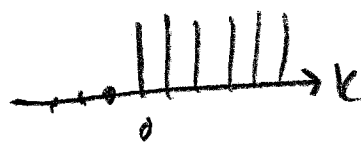
Overall, $N_0 = \text{lcm}(20, 5, 20) = \text{lcm}(2 \cdot 2 \cdot 5, 5, 2 \cdot 2 \cdot 5) \\ = 2 \cdot 2 \cdot 5 = \underline{\underline{20}}.$

2.46a)

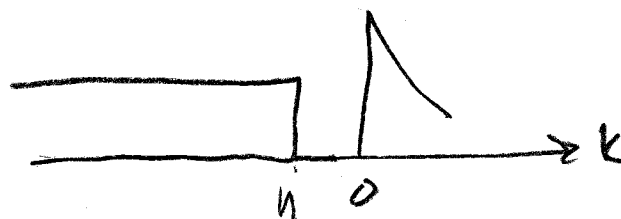
$$x_1[n] = \alpha^n u[n]$$

$$x_2[n] = u[n]$$

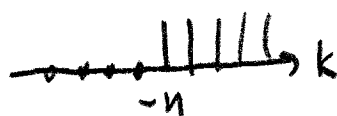
2.46a-1


 $x_2[k]$


$$\begin{aligned} x_1[n] * x_2[n] &= \sum_{k=-\infty}^{\infty} x_1[k] x_2[n-k] \\ &= \sum_{k=-\infty}^{\infty} x_1[n-k] x_2[k] \end{aligned}$$

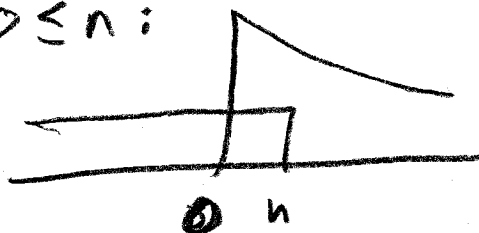
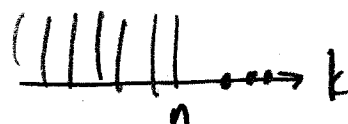
I) $n < 0$:

$$x_2[k-n] = x_2[n+k]$$



$$x_1[k] x_2[n-k] = 0 \quad \forall k$$

$$\Rightarrow y[n] = 0.$$

II) $0 \leq n$:

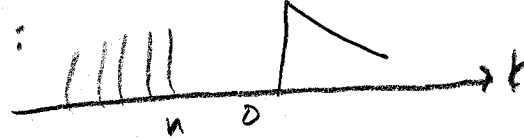
you get nonzero terms from $k=0$ to $k=n$:

$$\begin{aligned} y[n] &= \sum_{k=0}^n x_1[k] x_2[n-k] = \sum_{k=0}^n \alpha^k \cdot 1 \\ &= \sum_{k=0}^n \alpha^k = \frac{\alpha^0 - \alpha^{n+1}}{1 - \alpha} = \frac{1 - \alpha^{n+1}}{1 - \alpha} \end{aligned}$$

$$\text{All together: } y[n] = \begin{cases} 0, & n < 0 \\ \frac{1 - \alpha^{n+1}}{1 - \alpha}, & n \geq 0 \end{cases} = \frac{1 - \alpha^{n+1}}{1 - \alpha} u[n].$$

2.46b-1

$\chi, [v]$

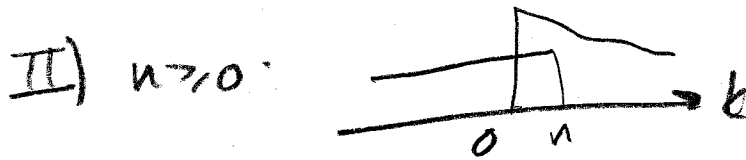


$$X_2[k]$$



$$\sum_{k=-\infty}^{\infty} x_1[k] x_2[n-k] = \sum_{k=-\infty}^{\infty} 0 = 0.$$

$$x_2[k-n] = x_2[n+k]$$



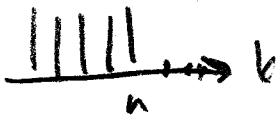
$$y[n] = \sum_{k=0}^n x_1[k] x_2[n-k]$$

$$= \sum_{k=0}^n k \alpha^k \cdot 1$$

$$= \frac{\alpha(1 - (n+1)\alpha^n + n\alpha^{n+1})}{(1-\alpha)^2}.$$

This formula
is given
on the
formula
sheet

$$x_2[n-k]$$

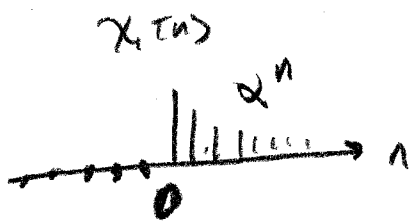


All together:

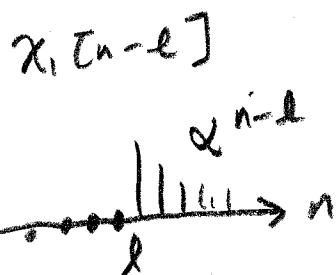
$$y[n] = \left\{ \frac{\alpha [1 - (n+1)\alpha^n + n\alpha^{n+1}]}{(1-\alpha)^2} \right\} u[n]$$

$$2.53a) \quad x_1[n] = \alpha^n u[n]$$

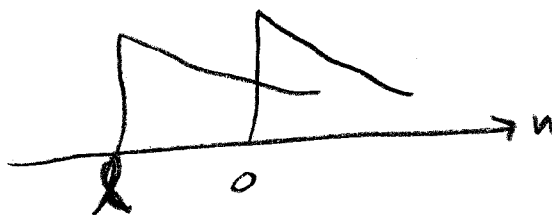
2.53a-1



$$\Gamma_{x_1 x_1}[l] = \sum_{n \in \mathbb{Z}} x_1[n] x_1[n-l]$$

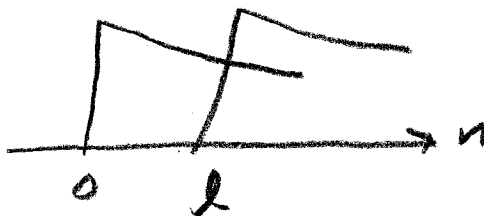


Case I) $l < 0$:



$$\begin{aligned} \Gamma_{xx}[l] &= \sum_{n=0}^{\infty} x_1[n] x_1[n-l] = \sum_{n=0}^{\infty} \alpha^n \alpha^{n-l} \\ &= \sum_{n=0}^{\infty} \alpha^n \alpha^n \alpha^{-l} = \alpha^{-l} \sum_{n=0}^{\infty} \alpha^{2n} \\ &= \alpha^{-l} \sum_{n=0}^{\infty} (\alpha^2)^n = \alpha^{-l} \frac{1}{1-\alpha^2} \\ &= \frac{\alpha^{-l}}{1-\alpha^2} \quad \text{provided } |\alpha^2| < 1. \end{aligned}$$

