

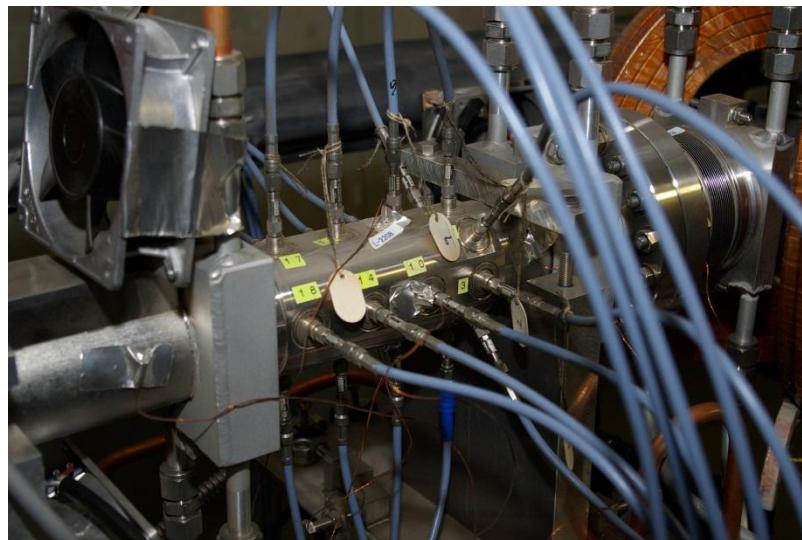
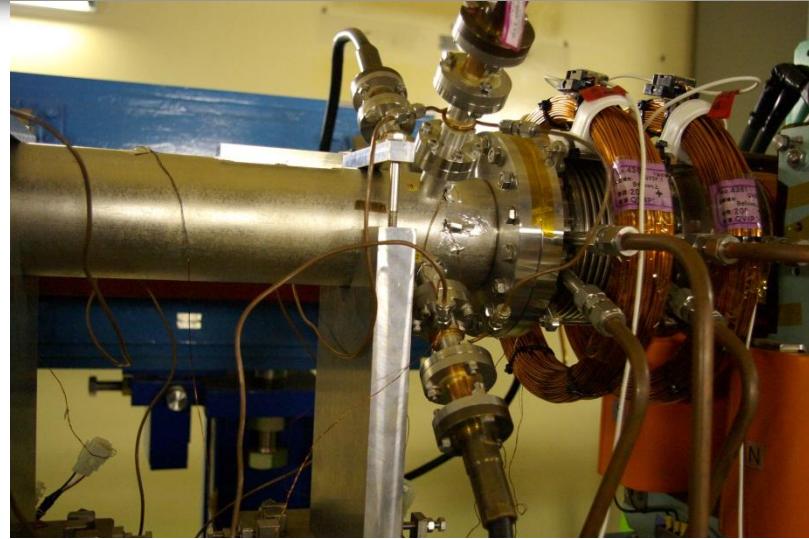
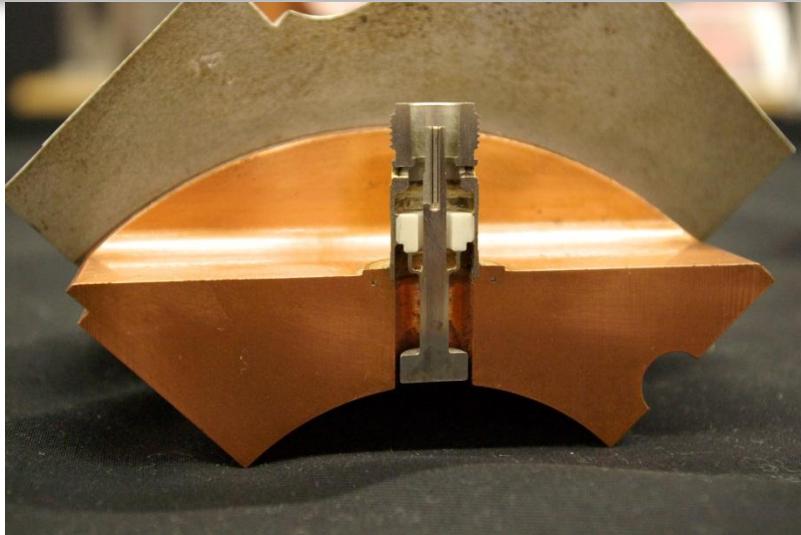
# BPM electrode and high power feedthrough –Special topics in wideband feedthrough–

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KEK Accelerator Laboratory

# Introduction

- Beam runs in vacuum chambers (mostly) made of good-conducting metal.
  - Vacuum chambers shield (fast) electro-magnetic wave from / to the beam.
  - Need some kind of “Feedthrough” to get / feed the electro-magnetic signal from / to beam.
- Wideband coaxial feedthrough
  - Button-type electrode
  - Stripline kicker / monitor
  - Plate-type position monitor
  - Ion (or electron cloud) clearing electrode

# Feedthrough



# Need to consider..

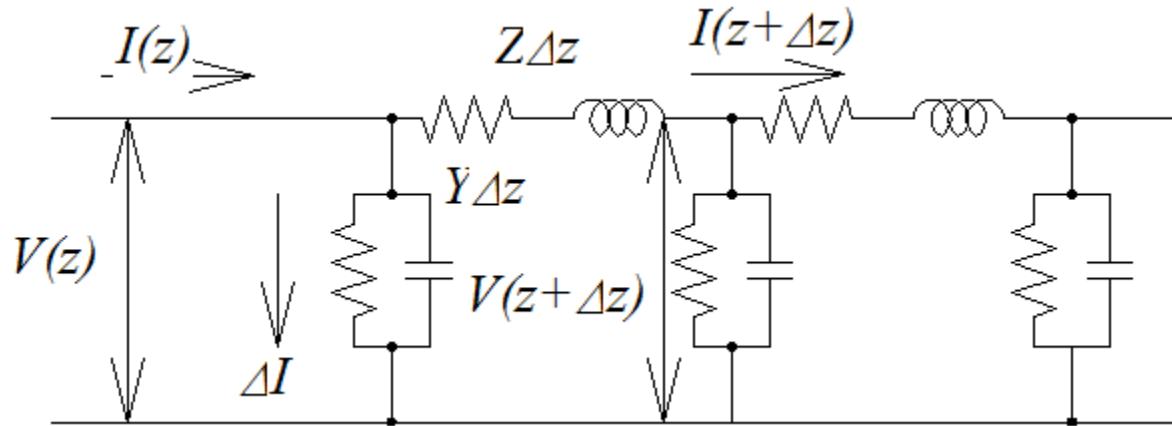
## ■ Specification of the feedthrough

- Structure
  - Total size
  - Kind of RF connector, vacuum structure...
  - Mechanical (and heating) toughness
- Frequency range, allowed SWR, reflection..
- Power range
  - Beam induced power
  - Supplied power from outside
  - allowed (V)SWR

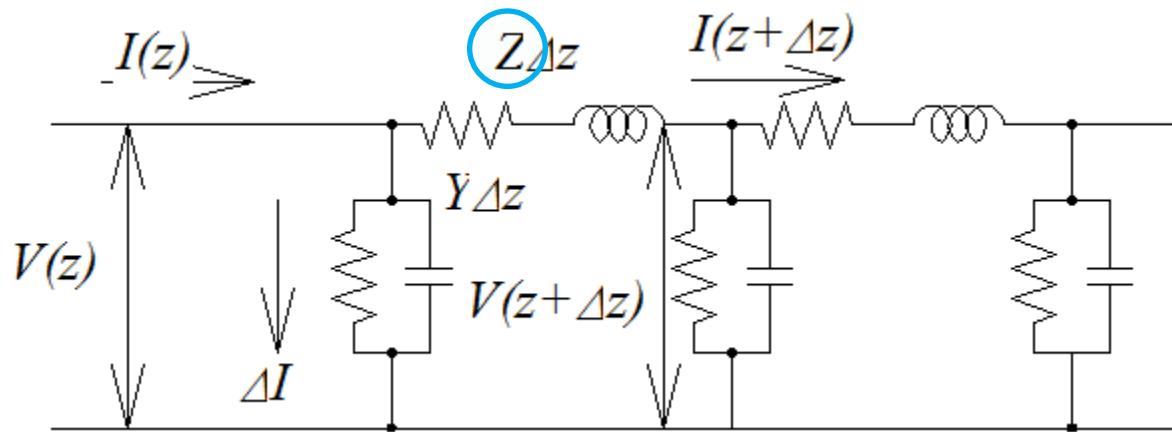
# In this tutorial

- **Review of transmission line theory**
  - S-Parameter, (V)SWR
  - Time domain behavior (TDR)
- **Frequency-domain simulation**
  - Ansys HFSS
- **Time-domain simulation**
  - GdfidL
- **Design, evaluation of the button electrodes**
- **Design, evaluation of the high-power feedthroughs**
- **Summary**

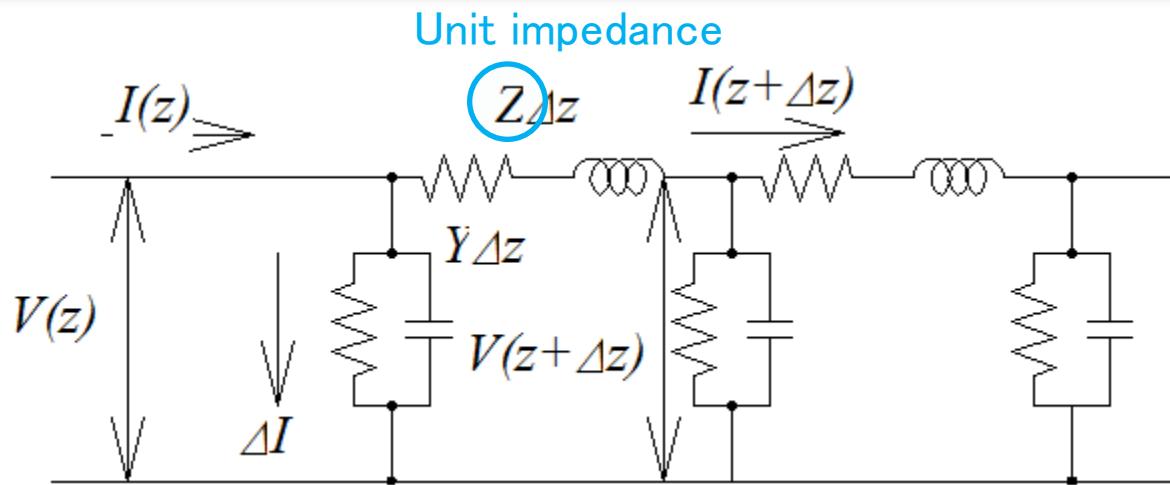
# Circuit representation of a uniform transmission line



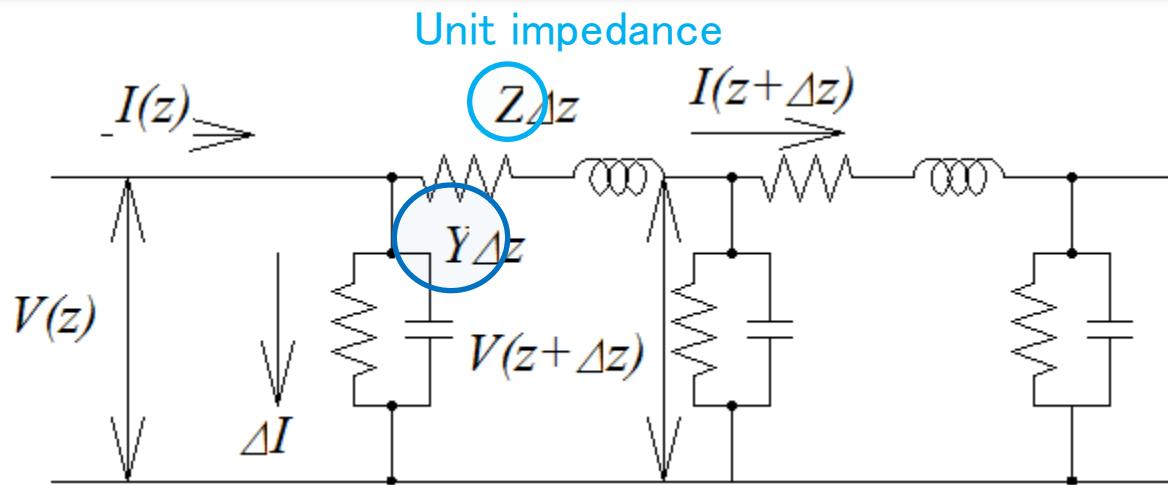
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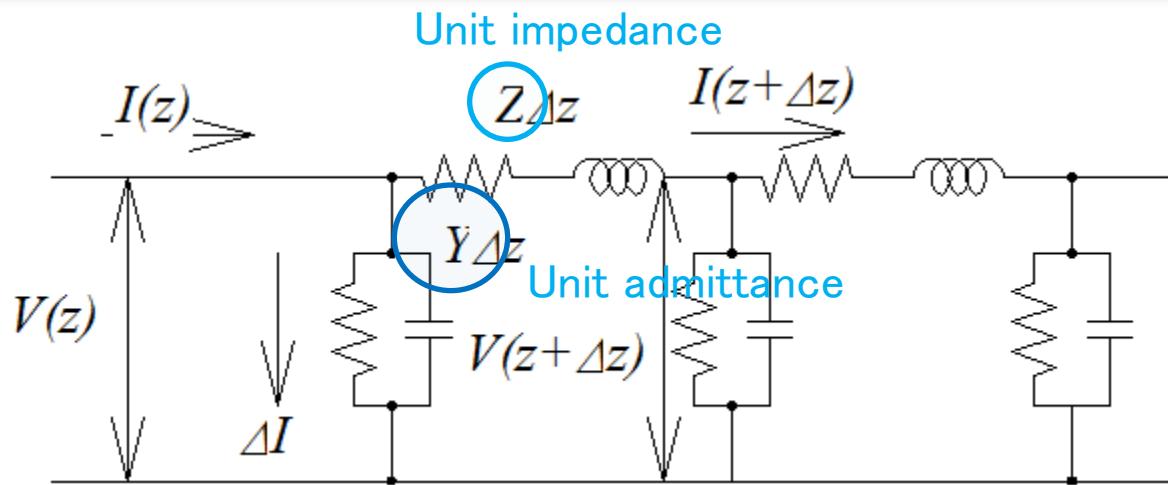
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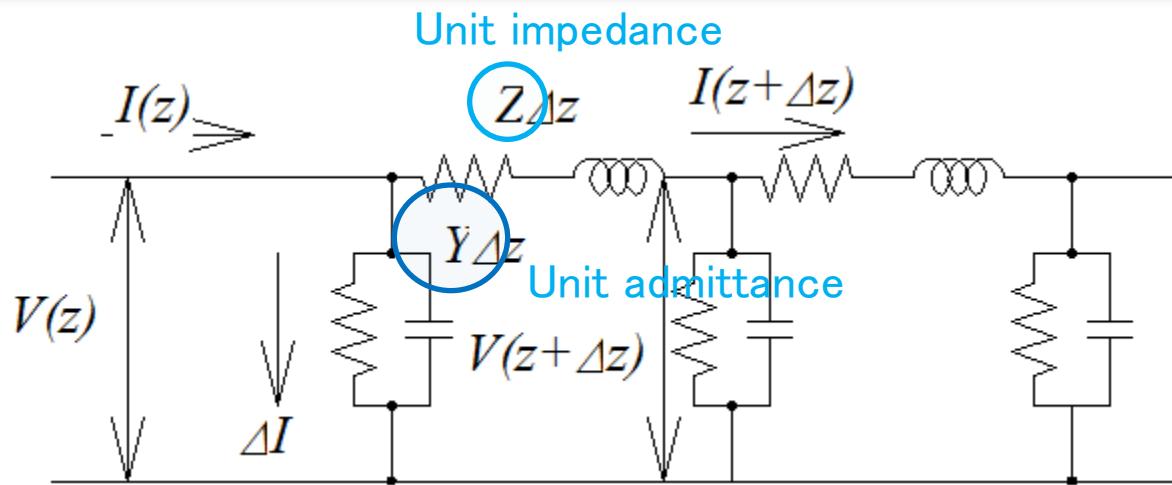
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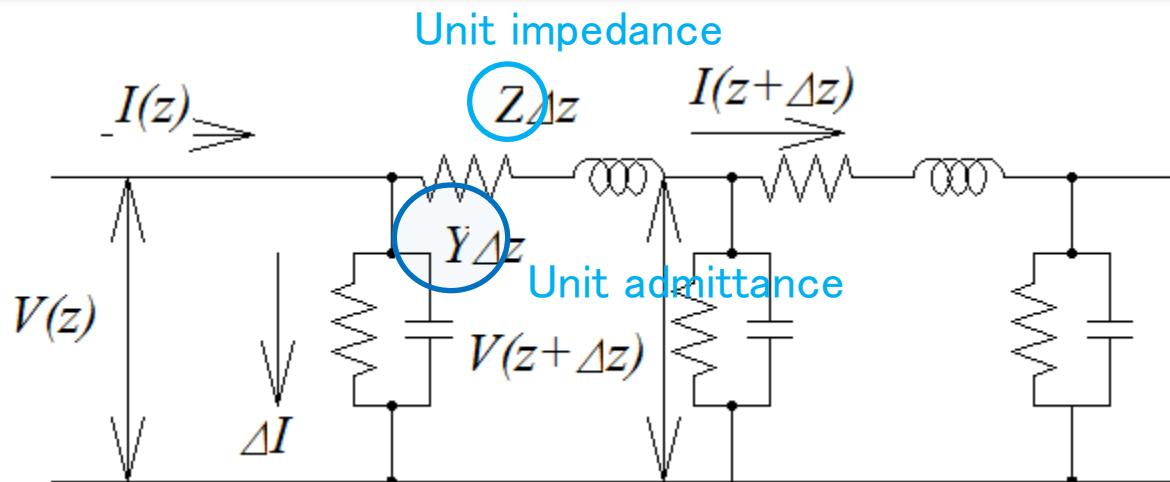
# Circuit representation of a uniform transmission line



$$\Delta V = -IZ\Delta z$$

$$\Delta I = -VY\Delta z$$

# Circuit representation of a uniform transmission line



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$$\frac{dV}{dz} = -IZ$$

$$\frac{dI}{dz} = -VY$$

$$\frac{d^2V}{dz^2} = \gamma^2 V, \frac{d^2I}{dz^2} = \gamma^2 I$$

$$\gamma^2 = ZY$$

# Transmission line (cont.)

- **Solution**

$$\begin{aligned} V &= V_1 e^{-\gamma z} + V_2 e^{\gamma z} \\ &= V_1 e^{(j\omega t - \gamma z)} + V_2 e^{(j\omega t + \gamma z)} \end{aligned}$$

- **Normal transmission line**

$$Z = R + jL\omega$$

$$Y = G + jC\omega$$

$$\gamma = \left( LC\omega^2 \right)^{\frac{1}{2}} \left[ 1 - \frac{RG}{LC\omega^2} - j \left( \frac{G}{C\omega} + \frac{R}{L\omega} \right) \right]^{\frac{1}{2}}$$

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Resistance   Inductance

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$$R / L\omega \ll 1$$

## Normal transmission line

$$G / C\omega \ll 1$$

Resistance Inductance

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Parallel conductance

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# Propagation constant

$$\gamma = \alpha + j\beta$$

$$\alpha = \frac{R}{2} \sqrt{\frac{C}{L}}, \beta = \omega \sqrt{LC}$$

$$V = V_1 e^{j(\omega t - \beta z)} e^{-\alpha z} + V_2 e^{j(\omega t + \beta z)} e^{\alpha z}$$

- **Phase velocity**

$$v_p = \frac{dz}{dt} = \frac{\omega}{\beta} = \frac{1}{\sqrt{LC}}$$

$$\beta = \frac{2\pi f}{f\lambda} = \frac{2\pi}{\lambda}$$

# Propagation constant

Attenuation constant

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Attenuation constant  
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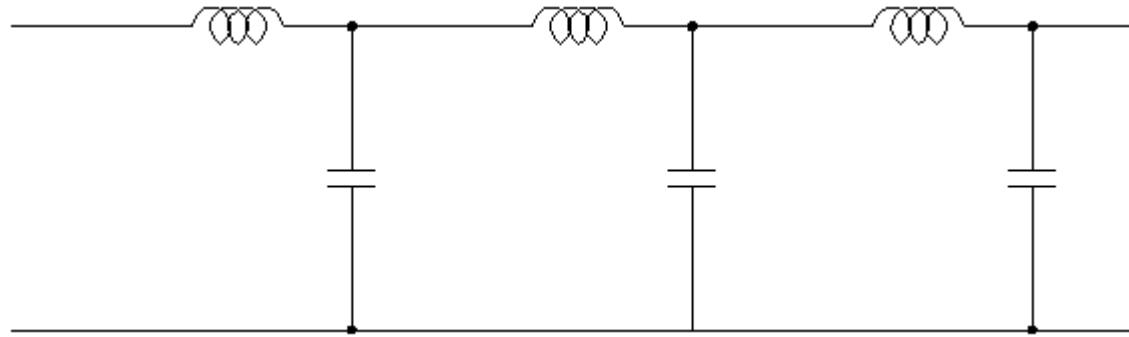
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# Loss less line



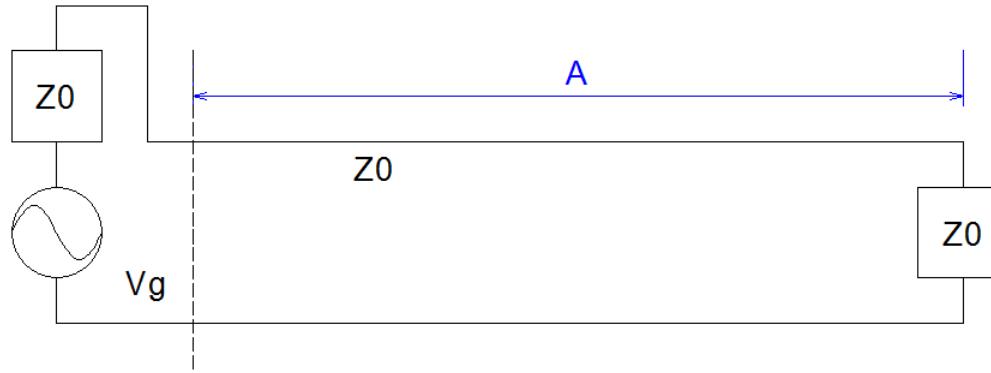
$$V = V_1 e^{-j\beta z} + V_2 e^{j\beta z}$$

$$I = \frac{1}{Z_0} (V_1 e^{-j\beta z} - V_2 e^{j\beta z})$$

$$Z_0 = \frac{Z}{\gamma} = \sqrt{\frac{L}{C}}$$

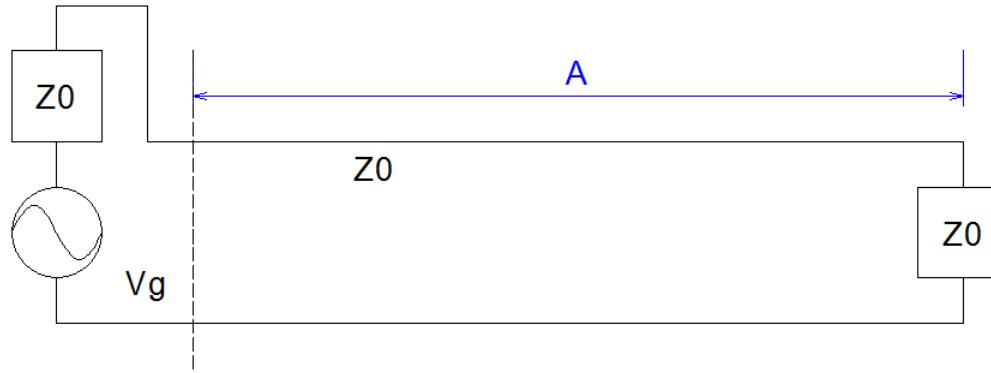
$$\beta = \omega \sqrt{LC}$$

# Impedance matched line



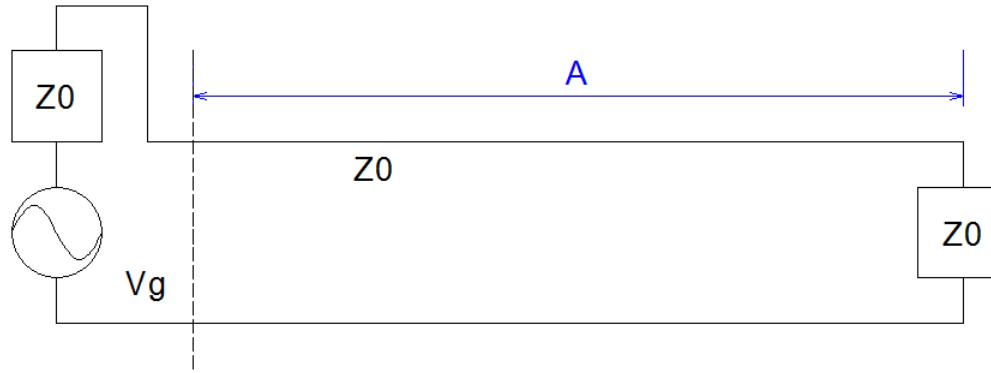
$$\frac{V(A)}{I(A)} = Z_0 = Z_0 \frac{V_1 e^{-j\beta A} + V_2 e^{j\beta A}}{V_1 e^{-j\beta A} - V_2 e^{j\beta A}}$$

# Impedance matched line



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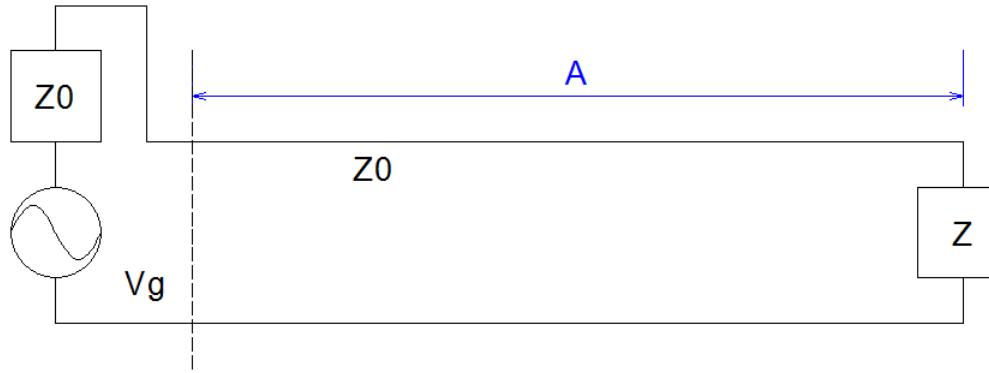
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- No reflection! (Only forwarding wave exist)

# Arbitrary termination

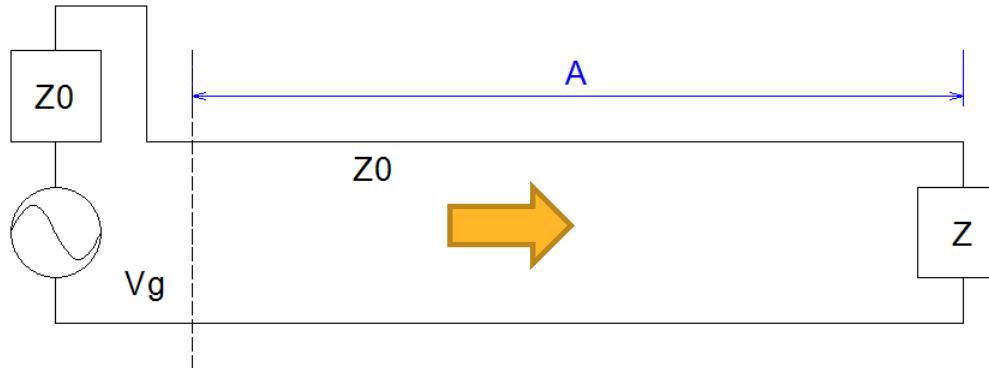


- To satisfy  $V(A)/I(A)=Z$ , we need both forwarding and reflecting wave.

$$V = V_1 e^{-j\beta z} \left( 1 + 2 \frac{|V_2|}{|V_1|} \cos \chi + \frac{|V_2|^2}{|V_1|^2} \right)^{\frac{1}{2}} \exp \left( j \tan^{-1} \frac{\frac{|V_2|}{|V_1|} \sin \chi}{1 + \frac{|V_2|}{|V_1|} \cos \chi} \right)$$

$$\chi = 2\beta z + \theta_2 - \theta_1$$

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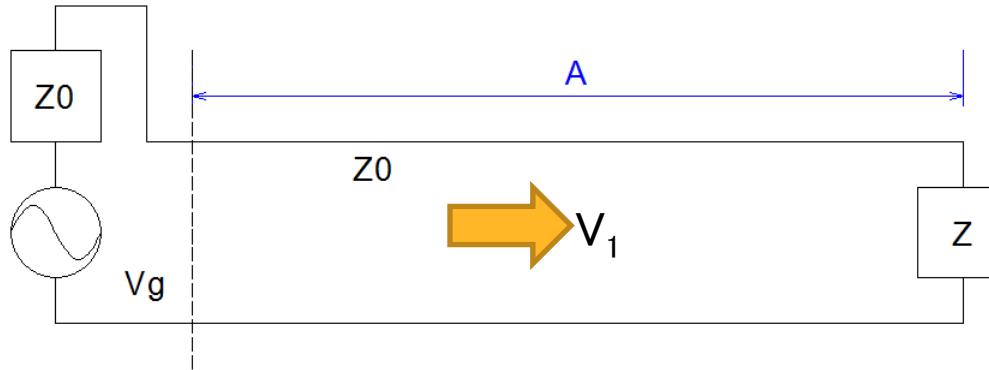


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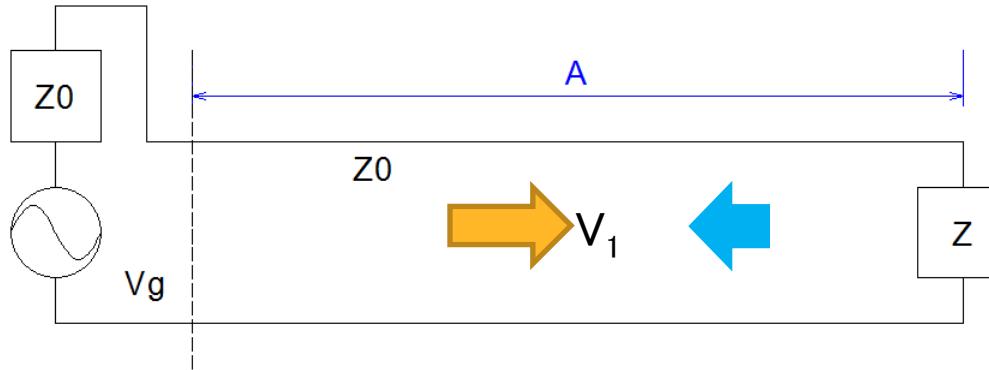


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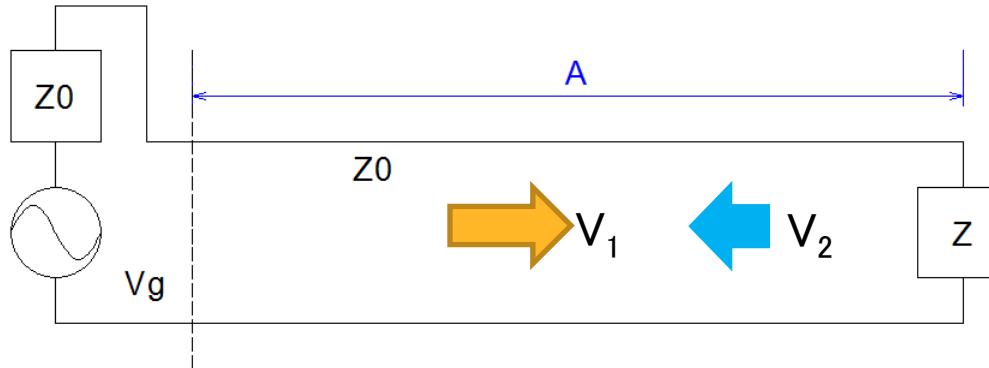


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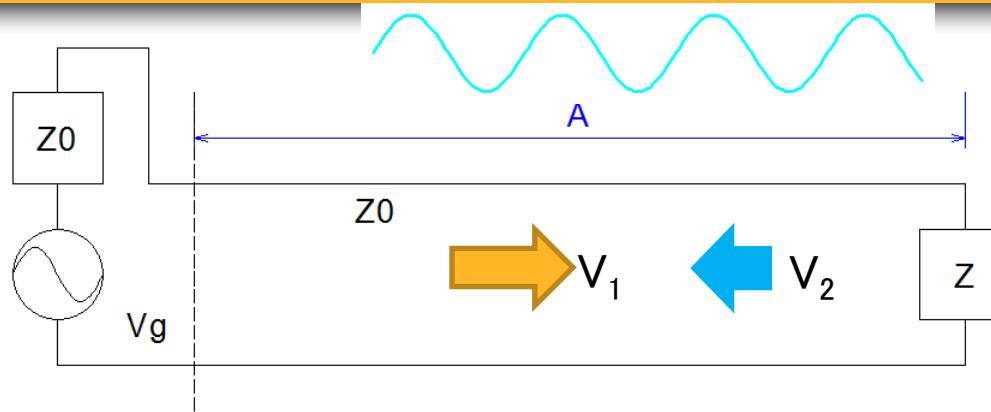


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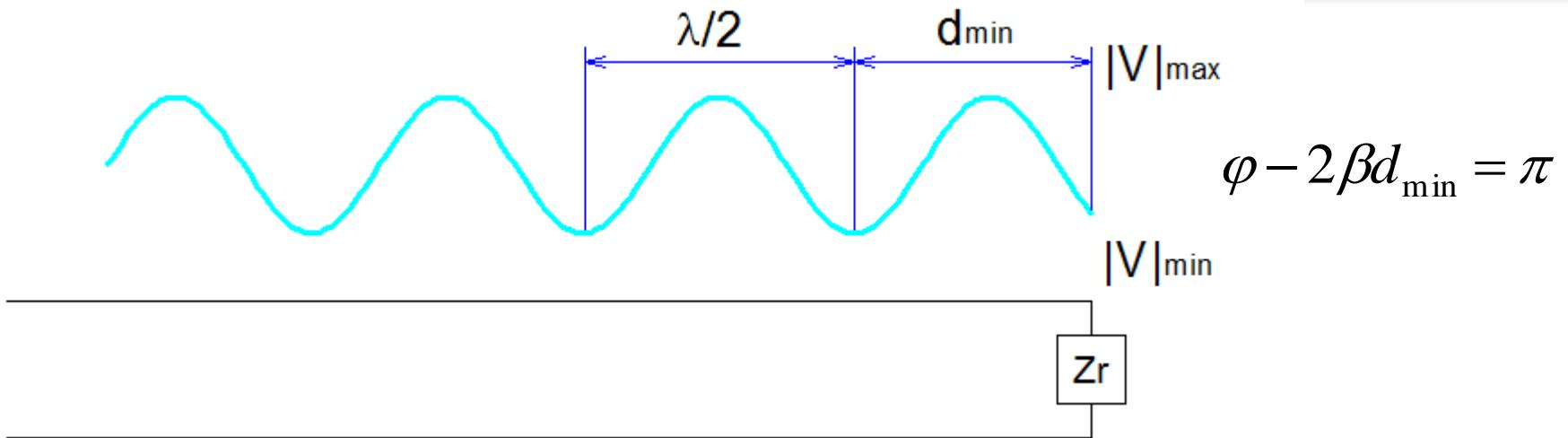


