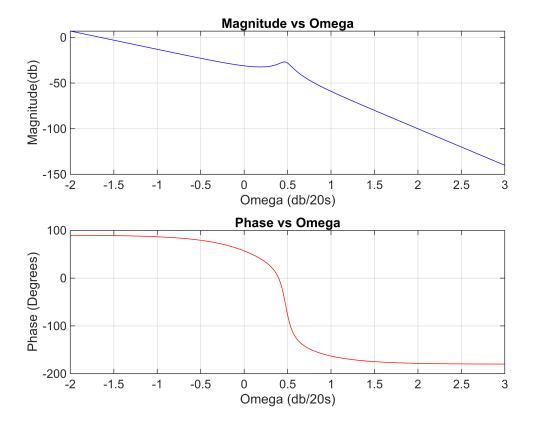
1.Bode extraction

1.1.Load the data

```
load('Data.mat');
omega = log10(Data.omega);
magnitude = 20.*log10(Data.magnitude);
phase = Data.phase;
```

1.2.Plot

```
figure;
subplot(2,1,1);
plot(omega, magnitude, 'b');  % Plot Magnitude vs Omega
xlabel('Omega (db/20s)');
ylabel('Magnitude(db)');
title('Magnitude vs Omega');
grid on;
subplot(2,1,2);
plot(omega, phase, 'r');  % Plot Phase vs Omega
xlabel('Omega (db/20s)');
ylabel('Phase (Degrees)');
title('Phase vs Omega');
grid on;
```



2. System identification

2.1.System type

As the slope in the Magnitude part is -20dB/dec, we assume that we have a pole in the Origin so obviously the answer to step function should be "0" and the system is type 1.

2.2.System order

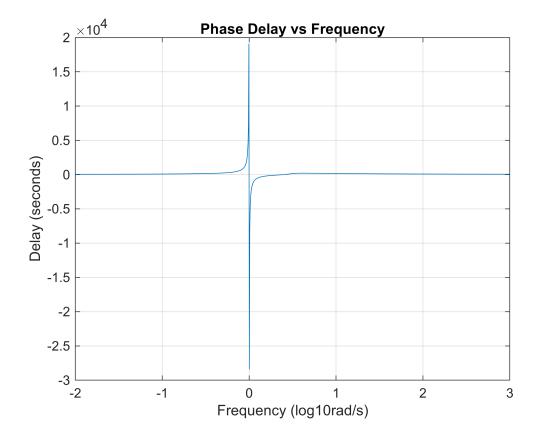
As explained in section 2.4 we have 3 poles so the system is from the third order.

2.3.System delay

To estimate the delay τ , you can calculate the slope of the phase curve at different frequencies.

```
delay = -phase ./ omega; % Phase delay at each frequency (in seconds)

figure;
plot(omega,delay) % Plot delay vs frequency (log scale for omega)
xlabel('Frequency (log10rad/s)');
ylabel('Delay (seconds)');
grid on;
title('Phase Delay vs Frequency');
```



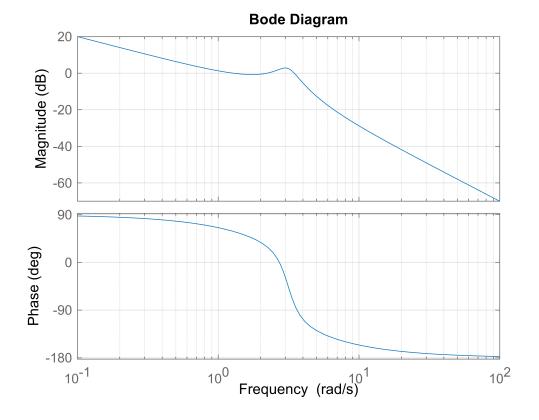
2.4. Minimum phase

As it can be seen in the Bode diagram, we start the magnitude with a -20dB/dec slope and a phase of 90 which shows that we have a "-s" and a second order func pole with positive middle coefficient and a zero at the same frequency, So it has a minimum phase zero.

3. Transfer function

The estimated transfer function is gonna be like the following

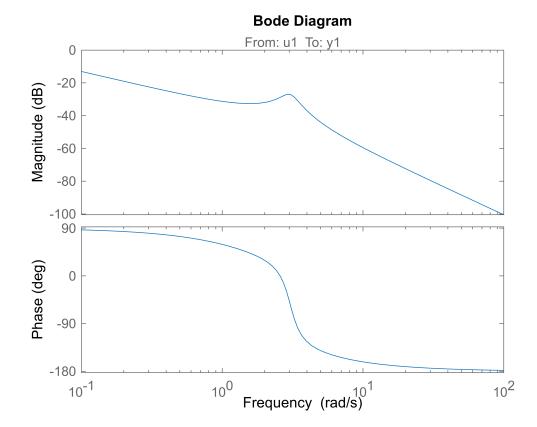
```
num = [1/pi -1];
den = conv([1/(pi^2) 1/(3*pi) 1], [1 0]);
sys = tf(num,den);
bode(sys);
grid on;
```



which actually looks the same as the upper expresion.

