
DSB-SC, SSB and VSB

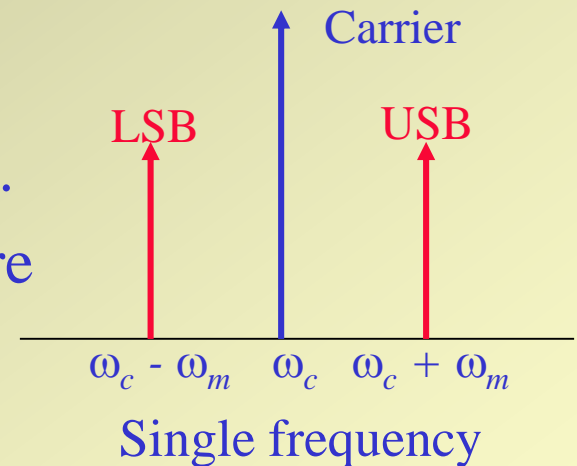
Contents

- Theory
- Implementation
 - Transmitter
 - Detector
 - Synchronous
 - Square
- Power analysis
- Summary

Double Side Band Suppressed Carrier

From AM spectrum:

- Carrier signal ω_c carries no information ω_m .
- Carrier signal consumes a lot of power more than 50%



Question: *Why transmit carrier at all?*

Ans:

Question: *Can one suppress the carrier?*

Ans.: *Yes, just transmit two side bands (i.e **DSB-SC**)*

But what is the penalty?

System complexity at the receiver

DSB-SC - *Theory*

General expression: $c(t) = [k_1 m(t) + C] \cos(\omega_c t + \phi_c)$

Let $k_1 = 1$, $C = 0$ and $\phi_c = 0$, the modulated carrier signal, therefore:

$$c(t) = m(t) \cos \omega_c t$$

Information signal $m(t) = E_m \cos \omega_m t$

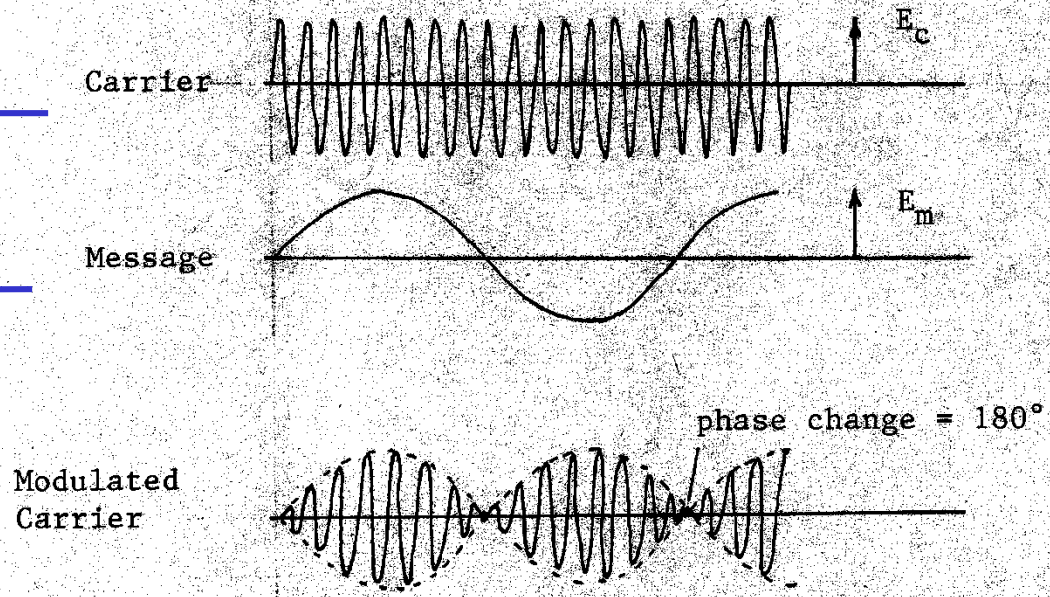
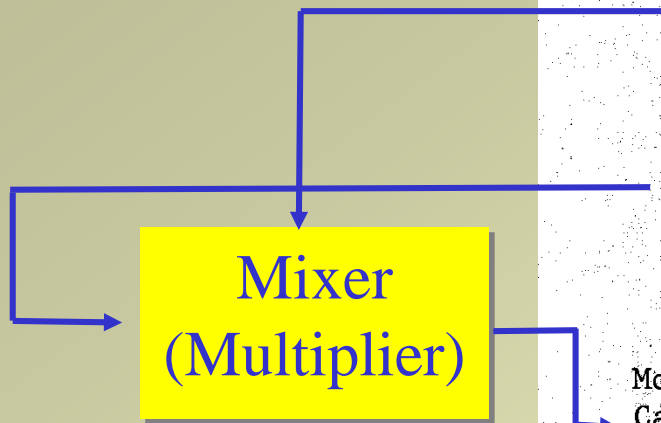
Thus

$$\begin{aligned} c(t) &= E_m \cos \omega_m t \cos \omega_c t \\ &= \frac{ME_c}{2} \cos(\omega_c + \omega_m)t + \frac{ME_c}{2} \cos(\omega_c - \omega_m)t \end{aligned}$$