Case II) $l \geq 0$



$$\Gamma_{xx}[l] = \sum_{n=l}^{\infty} x_1[n] x_1[n-l]$$

→

$$= \sum_{n=l}^{\infty} \alpha^n \alpha^{n-l} = \sum_{n=l}^{\infty} \alpha^n \alpha^n \alpha^{-l}$$

2.53a-2

$$= \alpha^{-l} \sum_{n=l}^{\infty} \alpha^{2n} = \alpha^{-l} \sum_{n=l}^{\infty} (\alpha^2)^n$$

$$= \alpha^{-l} \lim_{A \rightarrow \infty} \sum_{n=l}^A (\alpha^2)^n = \alpha^{-l} \lim_{A \rightarrow \infty} \frac{\alpha^{2l} - \alpha^{2A+2}}{1 - \alpha^2}$$

$$= (\text{provided } |\alpha^2| < 1) \quad \alpha^{-l} \frac{\alpha^{2l} - 0}{1 - \alpha^2}$$

$$= \frac{\alpha^l}{1 - \alpha^2}$$

All Together: $r_{xx}[l] = \begin{cases} \frac{\alpha^{-l}}{1 - \alpha^2}, & l < 0 \\ \frac{\alpha^l}{1 - \alpha^2}, & l \geq 0 \end{cases}$

or: $r_{xx}[l] = \frac{\alpha^{|l|}}{1 - \alpha^2}$

→ This function is clearly even in "l" because "l" appears only in an absolute value.

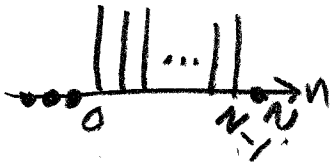
→ Since $|\alpha| < 1$ is required for convergence, $\alpha^{|l|}$ is monotonically decreasing in $|l|$ and the peak occurs at $l=0$.

$$2.53b) \quad x_2[n] = \begin{cases} 1, & 0 \leq n \leq N-1 \\ 0, & \text{other} \end{cases}$$

2.53b-1

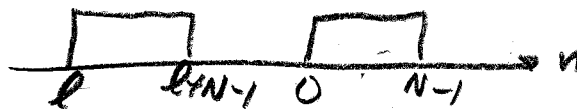
$$= u[n] - u[n-N]$$

$x_2[n]$



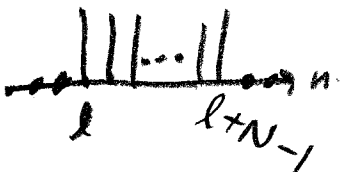
Case I) $l+N-1 < 0$:

$$l < 1-N = -(N-1):$$



$x_2[n-l]$

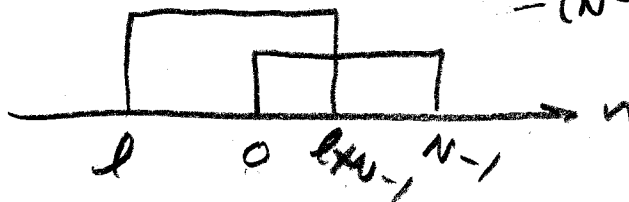
$$r_{xx}[l] = 0.$$



Case II) $l+N-1 \geq 0$ and $l+N-1 < N-1$:

$$l \geq 1-N = -(N-1) \text{ and } l < 0:$$

$$-(N-1) \leq l < 0:$$



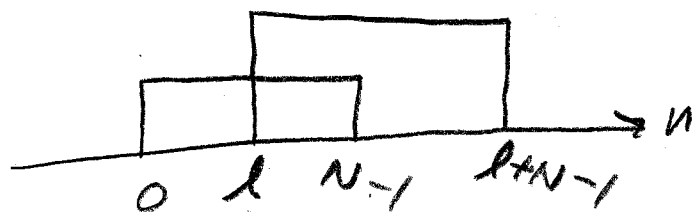
$$r_{xx}[l] = \sum_{n=0}^{l+N-1} x_2[n] x_2[n-l] = \sum_{n=0}^{l+N-1} 1 = N+l$$

Case III) $l+N-1 \geq N-1$ and $l < N$:

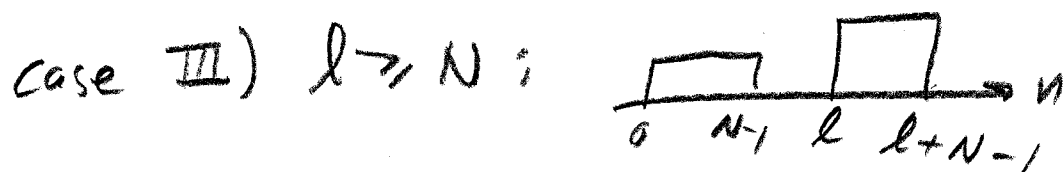
$$l \geq 0 \text{ and } l < N:$$

$$0 \leq l < N:$$





$$r_{xx}[l] = \sum_{n=l}^{N-1} x_2[n] x_2[n-l] = \sum_{n=l}^{N-1} 1 = N-l.$$



$$r_{xx}[l] = 0.$$

all Together: $r_{xx}[l] =$

$$\begin{cases} 0, & l < -(N-1) \\ N+l, & -(N-1) \leq l < 0 \\ N-l, & 0 \leq l < N \\ 0, & l \geq N \end{cases}$$

$$= \begin{cases} N-|l|, & |l| < N \\ 0, & |l| \geq N \end{cases}$$

→ Even in " l " because only $|l|$ appears.

→ For $l \in \mathbb{Z}$, $N-|l|$ is decreasing in $|l|$, so the max occurs when $l=0$.

