

LLRF controls and RF operations.

SRF 2019 Tutorial

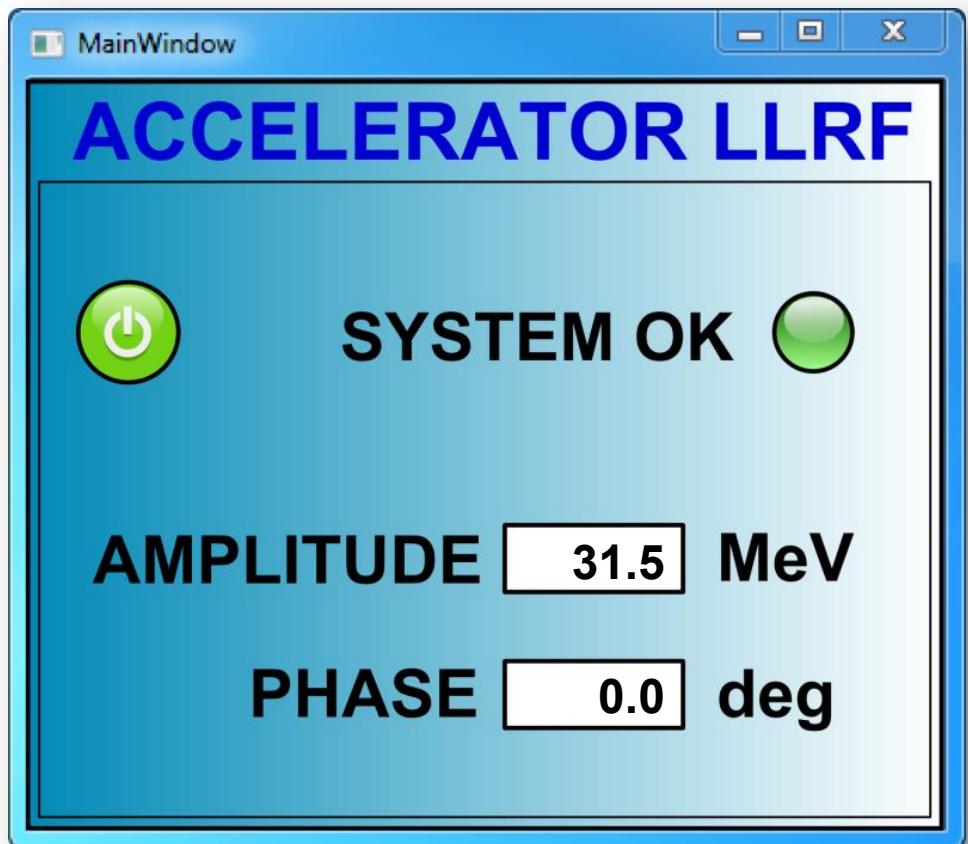


Julien BRANLARD
for the DESY LLRF team

Dresden, 29.06.2019

What is a LLRF system for?

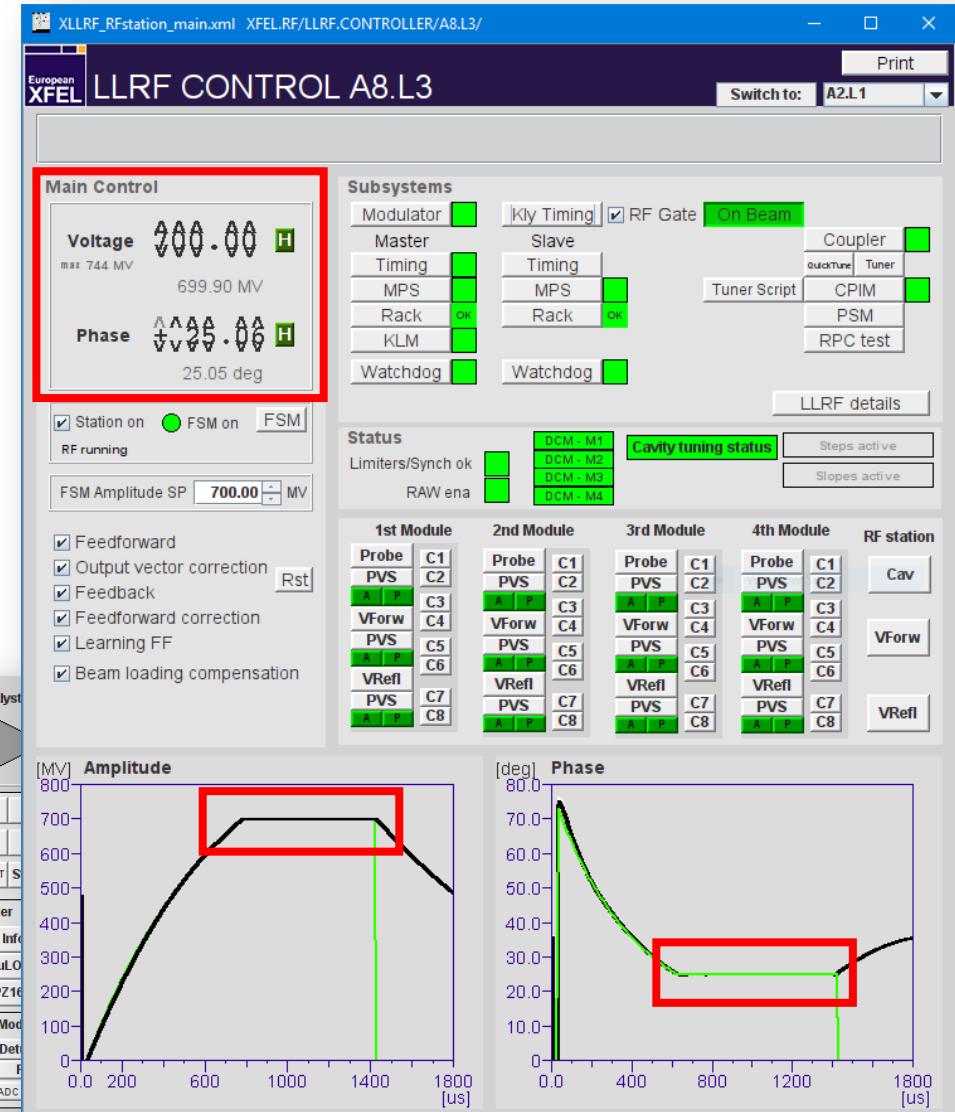
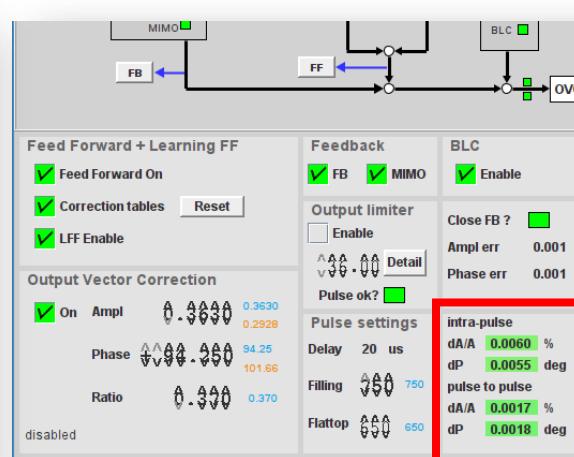
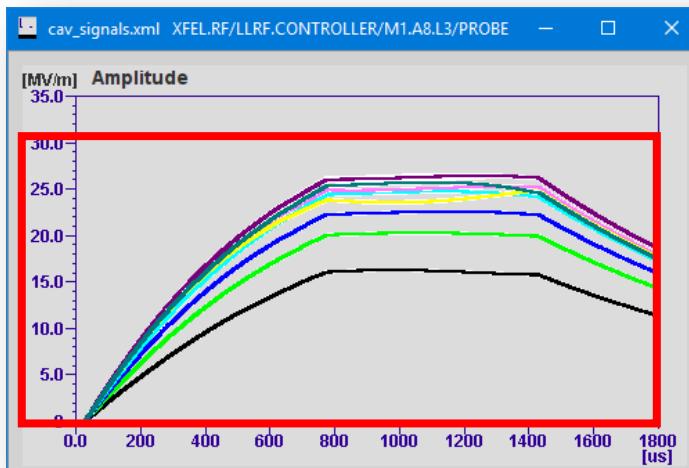
Interface to the “ultimate LLRF” system



What is a LLRF system for?

