

Exploring Fourier series from computational perspective

Igor Krzywda

April 3, 2021

Contents

| | | |
|----------|---|-----------|
| 1 | Introduction | 2 |
| 2 | What is Fourier series? | 2 |
| 2.1 | Deriving a_0 | 2 |
| 2.2 | Finding a_n and b_n | 3 |
| 2.2.1 | Finding a_n | 3 |
| 2.3 | Even and odd functions in context of Fourier series | 4 |
| 2.4 | Approximating a step function | 5 |
| 3 | Overview of the program | 6 |
| 3.1 | Taking input | 6 |
| 3.2 | Implementation of complex numbers | 7 |
| 3.3 | Discrete Fourier transform (DFT) | 8 |
| 3.3.1 | Expressing a point on a unit circle using complex numbers | 8 |
| 3.3.2 | Applying Discrete Fourier Transform | 9 |
| 3.3.3 | Why naive implementation of DFT is inefficient | 11 |
| 3.4 | Fourier series and making drawings | 12 |
| 3.4.1 | Implementation of IDFT in code | 13 |
| 4 | Drawings | 14 |
| 4.1 | Triangle | 15 |
| 4.2 | Cube | 16 |
| 4.3 | Pi | 17 |
| 5 | Appendices | 17 |
| 5.1 | Appendix A | 17 |
| 5.2 | Appendix B | 17 |
| 5.3 | Appendix C | 18 |
| 5.4 | Appendix D | 19 |
| 6 | Bibliography | 19 |

1 Introduction

One time when I was binging YouTube, I stumbled upon a video by Grant Sanderson of 3Blue1Brown on Fourier series which featured many visualizations, prime of whose were drawings on a complex plain which were generated using Fourier series. Having done some reading on Fourier series as well as Fourier transform, I decided that it would be a great idea to actually understand those concepts. Since for me the ultimate way of learning is solving problems and since I feel best at expressing ideas in code, I decided to set myself a challenge of programming a program that will redraw my sketches with Fourier series.

2 What is Fourier series?

The purpose of Fourier series is to approximate any periodic function with a sum of sines and cosines. The formula is given by (1)

$$S_{\infty}(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos \frac{2\pi nt}{P} + b_n \sin \frac{2\pi nt}{P} \quad (1)$$

From the formula we can observe two things, each sine and cosine in the sum has a weight assigned to it (a_n and b_n) and that n determines the frequency of the sinusoids. We can demistify the idea behind Fourier series by deriving formulas for all coefficients.

2.1 Deriving a_0

First coefficient we will find is a_0 , which is the only coefficient that stands on its own. It is added before the actual series in order to compensate for the original function not oscillating around x axis on the plot. It basically is a vertical translation. We can see it if we add any constant to $\sin(x)$, Figure 1.

Figure 1: red : $\sin(x)$, blue : $1 + \sin(x)$



The problem to solve is to find the axis against which the sum will oscillate. The solution to this problem is the average value of the original function over one period. The formula given for finding a_0 goes as follows (2)

$$a_0 = \frac{1}{P} \int_{\frac{P}{2}}^{-\frac{P}{2}} f(t) dt \quad (2)$$

The average can be found through this formula because the integral gives us the area under the graph of function $f(t)$ from $-\frac{P}{2}$ to $\frac{P}{2}$. If we have an area, we can express it with any figure that has this area, since we want an offset from the x axis, we can draw a rectangle whose base has length of P . Since we want to find out what the height of that rectangle is, we need to divide its area by its width, which is exactly what we are doing when dividing the integral by P .

2.2 Finding a_n and b_n

Coefficients a_n and b_n determine the amplitudes of sine waves of frequency n . Since the goal is to approximate some function $f(t)$ with a sum of weighted sines and cosines, let that be the starting point. (3)

$$f(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos \frac{2\pi nt}{P} + b_n \sin \frac{2\pi nt}{P} \quad (3)$$

In order to find the coefficients, we can exploit the properties of definite integrals of sine and cosine. To do that, we will need to expand the whole expression by either $\cos \frac{2\pi nt}{P}$ when looking for a_n , or by $\sin \frac{2\pi nt}{P}$ if we are looking for b_n .

2.2.1 Finding a_n

The expanded equation used to find a_n looks like this (4).

$$\begin{aligned} \int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \cos \frac{2\pi nt}{P} dt &= \int_{-\frac{P}{2}}^{\frac{P}{2}} a_0 \cos \frac{2\pi nt}{P} dt + \int_{-\frac{P}{2}}^{\frac{P}{2}} a_1 \cos \frac{2\pi t}{P} \cos \frac{2\pi nt}{P} dt + \int_{-\frac{P}{2}}^{\frac{P}{2}} b_1 \sin \frac{2\pi t}{P} \cos \frac{2\pi nt}{P} dt \\ &+ \int_{-\frac{P}{2}}^{\frac{P}{2}} a_2 \cos \frac{4\pi t}{P} \cos \frac{2\pi nt}{P} dt + \int_{-\frac{P}{2}}^{\frac{P}{2}} b_2 \sin \frac{4\pi t}{P} \cos \frac{2\pi nt}{P} dt \\ &\vdots \\ &+ \int_{-\frac{P}{2}}^{\frac{P}{2}} a_n \cos^2 \left(\frac{2\pi nt}{P} \right) dt + \int_{-\frac{P}{2}}^{\frac{P}{2}} b_n \sin \frac{2\pi nt}{P} \cos \frac{2\pi nt}{P} dt + \dots \end{aligned} \quad (4)$$

From this rather long expansion we can distill and solve for five cases appearing on the right side of the equation:

1.

$$\int_{-\frac{P}{2}}^{\frac{P}{2}} a_0 \cos \frac{2\pi nt}{P} dt = a_0 \cdot \left[\frac{P}{2\pi n} \sin \frac{2\pi nt}{P} \right]_{-\frac{P}{2}}^{\frac{P}{2}} = a_0 \cdot \frac{P}{2\pi n} (\sin(\pi n) - \sin(-\pi n)) = 0$$

2. For $m \in \mathbb{N}$ and $m \neq n$

$$\begin{aligned} \int_{-\frac{P}{2}}^{\frac{P}{2}} a_m \cos \frac{2\pi mt}{P} \cos \frac{2\pi nt}{P} dt &= a_m \cdot \int_{-\frac{P}{2}}^{\frac{P}{2}} \frac{1}{2} \left(\cos \frac{(m-n)2\pi t}{P} + \cos \frac{(m+n)2\pi t}{P} \right) dt \\ &= \frac{a_m}{2} \cdot \left[\frac{P}{(m-n)2\pi n} \sin \frac{(m-n)2\pi t}{P} + \frac{P}{(m+n)2\pi n} \sin \frac{(m+n)2\pi t}{P} \right]_{-\frac{P}{2}}^{\frac{P}{2}} \\ &= \frac{Pa_m}{4\pi} \left(\frac{1}{m-n} \sin((m-n)\pi) + \frac{1}{m+n} \sin((m+n)\pi) \right. \\ &\quad \left. - \frac{1}{m-n} \sin(-(m-n)\pi) - \frac{1}{m+n} \sin(-(m+n)\pi) \right) = 0 \end{aligned}$$

