#### **AERO 632: Design of Advance Flight Control System**

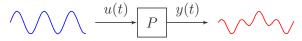
Norms for Signals and Systems

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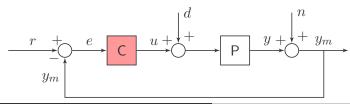
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## Norms for Signals

#### **Signals**



- We consider signals mapping  $(-\infty, \infty) \mapsto \mathbb{R}$
- Piecewise continuous
- lacksquare A signal may be zero for t < 0
- We worry about size of signal
- Helps specify performance
- Signal size  $\iff$  signal norm



#### 00000000 Norms

Signals

A norm must have the following 4 properties

- ||u|| > 0
- $\|u\| = 0 \iff u = 0$
- $\|a\boldsymbol{u}\| = |a|\|\boldsymbol{u}\|, \, \forall a \in \mathbb{R}$
- $lacksquare \|u+v\| \leq \|u\| + \|v\|$  triangle inequality

For  $\boldsymbol{u} \in \mathbb{R}^n$  and p > 1,

$$\|\boldsymbol{u}\|_p := (|u_1|^p + \dots + |u_n|^p)^{1/p}$$

Special case,

$$\|\boldsymbol{u}\|_{\infty} := \max_{i} |u_i|$$

## **Norms of Signals**

#### $\mathcal{L}_1$ Norm

The 1-norm of a signal u(t) is the integral of its absolute value:

$$||u(t)||_1 := \int_{-\infty}^{\infty} |u(t)| dt$$

#### $\mathcal{L}_2$ Norm

The 2-norm of a signal u(t) is

$$\|u(t)\|_2 := \left(\int_{-\infty}^\infty u(t)^2 dt
ight)^{1/2}$$
 associated with energy of signal

#### $\mathcal{L}_{\infty}$ Norm

The  $\infty$ -norm of a signal u(t) is the least upper bound of its absolute value:

$$||u(t)||_{\infty} := \sup_{t} |u(t)|$$

## **Power Signals**

The average power of u(t) is the average over time of its instantaneous power:

$$\lim_{T \to \infty} \frac{1}{2T} \int_{-T}^{T} u^2(t) dt$$

- $\blacksquare$  if limit exists, u(t) is called a power signal
- average power is then

$$\mathbf{pow}(u) := \left(\lim_{T \to \infty} \frac{1}{2T} \int_{-T}^{T} u^2(t) dt\right)^{1/2}$$

- **pow** $(\cdot)$  is not a norm
  - non zero signals can have  $pow(\cdot) = 0$

## **Vector Signals**

Signals 000000000

For 
$$\boldsymbol{u}(t):(-\infty,\infty)\mapsto\mathbb{R}^n$$
 and  $p>1$ 

$$\|u(t)\|_p := \left(\int_{-\infty}^{\infty} \sum_{i=1}^{n} |u_i(t)|^p dt\right)^{1/p}$$

Signals 000000000

Does finiteness of one norm imply finiteness of another?

$$\blacksquare \ \|u\|_2 < \infty \implies \mathbf{pow}(u) = 0$$

We have

$$\frac{1}{2T} \int_{-T}^{T} u^2(t)dt \le \frac{1}{2T} \int_{-\infty}^{\infty} u^2(t)dt = \frac{1}{2T} ||u||_2.$$

Right hand side tends to zero as  $T \to \infty$ 

## Finiteness of Norms (contd.)

■ If u is a power signal and  $||u||_{\infty} < \infty$ , then  $\mathbf{pow}(u) \leq ||u||_{\infty}$ .

We have

$$\frac{1}{2T} \int_{-T}^{T} u^2(t)dt \le ||u||_{\infty} \frac{1}{2T} \int_{-T}^{T} dt = ||u||_{\infty}$$

Let  $T \to \infty$ .