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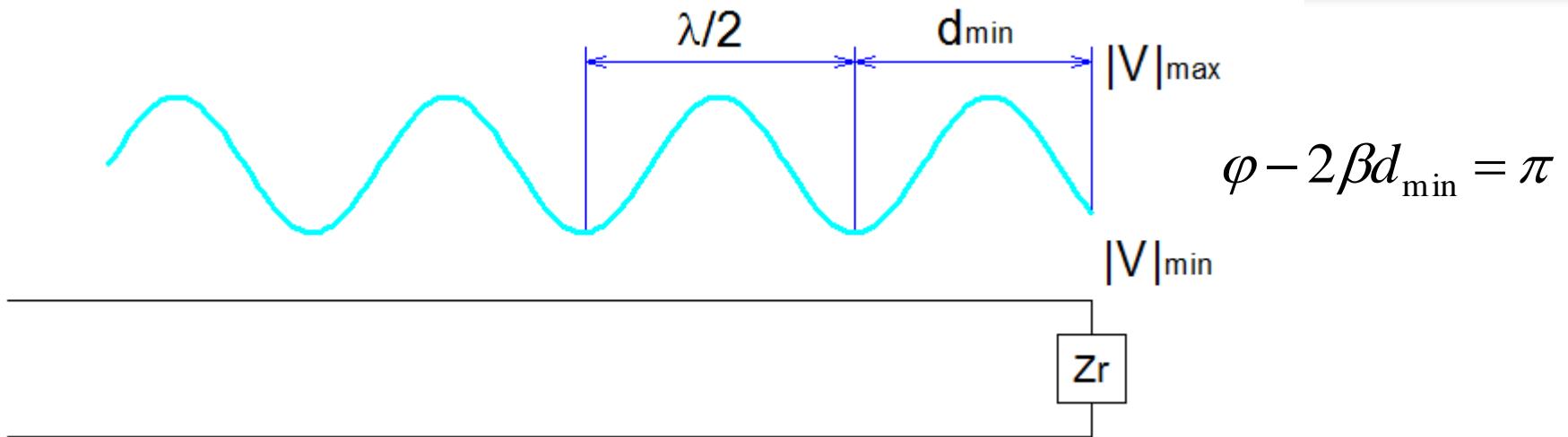
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# Standing wave ratio

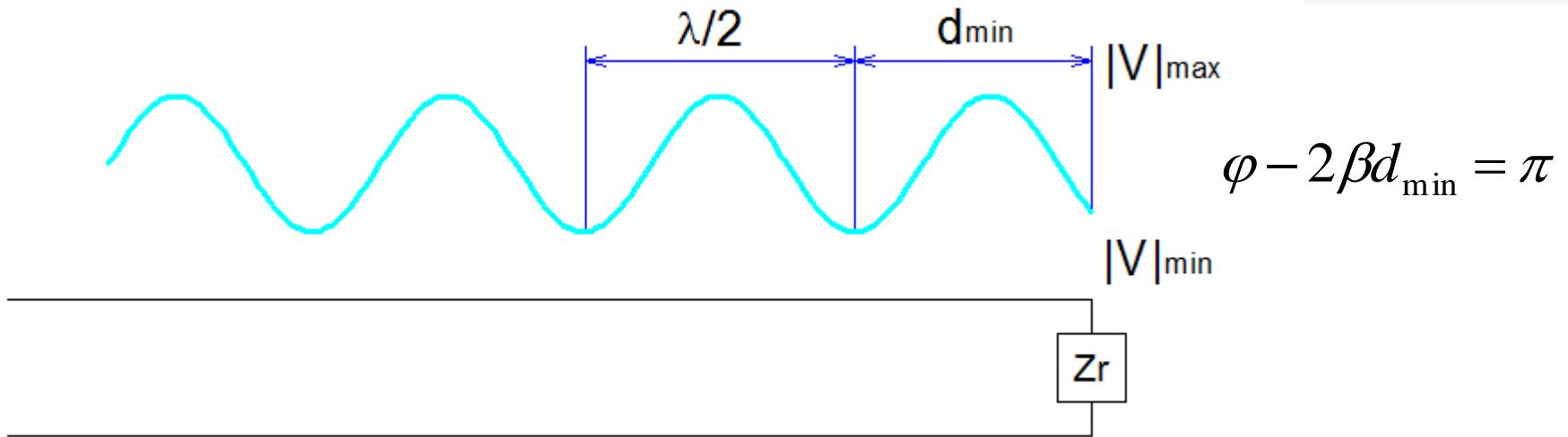


# Standing wave ratio



- Standing wave ratio (SWR)

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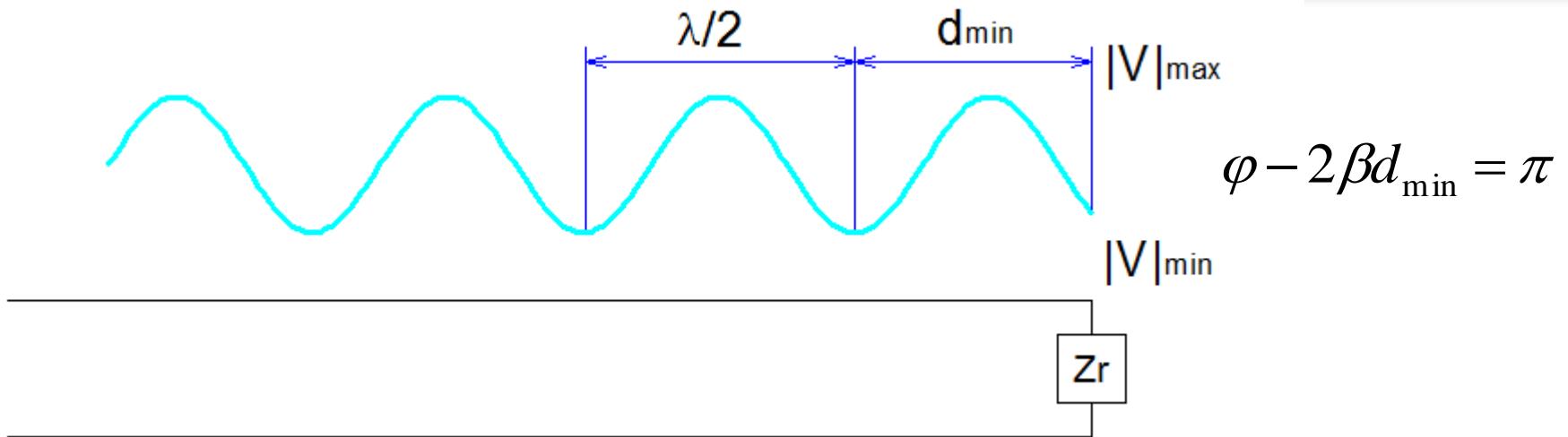


- Standing wave ratio (SWR)

$$S = \frac{|V|_{\max}}{|V|_{\min}} = \frac{1+|\rho|}{1-|\rho|}$$

$$|\rho| = \frac{S-1}{S+1}$$

# Standing wave ratio



- Standing wave ratio (SWR)

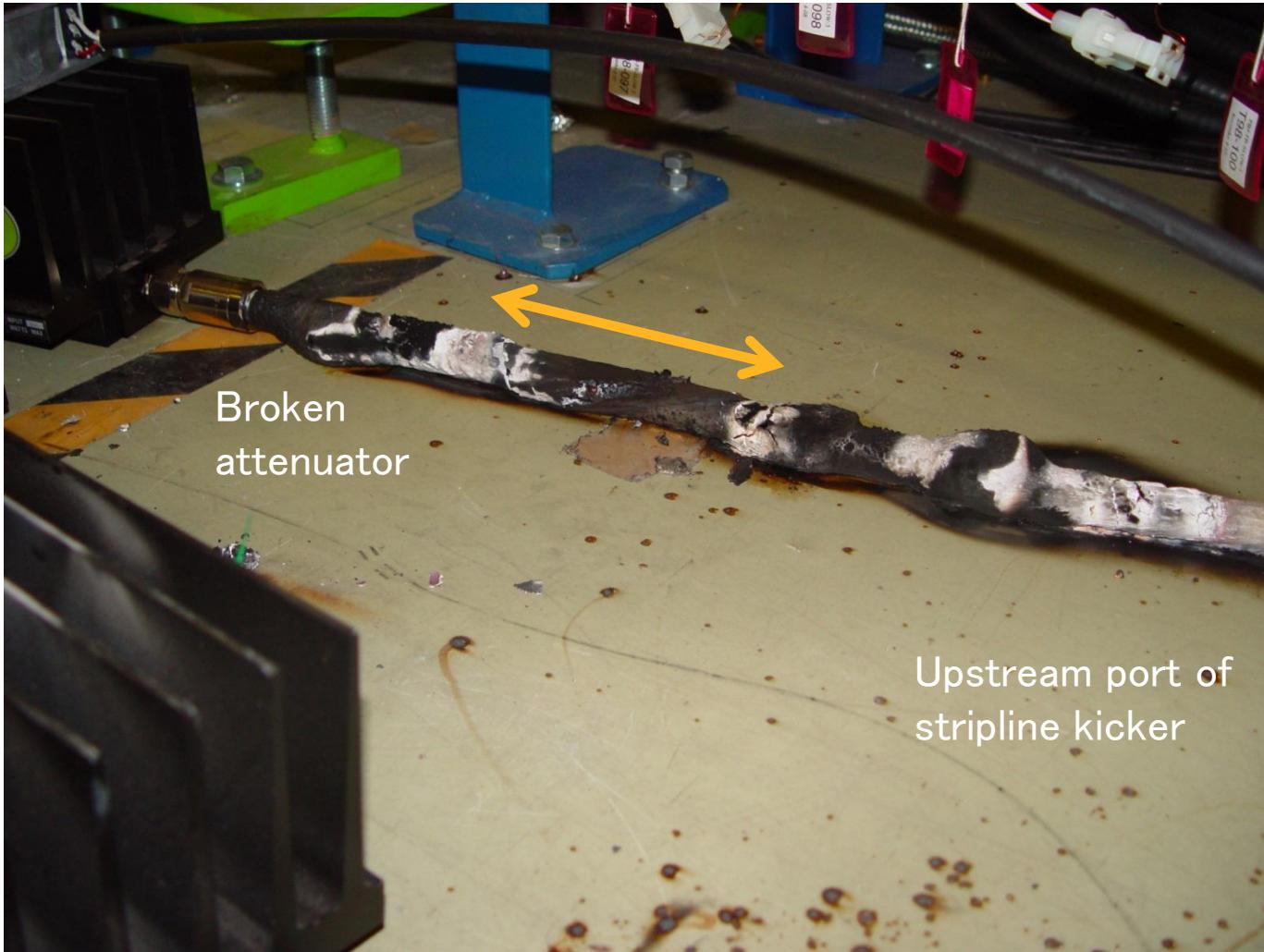
$$S = \frac{|V|_{\max}}{|V|_{\min}} = \frac{1+|\rho|}{1-|\rho|} \quad \text{ρ: reflection constant}$$

$$|\rho| = \frac{S-1}{S+1}$$

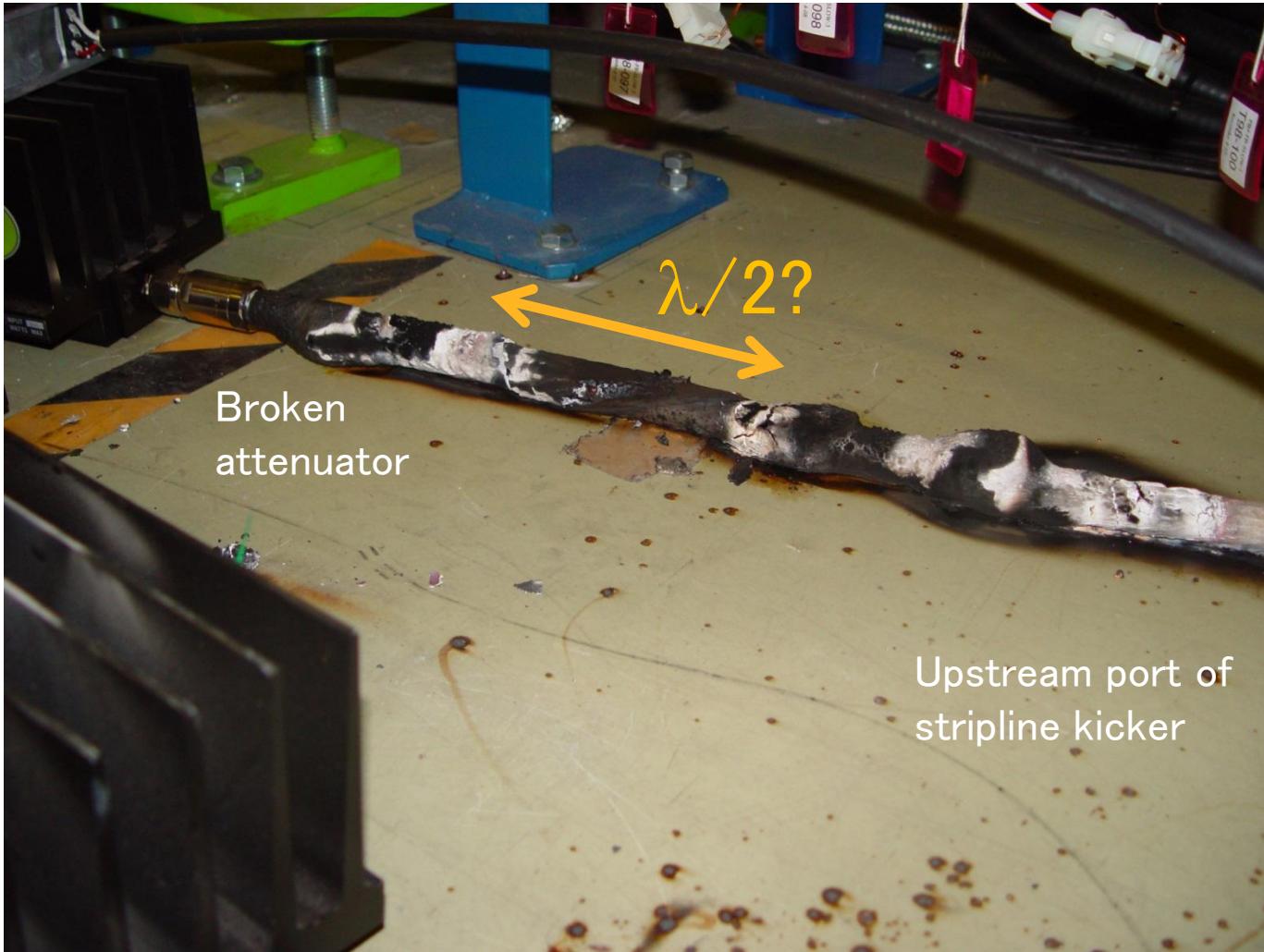
# Example of standing wave!



# Example of standing wave!



# Example of standing wave!



# Impedance of the load

- We can estimate the load impedance using

$$Z_r = Z_0 \frac{1 + \rho}{1 - \rho}$$

$$\begin{aligned} z_r &= \frac{Z_r}{Z_0} = \frac{1 + |\rho| e^{j\varphi}}{1 - |\rho| e^{j\varphi}} = \frac{(S+1) + (S-1)e^{j(2\beta d_{\min} + \pi)}}{(S+1) - (S-1)e^{j(2\beta d_{\min} + \pi)}} \\ &= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}} \end{aligned}$$

# Impedance of the load

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$$Z_r = Z_0 \frac{1+\rho}{1-\rho}$$

$$\begin{aligned} z_r &= \frac{Z_r}{Z_0} = \frac{1+|\rho|e^{j\varphi}}{1-|\rho|e^{j\varphi}} = \frac{(S+1)+(S-1)e^{j(2\beta d_{\min} + \pi)}}{(S+1)-(S-1)e^{j(2\beta d_{\min} + \pi)}} \\ &= \frac{1-jS \tan \beta d_{\min}}{S-j \tan \beta d_{\min}} \end{aligned}$$

# Impedance of the load

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Reflection  
constant

$$\begin{aligned} z_r &= \frac{Z_r}{Z_0} = \frac{1 + |\rho| e^{j\varphi}}{1 - |\rho| e^{j\varphi}} = \frac{(S+1) + (S-1)e^{j(2\beta d_{\min} + \pi)}}{(S+1) - (S-1)e^{j(2\beta d_{\min} + \pi)}} \\ &= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}} \end{aligned}$$

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# Impedance of the load

- We can estimate the load impedance using

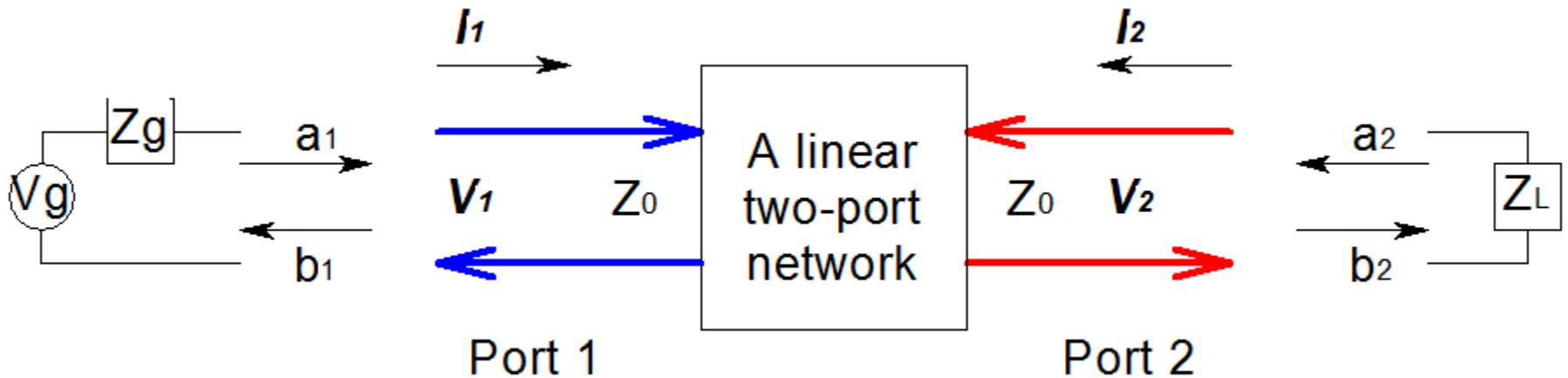
$$Z_r = Z_0 \frac{1 + \rho}{1 - \rho}$$

Reflection  
constant

SWR

$$z_r = \frac{Z_r}{Z_0} = \frac{1 + |\rho| e^{j\varphi}}{1 - |\rho| e^{j\varphi}} = \frac{(S+1) + (S-1)e^{j(2\beta d_{\min} + \pi)}}{(S+1) - (S-1)e^{j(2\beta d_{\min} + \pi)}}$$
$$= \frac{1 - jS \tan \beta d_{\min}}{S - j \tan \beta d_{\min}}$$

# Scattering Matrix(S-Parameters)



$$\begin{pmatrix} \mathbf{b}_1 \\ \mathbf{b}_2 \end{pmatrix} = \begin{pmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{pmatrix} \begin{pmatrix} \mathbf{a}_1 \\ \mathbf{a}_2 \end{pmatrix}$$

# S-parameters

- **Input Reflection Coefficient with  $Z_L=Z_0$**

$$S_{11} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

- **Output Reflection Coefficient with  $Z_G=Z_0$  and  $V_G=0$**

$$S_{22} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_2} \right|_{\mathbf{a}_1=0}$$

- **Forward Transmission Coefficient with  $Z_L=Z_0$**

$$S_{21} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

- **Reverse Transmission Coefficient with  $Z_G=Z_0$  and  $V_G=0$**

$$S_{12} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

# S-parameters

- Input Reflection Coefficient with  $Z_L = Z_0$

$$S_{11} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Input match

- Output Reflection Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{22} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_2} \right|_{\mathbf{a}_1=0}$$

- Forward Transmission Coefficient with  $Z_L = Z_0$

$$S_{21} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

- Reverse Transmission Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{12} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

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$$S_{11} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Input match

- Output Reflection Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{22} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_2} \right|_{\mathbf{a}_1=0}$$

Output match

- Forward Transmission Coefficient with  $Z_L = Z_0$

$$S_{21} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

- Reverse Transmission Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{12} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

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Input match

- Output Reflection Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{22} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_2} \right|_{\mathbf{a}_1=0}$$

Output match

- Forward Transmission Coefficient with  $Z_L = Z_0$

$$S_{21} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Gain or loss

- Reverse Transmission Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{12} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

# S-parameters

- Input Reflection Coefficient with  $Z_L = Z_0$

$$S_{11} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Input match

- Output Reflection Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

$$S_{22} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_2} \right|_{\mathbf{a}_1=0}$$

Output match

- Forward Transmission Coefficient with  $Z_L = Z_0$

$$S_{21} = \left. \frac{\mathbf{b}_2}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Gain or loss

- Reverse Transmission Coefficient with  $Z_G = Z_0$  and  $V_G = 0$

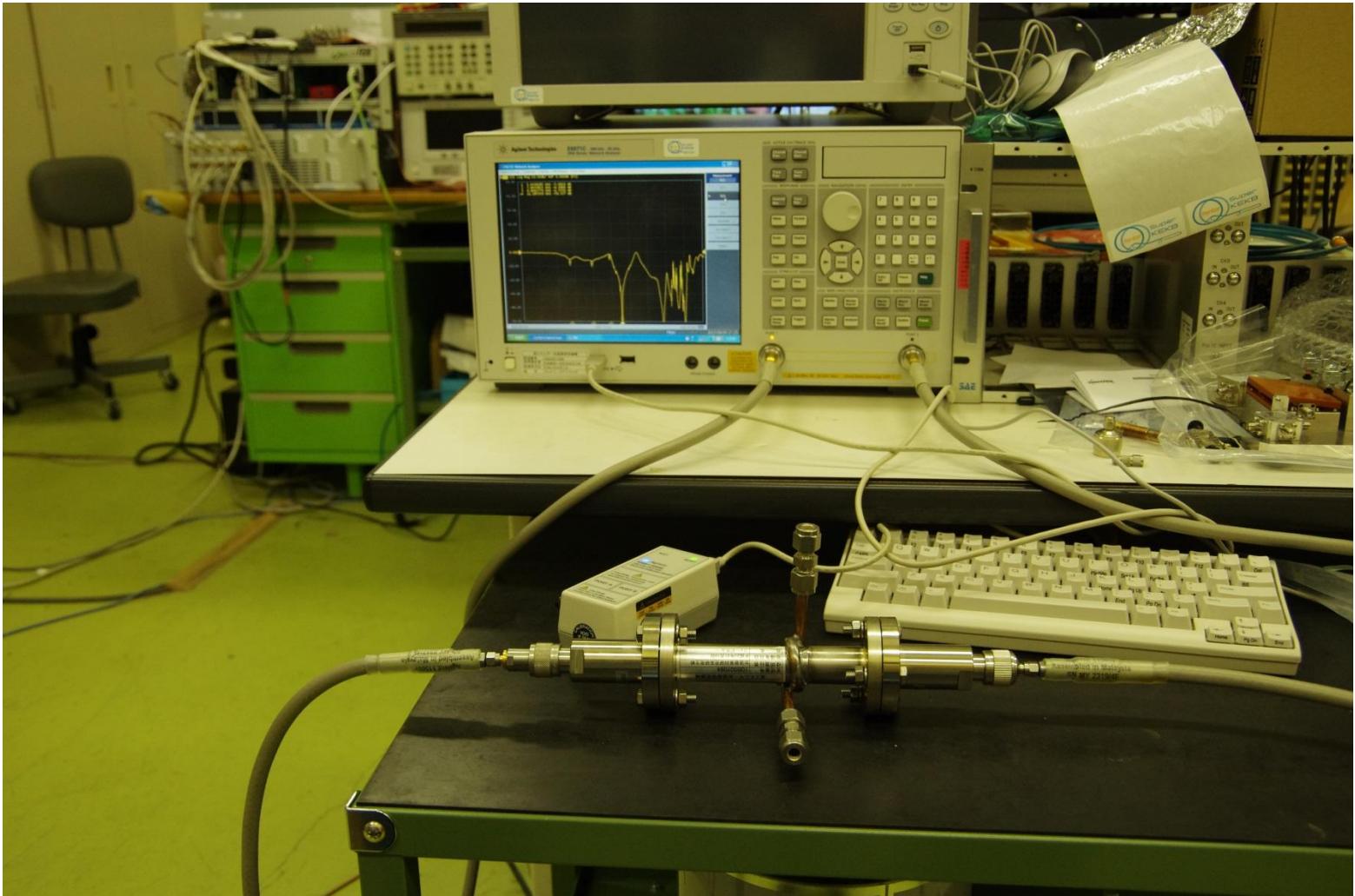
$$S_{12} = \left. \frac{\mathbf{b}_1}{\mathbf{a}_1} \right|_{\mathbf{a}_2=0}$$

Isolation

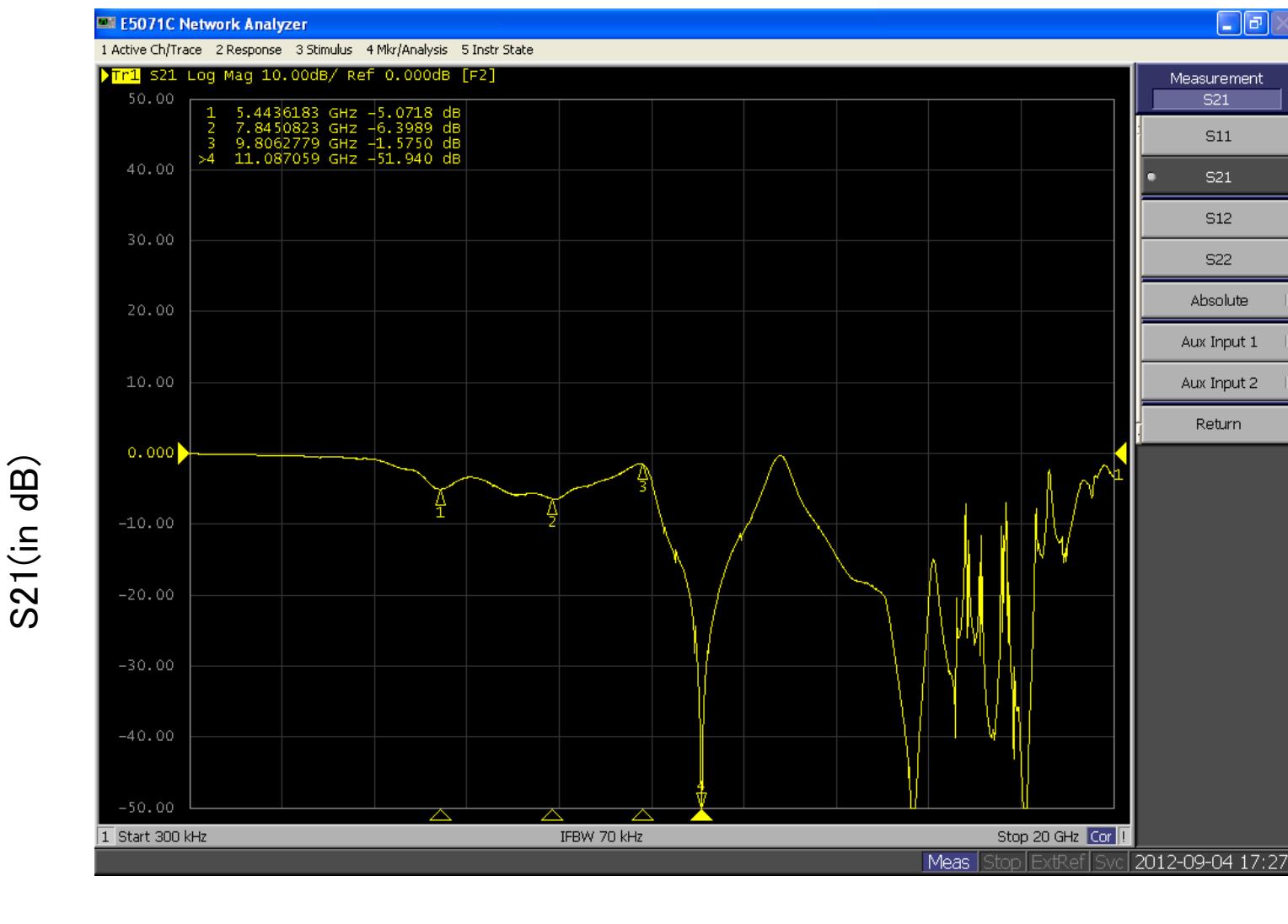
# Why Use S-Parameters?

- **relatively easy to obtain at high frequencies**
  - measure voltage traveling waves with a vector network analyzer
  - don't need shorts/opens which can cause active devices to oscillate or self-destruct
- **relate to familiar measurements (gain, loss, reflection coefficients...)**
- **can cascade S-parameters of multiple devices to predict system performance**
- **can compute H, Y or Z parameters from S-parameters if desired**
- **can easily import and use S-parameter files in electronic-simulation tools**

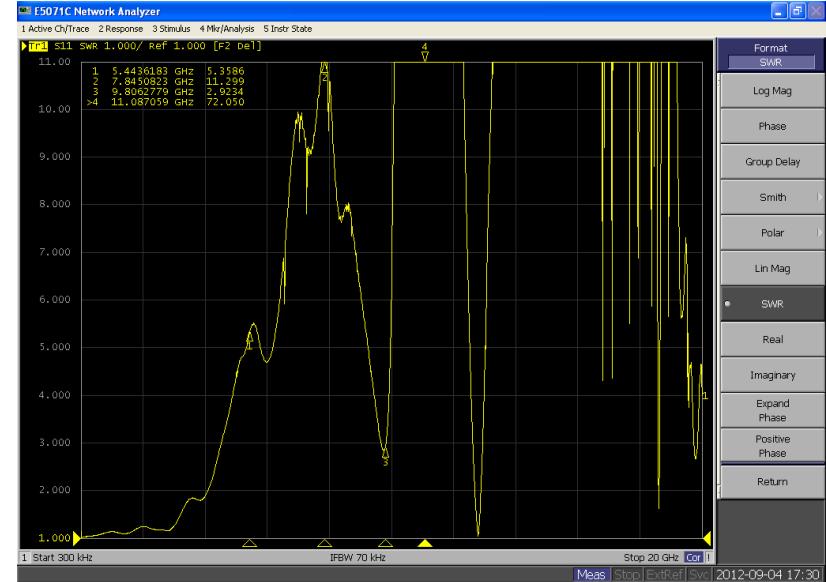
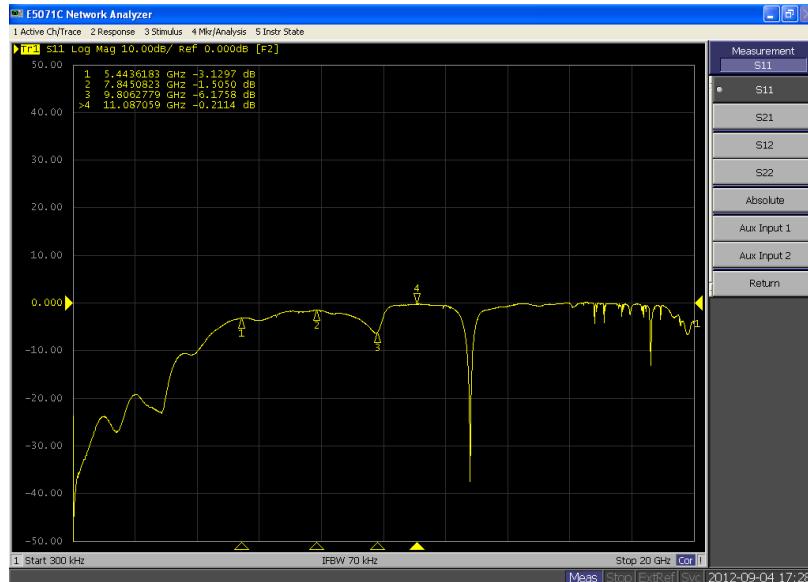
# RF Network Analyzer



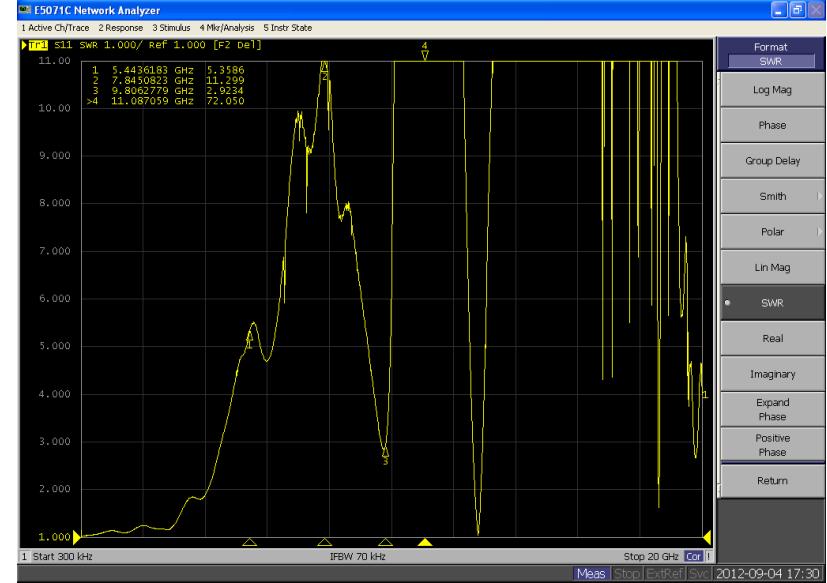
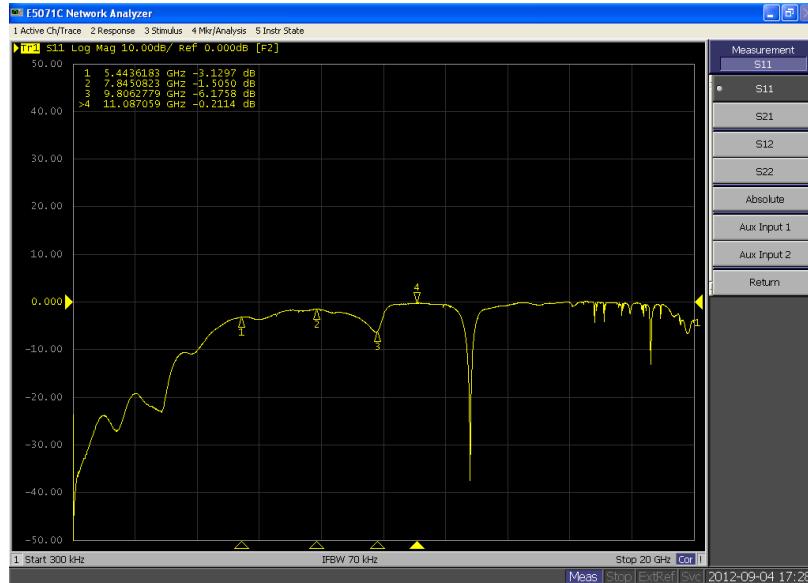
S21