3. For $m \in \mathbb{N}$ and $m \neq n$

$$\begin{aligned} \int_{-\frac{P}{2}}^{\frac{P}{2}} b_m \sin \frac{2\pi mt}{P} \cos \frac{2\pi nt}{P} dt &= b_m \cdot \int_{-\frac{P}{2}}^{\frac{P}{2}} \frac{1}{2} \left(\sin \frac{(m+n)2\pi t}{P} + \sin \frac{(m-n)2\pi t}{P} \right) dt \\ &= \frac{b_m}{2} \cdot \left[\frac{-P}{2\pi(m+n)} \cos \frac{(m+n)2\pi t}{P} + \frac{-P}{2\pi(m-n)} \cos \frac{(m-n)2\pi t}{P} \right]_{-\frac{P}{2}}^{\frac{P}{2}} \\ &= \frac{Pb_m}{4\pi} \left(\frac{-1}{m+n} \cos((m+n)\pi) + \frac{-1}{m-n} \cos((m-n)\pi) \right. \\ &\quad \left. - \frac{-1}{m+n} \cos(-(m+n)\pi) - \frac{-1}{m-n} \cos(-(m-n)\pi) \right) = 0 \end{aligned}$$

4.

$$\begin{aligned}
\int_{-\frac{P}{2}}^{\frac{P}{2}} b_n \sin \frac{2\pi nt}{P} \cos \frac{2\pi nt}{P} dt &= b_n \int_{-\frac{P}{2}}^{\frac{P}{2}} \frac{1}{2} \left(\sin \frac{4\pi nt}{P} + \sin(0) \right) dt \\
&= \frac{b_n}{2} \left[\frac{-P}{4\pi n} \cos \frac{4\pi nt}{P} - \cos(0) \right]_{-\frac{P}{2}}^{\frac{P}{2}} \\
&= \frac{b_n}{2} \left(\frac{-P}{4\pi n} \cos(2\pi n) - 1 - \frac{-P}{4\pi n} \cos(-2\pi n) + 1 \right) = 0
\end{aligned}$$

5.

$$\begin{aligned}
\int_{-\frac{P}{2}}^{\frac{P}{2}} \cos^2 \left(\frac{2\pi nt}{P} \right) dt &= \frac{1}{2} \int_{-\frac{P}{2}}^{\frac{P}{2}} 1 + \cos \frac{4\pi nt}{P} dt = \frac{1}{2} \cdot \left[t + \frac{P}{4\pi n} \sin \frac{4\pi nt}{P} \right]_{-\frac{P}{2}}^{\frac{P}{2}} \\
&= \frac{1}{2} \left(\frac{P}{2} - \frac{4\pi n}{P} \sin(2\pi n) + \frac{P}{2} + \frac{4\pi n}{P} \sin(-2\pi n) \right) = \frac{P}{2}
\end{aligned}$$

We can see that the only term after expansion and integration that does not amount to zero is $\int_{-\frac{P}{2}}^{\frac{P}{2}} \cos^2 \left(\frac{2\pi nt}{P} \right) dt$, which means that our equation will simplify to the following form (5)

$$\int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \cos \frac{2\pi nt}{P} dt = a_n \frac{P}{2} \quad (5)$$

So in order to find a_n , we will get the following (6)

$$a_n = \frac{2}{P} \int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \cos \frac{2\pi nt}{P} dt \quad (6)$$

Formula for b_n is very similar (7) and stems from the same logic as one of a_n , see Appendix A for calculations.

$$b_n = \frac{2}{P} \int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \sin \frac{2\pi nt}{P} dt \quad (7)$$

2.3 Even and odd functions in context of Fourier series

Last thing that will improve the efficiency with which coefficients can be computed is recognizing the fact whether the input function is an even or odd one. Even function can be described as $f(t) = f(-t)$ and an odd function can be described as $f(t) = -f(-t)$. This fact is important in integration, as there are formulas describing definite integrals of even and odd functions over an interval $[-a, a]$, which we are doing when computing the coefficients. The definite integral for an even function $g(t)$ is in equation (8).

$$\int_{-a}^a g(t) dt = 2 \int_{-a}^a g(t) dt \quad (8)$$

Definite integral for an odd function $h(t)$ is given in (9).

$$\int_{-a}^a h(t) dt = 0 \quad (9)$$

In context of Fourier series this is very useful, because if we look at sine and cosine functions, we can say that they are odd and even respectively:

- $\sin(\theta) = -\sin(-\theta)$
- $\cos(\theta) = \cos(-\theta)$

If the input function $f(t)$ is either even or odd, we need to pay attention to how the functions that describe a_n and b_n look like, because if this function is odd, we can establish from the start that it is zero. If both functions are the same in terms of being even or odd, then the product function will be even. If one function is even and other is odd, however, then the product function will be odd. Coming back to Fourier series, if we are computing coefficients a_n and b_n , we can state that:

- if function $f(t)$ is even, then

$$b_n = \frac{2}{P} \int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \sin \frac{2\pi nt}{P} dt = 0$$

- if function $f(t)$ is odd, then

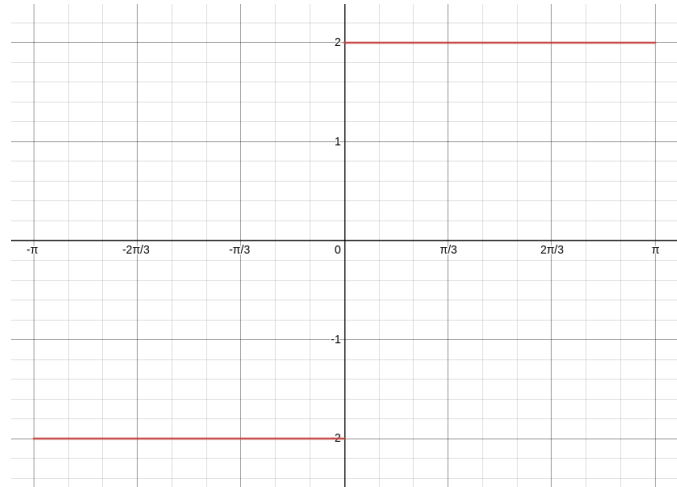
$$a_n = \frac{2}{P} \int_{-\frac{P}{2}}^{\frac{P}{2}} f(t) \cos \frac{2\pi nt}{P} dt = 0$$

2.4 Approximating a step function

Before diving straight to other concepts and the drawings, let us take a second and manually approximate a step function defined in (10) and plotted in Figure 2

$$\begin{aligned} f(t) &= \begin{cases} -2 & \text{if } -\pi \leq t < 0 \\ 2 & \text{if } 0 \leq t \leq \pi \end{cases} \\ f(t) &= f(t + 2\pi) \end{aligned} \tag{10}$$

Figure 2: Step function $f(t)$



The function $f(t)$ is odd, so we only need to calculate coefficients b_n , as we know that all weights a_n will amount to zero.