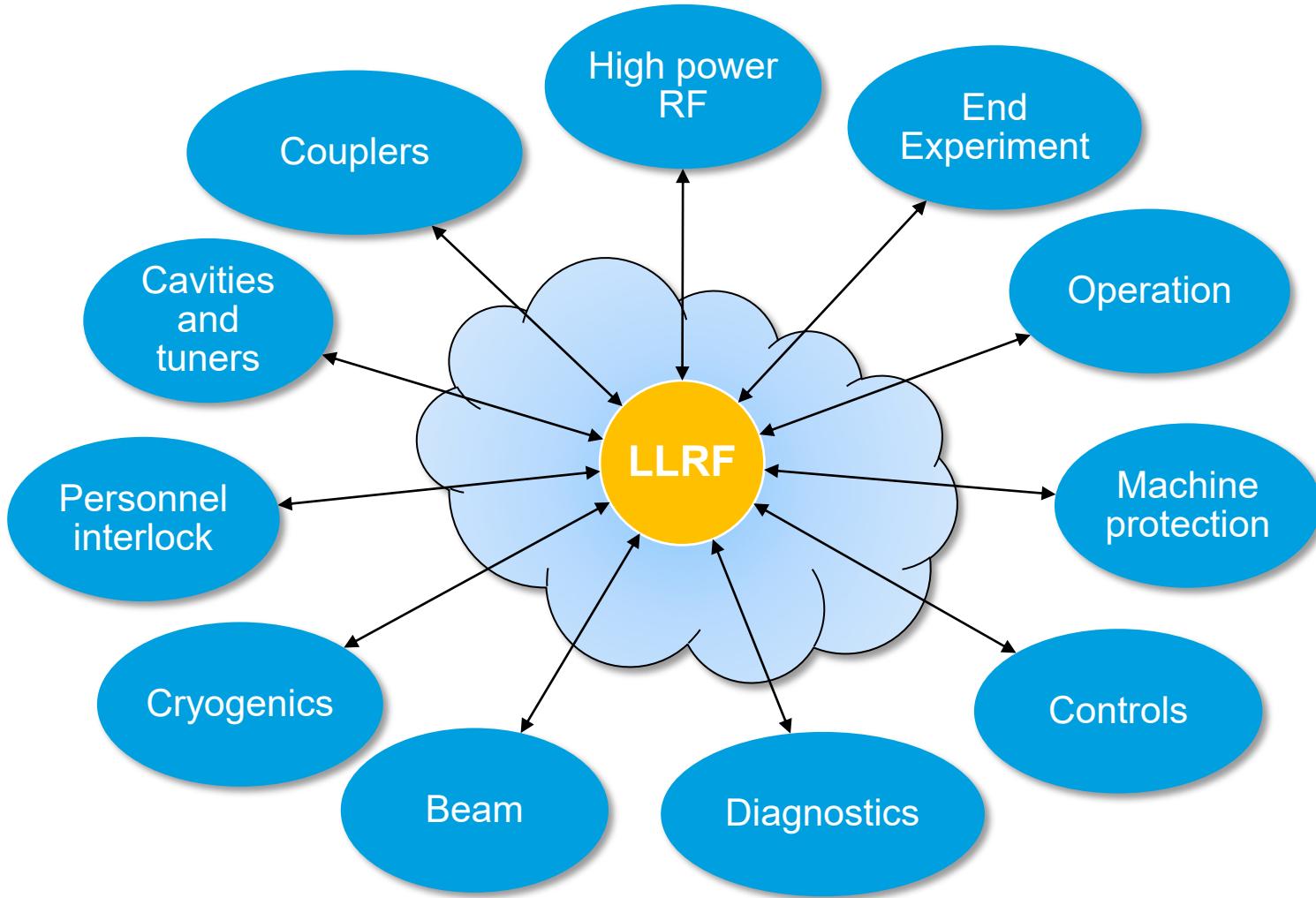
Main functionality

- It allows operator / other actors to **set** the **accelerating voltage and phase** for a given RF station
- It **Maintains** the desired amplitude and phase across the flat top at the **required stability**
- Provides **calibrated** (engineering units) **waveforms** for all cavities (Forward, Reflected, Probe)



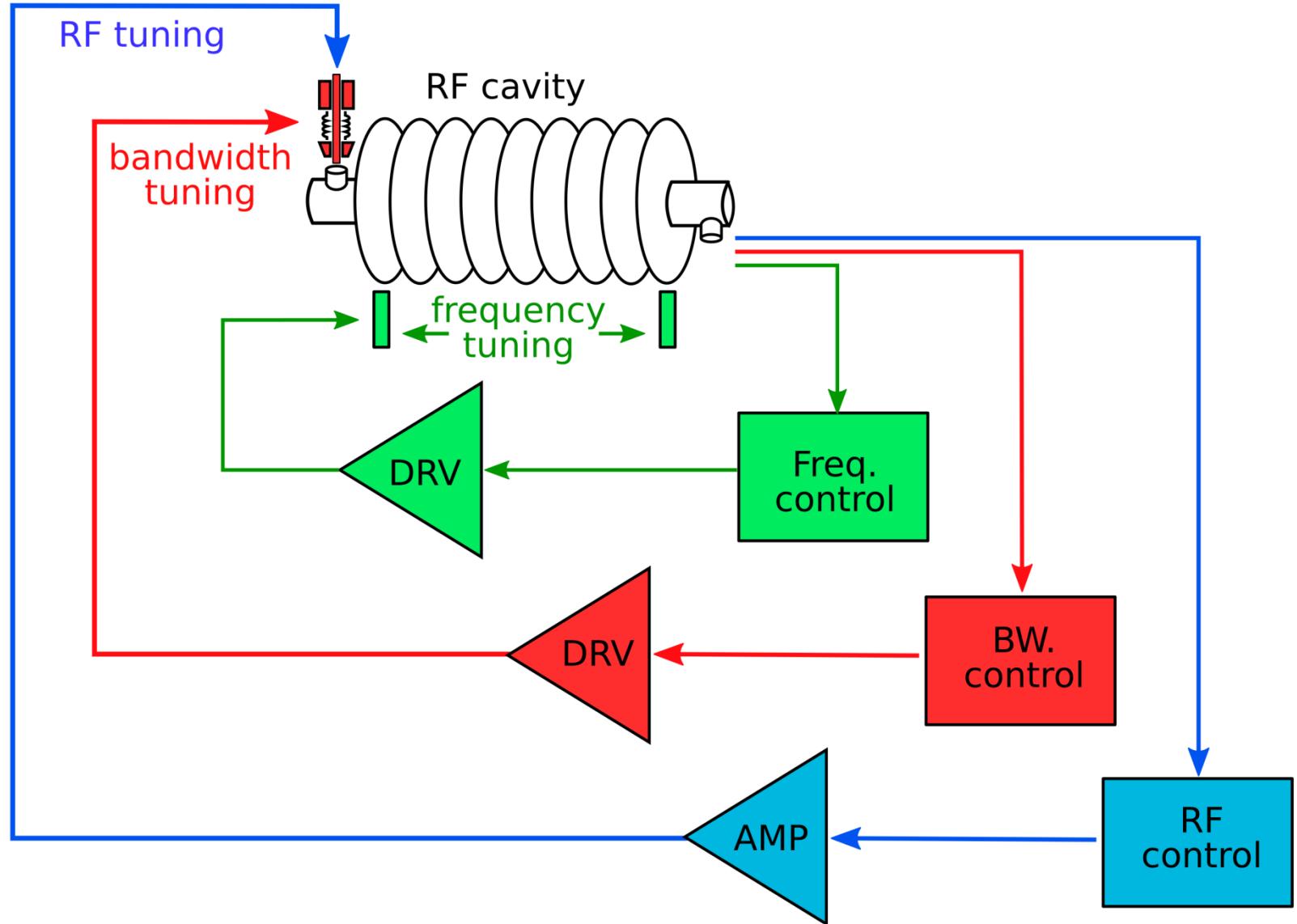
LLRF

Where does it start ? Where does it end ?



