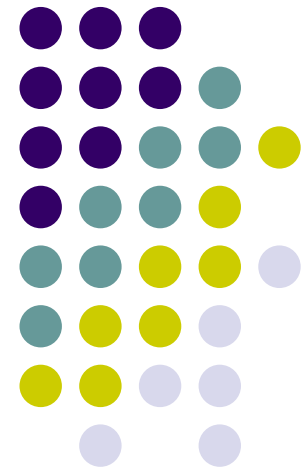


# Chapter 7

## Wavelets and Multiresolution Processing

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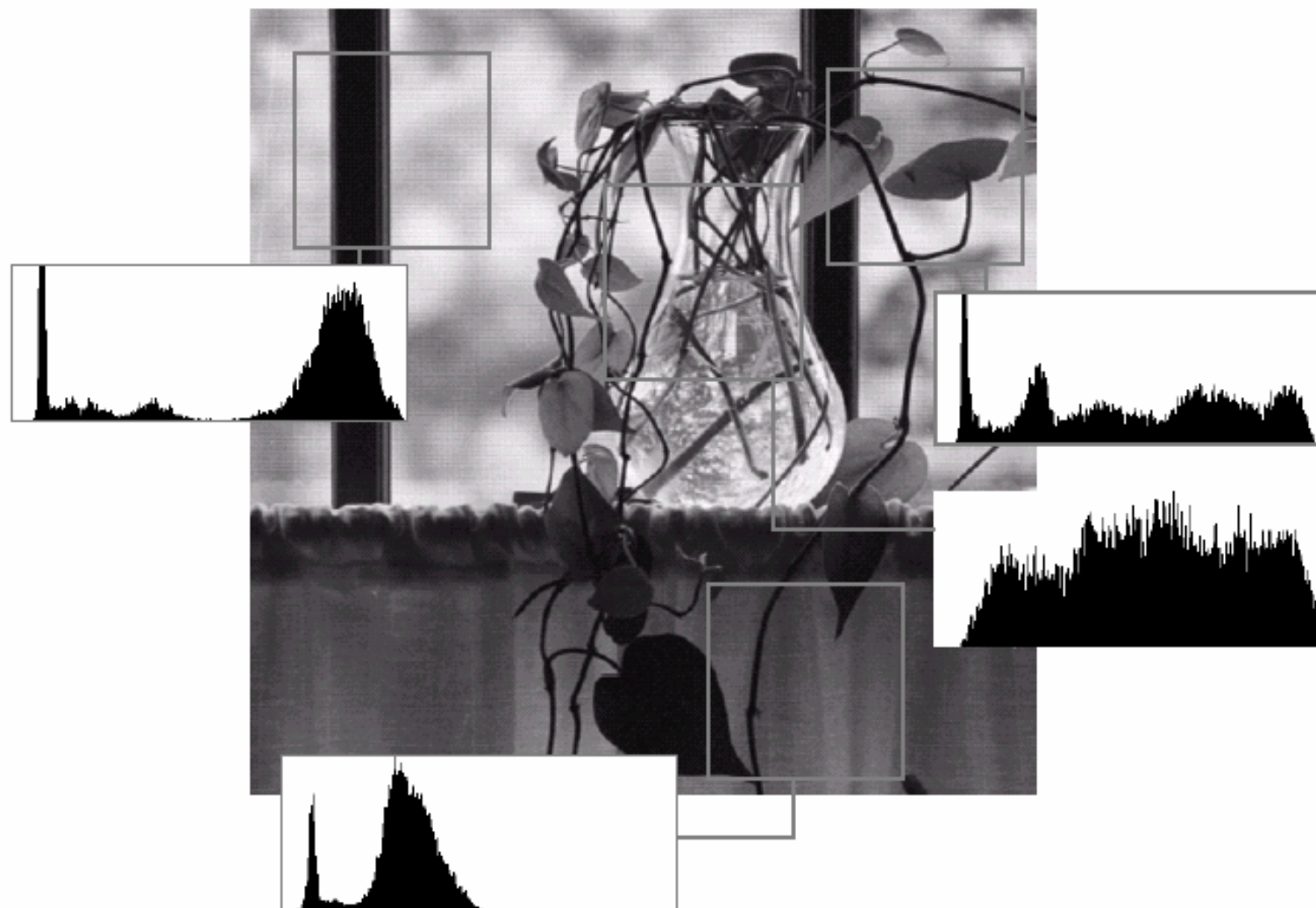




- Background
- Multiresolution Expansions
- Wavelet Transforms in One Dimension
- Wavelet Transforms in Two Dimensions



**FIGURE 7.1** A natural image and its local histogram variations.



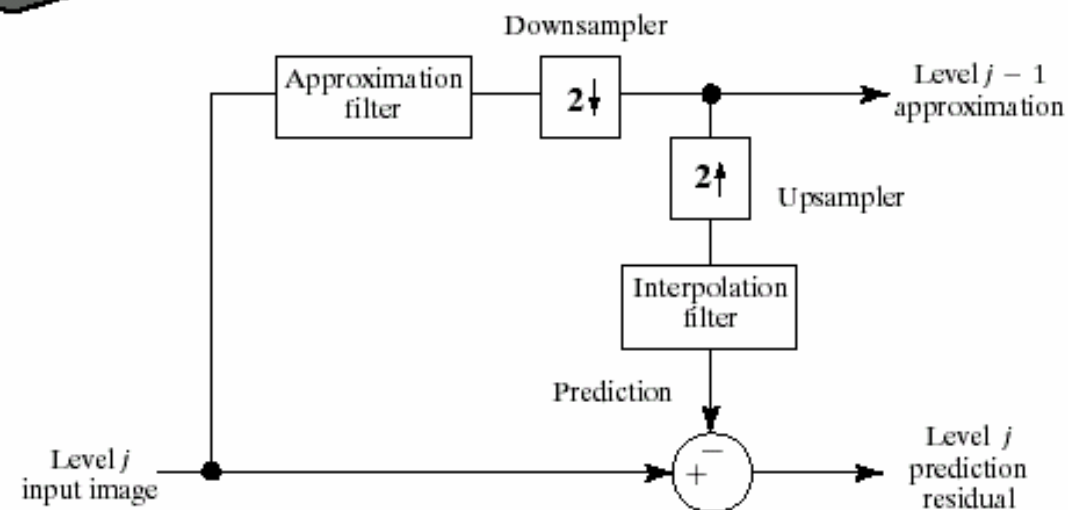
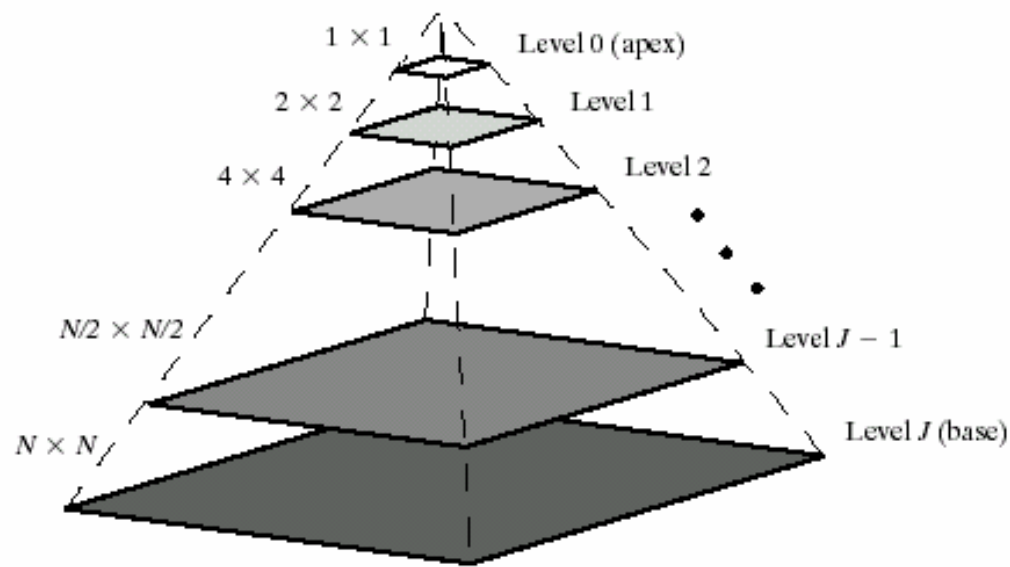


- Image Pyramids
- Subband Coding
- The Haar Transform



- The total number of elements in a P+1 level pyramid for P>0 is

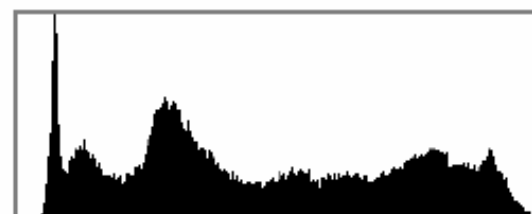
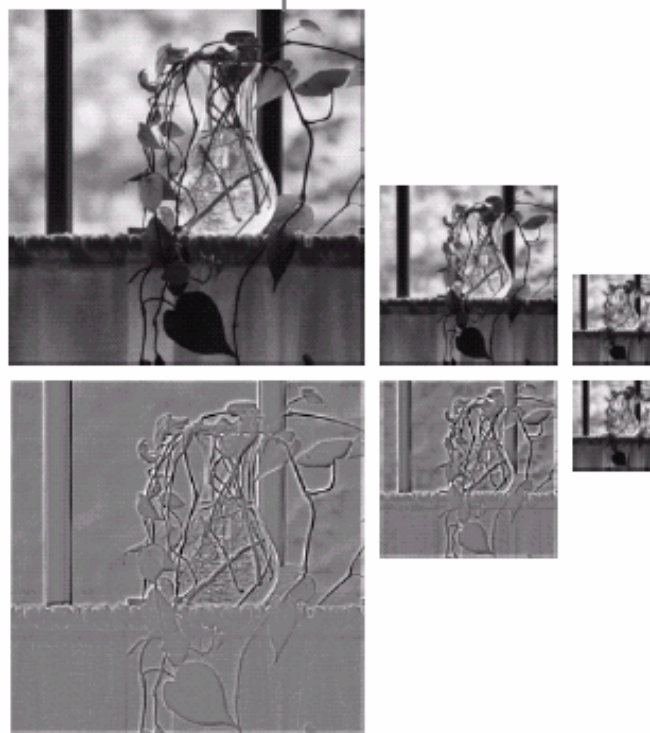
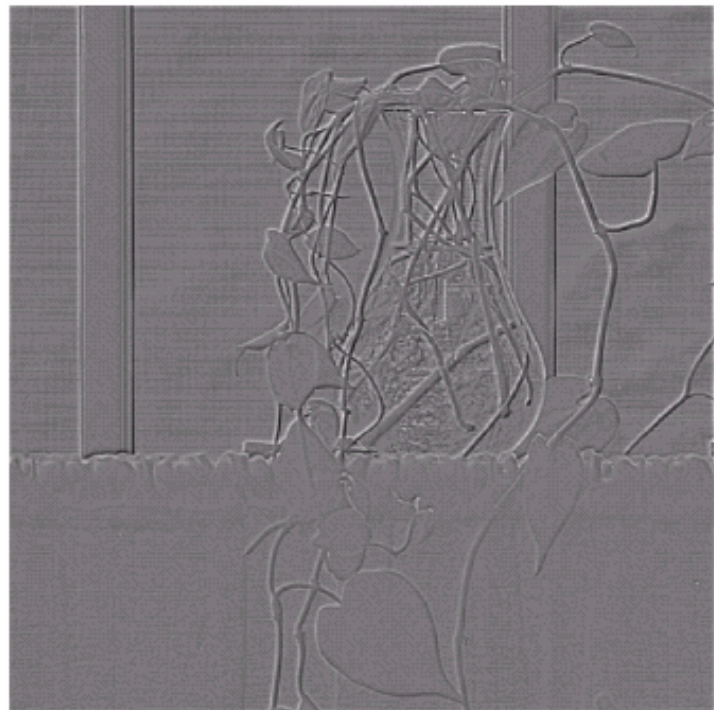
$$N^2 \left( 1 + \frac{1}{(4)^1} + \frac{1}{(4)^2} + \dots + \frac{1}{(4)^P} \right) \leq \frac{4}{3} N^2$$



a

b

**FIGURE 7.2** (a) A pyramidal image structure and (b) system block diagram for creating it.



a  
b

**FIGURE 7.3** Two image pyramids and their statistics: (a) a Gaussian (approximation) pyramid and (b) a Laplacian (prediction residual) pyramid.