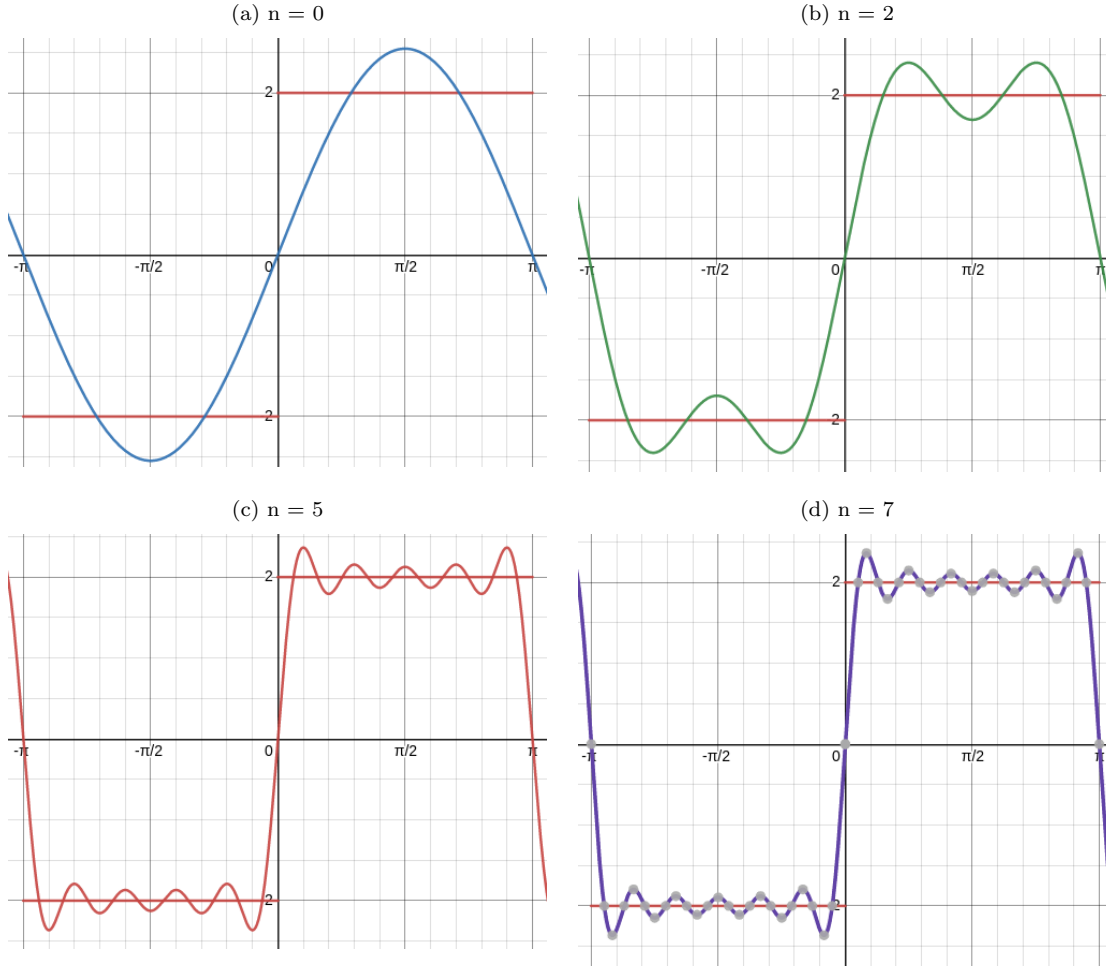
$$\begin{aligned} b_n &= \frac{2}{2\pi} \int_{-\pi}^{\pi} f(t) \sin(nt) dt = \frac{1}{\pi} \int_{-\pi}^0 \frac{2}{n} \cos(nt) + \frac{1}{\pi} \int_0^{\pi} \frac{-2}{n} \cos(nt) \\ &= \frac{2}{\pi n} \left([\cos(nt)]_{-\pi}^0 - [\cos(nt)]_0^{\pi} \right) \\ &= \frac{2}{\pi n} (\cos(0) - \cos(-n\pi) - \cos(n\pi) + \cos(0)) \\ &= \frac{2}{\pi n} (2 - 2\cos(n\pi)) = \frac{4}{\pi n} (1 - \cos(n\pi)) = \begin{cases} \frac{8}{n\pi} & \text{if } n \text{ is odd} \\ 0 & \text{if } n \text{ is even} \end{cases} \end{aligned} \tag{11}$$

The series for our step function looks like this (12).

$$S_{\infty}(t) = \sum_{n=1}^{\infty} \frac{8}{(2n-1)\pi} \sin((2n-1)t) \quad (12)$$

Now we can make a few plots of the series with increasing number of terms and see it converge to $f(t)$, Figure 11

Figure 3



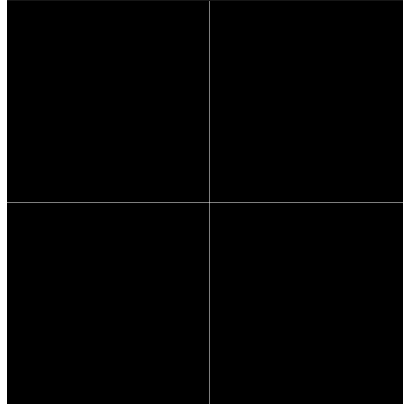
In the plots above we can clearly see the convergence of the series to the function $f(t)$, however the series does not converge well to points at $-\pi$, 0 and π , which shows that the approximations done with Fourier series are not fully precise.

3 Overview of the program

3.1 Taking input

Input is taken from the user in form of an array of (X, Y) coordinates on a window prompted using SFML (Simple and Fast Multimedia Library) for C++ programming language (Figure 4). If the mouse is left-clicked and moved, the coordinates are recorded and saved in the array, as well as printed on the window.

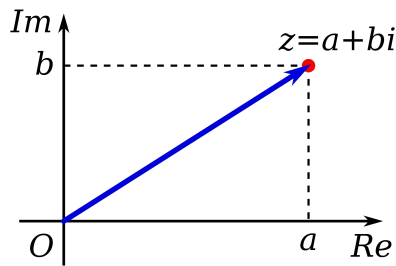
Figure 4



3.2 Implementation of complex numbers

The drawing is a set of points with x and y coordinates, however, all calculations done in the next stages need to be done on complex numbers. Akin to points that are described in terms of coordinates, complex numbers are also described in terms of two dimensions. One of them everyone knows really well, because this is the the real component, which, as the name suggests, consists of real numbers. The component that allows us to move vertically does not belong on the real number line because this number is $\sqrt{-1}$, usually denoted as i . So any complex number can be denoted as following: $a + bi$ and can be depicted like in figure 5 ¹

Figure 5



The implementation of complex numbers in code is a template structure containing two fields - real component (`re`) and imaginary (`im`).

```
template <class T>
struct Complex
{
    T re = 0,
    im = 0;
}
```

The most important thing, however, is overloading operators for this structure, which will make later code much more readable. Below is an example of overloading multiplication operator that will be used a lot in later code. The implementation stems from equation (13)

$$(a + bi) \cdot (c + di) = ac - bd + adi + cbi \quad (13)$$

```
Complex operator*(Complex n)
{
    Complex t;

    t.re = this->re * n.re - this->im * n.im;
```

¹Graphic taken from https://en.wikipedia.org/wiki/Complex_number


```

t.im = this->re * n.im + this->im * n.re;

*this = t;

return *this;
};

```

3.3 Discrete Fourier transform (DFT)

Next step in the program is to find the coefficients for the Fourier series. However, the coefficients will be a bit different, since we are working with a set of complex numbers rather than a function. This means that the series will have a different form, but first we need to find the weights, which can be found using DFT, which is defined by formula (14).

$$X_k = \sum_{n=0}^{N-1} x_n \cdot \left[\cos\left(\frac{2\pi}{N}kn\right) - i\sin\left(\frac{2\pi}{N}kn\right) \right] \quad (14)$$

Now let us walk through what each variable means:

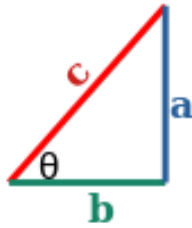
- X_k - output complex number of DFT
- N - size of input set
- k - frequency
- x_n - n^{th} complex number in the input set $\{x_0, x_1, \dots, x_{N-1}\}$

The first thing to go through is why we are multiplying our input complex number by a complex sum?