## Finiteness of Norms (contd.)

■ If  $||u||_1 < \infty$  and  $||u||_\infty < \infty$  then  $||u||_2 < \infty$ 

We have

$$\int_{-\infty}^{\infty} u^{2}(t)dt = \int_{-\infty}^{\infty} |u(t)| \cdot |u(t)|dt$$

$$\leq ||u||_{\infty} \int_{-\infty}^{\infty} |u(t)|dt$$

$$= ||u||_{\infty} ||u||_{1}$$

$$\leq \infty$$

# Norms for Systems

### System

$$u(t)$$
  $G$   $y(t)$ 

Systems 000000000000

#### We consider

- Linear
- Time invariant
- Causal
- Finite dimensional

#### In time domain

- $\blacksquare$  if u(t) is the input to the system and
- $\blacksquare$  y(t) is the output

#### System has the form

$$egin{array}{lcl} y &=& G*u ext{ convolution} \ &=& \int_{-\infty}^{\infty} G(t- au) u( au) d au & \mathcal{L}^{-1}\left\{\hat{G}(s)
ight\} := G(t) \end{array}$$

#### Causal

- A system is causal when the effect does not anticipate the cause; or zero input produces zero output
- Its output and internal states only depend on current and previous input values
- Physical systems are causal

contd.

#### Acausal

- A system whose output is nonzero when the past and present input signal is zero is said to be anticipative
- A system whose state and output depend also on input values from the future, besides the past or current input values, is called acausal
- Acausal systems can only exist as digital filters (digital signal processing).

contd.

#### Anti-Causal

- A system whose output depends only on future input values is anti-causal
- Derivative of a signal is anti-causal.

contd.

- Zeros are anticipative
- Poles are causal

Systems

- $\blacksquare$  Overall behavior depends on m and n.
- $\blacksquare$  Causal: n > m, strictly proper
- $\blacksquare$  Causal: n=m, still causal, but there is instantaneous transfer of information from input to output
- $\blacksquare$  Acausal: n < m

### **Example**

- $\blacksquare$  System  $G_1(s) = s$
- Input  $u(t) = \sin(\omega t)$ ,  $U(s) = \frac{\omega}{s^2 + \omega^2}$
- $y_1(t) = \mathcal{L}^{-1} \left\{ G_1(s)U(s) \right\} = \mathcal{L}^{-1} \left\{ \frac{s\omega}{s^2 + \omega^2} \right\} = \omega \cos(\omega t)$ , or

$$u(t)=\sin(\omega t)$$
  $y_1(t)=\omega\sin(\omega t+\pi/2)$   $=\omega u(t+\frac{\pi}{2\omega})$  output leads input, anticipatory

## **Example**

contd.

■ System 
$$G_2(s) = \frac{1}{s}$$

$$\blacksquare$$
 Input  $u(t)=\sin(\omega t)$ ,  $U(s)=\frac{\omega}{s^2+\omega^2}$ 

$$y_2(t) = \mathcal{L}^{-1} \left\{ G_2(s) U(s) \right\} = \mathcal{L}^{-1} \left\{ \frac{1}{s} \frac{\omega}{s^2 + \omega^2} \right\} = \frac{1}{\omega} - \frac{\cos(\omega t)}{\omega}, \text{ or }$$

$$u(t) = \sin(\omega t)$$
 
$$y_2(t) = \frac{1}{\omega} + \frac{\sin(\omega t - \pi/2)}{\omega}$$
 
$$= \frac{1}{\omega} + \frac{u(t - \frac{\pi}{2\omega})}{\omega}$$
 output lags input, causal

(contd.)