3.1.Optional first

```
systemIdentification;
bode(tf1);
```



```
tf1
```

4. Ruth hurwitz

```
syms s;

% Define the characteristic equation
char_eq = (1/(pi^2)) * s^3 + (1/(3*pi)) * s^2 + 1 * s;
```

```
% Get the coefficients as a numeric vector
coeffs = double(sym2poly(char_eq));  % Ensure numeric values

% Compute the Routh array
routh_table = routh_hurwitz(coeffs);

% Display Routh-Hurwitz Table in Live Script
disp('Routh-Hurwitz Table:');
```

Routh-Hurwitz Table:

```
disp(routh_table);

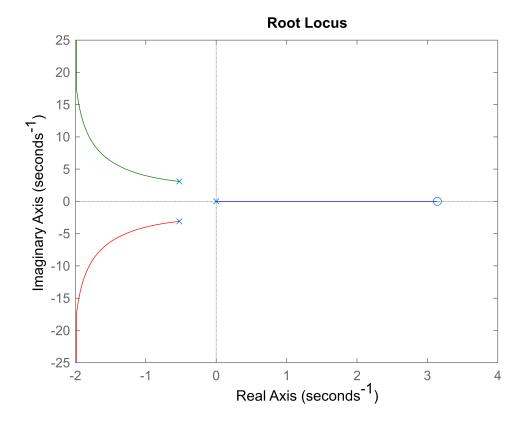
0.1013   1.0000
0.1061   0
1.0000   0
```

I tried to do the computing completly on matlab but I couldnt figure it out, the computing goes on by writing the characteristic equation considering K and being close loop. It converts to $\frac{s^3}{\pi^2} + \frac{s^2}{3\pi} + (1+k)s - k\pi$ which translates to the ruths tables first column to be $\frac{1}{\pi^2} - \frac{1}{3\pi} - (1+4k) - k\pi$ which means that k is to be $-\frac{1}{4} < k < 0$

5. Ruth locus analysis

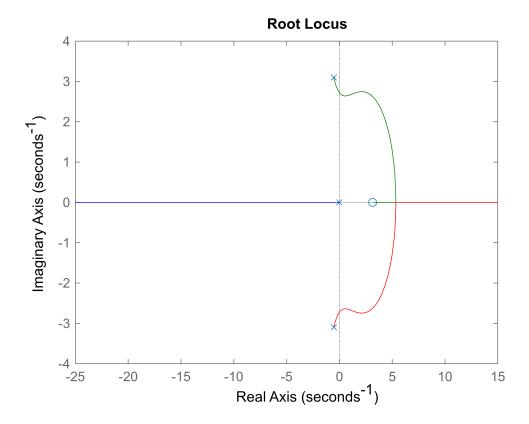
5.1.Ploting

```
rlocus(sys);
```



As seen, we have a complete branch in the ORHP which means that we can't manage stablizing the system using only the k coefficient.

rlocus(-sys);

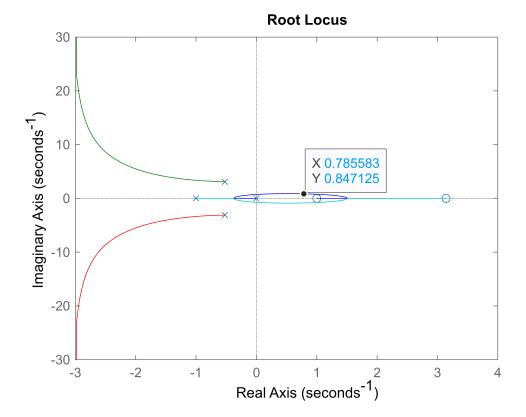


Working with the negative Ks, we have a really small gap in the branches to work with which actually matches what we saw in ruth's method.

5.2.Configure a PI or PD controller

We can add another minimum phase zero and another pole in the PLHP so that it leads to apearing a new branch moving from the OLHP to ORHP so that we find the chance to consider areas for K to bring all the poles in the OLHP so it shapes a PID.

```
num = conv([1/pi -1] , [1 -1]);
den = conv([1/(pi^2) 1/(3*pi) 1], [1 1 0]);
sys_rc = tf(num,den);
rlocus(sys_rc);
```



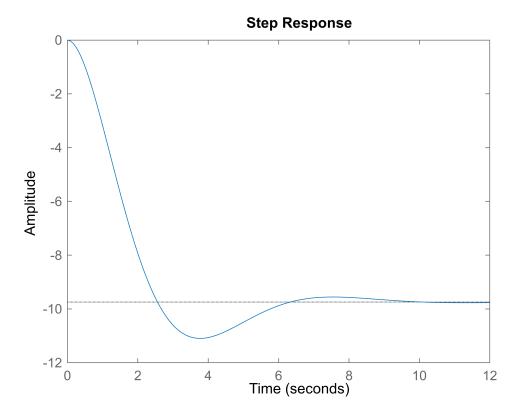
As you see now we have a little area in positive Ks to bring all the poles in the OLHP and make the system stable.