4.3a)

$$y[n] = x[n+3]$$

4.3a-1

(1) Linear: Let the input be $x_1[n]$. Then the output is $y_1[n] = H\{x_1[n]\} = x_1[n+3]$.

Now let the input be $x_2[n]$. Then the output is $y_2[n] = H\{x_2[n]\} = x_2[n+3]$.

Let $a, b \in \mathbb{C}$ be constants and let $x_3[n] = ax_1[n] + bx_2[n]$.

$$\begin{aligned}\text{Then } y_3[n] &= H\{x_3[n]\} \\ &= x_3[n+3] \\ &= ax_1[n+3] + bx_2[n+3] \\ &= ay_1[n] + by_2[n] \checkmark\end{aligned}$$

Therefore, the system is linear.

(2) Causal: Let the input be $x[n] = n$.

Then the output is $y[n] = H\{x[n]\} = x[n+3] = n+3$.

When $n=0$, we have $y[0] = x[3] = 3$, which depends on the future input $x[3]$.

Therefore, the system is not causal.

(3) Stable: Suppose $x[n]$ is a bounded 4.3a-2 input. Then $\exists B \in \mathbb{R}, B > 0$, s.t. $|x[n]| \leq B \forall n \in \mathbb{Z}$.

Now, when $x[n]$ is the input, the output is given by

$$y[n] = H\{x[n]\} = x[n+3].$$

$$\text{So } |y[n]| = |x[n+3]| \leq B.$$

Therefore, $y[n]$ is bounded by B .

Since every bounded input produces a bounded output, the system is stable.

(4) Shift Invariant: Let the input be $x_1[n]$. Then

$$\text{the output is } y_1[n] = H\{x_1[n]\} = x_1[n+3].$$

$$\text{Then } y_1[n-n_0] = x_1[n-n_0+3] \forall n_0 \in \mathbb{Z}.$$

Now let $x_2[n] = x_1[n-n_0]$, where $n_0 \in \mathbb{Z}$.

$$\text{Then } y_2[n] = H\{x_2[n]\} = x_2[n+3] = x_1[n-n_0+3]$$

$$= y_1[n-n_0] \checkmark$$

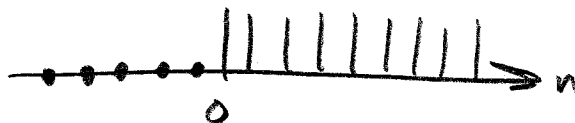
Therefore the system is shift invariant.

4.3b) $y[n] = x[2-n] + \alpha$, $\alpha \in \mathbb{C}$
 $\alpha \neq 0$.

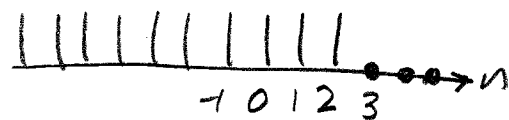
4.3b-1

(1) Linear: Let $x_1[n] = u[n]$

$u[n]$



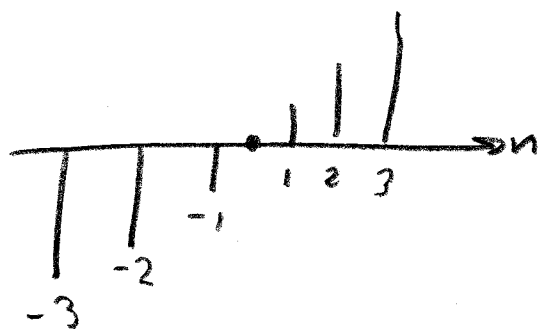
$$u[2-n] = u[-n-2]$$



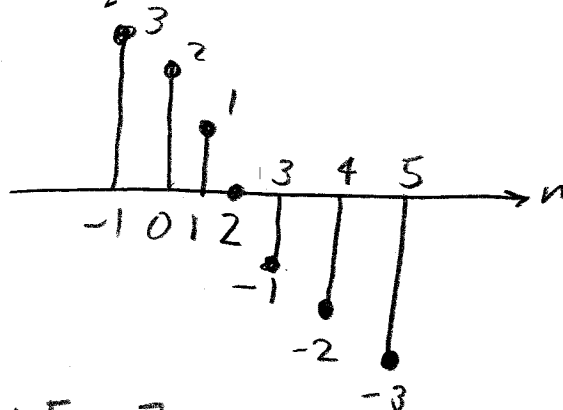
Then $y_1[n] = H\{x_1[n]\} = u[2-n] + \alpha$.

Let $x_2[n] = n$.

$$x_2[n] = n$$



$$x_2[2-n] = 2-n$$



Then $y_2[n] = H\{x_2[n]\} = x_2[2-n] + \alpha$
 $= 2-n + \alpha$

Now let $x_3[n] = x_1[n] + x_2[n] = u[n] + n$

Then $y_3[n] = H\{x_3[n]\} = x_3[2-n] + \alpha = u[2-n] + 2-n + \alpha$.

But $y_1[n] + y_2[n] = u[2-n] + 2-n + \underline{\underline{2\alpha}} \neq y_3[n]$.

So the system is not linear.

(2) Causal: Let the input be $x[n] = n$.

4.3b-2

Then, as in part (1), the output is given by

$$y[n] = H\{x[n]\} = x[2-n] + \alpha = 2-n+\alpha.$$

When $n=0$, we have $y[0] = x[2] + \alpha = 2+\alpha$.

Since $y[0]$ depends on the future input $x[2]$,
the system is not causal.

(3) Stable: Let $x[n]$ be a bounded input. Then
 $\exists B \in \mathbb{R}, B > 0$, s.t. $|x[n]| \leq B \quad \forall n \in \mathbb{Z}$.