#### Causality means

$$G(t) = 0$$
 for  $t < 0$ 

- ullet  $\hat{G}(s)$  is stable if it is analytic in the closed RHP residue theorem
- lacksquare proper if  $\hat{G}(j\infty)$  is finite deg of den  $\geq$  deg of num
- $\blacksquare$  strictly proper if  $\hat{G}(i\infty) = 0$  deg of den > deg of num
- **biproper**  $\hat{G}$  and  $\hat{G}^{-1}$  are both proper

## Norms of G

Definitions for SISO Systems

 $\mathcal{L}_2$  Norm

$$\|\hat{G}\|_2 := \left(\frac{1}{2\pi} \int_{-\infty}^{\infty} |\hat{G}(j\omega)|^2 d\omega\right)^{1/2}$$

 $\mathcal{L}_{\infty}$  Norm

$$\|\hat{G}\|_{\infty} := \sup_{\omega} |\hat{G}(j\omega)|$$
 peak value of  $|\hat{G}(j\omega)|$ 

#### Parseval's Theorem

If  $G(j\omega)$  is stable

$$\|\hat{G}\|_{2} = \left(\frac{1}{2\pi} \int_{-\infty}^{\infty} |\hat{G}(j\omega)|^{2} d\omega\right)^{1/2} = \left(\int_{-\infty}^{\infty} |G(t)|^{2} dt\right)^{1/2}.$$

## **Important Properties of System Norms**

#### Submultiplicative Property of $\infty$ -norm

$$\|\hat{G}\hat{H}\|_{\infty} \le \|\hat{G}\|_{\infty} \|\hat{H}\|_{\infty}$$

## **Important Properties of System Norms (contd.)**

#### Lemma 1

 $\|\hat{G}\|_2$  is finite iff  $\hat{G}$  is strictly proper and has no poles on the imaginary axis.

Proof: Look at transfer function of the type

$$\hat{G}(s) = \frac{(s+z_1)(s+z_2)\cdots(s+z_m)}{(s+p_1)(s+p_2)\cdots(s+p_n)}, n > m.$$

Argue area under  $|\hat{G}(j\omega)|^2$  is finite.

Or apply residue theorem

$$\left(\frac{1}{2\pi} \int_{-\infty}^{\infty} |\hat{G}(j\omega)|^2 d\omega\right)^{1/2} = \frac{1}{2\pi j} \oint_{\mathsf{LHP}} \hat{G}(-s) \hat{G}(s) ds.$$

## **Important Properties of System Norms (contd.)**

#### Lemma 2

 $\|\hat{G}\|_{\infty}$  is finite iff  $\hat{G}$  is **proper** and has no poles on the imaginary axis.

Proof: Look at transfer function of the type

$$\hat{G}(s) = \frac{(s+z_1)(s+z_2)\cdots(s+z_m)}{(s+p_1)(s+p_2)\cdots(s+p_n)}, n >= m.$$

Argue  $\sup_{\omega} |\hat{G}(j\omega)|$  is finite.

# Signal Spaces

- Describe performance in terms of norms of certain signals of interest
- Understand which norm is suitable
  - difference from control system performance perspective
- We will learn Hardy spaces  $\mathcal{H}_2$  and  $\mathcal{H}_{\infty}$

Signal Spaces

 measures of worst possible performance for many classes of input signals

#### Also called linear space

Elements  $u, v, w \in \mathcal{V} \subseteq \mathbb{C}^n$  (or  $\mathbb{R}^n$ ) satisfy the following 8 axioms

Associativity of addition

$$u + (v + w) = (u + v) + w$$

■ Commutativity of addition

$$u + v = v + u$$

■ Identity element of addition

$$0 + v = v, \forall v \in \mathcal{V}$$

Inverse element of addition

for every 
$$v \in \mathcal{V}, \exists -v \in \mathcal{V}: v + (-v) = 0$$

contd.

Compatibility of scalar multiplication

$$\alpha(\beta u) = (\alpha \beta)u$$

Identity of multiplication

$$1v = v$$

Distributivity of scalar multiplication wrt vector addition

$$\alpha(u+v) = \alpha u + \alpha v$$

Distributivity of scalar multiplication wrt field addition

$$(\alpha + \beta)u = \alpha u + \beta u$$

## **Normed Space**

- $\blacksquare$  Let  $\mathcal V$  be a vector space over  $\mathbb C$  or  $\mathbb R$
- Let  $\|\cdot\|$  be defined over  $\mathcal{V}$
- $\blacksquare$  Then  $\mathcal{V}$  is a normed space

#### Example 1

A vector space  $\mathbb{C}^n$  with any vector p-norm,  $\|\cdot\|$ , for 1 .