LLRF

LLRF controls



What I will try to cover in this talk

“RF controls crash course”

RF cavities

- Some basics about the cavity electrical and mechanical system and its modelling



RF power couplers

- Why, how ?
- Impact of changing the external coupling



Frequency tuners

- Fundamentals of tuners, slow, fast
- Some examples and practical considerations



LLRF system

- The basic blocks of LLRF
- Practical case simulation demo



RF operations

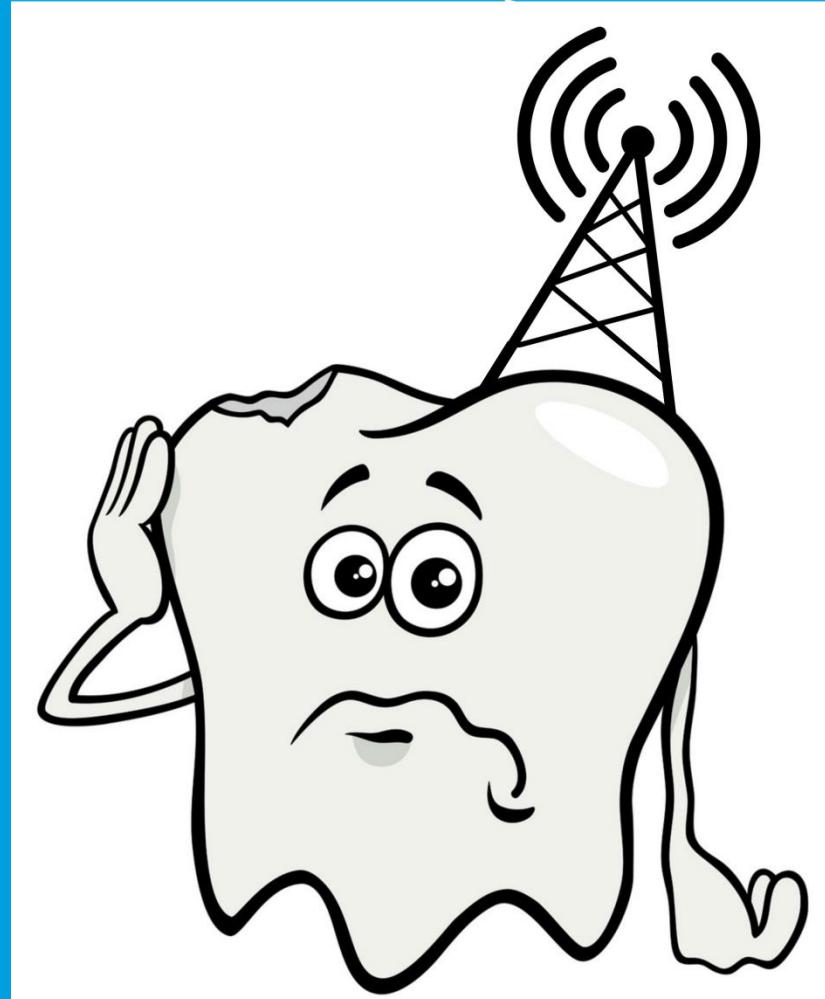
- Some examples of operation study or cases



Note:

Due to my background and work environment, most examples in this tutorial are inspired from [pulsed linac electron](#) machines.

RF cavities

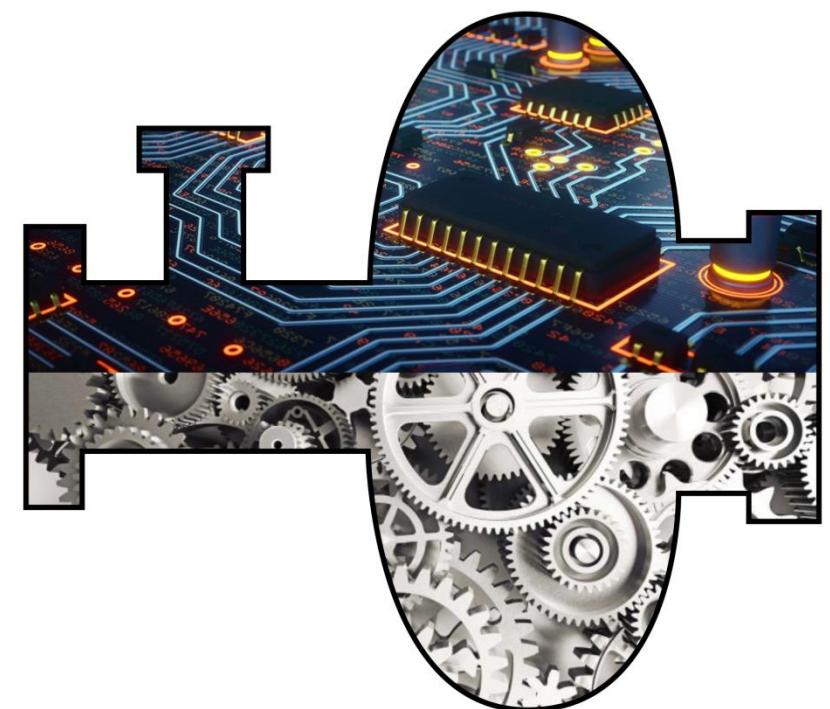


RF cavities

Copper (NRF) cavities



Niobium (SRF) cavities



Coupled **electrical mechanical** system

Cavity as an Electrical System

The standard RLC model (simplified)

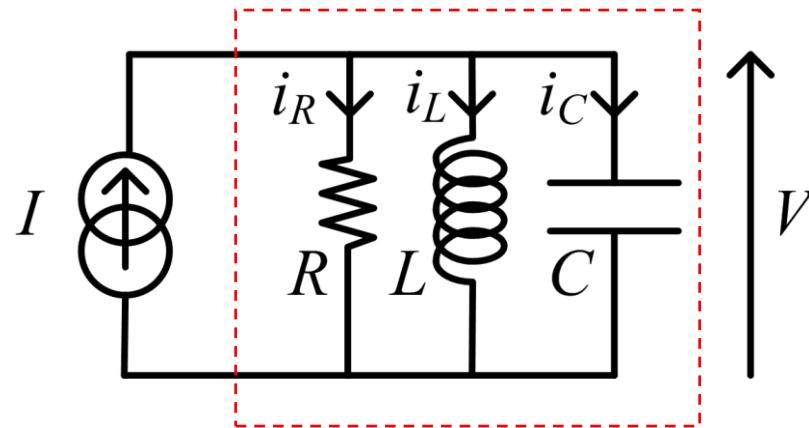
$$i_R + i_L + i_C = I \quad (1)$$

$$i_R = \frac{V}{R}$$

$$\frac{di_L}{dt} = \frac{1}{L}V$$

$$i_C = C \frac{dV}{dt}$$

(1)



(2)

$$(1) \rightarrow \frac{di_R}{dt} + \frac{di_L}{dt} + \frac{di_C}{dt} = \frac{dI}{dt}$$

$$(1) \& (2) \rightarrow \frac{1}{R} \frac{dV}{dt} + \frac{1}{L}V + C \frac{d^2V}{dt^2} = \frac{dI}{dt}$$

$$\rightarrow \frac{d^2V}{dt^2} + \frac{1}{RC} \times \frac{dV}{dt} + \frac{1}{LC} \times V = \frac{1}{C} \times \frac{dI}{dt}$$

$$\ddot{V}(t) + \frac{1}{RC} \dot{V}(t) + \frac{1}{LC} V(t) = \frac{1}{C} \dot{I}(t)$$