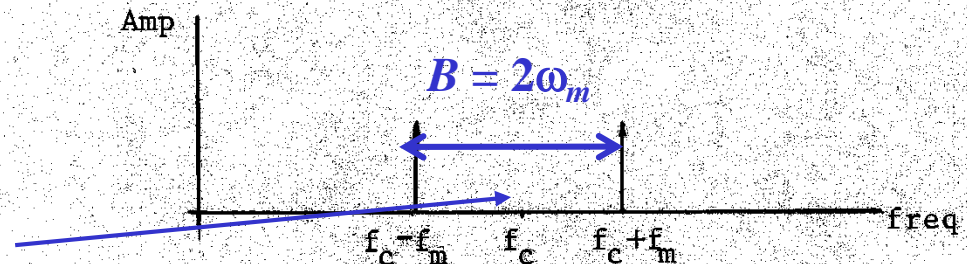
upper side band

lower side band

DSB-SC - Waveforms



a) DSB-SC waveforms



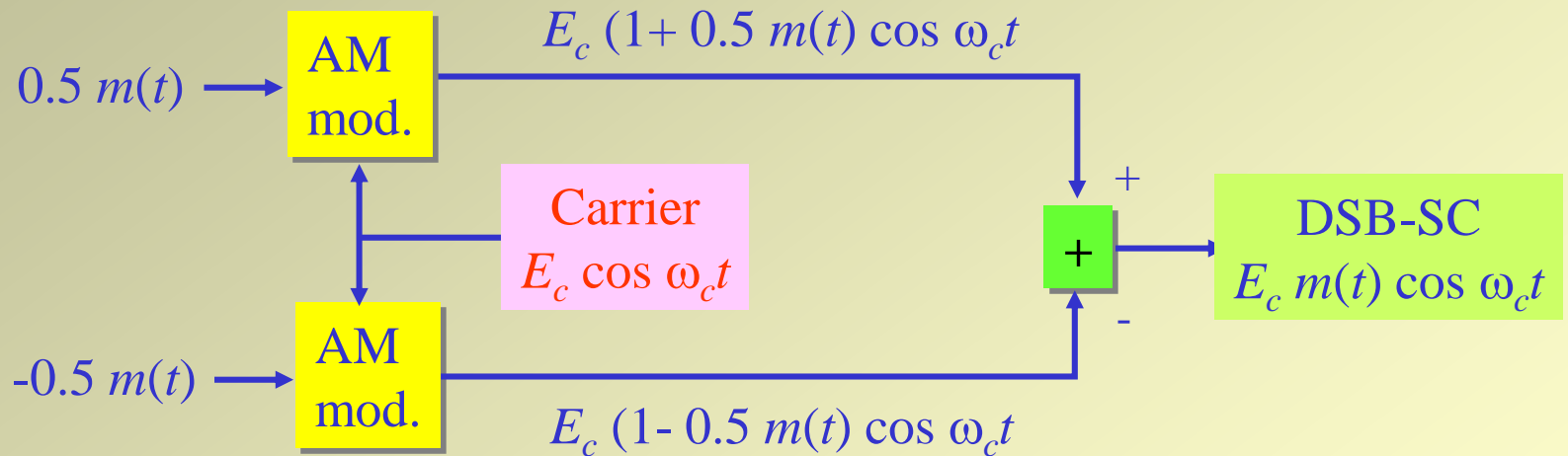
b) Frequency spectrum

Figure 2.6 DSB-SC amplitude modulation

Notice: No carrier frequency

DSB-SC - *Implementation*

- Balanced modulator



- Ring modulator
- Square-law modulator

DSB-SC - *Implementation*

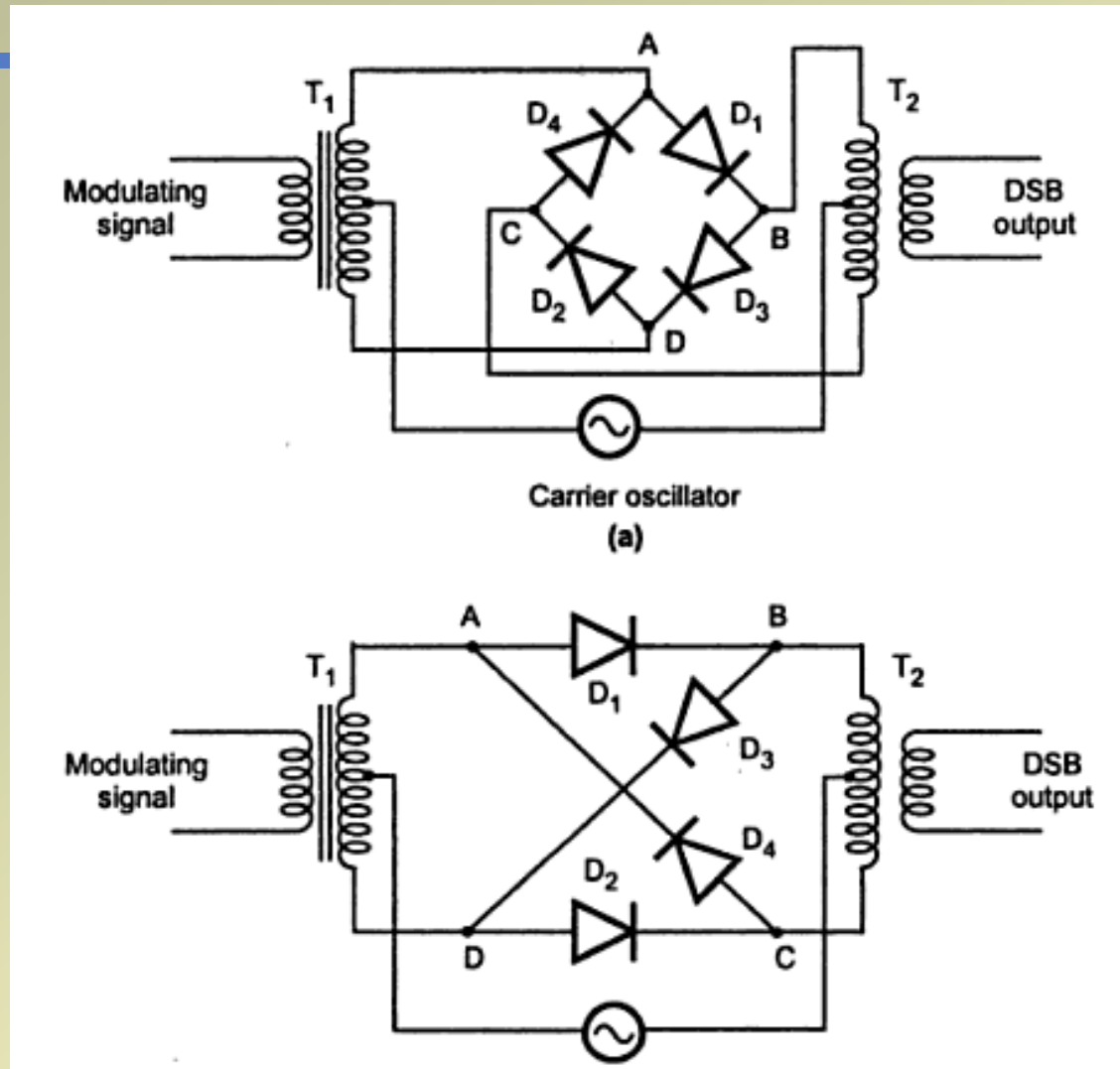
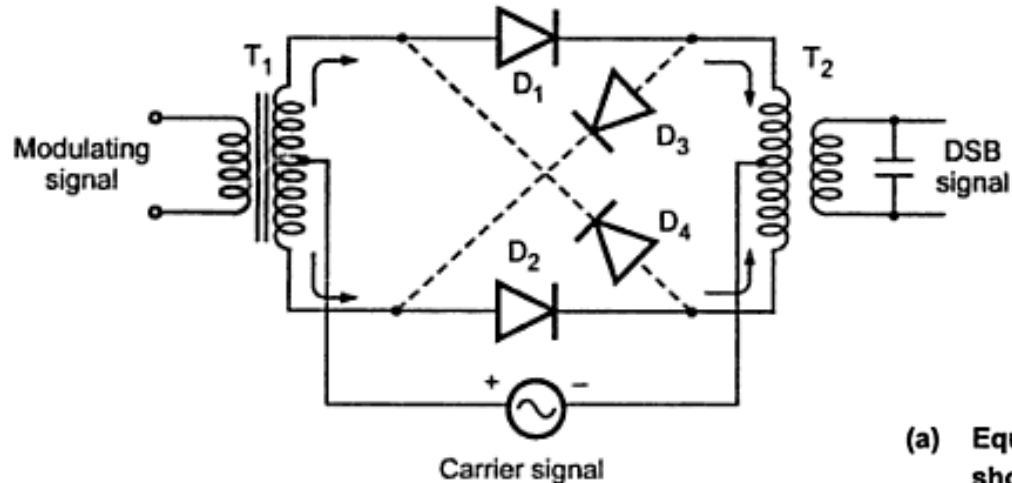


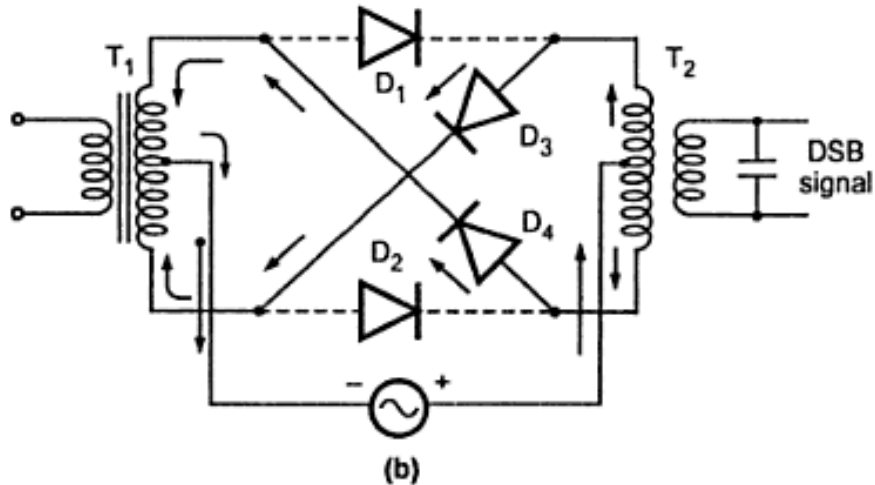
Fig: Balanced Ring Modulator

DSB-SC - *Implementation*



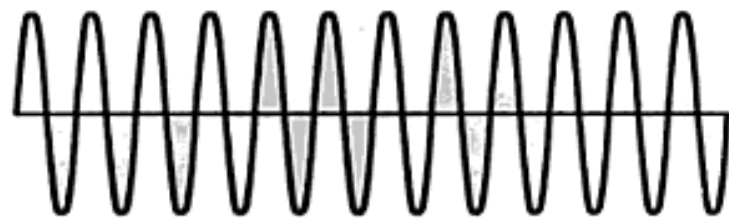
(a)

- (a) Equivalent circuit of balanced modulator showing positive half cycle of carrier
- (b) Equivalent circuit of balanced modulator showing negative half cycle of carrier

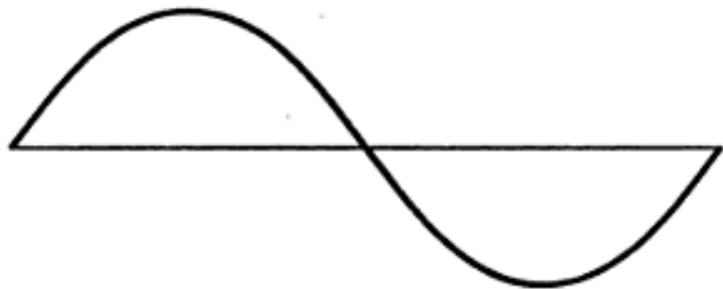


(b)

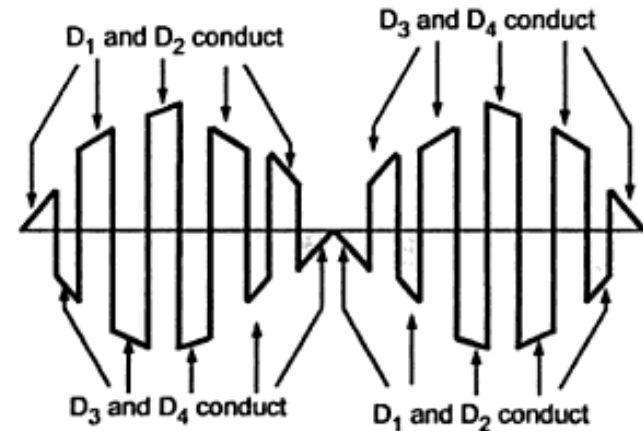
DSB-SC - *Implementation*



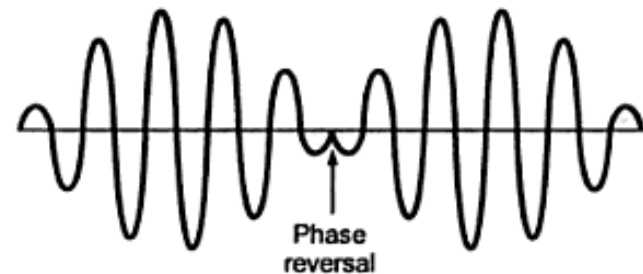
(a) Carrier



(b) Modulating signal



(c) DSB signal - primary T_2



(d) DSB output

Fig. 1.5.6 (a) A carrier signal (b) Sinusoidal modulating signal (c) DSB signal at primary of T_2 (d) DSB output at secondary of T_2