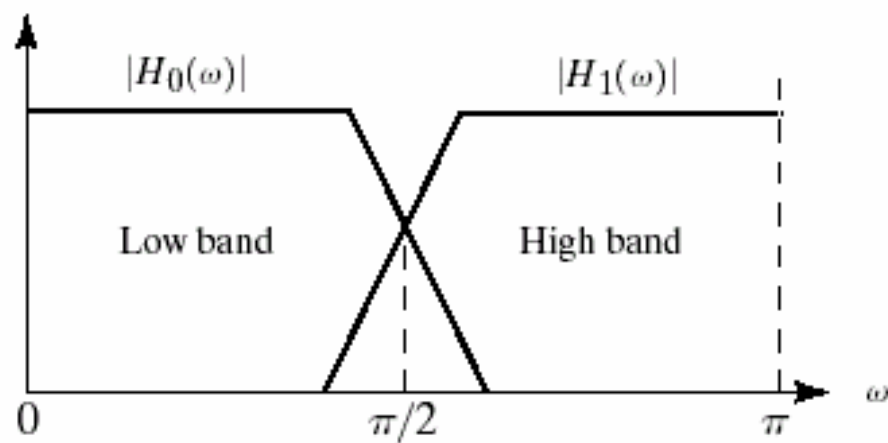
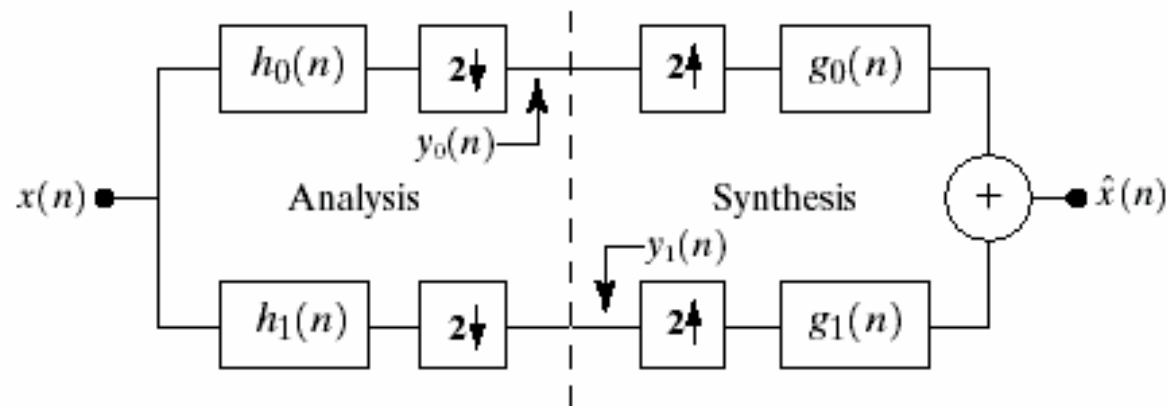


- Image Pyramids
- Subband Coding
- The Haar Transform



a  
b

**FIGURE 7.4** (a) A two-band filter bank for one-dimensional subband coding and decoding, and (b) its spectrum splitting properties.





- The Z-transform, a generalization of the discrete Fourier transform, is the ideal tool for studying discrete-time, sampled-data systems.
- The Z-transform of sequence  $x(n)$  for  $n=0,1,2,\dots$  is

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

- Where  $z$  is a complex variable.



- Downsampling by a factor of 2 in the time domain corresponds to the simple Z-domain operation

$$x_{down}(n) = x(2n) \Leftrightarrow X_{down}(z) = \frac{1}{2} [X(z^{1/2}) + X(-z^{1/2})] \quad (7.1-2)$$

- Upsampling-again by a factor of 2---is defined by the transform pair

$$x^{up}(n) = \begin{cases} x(n/2) & n = 0, 2, 4, \dots \\ 0 & \text{otherwise} \end{cases} \Leftrightarrow X^{up}(z) = x(z^2) \quad (7.1-3)$$



- If sequence  $x(n)$  is downsampled and subsequently upsampled to yield  $\hat{x}(n)$ , Eqs.(7.1-2) and (7.1-3) combine to yield

$$\hat{X}(z) = \frac{1}{2} [X(z) + X(-z)]$$

where  $\hat{x}(n) = Z^{-1}[\hat{X}(z)]$  is the downsampled-upsampled sequence.

- Its inverse Z-transform is

$$Z^{-1}[X(-z)] = (-1)^n x(n)$$



- We can express the system's output as

$$\begin{aligned}\hat{X}(z) &= \frac{1}{2} G_0(z) [H_0(z)X(z) + H_0(-z)X(-z)] \\ &\quad + \frac{1}{2} G_1(z) [H_1(z)X(z) + H_1(-z)X(-z)]\end{aligned}$$

- The output of filter  $h_0(n)$  is defined by the transform pair

$$h_0(n) * x(n) = \sum_k h_0(n-k)x(k) \Leftrightarrow H_0(z)X(z)$$



- As with Fourier transform, convolution in the time (or spatial domain is equivalent to multiplication in the Z-domain.

$$\begin{aligned}\hat{X}(z) = & \frac{1}{2} [H_0(z)G_0(z) + H_1(z)G_1(z)]X(z) \\ & + \frac{1}{2} [H_0(-z)G_0(z) + H_1(-z)G_1(z)]X(-z)\end{aligned}$$



- For error-free reconstruction of the input,  $\hat{x}(n) = x(n)$  and  $\hat{X}(z) = X(z)$ . Thus, we impose the following conditions:

$$H_0(-z)G_0(z) + H_1(-z)G_1(z) = 0$$

$$H_0(z)G_0(z) + H_1(z)G_1(z) = 2$$

To get

$$\begin{bmatrix} G_0(z) \\ G_1(z) \end{bmatrix} = \frac{2}{\det(H_m(z))} \begin{bmatrix} H_1(-z) \\ -H_0(-z) \end{bmatrix}$$

Where  $\det(H_m(z))$  denotes the determinant of  $H_m(z)$ .



$$\det(H_m(z)) = \alpha z^{-(2k+1)}$$

- Letting  $\alpha = 2$ , and taking the inverse Z-transform, we get

$$g_0(n) = (-1)^n h_1(n)$$

$$g_1(n) = (-1)^{n+1} h_0(n)$$

- Letting  $\alpha = -2$ , and taking the inverse Z-transform, we get

$$g_0(n) = (-1)^{n+1} h_1(n)$$

$$g_1(n) = (-1)^n h_0(n)$$



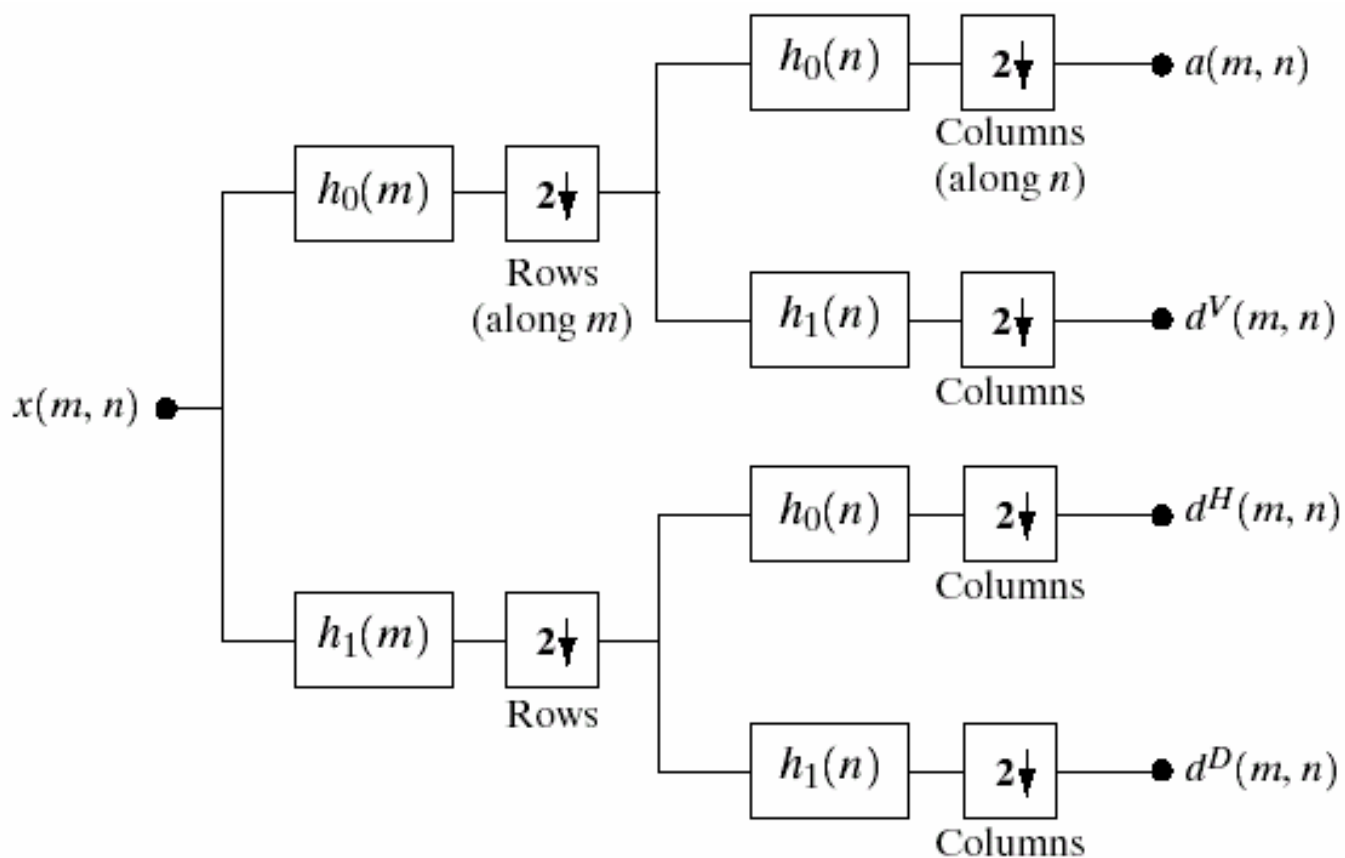


**TABLE 7.1**

Perfect reconstruction filter families.

| Filter   | QMF                        | CQF   | Orthonormal                                   |
|----------|----------------------------|---|---|
| $H_0(z)$ | $H_0^2(z) - H_0^2(-z) = 2$ | $H_0(z)H_0(z^{-1}) + H_0^2(-z)H_0(-z^{-1}) = 2$ | $G_0(z^{-1})$                                 |
| $H_1(z)$ | $H_0(-z)$                  | $z^{-1}H_0(-z^{-1})$                            | $G_1(z^{-1})$                                 |
| $G_0(z)$ | $H_0(z)$                   | $H_0(z^{-1})$                                   | $G_0(z)G_0(z^{-1}) + G_0(-z)G_0(-z^{-1}) = 2$ |
| $G_1(z)$ | $-H_0(-z)$                 | $zH_0(-z)$                                      | $-z^{-2K+1}G_0(-z^{-1})$                      |

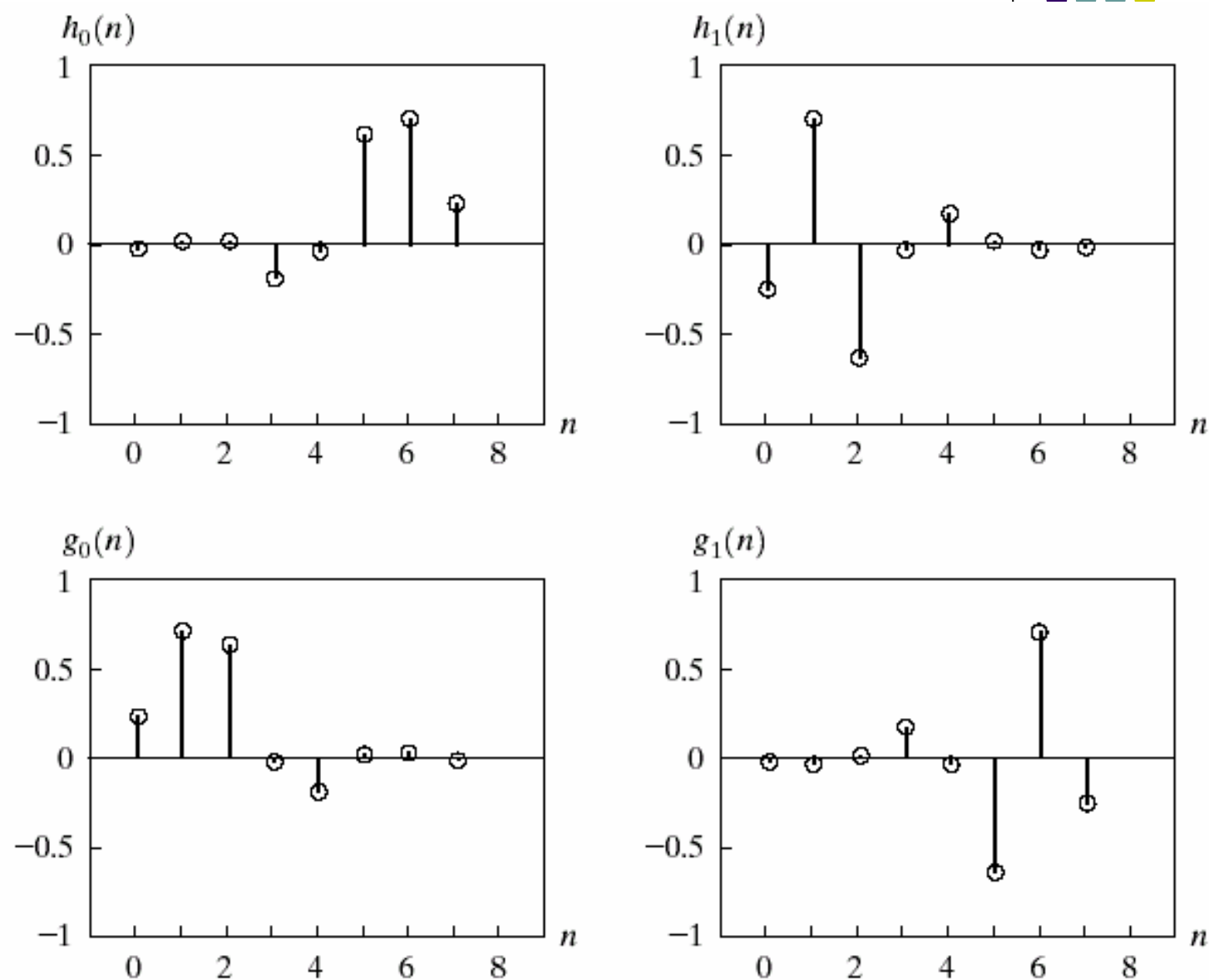
- Three general solution:
  - Quadrature mirror filters (OMFs)
  - Conjugate quadrature filters (CQFs)
  - Orthonormal

