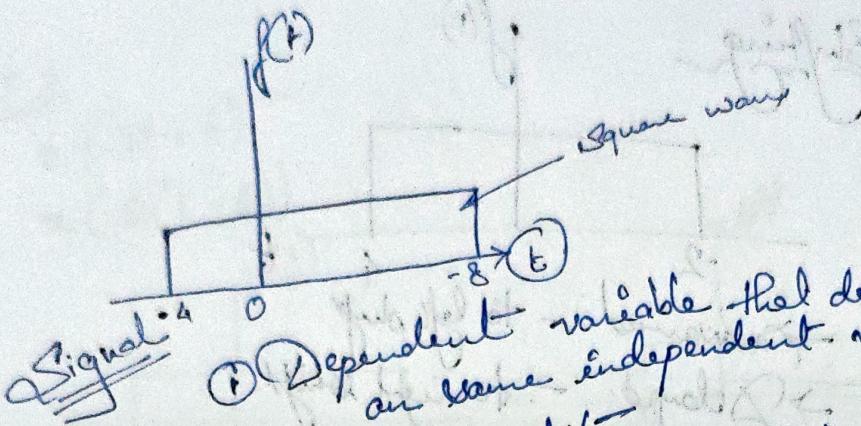


Signal & Systems



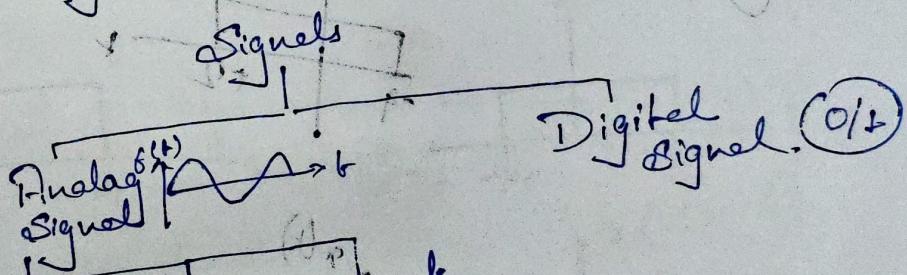
- defn of signal.
- ① Dependent variable that depends on some independent variable (like time, space, etc)
 - ② Physical quantity
 - ③ Have same information that can be displayed, manipulated and even conveyed.

System

① Conversion of one form of signal to another form of signal

Two types of Signals

An



Continuous
in time
Continuous
value

Discrete
in time
Continuous
value

Continuous
in time
Discrete
value

Discrete
time
Discrete
value

Digital
Signal. (0/1)

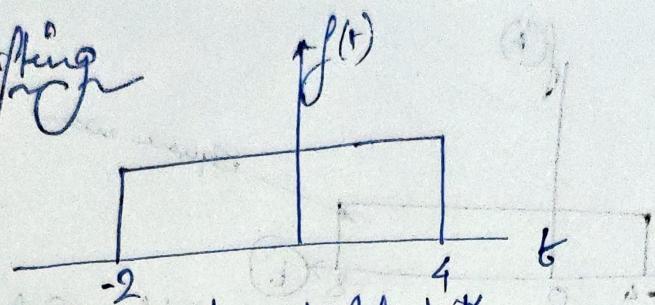
* A discrete signal is analog signal

Operations on Signal

- ① Trim Scaling
- ② Trim Shifting
- ③ Trim Reversal

① Time Shifting

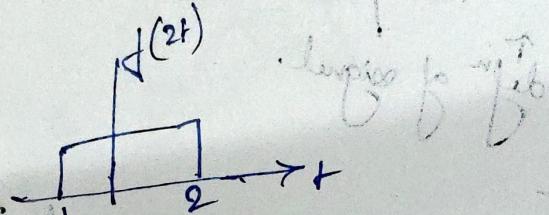
$$f(t) =$$



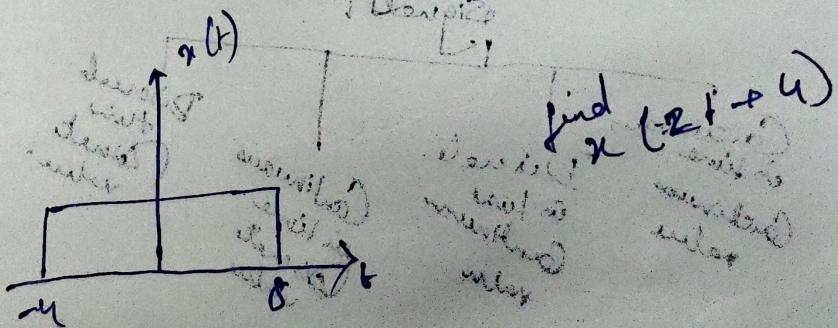
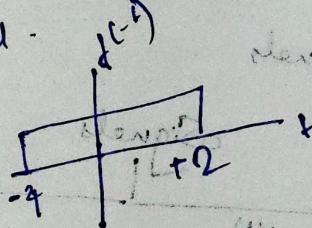
$f(t+3) \rightarrow$ advanced \rightarrow left shift

$f(t-3) \rightarrow$ delayed \rightarrow right shift

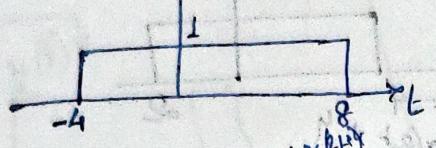
② Time Scaling



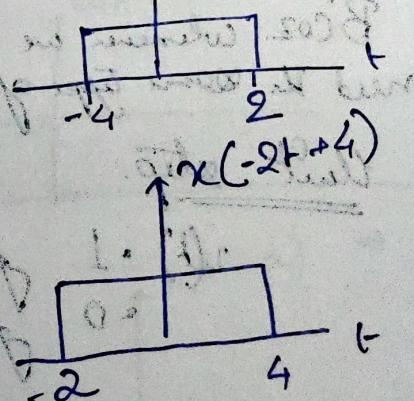
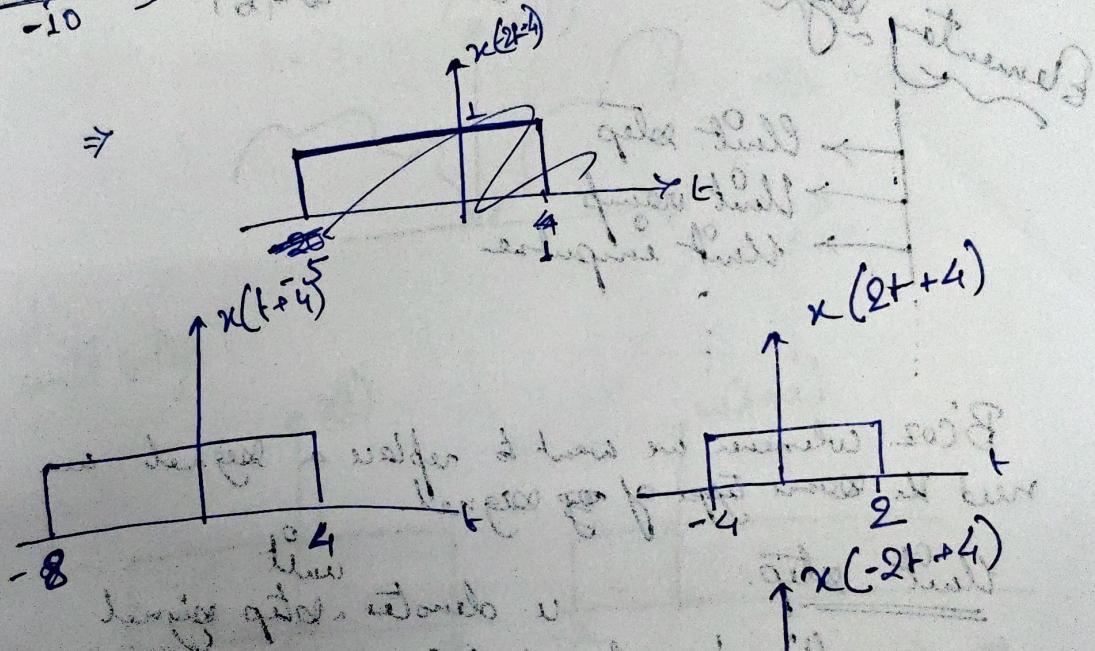
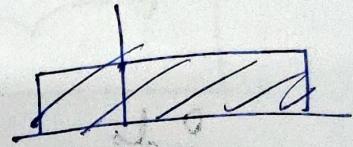
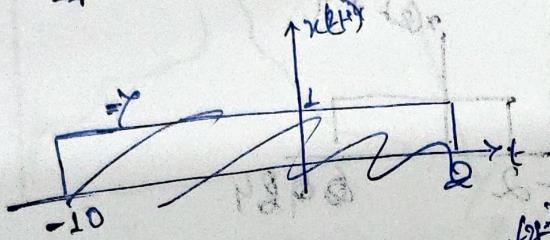
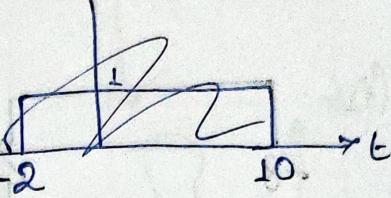
③ Time and reversal



find
 $x(-2t+4)$
 $x(-2(t-2))$
 $x(t)$

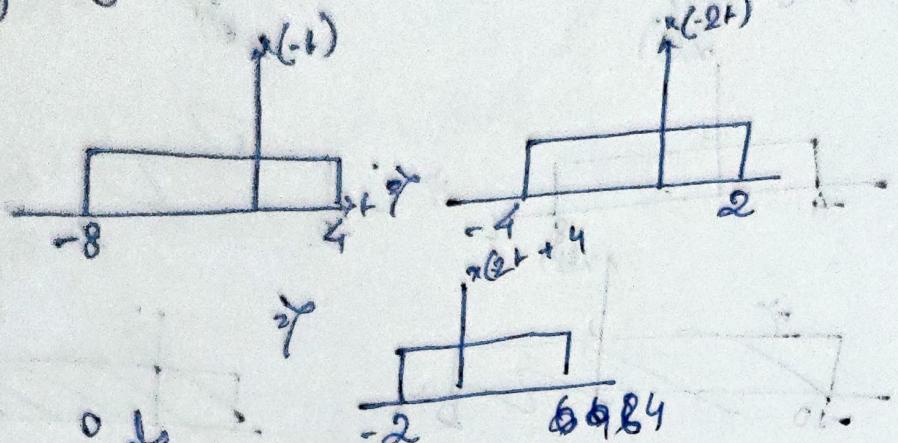


$(+) x(2t)$



reflect and shift towards right
 and subtract it from
 reflect about 0 and add
 because of this

from right to left.



Elementary signals

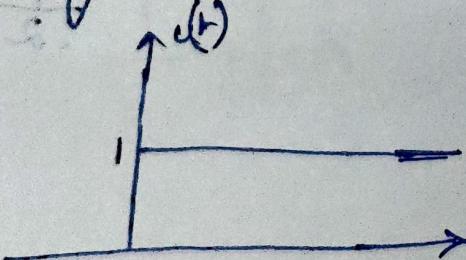
- Unit step
- Unit ramp
- Unit impulse

B'coz whenever we want to replace a signal we need the same type of ~~or~~ signal.