#### Example 2

Space C[a,b] of all bounded continuous functions becomes a norm space if

$$||f||_{\infty} := \sup_{t \in [a,b]} |f(t)|$$

is defined

■ A sequence  $\{x_n\}$  in a normed space  $\mathcal{V}$  is Cauchy sequence, if

$$||x_n - x_m|| \to 0 \text{ as } n, m \to 0.$$

■ A sequence  $\{x_n\}$  is said to converge to  $x \in \mathcal{V}$ , written  $x_n \to x$ . if

Signal Spaces

$$||x_n - x|| \to 0.$$

- $\blacksquare$  A normed space  $\mathcal{V}$  is said to be complete if every Cauchy sequence in  $\mathcal{V}$  converges in  $\mathcal{V}$ .
- A complete normed space is called a Banach space.

$$l_p[0,\infty)$$
 spaces for  $1 \leq p < \infty$ 

For each  $1 \leq p < \infty$ ,  $l_p[0,\infty)$  consists of all sequence  $x = (x_0, x_1, \cdots)$  such that

$$\sum_{i=0}^{\infty} |x_i|^p < \infty.$$

The associate norm is defined as

$$||x||_p := \left(\sum_{i=0}^{\infty} |x_i|^p\right)^{1/p}.$$

 $l_{\infty}[0,\infty)$  space

$$l_{\infty}[0,\infty)$$
 consists of all bounded sequence  $x=(x_0,x_1,\cdots)$ .

Signal Spaces

The  $l_{\infty}$  norm is defined as

$$||x||_{\infty} := \sup_{i} |x_i|.$$

 $\mathcal{L}_p(I)$  spaces for 1

For each  $1 \leq p \leq \infty$ ,  $\mathcal{L}_p(I)$  consists of all Lebesgue measurable functions x(t) defined on an interval  $I \subset \mathbb{R}$  such that

$$||x||_p := \left(\int_I |x(t)|^p \mu(dt)\right)^{1/p} < \infty, \text{ for } 1 \le p < \infty,$$

and

$$||x||_{\infty} := \operatorname{ess\,sup}_{t \in I} |x(t)|.$$

We will study  $\mathcal{L}_2(-\infty,0]$ ,  $\mathcal{L}_2[0,\infty)$ , and  $\mathcal{L}_2(-\infty,\infty)$  spaces in detail.

C[a,b] space

Consists of all continuous functions on the real interval [a, b] with the norm

Signal Spaces

$$||x||_{\infty} := \sup_{t \in [a,b]} |x(t)|.$$

Recall the inner product of vectors in Euclidean space  $\mathbb{C}^n$ :

Signal Spaces

$$\langle x, y \rangle := x^* y = \sum_{i=1}^n \bar{x}_i y_i, \forall x, y \in \mathbb{C}^n.$$

#### **Important Metric Notions & Geometric Properties**

- length, distance, angle
- energy

We can generalize beyond Euclidean space!

#### Genralization

Let  $\mathcal{V}$  be a vector space over  $\mathbb{C}$ . An inner product on  $\mathcal{V}$  is a complex value function

Signal Spaces

$$\langle \cdot, \cdot \rangle : \mathcal{V} \times \mathcal{V} \mapsto \mathbb{C}$$

such that for any  $\alpha, \beta \in \mathbb{C}$  and  $x, y, z \in \mathcal{V}$ 

- 1.  $\langle x, \alpha y + \beta z \rangle = \alpha \langle x, y \rangle + \beta \langle x, z \rangle$
- 2.  $\langle x, y \rangle = \langle y, x \rangle$
- 3.  $\langle x, x \rangle > 0$ , if  $x \neq 0$

A vector space with an inner product is called an inner product space.

