# MODEL PREDICTIVE CONTROL

### **HYBRID MPC**

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## **COURSE STRUCTURE**

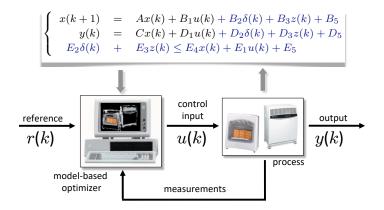
- ✓ Basic concepts of model predictive control (MPC) and linear MPC
- ✓ Linear time-varying and nonlinear MPC
- ✓ MPC computations: quadratic programming (QP), explicit MPC
  - Hybrid MPC
  - Stochastic MPC
  - Data-driven MPC

#### Course page:

http://cse.lab.imtlucca.it/~bemporad/mpc course.html



### HYBRID MODEL PREDICTIVE CONTROL



Use a hybrid dynamical model of the process to predict its future evolution and choose the "best" control action

Finite-horizon optimal control problem (regulation)

min 
$$\sum_{k=0}^{N-1} y_k' Q y_k + u_k' R u_k$$
s.t. 
$$\begin{cases} x_{k+1} &= A x_k + B_1 u_k + B_2 \delta_k + B_3 z_k + B_5 \\ y_k &= C x_k + D_1 u_k + D_2 \delta_k + D_3 z_k + D_5 \\ E_2 \delta_k &+ E_3 z_k \le E_4 x_k + E_1 u_k + E_5 \\ x_0 &= x(t) \end{cases}$$

$$Q=Q'\succ 0, R=R'\succ 0$$

- Treat  $u_k, \delta_k, z_k$  as free decision variables,  $k = 0, \dots, N-1$
- Predictions can be constructed exactly as in the linear case

$$x_k = A^k x_0 + \sum_{j=0}^{k-1} A^j (B_1 u_{k-1-j} + B_2 \delta_{k-1-j} + B_3 z_{k-1-j} + B_5)$$

• After substituting  $x_k,y_k$  the resulting optimization problem becomes the following Mixed-Integer Quadratic Programming (MIQP) problem

$$\begin{aligned} &\min_{\xi} && \frac{1}{2}\xi'H\xi + x'(t)F'\xi + \frac{1}{2}x'(t)Yx(t)\\ &\text{s.t.} && G\xi \leq W + Sx(t) \end{aligned}$$

• The optimization vector  $\xi=[u_0,\dots,u_{N-1},\delta_0,\dots,\delta_{N-1},z_0,\dots,z_{N-1}]$  has mixed real and binary components

$$u_k \in \mathbb{R}^{m_c} \times \{0, 1\}^{m_b}$$

$$\delta_k \in \{0, 1\}^{r_b}$$

$$z_k \in \mathbb{R}^{r_c}$$

$$\xi \in \mathbb{R}^{N(m_c + r_c)} \times \{0, 1\}^{N(m_b + r_b)}$$

## HYBRID MPC FOR REFERENCE TRACKING

• Consider the more general set-point tracking problem

$$\min_{\xi} \sum_{k=0}^{N-1} \|y_k - r\|_Q^2 + \|u_k - u_r\|_R^2 + \sigma (\|x_k - x_r\|_2^2 + \|\delta_k - \delta_r\|_2^2 + \|z_k - z_r\|_2^2)$$

s.t. MLD model equations

$$x_0 = x(t)$$
$$x_N = x_r$$

with  $\sigma>0$  and  $\|v\|_Q^2=v'Qv$ 

• The equilibrium  $(x_r,u_r,\delta_r,z_r)$  corresponding to r can be obtained by solving the following mixed-integer feasibility problem

$$x_r = Ax_r + B_1u_r + B_2\delta_r + B_3z_r + B_5$$

$$r = Cx_r + D_1u_r + D_2\delta_r + D_3z_r + D_5$$

$$E_2\delta_r + E_3z_r \le E_4x_r + E_1u_r + E_5$$

• Theorem. Let  $(x_r,u_r,\delta_r,z_r)$  be the equilibrium corresponding to r. Assume x(0) such that the MIQP problem is feasible at time t=0. Then  $\forall Q,R\succ 0$ ,  $\sigma>0$  the hybrid MPC closed-loop converges asymptotically

$$\lim_{t \to \infty} y(t) = r \qquad \qquad \lim_{t \to \infty} x(t) = x_r$$

$$\lim_{t \to \infty} \delta(t) = \delta_r$$

$$\lim_{t \to \infty} u(t) = u_r \qquad \qquad \lim_{t \to \infty} z(t) = z_r$$

and all constraints are fulfilled at each time  $t \geq 0$ .

- The proof easily follows from standard Lyapunov arguments (see next slide)
- Lyapunov asymptotic stability and exponential stability follows if proper terminal cost and constraints are imposed (Lazar, Heemels, Weiland, Bemporad, 2006)

## **CONVERGENCE PROOF**

- Main idea: Use the value function  $V^*(x(t))$  as a Lyapunov function
- Let  $\xi_t=[u_0^t,\dots,u_{N-1}^t,\delta_0^t,\dots,\delta_{N-1}^t,z_0^t,\dots,z_{N-1}^t]$  be the optimal sequence @t
- By construction @t+1  $\bar{\xi}=[u_1^t,\ldots,u_{N-1}^t,u_r,\delta_1^t,\ldots,\delta_{N-1}^t,\delta_r,z_0^t,\ldots,z_{N-1}^t,z_r]$  is feasible, as it satisfies all MLD constraints + terminal constraint  $x_N=x_r$
- $$\begin{split} \bullet \ \ \text{The cost of } \bar{\xi} \text{ is } V^*(x(t)) \|y(t) r\|_Q^2 \|u(t) u_r\|_R^2 \\ -\sigma \left( \|\delta(t) \delta_r\|_2^2 + \|z(t) z_r\|_2^2 + \|x(t) x_r\|_2^2 \right) & \\ \geq V^*(x(t+1)) \end{split}$$
- $V^*(x(t))$  is monotonically decreasing and  $\geq 0$ , so  $\exists \lim_{t \to \infty} V^*(x(t)) \in \mathbb{R}$
- $\bullet \ \ \text{Hence} \ \|y(t)-r\|_Q^2, \|u(t)-u_r\|_R^2, \|\delta(t)-\delta_r\|_2^2, \|z(t)-z_r\|_2^2, \|x(t)-x_r\|_2^2 \to 0$
- Since  $R,Q\succ 0, \lim_{t\rightarrow \infty}y(t)=r$  and all other variables converge.  $\qed$

Global optimum is not needed to prove convergence!

Finite-horizon optimal control problem using infinity norms

$$\min_{\xi} \sum_{k=0}^{N-1} \|Qy_k\|_{\infty} + \|Ru_k\|_{\infty}$$
s.t. 
$$\begin{cases} x_{k+1} &= Ax_k + B_1u_k + B_2\delta_k + B_3z_k + B_5 \\ y_k &= Cx_k + D_1u_k + D_2\delta_k + D_3z_k + D_5 \\ E_2\delta_k &+ E_3z_k \le E_4x_k + E_1u_k + E_5 \\ x_0 &= x(t) \end{cases} \qquad Q \in \mathbb{R}^{m_y \times n_y}$$

• Introduce additional variables  $\epsilon_k^y, \epsilon_k^u, k=0,\dots,N-1$ 

$$\left\{ \begin{array}{ll} \epsilon_k^y & \geq & \|Qy_k\|_\infty \\ \epsilon_k^u & \geq & \|Ru_k\|_\infty \end{array} \right. \qquad \left\{ \begin{array}{ll} \epsilon_k^y & \geq & \pm Q^i y_k \\ \epsilon_k^u & \geq & \pm R^i u_k \end{array} \right. \quad Q^i = i \mathrm{th} \ \mathrm{row} \ \mathrm{of} \ Q$$

• After substituting  $x_k, y_k$  the resulting optimization problem becomes the following Mixed-Integer Linear Programming (MILP) problem

$$\min_{\xi} \sum_{k=0}^{N-1} \epsilon_k^y + \epsilon_k^u$$
s.t.  $G\xi \le W + Sx(t)$ 

•  $\xi = [u_0, \dots, u_{N-1}, \delta_0, \dots, \delta_{N-1}, z_0, \dots, z_{N-1}, \epsilon_0^y, \epsilon_0^u, \dots, \epsilon_{N-1}^y, \epsilon_{N-1}^u]$  is the optimization vector, with mixed real and binary components

$$u_k \in \mathbb{R}^{m_c} \times \{0, 1\}^{m_b}$$

$$\delta_k \in \{0, 1\}^{r_b}$$

$$z_k \in \mathbb{R}^{r_c}$$

$$\epsilon_k^y, \epsilon_k^u \in \mathbb{R}$$

$$\xi \in \mathbb{R}^{N(m_c + r_c + 2)} \times \{0, 1\}^{N(m_b + r_b)}$$

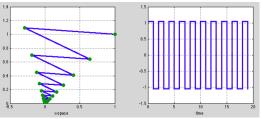
• Same approach applies to any convex piecewise affine stage cost

## HYBRID MPC EXAMPLE

PWA system:

$$\left\{ \begin{array}{rcl} x(t+1) & = & 0.8 \begin{bmatrix} \cos\alpha(t) & -\sin\alpha(t) \\ \sin\alpha(t) & \cos\alpha(t) \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t) \\ y(t) & = & \begin{bmatrix} 0 & 1 \end{bmatrix} x(t) \\ \alpha(t) & = & \begin{cases} -\frac{\pi}{3} & \text{if } \begin{bmatrix} 1 & 0 \end{bmatrix} x(t) \geq 0 \\ -\frac{\pi}{3} & \text{if } \begin{bmatrix} 1 & 0 \end{bmatrix} x(t) < 0 \end{array} \right.$$