The output is given by

$$y[n] = H\{x[n]\} = x[2-n] + \alpha,$$

$$\begin{aligned} \text{so } |y[n]| &= |x[2-n] + \alpha| \\ &\leq |x[2-n]| + |\alpha| \\ &\leq B + |\alpha|. \end{aligned}$$

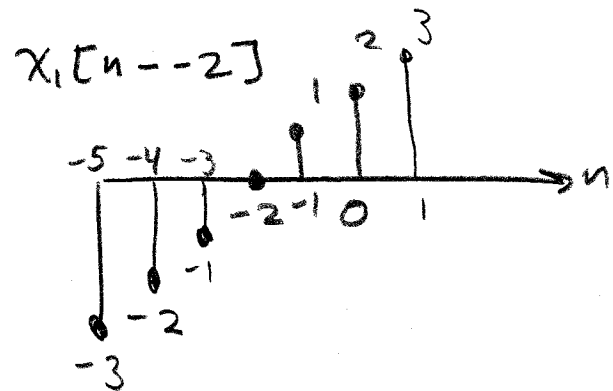
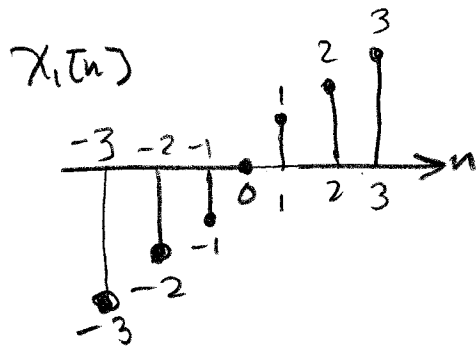
So $y[n]$ is bounded by $B + |\alpha|$.

Since every bounded input produces a bounded output,
the system is stable.

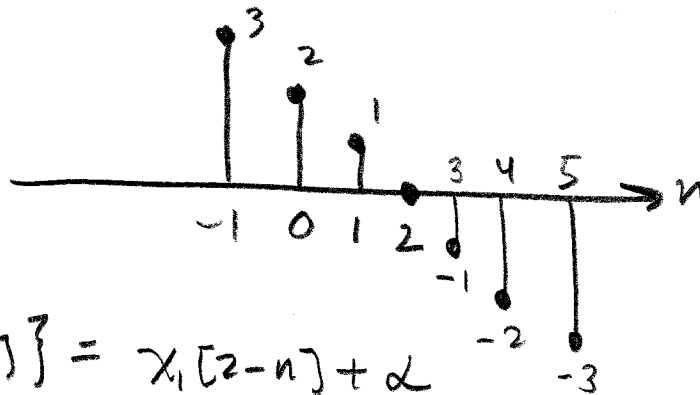
(4) Shift invariant:

4.36-3

Let $x_1[n] = n$.



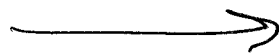
$$x_1[2-n] = x_1[-n-2] = 2-n$$



$$\begin{aligned} \text{Then } y_1[n] &= H\{x_1[n]\} = x_1[2-n] + \alpha \\ &= 2-n + \alpha. \end{aligned}$$

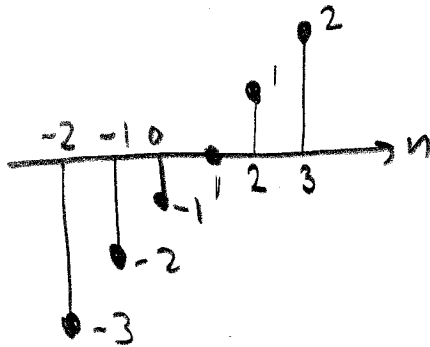
Let $n_0 = 1$. Then $y_1[n-n_0] = y_1[n-1]$

$$\begin{aligned} &= \text{[Plot of } y_1[n-1] \text{]} + \alpha \\ &= 3-n + \alpha. \end{aligned}$$

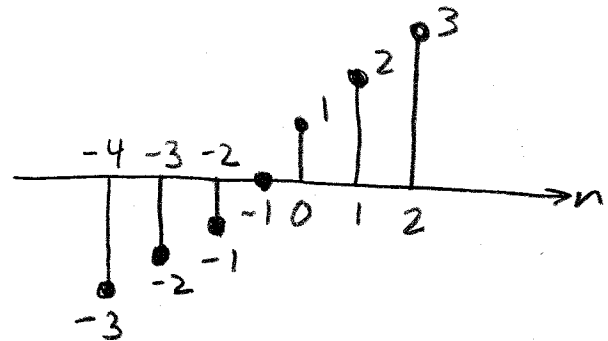


Now let $x_2[n] = x_1[n-n_0] = x_1[n-1] = n-1$: 4.3b-4

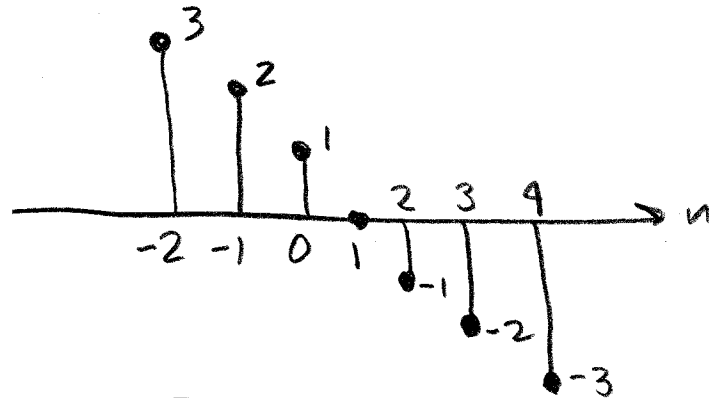
$$x_2[n] = n-1$$



$$x_2[n-2] = x_2[2+n] = n+1$$



$$x_2[-n-2] = x_2[2-n] = 1-n$$



Then $y_2[n] = H\{x_2[n]\} = x_2[2-n] + \alpha = 1-n + \alpha$.

Since $y_2[n] \neq y_1[n-1]$, the system is
not shift invariant.

$$4.3c) \quad y[n] = \ln(1 + |x[n]|)$$

4.3c-1

(1) Linear: Let the input be $x_1[n]$. Then the output is $y_1[n] = \ln(1 + |x_1[n]|)$.

Let the input be $x_2[n]$. Then the output is $y_2[n] = \ln(1 + |x_2[n]|)$.

Let $a, b \in \mathbb{C}$ be constants.

$$\text{Then } ay_1[n] + by_2[n] = a \ln(1 + |x_1[n]|) + b \ln(1 + |x_2[n]|).$$

$$\text{Now let } x_3[n] = ax_1[n] + bx_2[n]$$

$$\begin{aligned} \text{Then } y_3[n] &= \ln(1 + |x_3[n]|) \\ &= \ln(1 + |ax_1[n] + bx_2[n]|) \\ &\neq ay_1[n] + by_2[n] \text{ in general.} \end{aligned}$$

Therefore the system is not linear.