Unit step

u denotes ^{unit} step signal

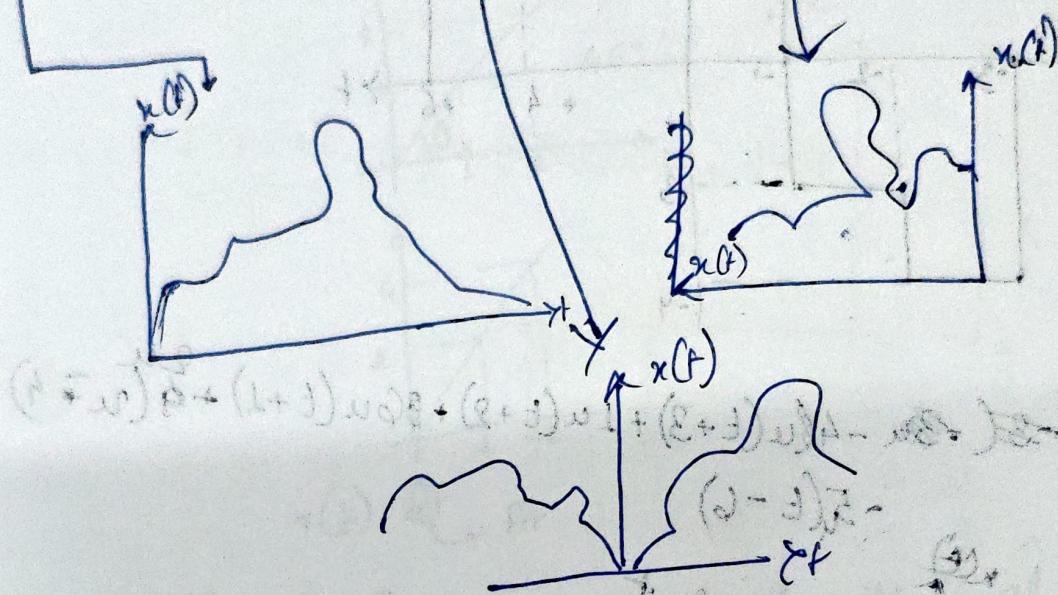
$$u(t) = \begin{cases} 1 & \text{for } t \geq 0 \\ 0 & \text{for } t < 0 \end{cases}$$



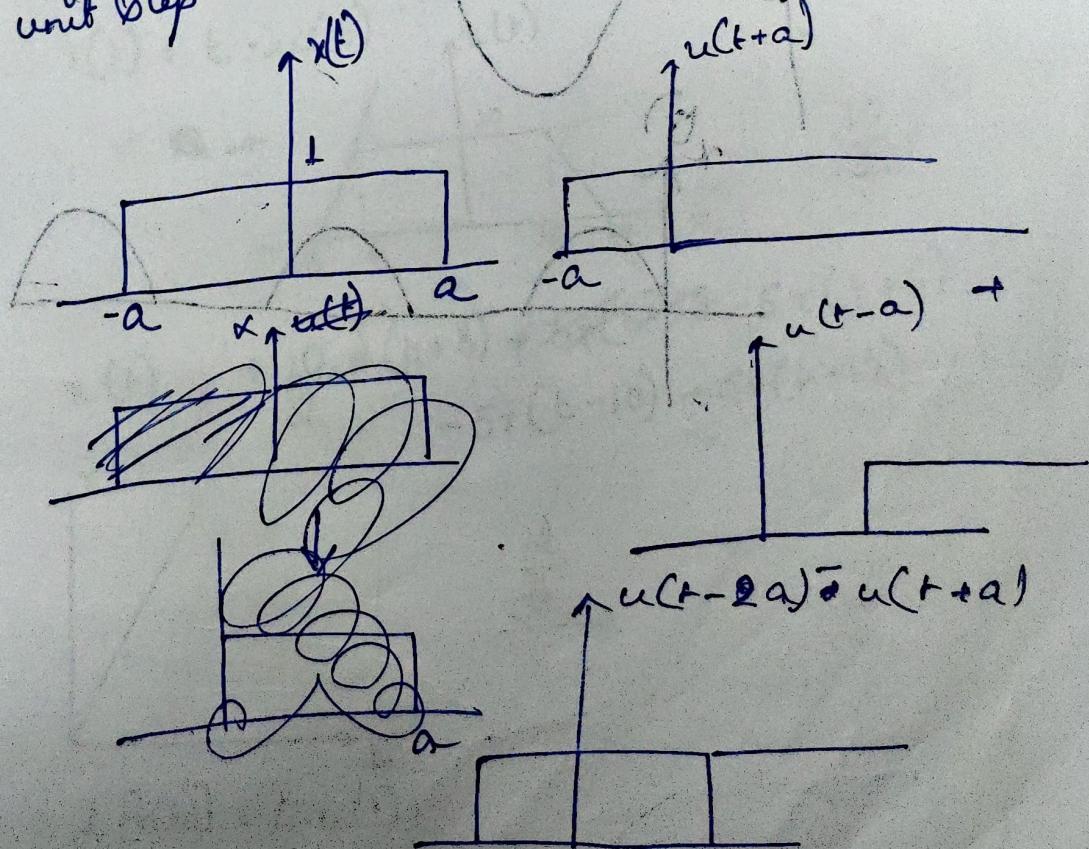
Scaling operation gives ⁶ same answer.

but for it starts a value other than 0 then that value will be divided

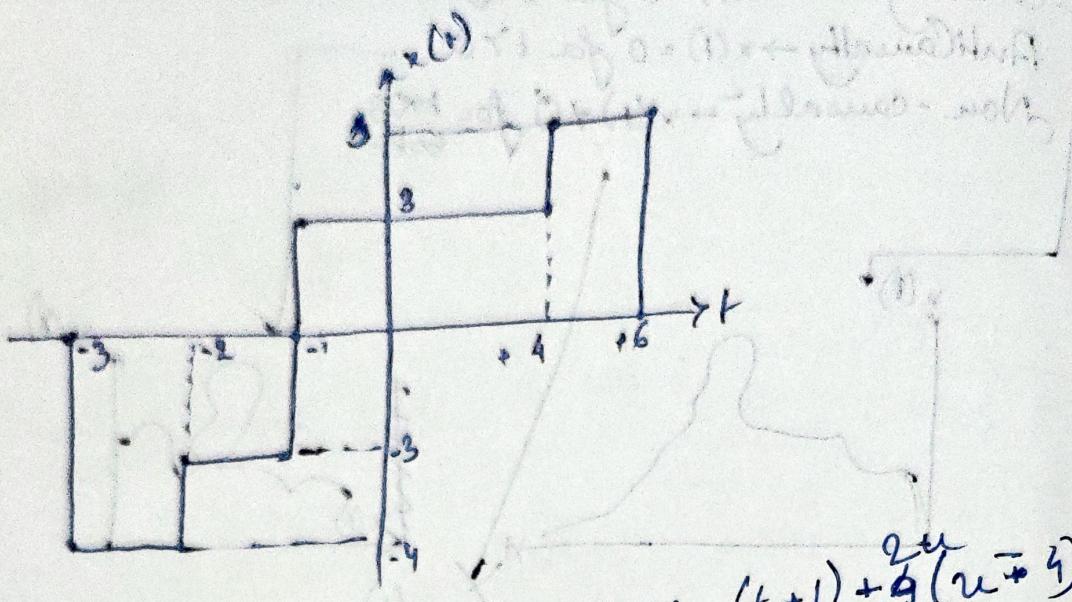
Causality $\rightarrow x(t) = 0$ for $t < 0$
 Anti-Causality $\rightarrow x(t) = 0$ for $t > 0$
 Non-causality $\rightarrow x(t) \neq 0$ for both



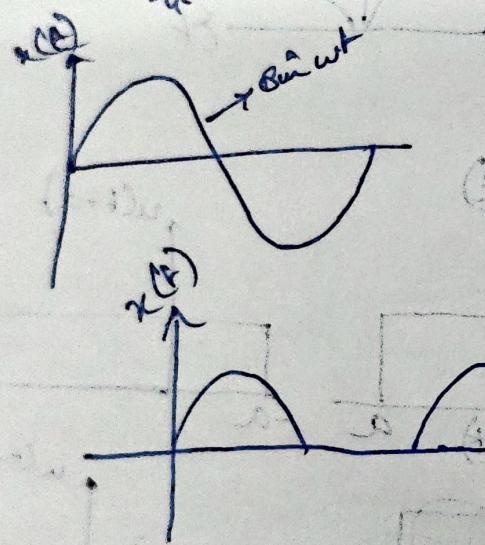
① To unit step



$$Au(t+a) = Au(t-a)$$

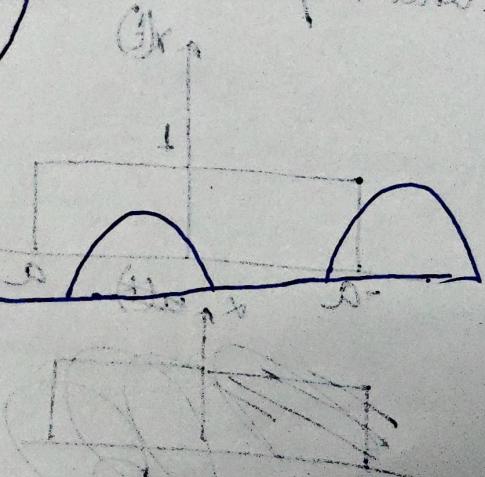


$$u(t) = -4u(t+3) + 1u(t+2) + 36u(t+1) + \frac{2}{5}u(t+4) - 5u(t-6)$$



$$u(t) = (\sin \omega t) = ?$$

Ques & draw. of P.O.



$\sin(\omega t) = \text{Ans}$

(Ans DWA & CWA)

Ramp signal

$$r(t) = \begin{cases} t & \text{for } t \geq 0 \\ 0 & \text{otherwise} \end{cases}$$