2nd order linear differential equation has a solution in the form of

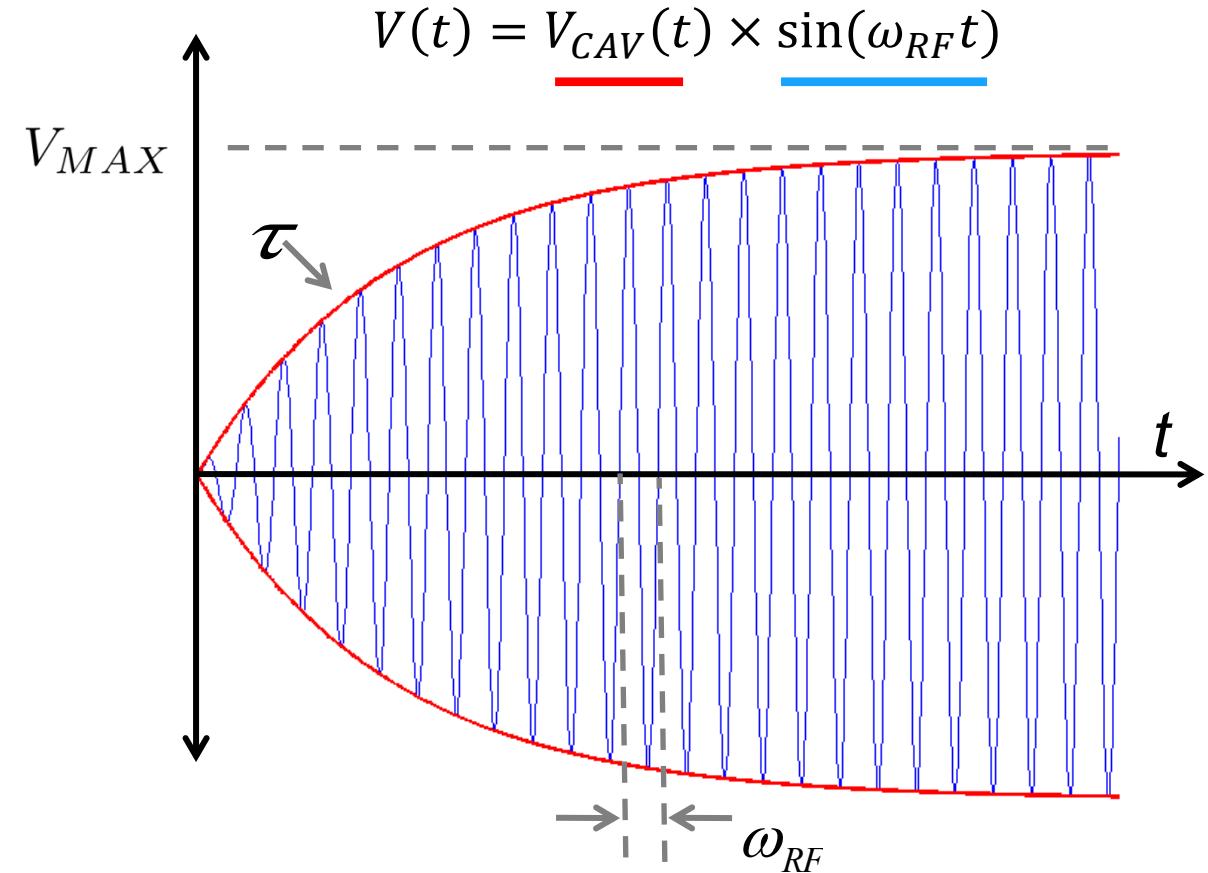
$$V(t) = V_{CAV}(t) \times \sin(\omega_{RF} t)$$

envelope

carrier

Cavity as an Electrical System

The envelope equation



V_{CAV} : envelope

$$V_{CAV} = V_{MAX}(1 - e^{-\frac{t}{\tau}})$$

τ is the **cavity time constant**

it depends on the **cavity bandwidth**

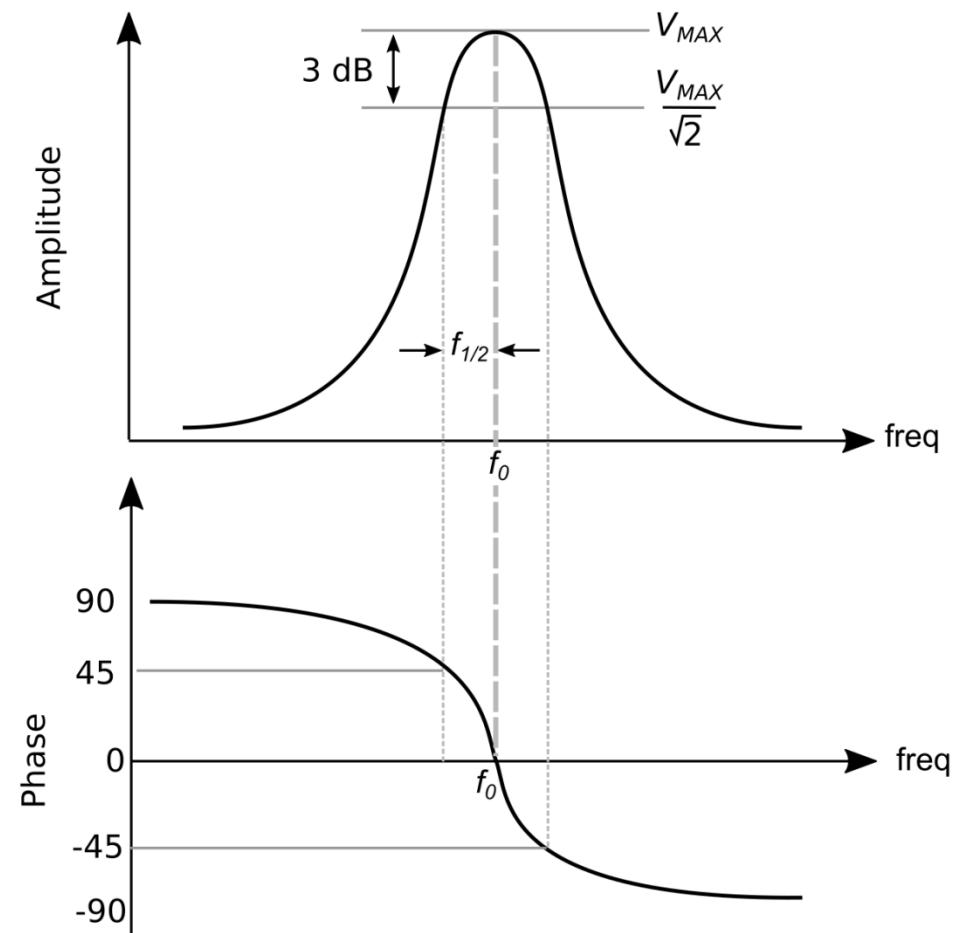
$$\tau = \frac{1}{2\pi f_{1/2}}$$

cavity half bandwidth
(i.e. half of the cavity bandwidth)

Cavity as an Electrical System

In the frequency domain

- Cavity behaves as a **band pass filter**
 - Center frequency f_0
 - Half bandwidth $f_{1/2}$

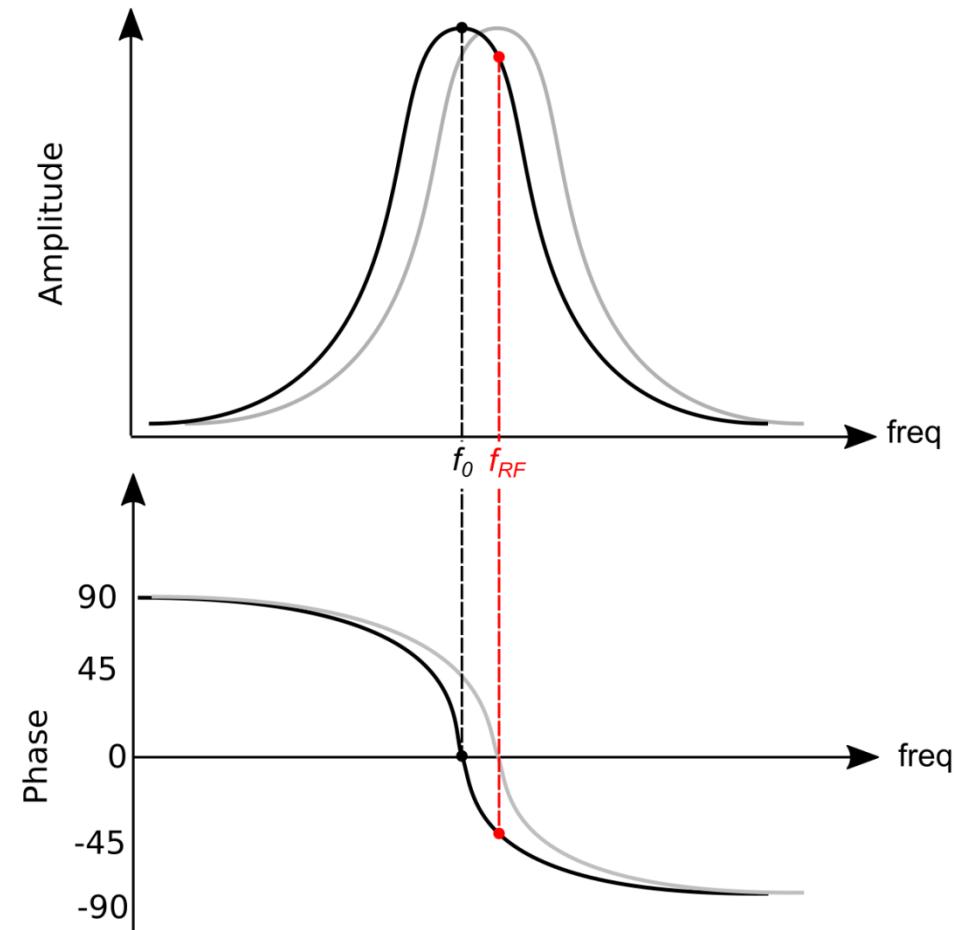


Cavity as an Electrical System

In frequency domain

- Cavity behaves as a **band pass filter**
 - Center frequency f_0
 - Half bandwidth $f_{1/2}$
- We can define **detuning** as the difference between the **cavity center frequency** (f_0 = resonance frequency) and the frequency of the **RF drive** (f_{RF})