DSB-SC - *Implementation*

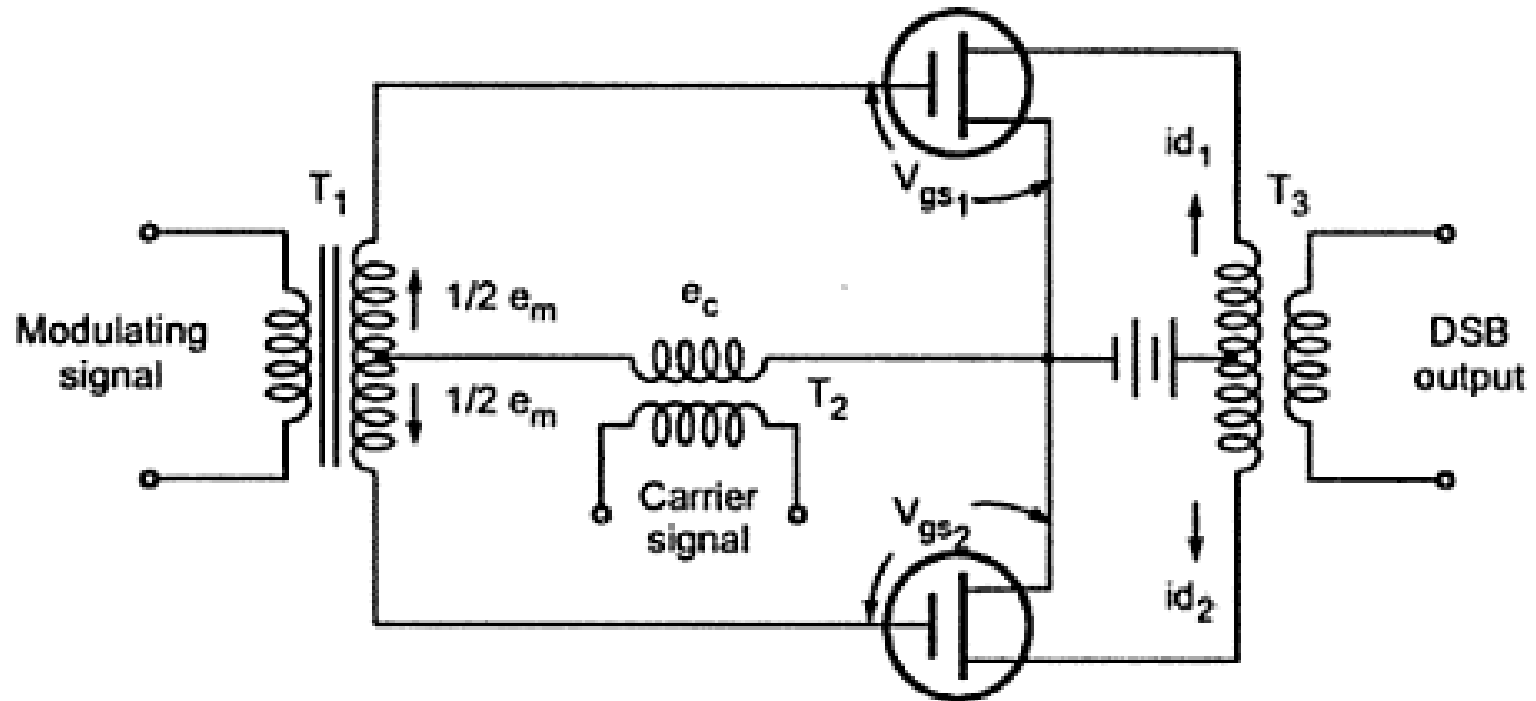
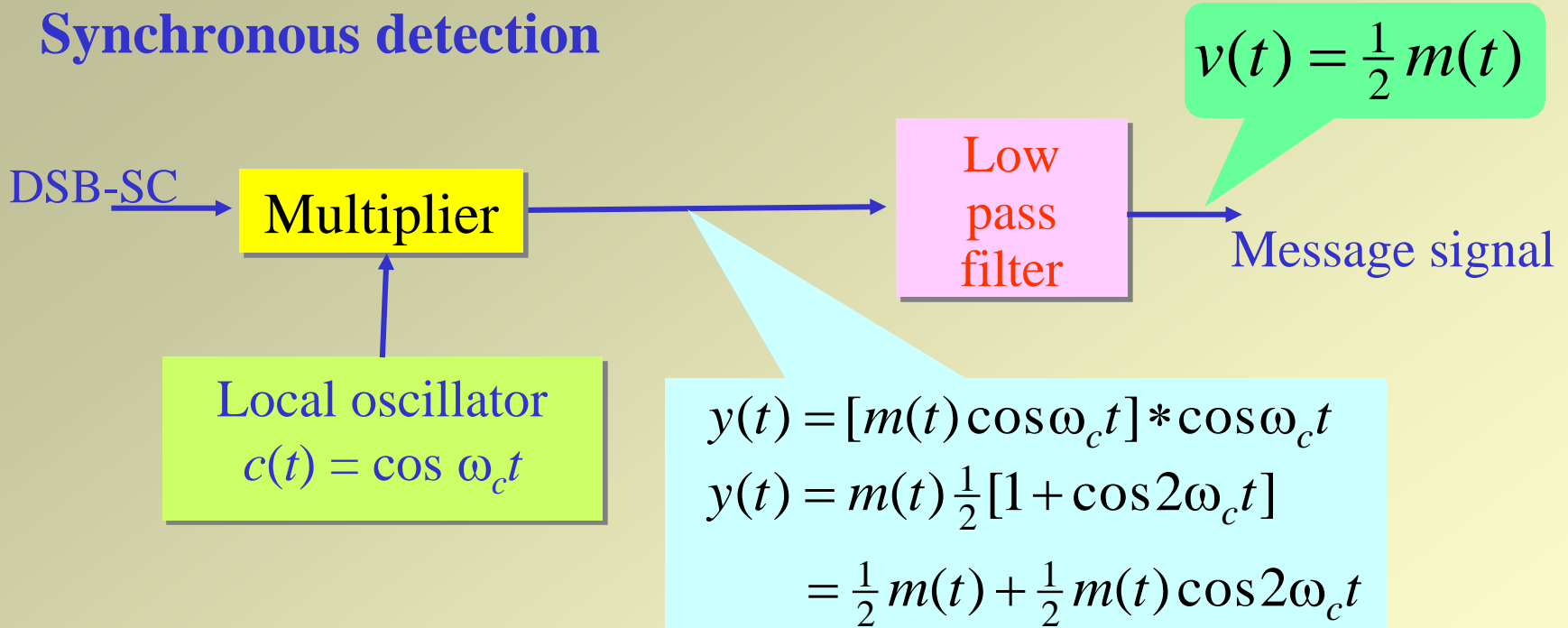


Fig. 1.5.7 Balanced modulator using FETs

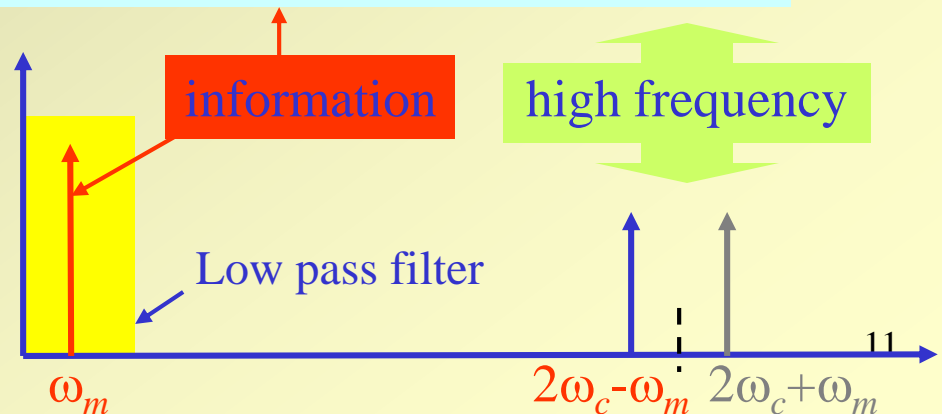
DSB-SC - *Detection*

- Synchronous detection



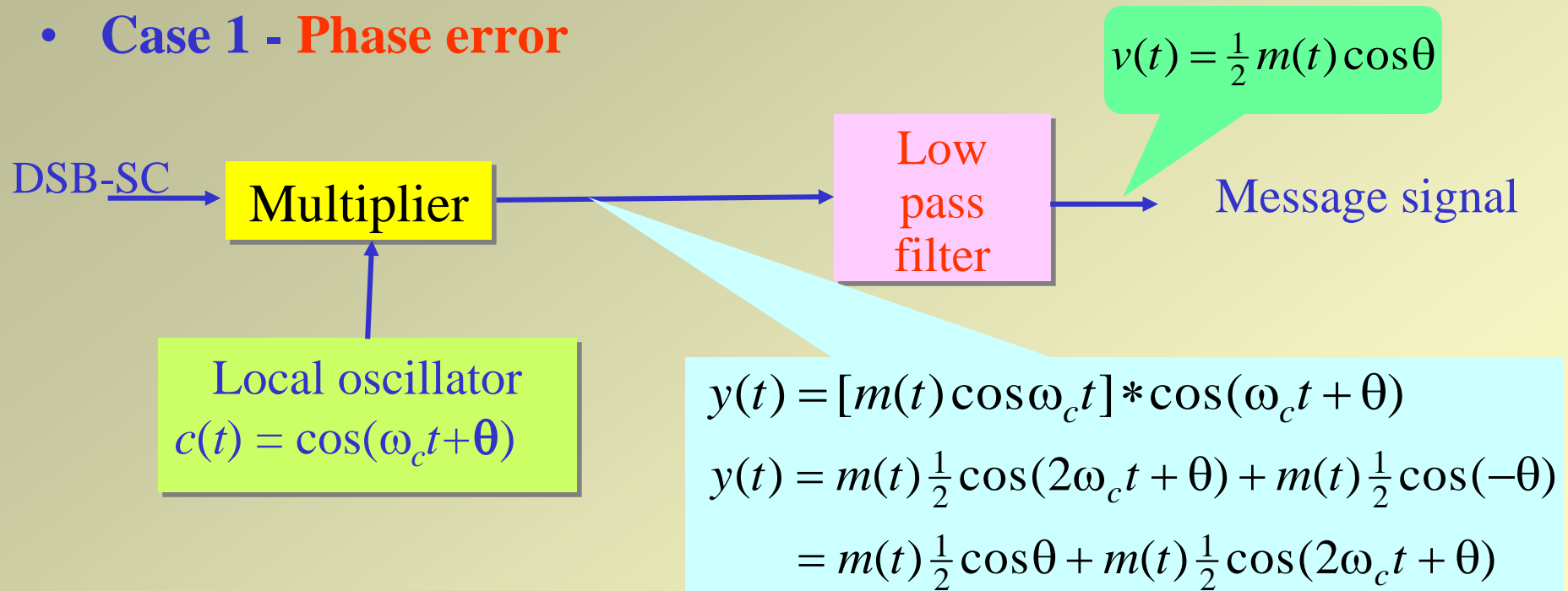
Condition:

- Local oscillator has the same **frequency** and **phase** as that of the carrier signal at the transmitter.