$$r(t)$$

$$t$$

$$2$$

$$1$$

$$0$$

$$-1$$

$$-2$$

$$-3$$

$$-4$$

$$-5$$

$$-6$$

$$-7$$

$$-8$$

$$-9$$

$$-10$$

$$-11$$

$$-12$$

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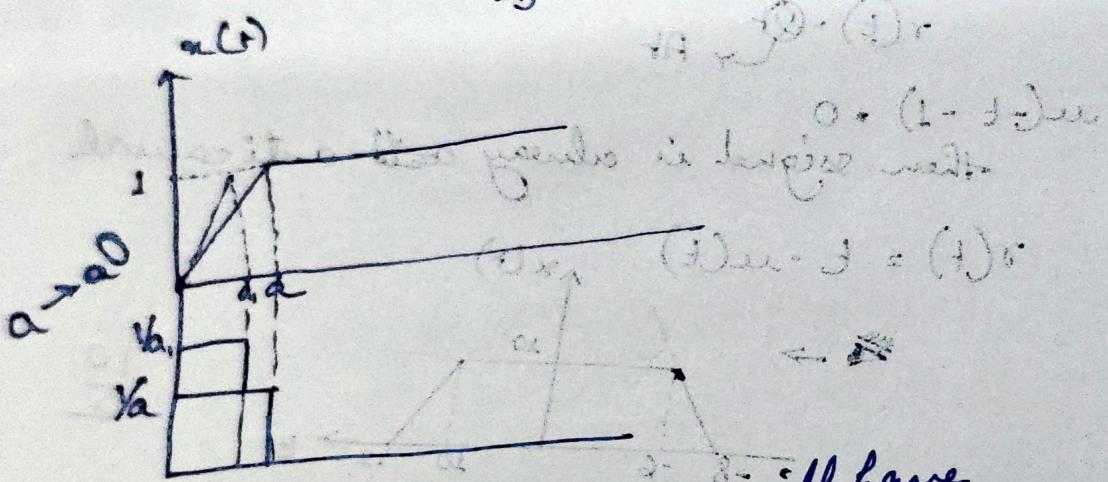
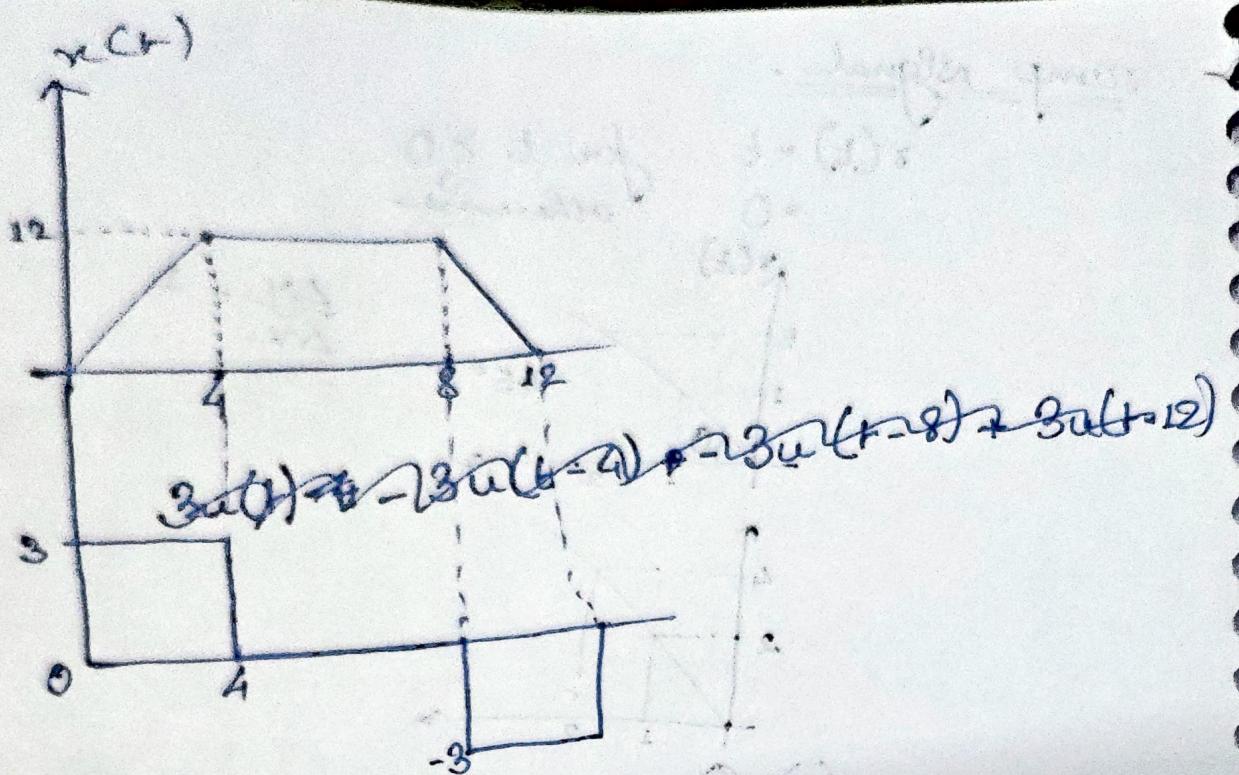
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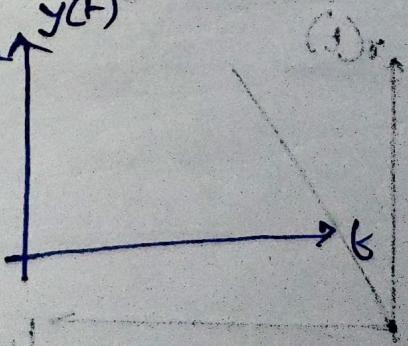
<math display



Derivative of a step signal we will have impulse signal $(\delta(t) \cdot \frac{d}{dt} u(t))$

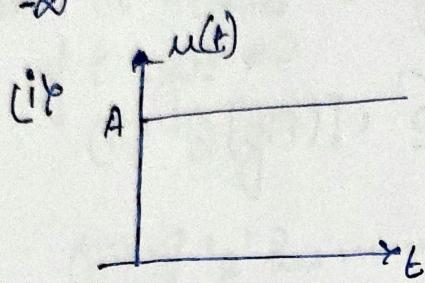
For

value = ∞
impulse
max area = 1 = $\delta(t)$

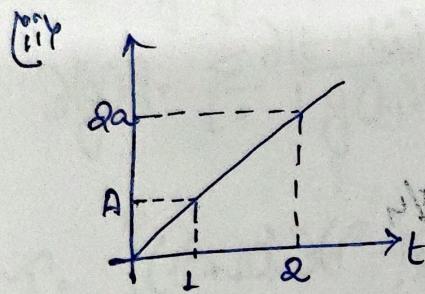


$\delta(t) = 1$ at $t=0$

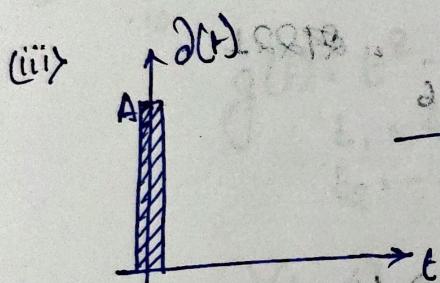
$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$



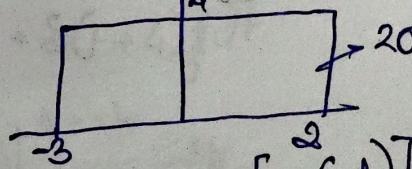
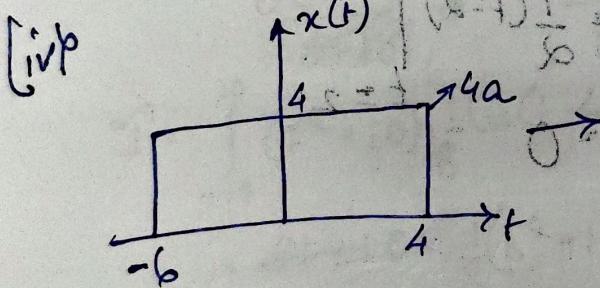
$\rightarrow A u(t) \rightarrow A \rightarrow \text{magnitude}$



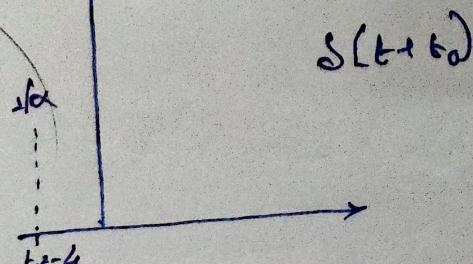
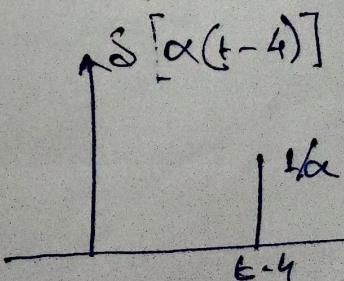
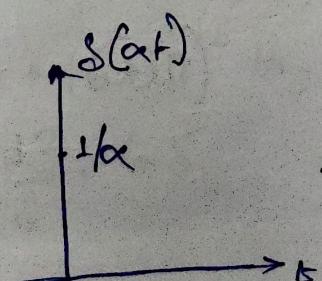
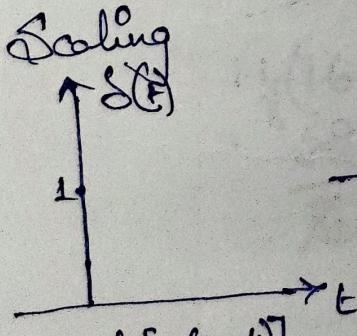
$\rightarrow A r(t) \rightarrow A \rightarrow \text{slope}$



$\rightarrow A \delta(t) \rightarrow A \rightarrow \text{rectangular}$



$$\text{area}[x(at)] = \frac{1}{\alpha} \text{area}[x(t)]$$



$$x(t) \cdot \delta(t - t_0) = x(t_0)$$

$$x(t) \cdot \delta(t + t_0) = x(-t_0)$$

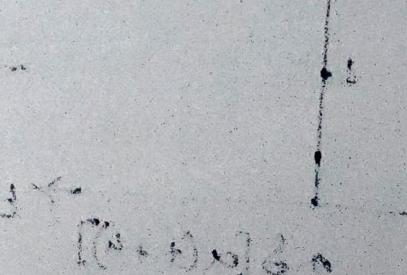
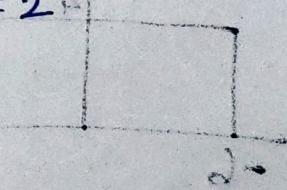
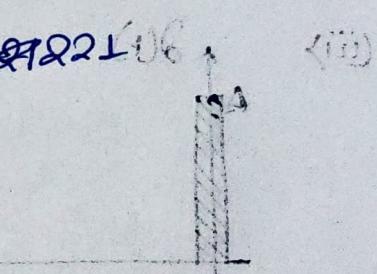
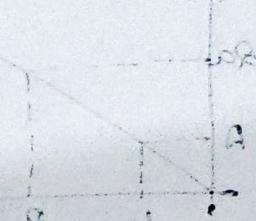
$$\begin{aligned} Q &= \int_{-6}^{+8} (t^2 + 4) \cdot \delta(t - 3) dt \\ &= [t^2 + 4] \Big|_{t=3} = 13 \end{aligned}$$

$$\int_{-\pi}^{\pi} \cos^2 t \cdot \delta(t - \pi/4) dt = \left. \cos^2 t \right|_{t=\pi/4} = 1$$

$$\int_{-3}^4 (t^3 + 5) \delta(t - 6) dt = [t^3 + 5] \Big|_{t=6} = 6^3 + 5 = 221$$

$$\int_{-6}^{+5} (t - 2) \delta(2t - 4) dt =$$

$$\int_{-6}^{+5} (t - 2) - \frac{1}{2} \delta(t - 2) dt = \frac{1}{2} (t - 2) \Big|_{t=2} = 0$$



$$\text{Property of integral response}$$

$$I = \int x(t) \cdot \partial g(t) \} dt$$

Condition (i) $\partial[g(t)] > 0$ if 'g' is nowhere 'zero' for any value of t , $g(t) \neq 0$

(ii) $\partial[g(t)] = \frac{\partial(t - t_0)}{g'(t_0)}$ if 'g' has one real root at $t=t_0$.

(iii) If 'g' has more than one real root:

$$\partial[g(t)] = \sum_{i=0}^n \frac{\partial(t - t_i)}{|g'(t_i)|} \leq \frac{\partial(t - t_0)}{|g'(t_0)|} + \frac{\partial(t - t_1)}{|g'(t_1)|} + \dots$$

$$I = \int_{-10}^{10} (t^2 + 10) \partial(t^2 - 16) dt$$

$$g(t) = t^2 - 16$$

$$\begin{aligned} t_1 &= 4 \\ t_2 &= -4 \end{aligned}$$

roots of t

$$\frac{\partial(t-4)}{|8|} + \frac{\partial(t+4)}{|8|} = \frac{\partial(t-4) + \partial(t+4)}{8}$$

$$I = \int_{-10}^{10} (t^2 + 10) \cdot \frac{1}{8} (S(t-4) + S(t+4)) dt$$

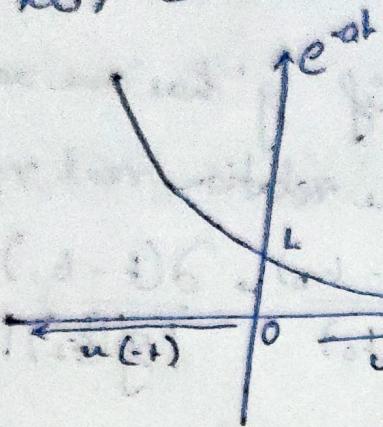
$$= \frac{1}{8} (t^2 + 10) \Big|_{t=-4}^{t=+4} dt$$

$$= \frac{1}{8} (16 + 10 + 16 + 10)$$

$$= \frac{52}{8} \cdot \frac{13}{2} \cdot 6.5$$

Exponential function

$$x(t) = e^{-at}$$



e^{-at} - Non-causal

to causal/anticausal

$$e^{-at} \cdot u(t)$$

$$x = e^{-at} u(t)$$

$$e^{-at} u(t) = e^{-a(t)} \cdot u(t) + e^{-a(t)} \cdot u(-t)$$

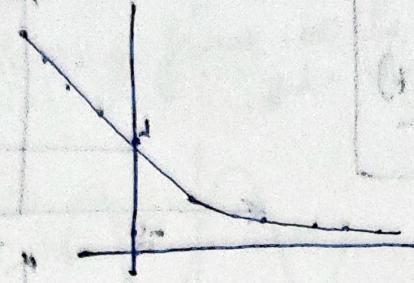
$$= e^{-at} \cdot u(t) + e^{at} \cdot u(-t)$$