$$\Delta\omega = \omega_0 - \omega_{RF} = 2\pi(f_0 - f_{RF})$$



Cavity as an Electrical System

Some figures of merit

Quality factor Q_0

$$Q_0 \equiv \frac{\text{Energy stored in cavity}}{\text{Energy dissipated in cavity walls per radian}} = \frac{\omega_0 U}{P_{diss}}$$

Shunt impedance R_{sh}

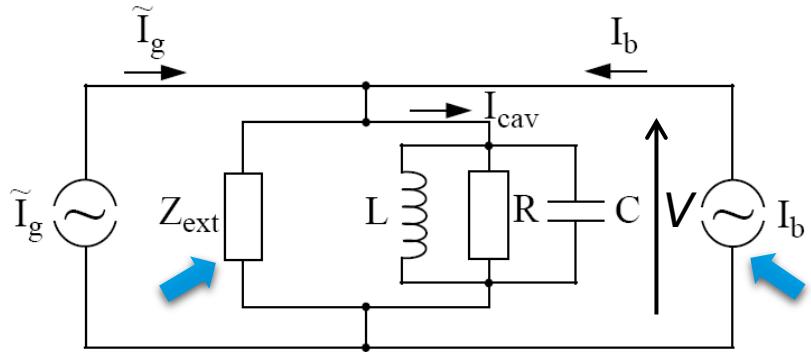
$$R_{sh} \equiv \frac{V_c^2}{P_{diss}} \quad \text{in } \Omega \quad V_c = \text{cavity "accelerating" voltage}$$

R_{sh}/Q_0 “R over Q”

$$\frac{R_{sh}}{Q_0} \quad \text{in } \Omega \quad \rightarrow \text{defined by cavity geometry only}$$

Cavity as an Electrical System

The standard RLC model (with coupling and beam)



- Same as before but now:
 - $I = I_g + I_b$
 - $R_L = R // Z_{ext}$

$$\ddot{V}(t) + \frac{1}{R_L C} \dot{V}(t) + \frac{1}{LC} V(t) = \frac{1}{C} \dot{I}(t)$$

$$\ddot{V}(t) + \frac{\omega_0}{Q_L} \dot{V}(t) + \omega_0^2 V(t) = \frac{\omega_0 R_L}{Q_L} \dot{I}(t)$$

using

$$\frac{1}{LC} = \omega_0^2$$

- Cavity characteristics

Coupling factor:

$$\beta = \frac{R}{Z_{ext}}$$

Loaded shunt impedance:

$$R_L = \left(\frac{1}{R} + \frac{1}{Z_{ext}} \right)^{-1} = \frac{R}{1 + \beta}$$

Loaded quality factor:

$$Q_L = \left(\frac{1}{Q_0} + \frac{1}{Q_{ext}} \right)^{-1} = \frac{Q_0}{1 + \beta}$$

Cavity as an Electrical System

State space representation

- Starting from the same differential equation

$$\ddot{\mathbf{V}}(t) + \frac{\omega_0}{Q_L} \dot{\mathbf{V}}(t) + \omega_0^2 \mathbf{V}(t) = \frac{\omega_0 R_L}{Q_L} \dot{\mathbf{I}}(t)$$

- Separate $V(t)$ and $I(t)$ into $\text{Re}\{\}$ and $\text{Im}\{\}$ parts

$$\mathbf{V}(t) = (V_r(t) + iV_i(t)) \cdot e^{i\omega t}$$

$$\mathbf{I}(t) = (I_r(t) + iI_i(t)) \cdot e^{i\omega t}$$

- Simplifications for SRF cavities, near resonance

$$\dot{V}_r + \omega_{1/2} V_r + \Delta\omega V_i = R_L \omega_{1/2} I_r$$

$$\dot{V}_i + \omega_{1/2} V_i - \Delta\omega V_r = R_L \omega_{1/2} I_i$$

- Introducing the following matrices:

$$\mathbf{A} = \begin{pmatrix} -\omega_{1/2} & -\Delta\omega \\ \Delta\omega & -\omega_{1/2} \end{pmatrix}$$

$$\mathbf{B} = \begin{pmatrix} R_L \omega_{1/2} & 0 \\ 0 & R_L \omega_{1/2} \end{pmatrix}$$

$$x = \begin{pmatrix} V_r \\ V_i \end{pmatrix}$$

$$u = \begin{pmatrix} I_r \\ I_i \end{pmatrix}$$

- The cavity equation can be expressed in the state space formalism

$$\dot{x}(t) = \mathbf{A} \cdot x(t) + \mathbf{B} \cdot u(t)$$

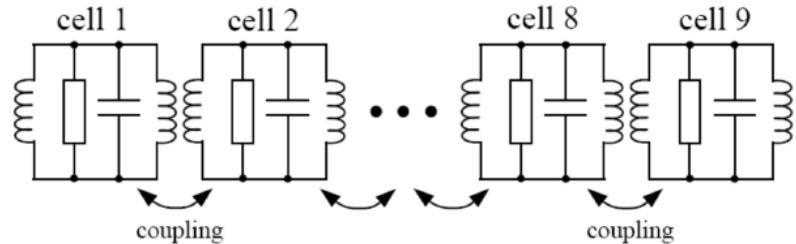
See: T. Schilcher's PhD Thesis for complete reference: "Vector Sum Control of Pulsed Accelerating Fields in Lorentz Force Detuned Superconducting Cavities", DESY, 1998

Cavity as an Electrical System

Sub-harmonic modes



http://tt.desy.de/desy_technologies/accelerators_magnets_und_cryogenic_technologies/weld_free_cavity/index_eng.html

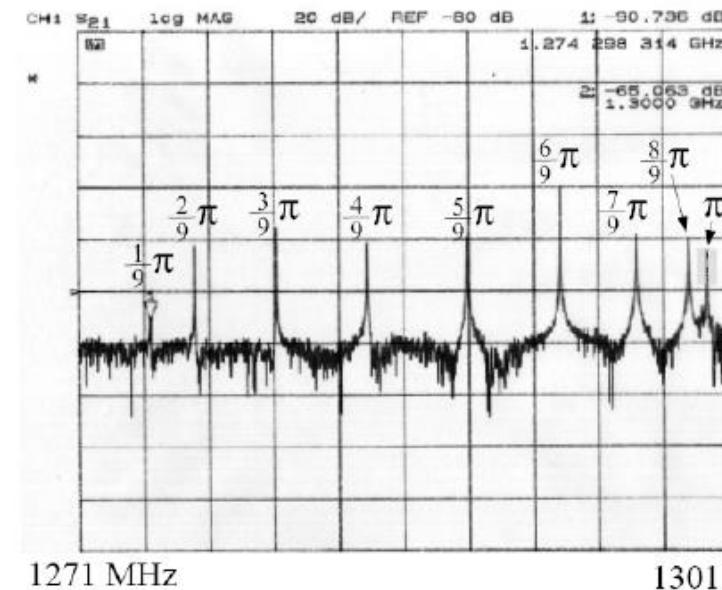


Modelled with 9 magnetically coupled resonators (RLC circuits)

- Pi mode is used for acceleration (TM010 mode)
- $8\pi/9$ mode only 800 kHz separated from operating frequency → may influence accelerating field stability

Parameters for SRF cavity

Operating frequency: 1.3 GHz
Length: 1.036 m
Aperture diameter: 70 mm
Cell to cell coupling: $\approx 2\%$
Quality factor Q_0 : $\approx 10^{10}$
 $r/Q := r_{sh}/Q_0$: 1036Ω