Open-loop simulation:



go to demo demos/hybrid/bm99sim.m

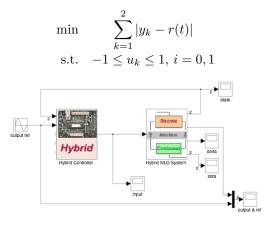
## **HYBRID MPC EXAMPLE**

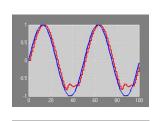
```
/* 2x2 PWA system - Example from the paper
   A. Bemporad and M. Morari, "Control of systems integrating logic, dynamics,
   and constraints, ' Automatica, vol. 35, no. 3, pp. 407-427, 1999.
   (C) 2003 by A. Bemporad, 2003 */
SYSTEM pwa {
INTERFACE {
            STATE { REAL x1 [-10,10];
                    REAL x2 [-10,10];}
            INPUT { REAL u [-1.1.1.1];}
            OUTPUT{ REAL y;}
            PARAMETER (
              REAL alpha = 1.0472; /* 60 deg in radiants */
              REAL C = cos(alpha);
              REAL S = sin(alpha);}
IMPLEMENTATION (
            AUX { REAL z1,z2;
                  BOOL sign; }
            AD { sign = x1 \le 0; }
            DA { z1 = \{IF \text{ sign THEN } 0.8*(C*x1+S*x2)\}
                        ELSE 0.8*(C*x1-S*x2) };
                  z2 = \{IF \text{ sign THEN } 0.8*(-S*x1+C*x2)\}
                        ELSE 0.8*(S*x1+C*x2) }; }
            CONTINUOUS { x1 = z1;
                         x2 = z2+u; }
            OUTPUT { v = x2; }
```

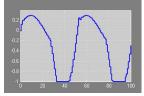
### go to demos/hybrid/bm99.hys

# HYBRID MPC EXAMPLE

• Closed-loop MPC results:







 Average CPU time to solve MILP:  $\approx 1$  ms/step (Macbook Pro 3GHz Intel Core i7 using GLPK)

## HYBRID MPC — TEMPERATURE CONTROL

#### >> C=hybcon(S,Q,N,limits,refs);

```
>> C

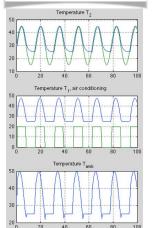
Hybrid controller based on MLD model S <heatcoolmodel.hys> [Inf-norm]

2 state measurement(s)
0 output reference(s)
0 input reference(s)
1 state reference(s)
0 reference(s) on auxiliary continuous z-variables
20 optimization variable(s) (8 continuous, 12 binary)
46 mixed-integer linear inequalities
sampling time = 0.5, MILP solver = 'glpk'

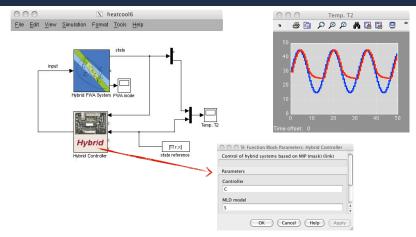
Type "struct(C)" for more details.
>>
```

>> [XX,UU,DD,ZZ,TT]=sim(C,S,r,x0,Tstop);

$$\min \sum_{k=1}^{2} \|x_{2k} - r(t)\|_{\infty}$$
 s.t. 
$$\begin{cases} x_{1k} \ge 25, \ k = 1, 2 \\ \text{MLD model} \end{cases}$$



## HYBRID MPC — TEMPERATURE CONTROL



• Average CPU time to solve MILP:  $\approx 1$  ms/step (Macbook Pro 3GHz Intel Core i7 using GLPK)

## MIXED-INTEGER PROGRAMMING SOLVERS

- Binary constraints make Mixed-Integer Programming (MIP) a hard problem ( $\mathcal{NP}$ -complete)
- However, excellent general purpose branch & bound / branch & cut solvers available for MILP and MIQP (Gurobi, CPLEX, FICO Xpress, GLPK, CBC, ...)

(more solvers/benchmarks: see http://plato.la.asu.edu/bench.html)

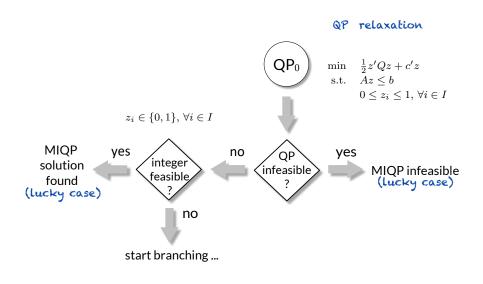
- MIQP approaches tailored to embedded hybrid MPC applications:
  - B&B + (dual) active set methods for QP
     (Leyffer, Fletcher, 1998) (Axehill, Hansson, 2006) (Bemporad, 2015) (Bemporad, Naik, 2018)
  - B&B + interior point methods: (Frick, Domahidi, Morari, 2015)
  - B&B + fast gradient projection: (Naik, Bemporad, 2017)
  - B&B + ADMM: (Stellato, Naik, Bemporad, Goulart, Boyd, 2018)
- No need to reach global optimum (see convergence proof), although performance may deteriorate

(Dakin, 1965)

• We want to solve the following MIQP

min 
$$V(z) \triangleq \frac{1}{2}z'Qz + c'z$$
  $z \in \mathbb{R}^n$   
s.t.  $Az \leq b$   $Q = Q' \succeq 0$   
 $z_i \in \{0,1\}, \forall i \in I$   $I \subseteq \{1,\dots,n\}$ 

- Branch & Bound (B&B) is the simplest (and most popular) approach to solve the problem to optimality
- Key idea:
  - for each binary variable  $z_i, i \in I$ , either set  $z_i = 0$ , or  $z_i = 1$ , or  $z_i \in [0, 1]$
  - solve the corresponding **QP relaxation** of the MIQP problem
  - use QP result to decide the next combination of fixed/relaxed variables



• Branching rule: pick the index i such that  $z_i$  is closest to  $\frac{1}{2}$  (max fractional part)

(Breu, Burdet, 1974

Solve two new QP relaxations

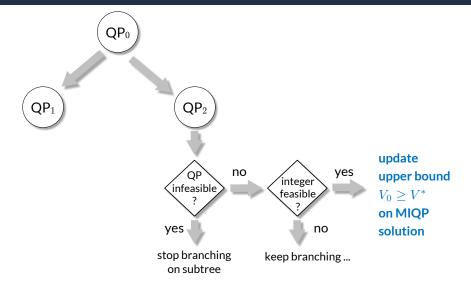
min 
$$\frac{1}{2}z'Qz + c'z$$
  $QP_1$   
s.t.  $Az \le b$   
 $z_i = 0$   
 $0 \le z_j \le 1, \forall j \in I, j \ne i$ 

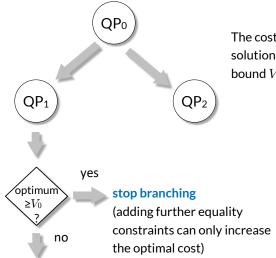
 $\bullet \ \ \mbox{Possibly exploit} \ \mbox{warm starting from} \ \mbox{QP}_0 \\ \mbox{when solving new relaxations} \ \mbox{QP}_1 \ \mbox{and} \ \mbox{QP}_2 \\$ 

$$\min \quad \frac{1}{2}z'Qz + c'z$$
s.t.  $Az \le b$ 

$$z_i = 1$$

$$0 \le z_j \le 1, \forall j \in I, j \ne i$$





The cost  $V_0$  of the best integer-feasible solution found so fare gives an upper bound  $V_0 \geq V^*$  on MIQP solution

keep branching ...

- While solving the QP relaxation, if the dual cost is available it gives a lower bound to the solution of the relaxed problem
- The QP solver can be stopped whenever the dual cost  $\geq V_0$ !

This may save a lot of computations

ullet When no further branching is possible, either the MIQP problem is recognized infeasible or the optimal solution  $z^*$  has been found

 B&B method + QP solver based on nonnegative least squares applied to solving the MIQP

$$\min_{z} \quad V(z) \triangleq \frac{1}{2}z'Qz + c'z 
\text{s.t.} \quad \ell \leq Az \leq u 
 Gz = g 
 \bar{A}_{i}z \in \{\bar{\ell}_{i}, \bar{u}_{i}\}, \ i = 1, \dots, q$$

- Binary constraints on z are a special case:  $\bar{\ell}_i=0, \bar{u}_i=1,$   $\bar{A}_i=[0\dots0\,1\,0\dots0]$
- Warm starting from parent node exploited when solving new QP relaxation
- QP solver interrupted when dual cost larger than best known upper-bound

# **SOLVING MIQP VIA NNLS**

#### Worst-case CPU time (ms) on random MIQP problems:

n	m	q	$NNLS_{LDL}$	$NNLS_{QR}$	GUROBI	CPLEX
10	5	2	2.3	1.2	1.4	8.0
10	100	2	5.7	3.3	6.1	31.4
50	25	5	4.2	6.1	14.1	30.1
50	200	10	68.8	104.4	114.6	294.1
100	50	2	4.6	10.2	37.2	69.2
100	200	15	137.5	365.7	259.8	547.8
150	100	5	15.6	49.2	157.2	260.1
150	300	20	1174.4	3970.4	1296.1	2123.9