Decay rate

$$e^{at} \cdot u(t) + 0.1 \cdot u(t) \cdot e^{-at}$$

$$= 2.0 \cdot 81 \cdot \frac{82}{80}$$

$$③ x(t) = e^{-at^2}$$

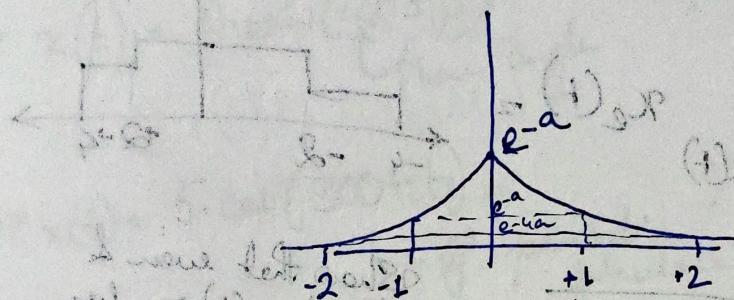


$$\begin{aligned} & (1) x + (3x - 1)x = (4)x \\ & 3x - 1 = 4x \\ & 3x = 1 \end{aligned}$$

$$e^{-at^2} \cdot u(t)$$

$$e^{-at^2} = e^{-a(t^2)} \cdot u(t) + e^{-a(t^2)} \cdot u(-t)$$

$$e^{-a(t^2)} \cdot u(t) + e^{-a(t^2)} \cdot u(-t)$$



It is not perfectly exponential &
the fig. is known gaussian structure
1) Start at 1 if coefficient to one
2) Both side change with $e^{-a(t^2)}$

Classification of Signals

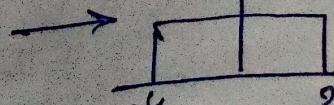
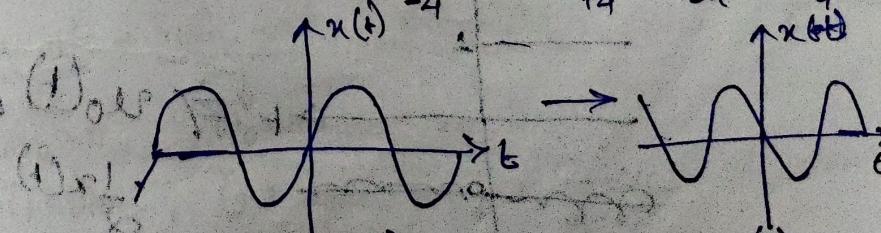
① Even and odd signals.

$$\text{Theorem } x(-t) = x(t) \rightarrow \text{even}$$

$$x(-t) = -x(t) \rightarrow \text{odd.}$$

$$x(t)$$

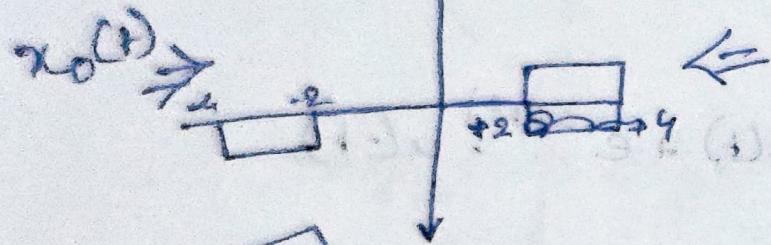
$$x(t)$$



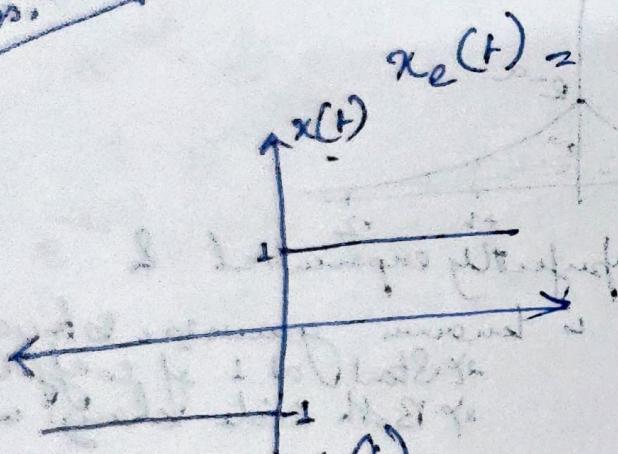
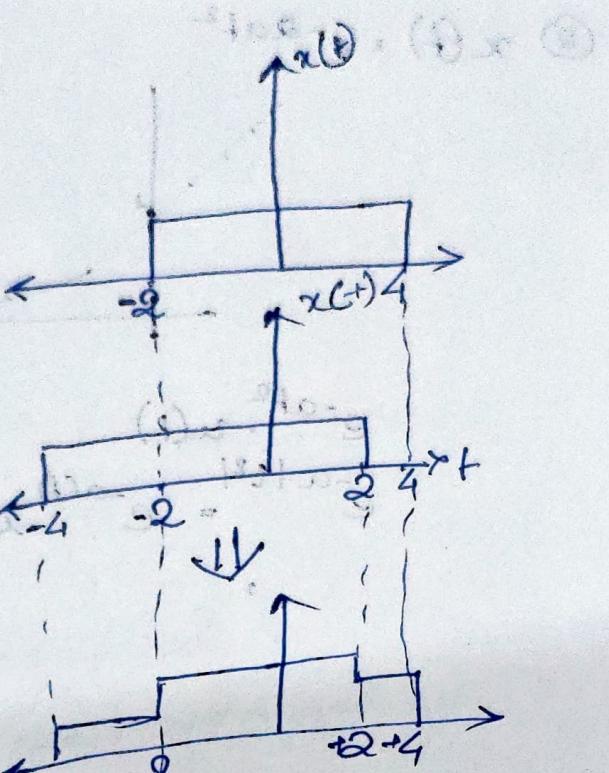
neither even nor odd.

$$x_e(t) = \frac{x(t) + x(-t)}{2}$$

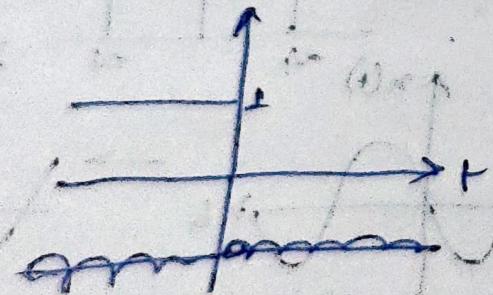
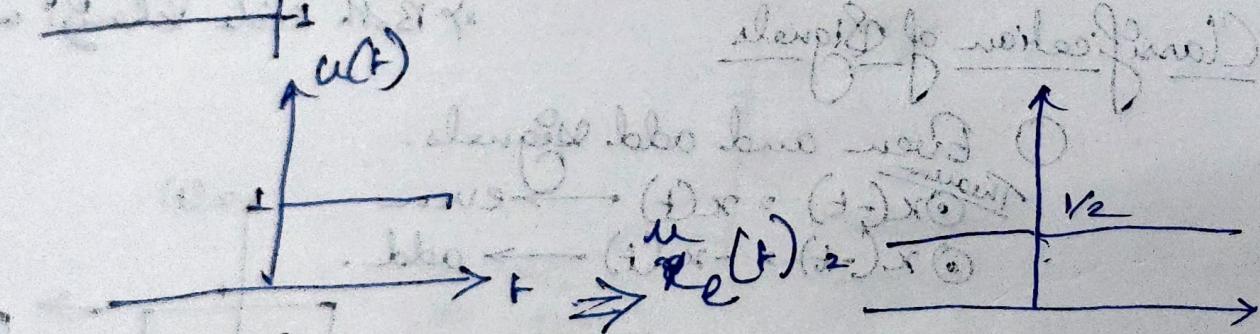
$$x_o(t) = \frac{x(t) - x(-t)}{2}$$



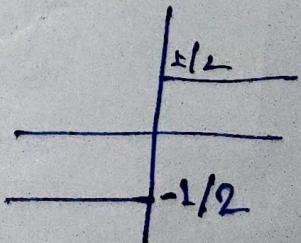
Q. pos. is imp



Show that even & odd part of $x(t)$ can be represented in terms of $x_e(t)$



$$x_o(t) = \frac{1}{2}x(t) - \frac{1}{2}x(-t)$$



Periodic signals
Any signal that repeats itself from after certain time
must maintain range from $-\infty$ to $+\infty$
since slightly has no effect in periodicity

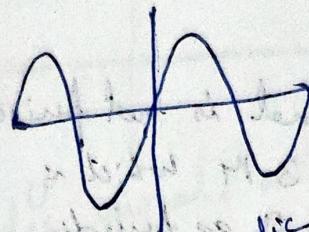


Fig. periodic signal

(i) $x(t) = A \cdot \sin(\omega t \pm \phi)$

amplitude

Phase angle

phase shift

phase angle related to angular frequency

(ii) $x(t) = 5 \cdot \cos(300\pi t)$

if periodic then fundamental period.

(iii) $x(t) = 5 \cdot \cos(20\pi t + \pi/2)$

Ans(ii) Step 1: Compose of generalized form i.e. $x(t) = A \sin(\omega t \pm \phi)$

$$\omega_0 = 300\pi$$

$$T = \frac{2\pi}{\omega_0}$$

$$\frac{2\pi}{300\pi} \text{ sec}$$

$\frac{1}{150}$ sec

fundamental period.

Suppose, $x(t) = 4 \cos \pi t + 3 \sin 2\pi t + 2 \sin 3\pi t$

$$\omega_0 = \pi, \omega_0 = 2\pi, \omega_0 = 3\pi$$

then $T_1 = 2 \text{ sec}, T_2 = 1 \text{ sec}, T_3 = \frac{2}{3} \text{ sec}$

$$\frac{\frac{1}{2}}{T_1} = \frac{1}{2} \text{ or } \frac{T_2}{T_1} = \frac{2}{3} \text{ or } \frac{T_1}{T_3} = \frac{2}{3}$$

$$\frac{T_2}{T_1} = \frac{T_3}{T_2} = \frac{T_1}{T_3} = \text{Rational}$$

Rational numbers

1. Ratio of p/q where p, q are integers
2. Decimal values must be terminating/repeating

Note

If value is π then signal is not periodic
 But system processes $\pi - 3.14$ which is periodic
 So the sys. considers it as periodic

$$x(t) = 4\cos \pi t + 3\sin 2\pi t + 2\sin 3\pi t$$

$$\text{fundamental period} = \text{LCM}\{T_1, T_2, T_3\}$$

$$= \text{LCM}\{2, 1, \frac{2}{3}\}$$

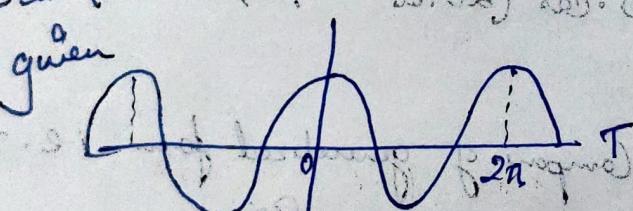
$$= \text{LCM}\{2, 1, 2\} = 2$$

$$= \frac{\text{HCF}\{1, 2, 3\}}{\text{HCF}\{1, 1, 3\}} = \frac{1}{1} = 2$$

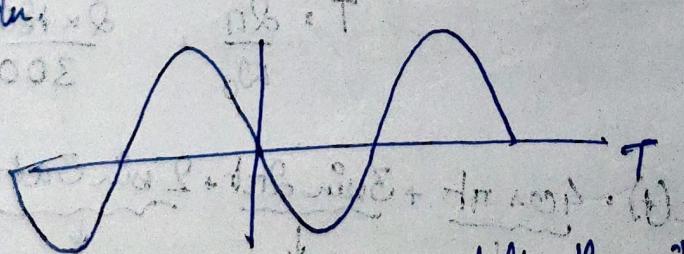
Fundamental period = 2

Q) $x(t) = 5\cos(20\pi t + \pi/2)$

given



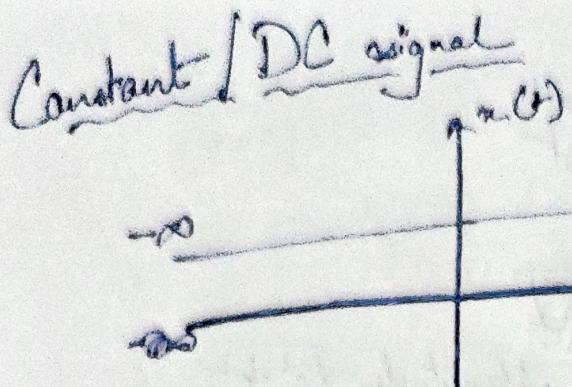
Solu.