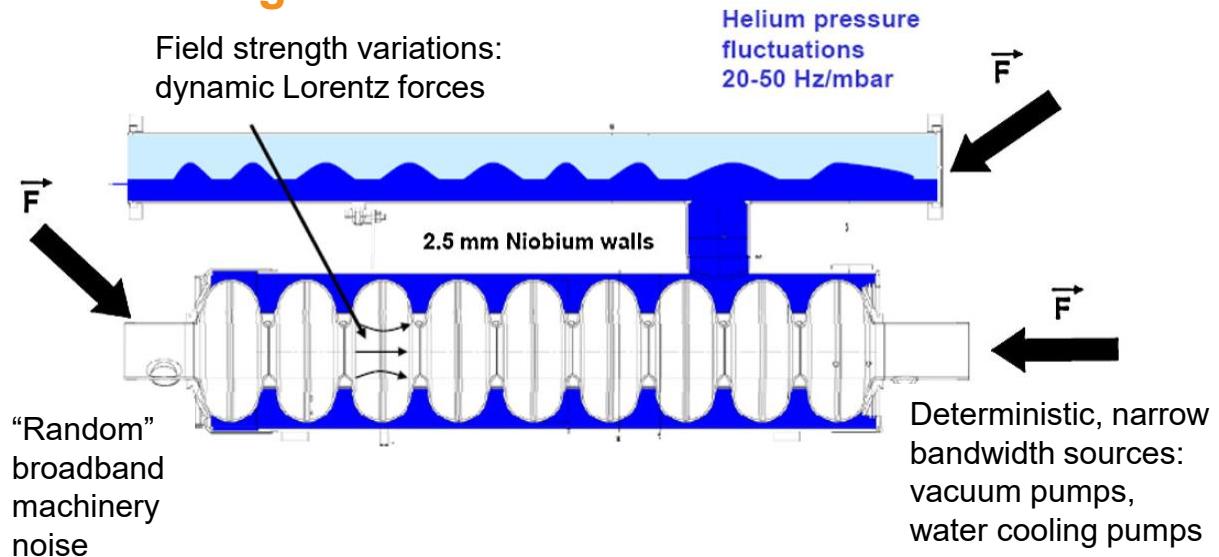


Courtesy: S. Pfeiffer

Cavity as a Mechanical System

Cavity mechanical modes, microphonics and sources of detuning

- Detuning can come from
 - Random external sources (external)
 - Deterministic external sources (external)
 - Helium bath fluctuations (external)
 - Lorentz force deformations (internal)
- Detuning **mechanical modes** modelled by 2nd order differential equation
- Each mode m has its own **frequency** (ω_m), **quality factor** (Q_m) and **coupling** to the cavity field (K_m)



Source: A. Neumann, "Analysis and active compensation of microphonics in continuous wave narrow-bandwidth superconducting cavities", PRST-AB, 2010

$$\frac{d^2\Delta f_m}{dt^2} + \frac{\omega_m}{Q_m} \frac{d\Delta f_m}{dt} + \omega_m^2 \Delta f_m = -\omega_m^2 K_m \left(\frac{V_{cav}}{l_{cav}} \right)^2 + \omega_m^2 \mu_m$$

frequency

detuning contribution of mode m

Lorentz force detuning constant

quality factor

LFD

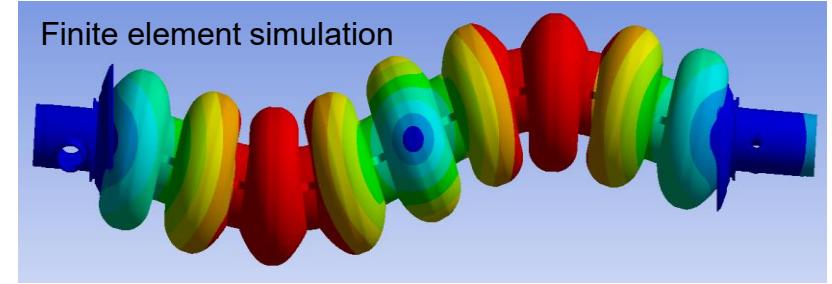
other mechanical driving forces (e.g. piezo)

Cavity as a Mechanical System

Cavity mechanical modes, microphonics and sources of detuning

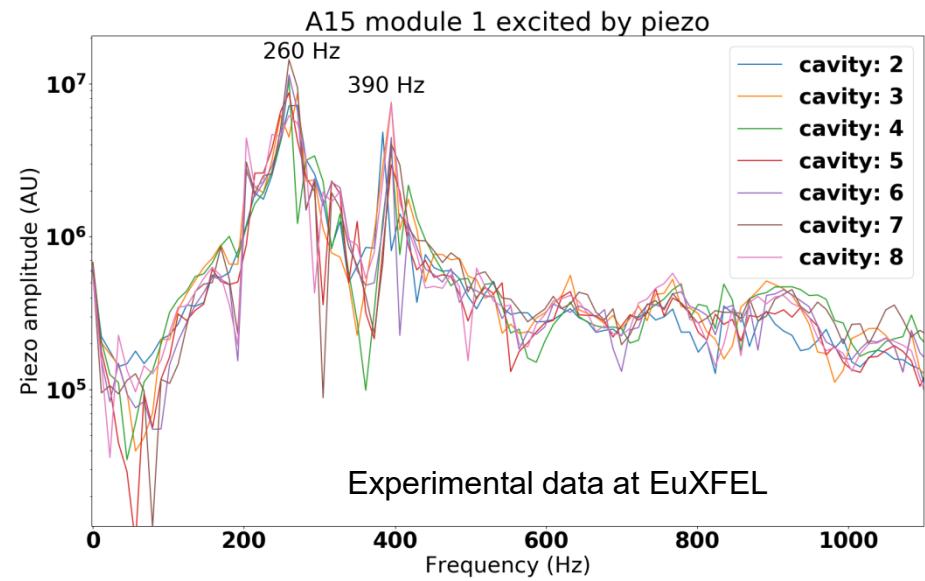
- Finite element analyses can accurately predict the **cavity eigenmodes**

Cavity first eigenmodes
59 – 60
150 – 152
202
266 – 270
391 – 398
402



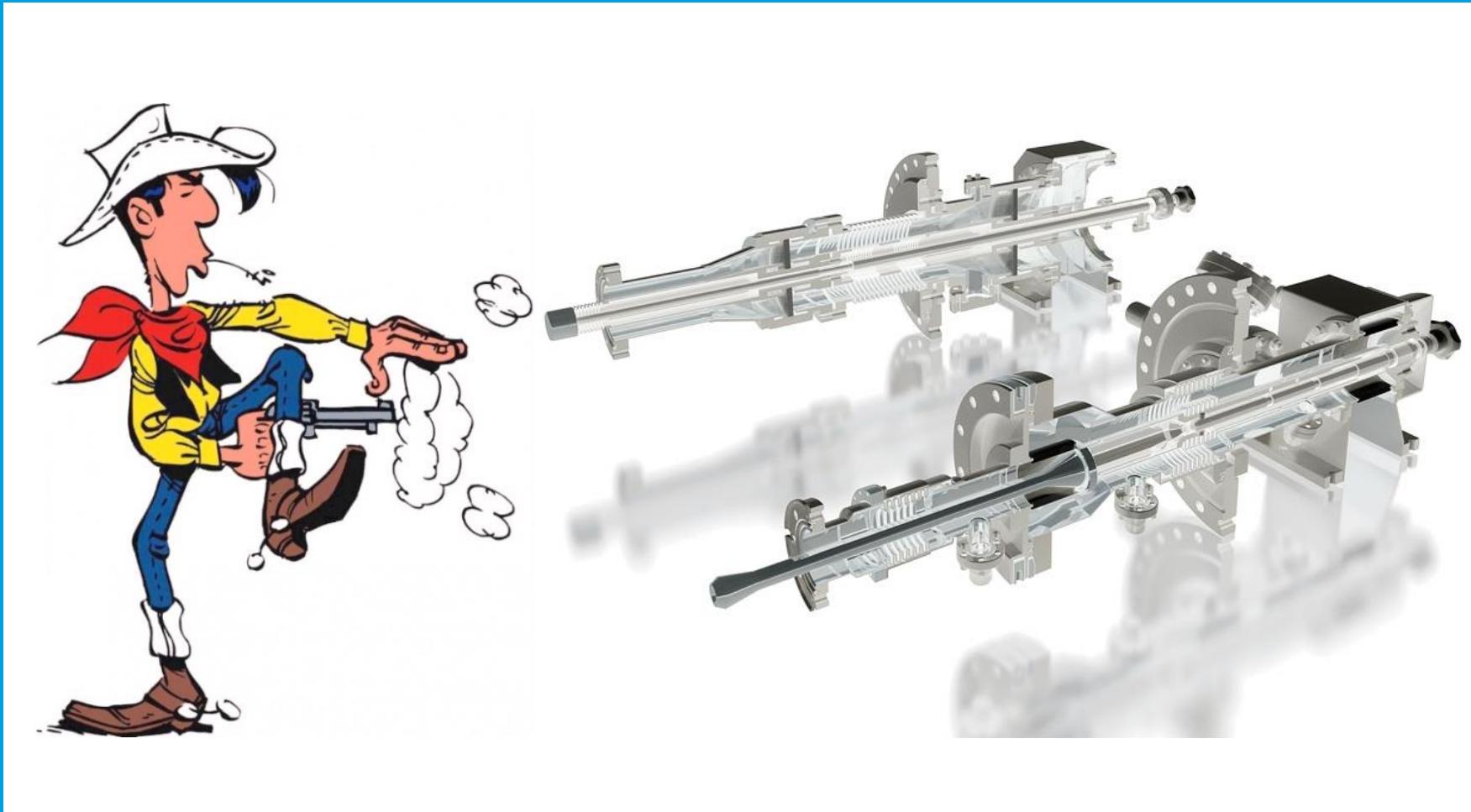
Source: S. Barbanotti, "Modal analysis of the XFEL 1.3 GHz cavity and cryomodule main components and comparison with measured data", SRF 2019