 $egin{array}{lll} n & = & \# \ {
m variables} \ m & = & \# \ {
m inequalities} \ q & = & \# \ {
m binary \, vars} \ {
m (no \ equalities)} \end{array}$ 

Compiled Embedded MATLAB code (QP solver) + MATLAB code (B&B) CPU results measured on Macbook Pro 3GHz Intel Core i7

NNLS-LDL = recursive LDL' factorization used to solve least-square problems in QP solver NNLS-QR = recursive QR factorization used instead (numerically more robust)

# **SOLVING MIQP VIA NNLS**

### Worst-case CPU time (ms) on random purely binary QP problems:

n	m	q	$NNLS_{LDL}$	$NNLS_{QR}$	GUROBI	CPLEX
2	10	2	5.1	4.0	0.7	8.4
4	20	4	8.9	4.3	4.5	16.7
8	40	8	19.2	18.0	37.1	14.7
12	60	12	59.7	57.8	82.3	47.9
20	100	20	483.5	457.7	566.8	99.6
25	250	25	110.4	93.3	1054.4	169.4
30	150	30	1645.4	1415.8	2156.2	184.5

### Worst-case CPU time (ms) on a hybrid MPC problem

N = prediction horizon	$\overline{N}$	$NNLS_{LDL}$	$NNLS_{QR}$	GUROBI	CPLEX
	2	2.2	2.3	1.2	3.0
MIQP regularized to make	3	3.4	3.9	2.0	6.5
$Q$ strictly $\succ 0$	4	5.0	6.5	2.6	8.1
(solution difference is negligible)	5	7.6	9.8	3.7	9.0
(	6	12.3	17.7	4.3	11.0
	7	20.5	30.5	5.8	13.1
	8	28.9	47.1	7.3	17.3
	9	38.8	62.5	9.5	18.9
	10	55.4	98.2	10.9	22.4
@ 2020 A D					

## **SOLVING MIQP VIA NNLS AND PROXIMAL-POINT ITERATIONS**

Bemporad, Naik, 2018)

 Robustified approach: use NNLS + proximal-point iterations to solve QP relaxations (Bemporad, 2018)

$$z_{k+1} = \arg\min_{z} \quad \frac{1}{2}z'Qz + c'z + \frac{\epsilon}{2}||z - z_{k}||_{2}^{2}$$
  
s.t.  $\ell \le Az \le u$   
 $Gz = q$ 

• CPU time (ms) on MIQP coming from hybrid MPC (bm99 demo):

For $N=10$ :	N	prox	prox-NNLS		prox-NNLS*		GUROBI		CPLEX	
30 real vars		avg	max	avg	max	avg	max	avg	max	
$10\mathrm{binaryvars}$	2	2.0	2.6	2.0	2.6	1.6	2.0	3.1	6.0	
160 inequalities	4	5.3	8.8	3.1	6.9	3.1	3.9	8.9	15.7	
	8	29.7	71.0	8.1	43.4	7.2	13.2	15.5	80.2	
prox-NNLS* = warm	10	76.2	146.1	14.4	103.2	11.1	17.6	35.1	95.3	
start of binary vars	12	155.8	410.8	26.9	263.4	14.9	31.2	61.7	103.7	
exploited	15	484.2	1242.3	61.7	766.9	25.9	109.8	89.9	181.1	

CPU time measured on Intel Core i7-4700MQ CPU 2.40 GHz

 $\bullet \;$  Consider again the MIQP problem with Hessian  $Q=Q'\succ 0$ 

$$\min_{z} \quad V(z) \triangleq \frac{1}{2}z'Qz + c'z$$
s.t.  $\ell \leq Az \leq u$ 

$$Gz = g$$

$$\bar{A}_{i}z \in \{\bar{\ell}_{i}, \bar{u}_{i}\}, i = 1, \dots, p$$

$$w^{k} = y^{k} + \beta_{k}(y^{k} - y^{k-1})$$
 $z^{k} = -Kw^{k} - Jx$ 
 $s^{k} = \dots$ 
 $y_{i}^{k+1} = \max\{w_{i}^{k} + s_{i}^{k}, 0\}, i \in I_{\text{ineq}}$ 

Use B&B and fast gradient projection to solve dual of QP relaxation

constraint is relaxed 
$$ar{A}_iz \leq ar{u}_i \rightarrow y_i^{k+1} = \max\left\{w_i^k + s_i^k, 0\right\} \quad (y_i \geq 0)$$
 constraint is fixed  $ar{A}_iz = ar{u}_i \rightarrow y_i^{k+1} = w_i^k + s_i^k \quad (y_i \leq 0)$  constraint is ignored  $ar{A}_iz = ar{\ell}_i \rightarrow y_i^{k+1} = 0 \quad (y_i = 0)$ 

## **FAST GRADIENT PROJECTION FOR MIQP**

Naik, Bemporad, 2017)

- Same dual QP matrices at each node, preconditioning computed only once
- Warm-start exploited, dual cost used to stop QP relaxations earlier
- Criterion based on Farkas lemma to detect QP infeasibility
- Numerical results (time in ms):

n	m	p	q	miqpGPAD	GUROBI
10	100	2	2	15.6	6.56
50	25	5	3	3.44	8.74
50	150	10	5	63.22	46.25
100	50	2	5	6.22	26.24
100	200	15	5	164.06	188.42
150	100	5	5	31.26	88.13
150	200	20	5	258.80	274.06
200	50	15	6	35.08	144.38

n = # variables m = # inequality constraints p = # binary constraints q = # equality constraints

CPU time measured on Intel Core i7-4700MQ CPU 2.40 GHz

### MIQP AND ADMM

• B&B + ADMM: solve QP relaxations via ADMM

(Stellato, Naik, Bemporad, Goulart, Boyd, 2018)

$$\begin{aligned} & \text{min} & & \frac{1}{2}x'Qx + c'x \\ & \text{s.t.} & & \ell \leq Ax \leq u \\ & & & A_ix \in \{\ell_i, u_i\}, \ i \in I \end{aligned}$$

Simpler heuristic approach: only perform one set of ADMM iterations

(Takapoui, Moehle, Boyd, Bemporad, 2017)

- Iterations converge to a (local) solution
- Similar heuristic idea also applicable to fast gradient methods (Naik, Bemporad, 2017)

## HEURISTIC ADMM METHOD FOR (SUBOPTIMAL) MIQP

(Takapoui, Moehle, Boyd, Bemporad, 2017)

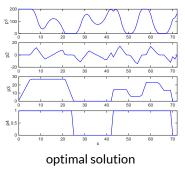
• Example: parallel hybrid electric vehicle control problem

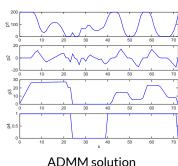


electrical power

energy stored in battery

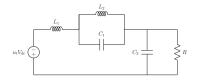
engine on/off





## HEURISTIC ADMM METHOD FOR (SUBOPTIMAL) MIQP

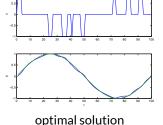
### **Example:** power converter control problem



minimize

$$\begin{array}{ll} \text{minimize} & \sum_{t=0}^T (v_{2,t}-v_{\mathrm{des}})^2 + \lambda |u_t-u_{t-1}| \\ \text{subject to} & \xi_{t+1} = G\xi_t + Hu_t \\ & \xi_0 = \xi_T \\ & u_0 = u_T \\ & u_t \in \{-1,0,1\} \end{array}$$

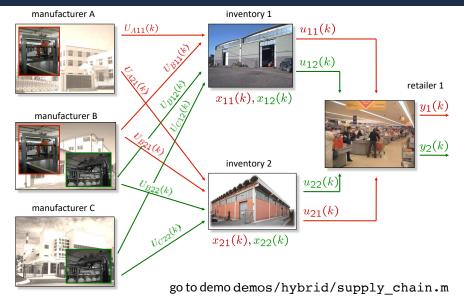
input voltage sign  $u_t$ 



ADMM solution

output voltage  $v_2$ 

## A SIMPLE EXAMPLE IN SUPPLY CHAIN MANAGEMENT



## **SUPPLY CHAIN MANAGEMENT - SYSTEM VARIABLES**

#### Continuous states:

 $x_{ij}(k)$  = amount of j hold in inventory i at time k (i=1,2,j=1,2)



#### Continuous outputs:

 $y_j(k)$  = amount of j sold at time k (j=1,2)

### Continuous inputs:

 $u_{ij}(k)$  = amount of j taken from inventory i at time k (i=1,2,j=1,2)

### • Binary inputs:

 $U_{Xij}(k)=1$  if manufacturer X produces and send j to inventory i at time k

## **SUPPLY CHAIN MANAGEMENT - CONSTRAINTS**

• Max capacity of inventory i:

$$0 \le \sum_{j=1}^{n} x_{ij} \le x_{Mi}$$

Max transportation from inventories:

$$0 \le u_{ij}(k) \le u_M$$

A product can only be sent to one inventory:

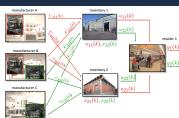
$$U_{A11}(k)$$
 and  $U_{A21}(k)$  cannot be both = 1  $U_{B11}(k)$  and  $U_{B21}(k)$  cannot be both = 1  $U_{B12}(k)$  and  $U_{B22}(k)$  cannot be both = 1  $U_{C12}(k)$  and  $U_{C22}(k)$  cannot be both = 1