If there is phase shift then it doesn't hamper the periodicity of signal.

$$W_0 = 20\pi$$

$$\text{Ans. } \frac{1}{T}, \frac{2\pi}{T}, \frac{1}{2\pi T}$$



periodic signals without periodic values

$$w(t) = 4 + 5 \cos(20\pi t + \pi/2)$$

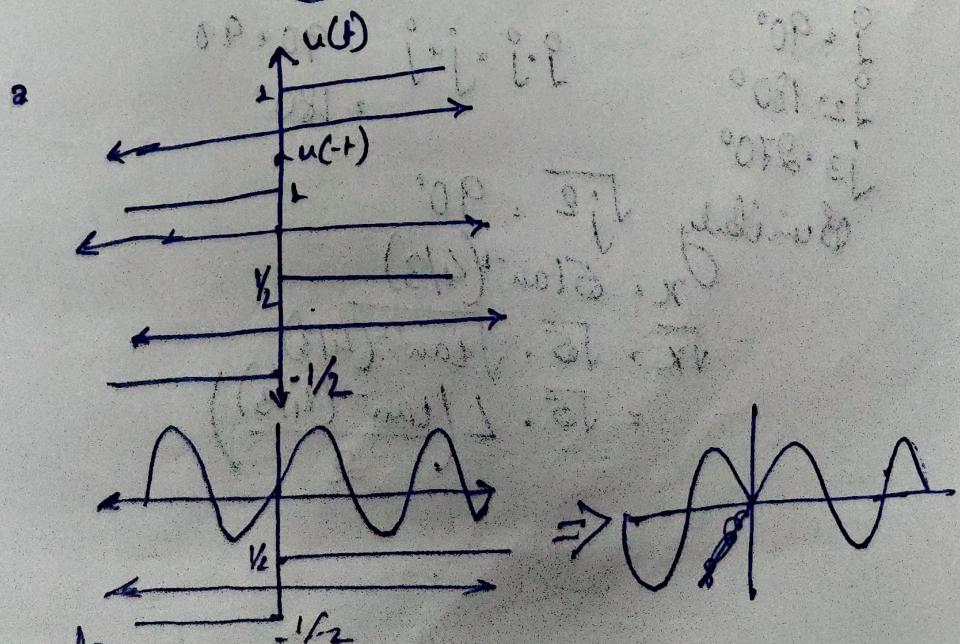
If this is periodic then
whole signal is periodic

$$u(t) = \text{Even of } [\sin 4\pi t \cdot u(t)]$$

$$\begin{aligned} x_1(t) &= \sin(4\pi t) \cdot u(t) = -\sin 4\pi t \cdot u(-t) \\ x_2(t) &= \sin(4\pi t) \cdot u(t) \end{aligned}$$

$$= \frac{\sin 4\pi t \cdot u(t) + \sin 4\pi t \cdot u(-t)}{2}$$

even part of $u(t)$



Proved
Not periodic.

Complex exponential Signal

$$x(t) = e^{j\omega_0 t} = \cos \omega_0 t + j \sin \omega_0 t$$

Real Imaginary

Magnitude = $\sqrt{a^2 + b^2}$
 phase = $\tan^{-1} \left(\frac{\sin \omega_0 t}{\cos \omega_0 t} \right)$

Magnitude = $\sqrt{a^2 + b^2}$
 phase angle = $\tan^{-1} (b/a)$

$$x(t) = e^{j\omega_0 t} + e^{j\omega_1 t} + e^{j\omega_2 t} + e^{j\omega_3 t} = \cos \omega_0 t + j \sin \omega_0 t$$

T_1 T_2 T_3 LCM (T_1, T_2, T_3)

Rectangular type

$$\begin{aligned} a &= \sqrt{a^2 + b^2} \cdot \cos \theta \\ \tan^{-1} (b/a) &= \theta \end{aligned}$$

$$r = \sqrt{3^2 + 4^2}$$

$$\begin{aligned} Q &= \tan^{-1} (b/a) \\ P &= \tan^{-1} 4/3 \end{aligned}$$

$$\begin{aligned} \theta &= 90^\circ \\ \theta_2 &= 180^\circ \\ \theta_3 &= 270^\circ \end{aligned}$$

$$j \cdot j = j^2 = -1$$

$$90^\circ + 90^\circ = 180^\circ$$

Similarly $\sqrt{j^2} = 90^\circ$

$$x = 5 \tan^{-1} (4/3)$$

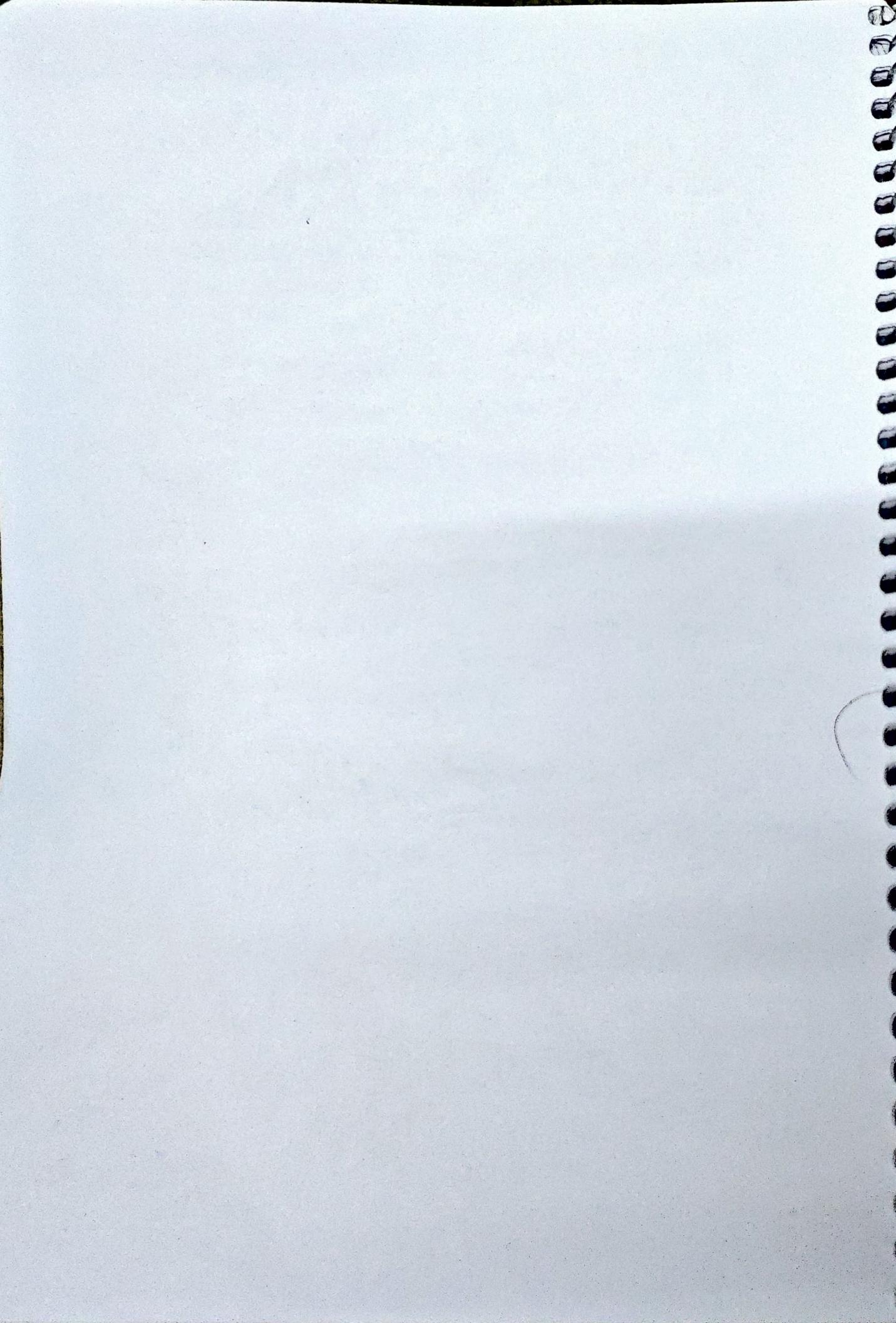
$$\begin{aligned} \sqrt{x} &= \sqrt{5} \cdot \sqrt{\tan^{-1} (4/3)} \\ &= \sqrt{5} \cdot \sqrt{\frac{\tan^{-1} (4/3)}{2}} \end{aligned}$$

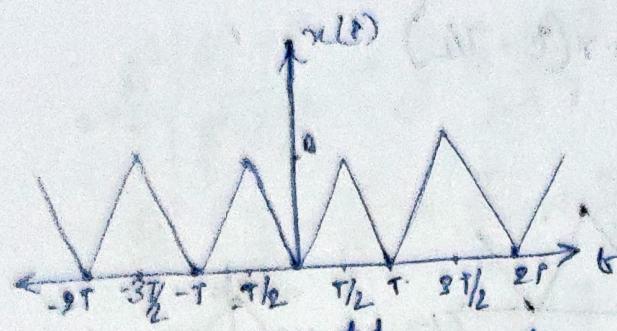
$$x = \frac{(4+5j)(3+6j)}{(8+4j)(9+3j)}$$

$$\therefore \frac{\sqrt{41} L \tan^{-1}(5/4) \cdot \sqrt{45} L \tan^{-1}(2)}{\sqrt{80} L \tan^{-1}(4/8) \cdot \sqrt{90} L \tan^{-1}(1/3)}$$

$$\therefore \frac{\sqrt{41} \cdot \sqrt{45}}{\sqrt{80} \cdot \sqrt{90}} \left(\frac{L \tan^{-1}(5/4) + L \tan^{-1}(2)}{L \tan^{-1}(1/2) + L \tan^{-1}(1/3)} \right) \text{ degree}$$

$$\therefore \frac{42.953}{84.852}$$





- * It must be bounded
- * If a signal is periodic then it is power signal.

$$P_x = \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt$$

$$y = mx + c$$

m = slope
c = intercept.

$$y_{ab} = \frac{2A \cdot t}{T} + 0 \text{ for } -T/2 \leq t < 0$$

$$y_{ab} = \frac{2A \cdot t}{T} + 0$$

$$y_{bc} = \frac{2A \cdot t}{T} + 0 \text{ for } 0 \leq t < T/2$$

$$y_{bc} = \frac{2A \cdot t}{T} + 0$$

$$P_x = \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt$$

$$= \frac{1}{T} \left[\int_0^{T/2} |x(t)|^2 dt + \int_{-T/2}^0 |x(t)|^2 dt \right]$$

$$= \frac{1}{T} \left[\int_0^{T/2} \left(\frac{2A \cdot t}{T} \right)^2 dt + \int_{-T/2}^0 \left(\frac{2A \cdot t}{T} \right)^2 dt \right]$$

$$= \frac{1}{T} \left[\frac{4A^2}{T^2} \left[\frac{t^3}{3} \right]_0^{T/2} + \frac{4A^2}{T^2} \left[\frac{t^3}{3} \right]_0^{-T/2} \right]$$