# S11 and SWR(VSWR)

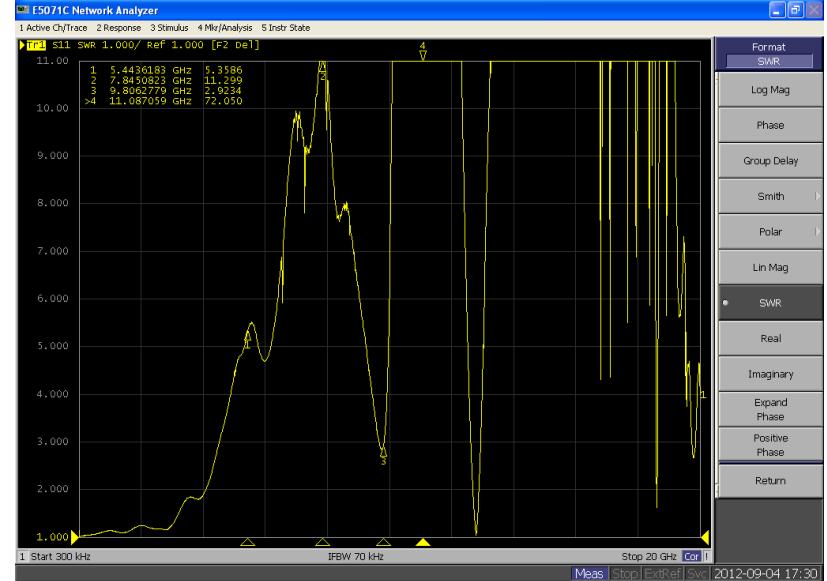
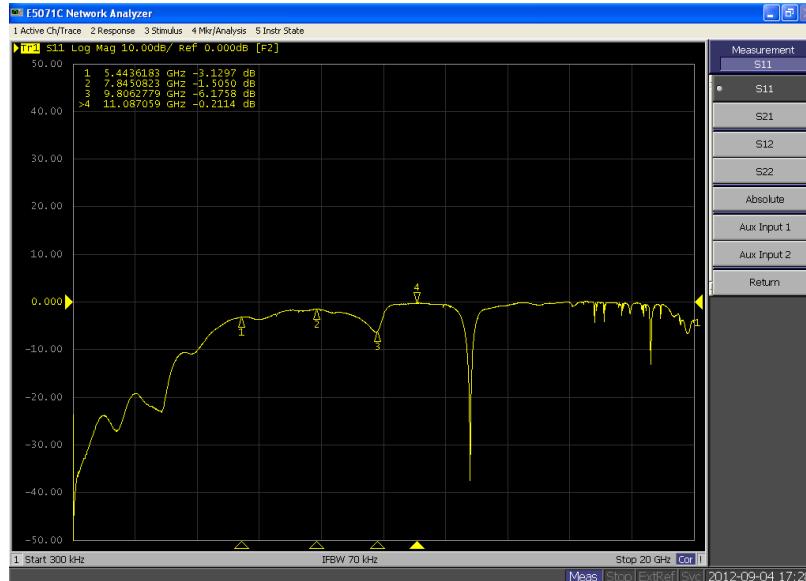


# S11 and SWR(VSWR)



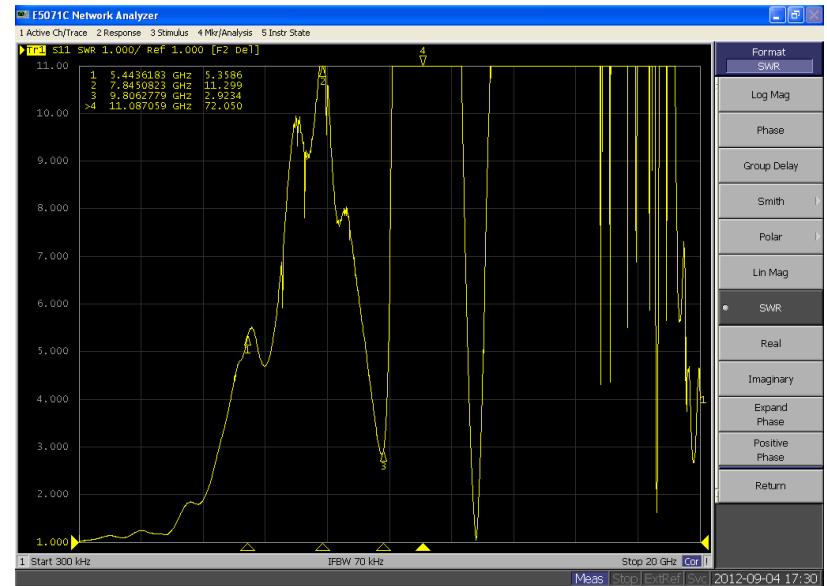
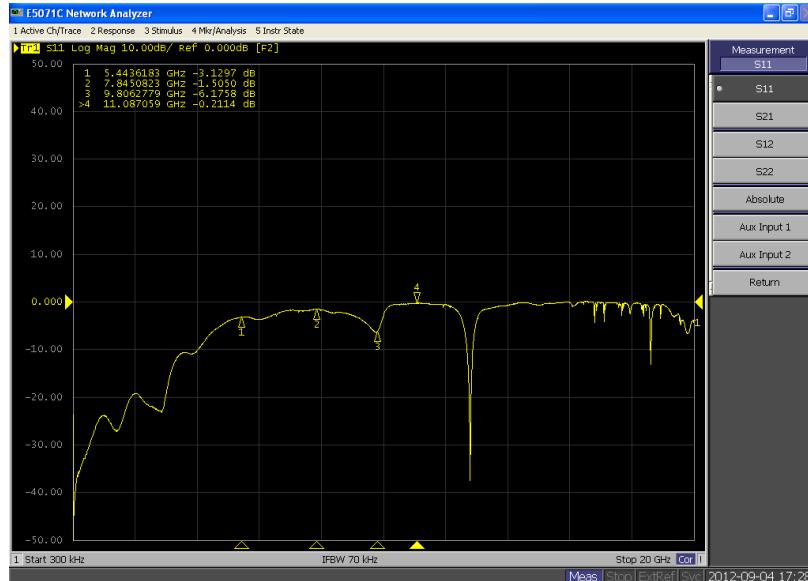
$$|20 \log_{10} |S_{11}| | \text{dB}$$

# S11 and SWR(VSWR)



- Return loss
- $$|20 \log_{10} |S_{11}|| \text{dB}$$

# S11 and SWR(VSWR)

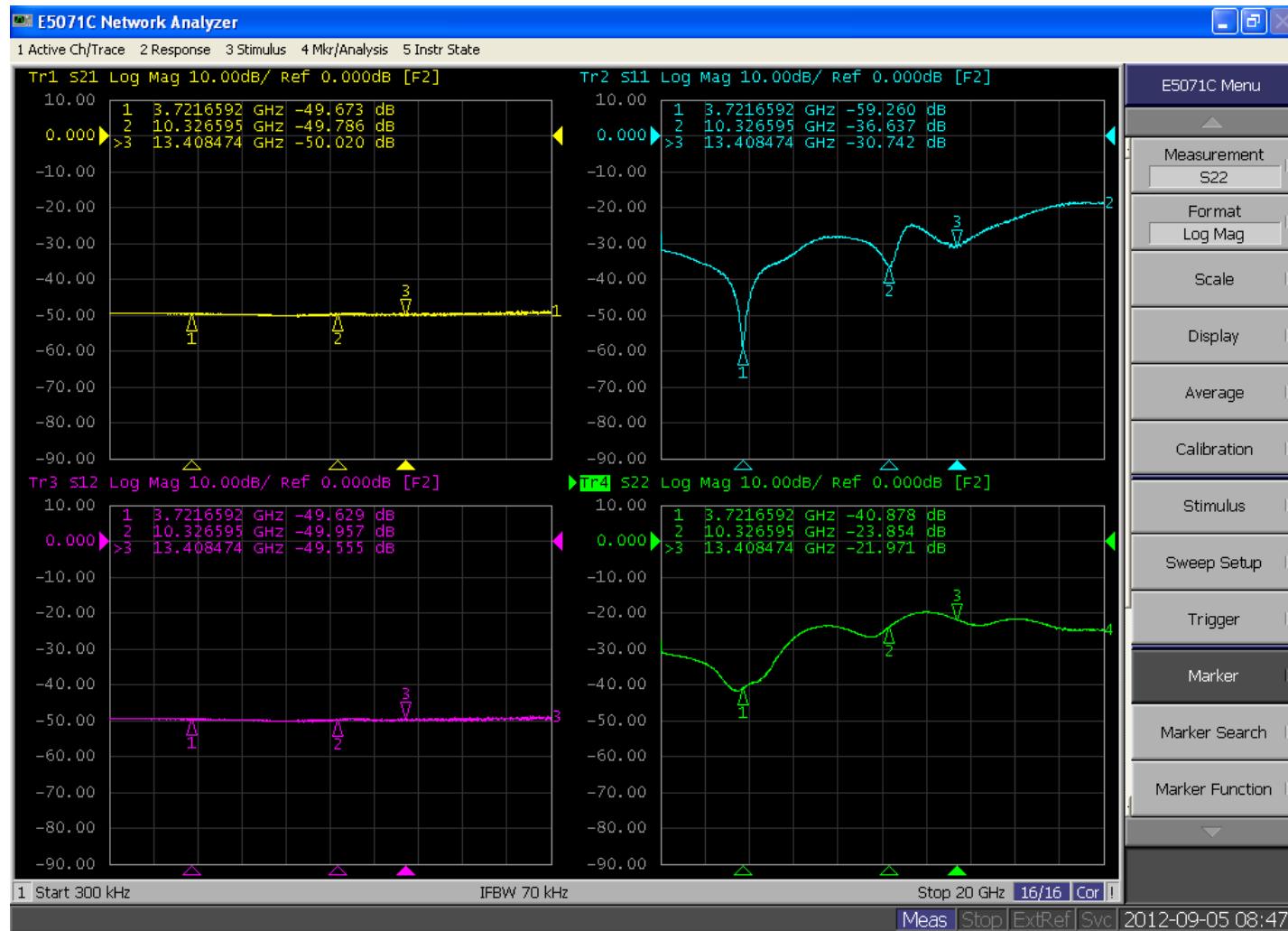


## Return loss

$$|20 \log_{10} |S_{11}|| \text{dB}$$

$$SWR_{input} = \frac{1 + |S_{11}|}{1 - |S_{11}|}$$

# Example (50dB attenuator)



# Example (800MHz Bessel LPF)



# Time domain behavior--TDR

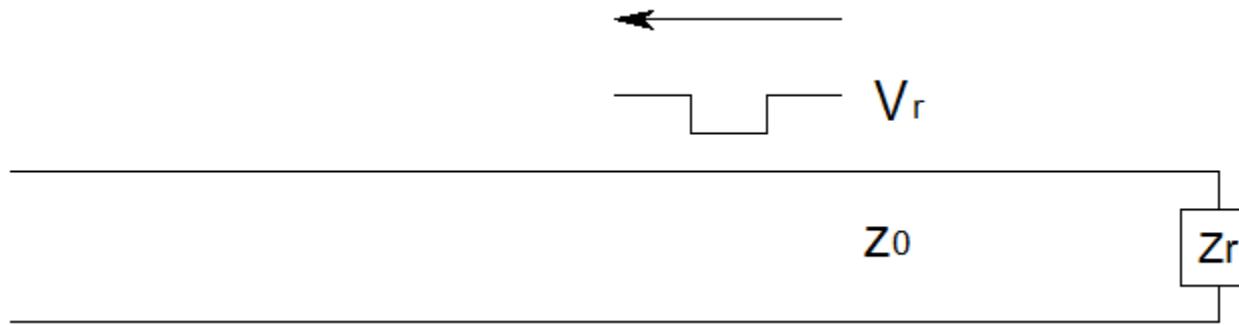


- Using voltage ratio between forward and reflecting wave

$$Z_r = \frac{1 + \rho}{1 - \rho} Z_0$$

$$\rho = \frac{V_r}{V_i}$$

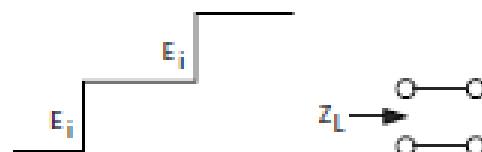
# Time domain behavior--TDR



- Using voltage ratio between forward and reflecting wave

$$Z_r = \frac{1+\rho}{1-\rho} Z_0$$

$$\rho = \frac{V_r}{V_i}$$



(A)  $E_r = E_i$

Therefore  $\frac{Z_L - Z_0}{Z_L + Z_0} = +1$

Which is true as  $Z_L \rightarrow \infty$

$\therefore Z = \text{Open Circuit}$

---

(A) Open Circuit Termination ( $Z_L = \infty$ )



(B)  $E_r = E_i$

Therefore  $\frac{Z_L - Z_0}{Z_L + Z_0} = -1$

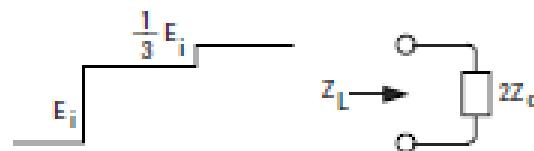
Which is only true for finite  $Z$

When  $Z_L = 0$

$\therefore Z = \text{Short Circuit}$

---

(B) Short Circuit Termination ( $Z_L = 0$ )



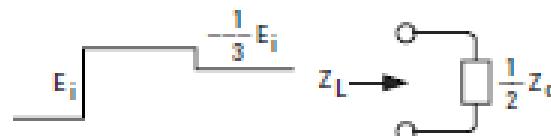
(C)  $E_r = +\frac{1}{3} E_i$

Therefore  $\frac{Z_L - Z_0}{Z_L + Z_0} = +\frac{1}{3}$

and  $Z_L = 2Z_0$

---

(C) Line Terminated in  $Z_L = 2Z_0$



(D) Line Terminated in  $Z_L = \frac{1}{2} Z_0$

(D)  $E_r = -\frac{1}{3} E_i$

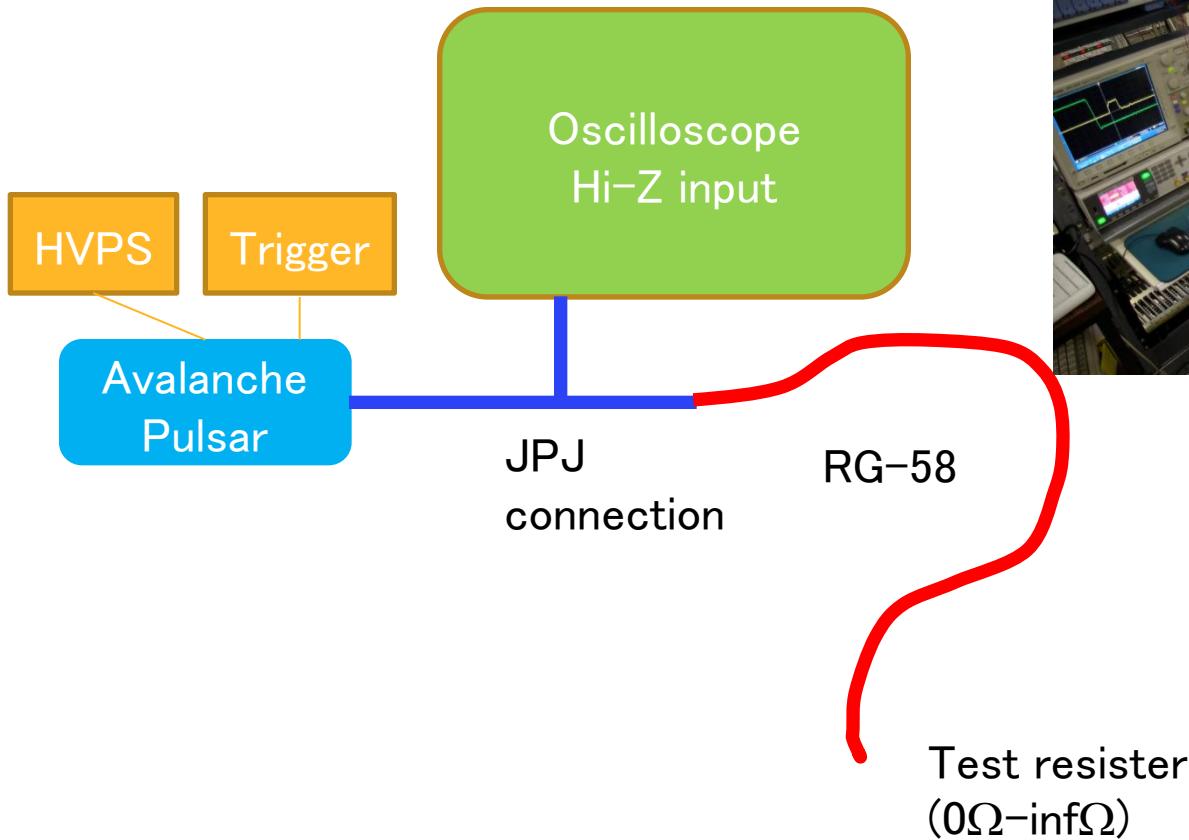
Therefore  $\frac{Z_L - Z_0}{Z_L + Z_0} = -\frac{1}{3}$

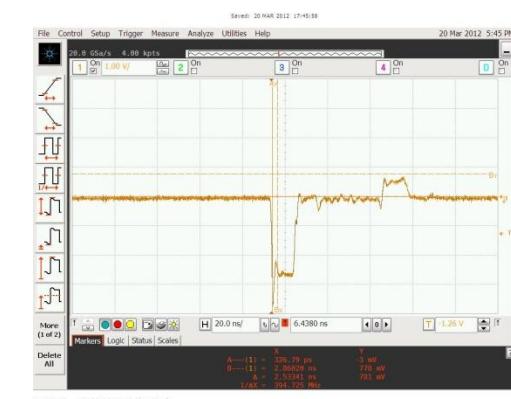
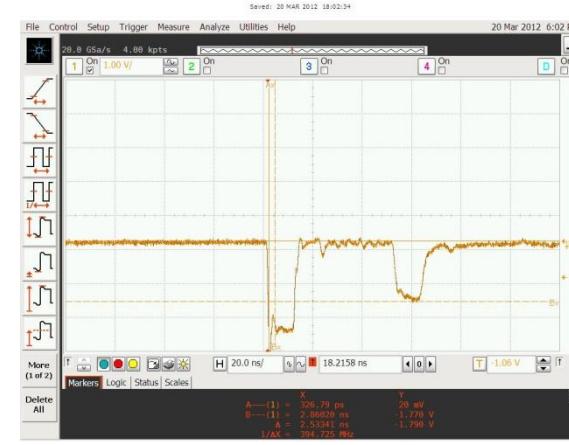
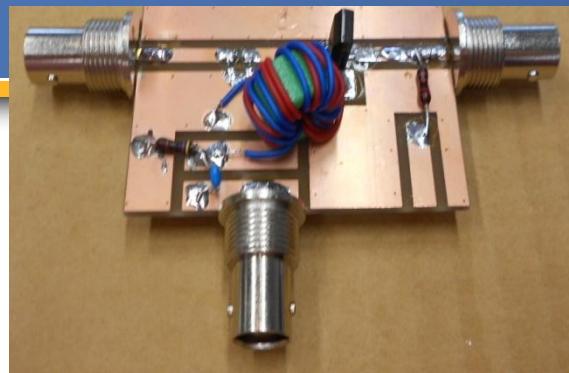
and  $Z_L = \frac{1}{2} Z_0$

---

# Science Camp 2011 (for High School students)







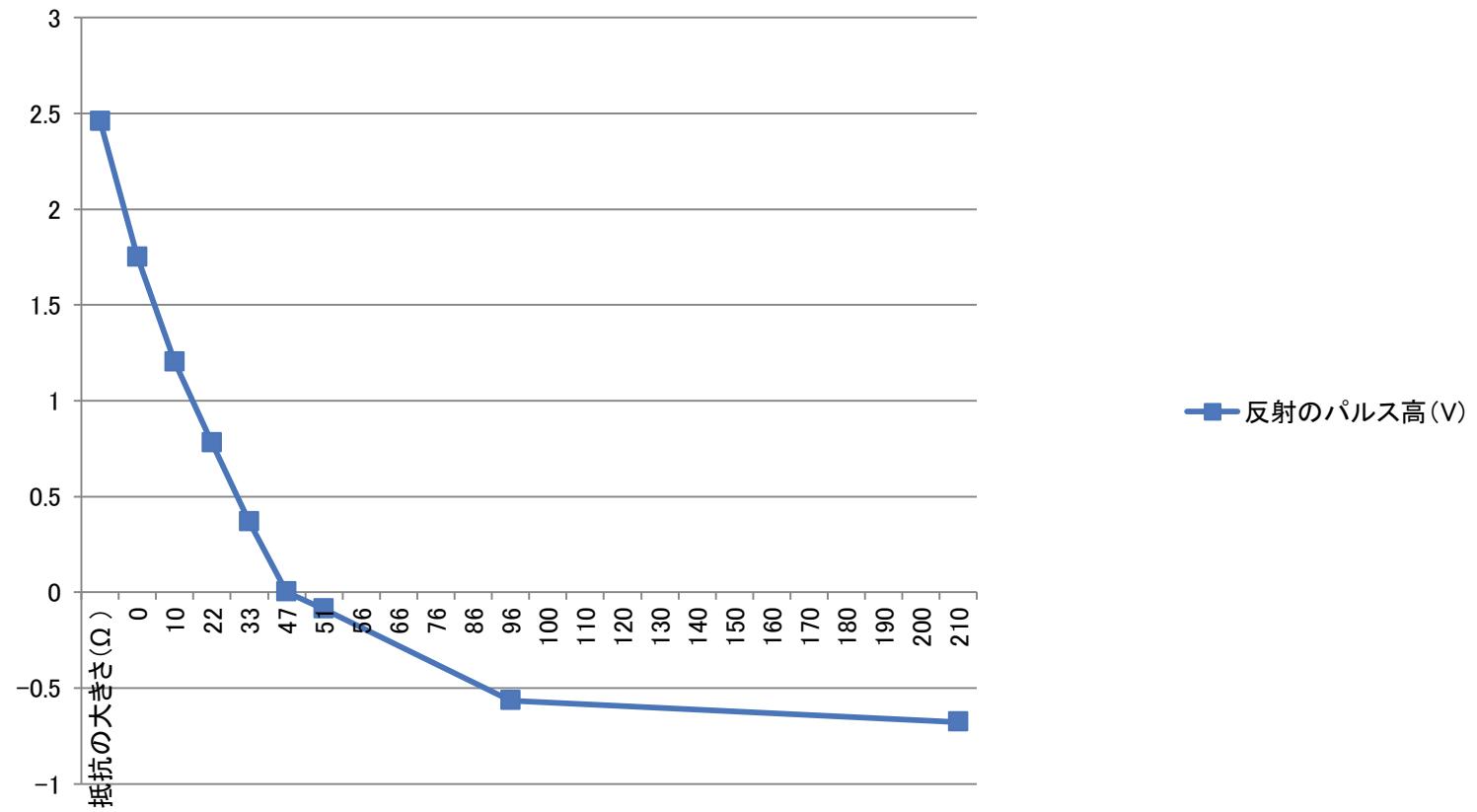
51Ω

470Ω

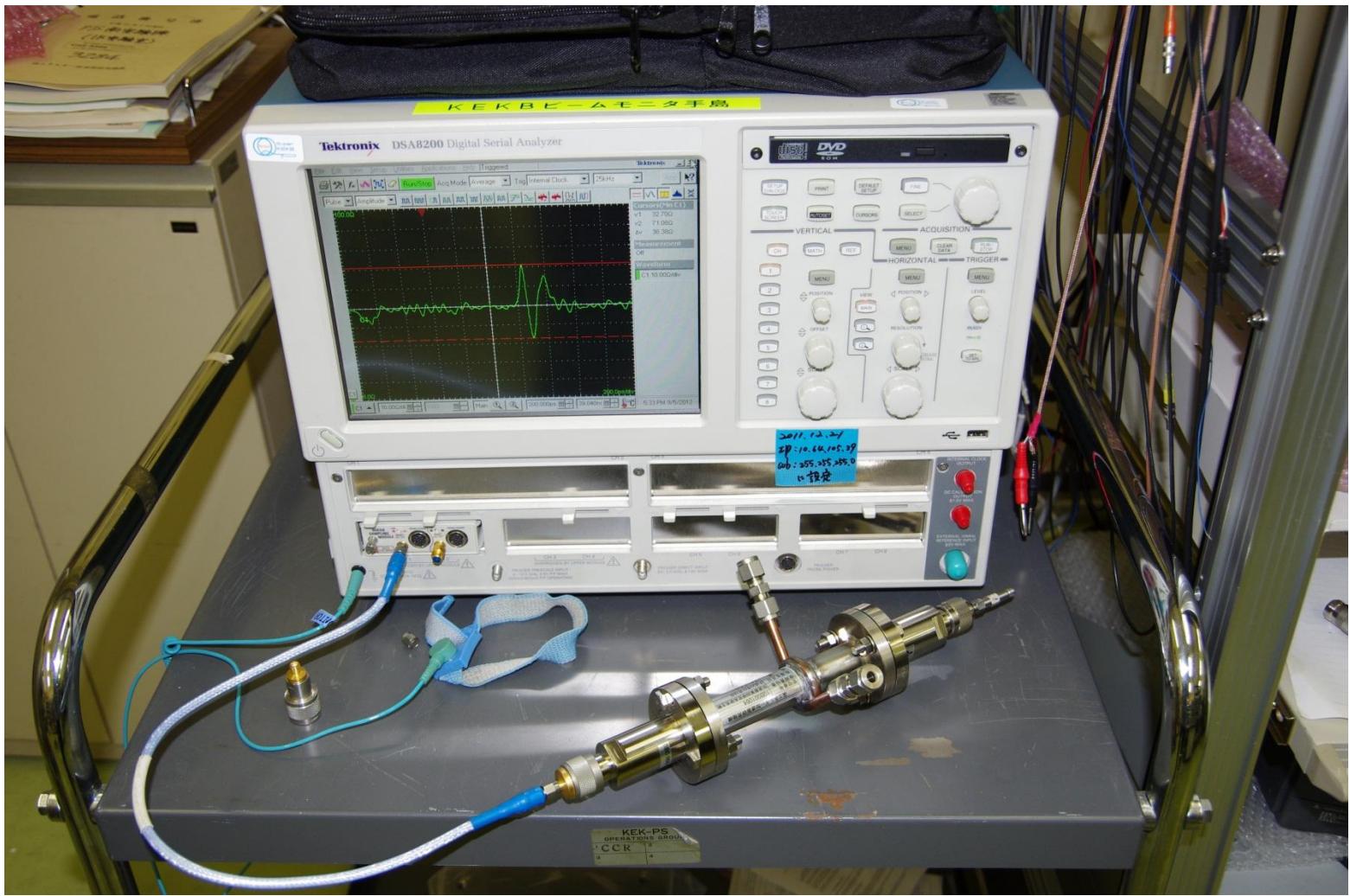
33Ω

# Matching of 50-ohm line

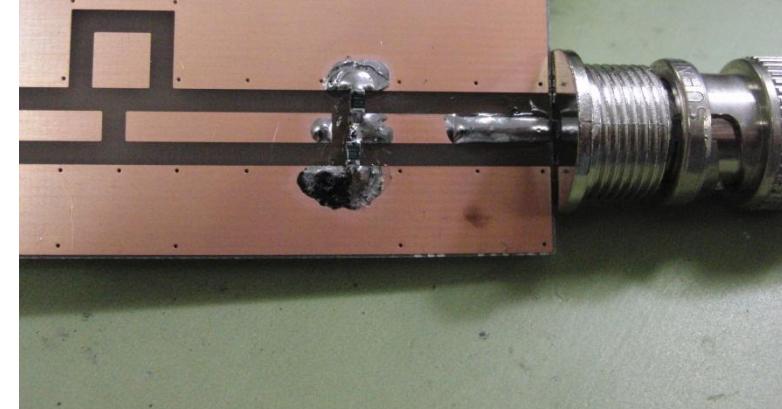
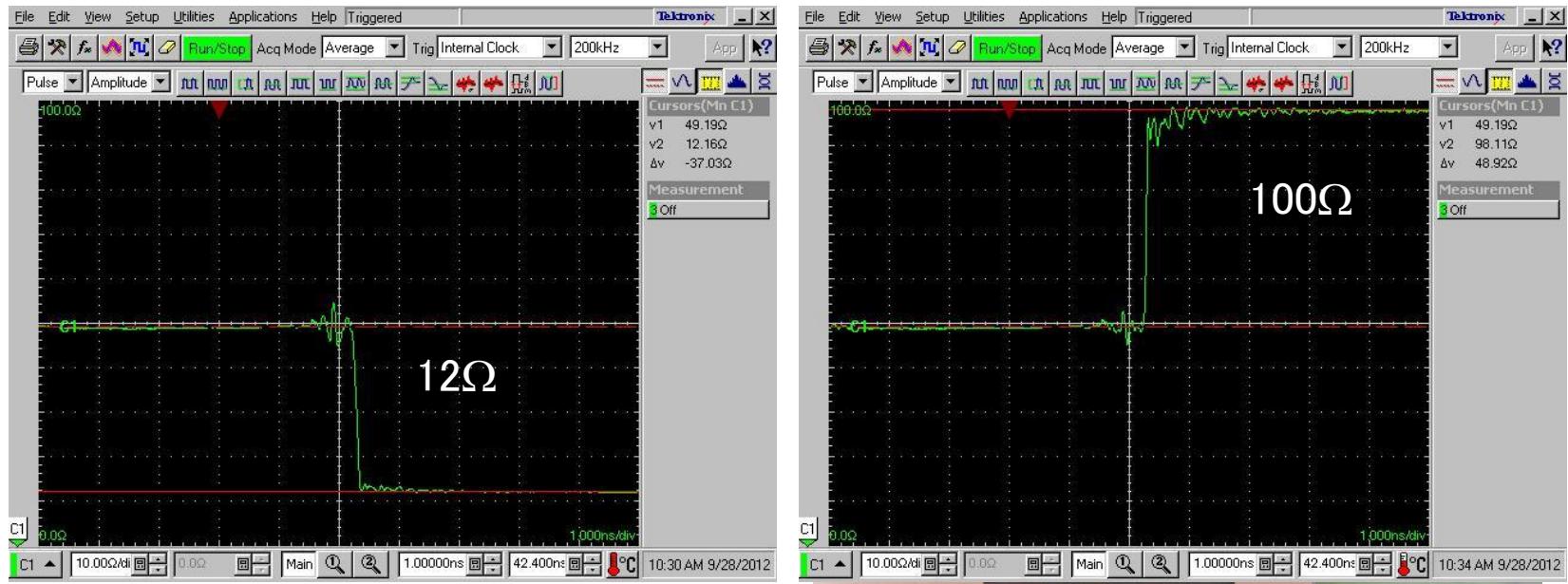
Reflection (V)



# Real TDR



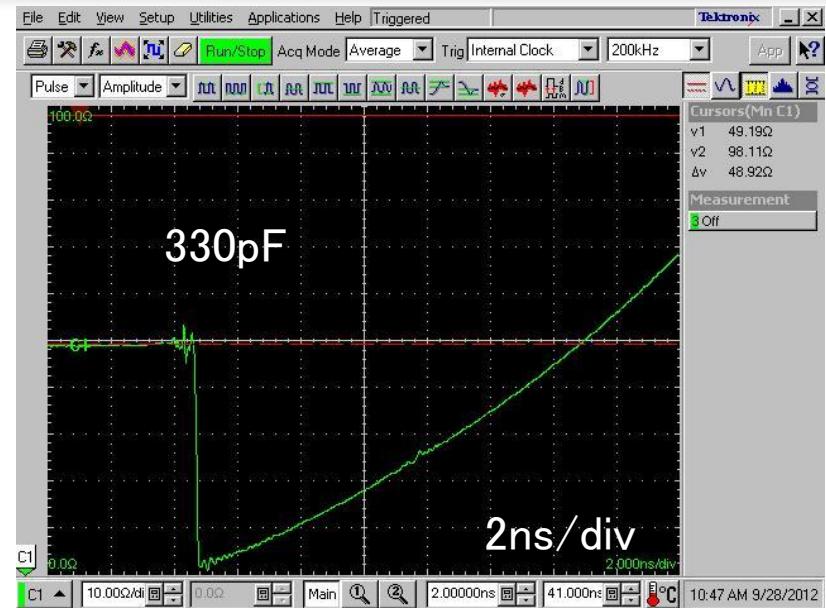
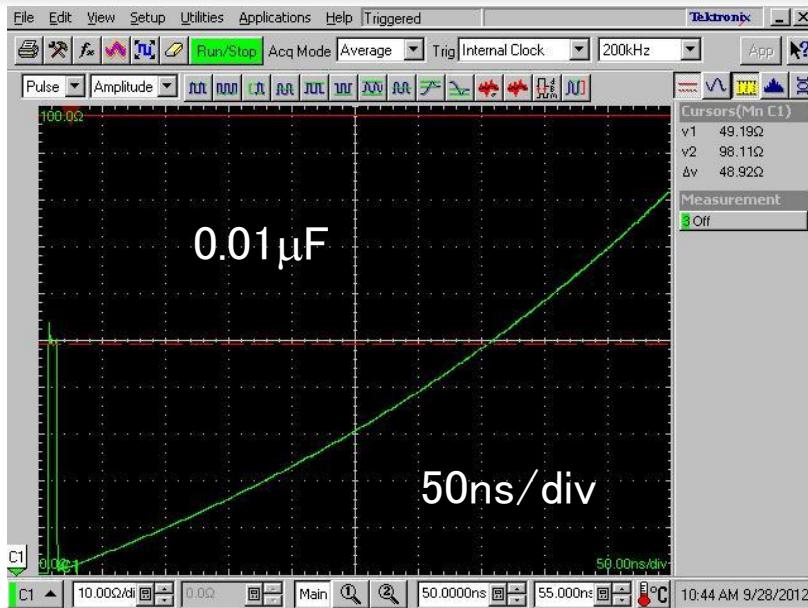
# TDR (chip register)