• A manufacturer can only produce one type of product at one time:  $[U_{B11}(k) \text{ or } U_{B21}(k) = 1], [U_{B12}(k) \text{ or } U_{B22}(k) = 1]$  cannot be both true

## **SUPPLY CHAIN MANAGEMENT - DYNAMICS**

• Let  $P_{A1}, P_{B1}, P_{B2}, P_{C2}$  = amount of product of type 1 (2) produced by A (B, C) in one time interval



Level of inventories

$$\begin{cases} x_{11}(k+1) &= x_{11}(k) + P_{A1}U_{A11}(k) + P_{B1}U_{B11}(k) - u_{11}(k) \\ x_{12}(k+1) &= x_{12}(k) + P_{B2}U_{B12}(k) + P_{C2}U_{C12}(k) - u_{12}(k) \\ x_{21}(k+1) &= x_{21}(k) + P_{A1}U_{A21}(k) + P_{B1}U_{B21}(k) - u_{21}(k) \\ x_{22}(k+1) &= x_{22}(k) + P_{B2}U_{B22}(k) + P_{C2}U_{C22}(k) - u_{22}(k) \end{cases}$$

Retailer: all items requested from inventories are sold

$$\begin{cases} y_1 &= u_{11} + u_{21} \\ y_2 &= u_{12} + u_{22} \end{cases}$$

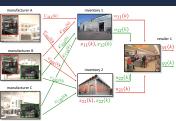
## SUPPLY CHAIN MANAGEMENT - HYSDEL CODE

```
SYSTEM supply chain{
INTERFACE (
                                                                                     manufacturer A
        STATE { REAL x11
                           [0,101;
                REAL x12
                           [0,10];
                REAL x21
                           [0,10];
                REAL x22 [0,10]; }
        INPUT { REAL u11 [0.10];
         REAL u12 [0.10];
         REAL u21 [0,10];
                                                                                                             inventory 2
         REAL u22 [0,10];
         BOOL UA11, UA21, UB11, UB12, UB21, UB22, UC12, UC22; }
                                                                                     manufacturer C
                                                                                                                       u_{21}(k)
        OUTPUT {REAL v1.v2;}
        PARAMETER { REAL PA1.PB1.PB2.PC2.xM1.xM2:}
IMPLEMENTATION (
        AUX { REAL zA11, zB11, zB12, zC12, zA21, zB21, zB22, zC22;}
                 zA11 = {IF UA11 THEN PA1 ELSE 0};
        DA {
                 zB11 = {IF UB11 THEN PB1 ELSE 0};
                 zB12 = {IF UB12 THEN PB2 ELSE 0};
                 zC12 = {IF UC12 THEN PC2 ELSE 0};
                                                                     CONTINUOUS \{x11 = x11 + zA11 + zB11 - u11\}
                 zA21 = {IF UA21 THEN PA1 ELSE 0};
                                                                                 x12 = x12 + zB12 + zC12 - u12;
                 zB21 = {IF UB21 THEN PB1 ELSE 0};
                                                                                 x21 = x21 + zA21 + zB21 - u21;
                 zB22 = {IF UB22 THEN PB2 ELSE 0};
                                                                                 x22 = x22 + zB22 + zC22 - u22; }
                 zC22 = \{IF\ UC22\ THEN\ PC2\ ELSE\ 0\}; \}
                                                                     OUTPUT (
                                                                                 y1 = u11 + u21;
                                                                                 y2 = u12 + u22; }
                                                                     MUST { ~ (UA11 & UA21) ;
                                                                              ~ (UC12 & UC22);
                                                                              ~((UB11 | UB21) & (UB12 | UB22));
                                                                              ~ (UB11 & UB21);
                                                                              ~(UB12 & UB22);
                                                                              x11+x12 \le xM1:
                                                                              x11+x12 >=0;
                                                                              x21+x22 <= xM2;
                                                                              x21+x22 >=0: }
                                                            } }
```

## **SUPPLY CHAIN MANAGEMENT - OBJECTIVES**

• Meet customer demand as much as possible:

$$y_1 \approx r_1, \quad y_2 \approx r_2$$



Minimize transportation costs

• Fulfill all constraints

## **SUPPLY CHAIN MANAGEMENT - PERFORMANCE INDEX**

$$\min \sum_{k=0}^{N-1} \frac{10(|y_{1,k}-r_1(t)|+|y_{2,k}-r_2(t)|+}{10(|y_{1,k}-r_1(t)|+|y_{2,k}-r_2(t)|+}$$
 shipping cost from inv. 1 to market 
$$\frac{4(|u_{11,k}|+|u_{12,k}|)}{2(|u_{21,k}|+|u_{22,k}|)} +$$
 shipping cost from inv. 2 to market 
$$\frac{2(|u_{21,k}|+|u_{22,k}|)}{1(|U_{A11,k}|+|U_{A21,k}|)} +$$
 cost from A to inventories 
$$\frac{1(|U_{B11,k}|+|U_{B12,k}|+U_{B21,k}|+|U_{B22,k}|)}{4(|U_{B11,k}|+|U_{B12,k}|+|U_{B21,k}|+|U_{B22,k}|)} +$$
 cost from C to inventories 
$$\frac{10(|U_{C12,k}|+|U_{C22,k}|)}{10(|U_{C12,k}|+|U_{C22,k}|)}$$

## **SUPPLY CHAIN MANAGEMENT - SIMULATION SETUP**

```
\begin{array}{c} u_{11}(k) \\ u_{12}(k) \\
```

```
>> C=hybcon(S,Q,N,limits,refs);
```

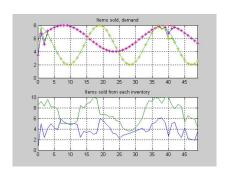
```
Bybrid controller based on MLD model S <supply_chain.hys>

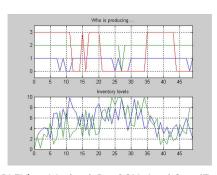
[Inf-norm]

4 state measurement(s)
2 output reference(s)
12 input reference(s)
0 state reference(s)
0 state reference(s)
44 optimization variable(s) (8 continuous z-variables

44 optimization variable(s) (8 continuous, 12 binary)
176 mixed-integer linear inequalities
sampling time = 1, MILP solver = 'glpk'
Type "struct(C)" for more details.
>>
```

## **SUPPLY CHAIN MANAGEMENT - SIMULATION RESULTS**

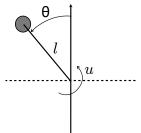




CPU time:  $\approx$  13 ms/sample (GLPK) or 9 ms (CPLEX) on Macbook Pro 3GHz Intel Core i7

# HYBRID MPC OF AN INVERTED PENDULUM

• Goal: swing the pendulum up



• Non-convex input constraint

$$u \in [-\tau_{\max}, -\tau_{\min}] \cup \{0\} \cup [\tau_{\min}, \tau_{\max}]$$

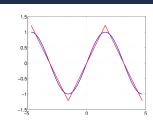
Nonlinear dynamical model

$$l^2 M\ddot{\theta} = Mgl\sin\theta - \beta\dot{\theta} + u$$

## INVERTED PENDULUM: NONLINEARITY

• Approximate  $\sin(\theta)$  as the piecewise linear function

$$\sin\theta \approx s \triangleq \left\{ \begin{array}{ll} -\alpha\theta - \gamma & \text{if} & \theta \leq -\frac{\pi}{2} \\ \alpha\theta & \text{if} & |\theta| \leq \frac{\pi}{2} \\ -\alpha\theta + \gamma & \text{if} & \theta \geq \frac{\pi}{2} \end{array} \right.$$



• Get optimal values for  $\alpha$  and  $\gamma$  by minimizing fit error

$$\min_{\alpha} \int_{0}^{\frac{\pi}{2}} (\alpha \theta - \sin(\theta))^{2} d\theta$$

$$= \frac{\theta}{2} - \frac{1}{2} \cos \theta \sin \theta - 2\alpha \sin \theta + \frac{1}{3} \alpha^{2} \theta^{3} + 2\alpha \theta \cos \theta \Big|_{0}^{\frac{\pi}{2}} = \frac{1}{24} \pi^{3} \alpha^{2} - 2\alpha + \frac{\pi}{4}$$

- Zeroing the derivative with respect to  $\alpha$  gives  $\alpha = \frac{24}{\pi^3}$
- Requiring s=0 for  $\theta=\pi$  gives  $\gamma=\frac{24}{\pi^2}$

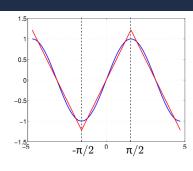
## INVERTED PENDULUM: NONLINEARITY

• Introduce the event variables

$$[\delta_3 = 1] \quad \leftrightarrow \quad [\theta \le -\frac{\pi}{2}]$$
$$[\delta_4 = 1] \quad \leftrightarrow \quad [\theta \ge \frac{\pi}{2}]$$

along with the logic constraint

$$[\delta_4=1] \to [\delta_3=0]$$



• Set 
$$s = \alpha \theta + s_3 + s_4$$
 with

$$s_3 = \left\{ egin{array}{ll} -2lpha heta - \gamma & \mbox{if } \delta_3 = 1 \\ 0 & \mbox{otherwise} \end{array} 
ight. \ s_4 = \left\{ egin{array}{ll} -2lpha heta + \gamma & \mbox{if } \delta_4 = 1 \\ 0 & \mbox{otherwise} \end{array} 
ight.$$