6.Control overshoot

Our closed loop transfer function will be $\frac{k\pi^2}{s^2 + \frac{\pi}{3}s + (k+1)\pi^2}$ which is a shape of $\frac{k\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}$ and as

 $M_p = e^{-\frac{\xi\pi}{\sqrt{1-\xi^2}}} \text{ it concludes that } 3.96 < \zeta < 5.8 \text{ and settling time with a } \delta = 2\% \text{ is equal to } \frac{4}{\xi\omega_n} \text{ so we conclude}$ that $\xi > \frac{4}{10\pi\sqrt{k+1}}$ and we can consider $-\frac{63}{64} > k > -\frac{3339}{3364}$ and we can find a k between these using an old rule :)))

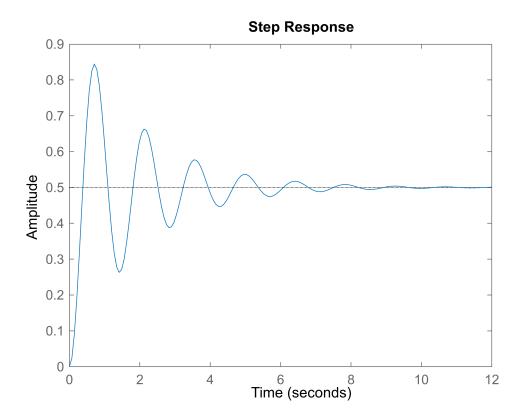
$$k = -\frac{63 + 3339}{64 + 3364}$$



stepinfo(sys_c)

```
ans = struct with fields:
    RiseTime: 1.7318
TransientTime: 5.8792
SettlingTime: 5.8792
SettlingMin: -11.0961
SettlingMax: -8.9610
    Overshoot: 13.8785
Undershoot: 0
    Peak: 11.0961
PeakTime: 3.7819
```

```
num = pi^2;
den = [1 pi/3 2*pi^2];
sys_uc = tf(num,den);
step(sys_uc);
```



stepinfo(sys_uc)

ans = struct with fields:

RiseTime: 0.2575
TransientTime: 7.2524
SettlingTime: 7.2524
SettlingMin: 0.2630
SettlingMax: 0.8443
Overshoot: 68.8609
Undershoot: 0

Peak: 0.8443 PeakTime: 0.7071

7. Controller design

7.1.Maximum of 2% for $e_{\rm sr}$

$$\frac{1}{s}S(s) = \frac{1}{s^2(1+L(s))} \text{ so } y(s) = \frac{1}{s^3(1+L(s))} \text{ and the } e_{sr} = \lim_{s \longrightarrow 0} \frac{1}{s\left(1+\frac{s-\pi}{s\left(\frac{s^2}{\pi^2}+\frac{s}{3\pi}+1\right)}\right)} = -\frac{1}{\pi} = \frac{1}{k_v} < \frac{1}{50} \text{ so we need a}$$

 $\frac{50}{\pi}$ gain to reach the wanted result.

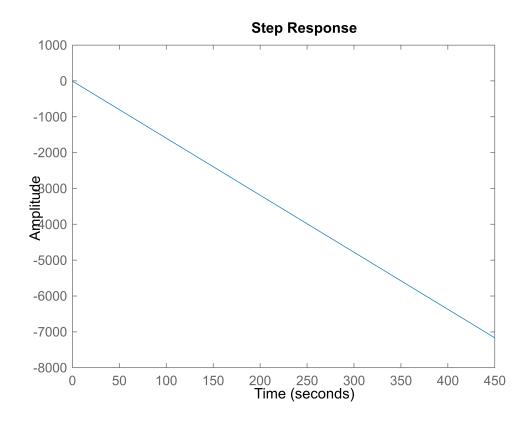
Also we add a zero at orgin.

```
K = (50/pi) -1;

T = (1/25)*(((K/0.05)^2)-1)^0.5;

num = conv([T 50/pi] , [1/pi -1 0]);

den = conv([1/(pi^2) 1/(3*pi) 1] , [1 0 0]); %having ramp response is equal to adding sys_fc = tf(num,den); %a pole at orgin then computing the step step(sys_fc); %response.
```



Thanks for your attention!

Iman Bidi 40116503 Linear Control Final project

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