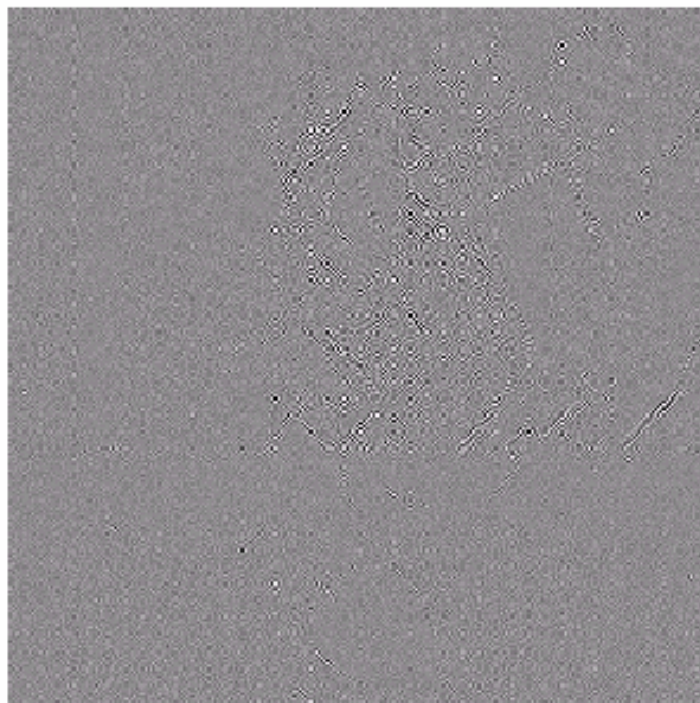
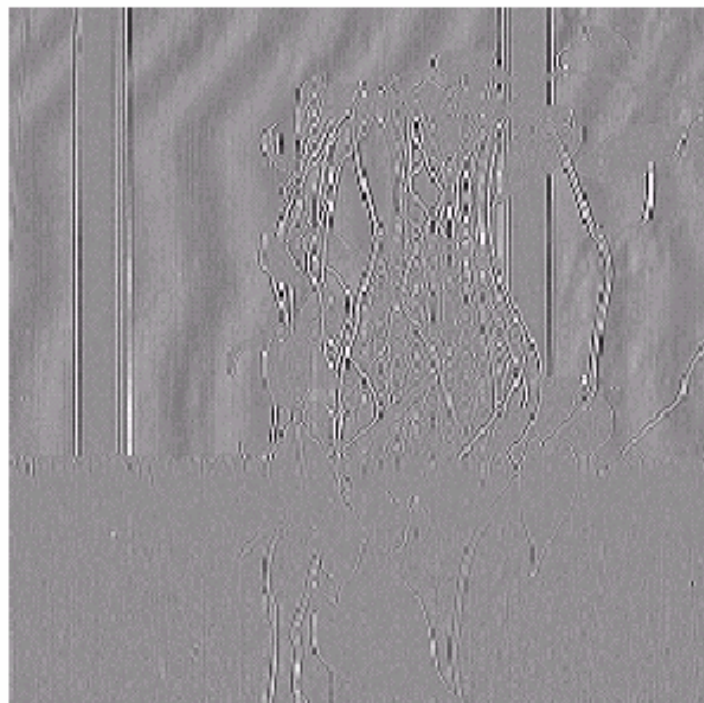
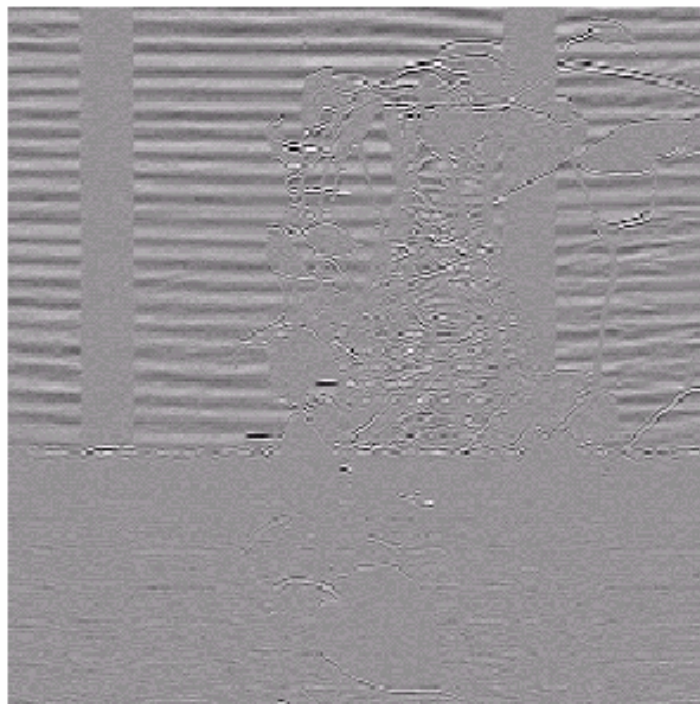


**FIGURE 7.5** A two-dimensional, four-band filter bank for subband image coding.



**FIGURE 7.6** The impulse responses of four 8-tap Daubechies orthonormal filters.





**FIGURE 7.7** A four-band split of the vase in Fig. 7.1 using the subband coding system of Fig. 7.5.



- Image Pyramids
- Subband Coding
- The Haar Transform



- The Haar transform can be expressed in matrix form

$$T = HFH^T$$

- Where
  - F is an N\*N image matrix,
  - H is an N\*N transformation matrix,
  - T is the resulting N\*N transform.



- For the Haar transform, transformation matrix  $H$  contains the Haar basis functions,  $h_k(z)$ . They are defined over the continuous, closed interval  $z \in [0,1]$  for  $k=0,1,2,\dots,N-1$ , where  $N = 2^n$ .
- To generate  $H$ , we define the integer  $k$  such that

$$k = 2^p + q - 1$$

where  $0 \leq p \leq n-1$ ,  $q = 0$  or  $1$  for  $p = 0$ .

$$1 \leq q \leq 2^p \text{ for } p \neq 0$$



- Then the Haar basis functions are

$$h_0(z) = h_{00}(z) = \frac{1}{\sqrt{N}}, \quad z \in [0,1]$$

- and

$$h_k(z) = h_{pq}(z) = \frac{1}{\sqrt{N}} \begin{cases} 2^{p/2} & (q-1)/2^p \leq z < (q-0.5)/2^p \\ -2^{p/2} & (q-0.5)/2^p \leq z < q/2^p \\ 0 & \text{otherwise, } z \in [0,1] \end{cases}$$





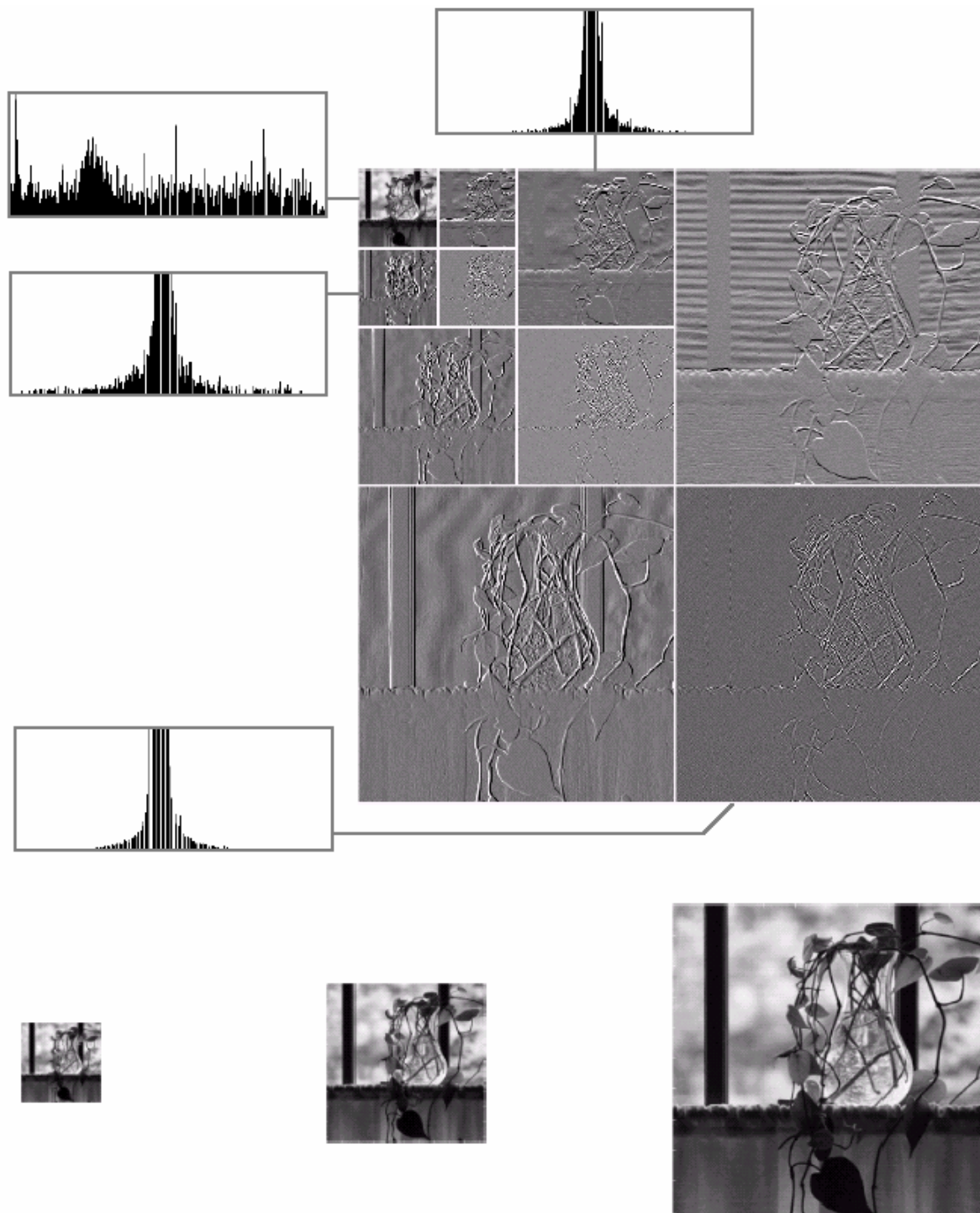
- The  $i$ th row of an  $N \times N$  Haar transformation matrix contains the elements of  $h_i(z)$  for  $z = 0/N, 1/N, 2/N, \dots, (N-1)/N$ .
- If  $N=4$ , for example  $k, q$ , and  $p$  assume the values

| k | p | q |
|---|---|---|
| 0 | 0 | 0 |
| 1 | 0 | 1 |
| 2 | 1 | 1 |
| 3 | 1 | 2 |



- The 4\*4 transformation matrix,  $H_4$ , is

$$H_4 = \frac{1}{\sqrt{4}} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & 1 & -1 & -1 \\ \sqrt{2} & -\sqrt{2} & 0 & 0 \\ 0 & 0 & \sqrt{2} & -\sqrt{2} \end{bmatrix}$$



a  
b c d

**FIGURE 7.8** (a) A discrete wavelet transform using Haar basis functions. Its local histogram variations are also shown; (b)–(d) Several different approximations ( $64 \times 64$ ,  $128 \times 128$ , and  $256 \times 256$ ) that can be obtained from (a).