3.3.1 Expressing a point on a unit circle using complex numbers

Let us take a right triangle with hypotenuse of length 1 (Figure 6) and angle θ between that hypotenuse (c) and the base (b) of triangle.

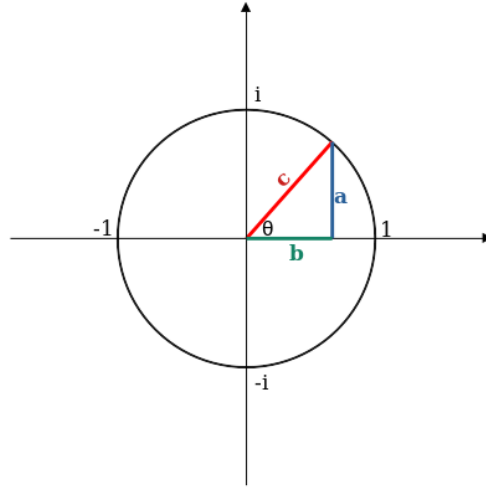
Figure 6



If we wanted to find the length of the base (b), it would be equal to the $\cos\theta$ since the hypotenuse has the length 1. Similarly, if we wanted to find the height of the triangle (a), it would be the $\sin\theta$.

Now we can inscribe such right triangle into a unit circle on a complex plane where the hypotenuse of the triangle would simultaneously be the radius of the circle (Figure 7) at angle θ between radius and the horizontal axis.

Figure 7



We can find the complex value for the point where the radius touches the circumference using the inscribed triangle, so $\cos\theta$ would be the horizontal coordinate and $i \cdot \sin\theta$ would be the vertical component since we are on complex plane. So the whole position can be described as $\cos\theta + i\sin\theta$. Conversely, if $\cos\theta + i\sin\theta$ are plotted for every θ between 0 and 2π , we get an unit circle on a complex plane.

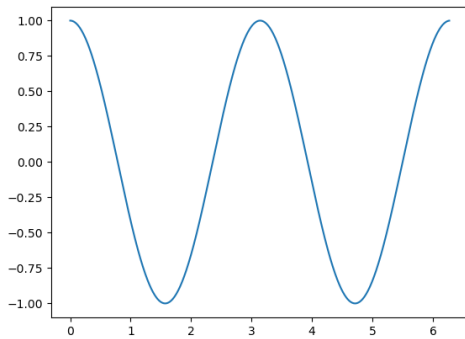
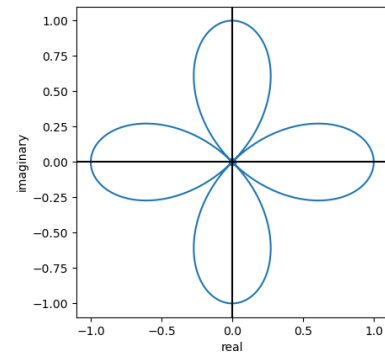
3.3.2 Applying Discrete Fourier Transform

The formula for DFT is a sum, however in order to visualize what is happening when applying DFT to a set of values, we will plot all values that are being summed, as well as the average of this sum (an average is taken normally when applying Fourier series, however, it can also be done during DFT, which will aid the readability of the plots). Let us apply Discrete Fourier Transform to an easy-to- visualize example. We will plot all results from expression (15)

$$x_n \cdot \left[\cos\left(\frac{2\pi}{N}kn\right) - i\sin\left(\frac{2\pi}{N}kn\right) \right] \quad (15)$$

where x_n are the elements of set, which is a discrete form of $\cos(2\theta)$ from 0 to 2π , and k (or frequency) is 1, which means that all values will be mapped within one revolution around the unit circle (Figure 8, where subfigure 8a is the plot of set x and subfigure 8b is the plot of results) of the expression (15)

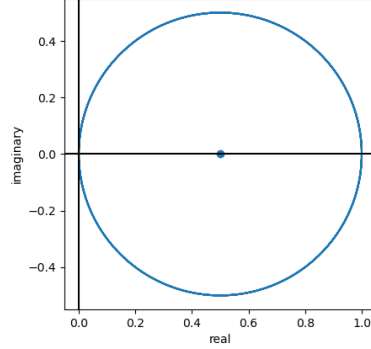
Figure 8

(a) $\cos(2t)$ (b) $\cos(2t)$ - wrapped

In plot 8b we can see that the values have aligned in a shape that is symmetric about real (horizontal) and imaginary (vertical) axes, which means for every element in the set, there is another one, which has an inverse value, which, after summing all values gives zero.

If we do the same operation, but now we change k to match the frequency of the input, which is 2, as the function $\cos(2t)$, that the input set is a discrete form of, completes two cycles in one period. If we plot all of those values along with the average of the sum, we get what is shown in Figure 9.

Figure 9

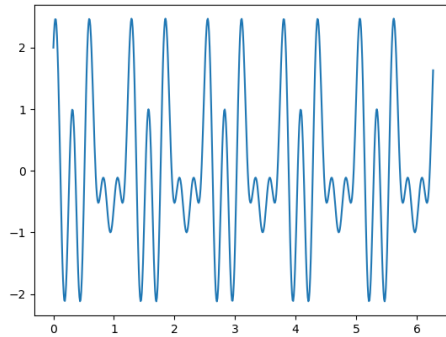


The shape that the values formed in this case is different, as it is only symmetric horizontally, which means that the imaginary components will cancel out, whereas the real components will give a non-zero value, which, after taking the average is in the center of this shape.

The results of DFT tell how much present a component that can be described as $\cos\left(\frac{2\pi}{N}kn\right) - i\sin\left(\frac{2\pi}{N}kn\right)$ is in the set. The concrete values are used later on in phase where Fourier series is used.