- The modes can be observed **experimentally**
 - Exciting a piezo with a frequency sweep
OR
 - Using the pulsed RF as external stimulus
 - Measuring the piezo sensor response



Source: A. Bellandi, "LLRF R&D towards CW operation of the European XFEL", Linac 2018

RF couplers



Coupler:
Lucky Luke:

Graphics Copyright: Rey.Hori / Mamoru Horiuchi. All rights reserved. ©Rey.Hori, 1995-2017
Graphics Copyright: Morris & Goscinny, 1946

Cavity coupling

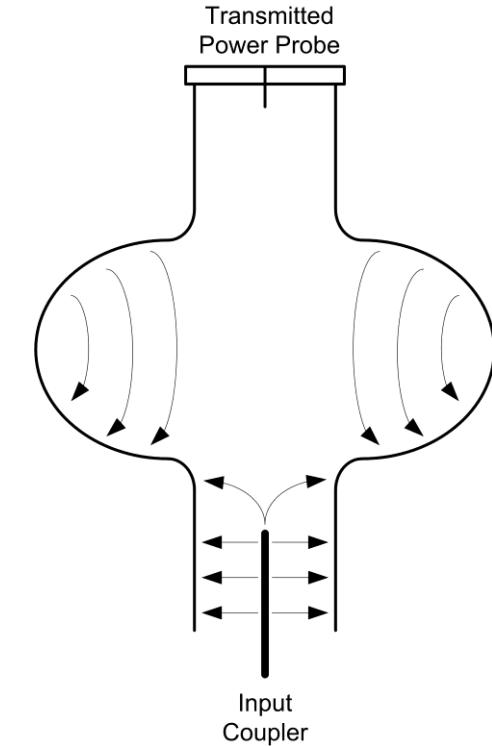
Coupling power in and out of a cavity

Input coupler

- An antenna carries power from an RF source to the cavity
- The strength of the **input coupler** is adjusted by changing the penetration of the center conductor

Output coupler (pick up)

- the **transmitted power probe** (fixed coupler) picks up power transmitted through the cavity



Measurable
“resulting” quality
factor (**loaded Q**)

$$Q_L = \left(\frac{1}{Q_{ext}} + \frac{1}{Q_0} \right)^{-1}$$

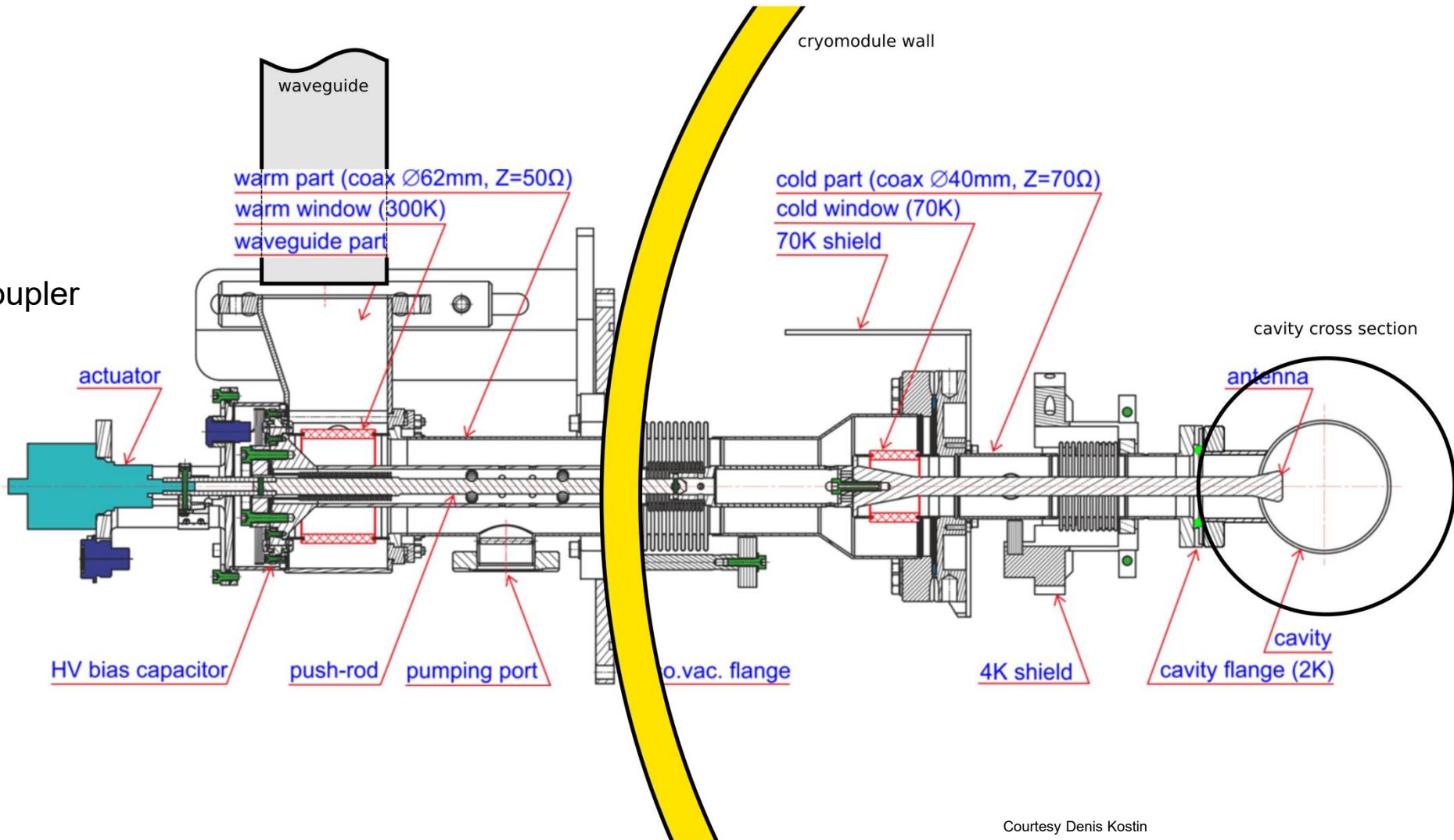
External quality
factor modified
by the coupler
antenna position