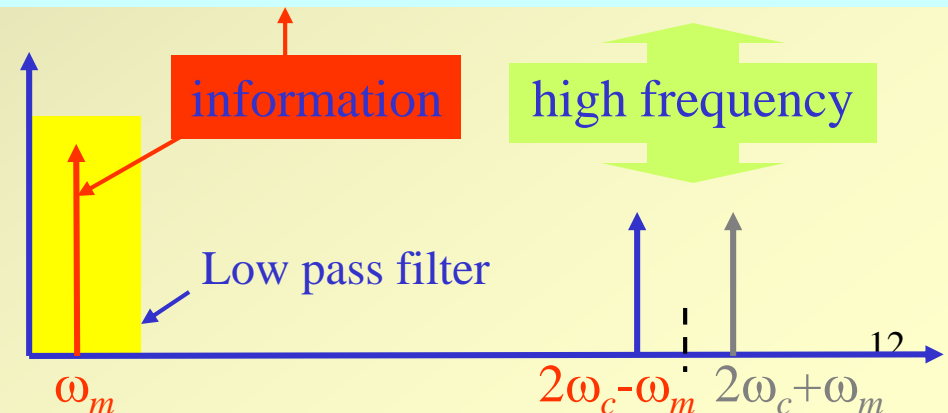
DSB-SC - *Synchronous Detection*

- Case 1 - Phase error



Condition:

- Local oscillator has the same **frequency** but *different phase* compared to carrier signal at the transmitter.



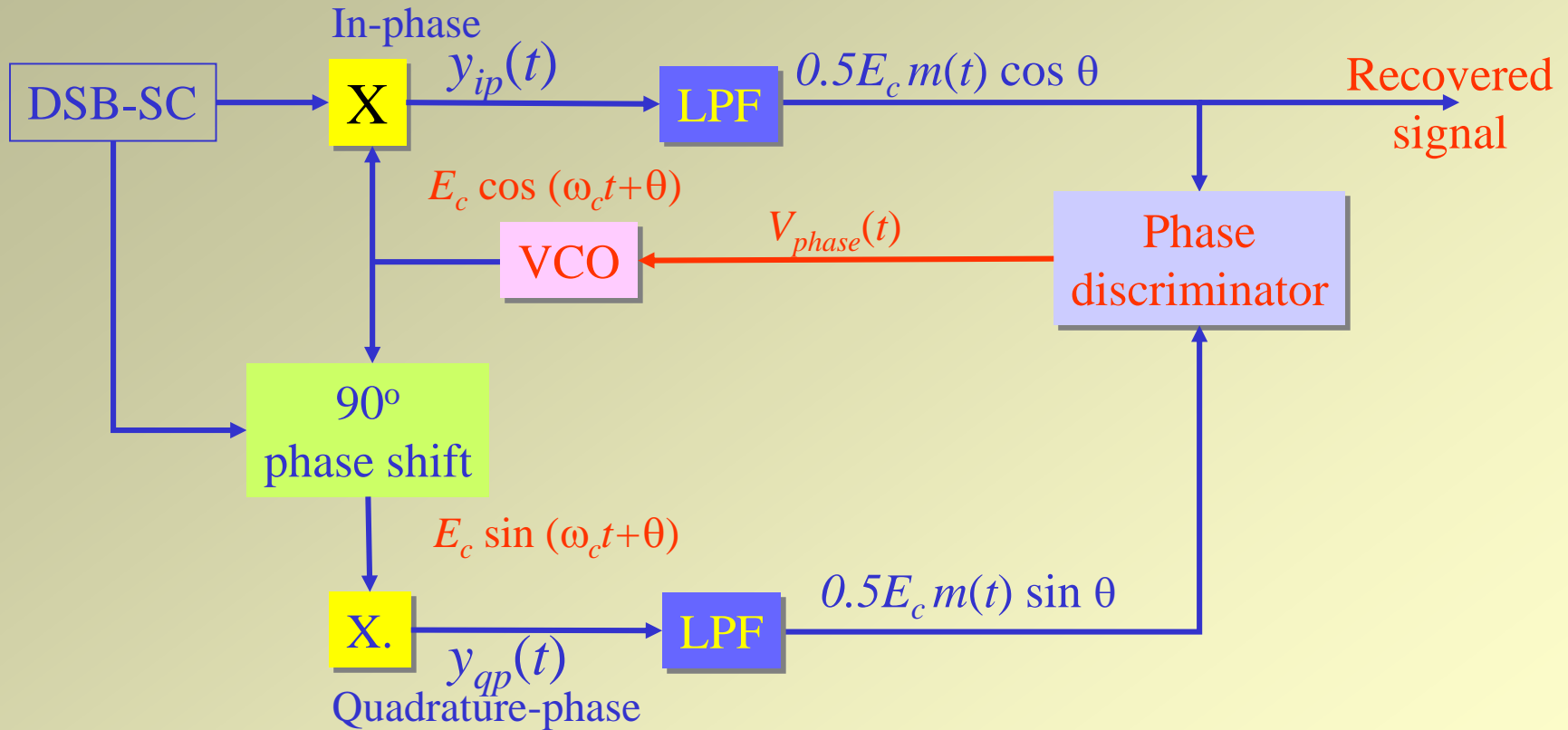
Demodulation of DSB-SC AM Signals

- Consequently, the output of the ideal lowpass filter

$$y_l(t) = \frac{1}{2} A_c m(t) \cos(\phi)$$

- Note that $m(t)$ is multiplied by $\cos(\phi)$
 - So the power in the demodulated signal is decreased by a factor of $\cos^2\phi$
 - Thus, the desired signal is scaled in amplitude by a factor that depends on the phase ϕ of the locally generated sinusoid
 1. When $\phi \neq 0$, the amplitude of the desired signal is reduced by the factor $\cos(\phi)$
 2. If $\phi = 45^\circ$, the amplitude of the signal is reduced by $\sqrt{2}$ and the power is reduced by a factor of two
 3. If $\phi = 90^\circ$, the desired signal component vanishes

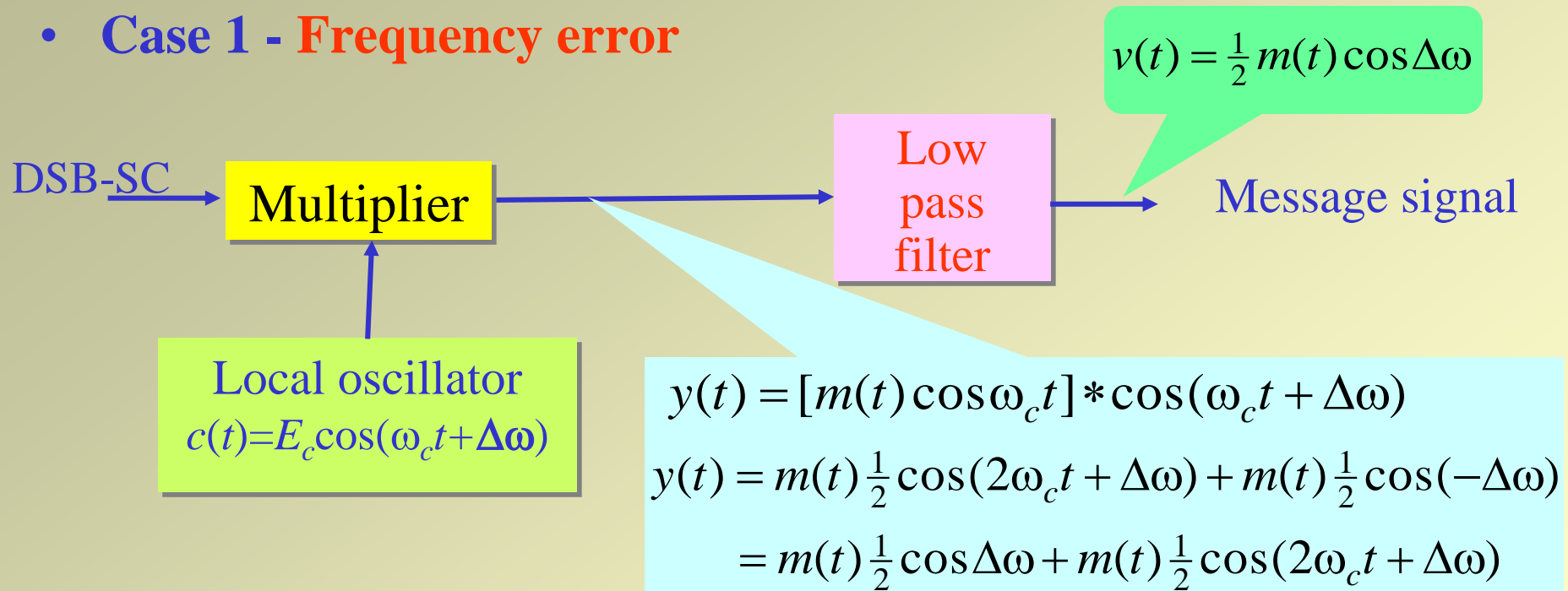
Phase Synchronisation - *Costa Loop*



- When there is no phase error. The quadrature component is zero
- When $\theta \neq 0$, $y_{ip}(t)$ decreases, while $y_{qp}(t)$ increases
- The out put of the phase discriminator is proportional to θ