(2) Causal: At time $n = n_0$, the value of the output $y[n_0]$ depends on the current value of the input $x[n_0]$, but not on past values of the input $x[n]$ for $n < n_0$ and not on future values of the input $x[n]$ for $n > n_0$. So the system is causal (and also memoryless).

(3) Stable: Let $x[n]$ be a bounded input.

4.3c-2

Then $\exists B \in \mathbb{R}$, $B > 0$, s.t. $|x[n]| \leq B \quad \forall n \in \mathbb{Z}$.

The output is given by

$$y[n] = \ln(1 + |x[n]|),$$

$$\text{So } |y[n]| = |\ln(1 + |x[n]|)|$$

$$\leq |\ln(1+B)| \quad \left(\begin{array}{l} \text{since } \ln x \text{ is monotonically} \\ \text{increasing for } x > 1. \end{array} \right)$$
$$= \ln(1+B)$$

Then $y[n]$ is bounded by $\ln(1+B)$.

Since every bounded input signal produces a bounded output signal, the system is stable.

(4) Shift Invariant: Let the input be $x_1[n]$. Then the output is $y_1[n] = H\{x_1[n]\} = \ln(1 + |x_1[n]|)$.
 $\forall n_0 \in \mathbb{Z}$, we have $y_1[n-n_0] = \ln(1 + |x_1[n-n_0]|)$.

Now let $n_0 \in \mathbb{Z}$ and let $x_2[n] = x_1[n-n_0]$. The output is $y_2[n] = H\{x_2[n]\} = H\{x_1[n-n_0]\} = \ln(1 + |x_1[n-n_0]|)$.

Since $y_2[n] = y_1[n-n_0]$, the system is shift invariant.

$$4.3d) \quad y[n] = \beta + \sum_{l=-1}^3 x[n-l]$$

4.3d-1

$$= \beta + x[n-3] + x[n-2] + x[n-1] + x[n] + x[n+1], \quad \beta \in \mathbb{C}, \beta \neq 0.$$

(1) Linear: Let $x_1[n] = 0$, e.g., $x_1[n]$ is the signal that is everywhere equal to zero.

$$\begin{aligned} \text{Then } y_1[n] &= H\{x_1[n]\} = \beta + \sum_{l=-1}^3 x_1[n-l] \\ &= \beta + 0 = \beta. \end{aligned}$$

Let $x_2[n] = 1$ (constant signal everywhere equal to one), Then

$$\begin{aligned} y_2[n] &= H\{x_2[n]\} = \beta + \sum_{l=-1}^3 x_2[n-l] \\ &= \beta + \sum_{l=-1}^3 1 = \beta + 5. \end{aligned}$$

Now let $x_3[n] = x_1[n] + x_2[n] = x_2[n] = 1$.

We have immediately from above that

$$y_3[n] = H\{x_3[n]\} = H\{x_2[n]\} = y_2[n] = \beta + 5.$$

$$\text{But } y_1[n] + y_2[n] = \beta + \beta + 5 = 2\beta + 5 \neq y_3[n].$$

So the system is not linear.

(2) causal: when $n=0$, we have

4.3d-2

$$y[0] = \beta + \sum_{l=-1}^3 x[n-l] = \beta + x[-3] + x[-2] + x[-1] + x[0] + x[1],$$

which depends on the future value $x[1]$ of the input $x[n]$ at $n=1$.

Therefore, the system is not causal.

(3) stable: Let $x[n]$ be a bounded input. Then $\exists B \in \mathbb{R}, B > 0$, s.t. $|x[n]| \leq B \quad \forall n \in \mathbb{Z}$.

The output is given by

$$y[n] = \beta + x[n-3] + x[n-2] + x[n-1] + x[n] + x[n+1],$$

$$\begin{aligned} \text{So } |y[n]| &= |\beta + x[n-3] + x[n-2] + x[n-1] + x[n] + x[n+1]| \\ &\leq |\beta| + |x[n-3]| + |x[n-2]| + |x[n-1]| + |x[n]| + |x[n+1]| \\ &\leq |\beta| + 5B. \end{aligned}$$

Therefore, $y[n]$ is bounded by $|\beta| + 5B$ when $x[n]$ is bounded by B .

Since every bounded input signal produces a bounded output signal, the system is stable.

(4) Shift Invariant:

4.3d-3

Let $x_1[n]$ be the input and let $n_0 \in \mathbb{Z}$.
The output is given by

$$y_1[n] = H\{x_1[n]\}$$

$$= \beta + x_1[n-3] + x_1[n-2] + x_1[n-1] + x_1[n] + x_1[n+1].$$

$$\text{So } y_1[n-n_0] = \beta + x_1[n-n_0-3] + x_1[n-n_0-2] + x_1[n-n_0-1] \\ + x_1[n-n_0] + x_1[n-n_0+1].$$

Now let $x_2[n] = x_1[n-n_0]$.

Then $y_2[n] = H\{x_2[n]\}$

$$= \beta + x_2[n-3] + x_2[n-2] + x_2[n-1] + x_2[n] + x_2[n+1]$$

$$= \beta + x_1[n-n_0-3] + x_1[n-n_0-2] + x_1[n-n_0-1] \\ + x_1[n-n_0] + x_1[n-n_0+1]$$

$$= y_1[n-n_0]. \checkmark$$

Therefore the system is shift invariant.

4.8) The system I/O relation is:

4.8-1

$$x[n] \rightarrow \boxed{H} \rightarrow y[n]$$

$$y[n] = x[n+1] - 2x[n] + x[n-1]$$

Let $x_1[n]$ be the input. Then we have

$$y_1[n] = H\{x_1[n]\} = x_1[n+1] - 2x_1[n] + x_1[n-1].$$

Let $x_2[n]$ be the input. Then

$$y_2[n] = H\{x_2[n]\} = x_2[n+1] - 2x_2[n] + x_2[n-1].$$

Let $c_1, c_2 \in \mathbb{C}$ be arbitrary constants.