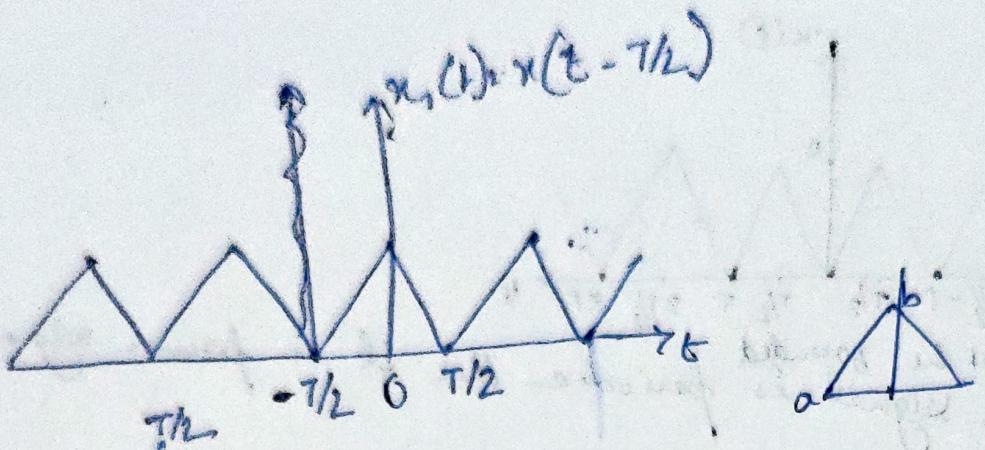
$$= \frac{1}{T} \cdot \frac{4A^2}{T^2} \left[\frac{T^3}{24} + \frac{T^3}{24} \right]$$

$$= \frac{4A^2}{T^2} \cdot \frac{2T^3}{24} = \frac{A^2}{3}$$

\approx power of signal

$$E_x = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

$$\therefore P_x \times \text{time} = \infty$$



$$P_x = \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt$$

$$= \frac{1}{T} \int_{-T/2}^{T/2} |x(t-T/2)|^2 dt$$

$$= \frac{1}{T} \left[\int_0^{T/2} |x(t-T/2)|^2 dt + \int_{-T/2}^0 |x(t-T/2)|^2 dt \right]$$

$$y_{ab} = \frac{2A}{T}$$

$$m_{ab} = \frac{2A}{T}$$

$$y_{bc} = \frac{2A}{T} \cdot b + A$$

$$y_{bc} = \frac{2A}{T} \cdot b + A$$

$$= \int_0^{T/2} \left(\frac{2A}{T} \cdot t + A \right)^2 dt$$

$$= \left[\frac{4A^2 \cdot t^2}{T^2} + A^2 + 2 \cdot \frac{2At}{T} \cdot A \right]_0^{T/2}$$

$$= \frac{4A^2 \cdot t^2}{T^2} + A^2 + \frac{4A^2 t}{T}$$

$$= \frac{1}{T} \cdot A^2 \left\{ \left[\frac{4t^2}{T^2} + 1 + \frac{4t}{T} \right] dt + \left[\frac{4t^2}{T^2} + 1 - \frac{4t}{T} \right] dt \right\}$$

$$= \frac{1}{T} \cdot A^2 \left\{ \left[\frac{4t^3}{3T^2} \right]_0^{T/2} + [6]_0^{T/2} + \frac{4[6t^2]}{T^2} \right\}$$

$$= \left[\frac{4t^3}{T^2 \cdot 3} \right]_0^{T/2} + [6]_0^{T/2} - \frac{4[6t^2]}{T^2} \Big|_{-T/2}^{T/2}$$

$$2 \frac{A^2}{T} \left\{ \frac{\frac{4}{3} T^3}{8 \cdot 3!} + \frac{T}{2} + \frac{4 T^2}{2 \cdot 4!} + \frac{4 T^3}{3! 2! 8!} + \frac{T}{2} - \frac{4 T^2}{8!} \right\}$$

$$\cdot \frac{A^2}{T} \left[\frac{T}{6} + \frac{T}{2} + \frac{T}{6} + \frac{T}{2} \right]$$

$$\cdot \frac{A^2}{T} \left[\frac{T + 3T + T + 3T}{6} \right] \cdot \frac{28T}{(8!)^2} \cdot \frac{2A^2}{3}$$

$$\frac{1908}{1152} = \frac{312}{1152} + \frac{31}{1152}$$

Q. $e^{2t} \cdot u(t)$

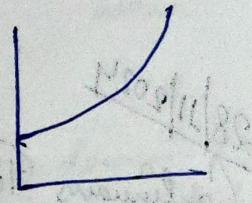
$$P_x = \frac{1}{T} \int_{-\pi/2}^{\pi/2} [x(t)]^2 dt$$

$$= |e^{At} \cdot u(t)|^2 dt$$

$$= e^{2At} \cdot u(t)^2 dt$$

m^2

$\int u^2 dt$
$E \rightarrow \infty$
$P = 0$



$$[(A)_x] P_{2,0} + [(A)_y] P_{1,0} = [(A)_{xy}] P_{1,0} + [(A)_{yx}] P_{2,0}$$

$$(A_{xy}) P_{1,0} + (A_{yx}) P_{2,0}$$

$(A)_{xy}$



$$\frac{18}{T} = 48$$

$$x(t) = 4 + 4\sin 500t + 8\cos 300nt$$

rms value of 4 = 4

$$4\sin 500t = 4^2/2$$

$$8\cos 300nt = 8^2/2$$

$$R_s = 4^2 + \left(\frac{4}{2}\right)^2 + \left(\frac{8}{2}\right)^2$$

$$= 16 + \frac{16}{4} + \frac{64}{4}$$

$$\frac{64+16+64}{4}$$

$$\begin{aligned} & 64 \\ & 16 \\ & 64 \\ & \hline 286 \\ & 324 \\ & \hline 4096 \end{aligned}$$

28/11/2024

Continuous Time Fourier Transform

Technique to convert a signal to frequency domain.

Limitation:

A periodic signal can be represented in frequency domain.

Linear combination of Harmonically related complex exponential

Linear combination

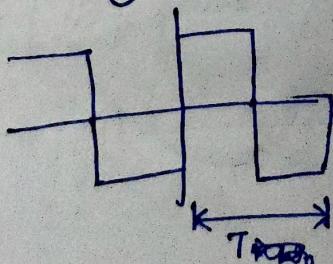
$$F[a_1x_1(t) + a_2x_2(t)] = a_1F[x_1(t)] + a_2F[x_2(t)]$$

$$= a_1y_1(t) + a_2y_2(t)$$

$$\Rightarrow [ay(t)]$$

Harmonically related

Considering periodic but not sinusoidal.



$$\Rightarrow w_0 = \frac{2\pi}{T}$$

If a sinusoidal signal which is represented with some frequency (ω_0) that signal will be called first harmonic of the given signal.

When it is $T/2$ it is second harmonic of the given signal i.e. $2\omega_0$.

Similarly, third harmonic is $3\omega_0$.
Harmonically related

$$sg(x) = \sin(\omega_0 t) + \sin(2\omega_0 t) + \sin(3\omega_0 t)$$

Complex exponential

$$e^{j\omega_0 t}, e^{j2\omega_0 t}, e^{j3\omega_0 t}$$

This are all harmonically related

$$e^{j\omega_0 t} = \text{sinusoidal wave}$$

Synthesis

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

$a_k \Rightarrow 1^{\text{st}} \text{ harmonic}$

$2^{\text{nd}} \text{ harmonic}$

$a_k \Rightarrow$ Fourier series coefficient

Analysis

$$a_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt$$

Synthesis

When periodic sawtooth signal is given

$$x(t) = A \cos \omega_0 t$$

$$x(t) = A \sin \omega_0 t$$

Step 1: Conversion to complex exponential

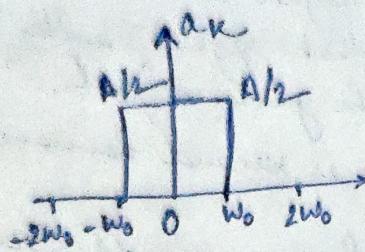
$$A \cdot (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

$$\frac{A e^{j\omega_0 t}}{2} + \frac{A e^{-j\omega_0 t}}{2}$$

$$a_1 = \frac{A}{2}, a_{-1} = \frac{A}{2} \leftarrow \text{coefficient of wave}$$

frequently domain then can be represented in form of phase & magnitude

As we need to calculate magnitude and phase

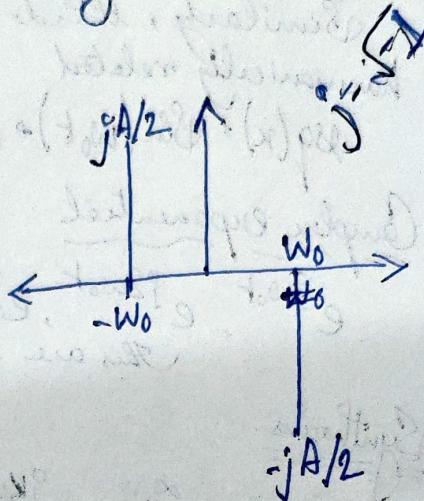


This is not the representation of magnitude & phase.
So we need to represent it in magnitude and phase.

$$\sin w_0 t = A \left(e^{jw_0 t} - e^{-jw_0 t} \right) / 2j$$

$$a_1 = \frac{A}{2j}, \quad a_2 = \frac{A}{2j}$$

$$= -\frac{jA}{2}, \quad = \frac{jA}{2}$$



$$x(t) = 4 + 2\cos \frac{2\pi}{3}t + 4\sin \frac{5\pi}{3}t$$

Calculating fundamental periodicity

$$w_0 = \frac{2\pi}{3}, \quad w_{0_2} = \frac{5\pi}{3}$$

$$T_1 = 3, \quad T_2 = 6/5$$

$$T = \text{LCM} \cdot \left[\frac{3}{1}, \frac{6}{5} \right]$$

$$\frac{6}{1} \text{ sec.}$$

\uparrow fundamental time

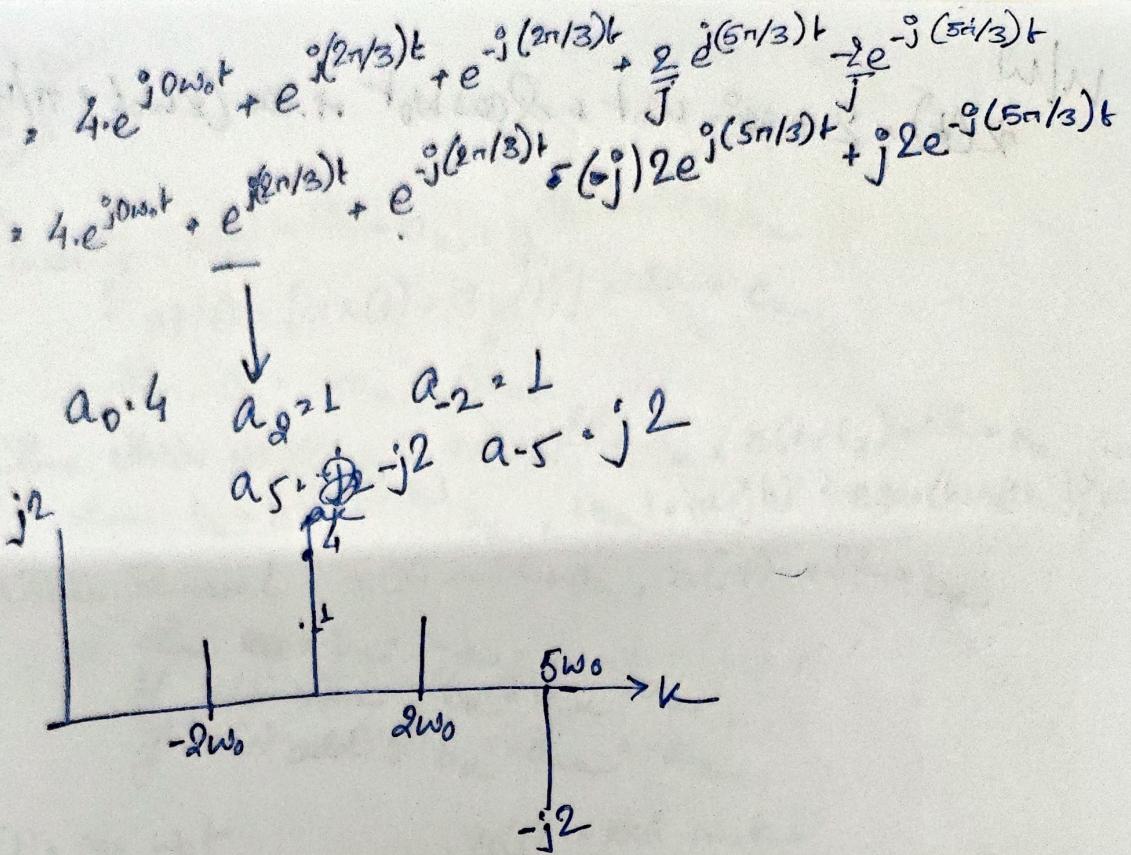
No. $\frac{6}{1} \cdot \frac{\pi}{3} = \text{fundamental period.}$