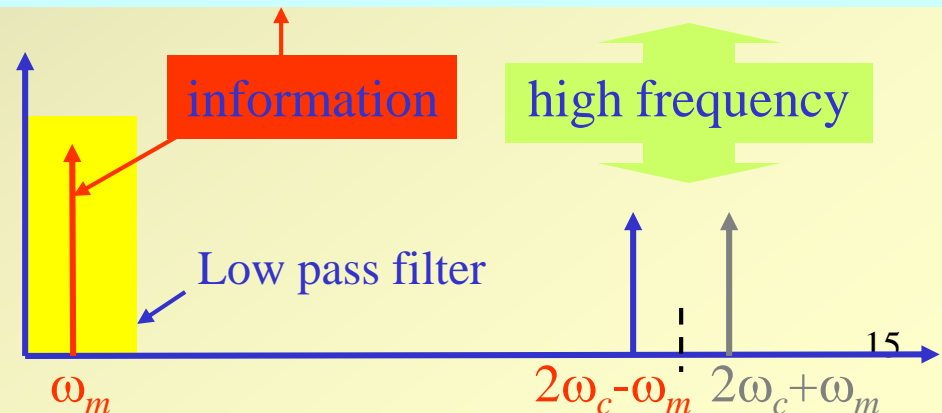
DSB-SC - *Synchronous Detection*

- Case 1 - Frequency error

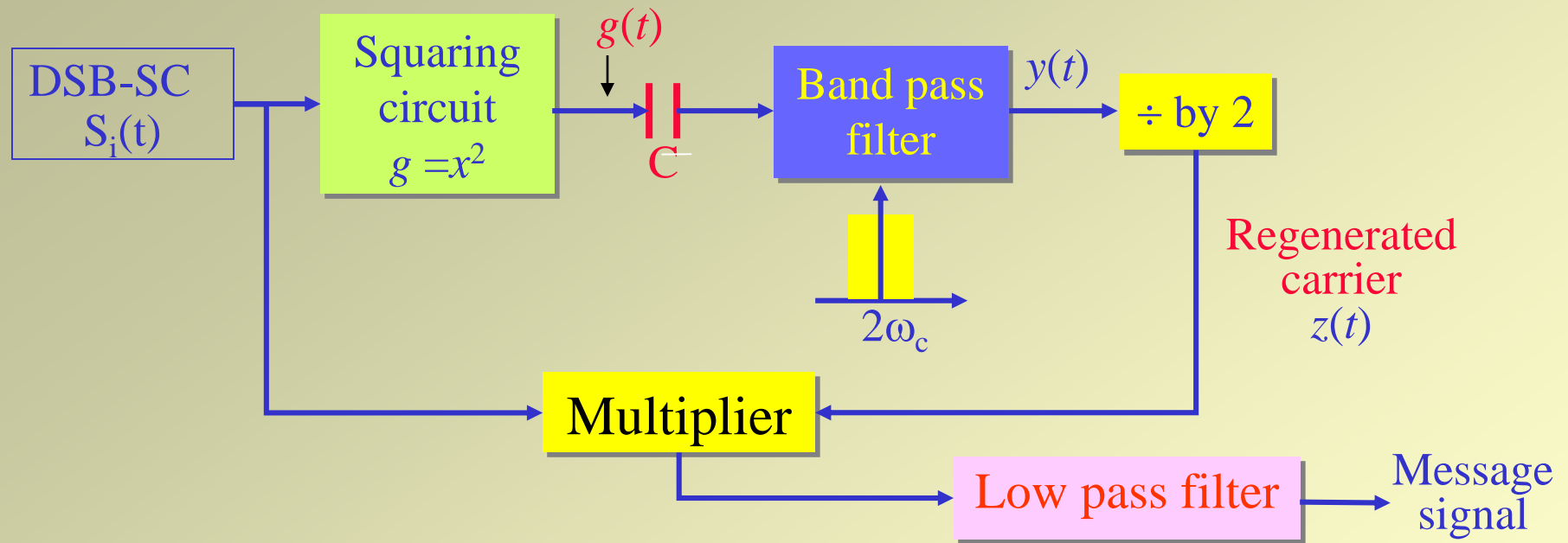


Condition:

- Local oscillator has the same **phase** but *different* **frequency** compared to carrier signal at the transmitter.



DSB-SC - *Square Detection*



$$\begin{aligned}
 g(t) &= S_i^2(t) = B^2 \cos^2 \omega_m t \cos^2 \omega_c t \\
 &= B^2 \left(\frac{1}{2} + \frac{1}{2} \cos 2\omega_m t \right) \left(\frac{1}{2} + \frac{1}{2} \cos 2\omega_c t \right) \\
 &= B^2/4 \left[1 + \frac{1}{2} \cos 2(\omega_c + \omega_m)t + \frac{1}{2} \cos 2(\omega_c - \omega_m)t + \cos 2\omega_m t + \underline{\cos 2\omega_c t} \right]
 \end{aligned}$$

$$y(t) = B^2/4 \cos 2\omega_c t$$

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$$z(t) = B^2/4 \cos \omega_c t$$

DSB-SC - *Power*

- The total power (or average power):

$$\begin{aligned} P_{T-DSB-SC} &= \frac{2}{R} \left[\frac{kA_c / \sqrt{2}}{2} \right]^2 \\ &= \frac{(kA_c)^2}{4R} \end{aligned}$$

- The maximum and peak envelop power

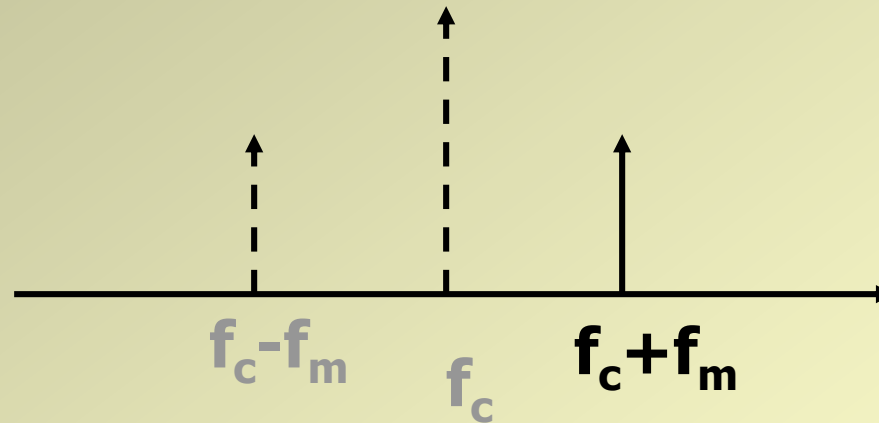
$$P_{P-DSB-SC} = \frac{(kA_c)^2}{R}$$

DSB-SC - *Summary*

- **Advantages:**
 - Lower power consumption
- **Disadvantage:**
 - Complex detection
- **Applications:**
 - Analogue TV systems: to transmit colour information
 - For transmitting *stereo* information in FM sound broadcast at VHF

Single Side band Suppress Carrier (SSB-SC)

Frequency spectrum:



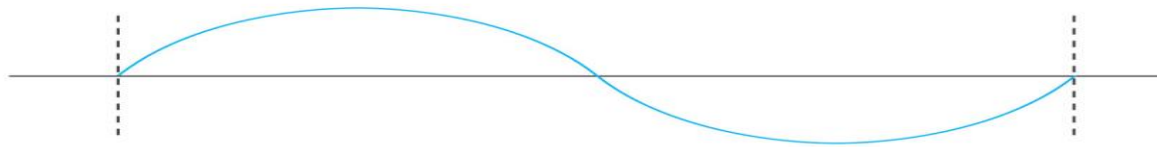
$$\text{Bandwidth} = f_{m(\max)}$$

$$\text{Total Power} = P_{\text{usb}}$$

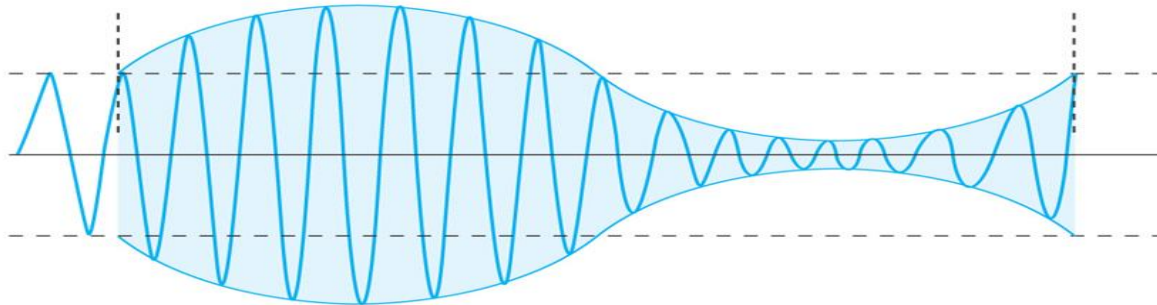
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- A form of amplitude modulation in which the carrier is totally suppressed and one of the sidebands removed.
- Therefore, SSBSC requires half as much bandwidth as conventional DSB AM and considerably less transmitted power

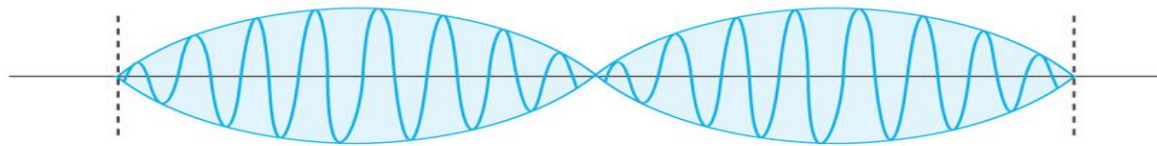
Comparison of time domain representation of three common AM transmission systems:



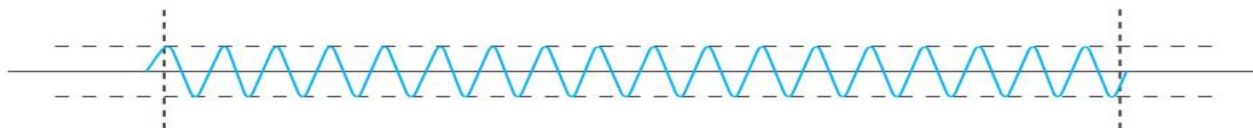
(a) Modulating signal



(b) DSBFC wave



(c) DSBSC wave



(d) SSBSC wave

Example 1

For an AM DSCFC wave with a peak unmodulated carrier voltage $V_c = 10 \text{ V}_p$, frequency of 100kHz, a load resistor of $R_L = 10 \Omega$, frequency of modulating signal of 10kHz and $m = 1$, determine the following

- i) Powers of the carrier and the upper and lower sidebands.
- ii) Total power of the modulated wave.
- iii) Bandwidth of the transmitted wave.
- iv) Draw the power and frequency spectrum.

Example 1..cont'd

- For the same given values, determine questions (ii)-(iv) for a AM DSB-SC, AM SSB-FC and AM SSB-SC systems. Determine also the percentage of power saved in each of the system design.