$$\begin{aligned} \text{Then } c_1 y_1[n] + c_2 y_2[n] &= c_1 x_1[n+1] - 2c_1 x_1[n] \\ &\quad + c_1 x_1[n-1] \\ &\quad + c_2 x_2[n+1] - 2c_2 x_2[n] \\ &\quad + c_2 x_2[n-1]. \end{aligned}$$

$$\text{Let } x_3[n] = c_1 x_1[n] + c_2 x_2[n].$$

$$\begin{aligned} \text{Then } y_3[n] &= H\{x_3[n]\} = x_3[n+1] - 2x_3[n] + x_3[n-1] \\ &= (c_1 x_1[n+1] + c_2 x_2[n+1]) \\ &\quad - 2(c_1 x_1[n] + c_2 x_2[n]) \\ &\quad + (c_1 x_1[n-1] + c_2 x_2[n-1]) \end{aligned}$$

$$= C_1 x_1[n+1] + C_2 x_2[n+1] \\ - 2C_1 x_1[n] - 2C_2 x_2[n] \\ + C_1 x_1[n-1] + C_2 x_2[n-1]$$

$$= C_1 x_1[n+1] - 2C_1 x_1[n] + C_1 x_1[n-1] \\ + C_2 x_2[n+1] - 2C_2 x_2[n] + C_2 x_2[n-1]$$

$$= C_1 (x_1[n+1] - 2x_1[n] + x_1[n-1]) \\ + C_2 (x_2[n+1] - 2x_2[n] + x_2[n-1])$$

$$= C_1 y_1[n] + C_2 y_2[n] \quad \checkmark$$

\Rightarrow The system is linear.

Let $x_1[n]$ be the input. Then

$$y_1[n] = H\{x_1[n]\} \\ = x_1[n+1] - 2x_1[n] + x_1[n-1]$$

Let $n_0 \in \mathbb{Z}$. Then

$$y_1[n-n_0] = x_1[n+1-n_0] - 2x_1[n-n_0] + x_1[n-1-n_0].$$



Now let $x_2[n] = x_1[n - n_0]$.

4.8-3

Then $y_2[n] = H\{x_2[n]\}$

$$\begin{aligned} &= x_2[n+1] - 2x_2[n] + x_2[n-1] \\ &= x_1[n - n_0 + 1] - 2x_1[n - n_0] + x_1[n - n_0 - 1] \\ &= x_1[n + 1 - n_0] - 2x_1[n - n_0] + x_1[n - 1 - n_0] \\ &= y_1[n - n_0] \checkmark \end{aligned}$$

The system is time invariant.

Consider the case $n=2$. At $n=2$,

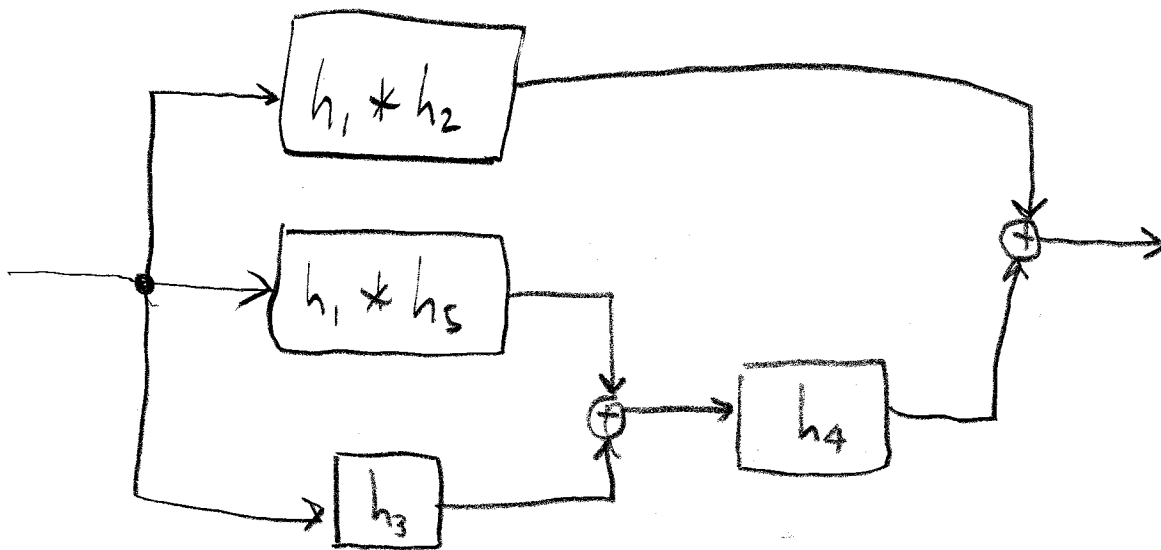
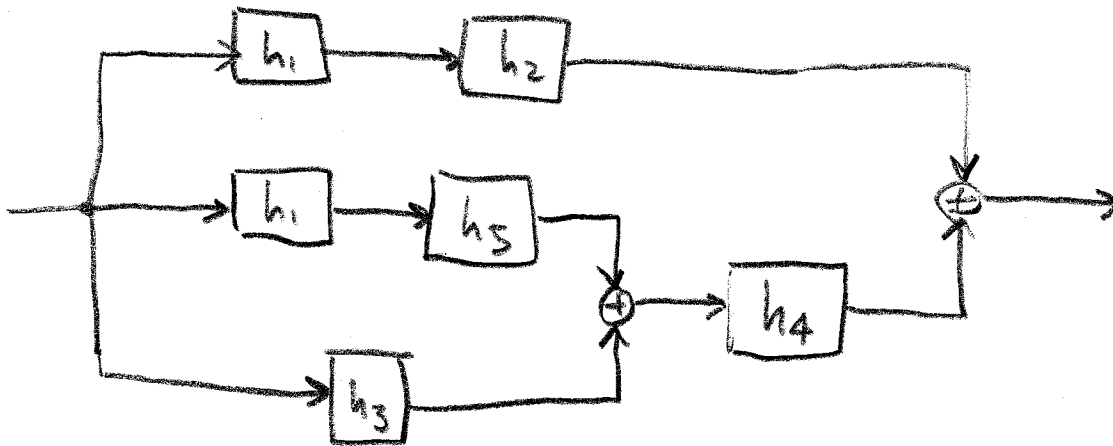
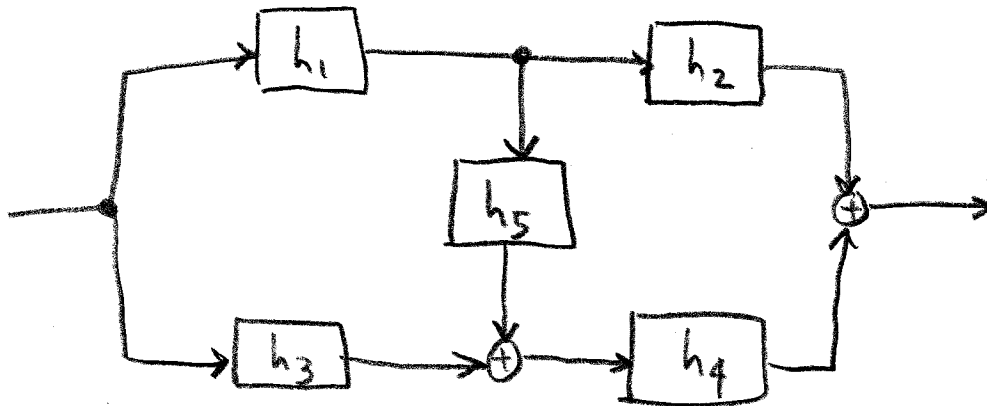
$$y[2] = x[3] - 2x[2] + x[1].$$