## INVERTED PENDULUM: NON-CONVEX CONSTRAINT

• To model the constraint  $u \in [-\tau_{\max}, -\tau_{\min}] \cup \{0\} \cup [\tau_{\min}, \tau_{\max}]$  introduce the auxiliary variable

$$\tau_A = \left\{ \begin{array}{ll} u & \text{if } -\tau_{\min} \leq u \leq \tau_{\min} \\ 0 & \text{otherwise} \end{array} \right.$$

and let  $u - \tau_A$  be the torque acting on the pendulum, with

$$u \in [-\tau_{\max}, \tau_{\max}]$$

• The input u has no effect on the dynamics for  $u \in [-\tau_{\min}, \tau_{\min}]$ . Hence, the solver will not choose values in that range if u is penalized in the MPC cost

## INVERTED PENDULUM: NON-CONVEX CONSTRAINT

• Introduce new event variables

$$\delta_2 = 0 \begin{vmatrix} \delta_1 = 1 \\ \delta_2 = 1 \end{vmatrix} \xrightarrow{\tau_{\text{min}}} u$$

$$[\delta_1 = 1] \leftrightarrow [u \le \tau_{\min}]$$
  
 $[\delta_2 = 1] \leftrightarrow [u \ge -\tau_{\min}]$ 

along with the logic constraint  $[\delta_1=0] o [\delta_2=1]$  and set

$$\tau_A = \left\{ \begin{array}{ll} u & \text{if } [\delta_1 = 1] \wedge [\delta_2 = 1] \\ 0 & \text{otherwise} \end{array} \right.$$

so that  $u - \tau_A$  is zero in for  $u \in [-\tau_{\min}, \tau_{\min}]$ 

## **INVERTED PENDULUM: DYNAMICS**

• Set  $x \triangleq \left[ \begin{smallmatrix} \theta \\ \dot{\theta} \end{smallmatrix} \right]$  ,  $y \triangleq \theta$  and transform into linear model

$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{g}{l}\alpha & -\frac{\beta}{l^2M} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ \frac{g}{l} & \frac{1}{l^2M} \end{bmatrix} \begin{bmatrix} s_3 + s_4 \\ u - \tau_A \end{bmatrix}$$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} x$$

• Discretize in time with sample time  $T_s=50~\mathrm{ms}$ 

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = A \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + B \begin{bmatrix} s_3(k) + s_4(k) \\ u(k) - \tau_A(k) \end{bmatrix}$$
$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} x(k)$$
$$A \triangleq e^{T_s A_c}, B \triangleq \int_0^{T_s} e^{tA_c} B_c dt$$

## INVERTED PENDULUM: HYSDEL MODEL

```
/* Hybrid model of a pendulum
   (C) 2012 by A. Bemporad, April 2012 */
SYSTEM hyb pendulum {
                                                 tauA = {IF d1 & d2 THEN u ELSE 0};
INTERFACE (
                                                 s3 = {IF d3 THEN -2*alpha*th-gamma ELSE 0};
                                                 s4 = {IF d4 THEN -2*alpha*th+gamma ELSE 0};
 STATE (
   REAL th [-2*pi,2*pi];
    REAL thdot [-20,201;
                                               CONTINUOUS (
 INPUT (
                                                       = a11*th+a12*thdot+b11*(s3+s4)+b12*(u-tauA);
    REAL u [-11,11];
                                                 thdot = a21*th+a22*thdot+b21*(s3+s4)+b22*(u-tauA):
 OUTPUT (
                                               OUTPUT {
    REAL y;
                                                 y = th:
 PARAMETER (
   REAL tau min,alpha,gamma;
                                               MUST (
   REAL all, al2, a21, a22, b11, b12, b21, b22;
                                                 d4->~d3:
                                                 ~d1->d2;
IMPLEMENTATION (
  AUX (
     REAL tauA.s3.s4;
     BOOL d1,d2,d3,d4;
  AD (
     d1 = u<=tau min;
     d2 = u = -tau min;
     d3 = th \le -0.5*pi;
     d4 = th >= 0.5*pi;
                                                     >> S=mld('pendulum', Ts);
```

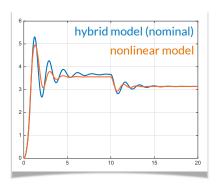
go to demo demos/hybrid/pendulum\_init.m

## INVERTED PENDULUM: MODEL VALIDATION

- Open-loop simulation from initial condition  $\theta(0)=0$ ,  $\dot{\theta}(0)=0$
- Input torque excitation

$$u(t) = \begin{cases} 2 \text{ Nm} & \text{if } 0 \le t \le 10 \text{ s} \\ 0 & \text{otherwise} \end{cases}$$

```
>> u0=2;
>> U=[2*ones(200,1);zeros(200,1)];
>> x0=[0;0];
```



## **INVERTED PENDULUM: MPC DESIGN**

• MPC cost function

$$\sum_{k=0}^{4} |y_k - r(t)| + |0.01u_k|$$

• MPC constraints  $u \in [-\tau_{\max}, \tau_{\max}]$ 

#### >> C=hybcon(S,Q,N,limits,refs);

```
>> C

Bybrid controller based on MLD model S <pendulum.hys> [Inf-norm]

2 state measurement(s)
1 output reference(s)
1 input reference(s)
0 state reference(s)
0 reference(s) on auxiliary continuous z-variables

55 optimization variable(s) (30 continuous, 25 binary)
155 mixed-integer linear inequalities
sampling time = 0.05, MILP solver = 'gurobi'

Type "struct(C)" for more details.
>>
```

```
>> refs.y=1;

>> refs.u=1;

>> Q.y=1;

>> Q.y=0.01;

>> Q.rho=Inf;

>> Q.norm=Inf;

>> N=5;

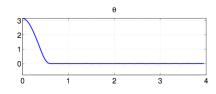
>> limits.umin=-10;

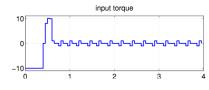
>> limits.umax=10;
```

## **INVERTED PENDULUM: CLOSED-LOOP RESULTS**

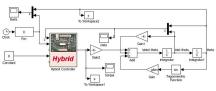
Nominal simulation

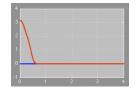
>> [X,U,D,Z,T,Y]=sim(C,S,r,x0,4);

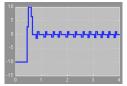




• Nonlinear simulation







#### CPU time:

51 ms per time step (GLPK)

22 ms per time step (CPLEX)

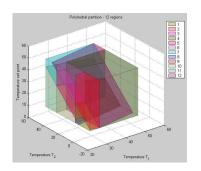
25 ms (GUROBI)

(Macbook Pro 3GHz Intel Core i7)



## **EXPLICIT HYBRID MPC**

- It is possible to write hybrid MPC laws in explicit form too!
- The explicit MPC law is still piecewise affine on polyhedra



(Bemporad, Borrelli, Morari, 2000)

(Mayne, ECC 2001)

(Belliporau, Frybrid Toolbox, 2003)

(Borrelli, Baotic, Bemporad, Morari, 2005)

Alessio, Bemporad, 2006

- The control law may be discontinuous, polyhedra may overlap
- Comparison of quadratic costs can be avoided by lifting the parameter space

(Fuchs, Axehill, Morari, 2015)

## **EXPLICIT HYBRID MPC (MLD FORMULATION)**

$$\min_{\xi} J(\xi, (x(t))) = \sum_{k=0}^{N-1} \|Qy_k\|_{\infty} + \|Ru_k\|_{\infty}$$
 
$$\begin{cases} x_{k+1} &= Ax_k + B_1u_k + B_2\delta_k + B_3z_k + B_5 \\ y_k &= Cx_k + D_1u_k + D_2\delta_k + D_3z_k + D_5 \\ E_2\delta_k + E_3z_k &\leq E_4x_k + E_1u_k + E_5 \\ x_0 &= x(t) \end{cases}$$

ullet On-line optimization: solve the problem for a given state x(t) as the <code>MILP</code>

$$\min_{\xi} \sum_{k=0}^{N-1} \epsilon_k^y + \epsilon_k^u$$
s.t.  $G\xi \le W + S(x(t))$ 

- Off-line optimization: solve the MILP in advance for all states x(t)
- multiparametric Mixed-Integer Linear Program (mp-MILP)

## **MULTIPARAMETRIC MILP**

• Consider the mp-MILP

$$\min_{\xi_c, \xi_d} \quad f'_c \xi_c + f'_d \xi_d$$
s.t.  $G_c \xi_c + G_d \xi_d \le W + S(x)$ 

$$\xi_c \in \mathbb{R}^{nc}$$

$$\xi_d \in \{0, 1\}^{n_d}$$

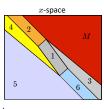
$$x \in \mathbb{R}^m$$

A mp-MILP can be solved by alternating MILPs and mp-LPs

(Dua, Pistikopoulos, 1999)

- The multiparametric solution  $\xi^*(x)$  is **PWA** (but possibly discontinuous)
- The MPC controller is piecewise affine in x=x(t)

$$u(x) = \left\{ \begin{array}{ccc} F_1x + g_1 & \text{if} & H_1x \leq K_1 \\ & \vdots & & \vdots \\ F_Mx + g_M & \text{if} & H_Mx \leq K_M \end{array} \right.$$



(More generally, the parameter vector  $\boldsymbol{x}$  includes states and reference signals)