Example 1..cont'd

○ Solution for DSBFC;

i)

$$P_c = \frac{(V_c / \sqrt{2})^2}{R} = \frac{V_c^2}{2R} = \frac{(10)^2}{2 * 10} = 5W$$
$$P_{usb} = P_{lsb} = \frac{k^2 P_c}{4} = 1.25W$$

ii)

$$P_t = P_c + \frac{k^2}{4} P_c + \frac{k^2}{4} P_c$$
$$= 5 + \frac{1^2}{4} (5) + \frac{1^2}{4} (5) = 7.5W$$

iii) Bandwidth = $2 \times f_{\text{mmax}} = 2(10\text{kHz}) = 20\text{kHz}$

Example 1..cont'd

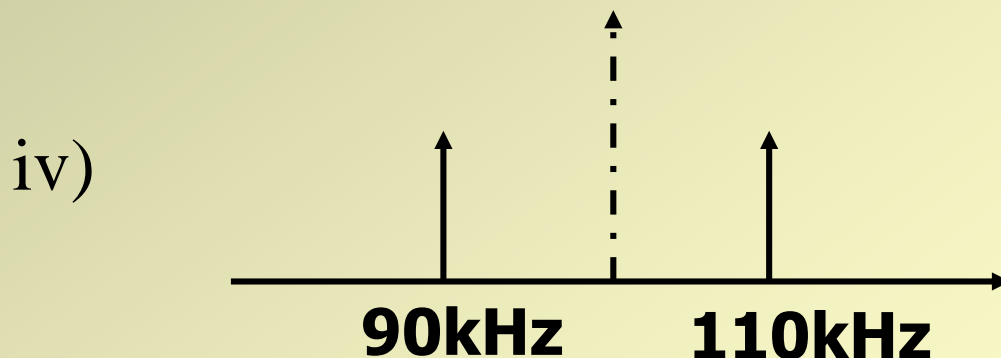
- Solution: For DSB-SC

$$\begin{aligned} \text{ii)} \quad P_t &= \frac{k^2}{4} P_c + \frac{k^2}{4} P_c \\ &= \frac{1^2}{4} (5) + \frac{1^2}{4} (5) = 2.5W \end{aligned}$$

$$\begin{aligned} Power_{\text{saved}} &= 7.5W - 2.5W \\ &= 5W \end{aligned}$$

$$\begin{aligned} \% Power_{\text{saved}} &= \frac{5W}{7.5W} \times 100\% \\ &= 66.67\% \end{aligned}$$

$$\text{iii)} \text{ Bandwidth} = 2 \times f_{\text{mmax}} = 2(10\text{kHz}) = 20\text{kHz}$$



Example 1..cont'd

- Solution: For SSB-FC

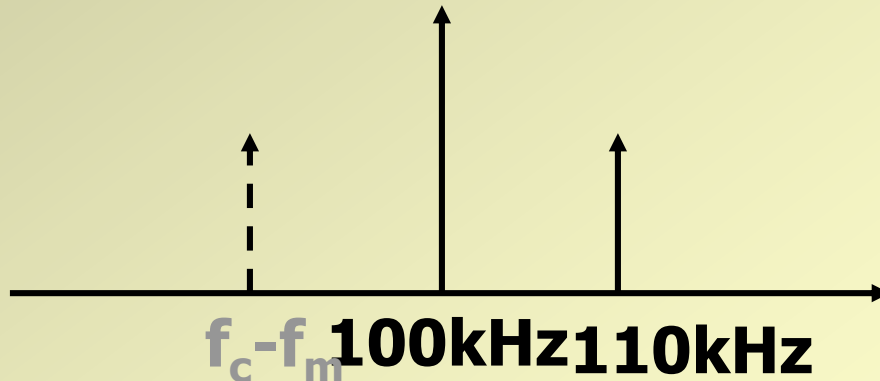
$$\begin{aligned} \text{ii)} \quad P_t &= P_c + \frac{k^2}{4} P_c \\ &= 5 + \frac{1^2}{4} (5) = 6.25W \end{aligned}$$

$$\begin{aligned} \text{Power}_{\text{saved}} &= 7.5W - 6.25W \\ &= 1.25W \end{aligned}$$

$$\begin{aligned} \% \text{Power}_{\text{saved}} &= \frac{1.25W}{7.5W} \times 100\% \\ &= 16.67\% \end{aligned}$$

$$\text{iii)} \text{ Bandwidth} = f_{\text{mmax}} = 10\text{kHz}$$

iv)



Example 1..cont'd

- Solution: For SSB-sC

ii)

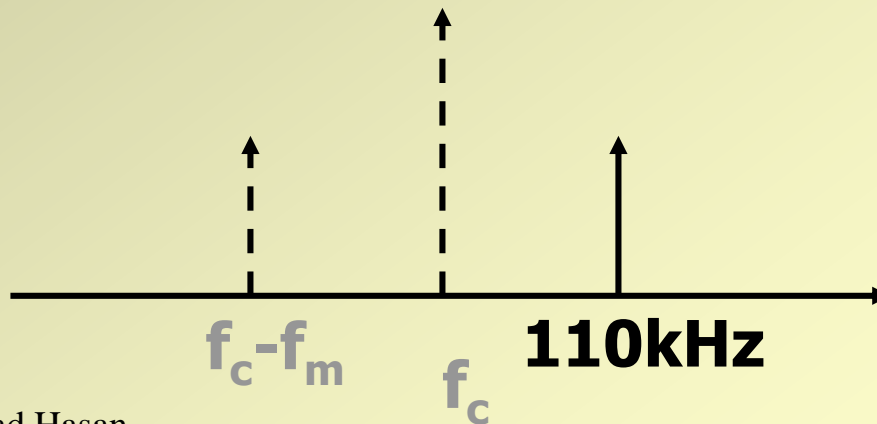
$$P_t = \frac{k^2}{4} P_c$$
$$= \frac{1^2}{4} (5) = 1.25W$$

$$Power_{saved} = 7.5W - 1.25W$$
$$= 6.25W$$

$$\% Power_{saved} = \frac{6.25W}{7.5W} \times 100\%$$
$$= 83.33\%$$

iii) Bandwidth = $f_{mmax} = 10\text{kHz}$

iv)

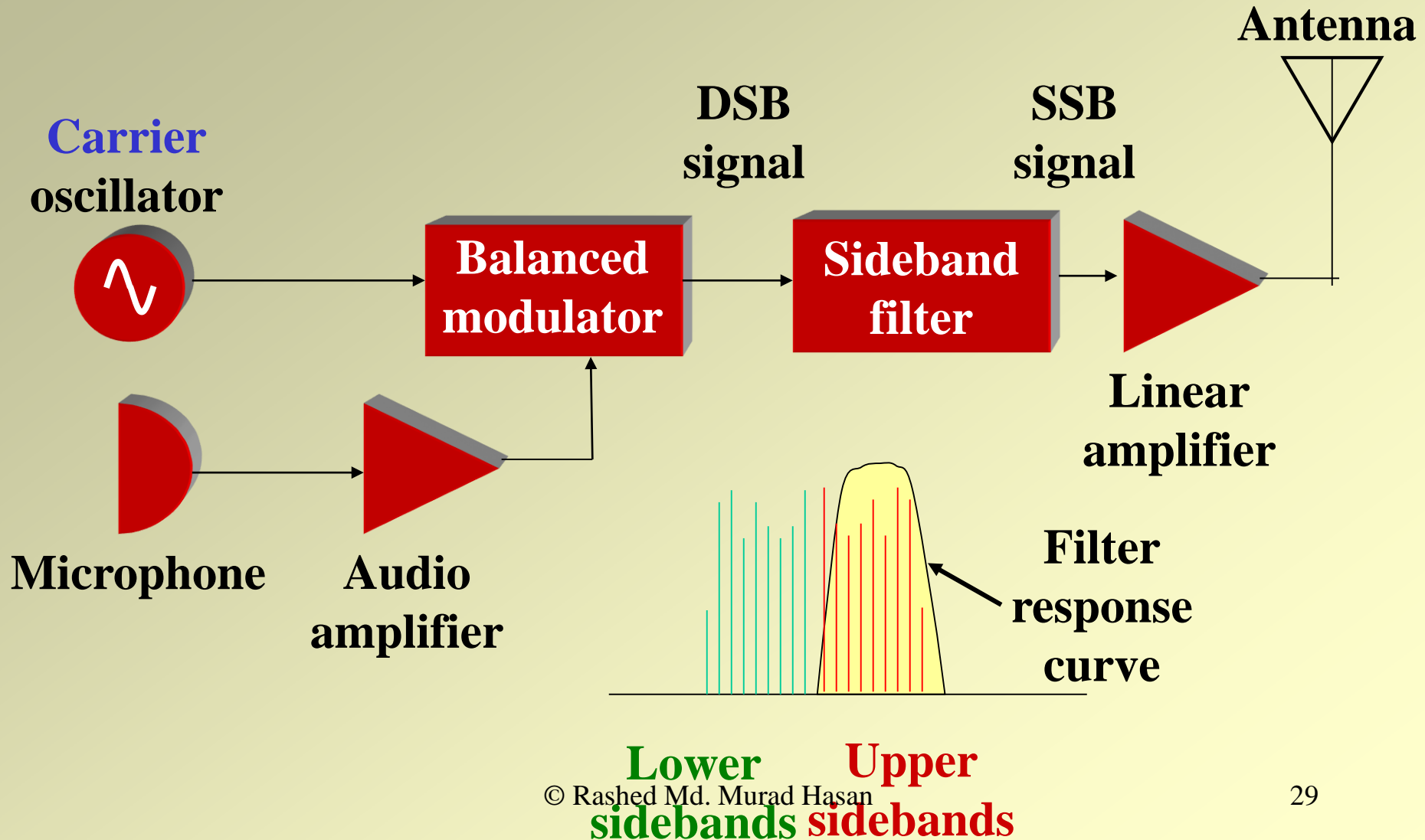


Methods of Generating SSB

i) Filtering method

- A filter removes the undesired sideband producing SSB.
- Balanced modulators is used to suppress the unwanted carrier and filters to suppress the unwanted sidebands
- Quartz crystal filters are the most widely used sideband filters since they are very selective and inexpensive.