\Rightarrow calculation of the output $y[2]$ requires the future input $x[3]$.

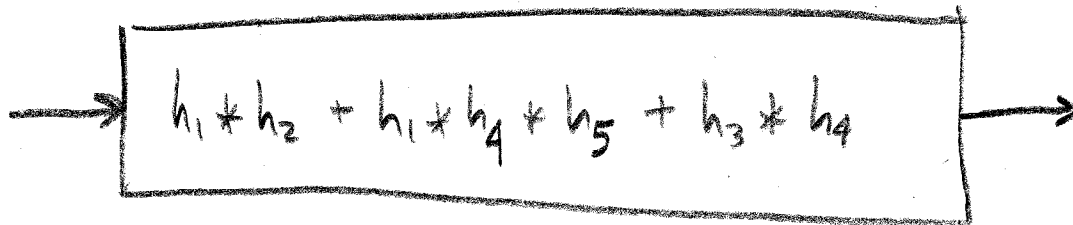
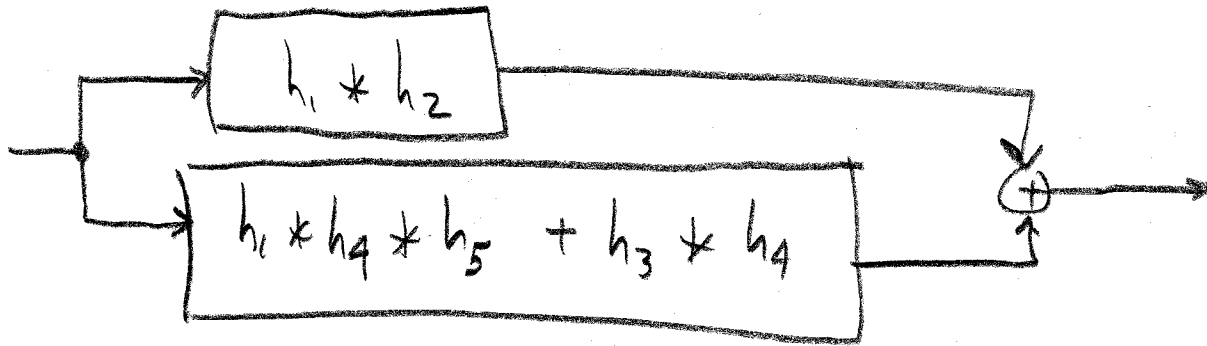
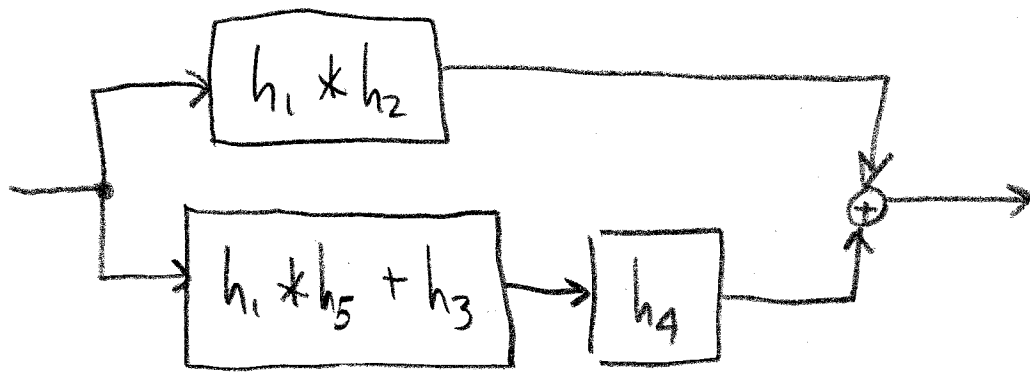
Therefore, the system is not causal.

4.30a)