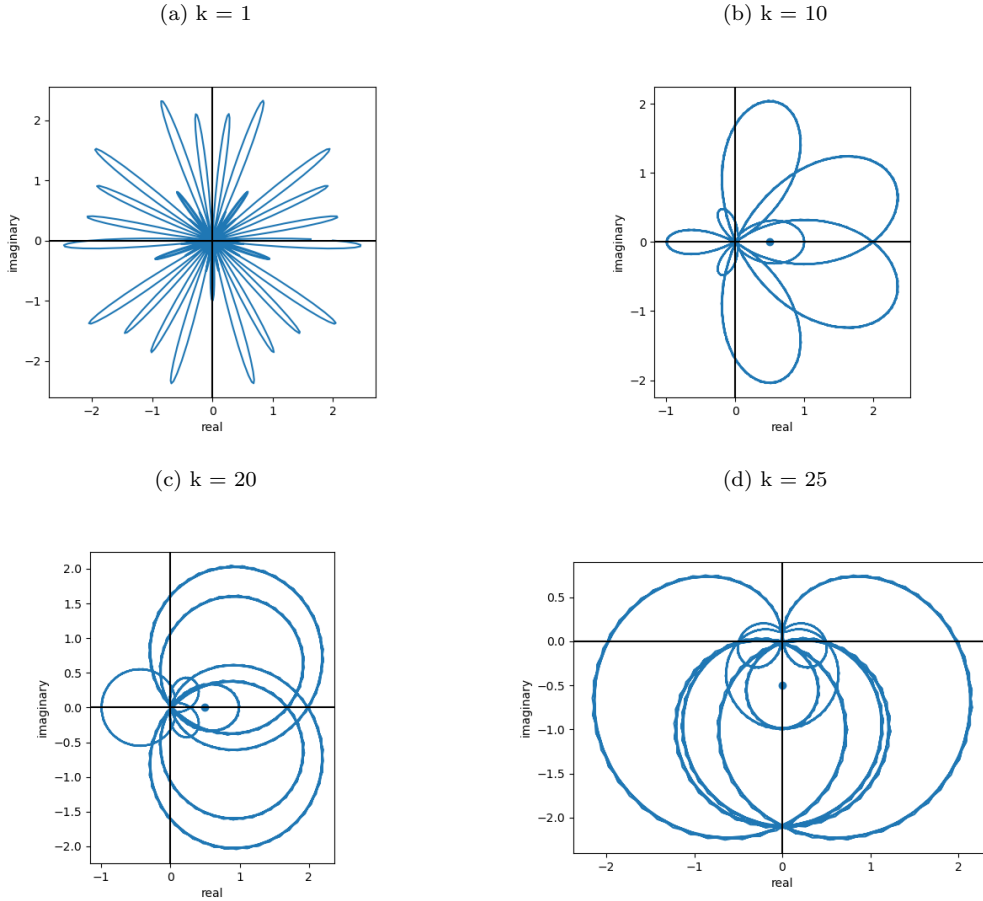
We can, however, take a look how the values will align when we will be applying DFT to sets that are a discrete form of sum of sinusoids. This is one of the most important uses of DFT, as sound signals are composed of many sine waves with different frequencies and DFT allows for finding those frequencies. As an example we shall take a function $f(t) = \cos(10t) + \cos(20t) + \sin(25t)$, whose plot is shown in Figure 10. We can expect non-zero values for $k = 10, 20, 25$.

Figure 10



Below, in Figure?? we can see the plots of all elements of input set multiplied by $\cos\left(\frac{2\pi}{N}kn\right) - i\sin\left(\frac{2\pi}{N}kn\right)$. k in the captions shows the value of k used in the expression and the dots in the plots are the averages of the results of DFT. $\frac{1}{N}$

Figure 11



We can see that for values of k that correspond to the frequencies of the sinusoids making up the original function, the plots bias towards one side, which causes the sums of all those values to be bigger than zero.

See Appendix B for source code of script used to generate all plots above.

3.3.3 Why naive implementation of DFT is inefficient

Below is implementation of DFT that is inside the source code of the program.

```
C_set *DFT(C_set input)
{
    C_set *out = new C_set;
    C temp, X_k;
    unsigned N = input.size();
    double a = (2 * M_PI) / N, x;

    for (unsigned k = 0; k < N; ++k)
    {
        for (unsigned n = 0; n < N; ++n)
        {
            x = a * n * k;
            temp = C(cos(x), -sin(x));

            X_k += (temp * input[n]);
        }
    }
}
```

```

        out->push_back(X_k);
        X_k = C(0,0);
    }

    return out;
}

```

Inside the function there are two for loops. The first one iterates through frequencies (k) as we want more coefficients than for one frequency. The inner loop is the exact implementation of the formula for DFT, where all terms from the input set are multiplied by $\cos\left(\frac{2\pi}{N}kn\right) - i\sin\left(\frac{2\pi}{N}kn\right)$ and added together to compute the coefficient for frequency k . This is also the source of the inefficiency of this algorithm. Because there are two nested loops that go from 0 to $N - 1$, the time complexity of this algorithm is $\mathcal{O}(n^2)$, which means that given an input of size n , the computer will make n^2 operations. This is very inefficient and slow for real-life applications of DFT, as the sizes of inputs for those applications can get very big. For example, a WAV file containing 1 second of audio signal has 44'100 samples. Processing a short song would take a lot of computing resources. This is why in most actual implementations of DFT, an algorithm called Fast Fourier Transform (FFT) is used, whose time complexity is $\mathcal{O}(n \log(n))$.

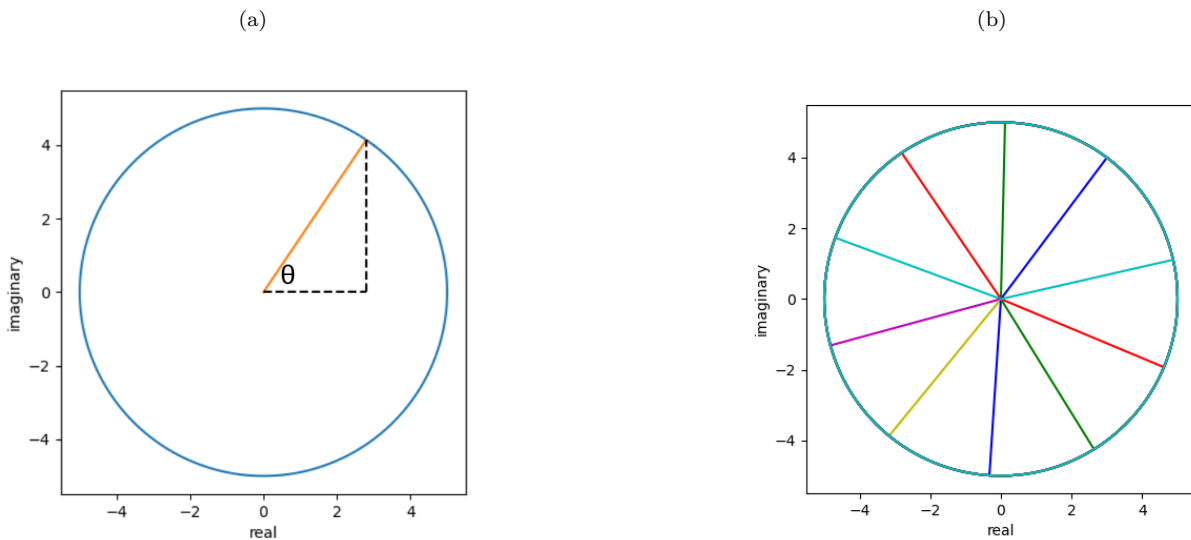
3.4 Fourier series and making drawings

The last stage of the program is to recreate the original drawing using the inverse discrete Fourier transform (IDFT), which is the same as discrete Fourier series². The formula is given in equation (16).