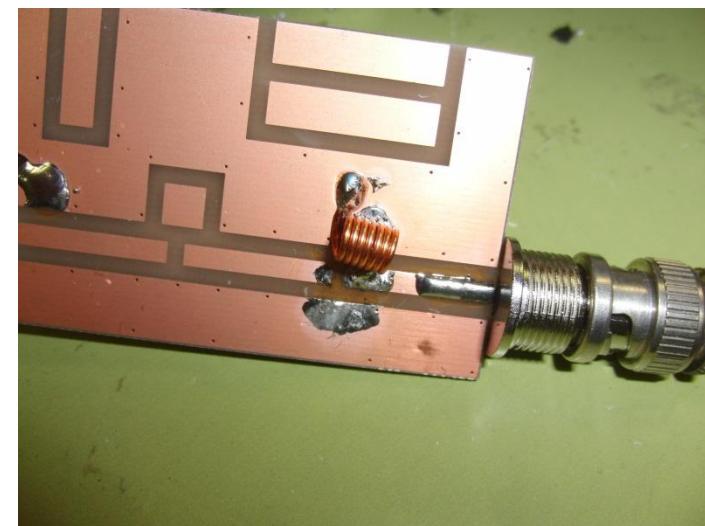
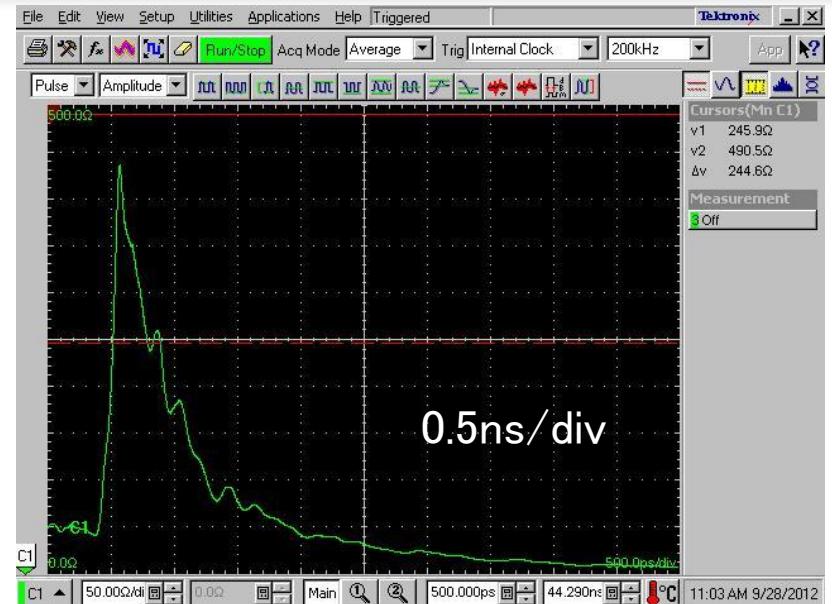
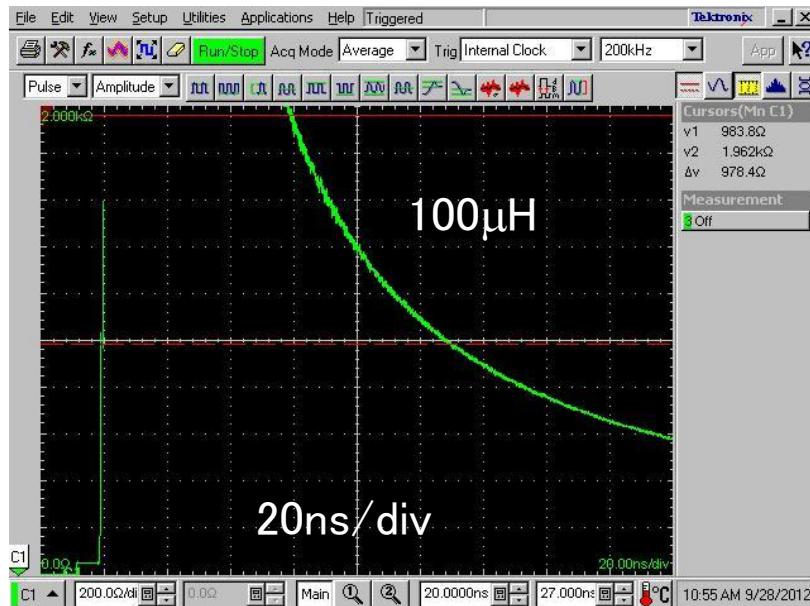
## TDR



# Capacitor



# Inductance



# Tips

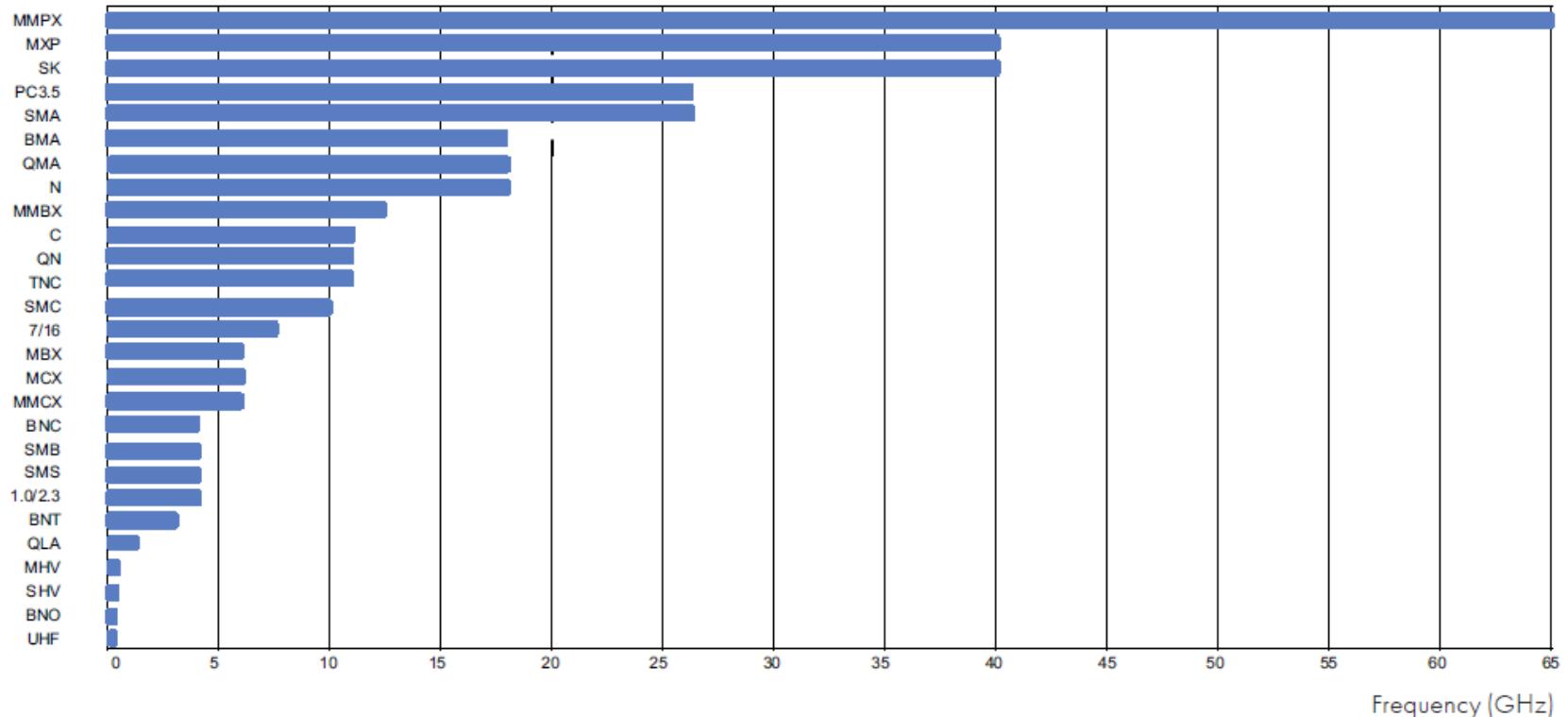
- Good matched two-ports (such as J-J structure ) is needed to measure the S-parameters of the feedthrough structure.
  - For the TDR measurements, open structure (button or rod on vacuum side) might usually be OK



# Ideal feedthrough?

- **Mechanical**
  - Completely satisfy required mechanical toughness
  - Satisfy vacuum requirements
- **Frequency domain**
  - $S_{21}$  (and  $S_{12}$ )  $\sim 1$  (0dB)
  - $S_{11}$  (and  $S_{22}$ )  $\sim -\infty$  dB
- **Time domain**
  - Impedance=50 Ohm, completely matched structure

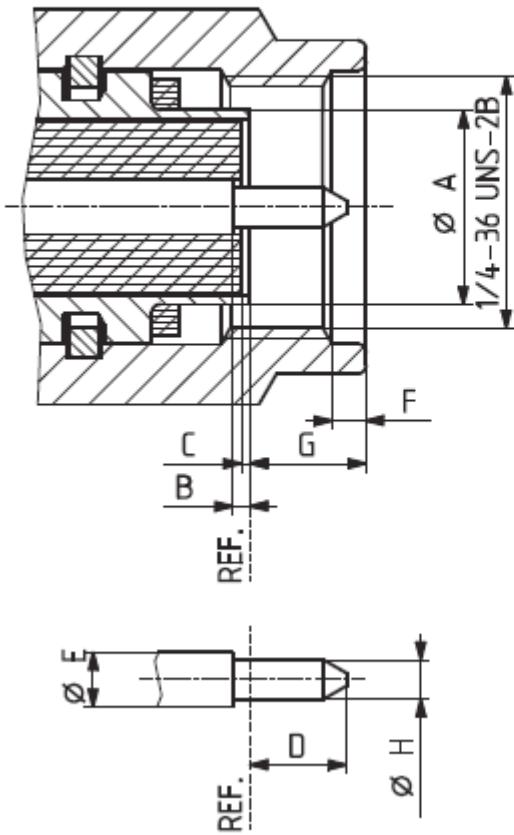
# RF connector



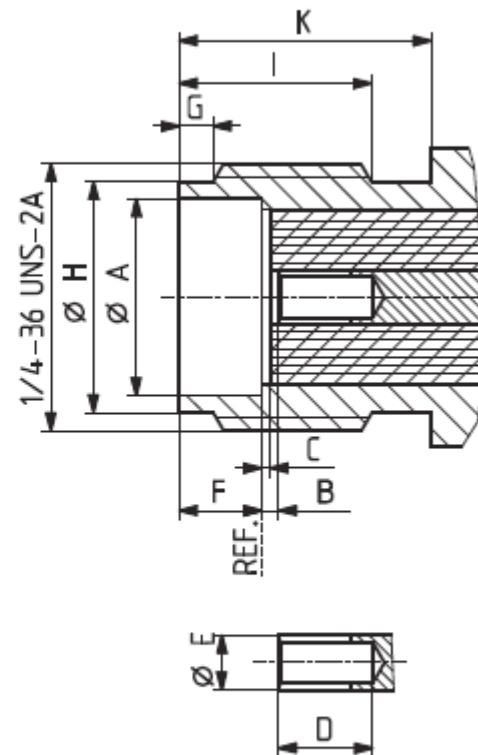
Frequency range of connector series.

# Plug(male) and Jack(female)

Plug (male)

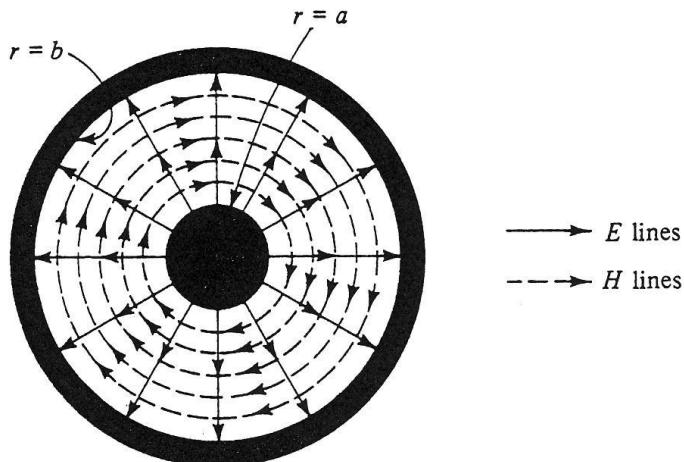


Jack (female)



# Size of the feedthrough

- Which connectors do you want to use?
  - N?, SMA?, BNC?, 7/16-DIN?, 7/8-EIA?



$$Z_0 = 60 \sqrt{\frac{\mu_R}{\epsilon_R}} \ln \frac{b}{a}$$

Attenuation Length (dB/m)

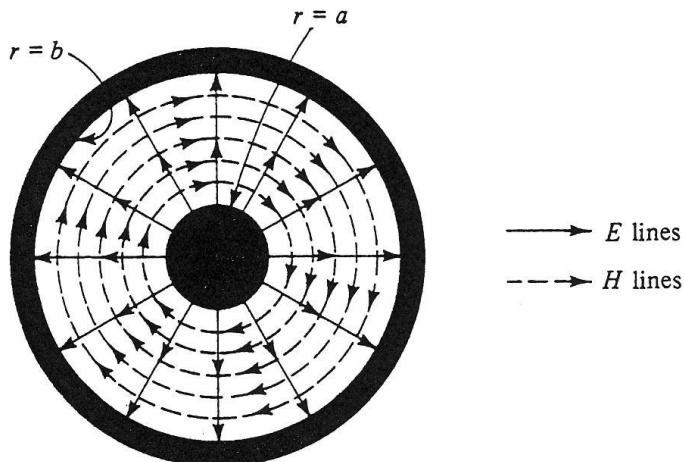
$$\alpha \approx \alpha_c + \alpha_d$$

$$\alpha_c = 13.6 \frac{\delta_s \sqrt{\epsilon_R} \{1 + (b/a)\}}{\lambda_0 b \ln(b/a)},$$

$$\alpha_d = 27.3 \frac{\sqrt{\epsilon_R}}{\lambda_0} \tan \delta$$

# Size of the feedthrough

- Which connectors do you want to use?
  - N?, SMA?, BNC?, 7/16-DIN?, 7/8-EIA?



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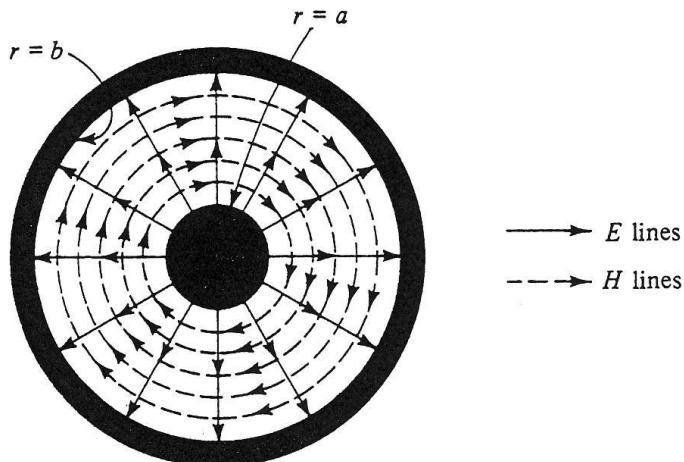
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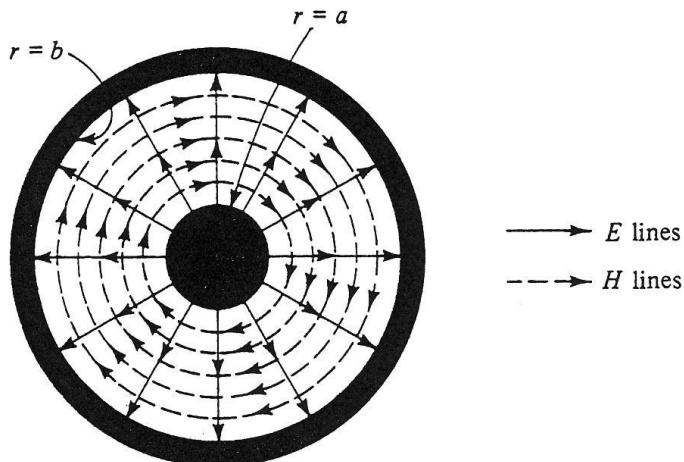
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# Size of the feedthrough

- Which connectors do you want to use?
  - N?, SMA?, BNC?, 7/16-DIN?, 7/8-EIA?



$$Z_0 = 60 \sqrt{\frac{\mu_R}{\epsilon_R}} \ln \frac{b}{a}$$

Attenuation Length (dB/m)

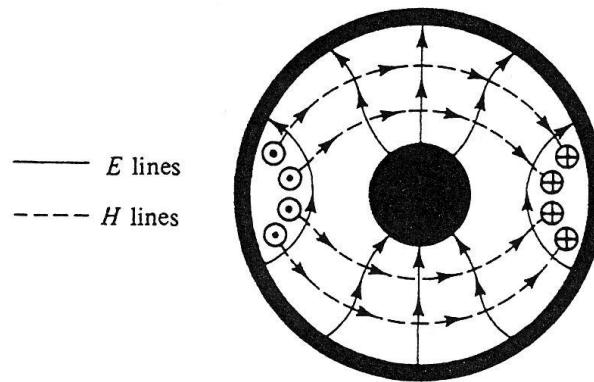
$$\alpha \approx \alpha_c + \alpha_d$$

$$\alpha_c = 13.6 \frac{\delta_s \sqrt{\epsilon_R} \left\{ 1 + (b/a) \right\}}{\lambda_0 b \ln(b/a)},$$

$$\alpha_d = 27.3 \frac{\sqrt{\epsilon_R}}{\lambda_0} \tan \delta$$

# Higher-mode propagation in coaxial lines.

## ■ TE<sub>11</sub> mode



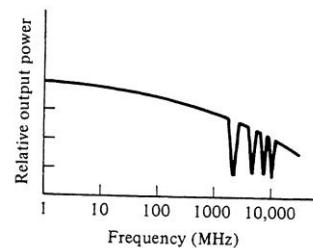
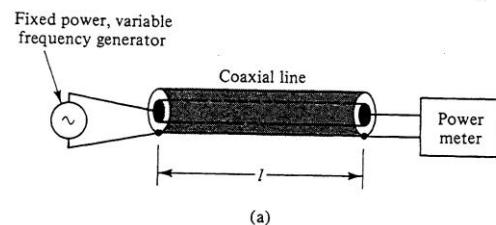
$$\lambda_c \approx \pi(a + b)$$

$a \equiv$  inner conductor radius  
 $b \equiv$  inner radius of the outer conductor

$$\lambda_c \approx \pi(a + b)$$

$$f_c \approx \frac{c}{\pi(a + b)\sqrt{\epsilon_R}}$$

Figure 5–3 The electromagnetic field pattern for the  $TE_{11}$  mode in a coaxial line.



- Type-N: 18GHz
- SMA: 34GHz
- RG-223: ~30GHz
- EIA 7/8": ~6.8GHz

# Connectors in the air

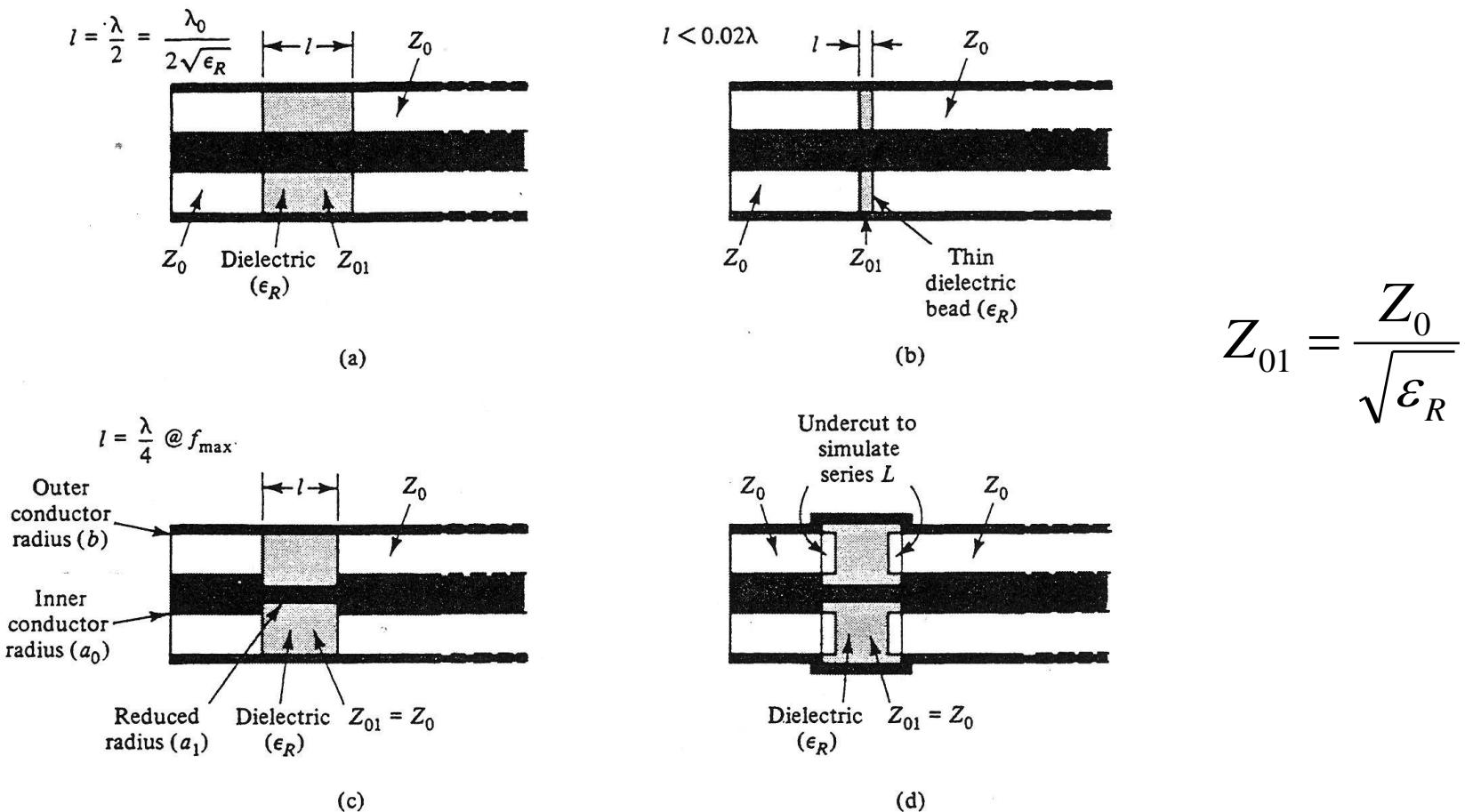


Figure 6-8 Four types of dielectric bead supports for coaxial lines.

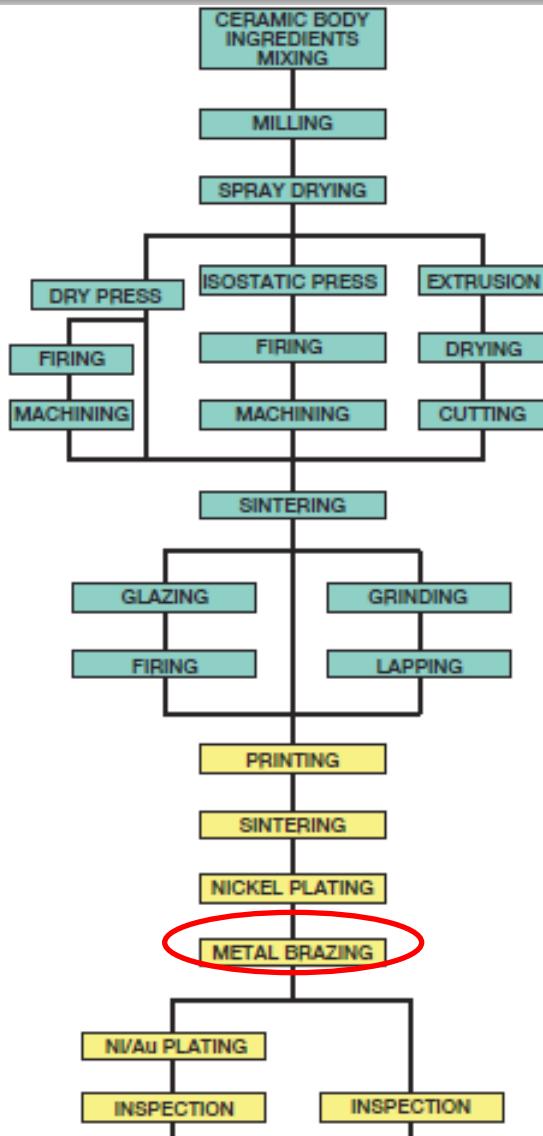
# Feedthrough for vacuum

- **Vacuum seal (and support for inner conductor)**
  - Alumina ceramic ( $\epsilon_R \sim 10$ ),  $Si_3N_4 (\sim 8)$
  - Glass ( $\epsilon_R \sim 4-5$ )
    - Much larger than the support used in the air ( $\sim 2$ )
- **Need to stand mechanical stress coming from**
  - Vacuum pressure
  - (huge) pressure when connecting the cable (or attenuator) to the feedthrough
  - Thermal stress
    - Baking
    - Heating by the beam power
- **Need to keep good RF contact under severe condition**
  - Heat cycle
  - Radiation, active gas

# Outer/inner conductor

- **Kovar (Nickel–cobalt ferrous alloy)**
  - Trademark of Carpenter Technology Corporation
  - Designed to be compatible with the thermal expansion characteristics of borosilicate glass
  - Ferromagnetics: Not so good for use near strong magnet.
  - Thermal conductivity : Low
- **Ti**
  - Non-magnetic material
- **Cu, Al, Stainless steel..**
  - Also be usable. Consult your ceramic company.

# Feedthrough process



## Brazing temperature

- ~800 degC
- After brazing process, metal parts will be well annealed – original characteristics might be changed.
  - No spring action at all!

## Braze to vacuum chamber OK?

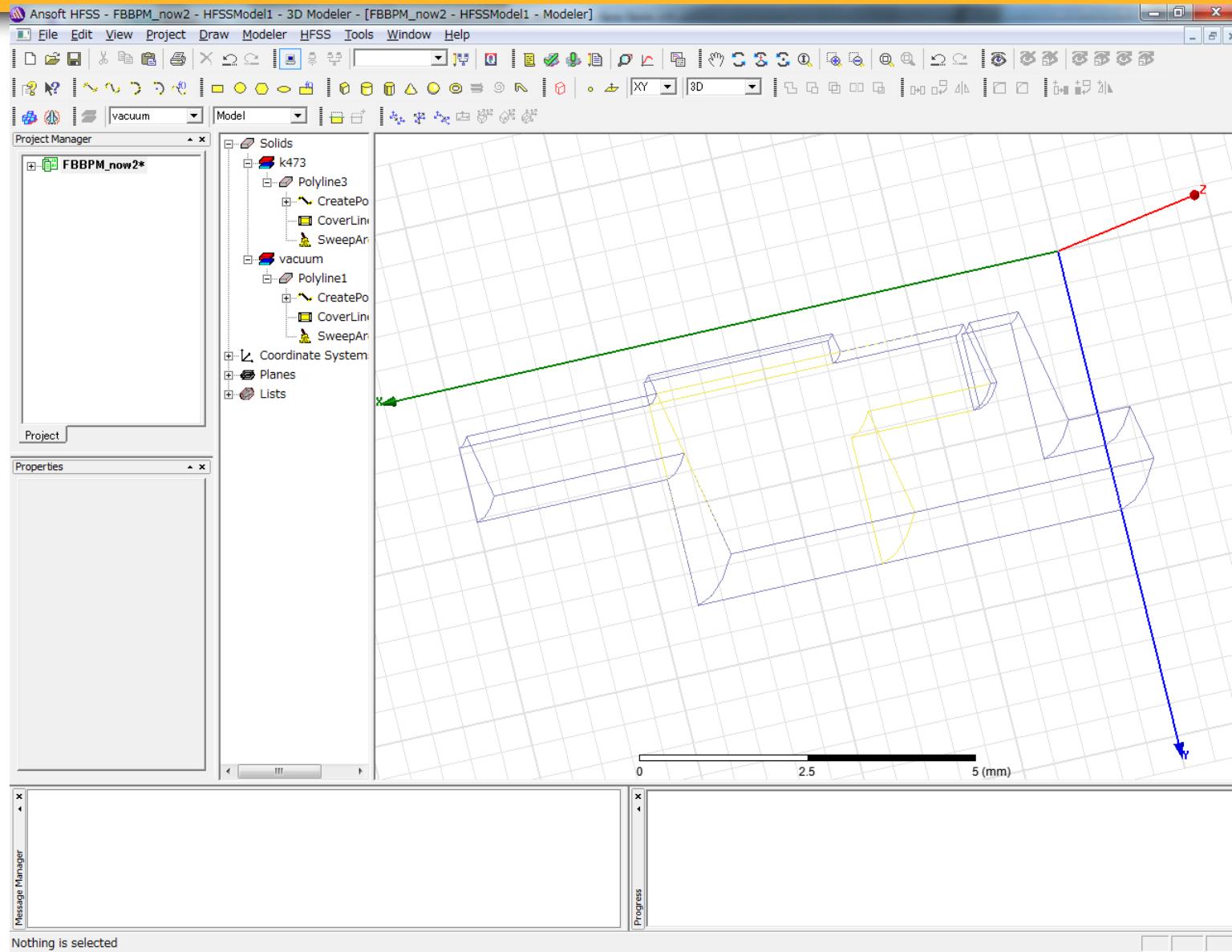
- Second (third?) brazing process
- There exist brazing materials with lower temperature
  - Not so easy to control the temperature

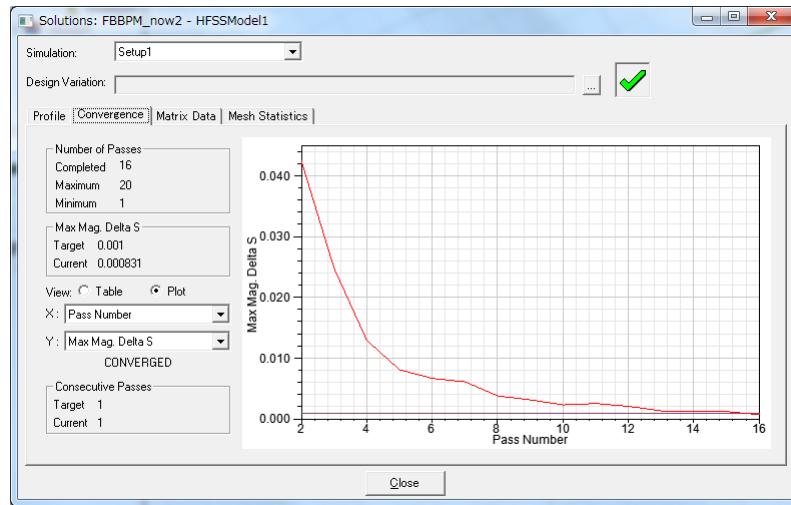
# Frequency domain design

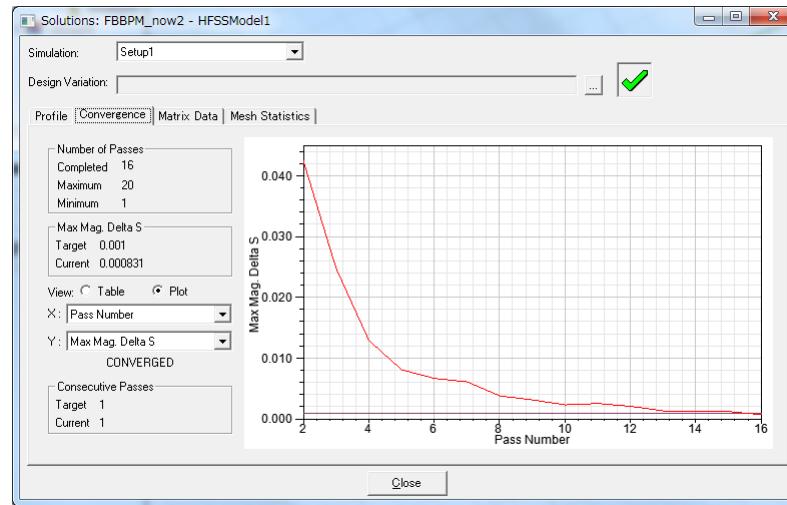
## ■ ANSYS (Ansoft) HFSS

- 3D full-wave electromagnetic field simulator
- Multiple state-of the-art solver (finite element method or integral equation method)
- calculate S-parameters, field patterns, eigen values, impedances..
- Good user interface, Good interface to other tools such as ANSYS, etc.
- Automatic mesh handing
- (Very) Fast simulation speed

# Example







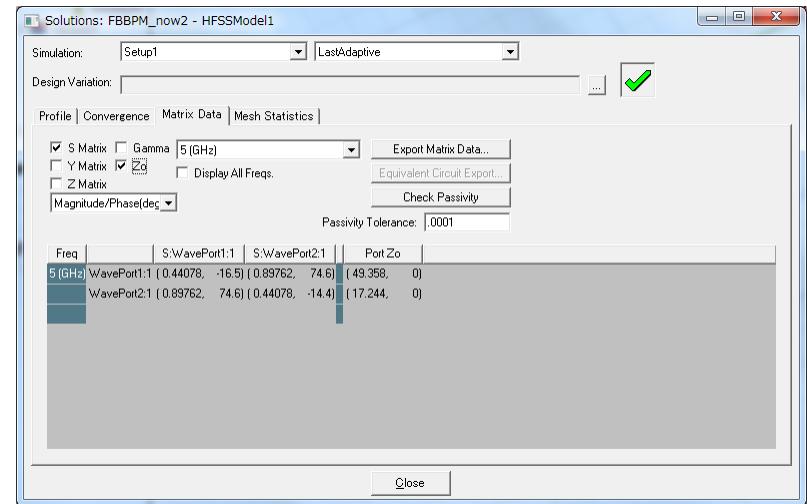
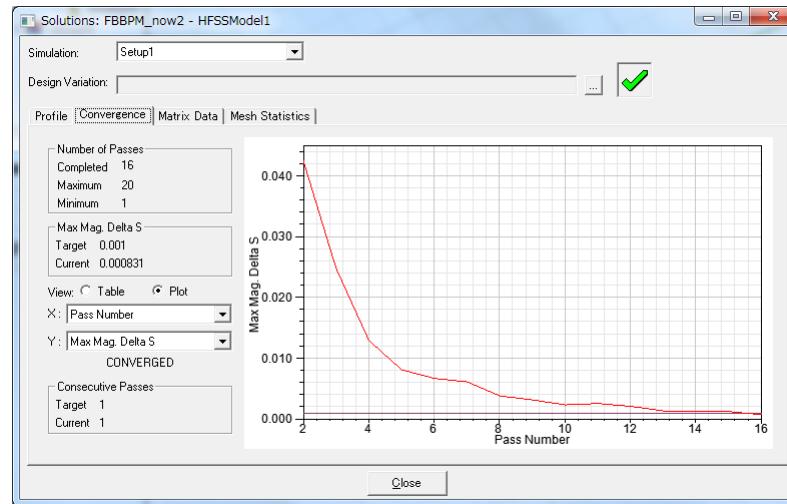
Solutions: FBBPM\_now2 - HFSSModel1

Simulation: Setup1 Design Variation:

Profile Convergence Matrix Data Mesh Statistics

Task	Real Time	CPU Time	Memory	Information
Frequency: 21.25 GHz				Full Solution # 19
Simulation Setup	00:00:00	00:00:00	35.4 M	Disk = 0 KBytes
Matrix Assembly	00:00:00	00:00:00	61 M	Disk = 0 KBytes, 9601 tetrahedra, WavePort1: 84 triangles, WavePort2: 90 triangles
Solver MRS4	00:00:01	00:00:02	130 M	Disk = 0 KBytes, matrix size 61092, matrix bandwidth 20.7
Field Recovery	00:00:00	00:00:00	130 M	Disk = 0 KBytes, 2 excitations
				Interpolation Error: S Matrix error 0 %
				Interpolating sweep converged
Solution Process				Elapsed time : 00:00:39, Hiss ComEngine Memory : 46.7 M
Total	00:00:21	00:00:40		Time: 09/10/2012 11:16:56, Status: Normal Completion