#### **EXPLICIT HYBRID MPC (PWA FORMULATION)**

Consider the MPC formulation using a PWA prediction model

$$\min_{\xi} J(\xi, x(t)) = \sum_{k=0}^{N-1} \|Qy_k\|_{\infty} + \|Ru_k\|_{\infty}$$
 subject to 
$$\begin{cases} x_{k+1} &= A_{i(k)}x_k + B_{i(k)}u_k + f_{i(k)} \\ y_k &= C_{i(k)}x_k + D_{i(k)}u_k + g_{i(k)} \\ & i(k) \text{ such that } H_{i(k)}x_k + W_{i(k)}u_k \leq K_{i(k)} \\ x_0 &= x(t) \end{cases}$$

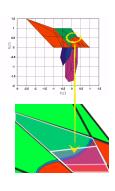
- Method #1: The explicit solution can be obtained by using a combination of dynamic programming (DP) and mpLP (Borrelli, Baotic, Bemporad, Morari, 2005)
- Clearly the explicit hybrid MPC law is again piecewise affine, as PWA systems MLD systems

# **EXPLICIT HYBRID MPC (PWA FORMULATION)**

Method #2: (Bemporad, Hybrid Toolbox, 2003)

(Alessio, Bemporad, 2006) (Mayne, ECC 2001) (Mayne, Rakovic, 2002)

- 1 Use backwards (=DP) reachability analysis for enumerating all feasible mode sequences  $I=\{i(0),i(1),\ldots,i(N)\}$
- 2 For each fixed sequence I, solve the explicit finite-time optimal control problem for the corresponding linear time-varying system (mpQP or mpLP)
- 3a Case of  $1/\infty$ -norms or convex PWA costs: Compare value functions and split regions
- 3b Case of quadratic costs: the partition may not be fully polyhedral, better keep overlapping polyhedra and compare on-line quadratic cost functions when overlaps are detected
- Comparison of quadratic costs can be avoided by lifting the parameter space (Fuchs, Axehill, Morari, 2015)



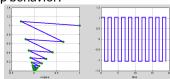
## **HYBRID MPC EXAMPLE - EXPLICIT VERSION**

PWA system:

$$\left\{ \begin{array}{rcl} x(t+1) & = & 0.8 \left[ \begin{array}{ccc} \cos\alpha(t) & -\sin\alpha(t) \\ \sin\alpha(t) & \cos\alpha(t) \end{array} \right] x(t) + \left[ \begin{array}{c} 0 \\ 1 \end{array} \right] u(t) \\ y(t) & = & \left[ \begin{array}{ccc} 0 & 1 \end{array} \right] x(t) \\ \alpha(t) & = & \left\{ \begin{array}{ccc} \frac{\pi}{3} & \text{if} & \left[ \begin{array}{ccc} 1 & 0 \end{array} \right] x(t) \geq 0 \\ -\frac{\pi}{3} & \text{if} & \left[ \begin{array}{ccc} 1 & 0 \end{array} \right] x(t) < 0 \end{array} \right. \end{array} \right.$$

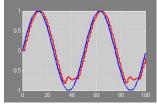
subject to 
$$-1 \le u(t) \le 1$$

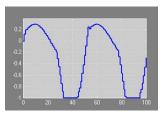
- MPC objective:  $\min \sum_{k=1}^{\infty} |y_k r(t)|$
- Open-loop behavior:



go to demo demos/hybrid/bm99sim.m

#### Closed-loop MPC



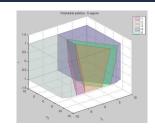


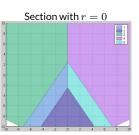
### HYBRID MPC EXAMPLE - EXPLICIT VERSION

$$u(x,r) = \left\{ \begin{array}{ll} \begin{bmatrix} 0.6928 - 0.4 \ 1 \end{bmatrix} \begin{bmatrix} x \\ r \end{bmatrix} & \text{if} & \begin{bmatrix} 0.6928 & -0.4 \ -0.4 & -0.6928 & 0 \ 0 & -1 & 0 \ -0.6928 & 0.4 & 1 \ 0 & 0 & 0 & 0 \ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x \\ r \end{bmatrix} \leq \begin{bmatrix} \frac{1}{10} \\ 10 \\ 10 \\ 10 \\ 10 \\ 10 \end{bmatrix} \begin{bmatrix} x \\ x \end{bmatrix} \leq \begin{bmatrix} \frac{1}{10} \\ 10 \\ 11 \\ 1e-006 \end{bmatrix} \end{bmatrix} \\ 1 & \text{if} & \begin{bmatrix} -0.6928 & 0.4 & -1 \\ 0.6928 & 0.4 & -1 \\ 0.6928 & -1 & 0 \end{bmatrix} \begin{bmatrix} x \\ x \end{bmatrix} \leq \begin{bmatrix} -1 \\ -1 \\ 10 \end{bmatrix} \\ 1 & 0 \end{bmatrix} \\ -1 & \text{if} & \begin{bmatrix} -0.4 & -0.6928 & 0 \\ 0.6928 & -1.4 & 0 \\ 0.6928 & -1.4 & 0 \\ 0.6928 & -1.4 & 0 \end{bmatrix} \begin{bmatrix} x \\ x \end{bmatrix} \leq \begin{bmatrix} 10 \\ -1 \\ 10 \end{bmatrix} \\ 1 & 0 \end{bmatrix} \\ -1 & \text{if} & \begin{bmatrix} -0.4 & -0.6928 & 0 \\ 0.6928 & -0.4 & 1 \\$$

goto to /demos/hybrid/bm99sim.m

Offline CPU time = 1.51 s (Macbook Pro 3GHz Intel Core i7)

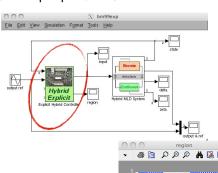


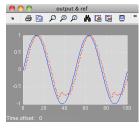


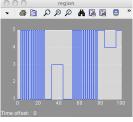
PWA law ≡ MPC law!

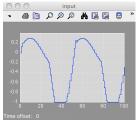
## HYBRID MPC EXAMPLE - EXPLICIT VERSION

#### • Closed-loop explicit MPC







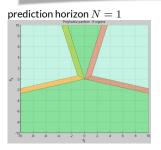


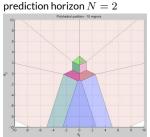
## **EXPLICIT PWA REGULATOR**

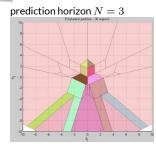
#### • MPC problem:

min 
$$10||x_N||_{\infty} + \sum_{k=0}^{N-1} 10||x_k||_{\infty} + ||u_k||_{\infty}$$
  
s.t. 
$$\begin{cases}
-1 & \leq u_k \leq 1, k = 0, \dots, N-1 \\
-10 & \leq x_k \leq 10, k = 1, \dots, N
\end{cases}$$

$$\begin{array}{rcl} Q & = & \left[ \begin{smallmatrix} 10 & 0 \\ 0 & 10 \end{smallmatrix} \right] \\ R & = & 1 \end{array}$$







go to demos/hybrid/bm99benchmark.m

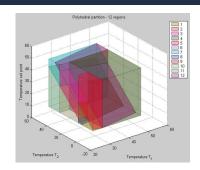
### EXPLICIT HYBRID MPC — TEMPERATURE CONTROL

#### >> E=expcon(C,range,options);

```
>> E

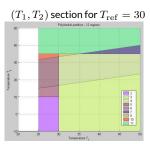
Explicit controller (based on hybrid controller C)
3 parameter(s)
1 input(s)
12 partition(s)
sampling time = 0.5

The controller is for hybrid systems (tracking)
This is a state-feedback controller.
Type "struct(E)" for more details.
>>
```

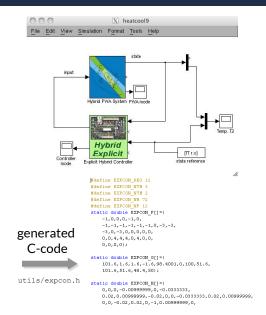


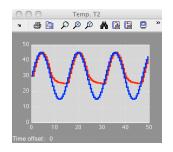
#### 384 numbers to store in memory

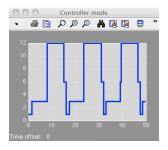
$$\min \sum_{k=0}^{2} \|x_{2k} - r(t)\|_{\infty}$$
 s.t. 
$$\begin{cases} x_{1k} \ge 25, \ k = 1, 2 \\ \text{hybrid model} \end{cases}$$



## EXPLICIT HYBRID MPC — TEMPERATURE CONTROL







### IMPLEMENTATION ASPECTS OF HYBRID MPC

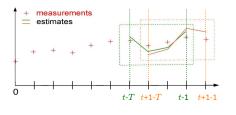
- Alternatives:
  - 1. solve MIP on-line
  - 2. evaluate a PWA function (explicit solution)
- Small problems (short horizon N=1,2, one or two inputs, 4-6 binary vars): explicit PWA control law is preferable
  - CPU time to evaluate the control law is shorter than by MIP
  - control code is simpler (no complex solver must be included in the control software!)
  - more insight in controller behavior
- Medium/large problems (longer horizon, many inputs and binary variables): on-line MIP is preferable



## STATE ESTIMATION / FAULT DETECTION

(Bemporad, Mignone, Morari, 1999) (Ferrari-Trecate, Mignone, Morari, 2002)

- Goal: estimate the state of a hybrid system from past I/O measurements
- Moving horizon estimation based on MLD models solves the problem