$$x_n = \frac{1}{N} \sum_{k=0}^{N-1} X_k \cdot \left[\cos\left(\frac{2\pi}{N}kn\right) + i\sin\left(\frac{2\pi}{N}kn\right) \right] \quad (16)$$

In discrete Fourier series, the means of coefficients acquired through DFT ($\frac{1}{N}X_k$) are used like weights a_0 , a_n and b_n in case of real form of Fourier series discussed in section 2. We already know that expression $\cos\theta + i\sin\theta$ returns complex value that represents a position at angle θ on a unit circle on a complex plane. The mean of the coefficient X_k regulates the radius and phase of such circle. In figure 12a, there is one such circle multiplied by $3 + 4i$. We can see that the radius of this circle is the hypotenuse of right triangle formed by real and imaginary component of the complex number. Angle θ on the plot is the phase shift caused by the term $3 + 4i$, which we can think of as the point from which the circle is drawn. Now, even though we are dealing with discrete values in IDFT, the values will “rotate” as n increases - Figure 12b. Additionally, each term has its k , which is the frequency, in this case meaning that values “rotate” k times within one period.

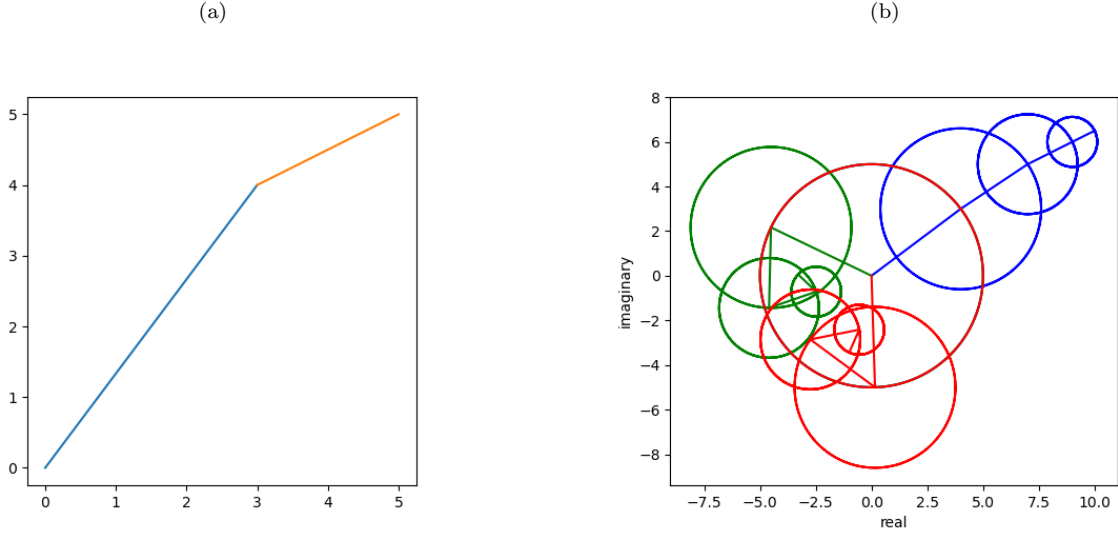
Figure 12



²terms IDFT and discrete Fourier series will be used interchangeably in this section

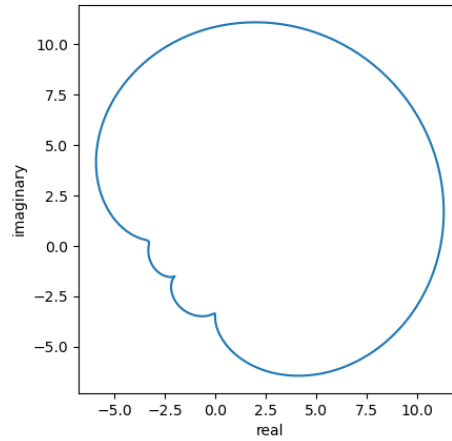
Because complex numbers are being added in the series, a good way to visualize this operation is by seeing complex numbers as lines that are “attached” together by their ends like in Figure 13a. Then, in the context of IDFT, we can think that each of those lines “rotate” by $\frac{2\pi}{N}k$ as we compute the next term - Figure 13b.

Figure 13



Since the plot is merely a representation of all terms summed in the series, the output value x_n is at the end of each of collection of lines representing complex numbers. If we now plot the values returned by discrete Fourier series with the same coefficients (Figure 14), we can see some shape - and such shapes will be the drawings generated by the program.

Figure 14



One last thing to point out is that, like in case of real form of Fourier series, the more terms we add, the closer the output will converge to the input. So what the program does to visualize this convergence is it simply regulates the number of terms in the series, however this will be presented in the renders from the program.

See Appendix C for the script used to generate all plots in the section

3.4.1 Implementation of IDFT in code

Below is an implementation of IDFT in the source code of the program.

```

void draw_IDFT(C_set &x_k, VertexArray &drawing, unsigned n)
{
    C temp, x_n;
    unsigned N = x_k.size();
    double step = (2 * M_PI) / N, theta;

    for (unsigned t = 0; t < N; ++t)
    {
        for (unsigned k = 0; k < n; ++k)
        {
            theta = step * t * k;
            temp = C(cos(theta), sin(theta));

            x_n += (temp * x_k[k]) / N;
        }

        drawing[t] = Vector2f(x_n.re + 400, -x_n.im + 400);
        x_n = C(0,0);
    }
}

```

The function takes a set of coefficients computed by DFT function (`x_k`), an array of coordinates to be put on the screen (`drawing`) and a variable `n`, which allows for controlling how many terms are being added in the series. The upper for loop iterates over one period in $N - 1$ steps in order to get coordinates for a full period. The inner loop iterates from 0 to n so that the user can regulate the number of arguments being added in the series and inside this loop is a direct implementation of IDFT. At the bottom, the complex number `x_n` is converted to coordinates that work with SFML and is saved to the `drawing` array.