- Background
- Multiresolution Expansions
- Wavelet Transforms in One Dimension
- Wavelet Transforms in Two Dimensions



- Series Expansion
- Scaling Functions
- Wavelet Functions



- A signal of function  $f(x)$  can often be better analyzed as a linear combination of expansion functions

$$f(x) = \sum_k \alpha_k \varphi_k(x)$$

- $k$  is an interger index of the finite or infinite sum;
- $\alpha_k$  are real-valued expansion coefficients;
- $\varphi_k(x)$  are real-valued expansion functions.



- These coefficients are computed by taking the integral inner products of the dual  $\tilde{\varphi}_k(x)$ 's and function  $f(x)$ . That is

$$\alpha_k = \langle \tilde{\varphi}_k(x), f(x) \rangle = \int \tilde{\varphi}_k^*(x) f(x) dx$$



- Series Expansion
- Scaling Functions
- Wavelet Functions





- The set of expansion functions composed of integer translations and binary scaling of the real, square-integrable function  $\varphi(x)$  ; that is, the set  $\{\varphi_{j,k}(x)\}$  where

$$\varphi_{j,k}(x) = 2^{j/2} \varphi(2^j x - k)$$



$$V_{j_0} = \overline{\text{Span}_k \{ \varphi_{j_0,k}(x) \}}$$

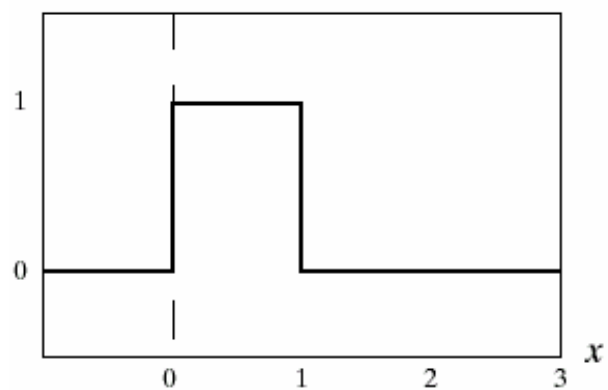
- If  $f(x) \in V_{j_0}$ , it can be written

$$f(x) = \sum_k \alpha_k \varphi_{j_0,k}(x)$$

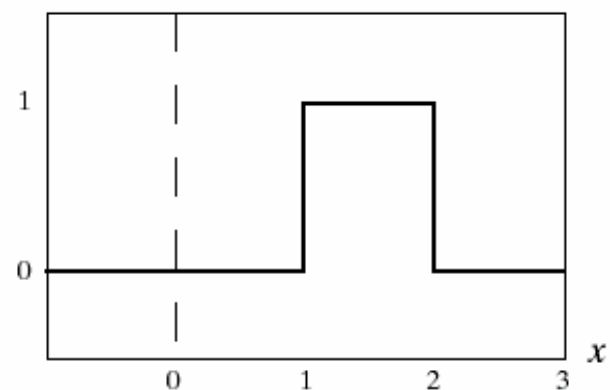
- We will denote the subspace spanned over  $k$  for any  $j$  as

$$V_j = \overline{\text{Span}_k \{ \varphi_{j,k}(x) \}}$$

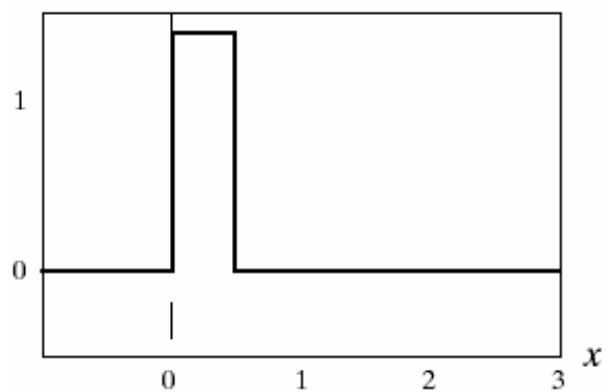
$$\varphi_{0,0}(x) = \varphi(x)$$



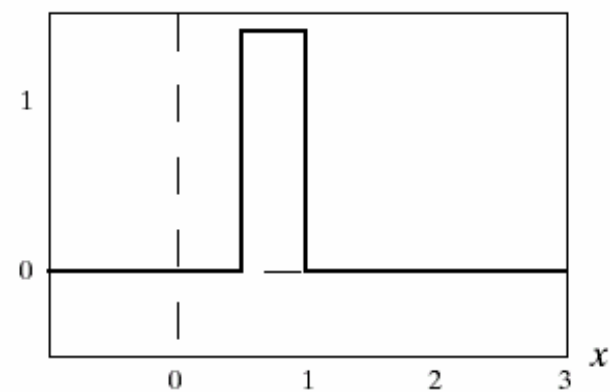
$$\varphi_{0,1}(x) = \varphi(x - 1)$$



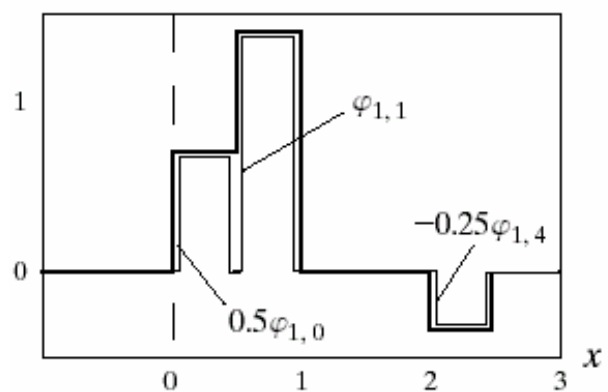
$$\varphi_{1,0}(x) = \sqrt{2} \varphi(2x)$$



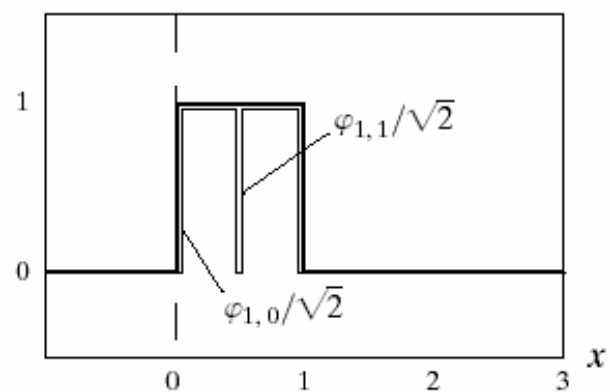
$$\varphi_{1,1}(x) = \sqrt{2} \varphi(2x - 1)$$



$$f(x) \in V_1$$



$$\varphi_{0,0}(x) \in V_1$$



|   |   |
|---|---|
| a | b |
| c | d |
| e | f |

**FIGURE 7.9** Haar scaling functions in  $V_0$  in  $V_1$ .

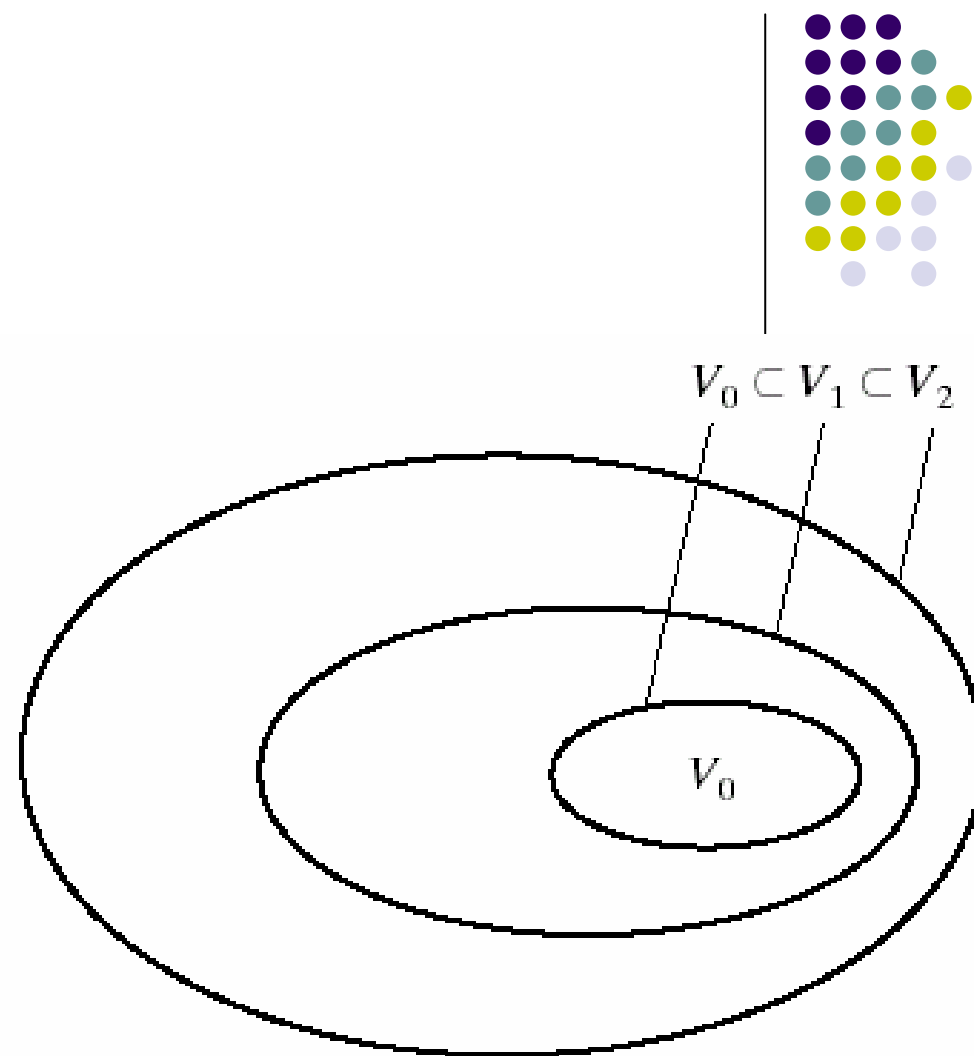




- The simple scaling function in the preceding example obeys the four fundamental requirements of multiresolution analysis:
  - MRA Requirement 1: The scaling function is orthogonal to its integer translates;
  - MRA Requirement 2: The subspaces spanned by the scaling function at low scales are nested within those spanned at higher scales.

**FIGURE 7.10** The nested function spaces spanned by a scaling function.

---





- MRA Requirement 3: The only function that is common to all  $V_j$  is  $f(x)=0$ .
- MRA Requirement 4: Any function can be represented with arbitrary precision.



- Series Expansion
- Scaling Functions
- Wavelet Functions



- Given a scaling function that meets the MRA requirements of the previous section, we can define a wavelet function  $\psi(x)$  that, together with its integer translates and binary scaling, spans the difference between any two adjacent scaling subspaces,  $V_j$  and  $V_{j+1}$ . We define the set  $\{\psi_{j,k}(x)\}$  of wavelets

$$\{\psi_{j,k}(x) = 2^{j/2} \psi(2^j x - k)\}$$





- As with scaling functions, we write

$$W_j = \overline{\text{Span}_k \{ \psi_{j,k}(x) \}}$$

- And note that if  $f(x) \in W_j$

$$f(x) = \sum_k \alpha_k \psi_{j,k}(x)$$

- The scaling and wavelet function subspaces are related by

$$V_{j+1} = V_j \oplus W_j$$



- We can now express the space of all measurable, square-integrable functions as

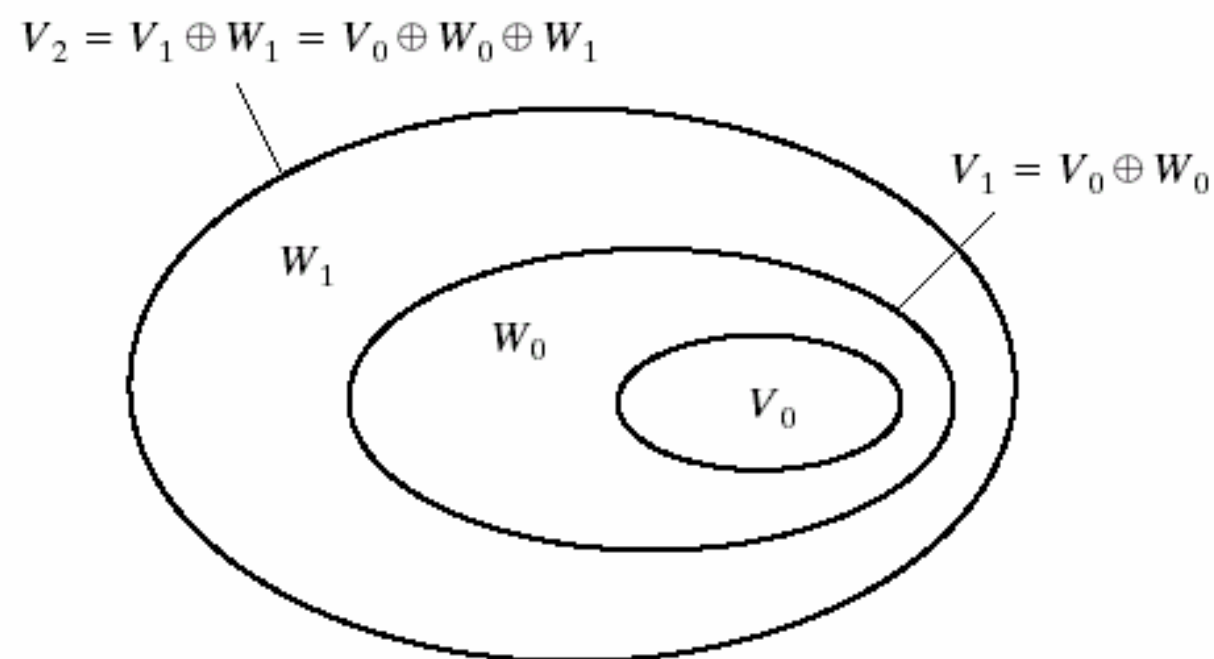
$$L^2(R) = V_0 \oplus W_0 \oplus W_1 \oplus \dots$$