## **Inner-Product Space**

Introduces Geometry

The inner-product defined as

$$\langle x, y \rangle := x^* y = \sum_{i=1}^n \bar{x}_i y_i, \forall x, y \in \mathbb{C}^n,$$

induces a norm

$$||x|| := \sqrt{\langle x, x \rangle}$$

#### **Geometric Properties**

 $\blacksquare$  Distance between vectors x, y

$$d(x,y) := ||x - y||.$$

■ Two vectors x, y in an inner-product space  $\mathcal{V}$  are orthogonal if

$$\langle x, y \rangle = 0.$$

■ Orthogonal to a set  $S \subset V$  if  $\langle x, y \rangle = 0$ ,  $\forall y \in S$ .

# **Inner-Product Space**

Important Properties

Let  $\mathcal{V}$  be an inner product space and let  $x, y \in \mathcal{V}$ . Then

Signal Spaces

- 1.  $|\langle x,y\rangle| \leq ||x|| ||y||$  Cauchy-Schwarz inequality.
  - Equality holds iff  $x = \alpha y$  for some constant  $\alpha$  or y = 0.
- 2.  $||x + y||^2 + ||x y||^2 = 2||x|| + 2||y|| Parallelogram law$
- 3.  $||x + y||^2 = ||x||^2 + ||y||^2$  if  $x \perp y$

## **Hilbert Space**

- A complete inner-product space with norm induced by its inner product
- Restricted class of Banach space
  - ► Banach space only norm
  - ► Hilbert space inner-product, which allows orthonormal bases, unitary operators, etc.
- Existence and uniqueness of best approximations in closed subspaces - very useful.

#### **Finite Dimensional Examples**

- $\blacksquare$   $\mathbb{C}^n$  with usual inner product
- $\blacksquare$   $\mathbb{C}^{n\times m}$  with inner-product

$$\langle A, B \rangle := \mathbf{tr} \left[ A^* B \right] = \sum_{i=1}^n \sum_{j=1}^m a_{ij} b_{ij}, \ \forall A, B \in \mathbb{C}^{n,m}$$

## Hilbert Space

$$l_2(-\infty,\infty)$$

Set of all real or complex square summable sequences

$$x = \{ \cdots, x_{-2}, x_{-1}, x_0, x_1, x_2, \cdots \},\$$

i.e.

$$\sum_{i=-\infty}^{\infty} |x_i|^2 < \infty,$$

with inner product defined as

$$\langle x, y \rangle := \sum_{i=-\infty}^{\infty} \bar{x}_i y_i,$$

for  $x, y \in l_2(-\infty, \infty)$ .  $x_i$  can be scalar, vector or matrix with norm

$$\langle x, y \rangle := \sum_{i=-\infty}^{\infty} \mathbf{tr} \left[ \bar{x}_i y_i \right].$$

## **Hilbert Space**

 $\mathcal{L}_2(I)$  for  $I \subset \mathbb{R}$ 

- $\blacksquare$   $\mathcal{L}_2(I)$  square integrable and Lebesgue measurable functions defined over interval  $I \subset \mathbb{R}$
- with inner product

$$\langle f, g \rangle := \int_{I} f(t)^{*} g(t) dt,$$

for  $f, g \in \mathcal{L}_2(I)$ .

For vector or matrix valued functions, the inner product is defined as

$$\langle f, g \rangle := \int_I \mathbf{tr} \left[ f(t)^* g(t) \right] dt.$$

# **Hardy Spaces**

 $\mathcal{L}_2(j\mathbb{R})$  Space

 $\mathcal{L}_2(j\mathbb{R})$  Space –  $\mathcal{L}_2$  is a Hilbert space of matrix-valued (or scalar-valued) complex function F on  $j\mathbb{R}$  such that

Signal Spaces

$$\int_{-\infty}^{\infty} \mathbf{tr} \left[ F^*(j\omega) F(j\omega) \right] d\omega < \infty,$$

with inner product

$$\langle F, G \rangle := \frac{1}{2\pi} \int_{-\infty}^{\infty} \mathbf{tr} \left[ F^*(j\omega) G(j\omega) \right] d\omega,$$

for  $F, G \in \mathcal{L}_2(j\mathbb{R})$ .