Export... Close



Solutions: FBBPM\_now2 - HFSSModel1

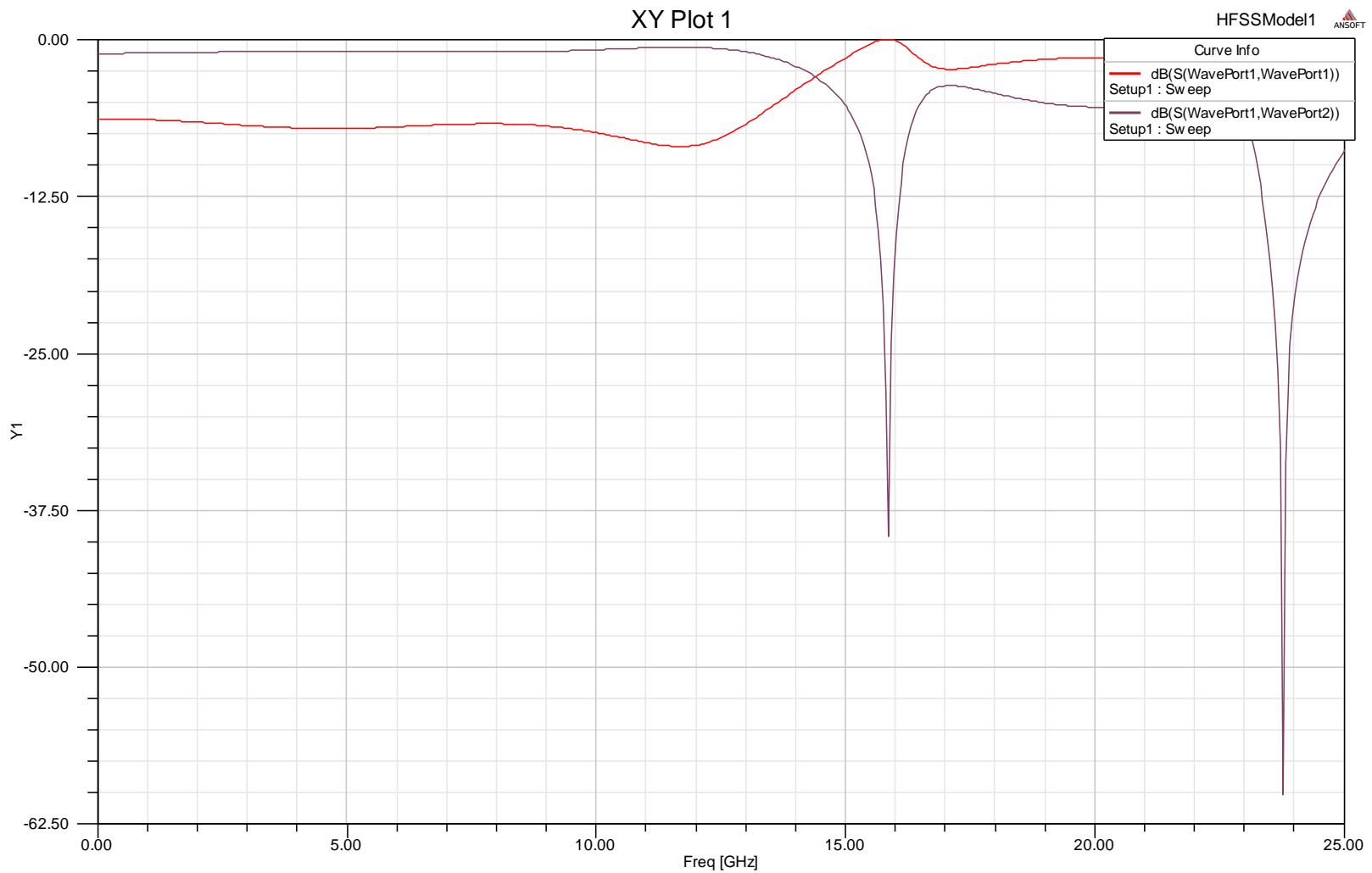
Simulation: Setup1 Design Variation:

Profile | Convergence | Matrix Data | Mesh Statistics |

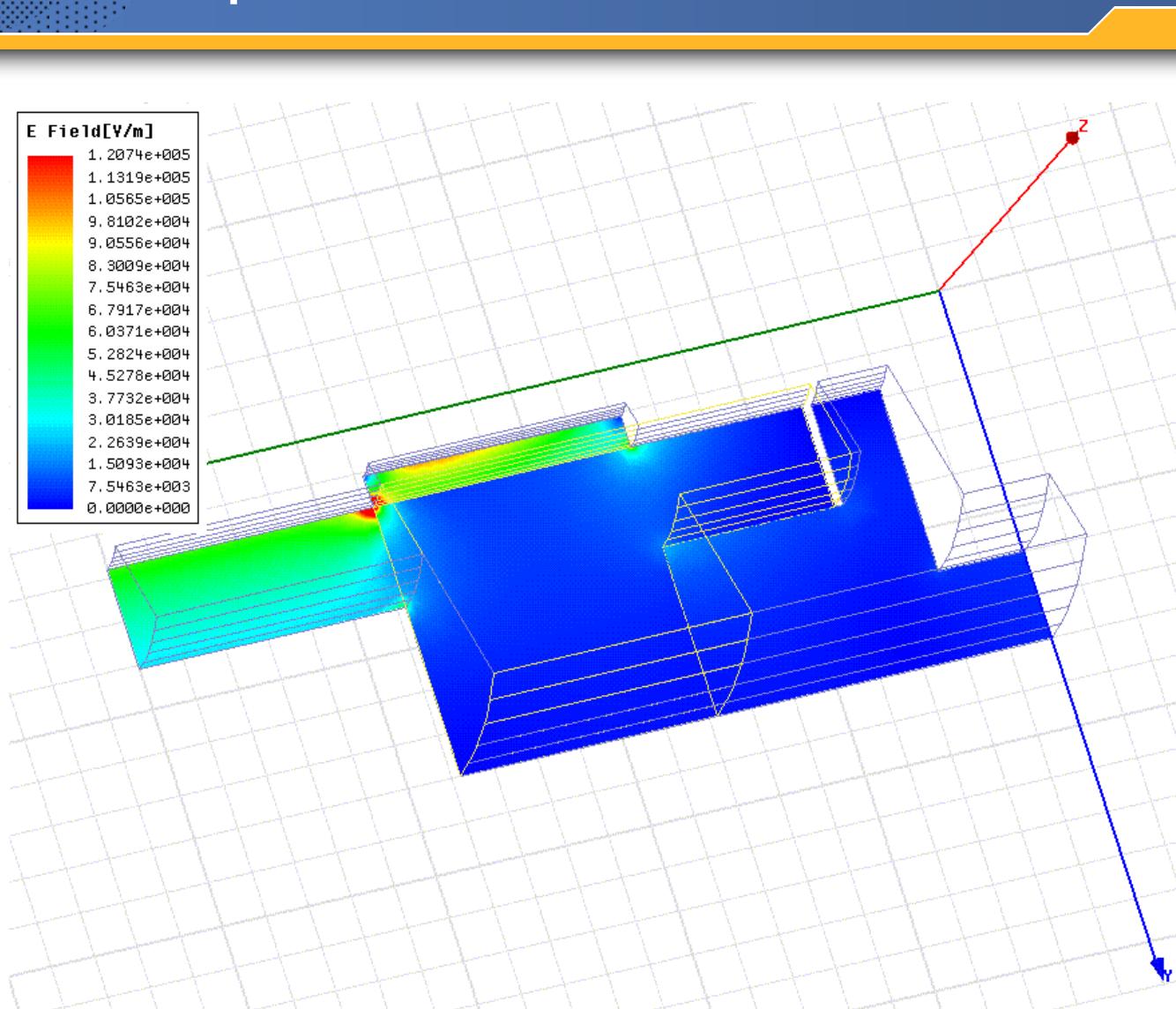
Task	Real Time	CPU Time	Memory	Information
Frequency: 21.25 GHz				Full Solution # 19
Simulation Setup	00:00:00	00:00:00	35.4 M	Disk = 0 KBytes
Matrix Assembly	00:00:00	00:00:00	61 M	Disk = 0 KBytes, 9601 tetrahedra, WavePort1: 84 triangles, WavePort2: 90 triangles
Solver MRS4	00:00:01	00:00:02	130 M	Disk = 0 KBytes, matrix size 61092, matrix bandwidth 20.7
Field Recovery	00:00:00	00:00:00	130 M	Disk = 0 KBytes, 2 excitations
				Interpolation Error: S Matrix error 0 %
				Interpolating sweep converged
Solution Process				Elapsed time : 00:00:39, Hiss ComEngine Memory : 46.7 M
Total	00:00:21	00:00:40		Time: 09/10/2012 11:16:56, Status: Normal Completion

Export... Close

# (Fast) Frequency sweep



# E-Field pattern



# Time domain simulation

- **GdfidL Electromagnetic Field simulator**
  - Time dependent Fields in loss-free or lossy structure
  - Fields may be excited by port mode or relativistic line charge
  - Resonant fields in loss-free or lossy structure
- **Calculated result**
  - S-parameters, including power (voltage) value
  - Wake potentials including loss factor, impedance etc.
  - Q values and Shunt impedance
- **Need (some)large-scale parallel computing resources**
  - KEKB: 256 cores, 512GB Linux cluster
  - Text-base user interface, longer simulation time

# Example of GdfidL input(1)

```
define(EL,1) define(MAG,2)  
define(INF, 1000)
```

```
define(PipeRadius, 25.40e-3/2)  
define(PipeStart, 100.0e-3)  
define(PipeEnd,100.0e-3)  
define(BpmAdd,13.5e-3)  
define(STPSZE,0.1e-3)
```

```
-general  
    outfile= /users/tobiyama/GD/bpm-CL22/temp/  
    scratchbase =/users/tobiyama/GD/scratch/  
-material  
    material=10  
    type=electric  
    material=15  
    type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0
```

```
-mesh  
    spacing=0.1e-3  
    pxlow= 0  
    pxhigh=(PipeRadius+BpmAdd-5*STPSZE)  
    pylow= 0  
    pyhigh=(PipeRadius+BpmAdd-5*STPSZE)  
    pzlow=-PipeStart+5*STPSZE  
    pzhigh=PipeEnd-5*STPSZE
```

```
    cxlow=magnetic, cxhigh=electric  
    cylow=magnetic, cyhigh=electric  
    czlow=electric, czhigh=electric
```

```
-brick  
    material=EL  
    xlow= -INF, xhigh=INF  
    ylow= -INF, yhigh=INF  
    zlow= -INF, zhigh=INF  
    doit
```

# Example of GdfidL input(1)

```
define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0
```

```
-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pzlow=-PipeStart+5*STPSZE
  pzhigh=PipeEnd-5*STPSZE
```

```
cxlow=magnetic, cxhigh=electric
cylow=magnetic, cyhigh=electric
czlow=electric, czhigh=electric
```

```
-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhight=INF
  doit
```

# Example of GdfidL input(1)

```
define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0
```

```
-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pzlow=-PipeStart+5*STPSZE
  pzhigh=PipeEnd-5*STPSZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhight=INF
  doit
```

# Example of GdfidL input(1)

```
define(EL,1) define(MAG,2)
define(INF, 1000)

define(PipeRadius, 25.40e-3/2)
define(PipeStart, 100.0e-3)
define(PipeEnd,100.0e-3)
define(BpmAdd,13.5e-3)
define(STPSZE,0.1e-3)

-general
  outfile= /users/tobiyama/GD/bpm-CL22/temp/
  scratchbase =/users/tobiyama/GD/scratch/
-material
  material=10
  type=electric
  material=15
  type=normal, epsr=9.7, muer=1, kappa=0, mkappa=0
```

```
-mesh
  spacing=0.1e-3
  pxlow= 0
  pxhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pylow= 0
  pyhigh=(PipeRadius+BpmAdd-5*STPSZE)
  pzlow=-PipeStart+5*STPSZE
  pzhigh=PipeEnd-5*STPSZE

  cxlow=magnetic, cxhigh=electric
  cylow=magnetic, cyhigh=electric
  czlow=electric, czhigh=electric

-brick
  material=EL
  xlow= -INF, xhigh=INF
  ylow= -INF, yhigh=INF
  zlow= -INF, zhight=INF
  doit
```

# Example of GdfidL input(2)

```
#  
# cave beam pipe  
#  
-gbor  
material=0  
origin=(0,0,0)  
zprimedirection=(0,0,1)  
rprimedirection=(1,0,0)  
range= (0,360)  
  
clear  
  
point(-PipeStart,0)  
point(-PipeStart,PipeRadius)  
point(PipeEnd,PipeRadius)  
point(PipeEnd,0)  
point(-PipeStart,0)  
show=all  
doit
```

```
#  
# bpm block  
#  
macro BpmBlock  
define(Dist, @arg1)  
define(Angle, (@arg2)*@pi/180)  
  
define(BpmDiskRadius, 6e-3/2)  
define(BpmRodRadius,1.8e-3/2)  
define(BpmairRadius,1.1e-3/2)  
define(BpmOutRadius1, 8e-3/2)  
define(BpmOutRadius2, 4.1e-3/2)  
define(BpmCera1Radius,4.1e-3/2)  
define(BpmCera2Radius,8e-3/2)  
define(BpmDiskLength,1e-3)  
define(BpmRod1Length,2.9e-3)  
define(BpmRod2Length,3e-3)  
define(BpmRod3Length,6.1e-3)  
  
define(Cerastart,1.9e-3)  
define(Cera1Length,2e-3)  
define(Cera2Length,3e-3)  
  
define(DP1, (5e-  
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))  
define(DP2, (DP1+Dist - 5e-3))
```

# Example of GdfidL input(2)

```
#  
# cave beam pipe  
#  
-gbor  
material=0  
origin=(0,0,0)  
zprimedirection=(0,0,1)  
rprimedirection=(1,0,0)  
range= (0,360)  
  
clear  
  
point(-PipeStart,0)  
point(-PipeStart,PipeRadius)  
point(PipeEnd,PipeRadius)  
point(PipeEnd,0)  
point(-PipeStart,0)  
show=all  
doit
```

```
#  
# bpm block  
#  
macro BpmBlock  
define(Dist, @arg1)  
define(Angle, (@arg2)*@pi/180)  
  
define(BpmDiskRadius, 6e-3/2)  
define(BpmRodRadius,1.8e-3/2)  
define(BpmairRadius,1.1e-3/2)  
define(BpmOutRadius1, 8e-3/2)  
define(BpmOutRadius2, 4.1e-3/2)  
define(BpmCera1Radius,4.1e-3/2)  
define(BpmCera2Radius,8e-3/2)  
define(BpmDiskLength,1e-3)  
define(BpmRod1Length,2.9e-3)  
define(BpmRod2Length,3e-3)  
define(BpmRod3Length,6.1e-3)  
  
define(Cerastart,1.9e-3)  
define(Cera1Length,2e-3)  
define(Cera2Length,3e-3)  
  
define(DP1, (5e-  
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))  
define(DP2, (DP1+Dist - 5e-3))
```

# Example of GdfidL input(2)

```
#  
# cave beam pipe  
#  
-gbor  
material=0  
origin=(0,0,0)  
zprimedirection=(0,0,1)  
rprimedirection=(1,0,0)  
range=(0,360)  
  
clear  
  
point(-PipeStart,0)  
point(-PipeStart,PipeRadius)  
point(PipeEnd,PipeRadius)  
point(PipeEnd,0)  
point(-PipeStart,0)  
show=all  
doit
```

```
#  
# bpm block  
#  
macro BpmBlock  
define(Dist, @arg1)  
define(Angle, (@arg2)*@pi/180)  
  
define(BpmDiskRadius, 6e-3/2)  
define(BpmRodRadius,1.8e-3/2)  
define(BpmairRadius,1.1e-3/2)  
define(BpmOutRadius1, 8e-3/2)  
define(BpmOutRadius2, 4.1e-3/2)  
define(BpmCera1Radius,4.1e-3/2)  
define(BpmCera2Radius,8e-3/2)  
define(BpmDiskLength,1e-3)  
define(BpmRod1Length,2.9e-3)  
define(BpmRod2Length,3e-3)  
define(BpmRod3Length,6.1e-3)  
  
define(Cerastart,1.9e-3)  
define(Cera1Length,2e-3)  
define(Cera2Length,3e-3)  
  
define(DP1, (5e-  
3+BpmDiskLength+BpmRod1Length+BpmRod2Length))  
define(DP2, (DP1+Dist - 5e-3))
```

# Example of GdfidL input(3)

```

#
-gccylinder
material=0, radius=BpmOutRadius1,length = DP1
origin=(cos(Angle)*(Dist - 5e-3), sin(Angle)*(Dist - 5e-3), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=0, radius=BpmOutRadius2
length= BpmRod3Length
origin=(cos(Angle)*(DP2),sin(Angle)*(DP2), 0)
direction=(cos(Angle),sin(Angle),0)
doit
#
# ceramic
#
-gccylinder
material=15, radius=BpmCera1Radius,length=Cera1Length
origin=(cos(Angle)*(Dist+Cerastart), sin(Angle)*(Dist+Cerastart), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=15, radius=BpmCera2Radius,length=Cera2Length
origin=(cos(Angle)*(Dist+Cerastart+Cera1Length),¥
         sin(Angle)*(Dist+Cerastart+Cera1Length), 0)
direction=(cos(Angle),sin(Angle),0)
doit

```

```

# air again
-gccylinder
material=0, radius=BpmRodRadius,length =
(Cera1Length+Cera2Length)
origin=(cos(Angle)*(Dist+Cerastart),sin(Angle)*(Dist-Cerastart) 0)
direction=(cos(Angle),sin(Angle),0)
doit
# bpm disk
-gccylinder
material=10, radius=BpmDiskRadius,length=BpmDiskLength
origin=(cos(Angle)*(Dist), sin(Angle)*(Dist), 0)
direction=(cos(Angle),sin(Angle),0)
doit
-gccylinder
material=10, radius=BpmRodRadius, length=BpmRod1Length
origin=(cos(Angle)*(Dist+BpmDiskLength),¥
         sin(Angle)*(Dist+BpmDiskLength), 0)
direction=(cos(Angle),sin(Angle),0)
doit

....still many definitions...

endmacro #BpmBlock

```

# Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)

-fdtd
-lcharge
  charge=1e-12
  sigma=6e-3
  xposition=0, yposition=0
  shigh=5
  showdata=no
-ports
  name=beamlow, plane=zlow, modes=0, npml=40, doit
  name=beamhigh, plane=zhigh, modes=0, npml=40, doit
  name=bpmyp, plane=yhigh, modes=1, npml=15, doit
  name=bpmxp, plane=xhigh, modes=1, npml=15, doit

-fdtd
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

# Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)

-fdtd
-lcharge
  charge=1e-12
  sigma=6e-3
  xposition=0, yposition=0
  shigh=5
  showdata=no
-ports
  name=beamlow, plane=zlow, modes=0, npml=40, doit
  name=beamhigh, plane=zhigh, modes=0, npml=40, doit
  name=bpmyp, plane=yhigh, modes=1, npml=15, doit
  name=bpmxp, plane=xhigh, modes=1, npml=15, doit

-fdtd
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

# Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)

-fdtd
-lcharge
    charge=1e-12
    sigma=6e-3
    xposition=0, yposition=0
    shigh=5
    showdata=no
-ports
    name=beamlow, plane=zlow, modes=0, npml=40, doit
    name=beamhigh, plane=zhigh, modes=0, npml=40, doit
    name=bpmyp, plane=yhigh, modes=1, npml=15, doit
    name=bpmxp, plane=xhigh, modes=1, npml=15, doit

-fdtd
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

# Example of GdfidL input(4)

```
call BpmBlock(13.2e-3,90)
call BpmBlock(13.2e-3,0)
call BpmBlock(13.2e-3,-90)
call BpmBlock(13.2e-3,180)
```

```
-fdtd
```

```
-lcharge
```

```
charge=1e-12
sigma=6e-3
xposition=0, yposition=0
shigh=5
showdata=no
```

```
-ports
```

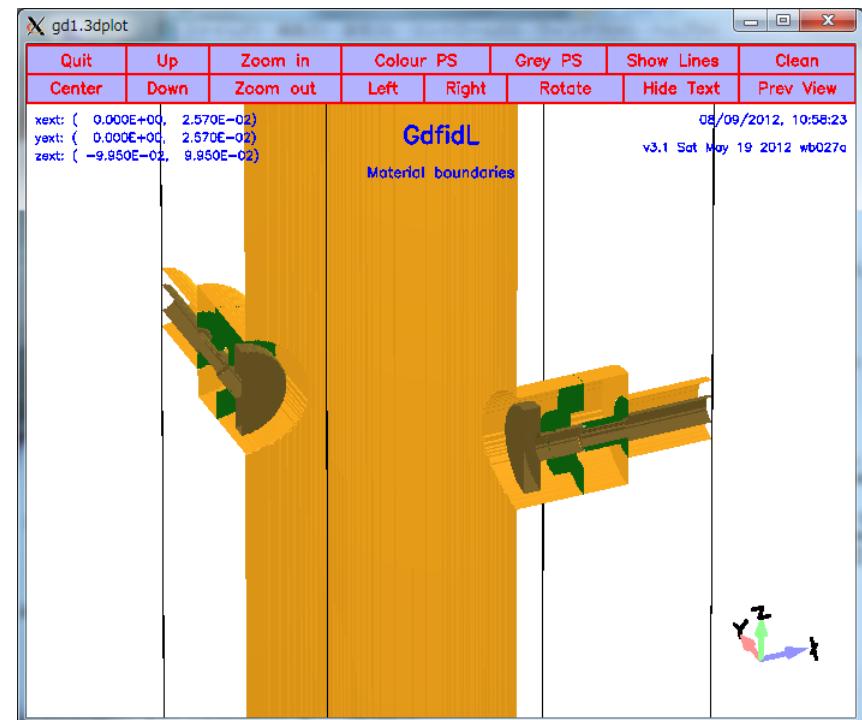
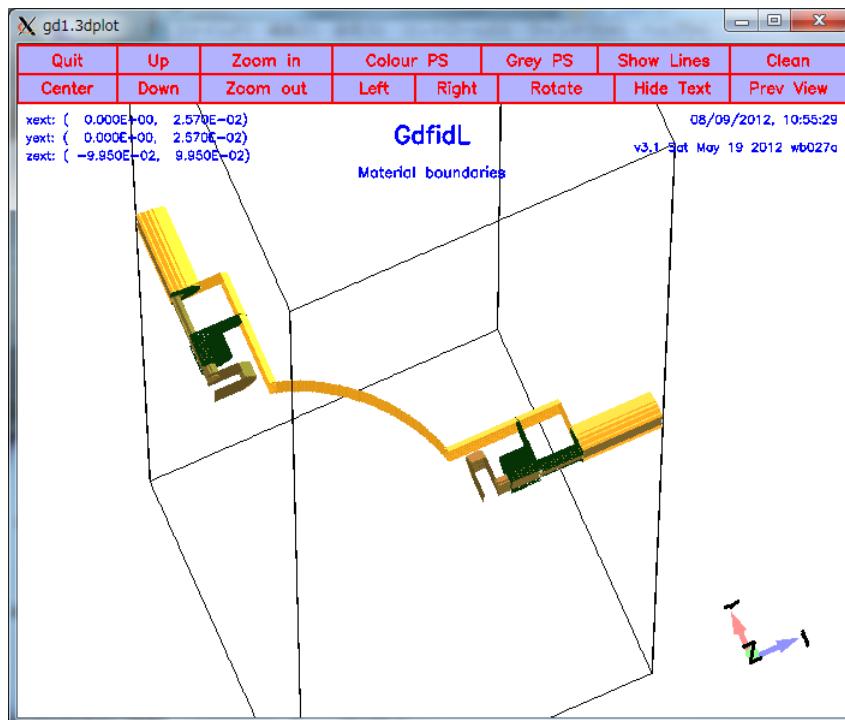
```
name=beamlow, plane=zlow, modes=0, npml=40, doit
name=beamhigh, plane=zhigh, modes=0, npml=40, doit
name=bpmyp, plane=yhigh, modes=1, npml=15, doit
name=bpmxp, plane=xhigh, modes=1, npml=15, doit
```

```
-fdtd
```

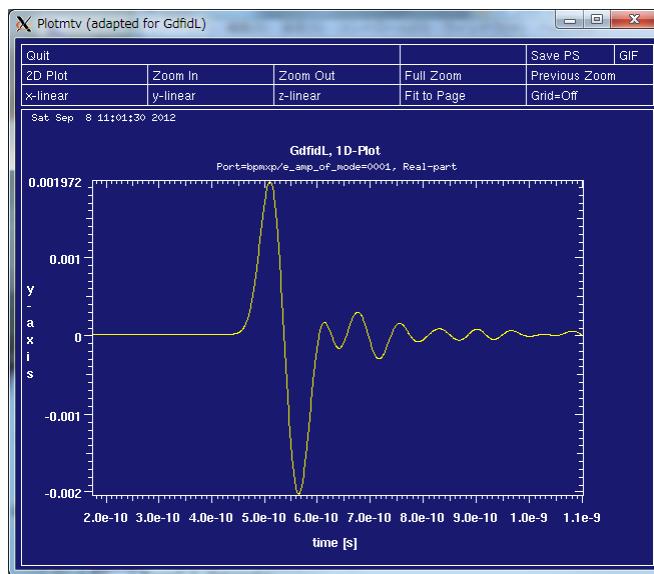
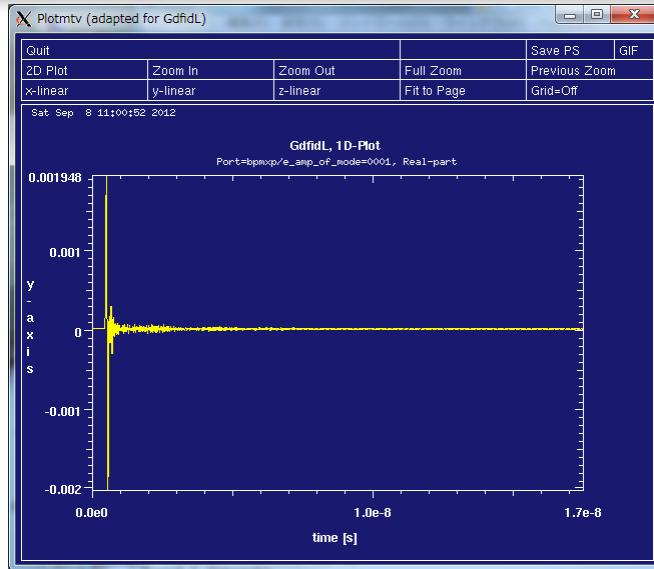
```
doit
```

- **mesh 0.1mm**
- **bunch length 6mm**
- **beam=(0,0) [center]**
- **wake up to 5m**

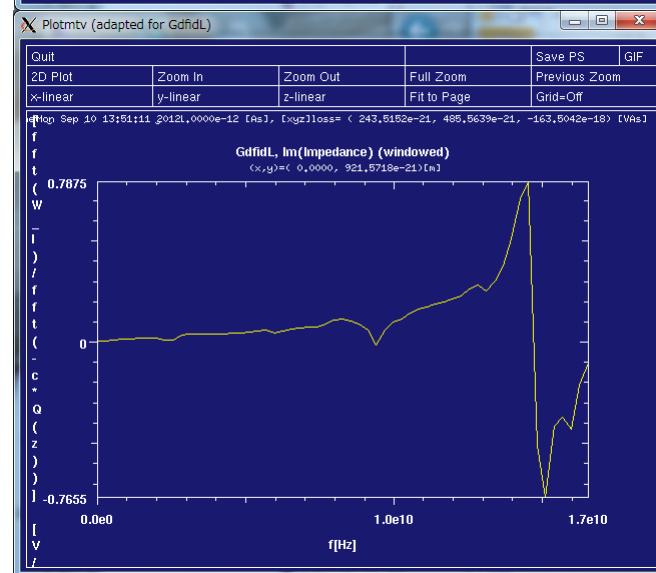
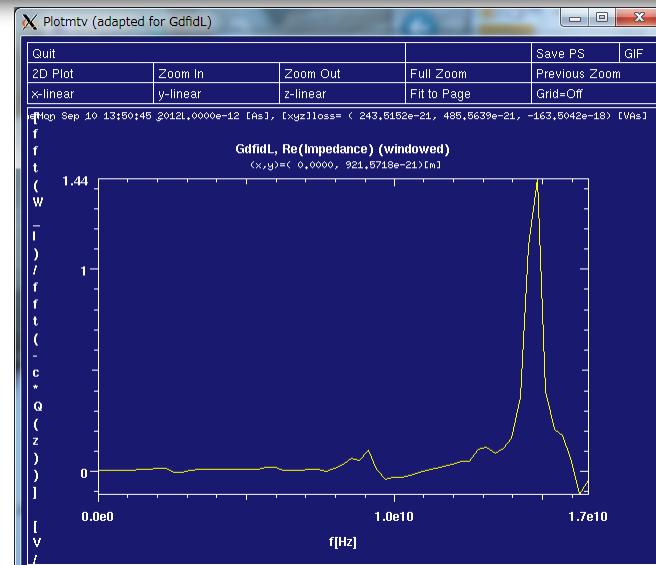
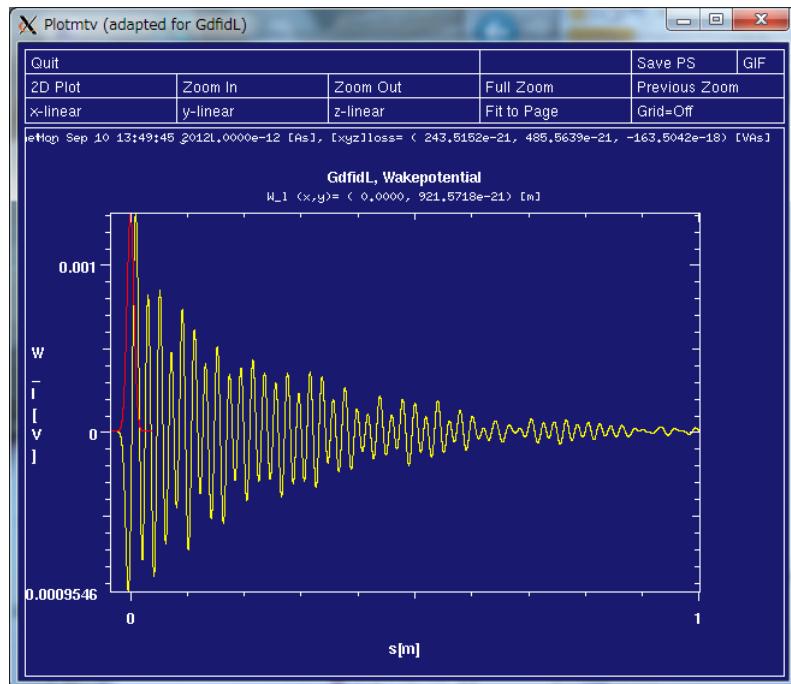
# Model view



# Post processing (using gd1.pp)

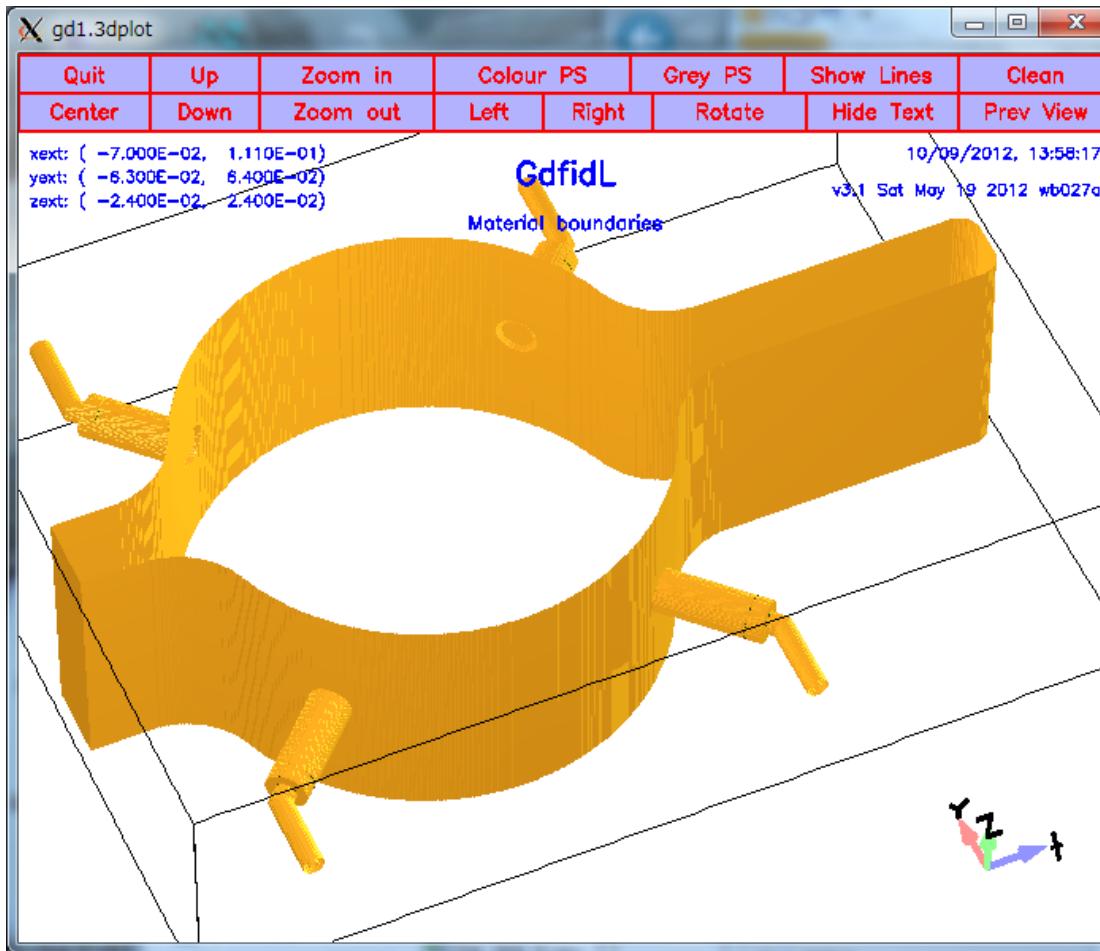


# wakes, impedances

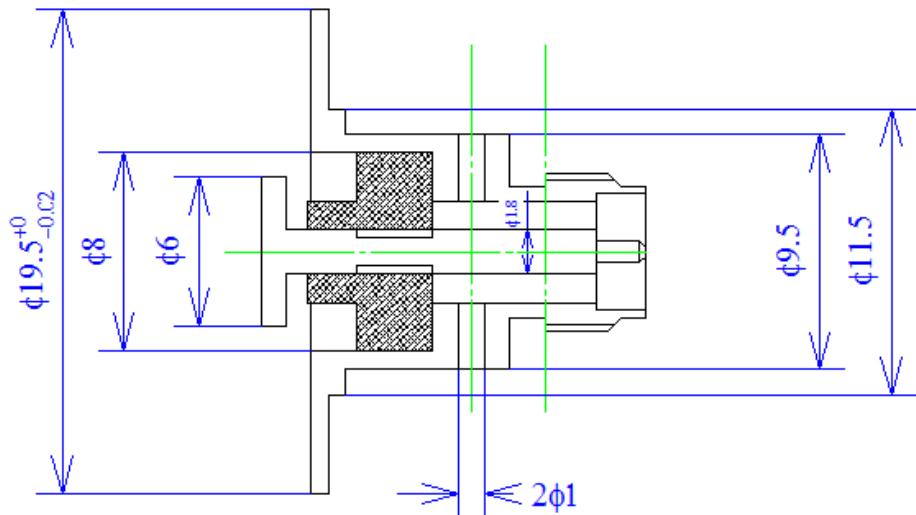


# Tips

- Ports must be placed orthogonally to the boundaries.