- or

$$L^2(R) = V_1 \oplus W_1 \oplus W_2 \oplus \dots$$



**FIGURE 7.11** The relationship between scaling and wavelet function spaces.

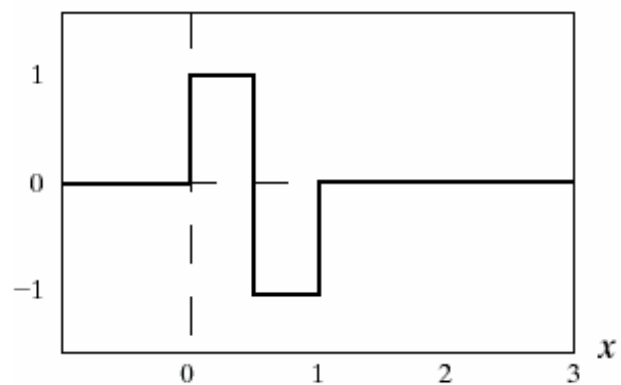




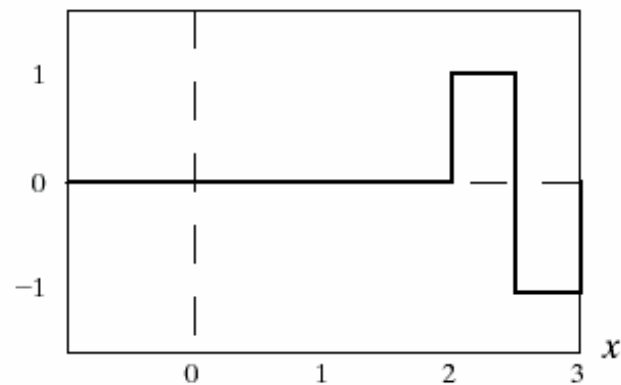
- The Haar wavelet function is

$$\psi(x) = \begin{cases} 1 & 0 \leq x < 0.5 \\ -1 & 0.5 \leq x < 1 \\ 0 & \text{elsewhere} \end{cases}$$

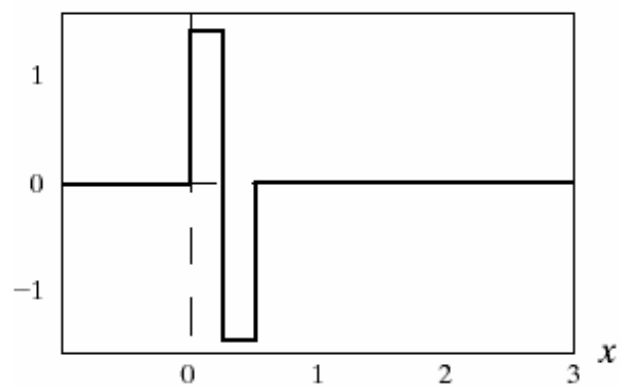
$$\psi(x) = \psi_{0,0}(x)$$



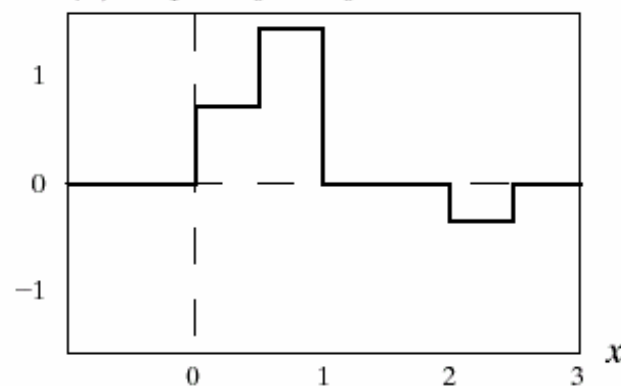
$$\psi_{0,2}(x) = \psi(x - 2)$$



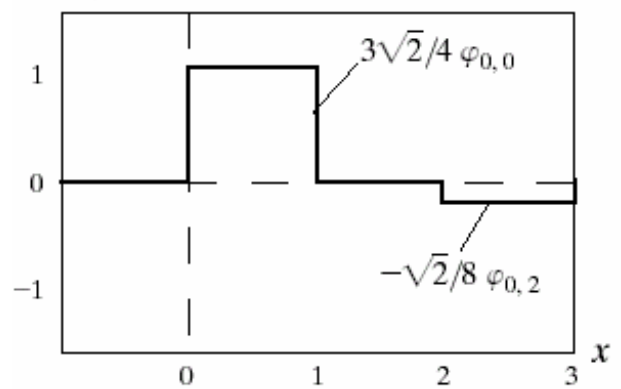
$$\psi_{1,0}(x) = \sqrt{2} \psi(2x)$$



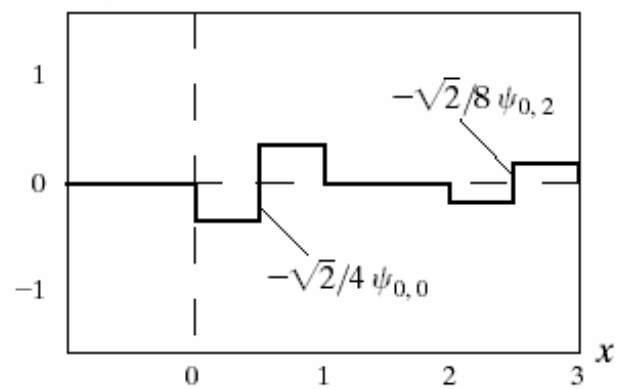
$$f(x) \in V_1 = V_0 \oplus W_0$$



$$f_a(x) \in V_0$$



$$f_d(x) \in W_0$$



|   |   |
|---|---|
| a | b |
| c | d |
| e | f |

**FIGURE 7.12** Haar wavelet functions in  $W_0$  and  $W_1$ .





- Background
- Multiresolution Expansions
- Wavelet Transforms in One Dimension
- Wavelet Transforms in Two Dimensions



- The Wavelet Series Expansions
- The Discrete Wavelet Transform
- The Continuous Wavelet Transform



- Defining the wavelet series expansion of function  $f(x) \in L^2(R)$  relative to wavelet  $\psi(x)$  and scaling function  $\varphi(x)$  .  $f(x)$  can be written as

$$f(x) = \sum_k c_{j_0}(k) \varphi_{j_0,k}(x) + \sum_{j=j_0}^{\infty} d_j(k) \psi_{j,k}(x)$$

- $c_{j_0}(k)$ 's : the approximation or scaling coefficients;
- $d_j(k)$ 's : the detail or wavelet coefficients.



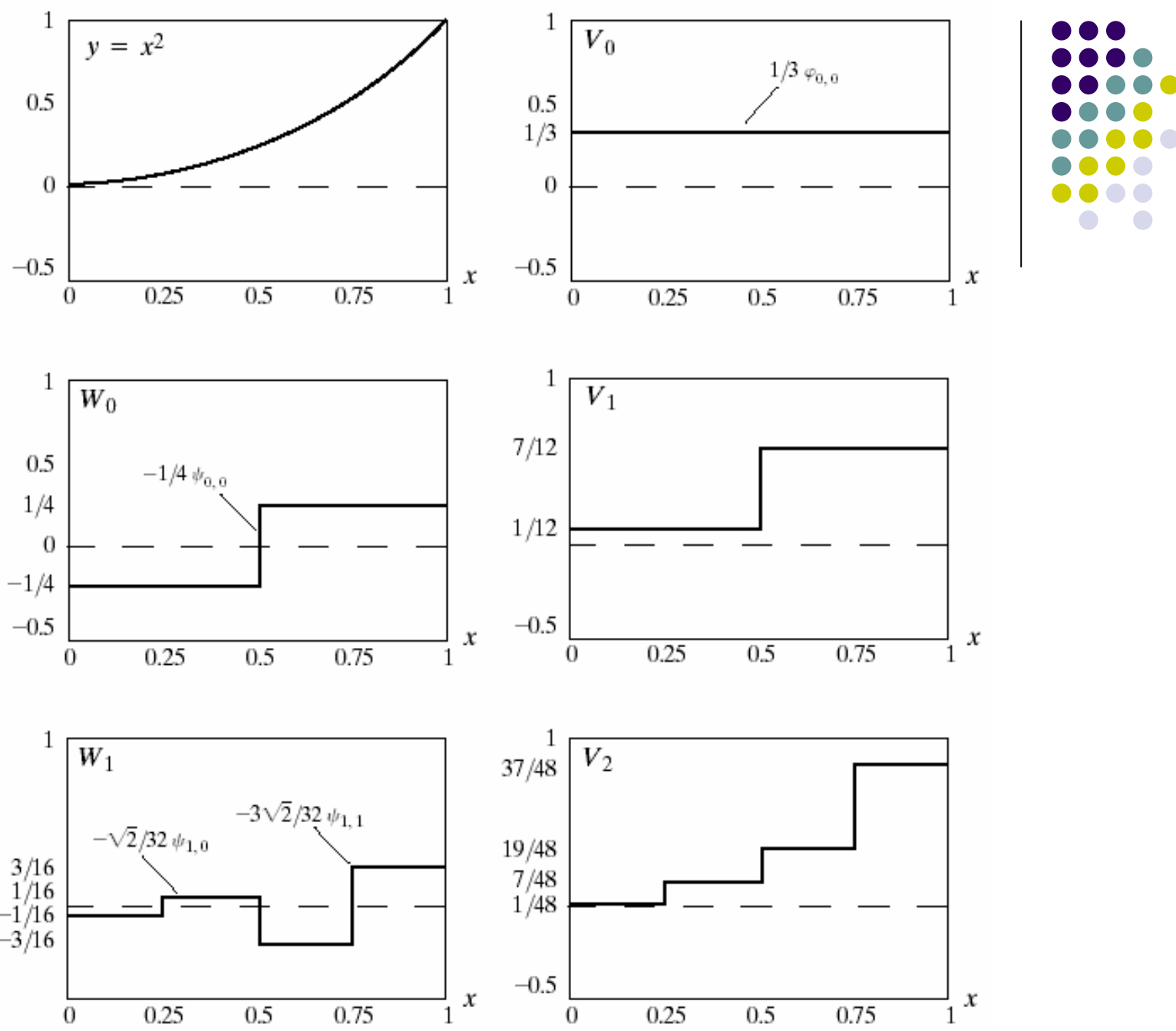


- If the expansion functions form an orthonormal basis or tight frame, the expansion coefficients are calculated as

$$c_{j_0}(k) = \langle f(x), \varphi_{j_0,k}(x) \rangle = \int f(x) \varphi_{j_0,k}(x) dx$$

- and

$$d_j(k) = \langle f(x), \psi_{j,k}(x) \rangle = \int f(x) \psi_{j,k}(x) dx$$



**FIGURE 7.13** A wavelet series expansion of  $y = x^2$  using Haar wavelets.



- The Wavelet Series Expansions
- The Discrete Wavelet Transform
- The Continuous Wavelet Transform



- If the function being expanded is a sequence of numbers, like samples of a continuous function  $f(x)$ , the resulting coefficients are called the discrete wavelet transform(DWT) of  $f(x)$ .

$$W_{\varphi}(j_0, k) = \frac{1}{\sqrt{M}} \sum_x f(x) \varphi_{j_0, k}(x)$$

$$W_{\psi}(j, k) = \frac{1}{\sqrt{M}} \sum_x f(x) \psi_{j, k}(x)$$

- and

$$f(x) = \frac{1}{\sqrt{M}} \sum_k W_{\varphi}(j_0, k) \varphi_{j_0, k}(x) + \frac{1}{\sqrt{M}} \sum_{j=j_0}^{\infty} \sum_k W_{\psi}(j, k) \psi_{j, k}(x)$$



- Consider the discrete function of four points:  
 $f(0)=1$ ,  $f(1)=4$ ,  $f(2)=-3$ , and  $f(3)=0$
- Since  $M=4$ ,  $J=2$  and, with  $j_0=0$ , the summations are performed over  
 $x=0,1,2,3$ ,  
 $j=0,1$ , and  
 $k=0$  for  $j=0$   
or  $k=0,1$  for  $j=1$ .