MLD model augmented

by

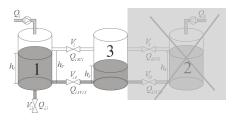
- state disturbance  $\xi \in \mathbb{R}^n$
- output disturbance  $\zeta \in \mathbb{R}^p$
- At each time t get the estimate  $\hat{x}(t)$  by solving the MIQP

$$\begin{aligned} \min_{\hat{x}(t-T|t)} & & \sum_{k=0}^{T} \|\hat{y}(t-k|t) - y(t-k)\|_2^2 + \dots \\ \text{s.t.} & & \text{constraints on } \hat{x}(t-T+k|t), \hat{y}(t-T+k|t) \end{aligned}$$

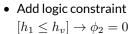
• For fault detection also include unknown binary disturbances  $\phi \in \{0,1\}^{n_f}$ 

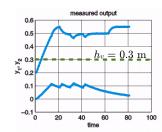
## MHE EXAMPLE - THREE TANK SYSTEM

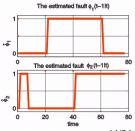
- $\bullet \ \ {\sf Can \, only \, measure \, tank \, levels} \, h_1, h_2$
- The system has two faults:
  - $\phi_1$ : leak in tank 1 between 20 s  $\leq t \leq$  60 s
  - $\phi_2$ : valve  $V_1$  blocked for  $t \geq 40 \, \mathrm{s}$



(COSY benchmark problem)

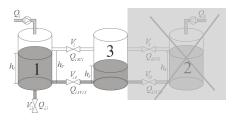




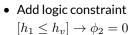


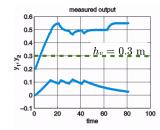
## MHE EXAMPLE - THREE TANK SYSTEM

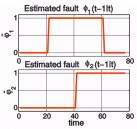
- $\bullet \ \ {\sf Can\,only\,measure\,tank\,levels}\, h_1, h_2$
- The system has two faults:
  - $\phi_1$ : leak in tank 1 between 20 s  $\leq t \leq$  60 s
  - $\phi_2$ : valve  $V_1$  blocked for  $t \geq$  40 s



(COSY benchmark problem)







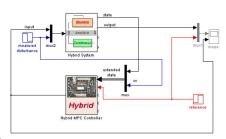


### **MEASURED DISTURBANCES**

- ullet A measured disturbance v(t) enters the hybrid system
- Augment the hybrid prediction model with the constant state

$$\begin{array}{rcl} x_{k+1}^v & = & x_k^v \\ x_0^v & = & v(t) \end{array}$$

HYSDEL model



• Same trick applies to linear MPC

go to demo demos/hybrid/hyb\_meas\_dist.m

# REFERENCE TRACKING

Hybrid MPC formulation for reference tracking

$$\begin{aligned} & \min & & \sum_{k=0}^{N-1} \|W^y(y_{k+1} - r(t))\|_2^2 + \|W^{\Delta u} \Delta u_k\|_2^2 \\ & \text{s.t.} & & \text{hybrid dynamics} \\ & & \Delta u_k = u_k - u_{k-1}, \ k = 0, \dots, N-1, \ u_{-1} = u(t-1) \\ & & u_{\min} \leq u_k \leq u_{\max}, \ k = 0, \dots, N-1 \\ & & y_{\min} \leq y_k \leq y_{\max}, \ k = 1, \dots, N \\ & & \Delta u_{\min} \leq \Delta u_k \leq \Delta u_{\max}, \ k = 0, \dots, N-1 \end{aligned}$$

The resulting optimization problem is the MIQP

$$\min_{\xi} \quad J(\xi, x(t)) = \frac{1}{2} \xi' H \xi + [x'(t) \, r'(t) \, u'(t-1)] F \xi$$
s.t.  $G \xi \leq W + S \begin{bmatrix} x(t) \\ r(t) \\ u(t-1) \end{bmatrix}$ 

$$\xi = \begin{bmatrix} \frac{\Delta u_0}{\delta_0} \\ \frac{\delta_0}{z_0} \\ \vdots \\ \frac{\Delta u_{N-1}}{\delta_{N-1}} \\ \frac{\delta_{N-1}}{z_{N-1}} \end{bmatrix}$$

• Same trick as in linear MPC

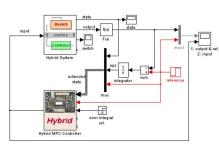
## INTEGRAL ACTION

Augment hybrid prediction model with integrals of output tracking errors

$$\epsilon_{k+1} = \epsilon_k + T_s(r(t) - y_k)$$

- ullet Treat set point r(t) as a measured disturbance (= constant state)
- Add weight on  $\epsilon_k$  in cost function
- HYSDEL model:

Same trick applies to linear MPC



go to demo demos/hybrid/hyb\_integral\_action.m

#### TIME-VARYING CONSTRAINTS

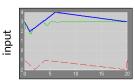
• Consider the time-varying constraint

$$u(t) \le u_{\max}(t)$$



Augment the hybrid prediction model with the constant state

$$\begin{array}{rcl} x_{k+1}^u & = & x_k^u \\ x_0^u & = & u_{\max}(t) \end{array}$$



and output  $y_k^u = x^u(k) - u_k$ , subject to the constraint  $y_k^u \geq 0, k = 0, 1, \dots, N$ 

• Same trick applies to linear MPC

go to demo demos/linear/varbounds.m

• Alternative: in HYSDEL simply impose MUST {u <= xu;}

- $\bullet \ \ \text{Measured disturbance} \ v(t) \ \text{is known} \ M \ \text{steps in advance} \\$
- Augment the model with the following buffer dynamics

$$\left\{ \begin{array}{lll} x_{k+1}^{M-1} & = & x_k^{M-2} \\ x_{k+1}^{M-2} & = & x_k^{M-3} \\ & \vdots & & \text{with initial condition} \\ x_{k+1}^1 & = & x_k^0 \\ x_{k+1}^0 & = & x_k^0 \end{array} \right. \qquad \left\{ \begin{array}{lll} x_0^{M-1} & = & v(t) \\ x_0^{M-2} & = & v(t+1) \\ \vdots & = & \vdots \\ x_0^1 & = & v(t+M-2) \\ x_0^0 & = & v(t+M-1) \end{array} \right.$$

• The predicted state  $x^{M-1}$  of the buffer is

$$x_k^{M-1} = \begin{cases} v(t+k) & k = 0, \dots, M-1 \\ v(t+M-1) & k = M, \dots, N-1 \end{cases}$$

- ullet Preview of reference signal r(t+k) can be dealt with in a similar way
- Same trick applies to linear MPC

#### DELAYS - METHOD #1

Hybrid model with delays

$$x(t+1) = Ax(t) + B_1u(t-\tau) + B_2\delta(t) + B_3z(t) + B_5$$
  

$$E_2\delta(t) + E_3z(t) \le E_1u(t-\tau) + E_4x(t) + E_5$$

• Map delays to poles in z = 0:

$$x_{k}(t) \triangleq u(t-k) \Rightarrow x_{k}(t+1) = x_{k-1}(t), \ k = 1, \dots, \tau$$

$$\begin{bmatrix} x^{(t+1)} \\ x_{\tau}(t+1) \\ x_{\tau-1}(t+1) \\ \vdots \\ x_{1}(t+1) \end{bmatrix} = \begin{bmatrix} A & B_{1} & 0 & 0 & \dots & 0 \\ 0 & 0 & I_{m} & 0 & \dots & 0 \\ 0 & 0 & I_{m} & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & 0 & 0 & \dots & 0 \end{bmatrix} \begin{bmatrix} x^{(t)} \\ x_{\tau}(t) \\ x_{\tau-1}(t) \\ \vdots \\ x_{1}(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ \vdots \\ I_{m} \end{bmatrix} u(t) + \begin{bmatrix} B_{2} \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} \delta(t) + \begin{bmatrix} B_{3} \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} z(t) + \begin{bmatrix} B_{5} \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix}$$

- Apply MPC to the extended MLD system
- Same trick as in linear MPC

#### DELAYS - METHOD #2

• Delay-free model:

$$\bar{x}(t) \triangleq x(t+\tau) \Longrightarrow \begin{cases} \bar{x}(t+1) = A\bar{x}(t) + B_1 u(t) + B_2 \bar{\delta}(t) + B_3 \bar{z}(t) + B_5 \\ E_2 \bar{\delta}(t) + E_3 \bar{z}(t) \leq E_1 u(t) + E_4 \bar{x}(t) + E_5 \end{cases}$$

- $\bullet \;\; \text{Design MPC}$  for delay-free model,  $u(t) = f_{\mathrm{MPC}}(\bar{x}(t))$
- Compute the predicted state

$$\bar{x}(t) = \hat{x}(t+\tau) = A^{\tau}x(t) + \sum_{j=1}^{\tau-1} A^{j}(B_{1} \underbrace{u(t-1-j)}_{\text{past inputs!}} + B_{2}\bar{\delta}(t+j) + B_{3}\bar{z}(t+j) + B_{5})$$

where  $\bar{\delta}(t+j)$ ,  $\bar{z}(t+j)$  are obtained from MLD inequalities or by simulation

• Compute the MPC control move  $u(t) = f_{\mathrm{MPC}}(\hat{x}(t+\tau))$ 

#### **CHOICE CONSTRAINTS**

- Logic constraint: make one or more choices out of a set of alternatives:
  - make at most one choice:  $\delta_1 + \delta_2 + \delta_3 \leq 1$
  - make at least two choices:  $\delta_1 + \delta_2 + \delta_3 \geq 2$
  - exclusive or constraint:  $\delta_1 + \delta_2 + \delta_3 = 1$
- More generally:

$$\sum_{i=1}^N \delta_i \leq m \qquad \text{choose at most } m \text{ items out of } N$$
 
$$\sum_{i=1}^N \delta_i = m \qquad \text{choose exactly } m \text{ items out of } N$$
 
$$\sum_{i=1}^N \delta_i \geq m \qquad \text{choose at least } m \text{ items out of } N$$