This means $x(t)$ will repeat after $T/3$

\uparrow 1st harmonic frequency

Conversion to exponential form.

$$4e^{j0w_0 t} + \frac{1}{2} \left(e^{j2\pi/3 t} + e^{-j2\pi/3 t} \right) + 4 \left(\frac{e^{j5\pi/3 t} - e^{-j5\pi/3 t}}{2} \right)$$

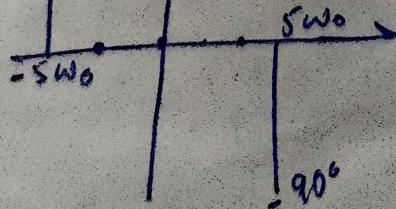


This is four series calculating magnitude & phase response.

	Values	$ a_k \leftarrow$ magnitude	$\angle a_k \leftarrow$ phase angle
a_0	1	1	0°
a_2	1	1	0°
a_{-2}	1	1	0°
a_5	$-j2$	2	-90°
a_{-5}	$+j2$	2	90°

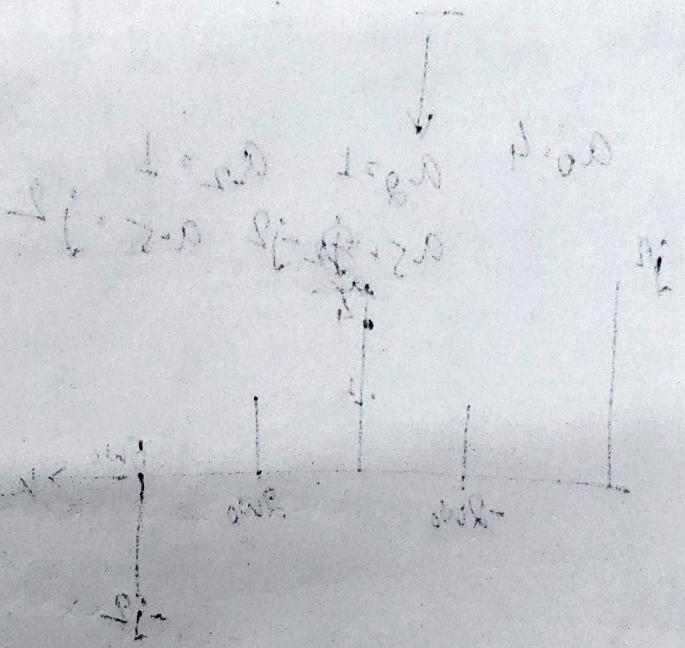
No complex term so phase angle is 0°

Draw the graph magnitude & phase angle vs values.

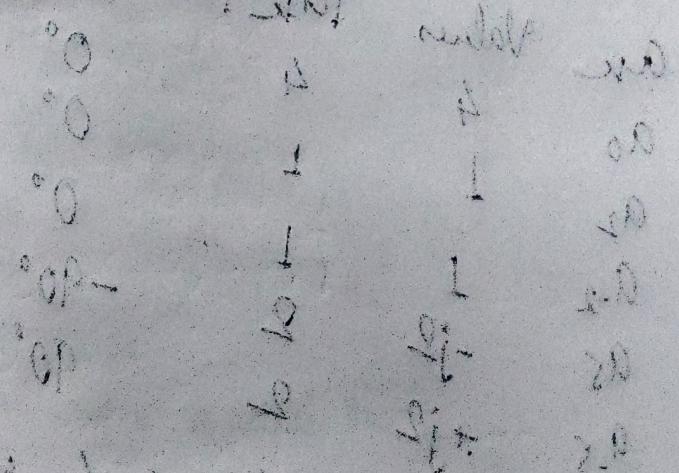


$$u/w$$

$$x(t) = 2 + \sin \omega_0 t + 2 \cos \omega_0 t + \cos(2\omega_0 t + \pi/4)$$



graph shows wave sum is all
sum of regular sum of individual patches



graph shows wave sum is all sum of patches

Properties
of Fourier Series
Coefficient

1) Linearity : $x(t) \xleftrightarrow{FS} a_k, y(t) \xleftrightarrow{FS} b_k$
 $z(t) = [\alpha x(t) + \beta y(t)] \xleftrightarrow{FS} c_k$

then, $c_k = \alpha a_k + \beta b_k$

2) Time Shifting : $x(t) \xleftrightarrow{FS} a_k, x(t \pm t_0) \xleftrightarrow{FS} b_k$ (for $t_0 \neq 0$)
 then $b_k = e^{\pm jk\omega_0 t_0} \cdot a_k, |b_k| = (a_k)^2 e^{\pm jk\omega_0 t_0}$

3) Time Reversal : $x(t) \xleftrightarrow{FS} a_k, x(-t) \xleftrightarrow{FS} b_k$

then $b_k = a_{-k}$

if $x(t)$ even : $b_k = a_{-k} = a_k$

if $x(t)$ odd : $b_k = a_{-k} = -a_k$

Suppose

Q1 $x(t) = \cos \omega_0 t$

$$= \frac{a_k}{2} e^{j\omega_0 t} + \frac{a_k}{2} e^{-j\omega_0 t}$$

$a_1 = \frac{1}{2}$

$a_{-1} = \frac{1}{2}$

$$y(t) = \sin \omega_0 t$$

$$= \frac{b_k}{2} e^{j\omega_0 t} - \frac{b_k}{2} e^{-j\omega_0 t}$$

$$\alpha = 2$$

$$\beta = 2$$

$$c_{k=1}(t) = \alpha \left(\frac{1}{2}\right) + \beta \left(-\frac{1}{2}\right) \quad c_{k=-1} = \frac{1}{2} + \frac{j}{2}$$

$$c_{k+1}(t) = \frac{\frac{1}{2} + \frac{-j}{2}}{2} \quad c_{-k} = \frac{\frac{1}{2} + \frac{j}{2}}{2}$$

Q2 $x(t + t_0) \rightarrow x(t - t_0) \xleftrightarrow{FS} b_k$

if $x(t + t_0) \leftrightarrow e^{jk\omega_0 t_0} \cdot a_k$

$x(t - t_0) \leftrightarrow e^{-jk\omega_0 t_0} \cdot a_k$

$e^{jk\omega_0 t_0} \cdot a_k + e^{-jk\omega_0 t_0} \cdot a_k$

$$b_k = \frac{2e^{jk\omega_0 t_0} \cdot a_k (e^{-jk\omega_0 t_0} + e^{jk\omega_0 t_0})}{2}$$

Multi & divide by 2

$b_k = 2a_k \cos(k\omega_0 t_0)$

$$|b_k| = \sqrt{[\cos(k\omega_0 t_0)]^2 + [j \sin(k\omega_0 t_0)]^2}$$

$$b_n = e^{j\omega_0(\pm t_0)} + a_n$$

- $\cos(\omega_0(\pm t_0)) + j \sin(\omega_0(\pm t_0))$

- $\tan^{-1} \frac{\sin \omega_0(\pm t_0)}{\cos \omega_0(\pm t_0)} + a_n$

- $a_n + \omega_0(\pm t_0)$

Q $x(1-t) \leftrightarrow b_n \leftrightarrow x(t-1)$

$$\underline{x(1-t)} \quad x(t) = a_n$$

$$x(-t) = \underline{b_n} \quad b_n = a_{-n}$$

$$x(-t+1) = \underline{e^{j\omega_0}} \cdot a_{-n}$$

$$x(t) = a_n$$

$$x(t-1) = \underline{e^{j\omega_0}} \cdot a_n$$

$$(e^{j\omega_0} \cdot a_{-n} + e^{-j\omega_0} \cdot a_n)$$

$$\cancel{e^{j\omega_0}} \quad e^{-j\omega_0} (a_{-n} + a_n)$$

$$\text{even: } b_n = e^{-j\omega_0} [a_n + a_{-n}] \\ = 2a_n e^{-j\omega_0}$$

$$\text{odd: } b_n = e^{-j\omega_0} [a_n - a_{-n}] \\ = 0$$

Purely even signal.

Q Time scaling

$$\begin{aligned} x(t) &\leftrightarrow a_n \\ x(Tt) &\leftrightarrow b_n \\ b_n &\leftrightarrow a_n \end{aligned}$$

Coefficient of $\cos 2\omega_0 t$

$$\frac{e^{j2\omega_0 t} + e^{-j2\omega_0 t}}{2}$$

$$a_2 = \frac{1}{2} + \frac{a_2 - \frac{1}{2}}{2}$$

for $2\omega_0 T$

Now time scaling -

~~$$a_2 = a_1 + \frac{1}{2} + a_{-1} \left(\frac{z-1}{2} \right)$$~~

Coefficient same but value of k changes.

If can change frequency, changing coefficient position

10/12/2024
Analysis & Equation

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

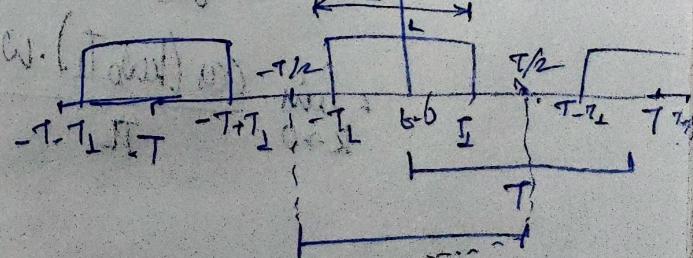
For synthesis of Fourier series - Periodic sinusoidal