- We find that

$$W_{\varphi}(0,0) = \frac{1}{2} \sum_{x=0}^3 f(x) \varphi_{0,0}(x) = \frac{1}{2} [1 \cdot 1 + 4 \cdot 1 - 3 \cdot 1 + 0 \cdot 1] = 1$$

$$W_{\psi}(0,0) = \frac{1}{2} [1 \cdot 1 + 4 \cdot 1 - 3 \cdot (-1) + 0 \cdot (-1)] = 4$$

$$W_{\psi}(1,0) = \frac{1}{2} [1 \cdot \sqrt{2} + 4 \cdot (-\sqrt{2}) - 3 \cdot 0 + 0 \cdot 0] = -1.5\sqrt{2}$$

$$W_{\psi}(1,1) = \frac{1}{2} [1 \cdot 0 + 4 \cdot 0 - 3 \cdot \sqrt{2} + 0 \cdot (-\sqrt{2})] = -1.5\sqrt{2}$$



$$f(x) = \frac{1}{2} [W_{\varphi}(0,0)\varphi_{0,0}(x) + W_{\psi}(0,0)\psi_{0,0}(x) + W_{\psi}(1,0)\psi_{1,0}(x) + W_{\psi}(1,1)\psi_{1,1}(x)]$$

- For  $x=0,1,2,3$ . If  $x=0$ , for instance,

$$f(0) = \frac{1}{2} [1 \cdot 1 + 4 \cdot 1 - 1.5\sqrt{2} \cdot (\sqrt{2}) - 1.5\sqrt{2} \cdot 0] = 1$$



- The Wavelet Series Expansions
- The Discrete Wavelet Transform
- The Continuous Wavelet Transform





- The continuous wavelet transform of a continuous, square-integrable function,  $f(x)$ , relative to a real-valued wavelet,  $\psi(x)$ , is

$$W_{\psi}(s, \tau) = \int_{-\infty}^{\infty} f(x) \psi_{s, \tau}(x) dx$$

- Where

$$\psi_{s, \tau}(x) = \frac{1}{\sqrt{s}} \psi\left(\frac{x - \tau}{s}\right)$$

- And  $s$  and  $\tau$  are called scale and translation parameters.



- Given  $W_{\psi}(s, \tau)$ ,  $f(x)$  can be obtained using the inverse continuous wavelet transform

$$f(x) = \frac{1}{C_{\psi}} \int_0^{\infty} \int_{-\infty}^{\infty} W_{\psi}(s, \tau) \frac{\psi_{s,\tau}(x)}{s^2} d\tau ds$$

- Where

$$C_{\psi} = \int_{-\infty}^{\infty} \frac{|\Psi(u)|^2}{|u|} du$$

- And  $\Psi(u)$  is the Fourier transform of  $\psi(x)$  .

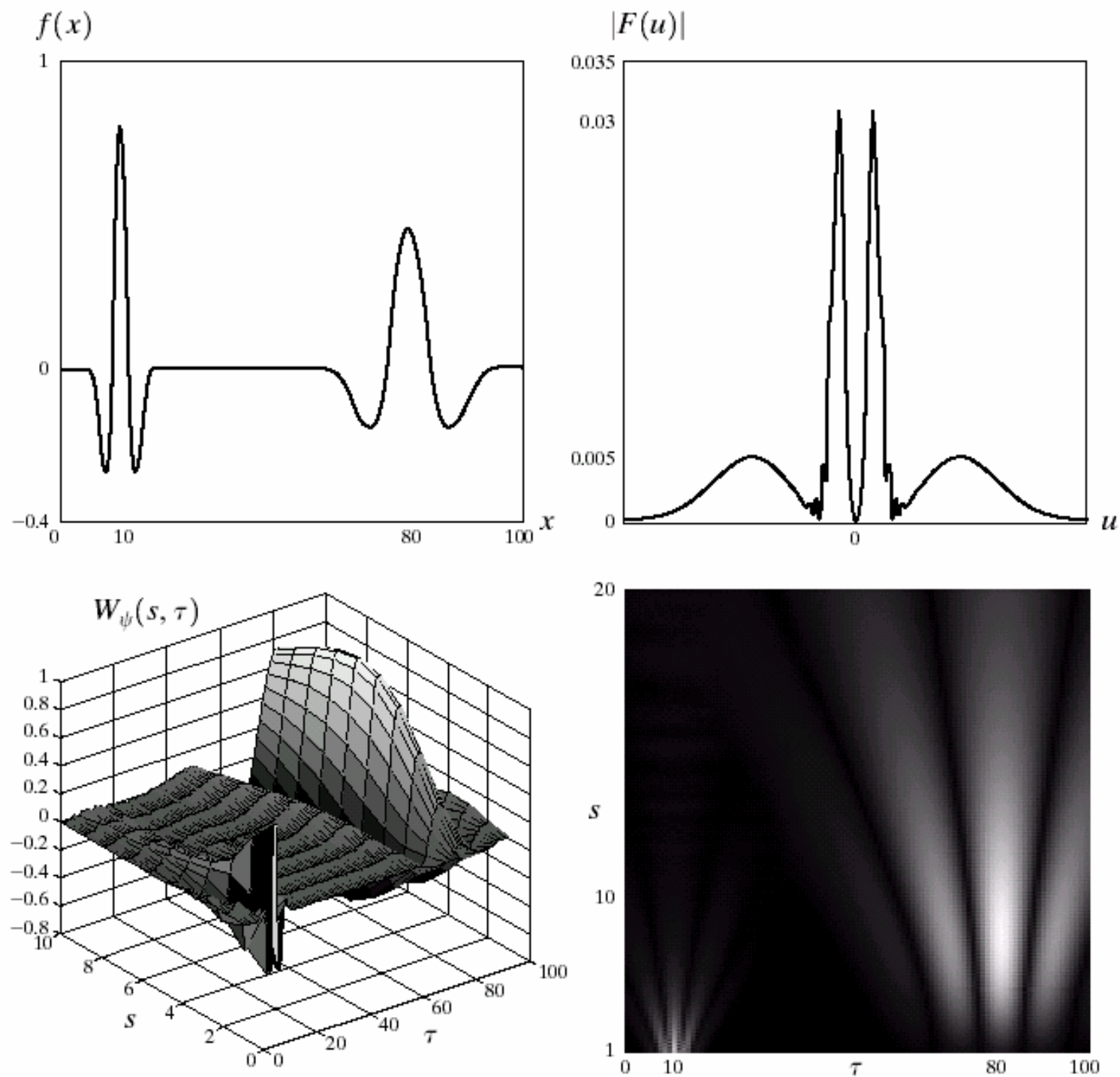


- The Mexican hat wavelet

$$\psi(x) = \left( \frac{2}{\sqrt{3}} \pi^{-1/4} \right) (1 - x^2) e^{-x^2/2}$$

a b  
c d

**FIGURE 7.14** The continuous wavelet transform (c and d) and Fourier spectrum (b) of a continuous one-dimensional function (a).





- Background
- Multiresolution Expansions
- Wavelet Transforms in One Dimension
- Wavelet Transforms in Two Dimensions



- In two dimensions, a two-dimensional scaling function,  $\varphi(x, y)$ , and three two-dimensional wavelet,  $\psi^H(x, y)$ ,  $\psi^V(x, y)$  and  $\psi^V(x, y)$ , are required.



- Excluding products that produce one-dimensional results, like  $\varphi(x)\psi(x)$ , the four remaining products produce the separable scaling function

$$\varphi(x, y) = \varphi(x)\varphi(y)$$

- And separable, “directionally sensitive” wavelets

$$\psi^H(x, y) = \psi(x)\varphi(y)$$

$$\psi^V(x, y) = \varphi(x)\psi(y)$$

$$\psi^D(x, y) = \psi(x)\psi(y)$$



- The scaled and translated basis functions:

$$\varphi_{j,m,n}(x, y) = 2^{j/2} \varphi(2^j x - m, 2^j y - n)$$

$$\psi^i_{j,m,n}(x, y) = 2^{j/2} \psi(2^j x - m, 2^j y - n), \quad i = \{H, V, D\}$$





- The discrete wavelet transform of function  $f(x,y)$  of size  $M*N$  is then

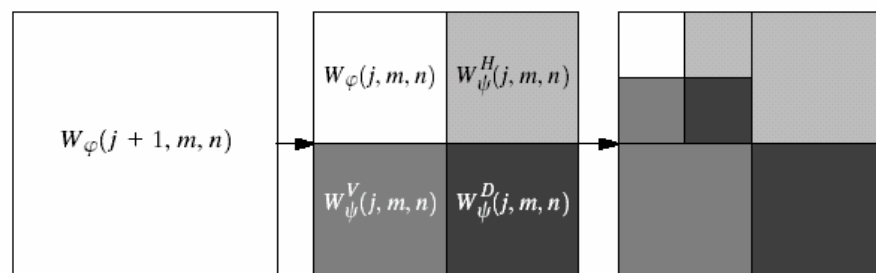
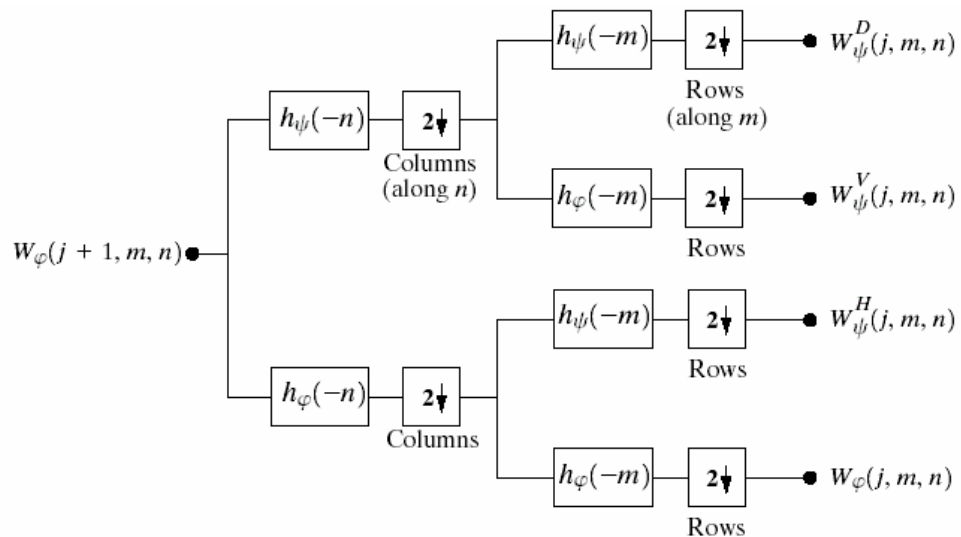
$$W_{\varphi}(j_0, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \varphi_{j_0, m, n}(x, y)$$

$$W_{\psi}^i(j, m, n) = \frac{1}{\sqrt{MN}} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \psi^i_{j, m, n}(x, y) \quad i = \{H, V, D\}$$