# Button electrode (KEKB-FB)



Feedback用SMAフィードスルーC型概略図

作図: 飛山真理 19/Mar/2004  
縮尺4:1



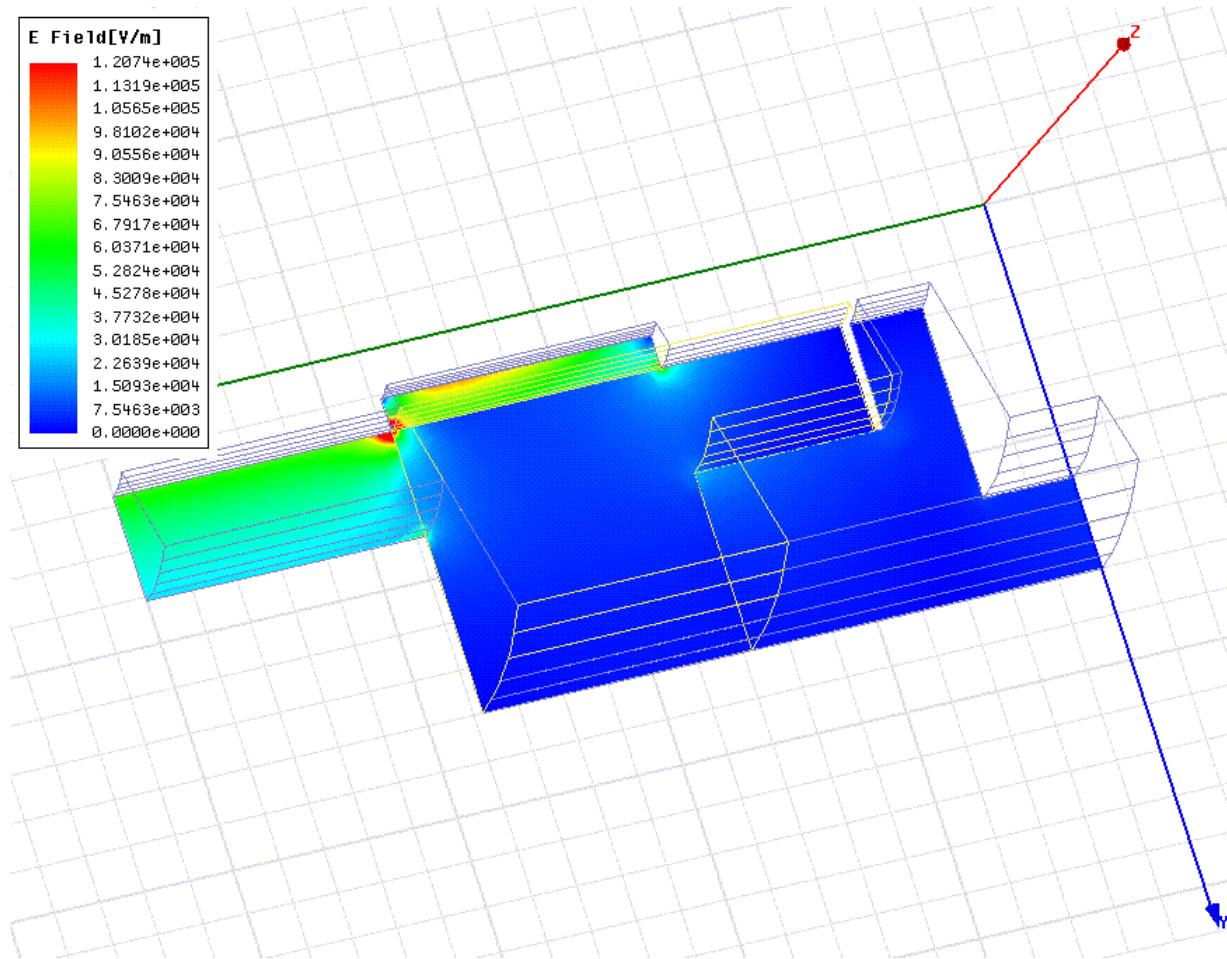
# Tips

- **Using normal SMA–J connector (or N–J, too) usually encounter difficulty in RF contact at the split–pin of the center conductor due to brazing process – No spring action might be expected after the process.**
  - Using P–structure.. not easy to handle..
  - Use pin on the J–structure (“Reverse type J”)
- **Gold–plating of at the RF connector is desirable**
  - Not easy to maintain the thickness of the gold during brazing process (Gold will be diffused into the bulk metal..)

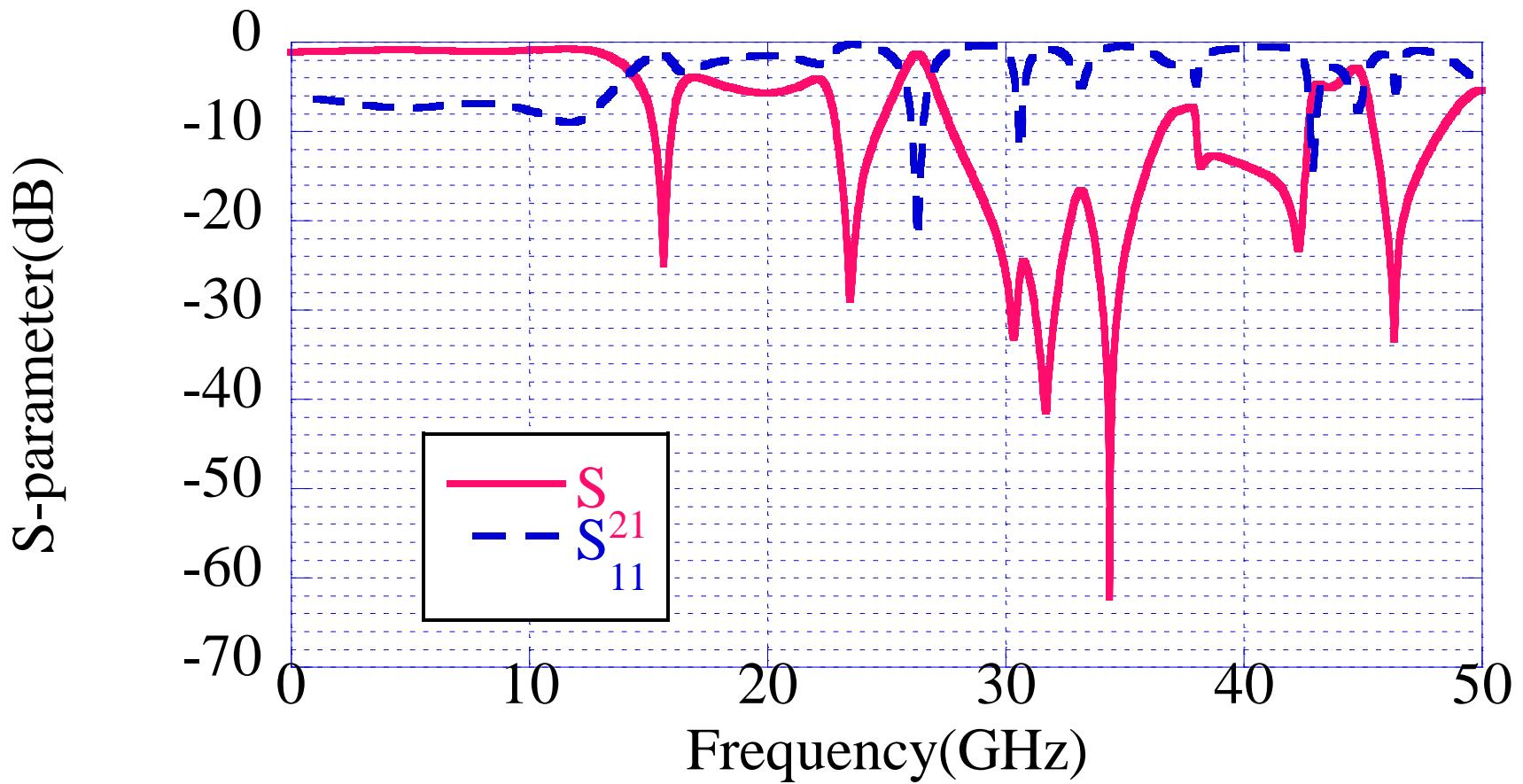
# SMA-J and SMA-J(R)



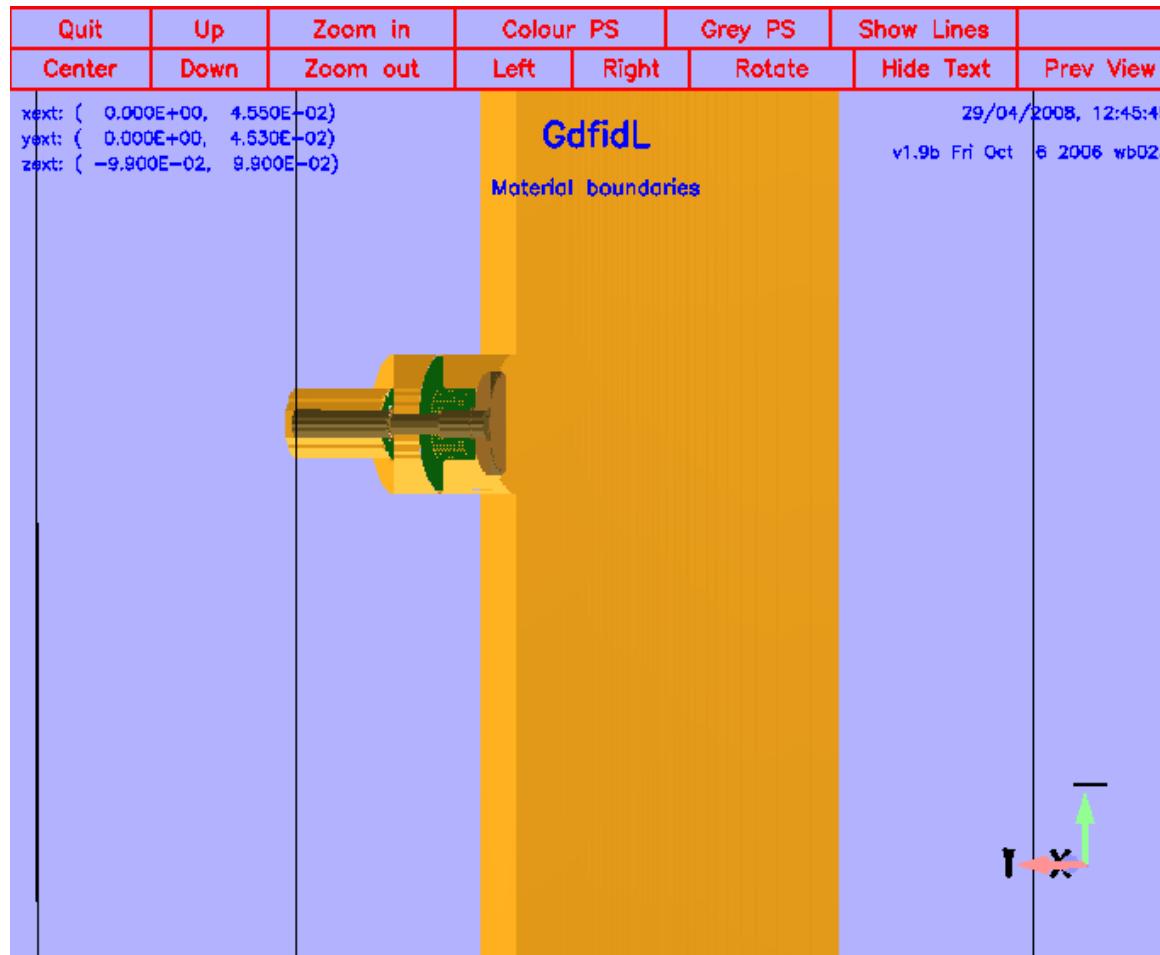
# HFSS model



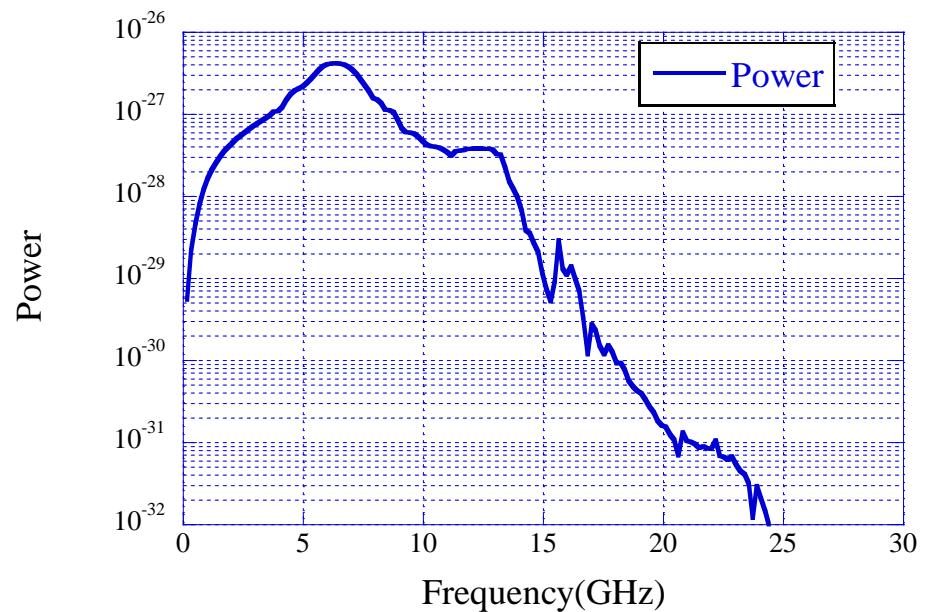
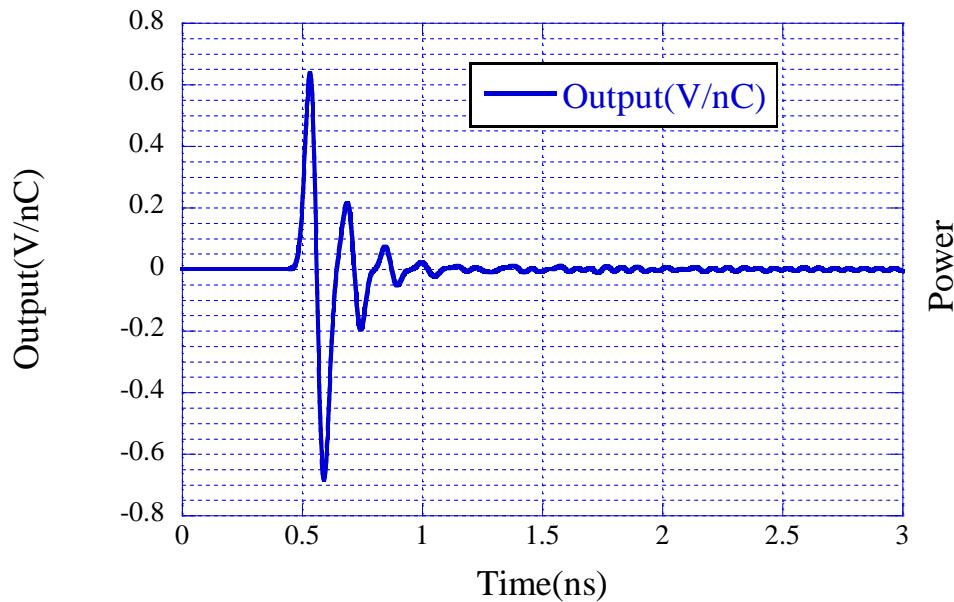
# Simulated S-Parameters



# GdfidL model



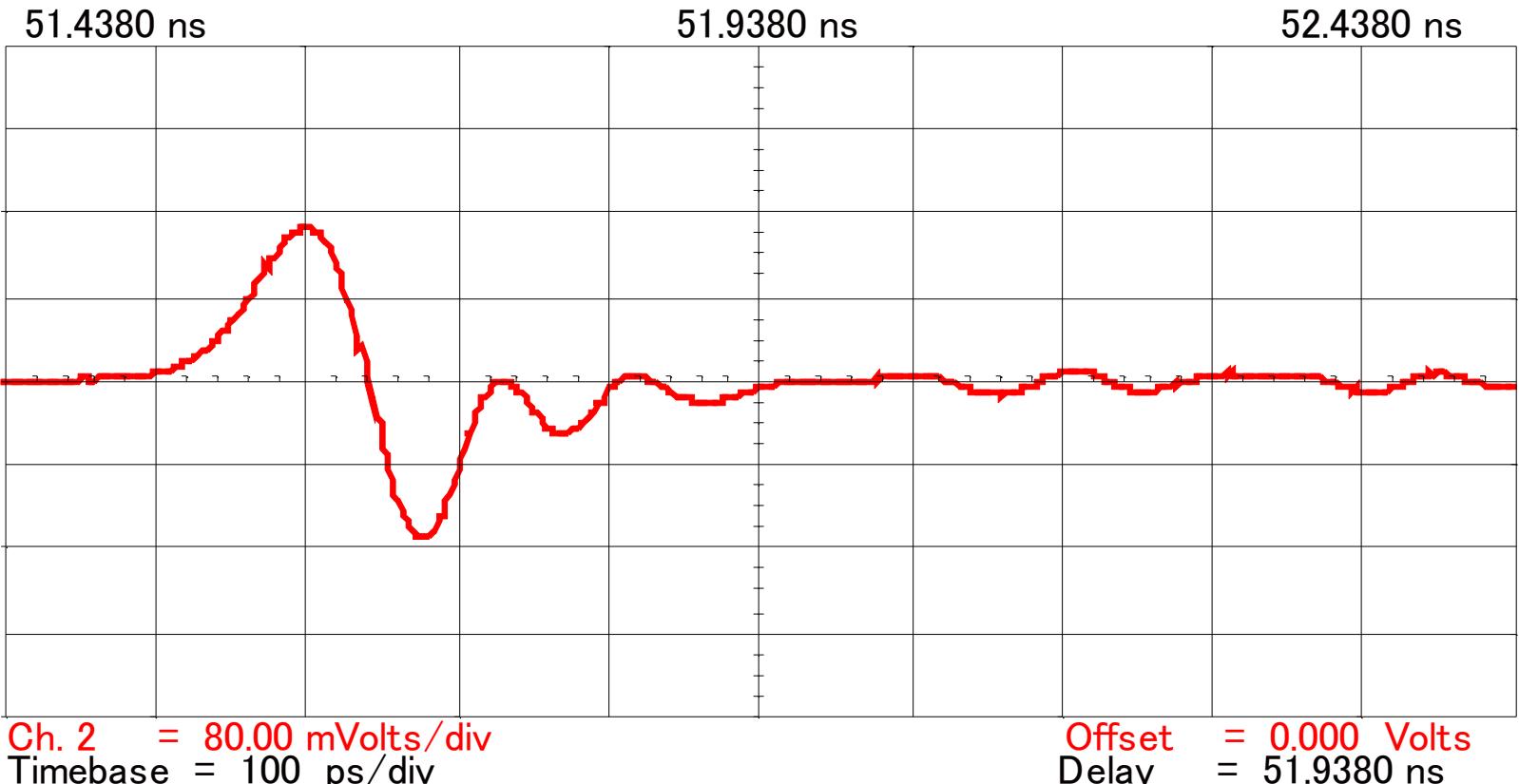
# Simulated Port output



# Tips

- Need to pay attention to the possible trapped modes in the sealing structure, which may damage the seal on high beam current operation.
  - Use Eigen-mode solver to estimate the frequency of the trapped mode.
  - Fine frequency sweep to find dip structure in S11
  - Coupling to coaxial mode might be difficult.
- Need to pay attention to the multipactoring around the button gap, especially for high single-bunch current machine.
- Longitudinal beam coupling impedance might not be negligible on large circumference, high beam current machine.

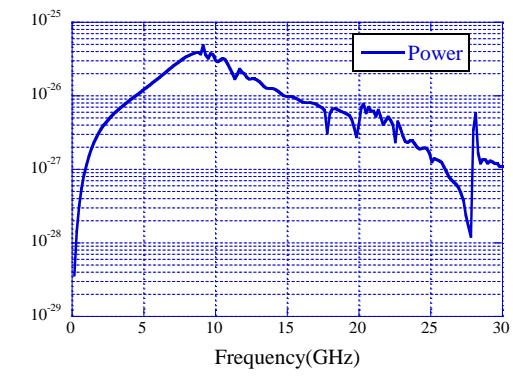
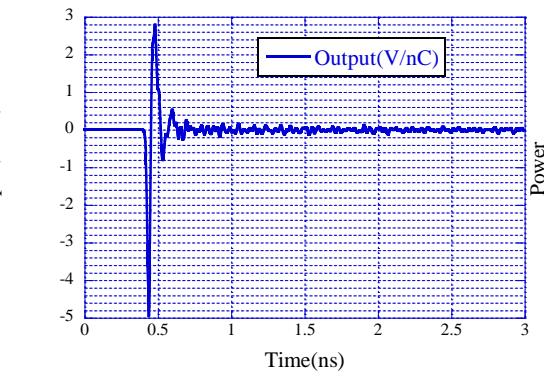
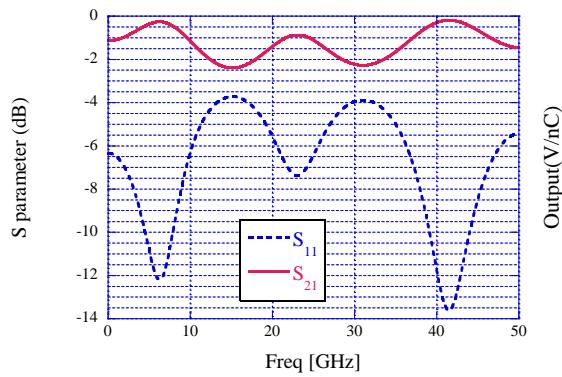
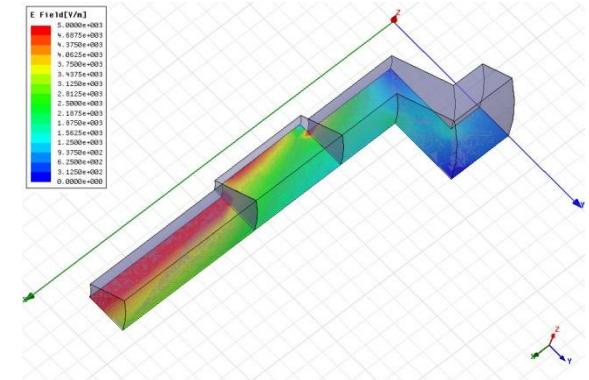
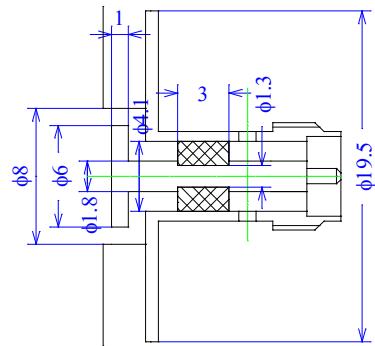
# Measured signal (20GHz scope)



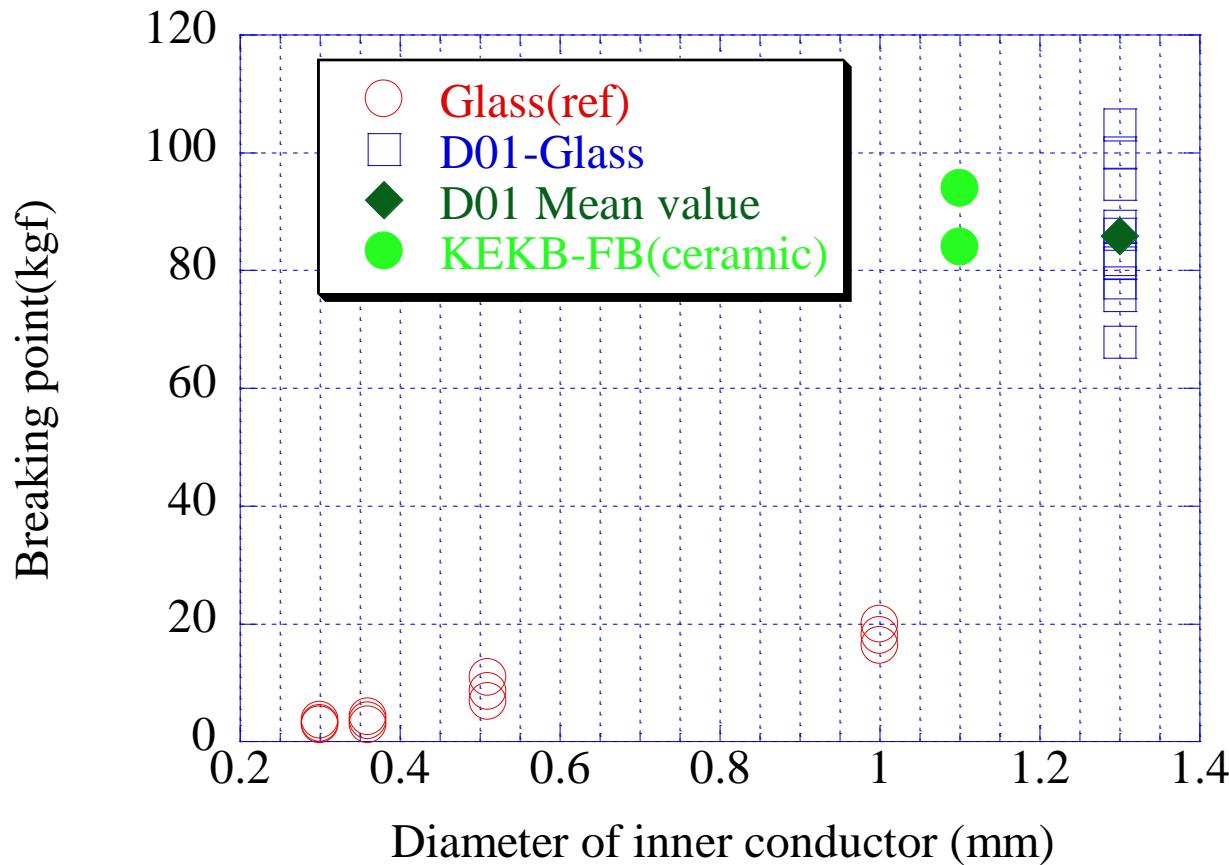
9nC/bunch LER 43dB attenuation, LER upstream

Maximum beam current at KEKB:  $2A + 1.3A = 3.3A$  with large horizontal offset

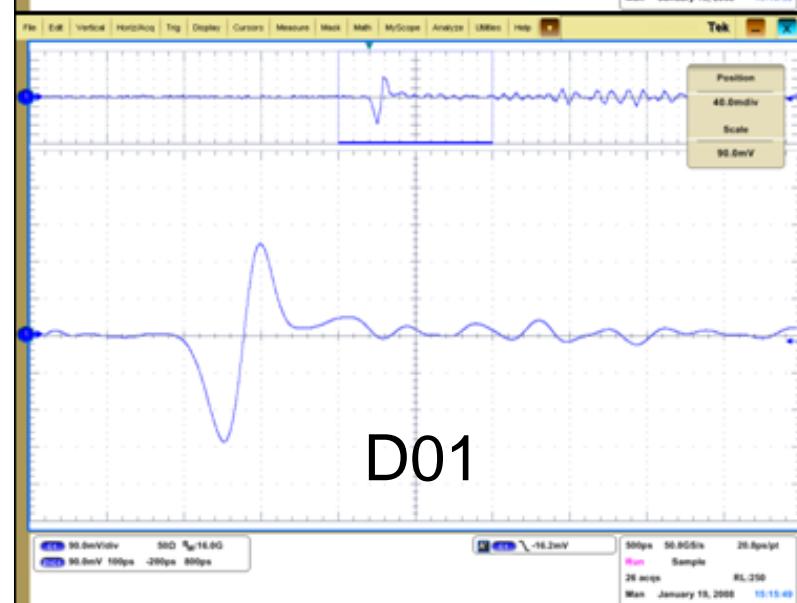
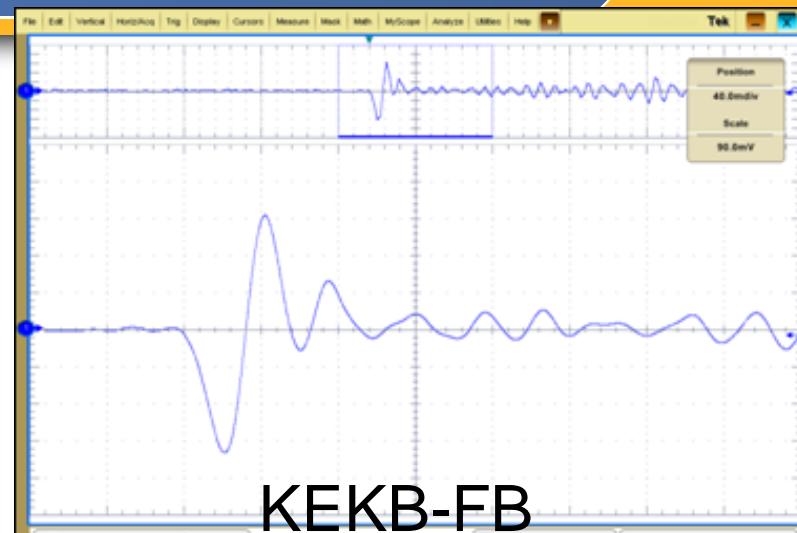
# Feedthrough with low $\epsilon_r(\sim 4)$ sealing



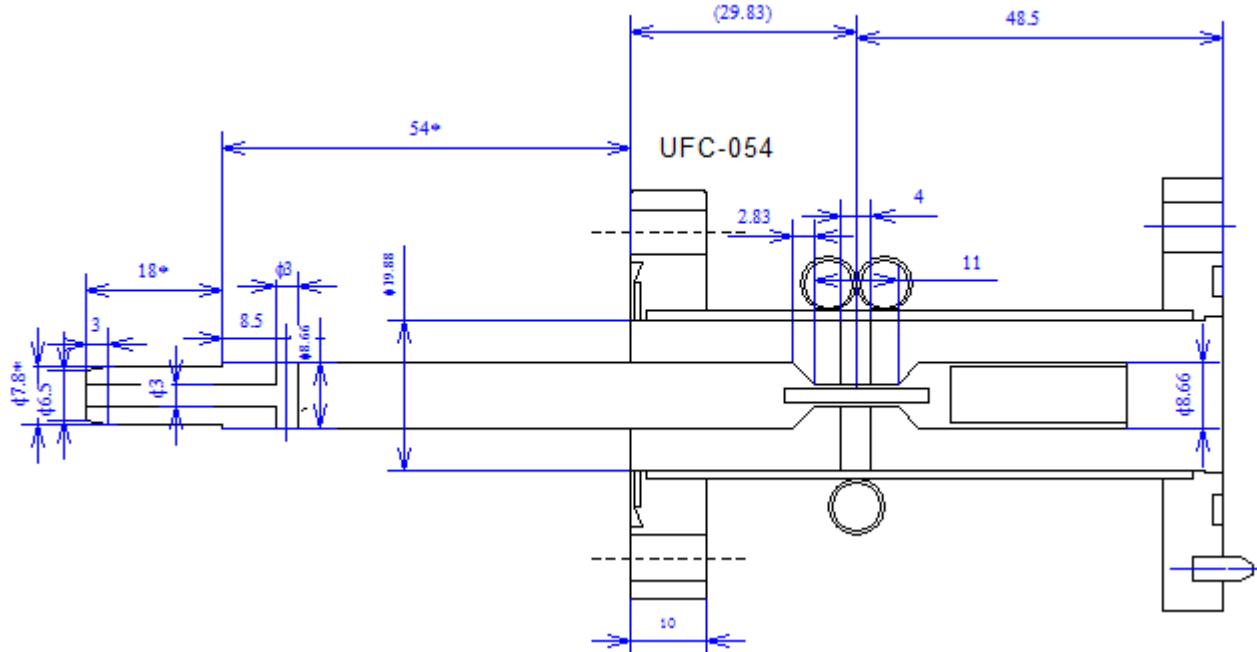
# Center pin pulling-out test



# Beam test using Linac short pulse



# High power feedthrough

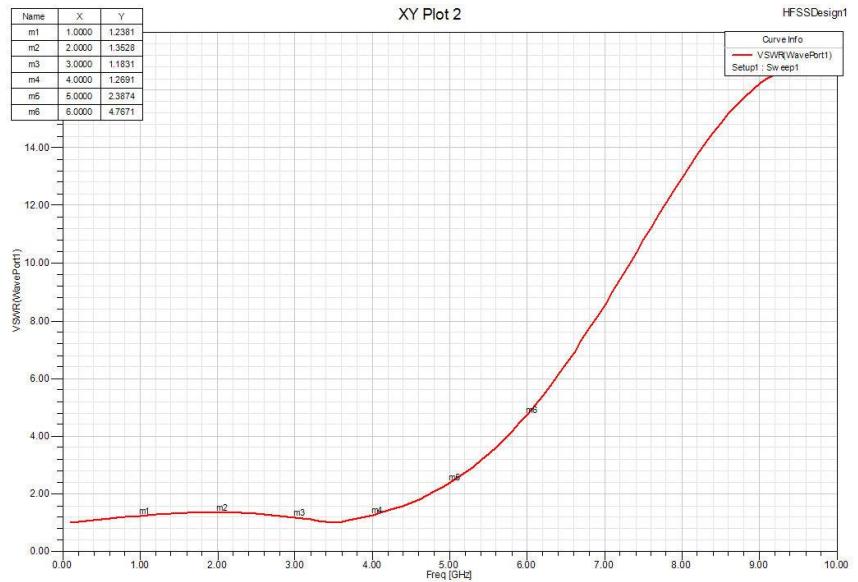
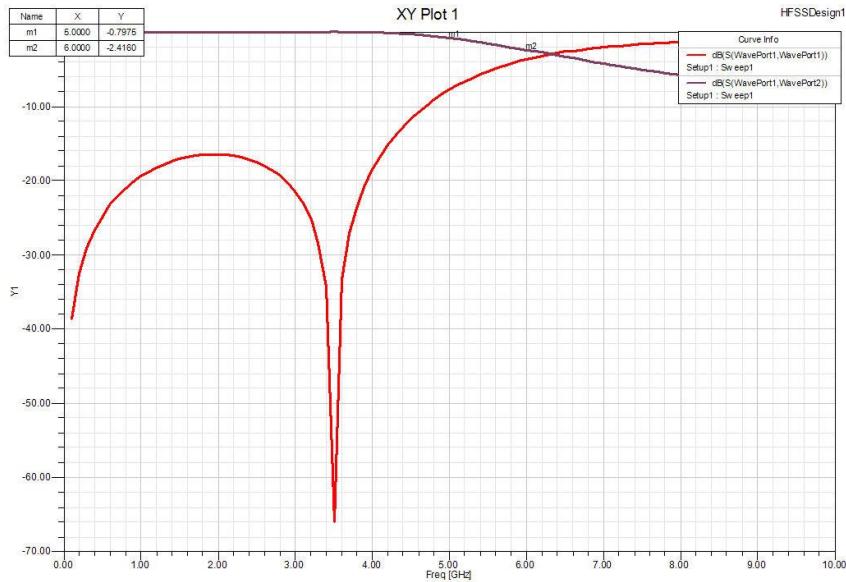