#### $\mathcal{RL}_2(j\mathbb{R})$

All real rational strictly proper transfer matrices with no poles on the imaginary axis.

# **Hardy Spaces**

 $\mathcal{H}_2$  Space

 $\mathcal{H}_2$  Space – Closed subspace of  $\mathcal{L}_2(j\mathbb{R})$  with matrix functions F(s)analytic in Re(s) > 0.

Norm is defined as

$$||F||_2^2 = \frac{1}{2\pi} \int_{-\infty}^{\infty} \mathbf{tr} \left[ F^*(j\omega) F(j\omega) \right] d\omega.$$

■ Computation of  $\mathcal{H}_2$  norm is same as  $\mathcal{L}_2(j\mathbb{R})$ 

#### $\mathcal{RH}_2$

Real rational subspace of  $\mathcal{H}_2$ , which consists of all strictly proper and real stable transfer matrices, is denoted by  $\mathcal{RH}_2$ .

# **Hardy Spaces**

 $\mathcal{L}_{\infty}(j\mathbb{R})$  Space

 $\mathcal{L}_{\infty}(j\mathbb{R})$  Space – is a Banach space of matrix-valued (or scalar-valued) functions that are (essentially) bounded on  $j\mathbb{R}$ , with norm

Signal Spaces

$$||F||_{\infty} := \operatorname{ess\,sup}_{\omega \in \mathbb{R}} \bar{\sigma} \left[ F(j\omega) \right].$$

#### $\mathcal{RL}_{\infty}(j\mathbb{R})$

All proper and real rational transfer matrices with no poles on the imaginary axis.

# **Hardy Space**

 $\mathcal{H}_{\infty}$  Space

 $\mathcal{H}_{\infty}$  Space – is a closed subspace of  $\mathcal{L}_{\infty}$  space with functions that are analytic and bounded in the open right-half plane.

Signal Spaces

The  $\mathcal{H}_{\infty}$  norm is defined as

$$\|F\|_{\infty} := \sup_{\operatorname{Re}(s)>0} \bar{\sigma}\left[F(s)\right] = \sup_{\omega \in \mathbb{R}} \bar{\sigma}\left[F(j\omega)\right].$$

 $\mathcal{RH}_{\infty}$ 

Real rational subspace of  $\mathcal{H}_{\infty}$ , which consists of all proper and real rational stable transfer matrices.

# Input-Output Relationships

## How big is output?

$$u(t)$$
  $G$   $y(t)$ 

Interesting Question: If we know how big the input is, how big is the output going to be?

## **Bounded Input Bounded Output**



- Given  $|u(t)| \le u_{\max} < \infty$ , what can we say about  $\max |y(t)|$ ?
- Recall

$$Y(s) = G(s)U(s) \implies y(t) = \int_{-\infty}^{\infty} h(\tau)u(t-\tau)d\tau.$$

Therefore.

$$|y(t)| = \left| \int hud\tau \right| \le \int |h||u|d\tau \le u_{\max} \int |h(\tau)|d\tau.$$

#### **Bound on output** y(t)

$$\max_{t} |y(t)| \le u_{\max} \int |h(\tau)| d\tau$$

## **Bounded Input Bounded Output (contd.)**



$$\max_{t} |y(t)| \le u_{\max} \int |h(\tau)| d\tau$$

#### **BIBO Stability**

If and only if

$$\int |h(\tau)|d\tau < \infty.$$

(LTI): **Re**  $p_i < 0 \implies BIBO$  stability

 $|y(t)| < y_{\text{max}}$  is not enough!