Block diagram of filtering method

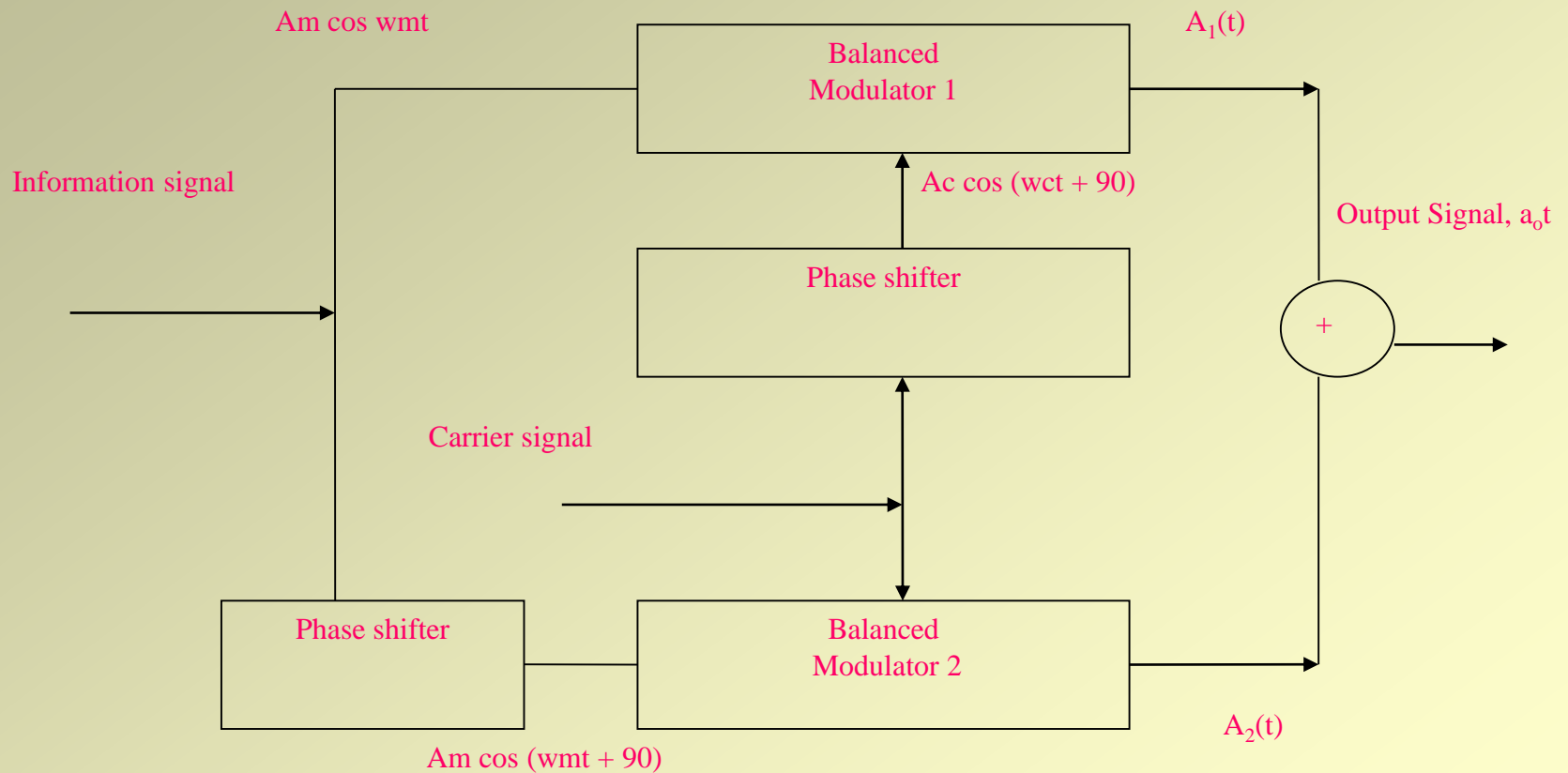


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ii) Phasing method using two balance modulator

- Another way to produce SSB uses a phase shift method to eliminate one sideband.
- Two balanced modulators driven by carriers and modulating signals 90° out of phase produce DSB.
- Adding the two DSB signals together results in one sideband being cancelled out.

Block diagram of phasing method



Mathematical analysis of phasing method

$$a_0(t) = a_1(t) + a_2(t) \dots \dots \dots > (1)$$

$$a_1(t) = A_c \cos(w_c t + 90^0) * A_m \cos w_m t$$

$$= \frac{1}{2} \{ A_c A_m \cos(w_c t + 90^0 - w_m t) + A_c A_m \cos(w_c t + 90^0 + w_m t) \} \dots \dots > (2)$$

$$a_2(t) = A_c \cos(w_c t) * A_m \cos(w_m t + 90^0)$$

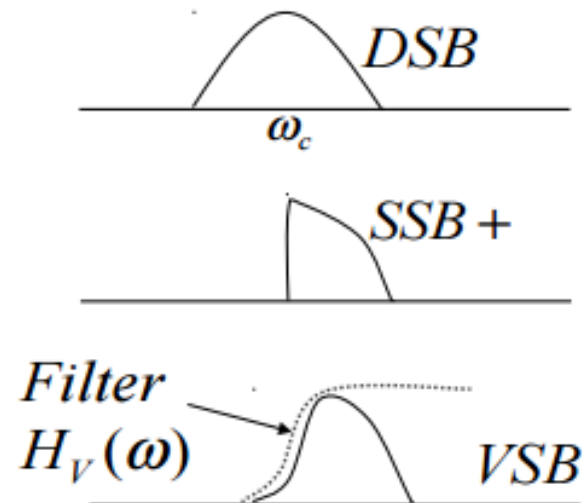
$$= \frac{1}{2} \{ A_c A_m \cos(w_c t - 90^0 - w_m t) + A_c A_m \cos(w_c t + 90^0 + w_m t) \} \dots \dots > (3)$$

$$a_0(t) = (2) + (3)$$

$$= \{ A_c A_m \cos(w_c t + 90^0 + w_m t) \}$$

Vestigial-Sideband Modulation (VSB)

- ◆ DSB modulation is wasteful of bandwidth.
- ◆ SSB is difficult to generate when modulating signal bandwidth is wide or has low frequency components.
- ◆ VSB is a compromise :
 - Partial suppression of one sideband reduces bandwidth
 - Filtering requirements are not as severe as SSB
 - Can demodulate via synchronous detection or envelope detection if “large carrier” transmitted



Vestigial-Sideband Modulation (VSB)

- ◆ Transmitted signal

$$\Phi_{VSB}(\omega) = \left[\frac{1}{2} F(\omega - \omega_c) + \frac{1}{2} F(\omega + \omega_c) \right] H_V(\omega)$$

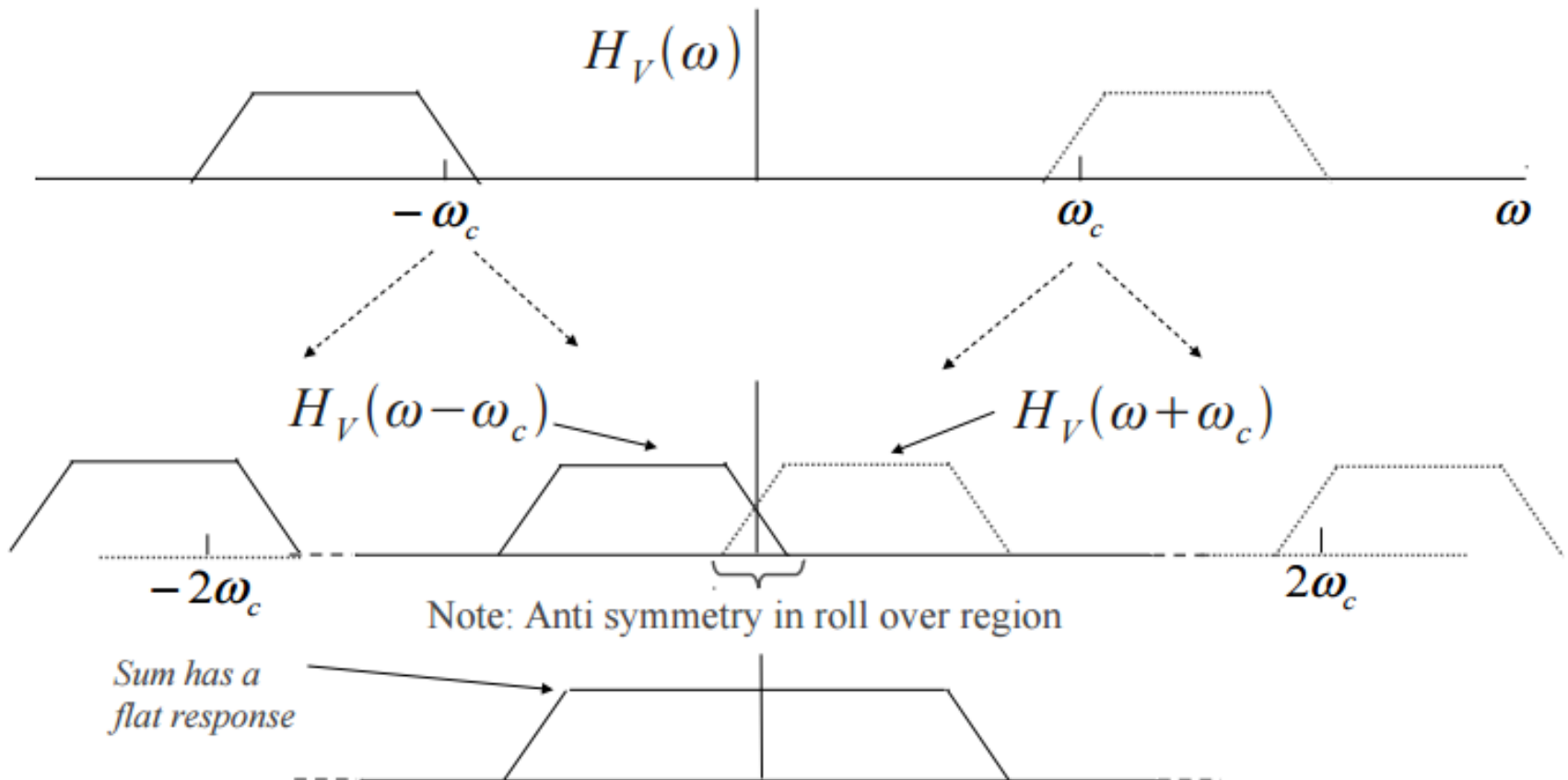
- ◆ We must design the “roll off” of $H_V(\omega)$ such that simple synchronous detection results in perfect signal recovery.
- ◆ To design $H_V(\omega)$, need to consider demodulation.