- Max CW power  $\sim$  5kW
  - Not so wideband (DC–3GHz)
  - SWR through pass band <1.1

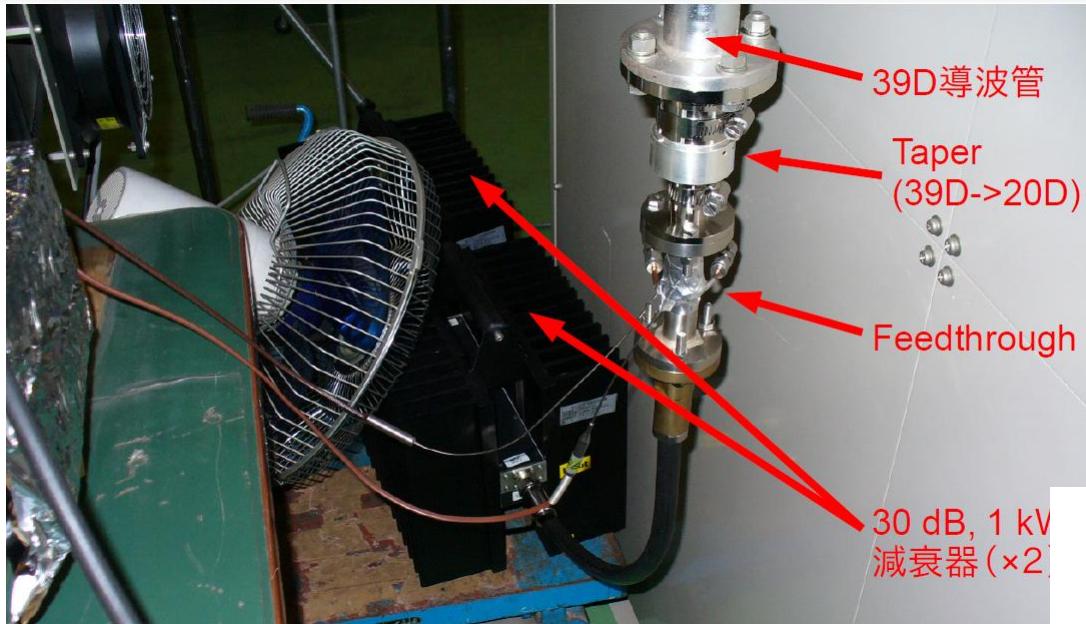
# Simulated S-Parameters

S21 and S11

VSWR(S11)



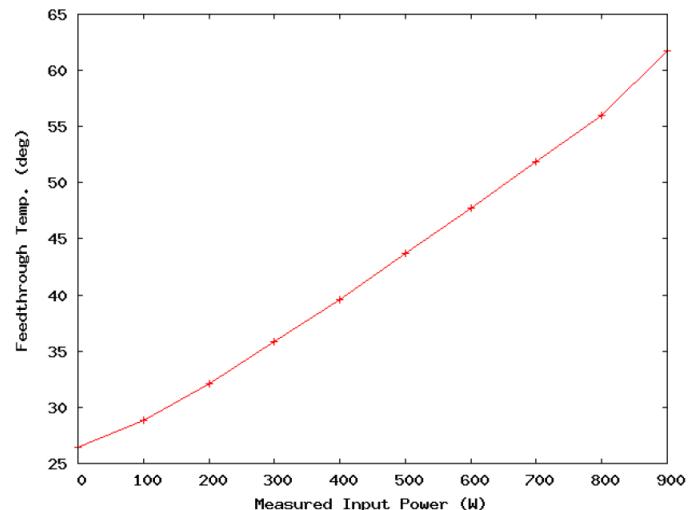
# High power test



1.3GHz power source

- Temperature rise : 4 deg/100W  
(without water cooling)

Feedthrough 温度 対 入力 パワー



冷却水がなくても900 Wまでは異常無し。

# Tips

- To enhance thermal conductivity between the center conductor and the outer wall, it might be considered to fill BN ( $\varepsilon_R \sim 4$ ) at the air side.
- $\text{Si}_3\text{N}_4$  has lower  $\varepsilon_R$  and has better (lower) loss tangent
  - Expensive
  - Mechanically weaker than alumina-ceramic

# Summary

- **Briefly reviewed the transmission-line theory**
  - S-Parameters, SWR, Time-domain response
- **Introduced E-M simulation software**
  - HFSS (Frequency domain)
  - GdfidL (Time domain)
- **Shown several examples**
  - KEKB-FB button electrode
  - SuperKEKB-FB button electrode
  - High power feedthrough for SuperKEKB longitudinal kicker.

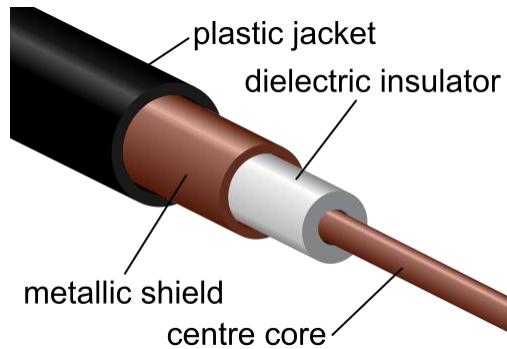
# Reference

- “Microwave Engineering Passive Circuits”,  
by Peter A. Rizzi, ISBN 0135867029
- “Introduction to Microwave Theory”,  
by Harry A. Atwater, ISBN 0898741920
- Ansys HFSS,
  - <http://www.ansys.com/Products/Simulation+Technology/Electromagnetics/High-Performance+Electronic+Design/ANSYS+HFSS>
- GdfidL, <http://www.gdfidl.de/>
- Kyocera Corporation
  - <http://global.kyocera.com/prdct/fc/product/usage/vacuum/index.html>
- Orient Microwave Corp.
  - <http://www.orient-microwave.com/english/index.html>
- SUHNER
  - <http://ipaper.ipapercms.dk/HUBERSUHNER/Technologies/Radiofrequency/RFConnectorsEN/>

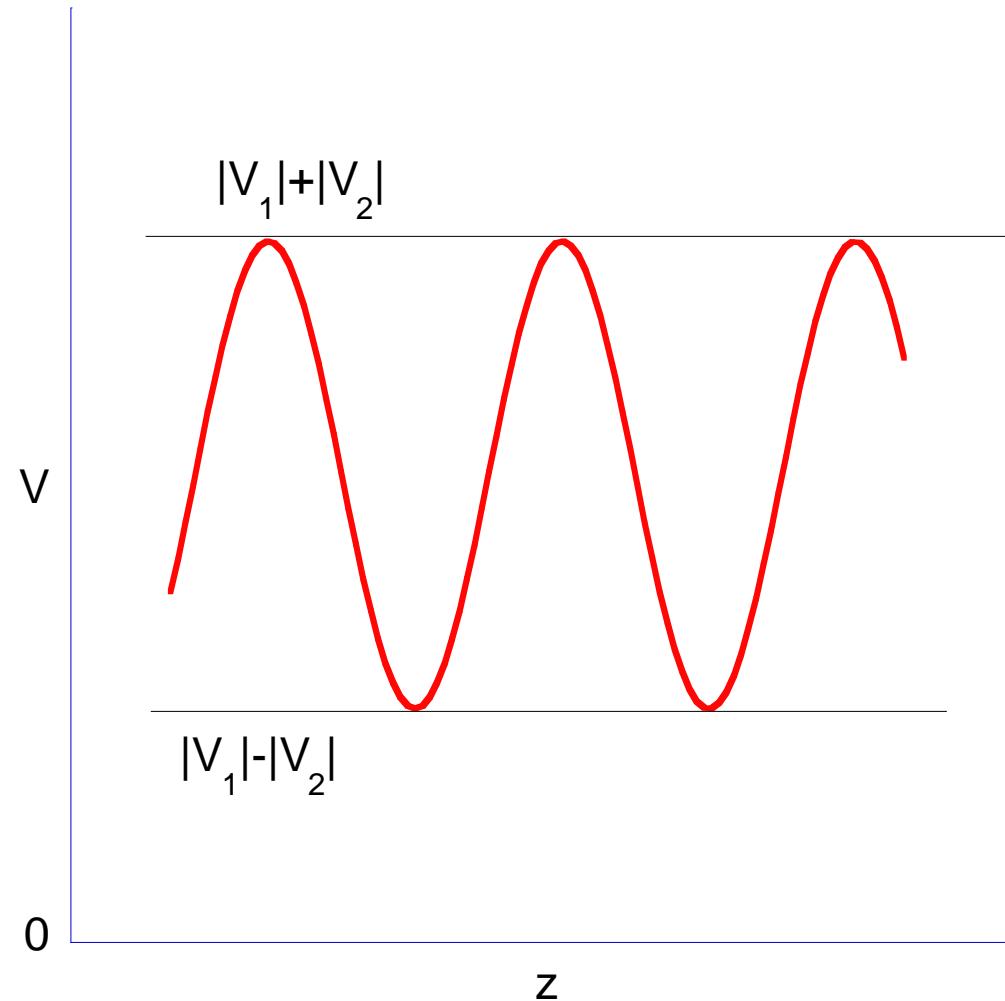
# backup

# Transmission line

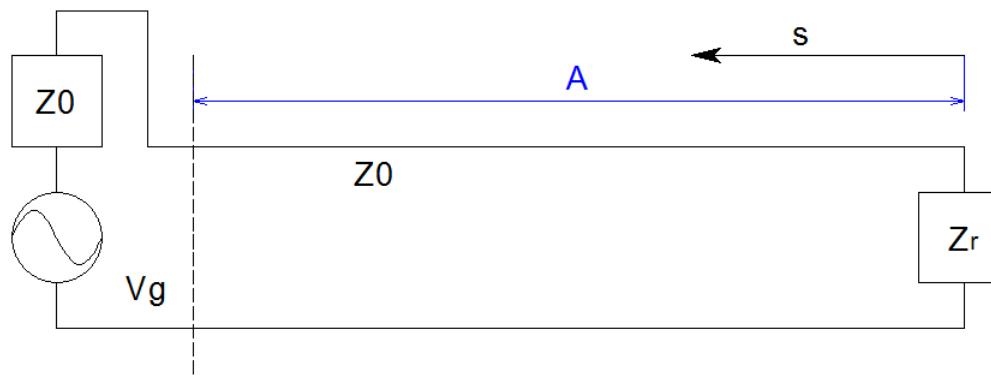
- Coaxial cable



# Standing wave

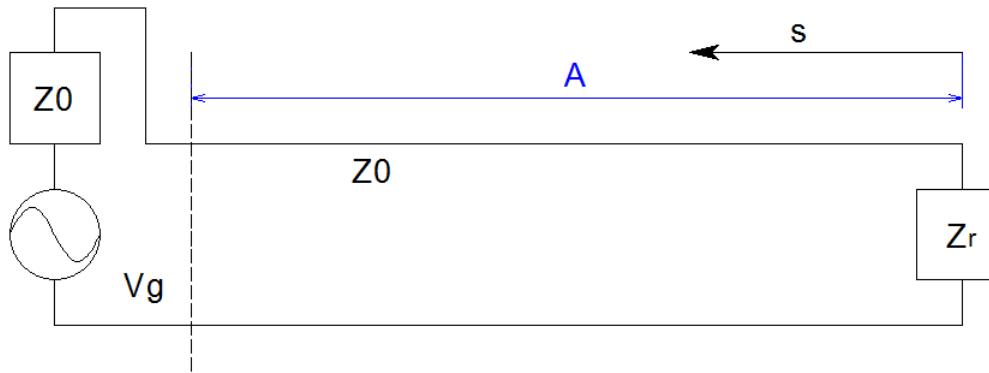


# Standing wave



$$\begin{aligned} V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V'_1 e^{j\beta s} + V'_2 e^{-j\beta s} \\ &= V'_1 e^{j\beta s} \left(1 + \rho e^{-2j\beta s}\right) \end{aligned}$$

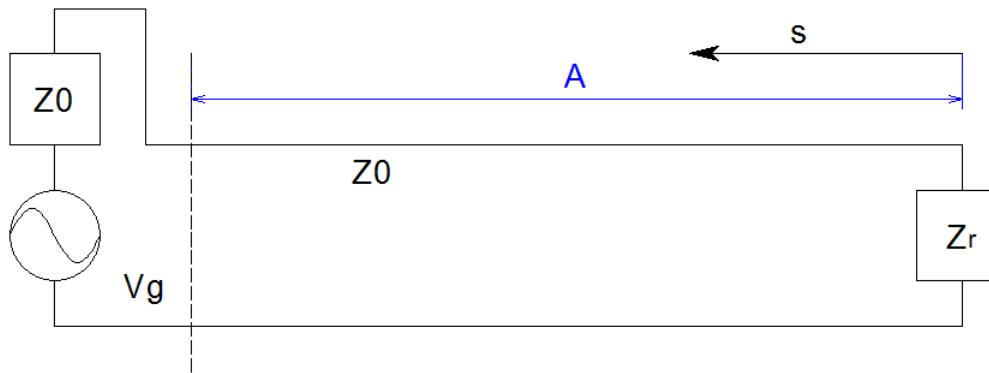
# Standing wave



$$\begin{aligned} V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V'_1 e^{j\beta s} + V'_2 e^{-j\beta s} \\ &= V'_1 e^{j\beta s} \left(1 + \rho e^{-2j\beta s}\right) \end{aligned}$$

- $\rho$ : reflection constant

# Standing wave



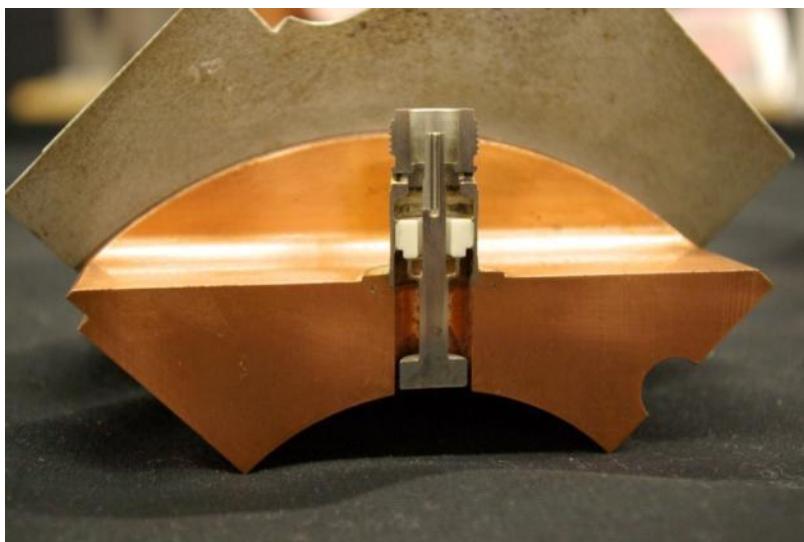
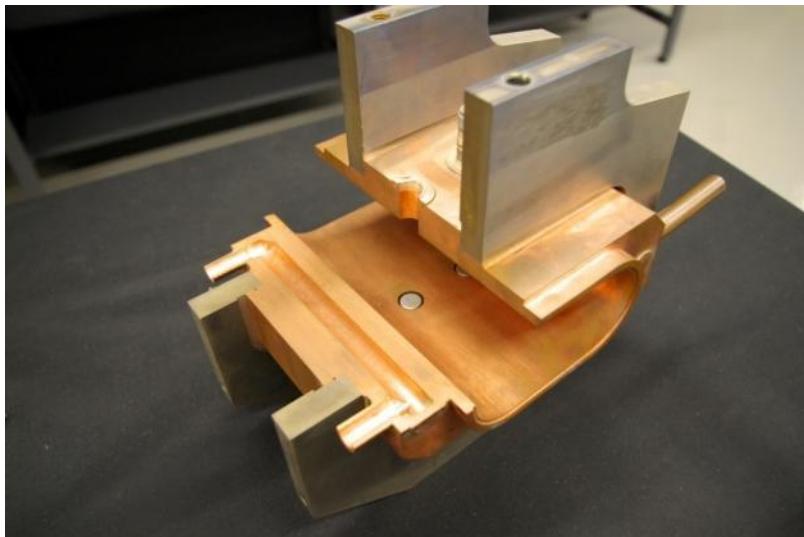
$$\begin{aligned} V &= V_1 e^{-j\beta(A-s)} + V_2 e^{j\beta(A-s)} = V'_1 e^{j\beta s} + V'_2 e^{-j\beta s} \\ &= V'_1 e^{j\beta s} \left(1 + \rho e^{-2j\beta s}\right) \end{aligned}$$

- $\rho$ : reflection constant

$$\rho = \frac{V'_2}{V'_1} = |\rho| e^{j\varphi}$$

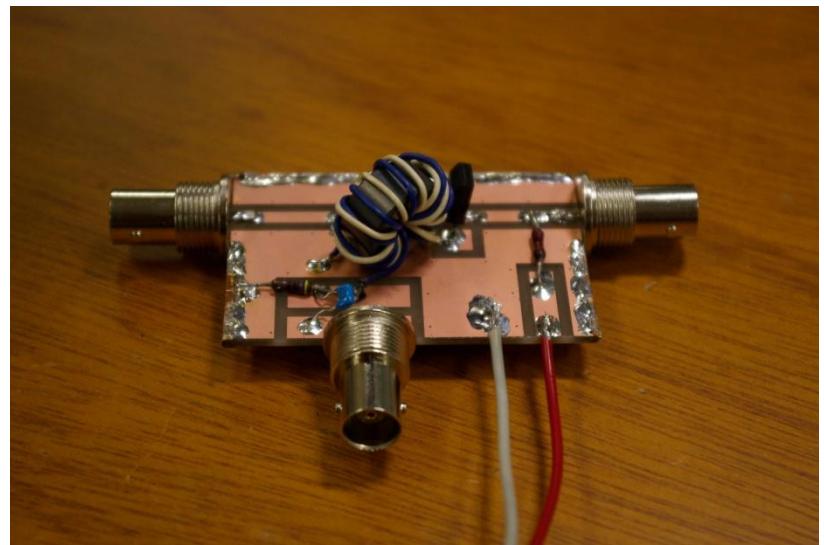
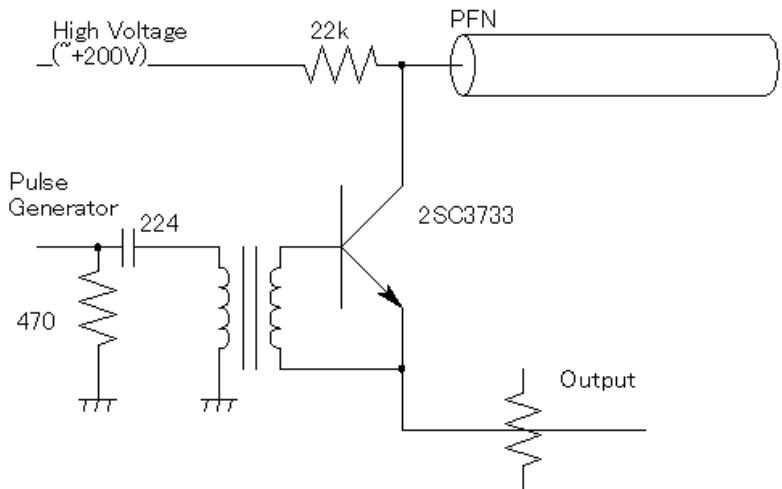
$$V = V'_1 e^{j\beta s} \left(1 + |\rho| e^{j(\varphi - 2\beta s)}\right)$$

# 例：ボタン電極型ビーム位置モニタ

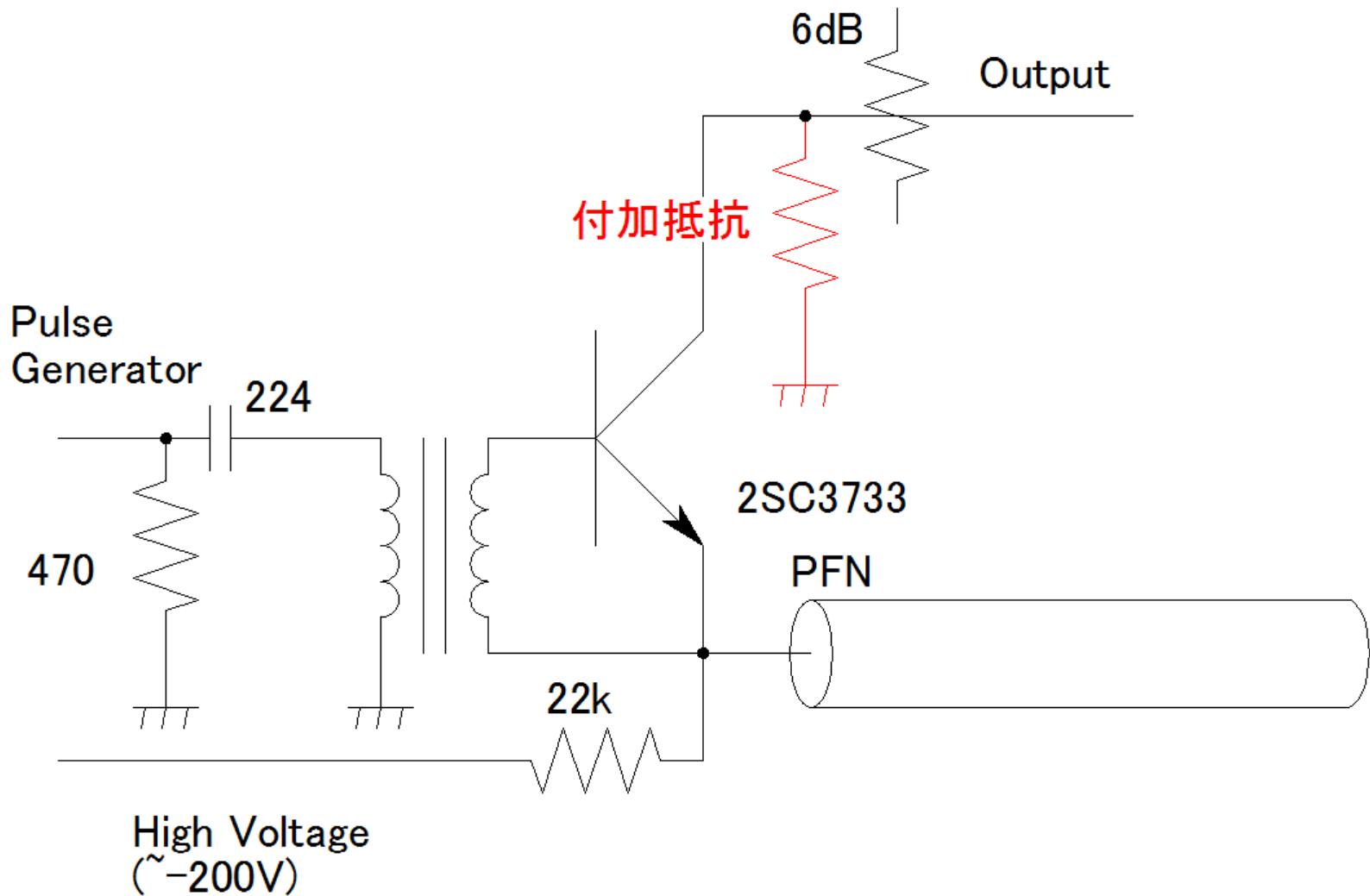


四つの電極に出てきた電荷の差から、バンチの重心位置を求めるためのモニターです。レンズの働きをする四極電磁石に固定してあります。

# アバランシェトランジスタパルサー(正極性)

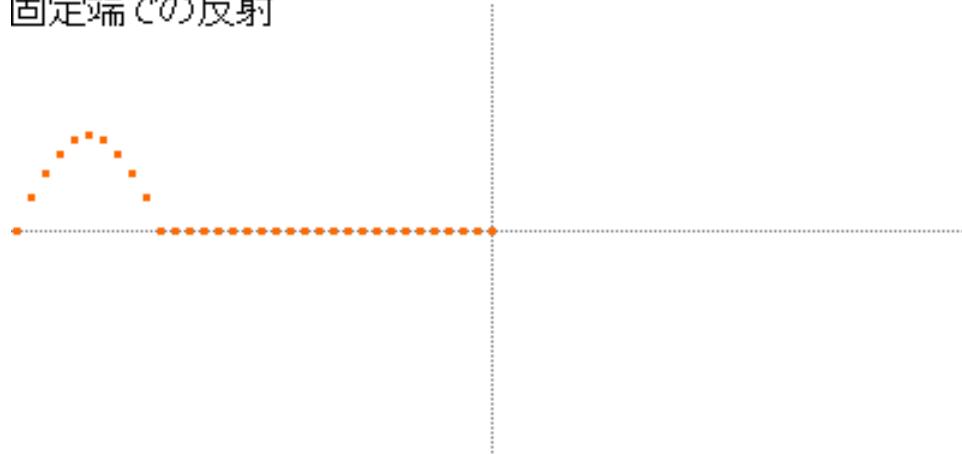


# アバランシェパルサー(負極性)

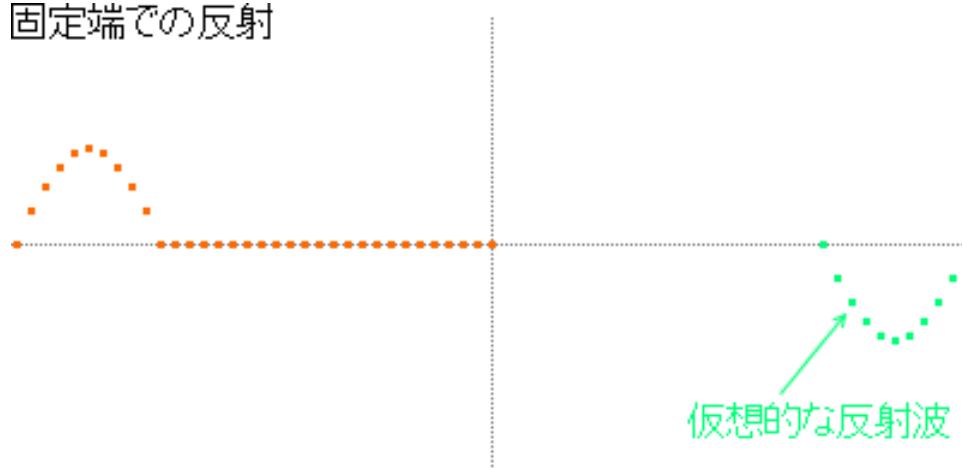


# 固定端での反射

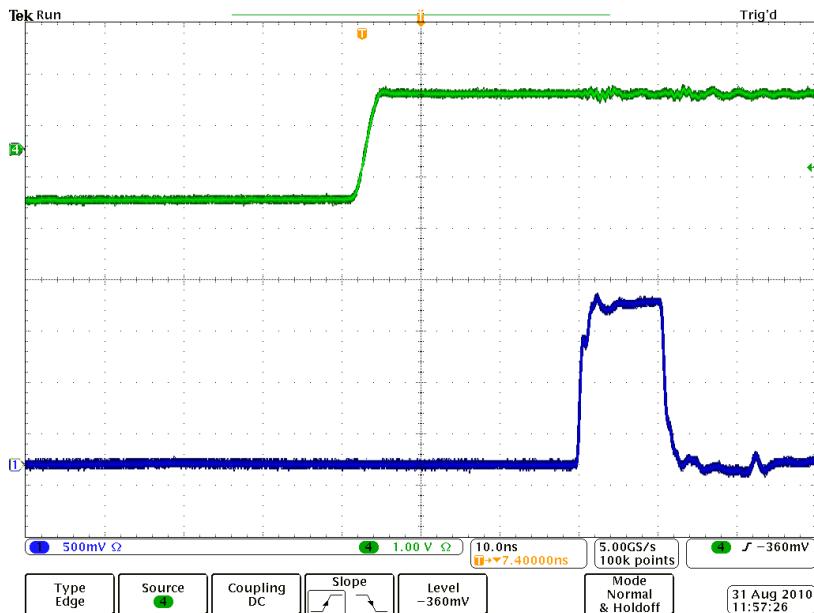
固定端での反射



固定端での反射

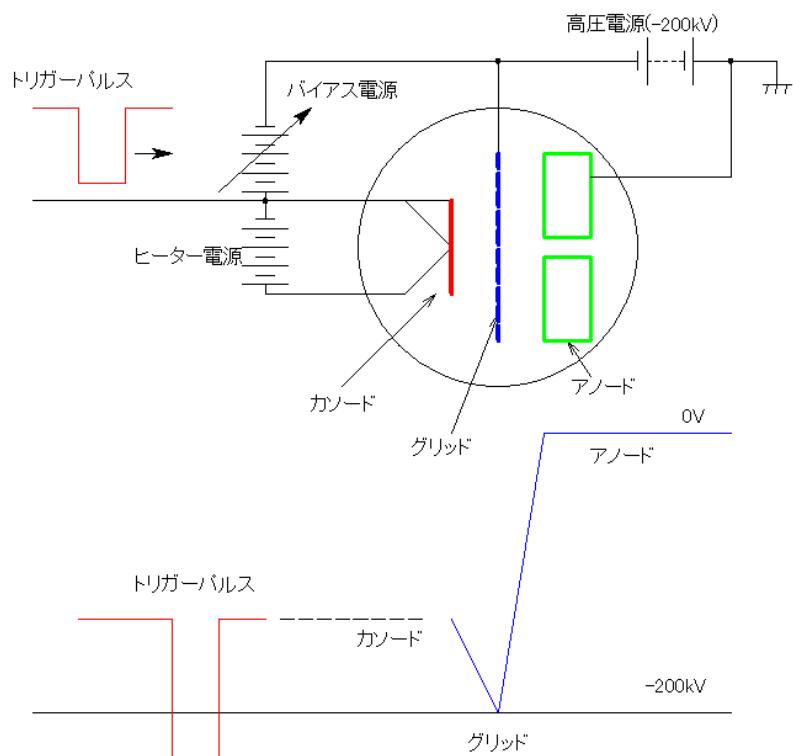


# 出力は

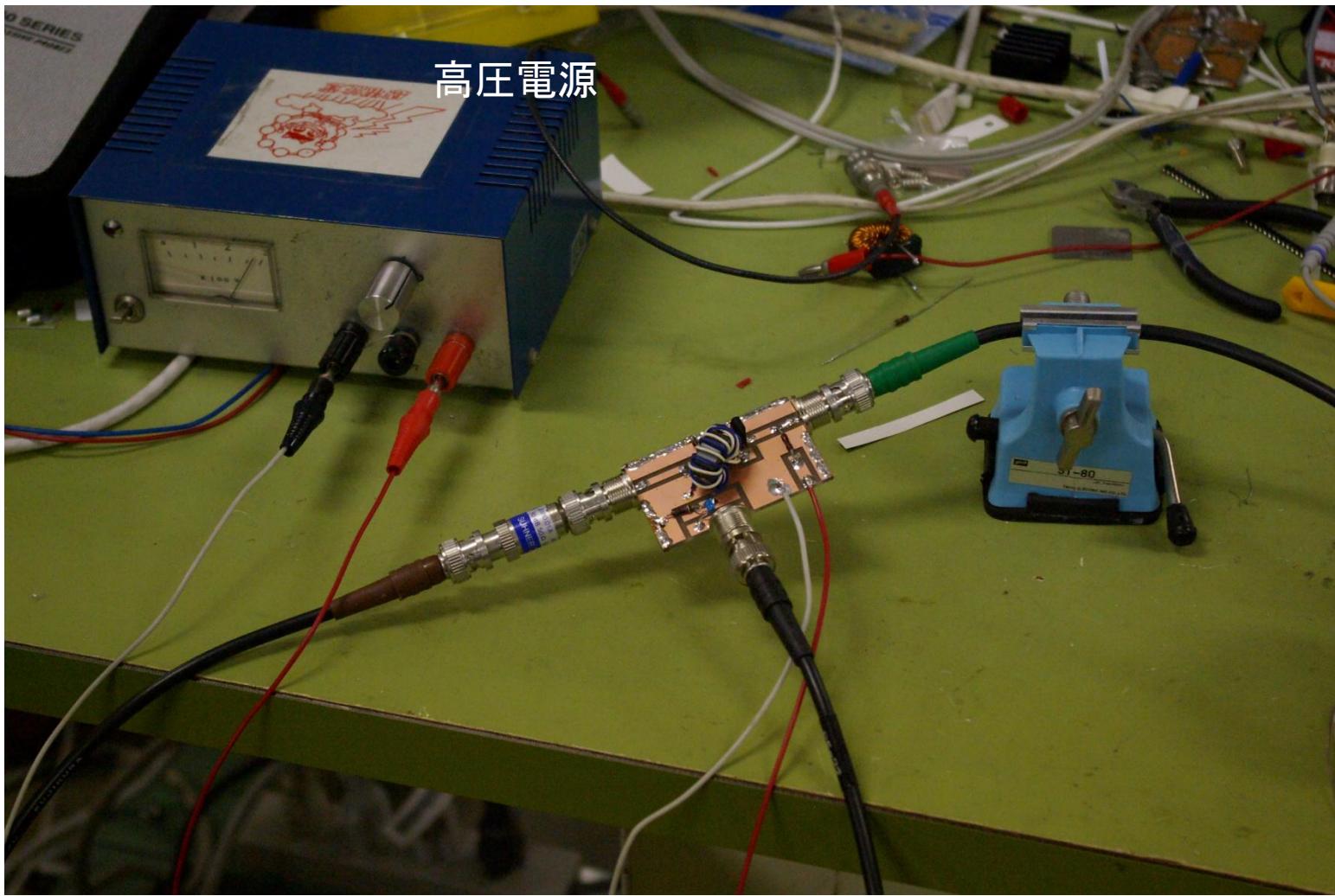


- トリガーが入ってトランジスタがONになり、約10ns後にOFFになっている。
- このとき、PFNには1mの同軸線をつけていた。ここにたまっていた高電圧が抜けて、パルス高が0に戻った。