Output norms for two candidate input signals

	$u(t) = \delta(t)$	$u(t) = \sin(\omega t)$
$  y  _2$	$\ \hat{G}(j\omega)\ _2$	$\infty$
$  y  _{\infty}$	$\ \hat{G}(j\omega)\ _{\infty}$	$ \hat{G}(j\omega) $
pow(y)	0	$\frac{1}{\sqrt{2}} \hat{G}(j\omega) $

## Input-Output Norms (contd.)

#### **System Gains**

- Input signal size is given
- What is the output signal size?

	$  u  _{2}$	$  u  _{\infty}$	pow(u)
$  y  _2$	$\ \hat{G}(j\omega)\ _{\infty}$	$\infty$	$\infty$
$  y  _{\infty}$	$\ \hat{G}(j\omega)\ _2$	$  G(t)  _1$	$\infty$
pow(y)	0	$\leq \ \hat{G}(j\omega)\ _{\infty}$	$\ \hat{G}(j\omega)\ _{\infty}$

∞-norm of system is pretty useful

# Computation of Norms

## **Computation of Norms**

Best computed in state-space realization of system

**State Space Model:** General MIMO LTI system modeled as

$$\dot{x} = Ax + Bu,$$
  
$$y = Cx + Du,$$

where  $x \in \mathbb{R}^n$ ,  $u \in \mathbb{R}^m$ ,  $y \in \mathbb{R}^p$ .

#### Transfer Function

$$\hat{G}(s) = D + C(sI - A)^{-1}B$$
 strictly proper when  $D = 0$ 

#### **Impulse Response**

$$G(t) = \mathcal{L}^{-1} \{ C(sI - A)^{-1}B \} = Ce^{tA}B.$$

# $\mathcal{H}_2$ Norm

MIMO Systems

$$\begin{split} \|\hat{G}(j\omega)\|_2^2 &= \frac{1}{2\pi} \int_{-\infty}^{\infty} \mathbf{tr} \left[ \hat{G}^*(j\omega) \hat{G}(j\omega) \right] \text{ for matrix transfer function} \\ &= \|G(t)\|_2^2 \text{ Parseval} \\ &= \int_0^{\infty} \mathbf{tr} \left[ Ce^{tA}BB^Te^{tA^T}C^T \right] dt \\ &= \mathbf{tr} \left[ C\underbrace{\left( \int_0^{\infty} e^{tA}BB^Te^{tA^T}dt \right)}_{L_c} C^T \right] L_c = \text{controllability Gramian} \end{split}$$

 $= \mathbf{tr} \left[ C L_c C^T \right]$ 

MIMO Systems

#### For any matrix M

$$\begin{aligned} \mathbf{tr}\left[M^*M\right] &= \mathbf{tr}\left[MM^*\right] \\ &\implies \|\hat{G}(j\omega)\|_2^2 &= \mathbf{tr}\left[B^T\underbrace{\left(\int_0^\infty e^{tA^T}C^TCe^{tA}dt\right)}_{L_o}B\right] \\ &= \mathbf{tr}\left[B^TL_oB\right] \ L_o = \text{observability Gramian} \end{aligned}$$

### $\mathcal{H}_2$ Norm of $\hat{G}(j\omega)$

$$\|\hat{G}(j\omega)\|_2^2 = \mathbf{tr} \left[ C L_c C^T \right] = \mathbf{tr} \left[ B^T L_o B \right].$$

## $\mathcal{L}_2$ Norm

How to determine  $L_c$  and  $L_o$ ?

They are solutions of the following equation

$$AL_c + L_c A^T + BB^T = 0, A^T L_o + L_o A + C^T C = 0.$$

#### Proof:

From definition,

$$L_o := \int_0^\infty e^{tA^T} C^T C e^{tA} dt$$

Instead.

$$L_o(t) = \int_0^t e^{\tau A^T} C^T C e^{\tau A} d\tau.$$

Change of variable  $\tau := t - \xi$ ,

$$L_o(t) = \int_0^t e^{(t-\xi)A^T} C^T C e^{(t-\xi)A} d\xi.$$

## $\mathcal{L}_2$ Norm

How to determine  $L_c$  and  $L_a$ ?