#### "NO-GOOD" CONSTRAINTS

- Given a binary vector  $\bar{\delta} \in \{0,1\}^n$  we want to impose the constraint

$$\delta 
eq \bar{\delta}$$

- This may be useful for example to extract different solutions from an MIP that has multiple optima
- The "no-good" condition can be expressed equivalently as

$$\sum_{i \in T} \delta_i - \sum_{i \in F} \delta_i \le -1 + \sum_{i=1}^n \bar{\delta}_i \qquad F = \{i : \bar{\delta}_i = 0\}$$
$$T = \{i : \bar{\delta}_i = 1\}$$

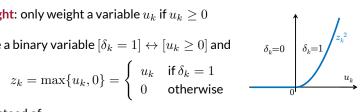
or

$$\sum_{i=1}^{n} (2\bar{\delta}_i - 1)\delta_i \le \sum_{i=1}^{n} \bar{\delta}_i - 1$$

#### ASYMMETRIC WEIGHTS

- **Asymmetric weight:** only weight a variable  $u_k$  if  $u_k \geq 0$
- We can introduce a binary variable  $[\delta_k = 1] \leftrightarrow [u_k \ge 0]$  and

$$z_k = \max\{u_k, 0\} = \begin{cases} u_k & \text{if } \delta_k = 1\\ 0 & \text{otherwise} \end{cases}$$



then weight  $z_k$  instead of  $u_k$ 

**Better solution**: only introduce auxiliary variable  $z_k$  and optimize

min 
$$(...) + \sum_{k=0}^{N-1} z_k^2$$
  
s.t.  $z_k \ge u_k$   
 $z_k \ge 0$ 

- Similar approach when  $\|\cdot\|_{\infty}$  or  $\|\cdot\|_{1}$  are used as penalties
- Same trick applies to linear MPC

#### **GENERAL REMARKS ABOUT MIP MODELING**

- The complexity of solving a mixed-integer program largely depends on the number of integer (binary) variables involved in the problem
- Hence, when creating a hybrid model one has to

Be thrifty with binary variables!

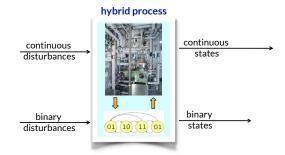
- Adding logical constraints usually helps
- Generally speaking

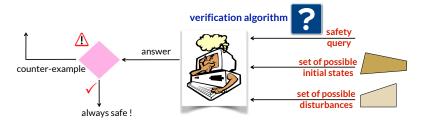
modeling is an art





### **HYBRID VERIFICATION PROBLEM**





#### **VERIFICATION ALGORITHM #1**

- Query: Is the target set  $X_f$  reachable after N steps from some initial state  $x_0 \in X_0$  for some input  $u_0, \ldots, u_{N-1} \in U$ ?
- The query can be answered by solving the mixed-integer feasibility test

$$\min_{\xi} 0$$
s.t.  $x_{k+1} = Ax_k + B_1 u_k + B_2 \delta_k + B_3 z_k + B_5$ 

$$E_2 \delta_k + E_3 z_k \le E_4 x_k + E_1 u_k + E_5$$

$$S_u u_k \le T_u \quad (u_k \in U), \quad k = 0, 1, \dots, N - 1$$

$$S_0 x_0 \le T_0 \quad (x_0 \in X_0)$$

$$S_f x_N \le T_f \quad (x_N \in X_f)$$

with respect to 
$$\xi=[x_0,\ldots,x_N,u_0,\ldots,u_{N-1},\delta_0,\ldots,\delta_{N-1},z_0,\ldots,z_{N-1}]$$

- Other approaches:
  - Exploit structure and use polyhedral computation (Torrisi, 2003)
  - Use abstractions (LPs) + SAT solvers (Giorgetti, Pappas, Bemporad)

# **VERIFICATION EXAMPLE**

• MLD model: room temperature control system



Set of unsafe states:

$$X_f = \left\{ \begin{bmatrix} T_1 \\ T_2 \end{bmatrix} : 10 \le T_1, T_2 \le 15 \right\}$$

Set of initial states:

$$X_0 = \left\{ \begin{bmatrix} T_1 \\ T_2 \end{bmatrix} : 35 \le T_1, T_2 \le 40 \right\}$$

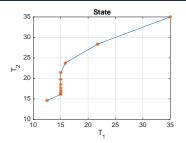
• Set of possible inputs:

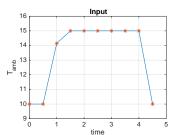
$$U = \{T_{\rm amb} : 10 \le T_{\rm amb} \le 30\}$$

• Time horizon: N=10 steps

>> [flag,x0,U]=reach(S,N,Xf,X0,umin,umax);

## **VERIFICATION EXAMPLE**





$$U = \{T_{\rm amb} : 10 \le T_{\rm amb} \le 30\}$$

```
>> umin=20;
>> reach(S,N,Xf,X0,umin,umax);
Hybrid Toolbox v.1.4.2 [February 2, 2020]
Elapsed time is 0.023282 seconds.

Xf is not reachable from X0
>>
```

$$U = \{T_{\rm amb} : 20 \le T_{\rm amb} \le 30\}$$

### **VERIFICATION ALGORITHM #2**

- Query: Is the target set  $X_f$  reachable within N steps from some initial state  $x_0 \in X_0$  for some input  $u_0, \dots, u_{N-1} \in U$ ?
- Augment the MLD system to register the entrance of the target (unsafe) set  $X_f = \{x: A_f x \leq b_f\}$ :
  - Add a new variable  $\delta_k^f$  , with  $[\delta_k^f=1] o [A_f x_{k+1} \le b_f]$

$$\underbrace{ A_f(Ax_k + B_1u_k + B_2\delta_k + B_3z_k + B_5) \leq b_f + M(1 - \delta_k^f) }_{\text{big-N}}$$

- Add the constraint  $\sum_{k=0}^{N-1} \delta_k^f \geq 1$  (i.e.,  $x_k \in X_f$  for at least one k)
- Solve MILP feasibility test

# A MORE COMPLEX VERIFICATION EXAMPLE

• States  $x_1, x_2, x_3 \in \mathbb{R}, x_4, x_5 \in \{0, 1\}$ , inputs  $u_1, u_2 \in \mathbb{R}, u_3 \in \{0, 1\}$ 

$$\begin{aligned} & [\delta_1 = 1] \leftrightarrow [x_1 \leq 0] \\ \bullet & \text{Events:} & [\delta_2 = 1] \leftrightarrow [x_2 \geq 1] \\ & [\delta_3 = 1] \leftrightarrow [x_3 - x_2 \leq 1] \end{aligned}$$

#### Switched dynamics

$$\begin{array}{lll} x_1(k+1) & = & \left\{ \begin{array}{l} 0.1x_1(k) + 0.5x_2(k) & \text{if } (\delta_1(k) \wedge \delta_2(k)) \vee x_4(k) \text{ true} \\ -0.3x_3(k) - x_1(k) + u_1(k) & \text{otherwise} \end{array} \right. \\ x_2(k+1) & = & \left\{ \begin{array}{l} -0.8x_1(k) + 0.7x_3(k) - u_1(k) - u_2(k) & \text{if } \delta_3(k) \vee x_5(k) \text{ true} \\ -0.7x_1(k) - 2x_2(k) & \text{otherwise} \end{array} \right. \\ x_3(k+1) & = & \left\{ \begin{array}{l} -0.1x_3(k) + u_2(k) & \text{if } (\delta_3(k) \wedge x_5(k)) \vee (\delta_1(k) \wedge x_4(k)) \text{ true} \\ x_3(k) - 0.5x_1(k) - 2u_1(k) & \text{otherwise} \end{array} \right. \end{array}$$

Automaton

$$x_4(k+1) = \delta_1(k) \wedge x_4(k) x_5(k+1) = ((x_4(k) \vee x_5(k)) \wedge (\delta_1(k) \vee \delta_2(k)) \vee (\delta_3(k) \wedge u_3(k))$$

#### A MORE COMPLEX VERIFICATION EXAMPLE

• Query: Verify if it possible that, starting from the set  $X_0$ 

$$X_0 = \{x : -0.1 \le x_1, x_3 \le 0.1, x_2 = 0.1, x_4, x_5 \in \{0, 1\}\}\$$

the state  $x(k) \in X_f$ 

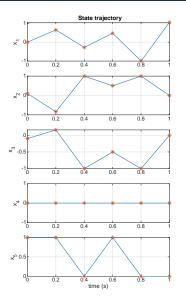
$$X_f = \{x: -1 \le x_1, x_3 \le 1, 0.5 \le x_2 \le 1, x_4, x_5 \in \{0, 1\}\}\$$

at some  $k \leq N$ , N = 5, under the restriction that  $\forall k \leq N$ 

$$x_3(k)+x_2(k)\leq 0$$
  $\delta_1(k)\vee\delta_2(k)\vee x_5(k)=$  true  $\neg x_4(k)\vee x_5(k)=$  true

go to demo demos/hybrid/reachtest.m

# A MORE COMPLEX VERIFICATION EXAMPLE



```
>> reachtest
Hybrid Toolbox v.1.4.2 [February 2, 2020]
Elapsed time is 0.038049 seconds.
>> reachtime
reachtime =

3
4
>>
```

The set  $X_f$  is reached by x(k) at time steps k = 3, 4