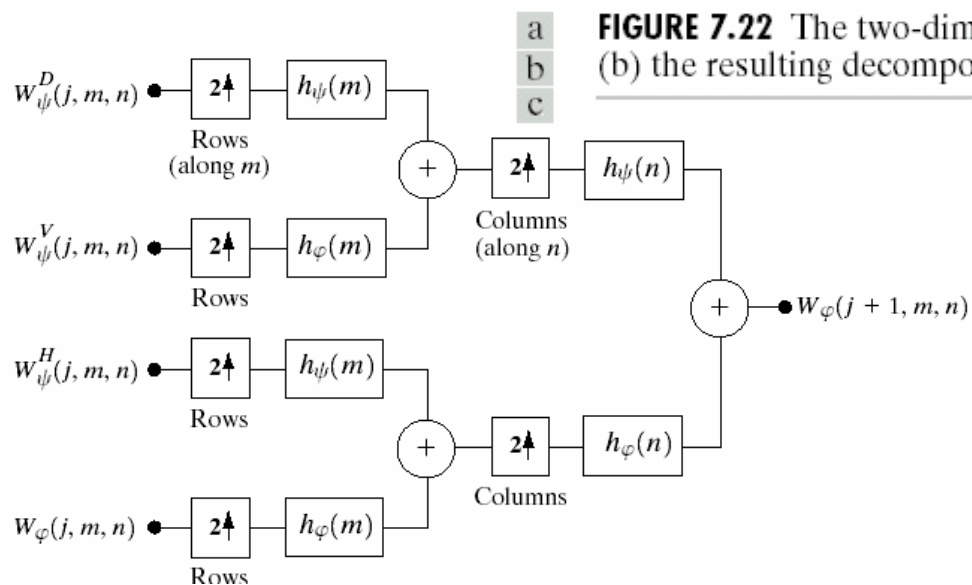


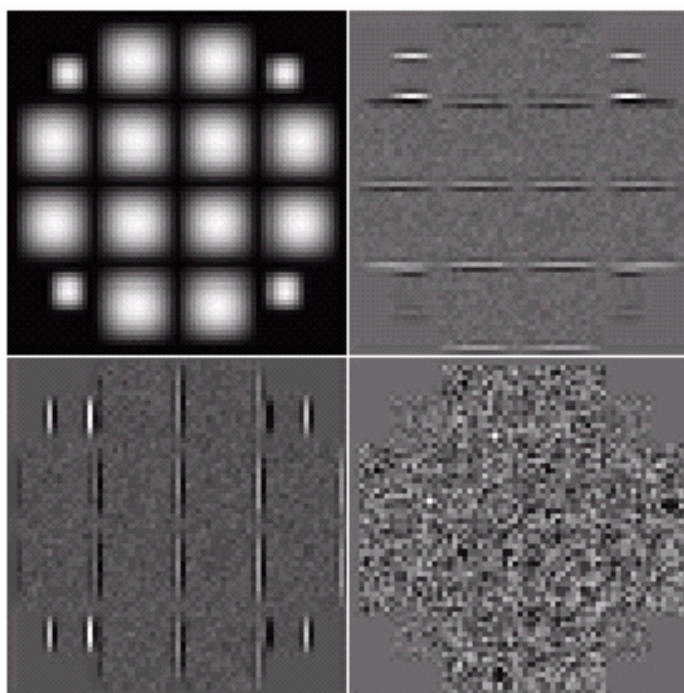
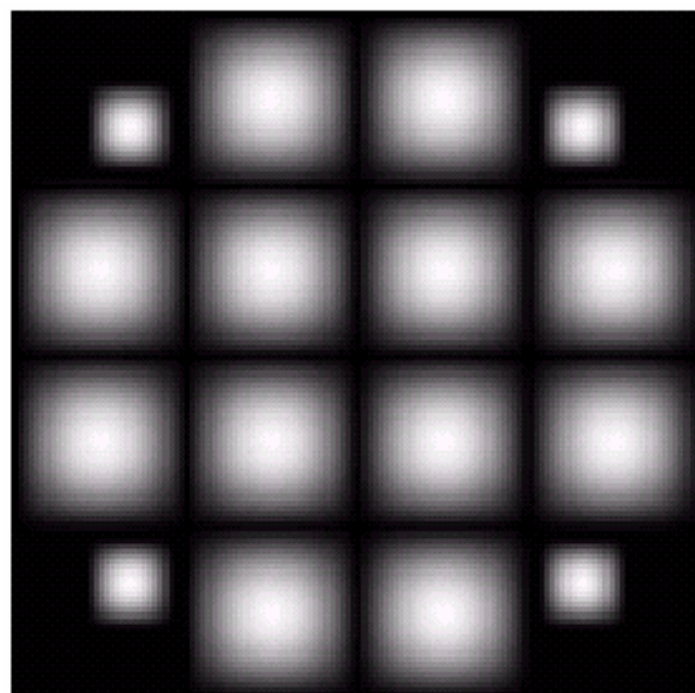
- Given the  $W_\varphi$  and  $W_\psi^i$ ,  $f(x,y)$  is obtained via the inverse discrete wavelet transform

$$f(x, y) = \frac{1}{\sqrt{MN}} \sum_m \sum_n W_\varphi(j_0, m, n) \varphi_{j_0, m, n}(x, y) \\ + \frac{1}{\sqrt{MN}} \sum_{i=H,V,D} \sum_{j=j_0}^{\infty} \sum_m \sum_n W_\psi^i(j, m, n) \psi_{j, m, n}^i(x, y)$$



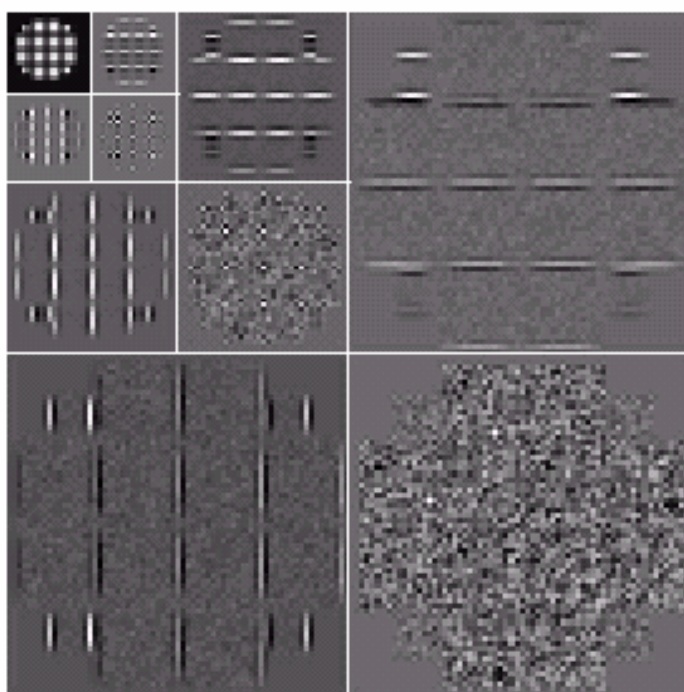
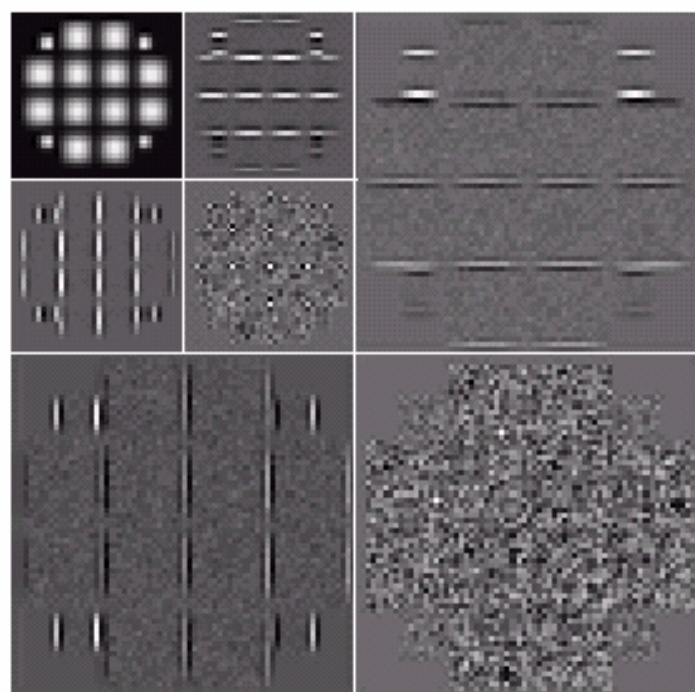
**FIGURE 7.22** The two-dimensional fast wavelet transform: (a) the analysis filter bank; (b) the resulting decomposition; and (c) the synthesis filter bank.





|   |   |
|---|---|
| a | b |
| c | d |

**FIGURE 7.23** A three-scale FWT.



a b  
c d  
e f  
g

**FIGURE 7.24**

Fourth-order symlets:  
(a)–(b) decomposition filters;  
(c)–(d) reconstruction filters;  
(e) the one-dimensional wavelet;  
(f) the one-dimensional scaling function;  
and (g) one of three two-dimensional wavelets,  
 $\psi^H(x, y)$ .

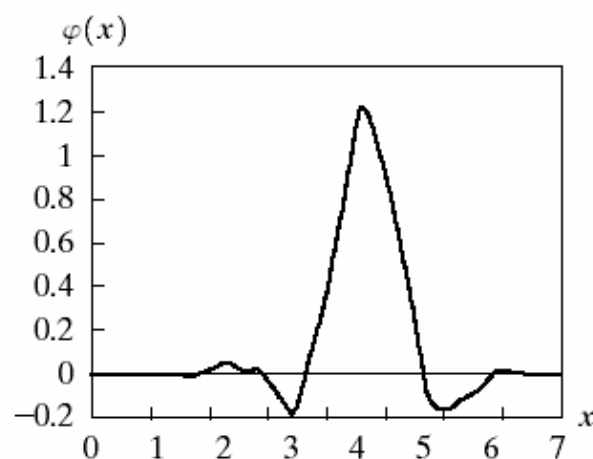
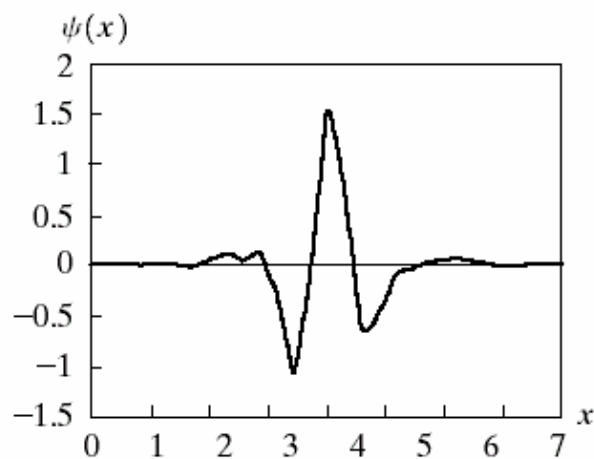
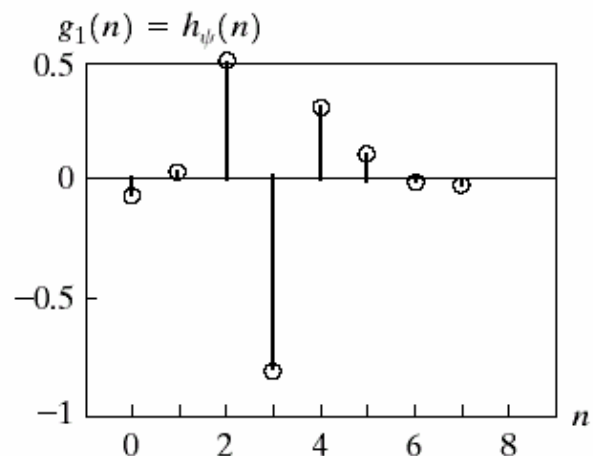
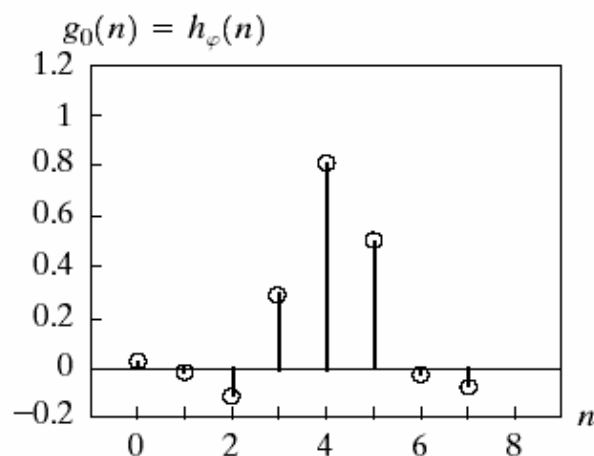
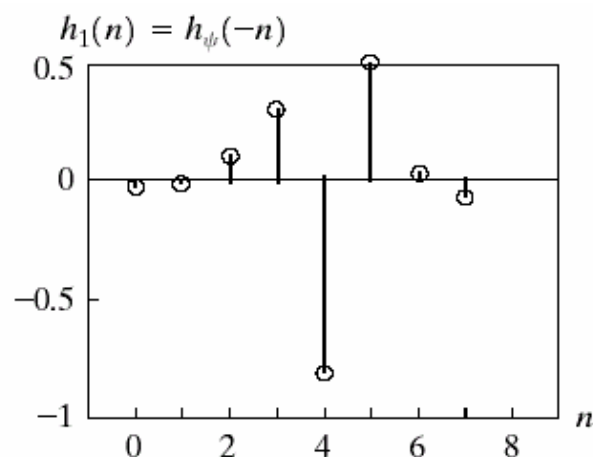
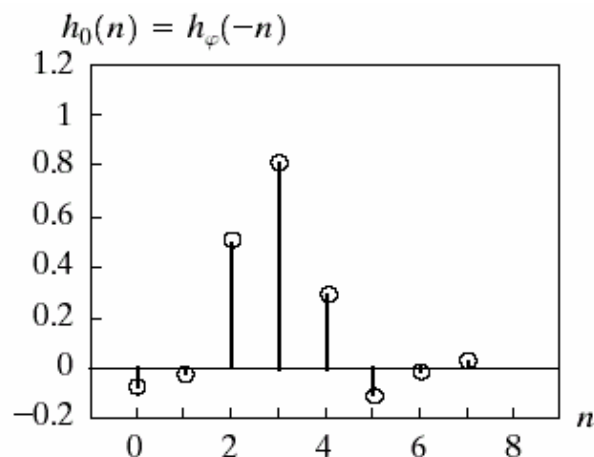
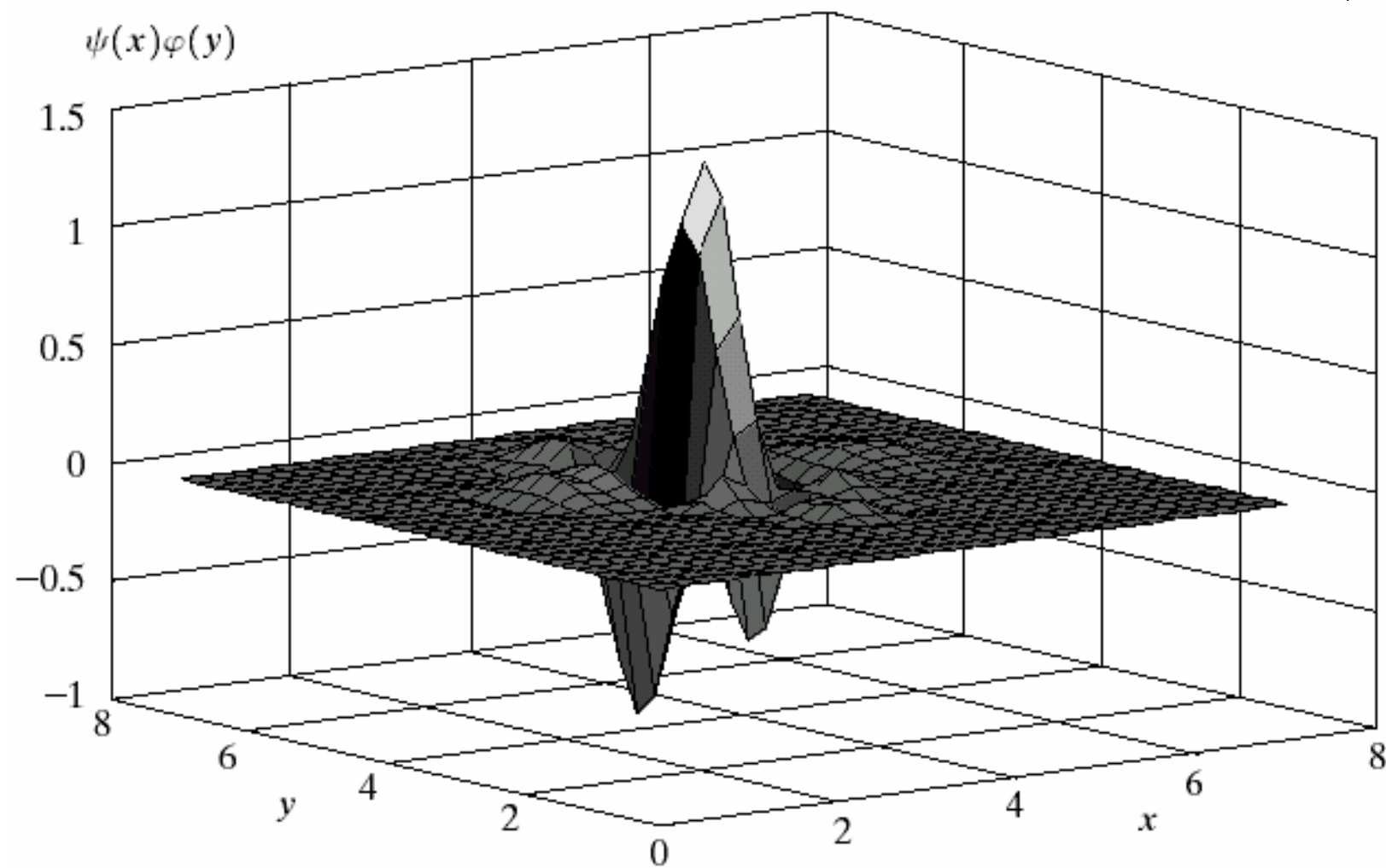
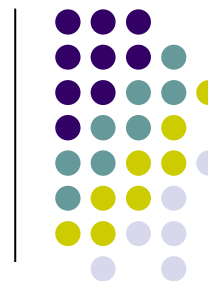