$$e_0(t) = [\phi_{VSB}(t) \cos \omega_c t]_{LPF}$$

$$\begin{aligned} e_0(\omega) &= \left[\frac{1}{2} \Phi_{VSB}(\omega + \omega_c) + \frac{1}{2} \Phi_{VSB}(\omega - \omega_c) \right]_{LPF} \\ &= \frac{1}{4} F(\omega) H_V(\omega + \omega_c) + \frac{1}{4} F(\omega) H_V(\omega - \omega_c) \\ &= \frac{1}{4} F(\omega) [H_V(\omega + \omega_c) + H_V(\omega - \omega_c)] \end{aligned}$$

Vestigial-Sideband Demodulation

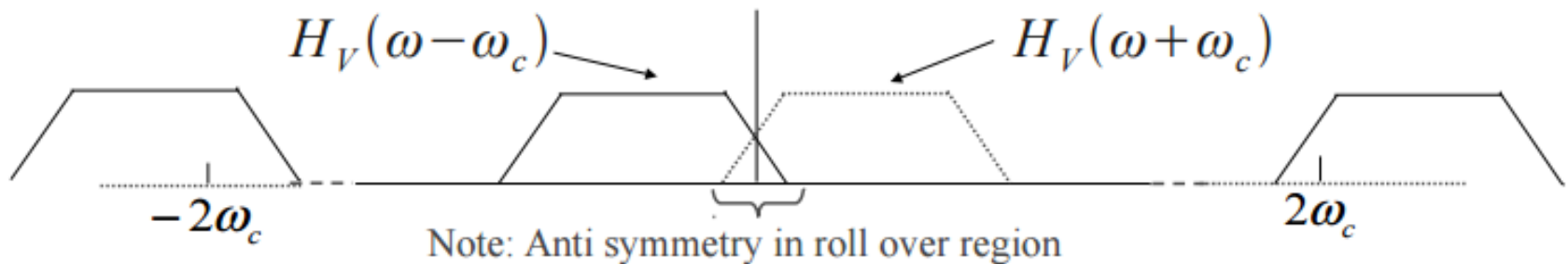
Consider the demodulation process



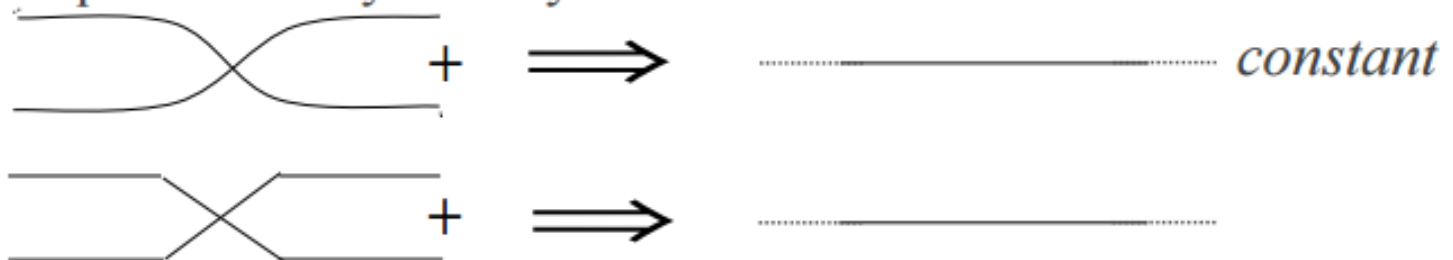
Vestigial-Sideband Modulation (VSB)

- ◆ Thus we need to arrange
 $[H_V(\omega + \omega_c) + H_V(\omega - \omega_c)] = \text{constant}$

The illustration shows the spectrum after the mixer.



- ◆ Examples of antisymmetry roll off.



- ◆ VSB is used for analogue television transmissions to conserve bandwidth.
- ◆ South Africa uses the “PAL-I” standard for analogue TV broadcasts, same as UK. It is similar to the NTSC standard used in the USA.
- ◆ There are many different variants of the PAL standard.

VSB - used for TV signal

*NTSC
(USA Standard)
~~substandard~~
B = 6 MHz*

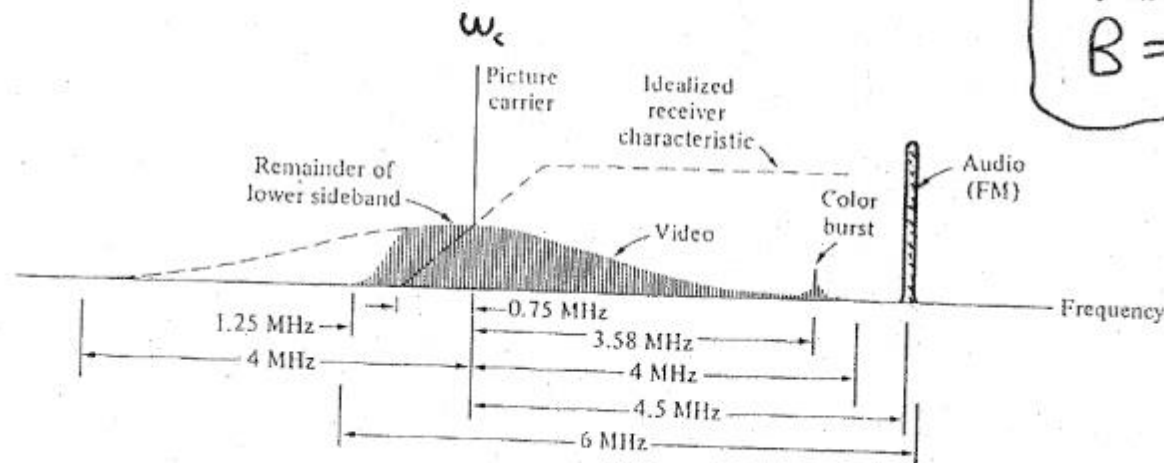


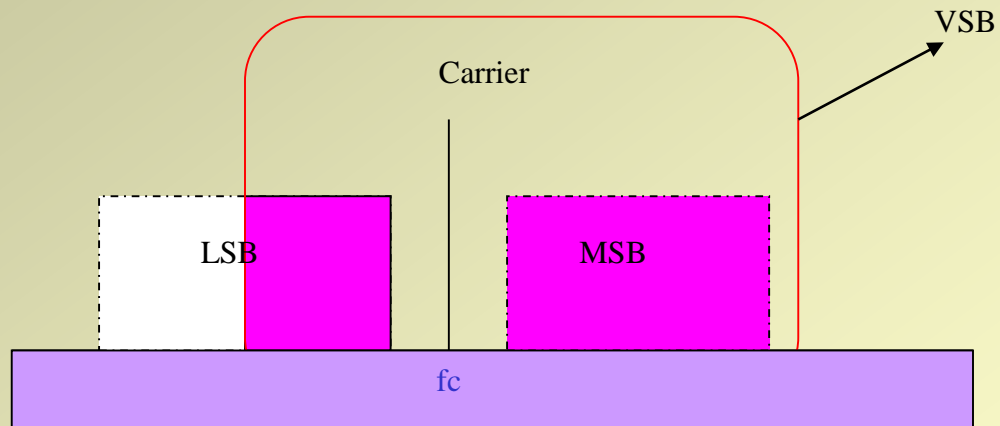
Figure 5.38 Spectrum of a public television transmission.

Cont'd

- AM wave is applied to a vestigial sideband filter, producing a modulation scheme – VSB + C
- Mainly used for television video transmission.

Cont'd...vsb

➤ VSB Frequency Spectrum



Cont'd...vsb

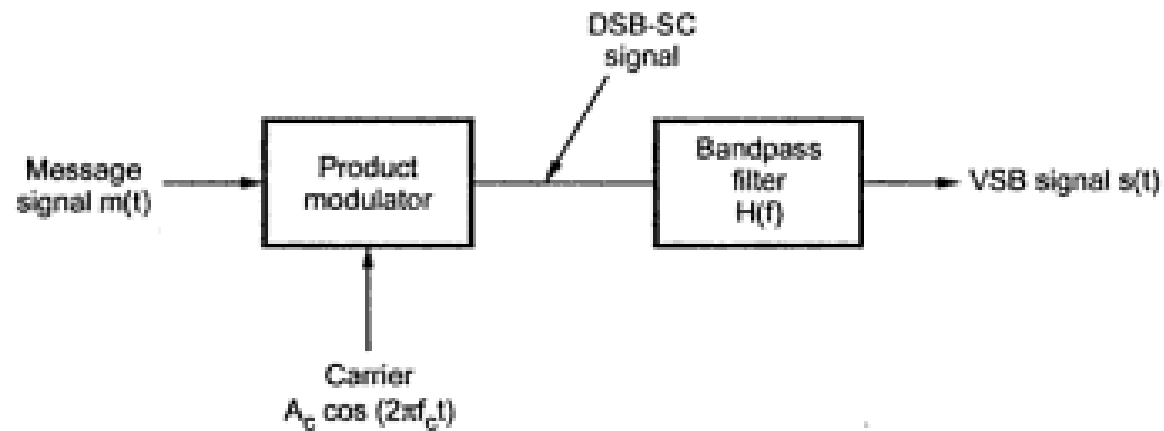


Fig. 1.6.1 Generation of VSB signal

Advantages/Benefits of SSB

- Power consumption
- Bandwidth conservation
- Selective fading
- Noise reduction

Disadvantages of SSB

Vestigial-Sideband Modulation (VSB)

- ◆ DSB modulation is wasteful of bandwidth.
- ◆ SSB is difficult to generate when modulating signal bandwidth is wide or has low frequency components.
- ◆ VSB is a compromise :
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