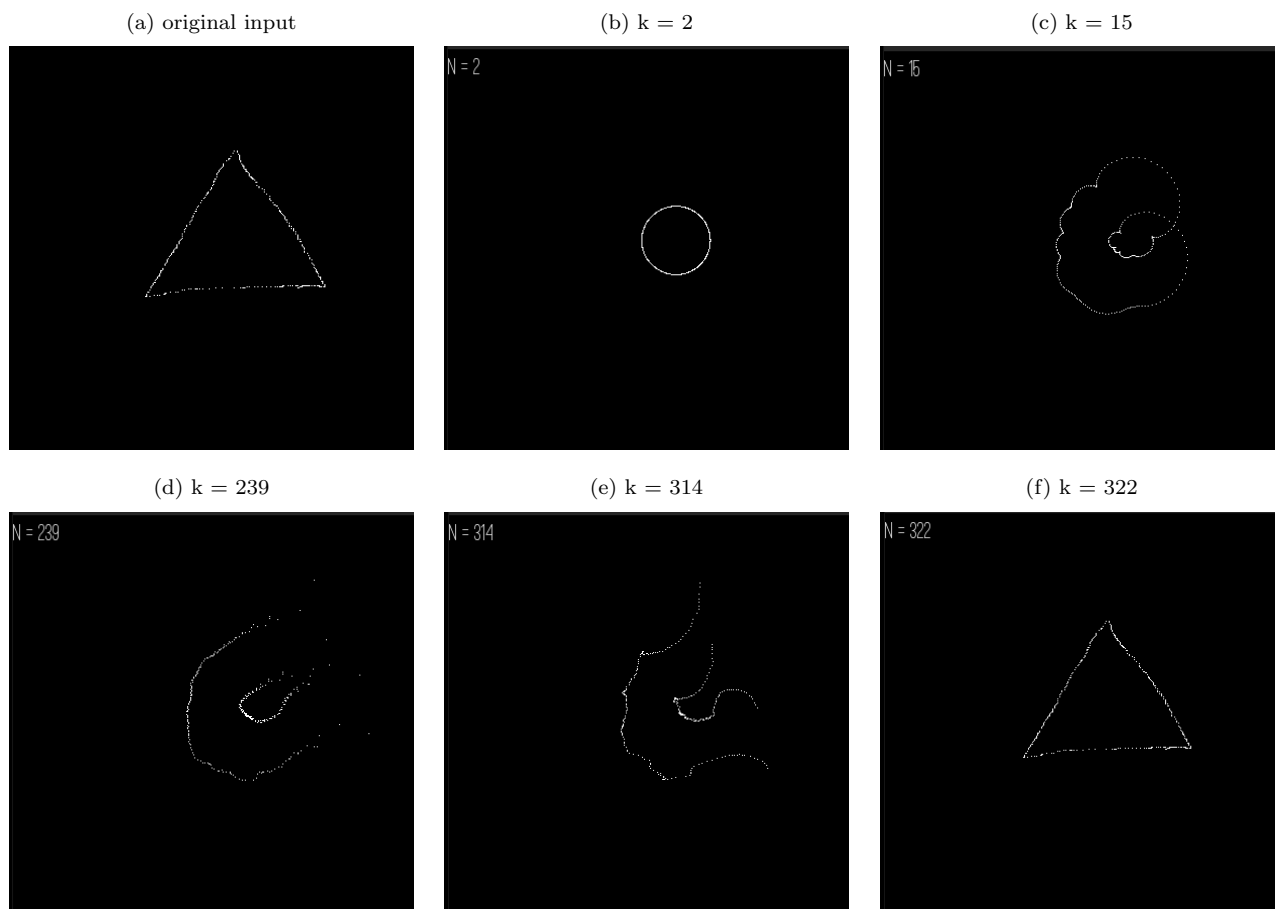
See Appendix D for instructions to compile the program on UNIX machine

4 Drawings

Below are sample examples of the program working. A few frames from the program have been picked in order to best depict the working of both program and discrete Fourier series. N in the top-left corner is the size of set containing DFT coefficients.