Fig. 7.24 (Con't)

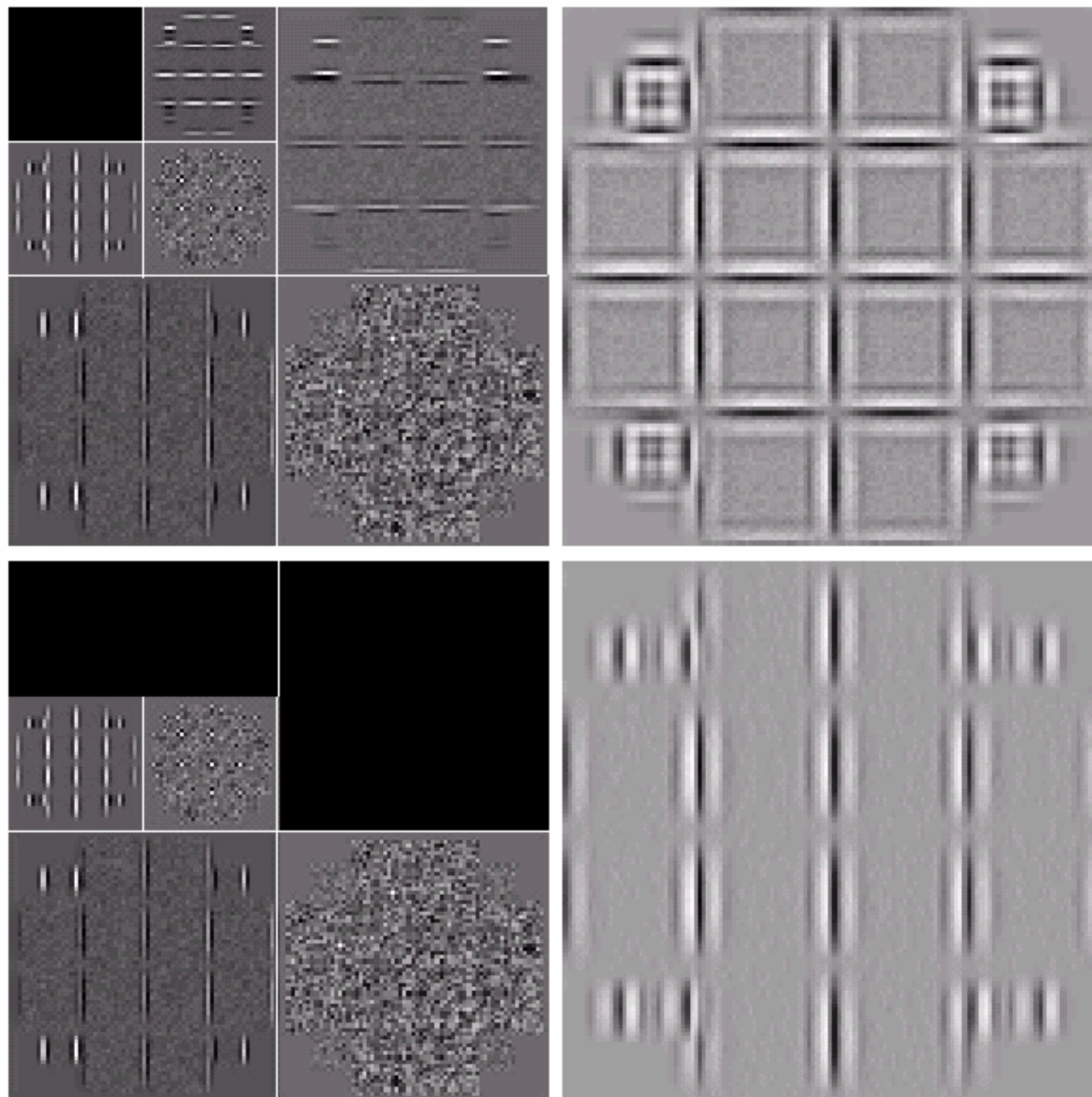


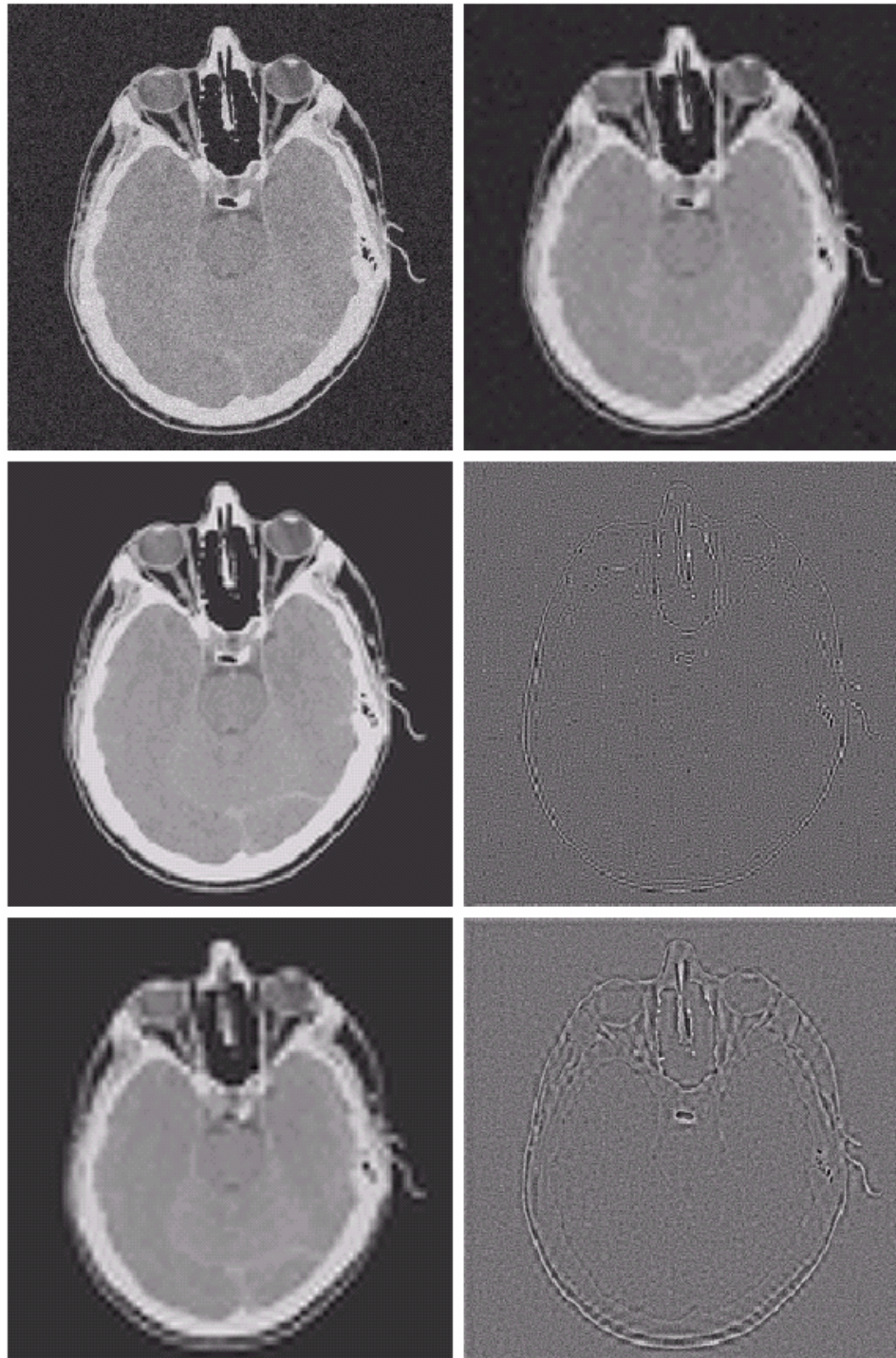


|   |   |
|---|---|
| a | b |
| c | d |

**FIGURE 7.25**

Modifying a DWT for edge detection: (a) and (c) two-scale decompositions with selected coefficients deleted; (b) and (d) the corresponding reconstructions.





a b  
c d  
e f

**FIGURE 7.26**  
Modifying a DWT  
for noise removal:  
(a) a noisy MRI  
of a human head;  
(b), (c) and  
(e) various  
reconstructions  
after thresholding  
the detail  
coefficients; (d)  
and (f) the  
information  
removed during  
the reconstruction  
of (c) and (e).  
(Original image  
courtesy  
Vanderbilt  
University  
Medical Center.)