If not sinusoidal then use analysis equation -

$$\therefore a_k = \frac{1}{T} \int_{-T_L}^T x(t) e^{-jk\omega_0 t} dt$$

$$= \frac{1}{T} \int_{-T_L}^{T_2} e^{-jk\omega_0 t} dt$$

$$= -\frac{1}{T} \left[\frac{e^{-jk\omega_0 t}}{-jk\omega_0} \right]_{-T_L}^{T_2}$$



$$\boxed{x(t) = L}$$

$$\boxed{\frac{e^{-jk\omega_0 t}}{-jk\omega_0} \Big|_{-T_L}^{T_2}}$$

$$= -\frac{1}{T} \left[\frac{e^{jk\omega_0 T_1}}{jk\omega_0} - \frac{e^{jk\omega_0 T_2}}{jk\omega_0} \right]$$

$$+ \frac{1}{T} \left[\frac{e^{jk\omega_0 T_2}}{2jk\omega_0} - e^{-jk\omega_0 T_2} \right]$$

$$\frac{2}{T\omega_0} \left[\frac{e^{j\omega_0 T_1} - e^{-j\omega_0 T_2}}{2} \right]$$

$$\frac{2}{T\omega_0} \sin(\omega_0 T_1)$$

$$\frac{2}{T\omega_0} \sin\left(\omega_0 \frac{2\pi}{T} T_1\right)$$

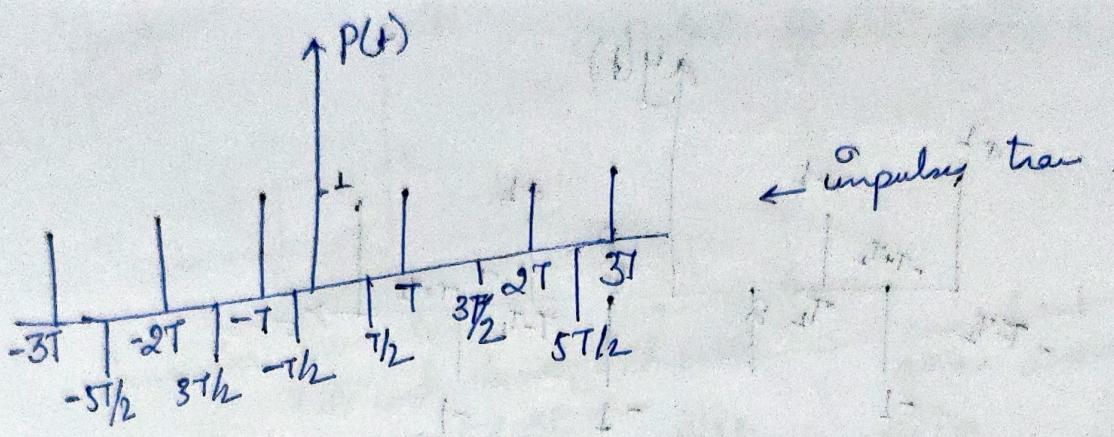
$$\frac{1}{k\pi} \sin\left(\frac{2\pi k T_1}{T}\right) = \boxed{\frac{1}{k\pi} \sin(\omega_0 k T_1)}$$

Pulse width
* $2T_1$
T is time period.

if $k=0$, use L'Hospital Rule
w.r.t k

$$a_0 = \lim_{k \rightarrow 0} \frac{\frac{d}{dk} \sin(k\omega_0 T_1)}{\frac{d}{dk} \cdot \pi k}$$

$$\lim_{k \rightarrow 0} \frac{\cos(k\omega_0 T_1) \cdot \omega_0 T_1}{\pi} \cdot \frac{\omega_0 T_1}{\pi}$$



$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) \cdot e^{-j k \omega_0 t} dt$$

$$= \frac{1}{T} \int_{-T/2}^{T/2} S(t) \cdot e^{-j k \omega_0 t} dt$$

$$= \frac{1}{T} \left[\frac{e^{-j k \omega_0 t}}{-j k \omega_0} \right]_{-T/2}^{T/2}$$

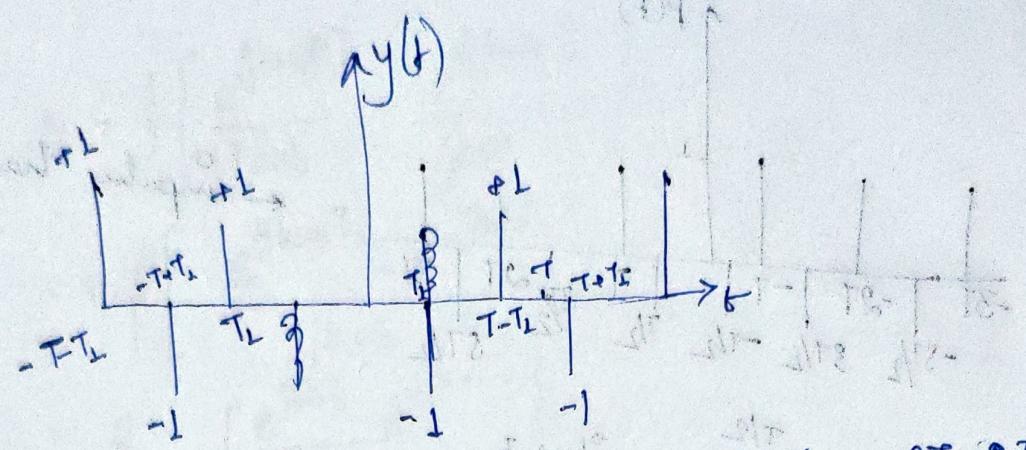
$$= \frac{1}{T} \left[\frac{e^{-j k \omega_0 T/2}}{-j k \omega_0} - \frac{e^{+j k \omega_0 T/2}}{-j k \omega_0} \right]$$

$$= \frac{1}{T} \quad (\text{independent of } k)$$

$$P(t) = \sum S(t - kT)$$

$$= S_t \quad \text{when } k=0$$

$$= S_t$$



$$\frac{P(t-T_L)}{P(t+T_L)} \xrightarrow{\text{a}} -2T_L + T_L \quad -T + T_L \quad T_L \quad T + T_L \quad 2T + 2T_L$$

$$\frac{P(t+T_L)}{P(t-T_L)} \xrightarrow{\text{a}} -2T - T_L \quad -T - T_L \quad -T_L \quad T - T_L \quad 2T - T_L$$

$$P(t+T_L) - P(t-T_L) \rightarrow y(t)$$

$$P(t) \xleftrightarrow{\text{F.S.C.}} \frac{1}{T}$$

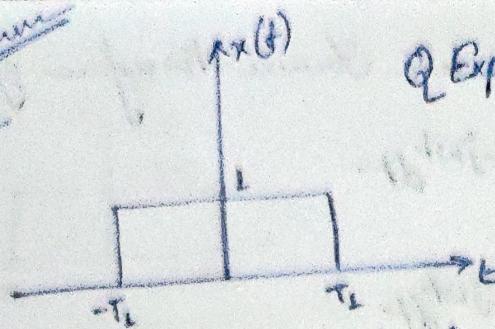
$$P(t+T_L) \xleftrightarrow{\text{F.S.C.}} e^{jk\omega_0 T_L} \frac{1}{T}$$

$$P(t-T_L) \xleftrightarrow{\text{F.I.C.}} e^{-jk\omega_0 T_L} \frac{1}{T}$$

$$\Rightarrow \frac{2j}{T} \left(\frac{e^{jk\omega_0 T_L}}{2T} - \frac{-e^{-jk\omega_0 T_L}}{2T} \right)$$

$$\Rightarrow \frac{2j}{T} \sin(k\omega_0 T_L)$$

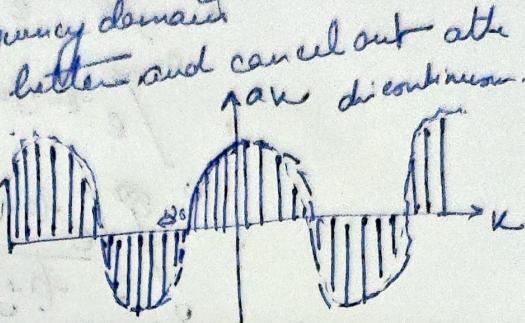
Priyadarshini
17/12/2021



Q Explain by FSC from RT

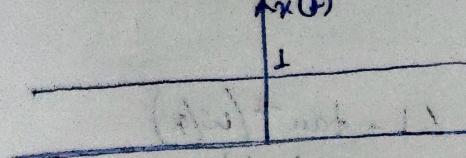
Conversion from time to frequency domain
to understand it better and cancel out other unnecessary noise

$$\text{Soln: } a_k = \frac{\sin k\omega_0 T_1}{\pi k}$$

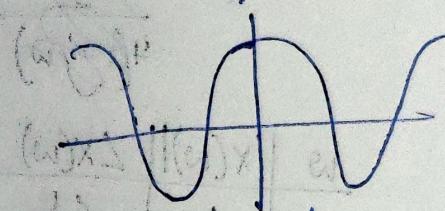


must $\omega_0 \rightarrow 0$ then gap = 0

$$\hookrightarrow T = \frac{2\pi}{\omega_0} \rightarrow \infty$$



\Downarrow Fourier Transform



Analytic Fourier

Consider $T \rightarrow \infty$ we calculate Fourier Transform.

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-j\omega_0 t} dt$$

$$a_k = \frac{1}{T} \int_{-\infty}^{\infty} x(t) e^{-j\omega_0 t} dt$$

$$\lim_{T \rightarrow \infty} (T a_k) = \int_{-\infty}^{\infty} x(t) e^{-j\omega_0 t} dt \rightarrow \text{Fourier transform.}$$

$$\boxed{x(\omega) = \lim_{T \rightarrow \infty} (T \cdot a_k) = \int_{-\infty}^{\infty} x(t) e^{-j\omega_0 t} dt}$$

use this for any coefficient

$$\text{Fourier Transform} \rightarrow x(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega_0 t} dt \rightarrow x(0) = \int_{-\infty}^{\infty} x(t) dt \text{ gives area in time domain}$$

$$\text{Inverse Fourier Transform} \rightarrow x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) e^{j\omega_0 t} d\omega \rightarrow x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) d\omega \text{ gives area in frequency domain}$$

$x(t) = e^{-at} \cdot u(t)$. Calculate Fourier transform of the signal

$$\hat{x}(w) = \int_{-\infty}^{\infty} x(t) \cdot e^{-jwt} dt$$

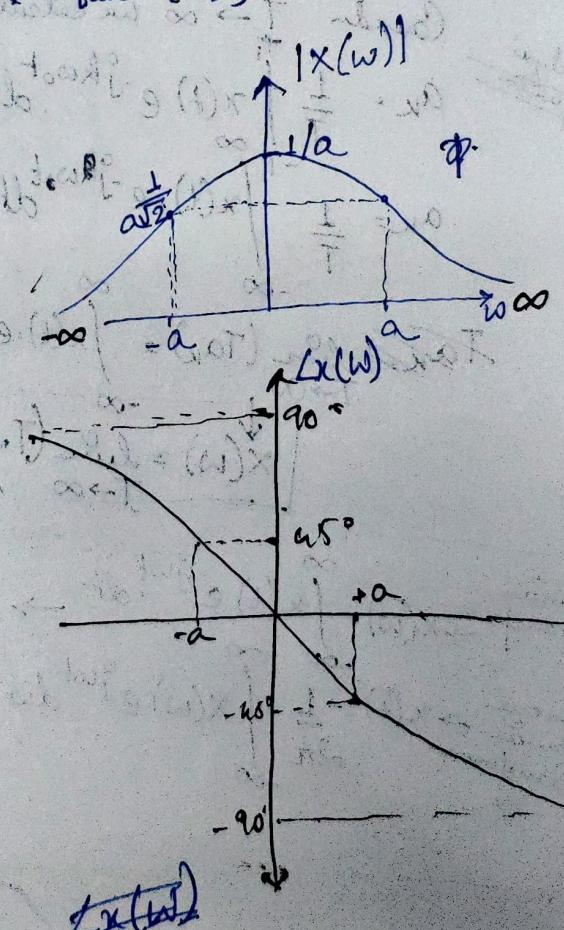
$$= \int_{-\infty}^{\infty} e^{-at} \cdot e^{-jwt} dt$$

$$= \int_0^{\infty} e^{-(a+jw)t} dt$$

$$= \left[\frac{e^{-(a+jw)t}}{-a-jw} \right]_0^{\infty}$$

$$\hat{x}(jw) = \frac{1}{a+jw}$$

w	$ x(w) $	$\angle x(w)$
0	$\frac{1}{\sqrt{a^2+w^2}}$	0°
$-a$	$\frac{1}{a}$	45°
$+a$	$\frac{1}{a\sqrt{2}}$	-45°
$-\infty$	0	$+90^\circ$
$+\infty$	0	-90°



Ex(1)

$$\underline{\text{H/W}} \quad Q_1 x(t) = e^{-\alpha |t|}$$