4.30a-1



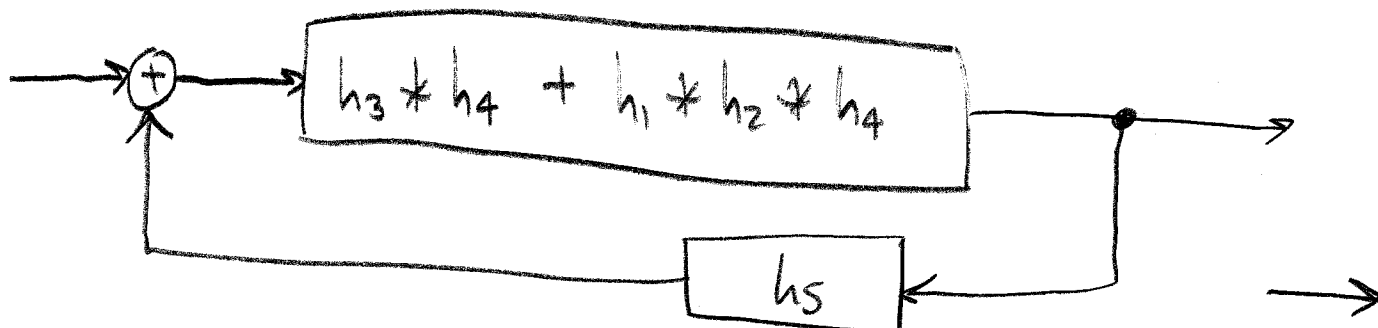
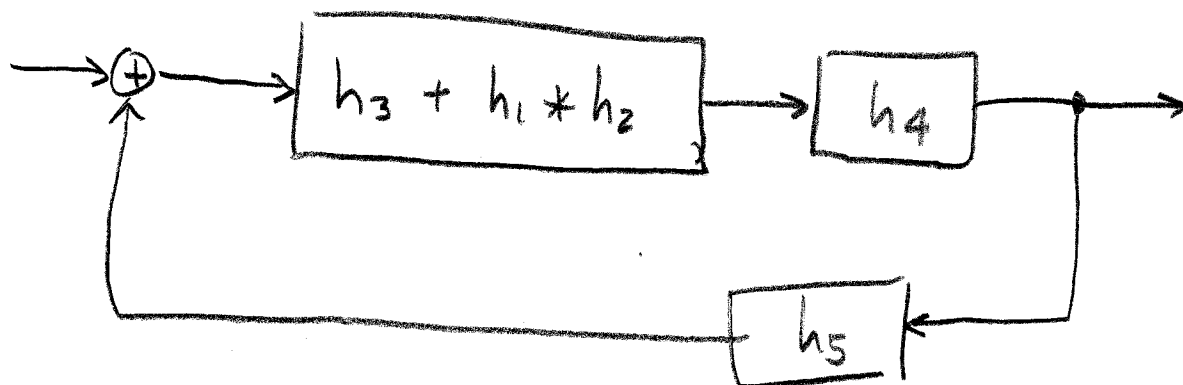
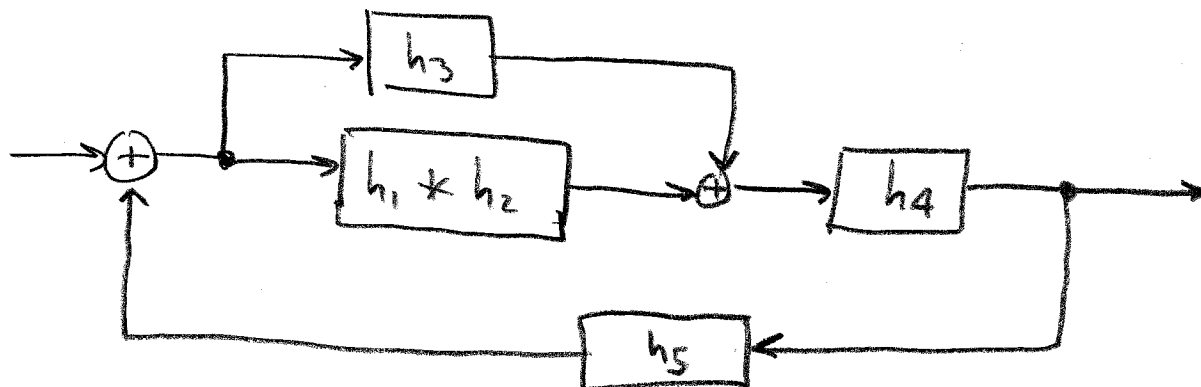
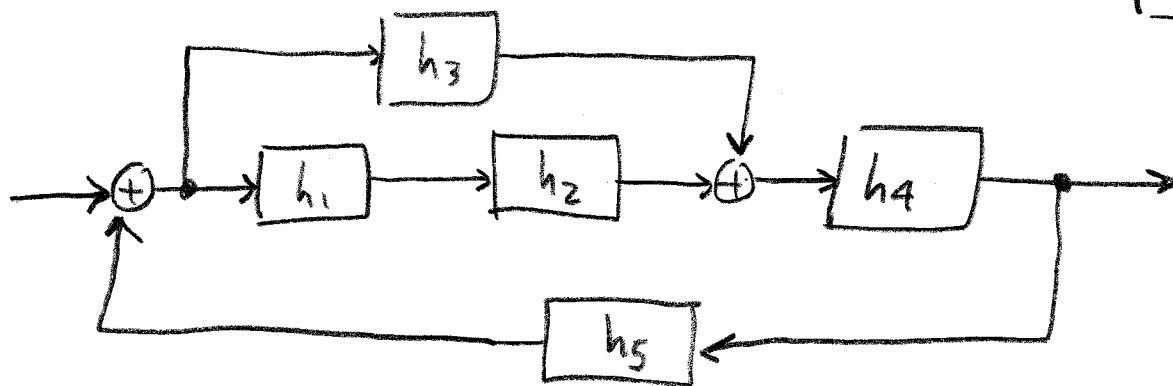
4.30a-2



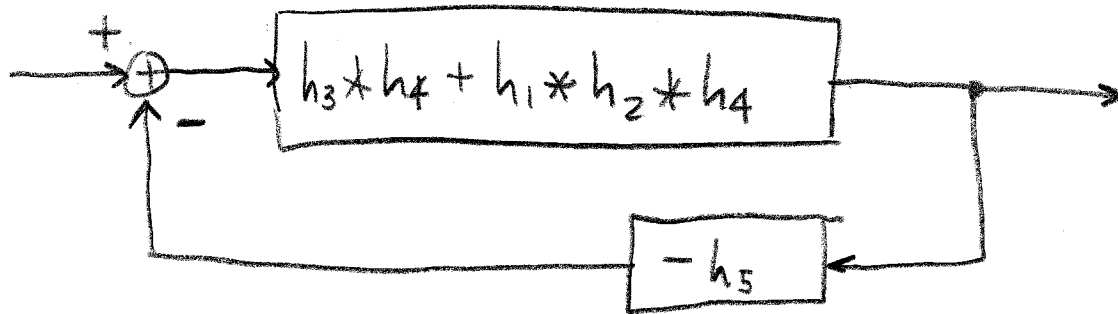
$$h[n] = h_1[n] * h_2[n] + h_1[n] * h_4[n] * h_5[n] + h_3[n] * h_4[n]$$

4.30b)

4.30b-1



4.30b-2



Forward path gain: $H_3(z)H_4(z) + H_1(z)H_2(z)H_4(z)$

Reverse path gain: $-H_5(z)$

$$H(z) = \frac{\text{Forward Gain}}{1 + (\text{Forward Gain})(\text{Reverse Gain})}$$

$$= \frac{H_3(z)H_4(z) + H_1(z)H_2(z)H_4(z)}{1 - H_3(z)H_4(z)H_5(z) - H_1(z)H_2(z)H_4(z)H_5(z)}$$

$$h[n] = \mathcal{Z}^{-1}\{H(z)\}$$

(no general closed form solution in time domain)