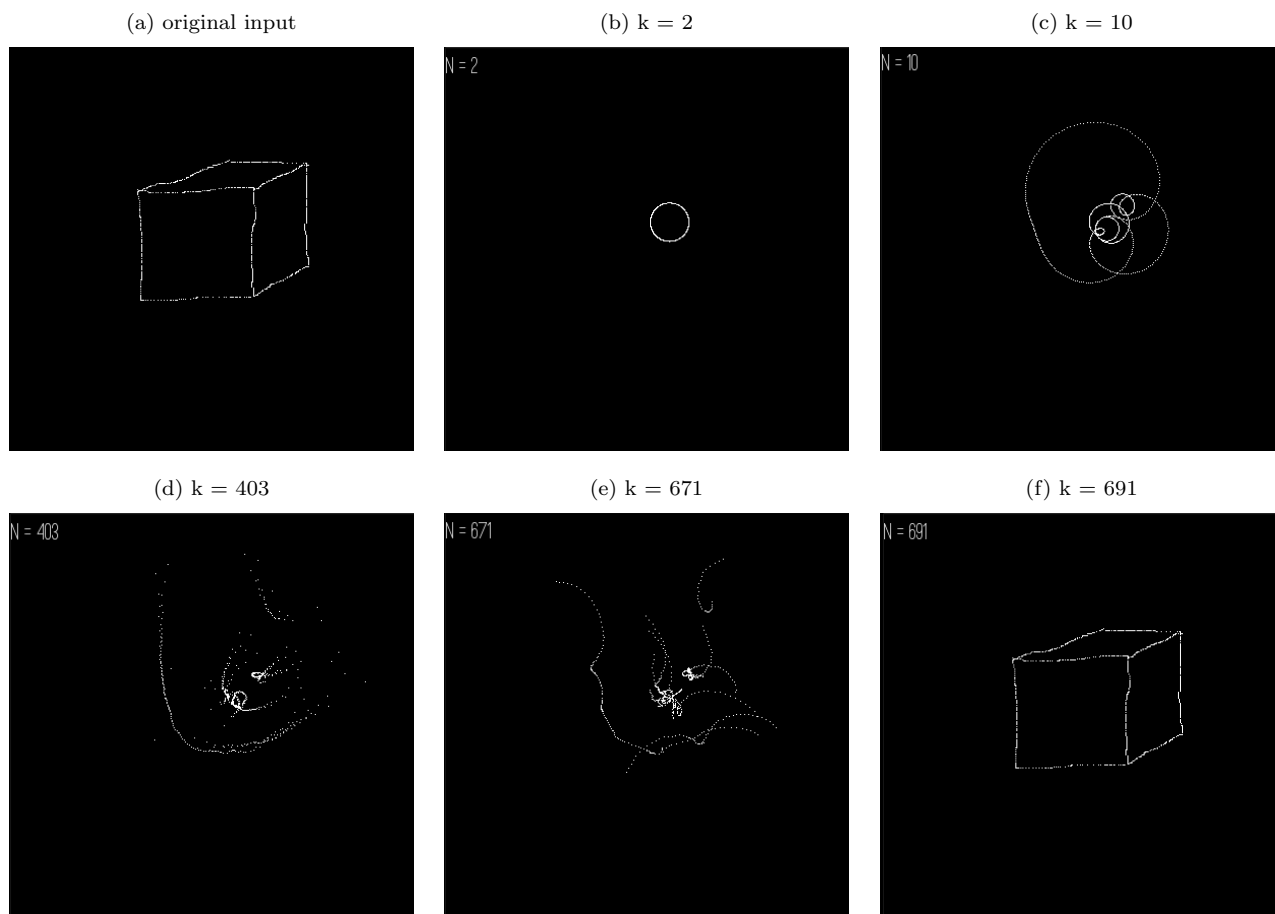
4.1 Triangle

Figure 15



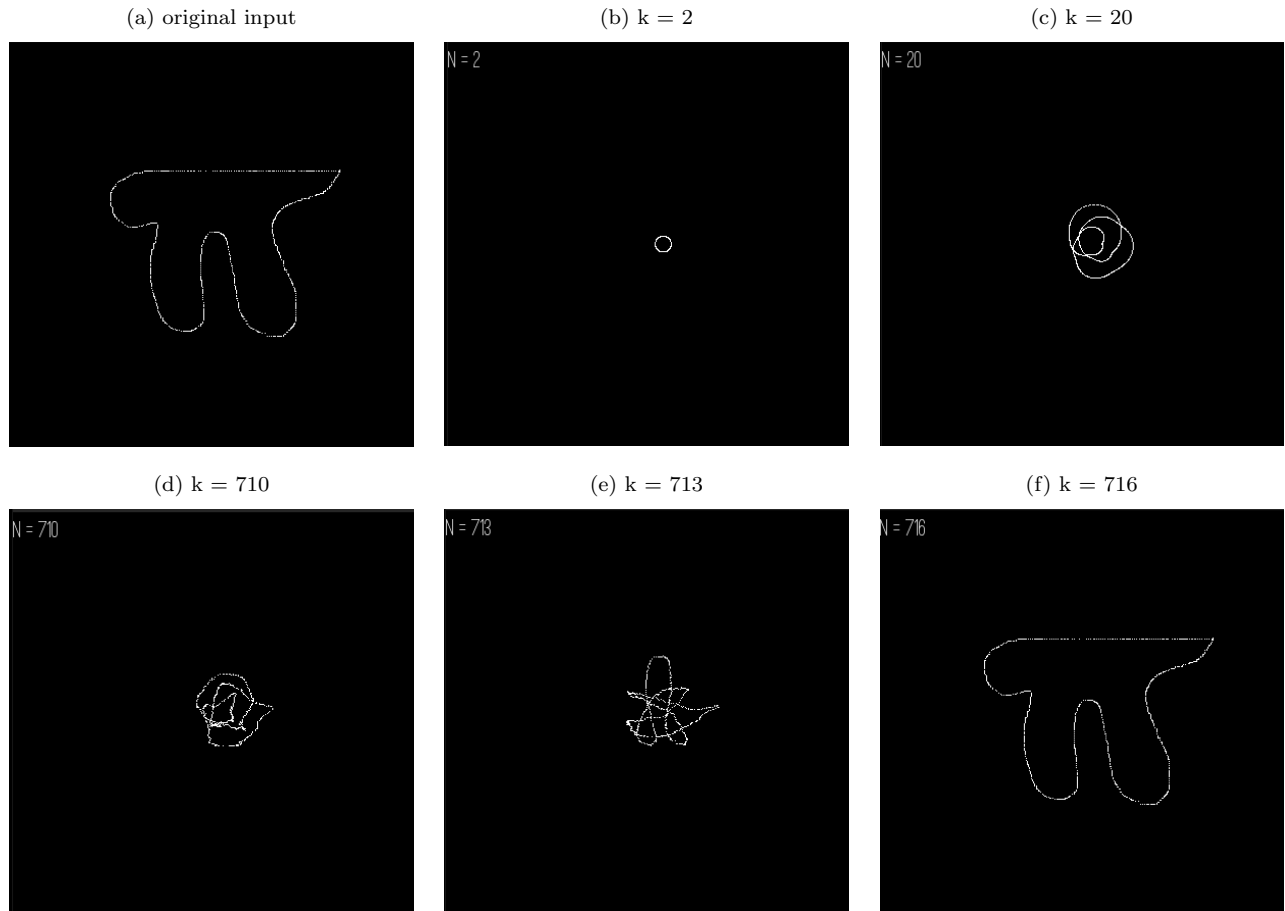
4.2 Cube

Figure 16



4.3 Pi

Figure 17



5 Appendices

5.1 Appendix A

5.2 Appendix B

Python script used to depict workings of DFT.

```
def wrap_function(func_in, freq):  
  
    N = len(func_in)  
    data = np.zeros(N, dtype = np.complex_)  
    theta = (6.28 / N) * freq  
    coeff = complex(0,0)  
  
    for n in range(0, N):  
        c = complex(m.cos(theta * n), -(m.sin(theta * n)))  
        data[n] = func_in[n] * c  
        coeff += data[n]  
  
    coeff /= N  
  
    return (data, coeff)
```

```
def plot_complex(data, name):

    plt.plot(data[0].real, data[0].imag)
    plt.scatter(data[1].real, data[1].imag)
    plt.axhline(y=0, color='k')
    plt.axvline(x=0, color='k')
    plt.savefig(sys.argv[1])

    plt.show()
```

Modules needed to run the script:

- numpy
- matplotlib
- math

5.3 Appendix C

Python script used to generate data for plots used in explaining IDFT.

```
def generate_fs(coeff, freq, offset):

    fs = np.fromfunction(
        np.vectorize(lambda i :
            complex(np.cos((i * freq) / 100),
                np.sin((i * freq) / 100)) * coeff + offset),
        (629,),
        dtype=complex)

    return fs
```

Python script used to plot data.

```
def plot_fs(coeff, epicycles, n, step):

    colors = ['b', 'g', 'r', 'c', 'm', 'y']

    if epicycles:
        for k in range(n):
            offset = complex(0,0)
            fs = list()

            for i in range(0, len(coeff)):
                fs.append(generate_fs(coeff[i], i+1, offset))
                offset += (fs[i][k * step] - offset)

            offset = complex(0,0)

            for f in fs:
                color = colors[k % len(colors)]
                plt.plot(f.real, f.imag, color)
                plt.plot([offset.real, f[k * step].real], [offset.imag, f[k * step].imag], color)
                offset += (f[k * step] - offset)
    else:
        fs = list()
```

```

for i in range(0, len(coeff)):
    fs.append(generate_fs(coeff[i], i+1, 0))
out = sum(fs)
plt.plot(out.real, out.imag)

plt.gca().set_aspect('equal', adjustable='box')
plt.xlabel('real')
plt.ylabel('imaginary')
plt.show()
plt.savefig('dfs_explainer.png')

```

5.4 Appendix D

Prerequisites

- gcc / Clang / LLVM compiler for C++11 and higher
- SFML media library

Compilation and use guide for Fourier drawer program for UNIX machines:

1. Clone git repository to your machine from the link: https://github.com/ikrzywda/fourier_drawer
2. change directory to `fourier_drawer/drawer`
3. use `make` to compile the program
4. type `./fourier_drawer` inside `drawer` directory to launch the program

Controls

- **left mouse button** to draw inside Fourier sketchpad window
- **Enter** to accept input given in Fourier sketchpad window and to terminate the program when in Fourier drawer window
- **K** to increase the number of arguments in Fourier drawer window
- **J** to decrease the number of arguments in Fourier drawer window

6 Bibliography

- “But what is the Fourier Transform? A visual introduction.” YouTube, uploaded by 3Blue1Brown, <https://www.youtube.com/watch?v=spUNpyF58BY>
- “But what is a Fourier series? From heat flow to circle drawings — DE4” YouTube, uploaded by 3Blue1Brown, <https://www.youtube.com/watch?v=r6sGWTCMz2k>
- “Electrical engineering : Signals and systems : Fourier series” Khan Academy, <https://www.khanacademy.org/science/electrical-engineering/ee-signals#ee-fourier-series>
- “Discrete Fourier transform”, Wikipedia, https://en.wikipedia.org/wiki/Discrete_Fourier_transform