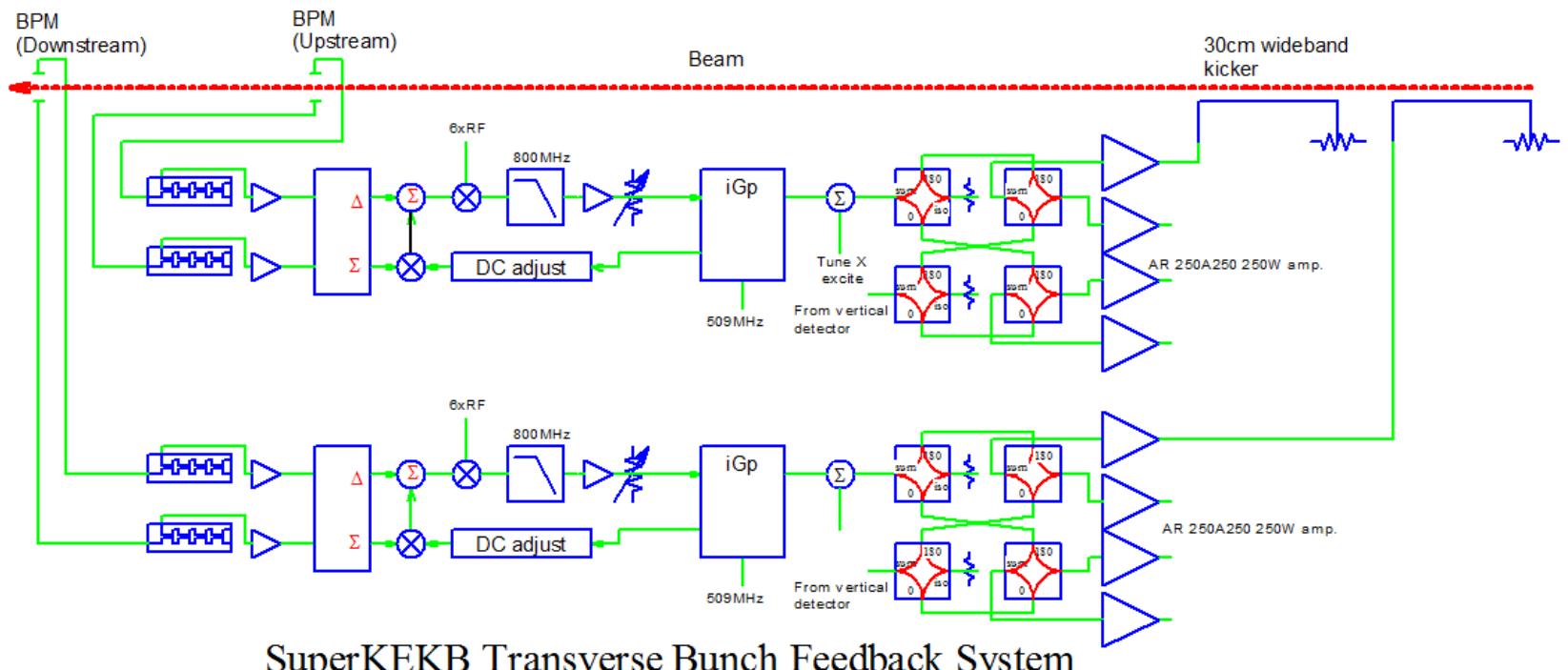
# アバランシェパルサーの使用例



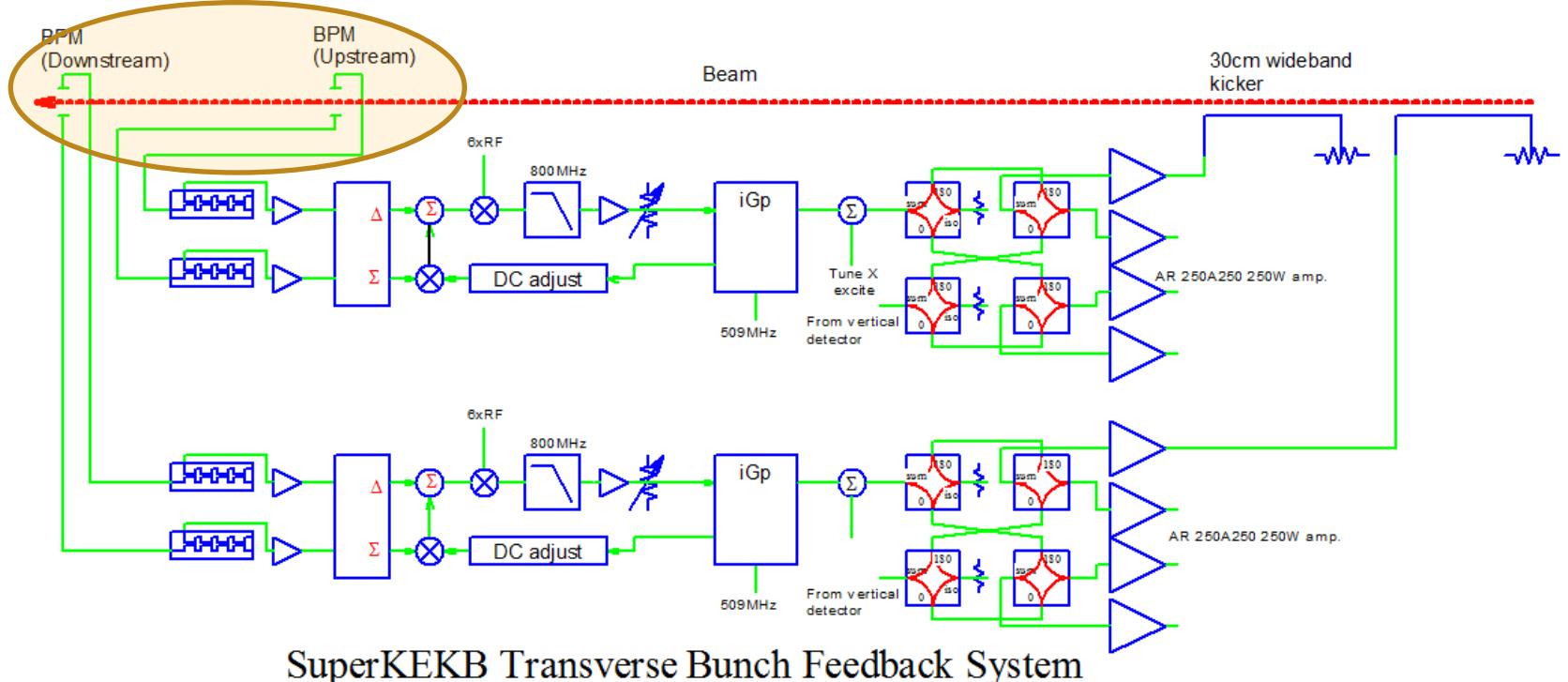
- 線形加速器の熱電子銃から出る電子ビームの長さを非常に狭く(1ns以下とか)したいとき、アバランシェパルサーを使ってビームが出るタイミングを制御します



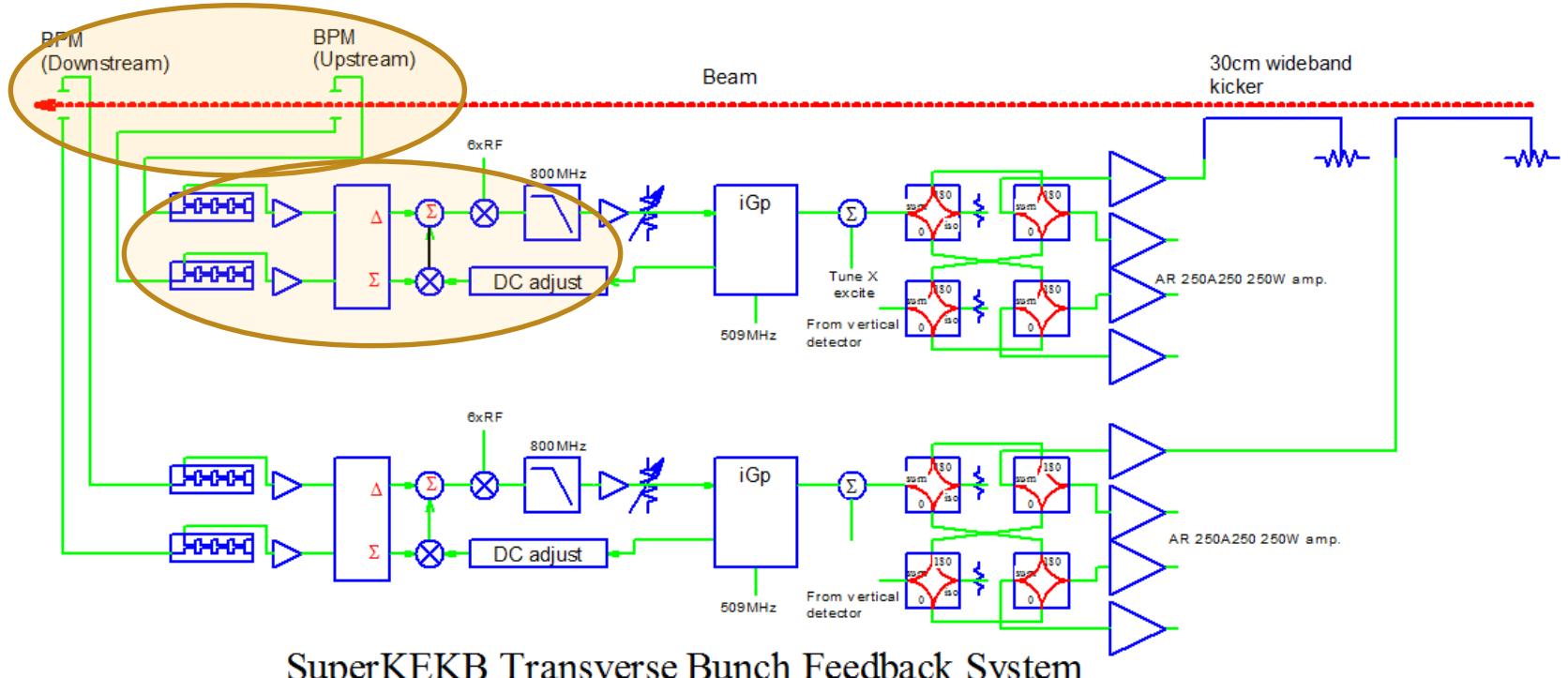
# SuperKEKB Transverse FB plan



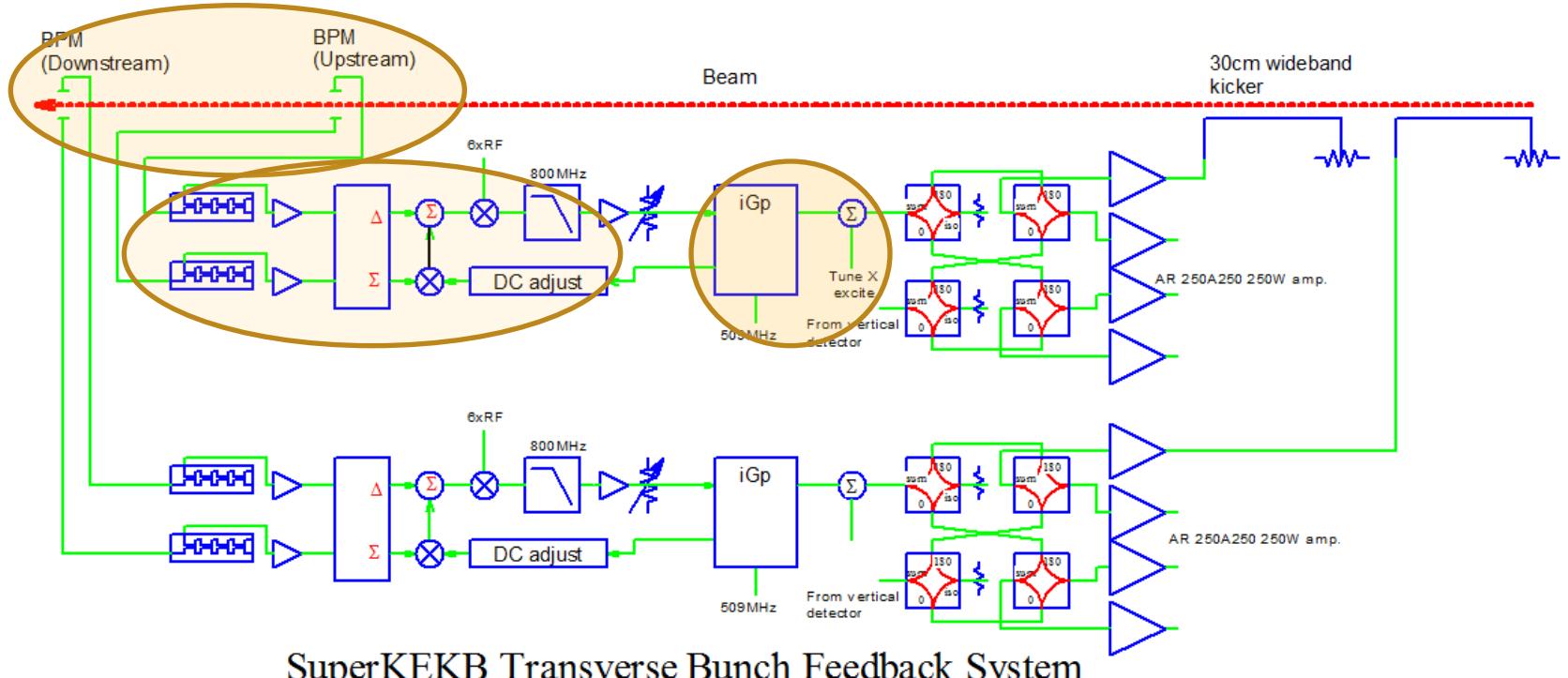
# SuperKEKB Transverse FB plan



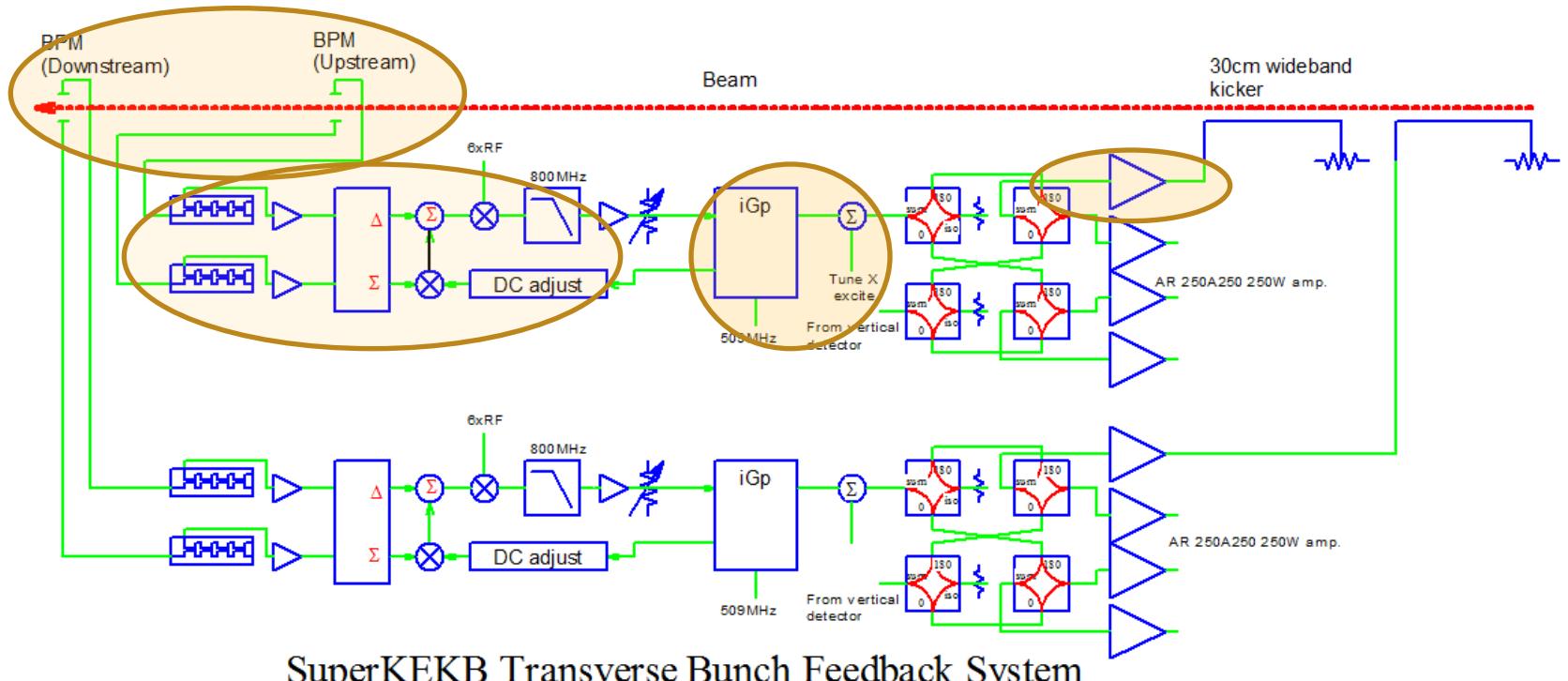
# SuperKEKB Transverse FB plan



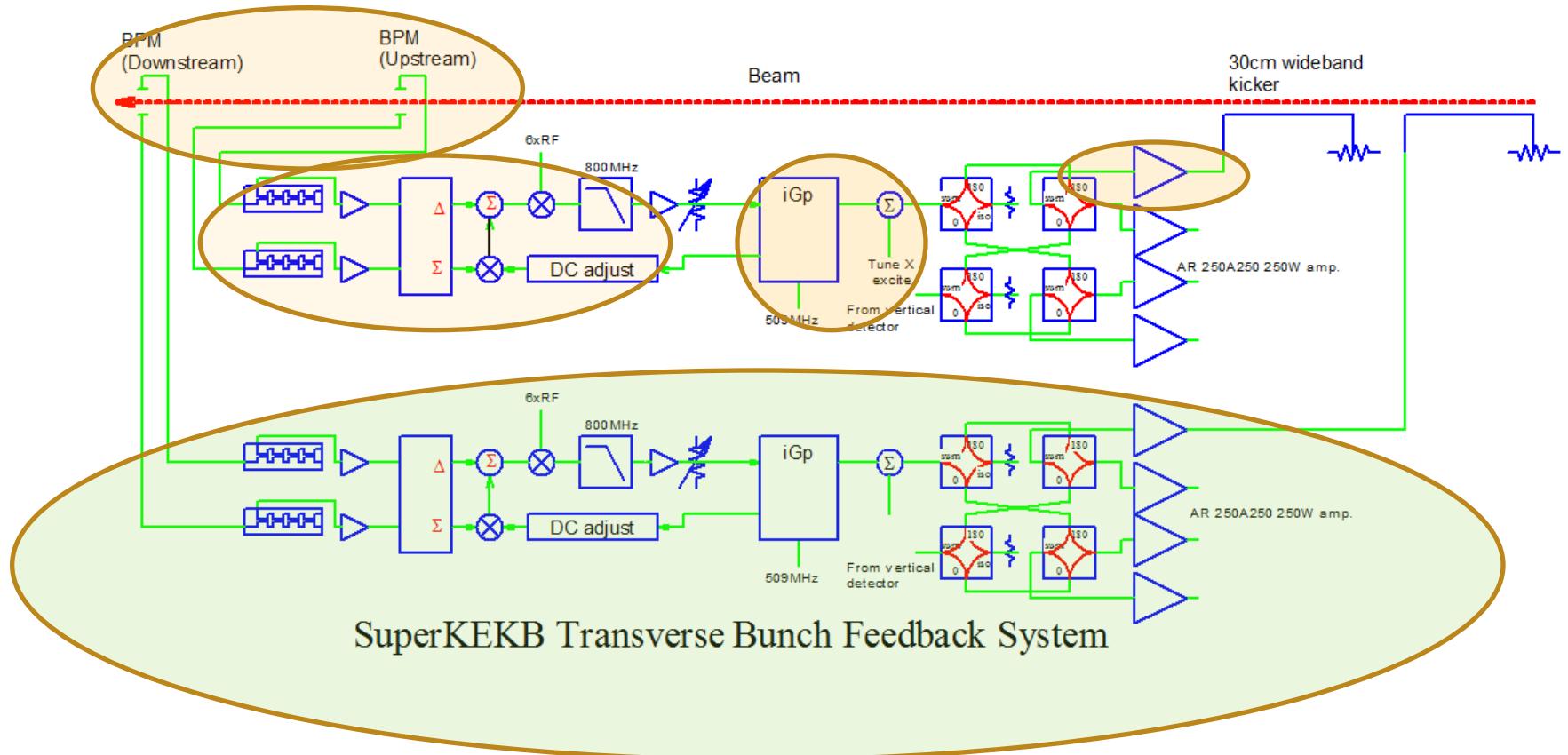
# SuperKEKB Transverse FB plan



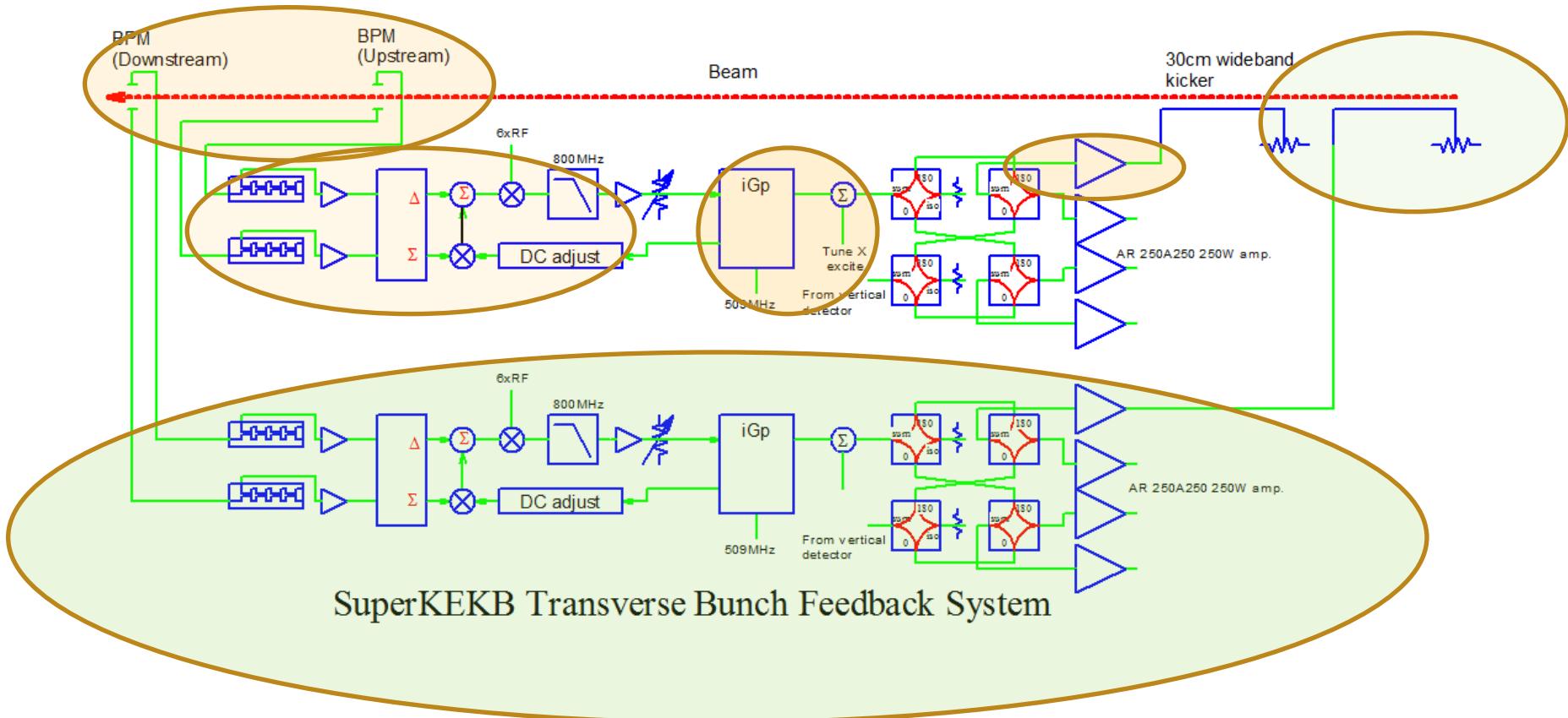
# SuperKEKB Transverse FB plan



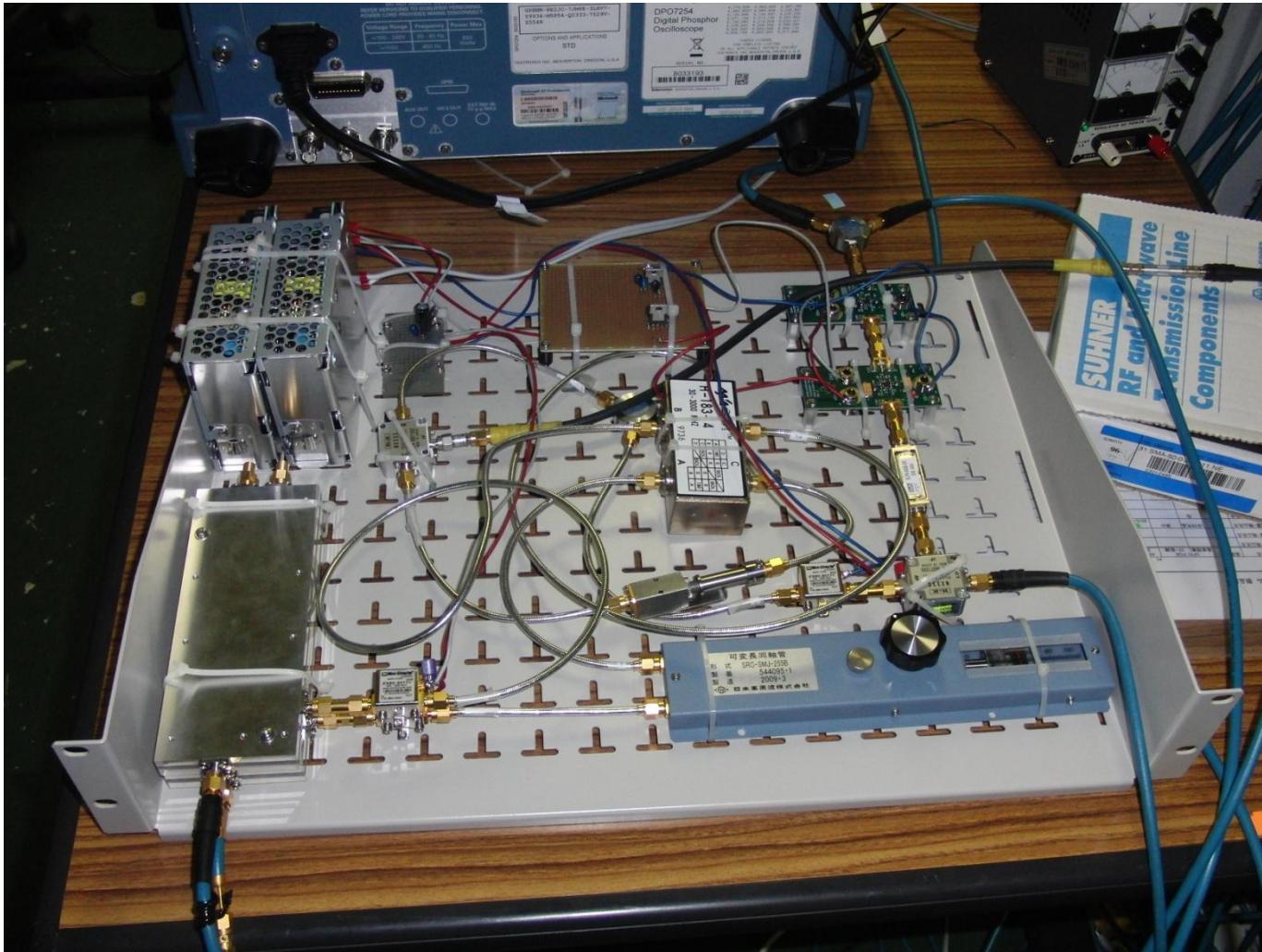
# SuperKEKB Transverse FB plan



# SuperKEKB Transverse FB plan



# Bunch position detector prototype



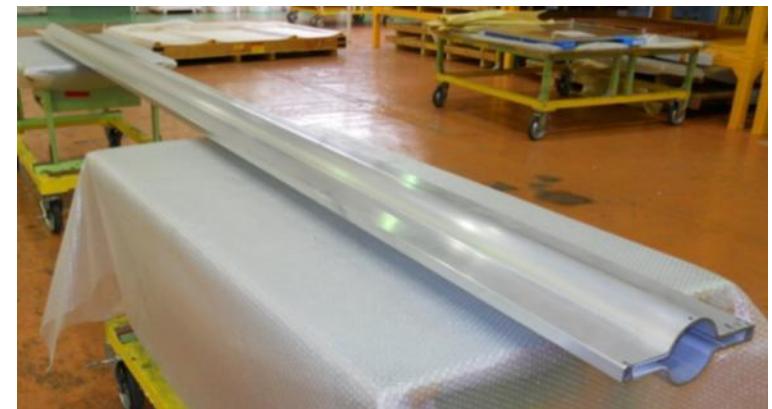
# Vacuum chamber

- Aluminum alloy antechamber
- cutoff frequency <1 GHz

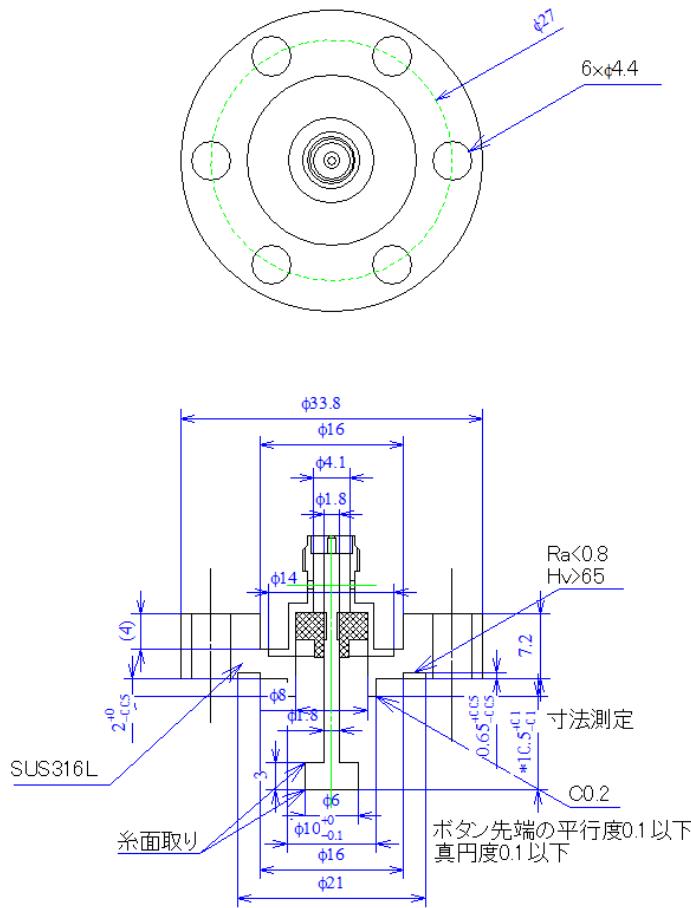
Aluminum-alloy duct



Aluminum-alloy duct

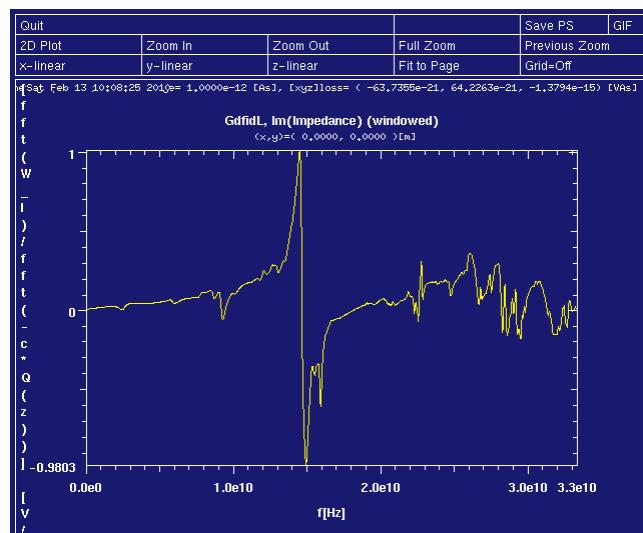
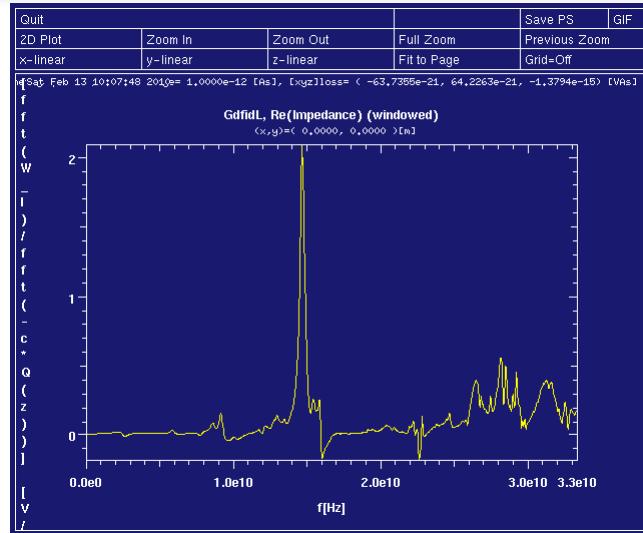
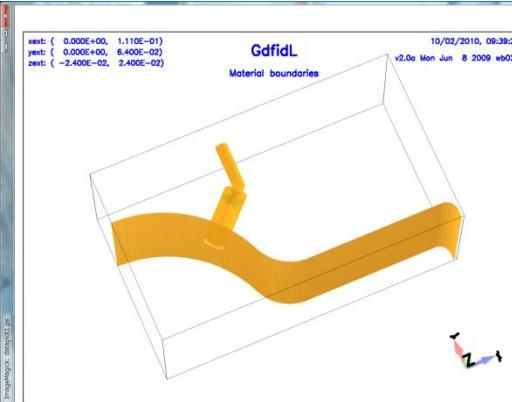


# BPM head

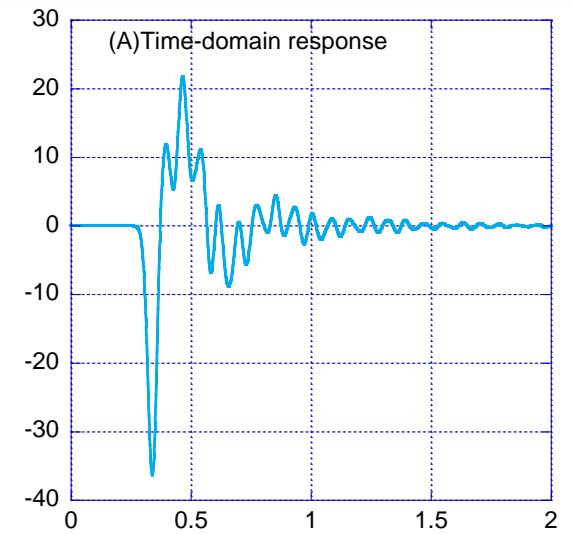


SuperKEKB用BPM model-E1  
作図: M.Tobiyama 6/Oct/2006  
修正: M.Tobiyama 8/Nov/2006

# Impedance/button output simulation



Output(V/mA)



Amplitude