Take time-derivative.

$$\frac{dL_o(t)}{dt} = \frac{d}{dt} \int_0^t e^{(t-\xi)A^T} C^T C e^{(t-\xi)A} d\xi.$$

Differentiation under integral sign:

$$\frac{d}{dx} \int_{a(x)}^{b(x)} f(x,y) dy$$

$$= f(x,b(x)) \frac{db(x)}{dx} - f(x,a(x)) \frac{da(x)}{dx} + \int_{a(x)}^{b(x)} \frac{\partial f(x,y)}{\partial x} dy.$$

### $\mathcal{L}_2$ Norm

How to determine  $L_c$  and  $L_a$ ?

$$\implies \frac{dL_o(t)}{dt} = A^T \left( \int_0^t e^{(t-\xi)A^T} C^T C e^{(t-\xi)A} d\xi \right) + \left( \int_0^t e^{(t-\xi)A^T} C^T C e^{(t-\xi)A} d\xi \right) A + C^T C e^{(t-\xi)A} d\xi$$

Or

$$\frac{dL_o(t)}{dt} = A^T L_o + L_o A + C^T C.$$

 $L_o(t)$  is smooth, therefore

$$\lim_{t \to \infty} L_o(t) = L_o \implies \lim_{t \to \infty} \frac{dL_o(t)}{dt} = 0.$$

Thefore,  $L_o$  satisfies

$$A^T L_0 + L_0 A + C^T C = 0.$$



#### Recall

$$\|\hat{G}(j\omega)\|_{\infty} := \operatorname{ess\,sup}_{\omega} \bar{\sigma} \left[\hat{G}(j\omega)\right]$$

- Requires a search
- Estimate can be determined using bisection algorithm
  - Set up a grid of frequency points

$$\{\omega_1,\cdots\omega_N\}.$$

• Estimate of  $\|\hat{G}(j\omega)\|_{\infty}$  is then.

$$\max_{1 \le k \le N} \bar{\sigma} \left[ \hat{G}(j\omega_k) \right].$$

lacksquare Or read it from the plot of  $\bar{\sigma} \left| \hat{G}(j\omega) \right|$ .

## $\mathcal{RL}_{\infty}$ Norm

Bisection Algorithm

#### Lemma

Let  $\gamma>0$  and

$$G(s) = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \in \mathcal{RL}_{\infty}.$$

Then  $\|\hat{G}(j\omega)\|_{\infty}<\gamma$  iff  $\bar{\sigma}\left[D\right]<\gamma$  and H has no eigen values on the imaginary axis where

$$H := \begin{bmatrix} A + BR^{-1}D^TC & BR^{-1}B^T \\ -C^T(I + DR^{-1}D^T)C & -(A + BR^{-1}D^TC)^T \end{bmatrix},$$

and

$$R := \gamma^2 I - D^T D.$$

#### **Proof:**

See Robust and Optimal Control, K. Zhou, J.C. Doyle, K. Glover, Ch. 4, pg 115.

#### Bisection Algorithm

1. Select an upper bound  $\gamma_u$  and lower bound  $\gamma_l$  such that

$$\gamma_l \le \|\hat{G}(j\omega)\|_{\infty} \le \gamma_u.$$

- 2. If  $(\gamma_u \gamma_l)/\gamma_l \le \epsilon$  STOP;  $\|\hat{G}(j\omega)\|_{\infty} = (\gamma_u + \gamma_l)/2$ .
- 3. Else  $\gamma = (\gamma_u + \gamma_l)/2$
- 4. Test if  $\|\hat{G}(j\omega)\|_{\infty} \leq \gamma$  by calculating eigen values of H for given  $\gamma$
- 5. If H has an eigen value on  $j\mathbb{R}$ ,  $\gamma_l = \gamma$ , else  $\gamma_u = \gamma$
- 6. Goto step 2.