Cavity **unloaded**
quality factor

Cavity coupling

Example: fundamental input power coupler (FPC) for EuXFEL

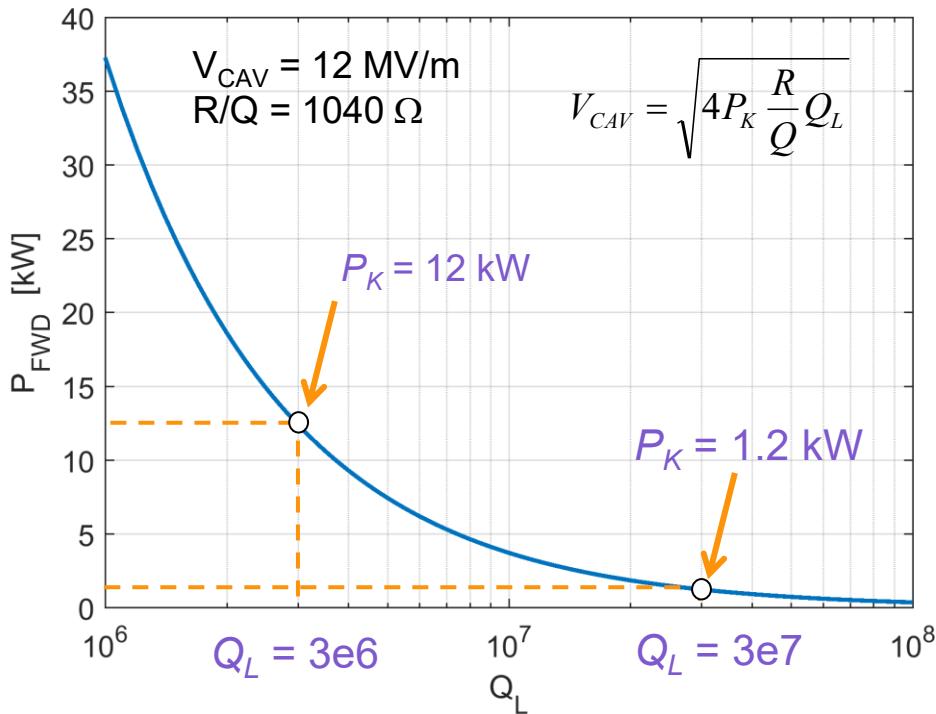
TTF –type coupler



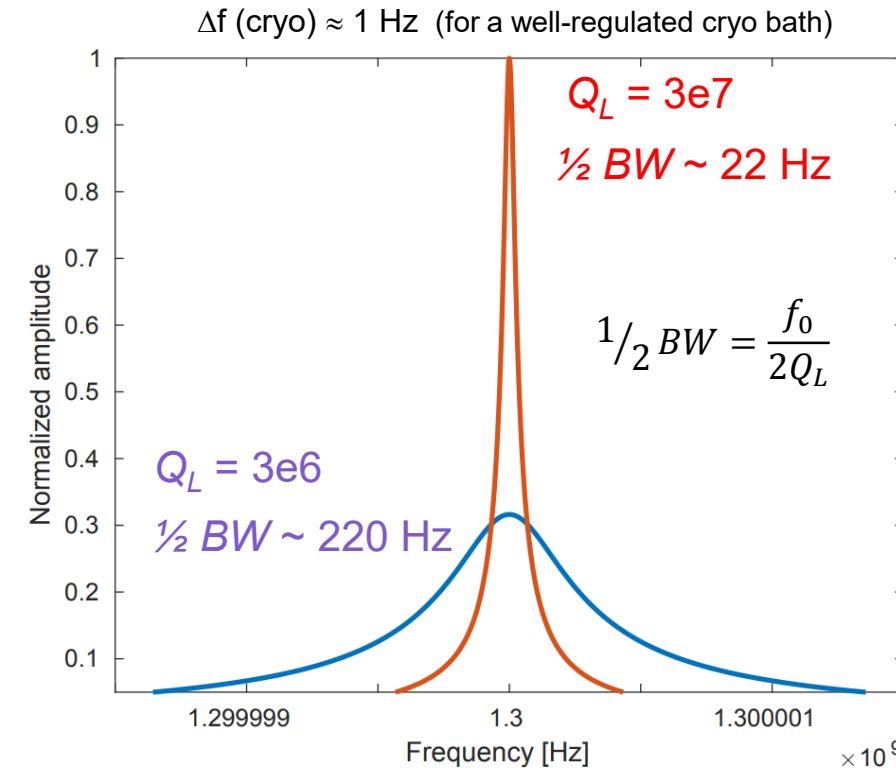
Courtesy Denis Kostin

Cavity coupling

Changing Q_{ext}



With higher Q_L , **less power** is required to achieve same gradient.

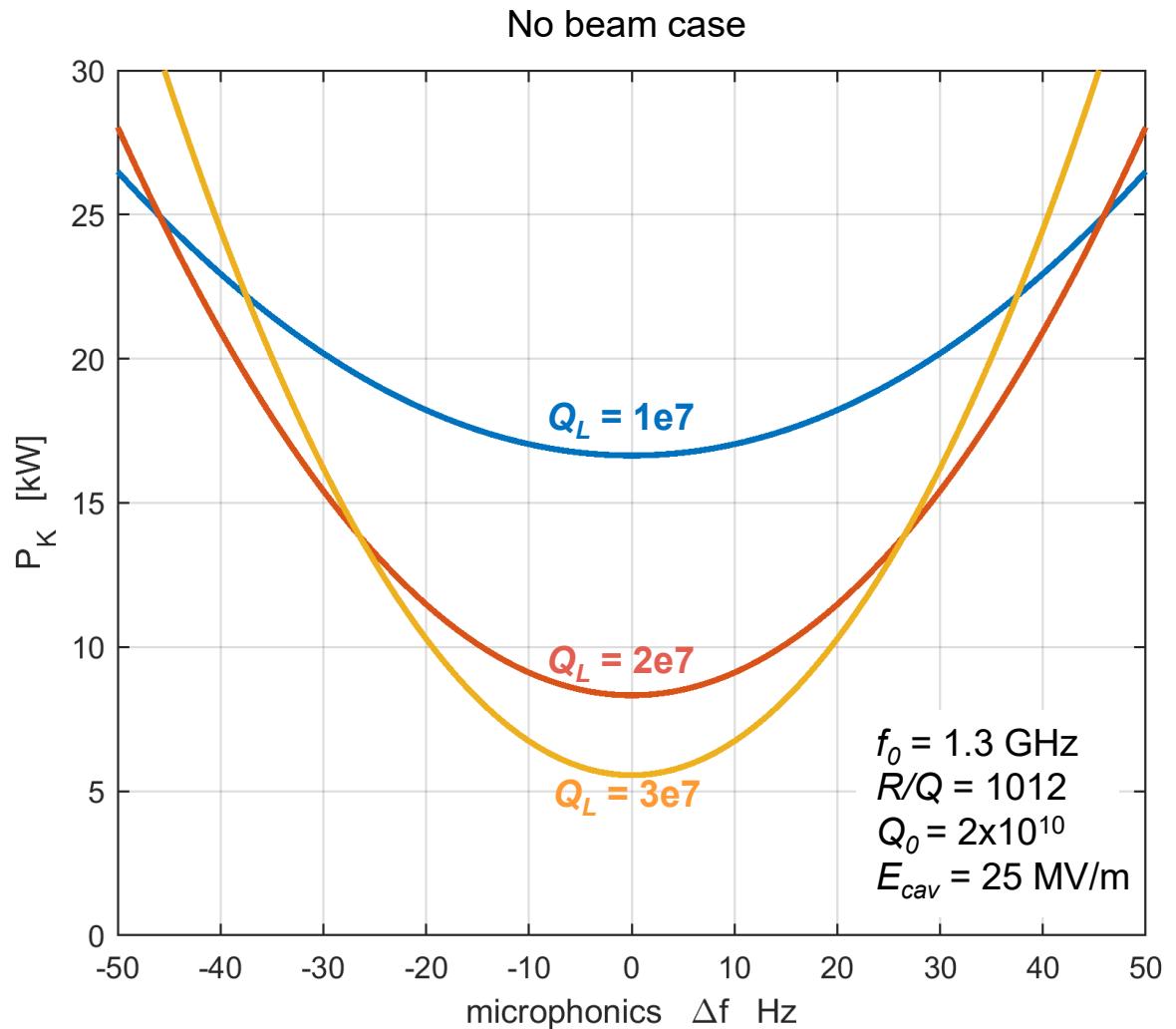


Increasing Q_L decreases the cavity bandwidth, making it **more sensitive to microphonics**.

Cavity coupling

Q_L , microphonics and power

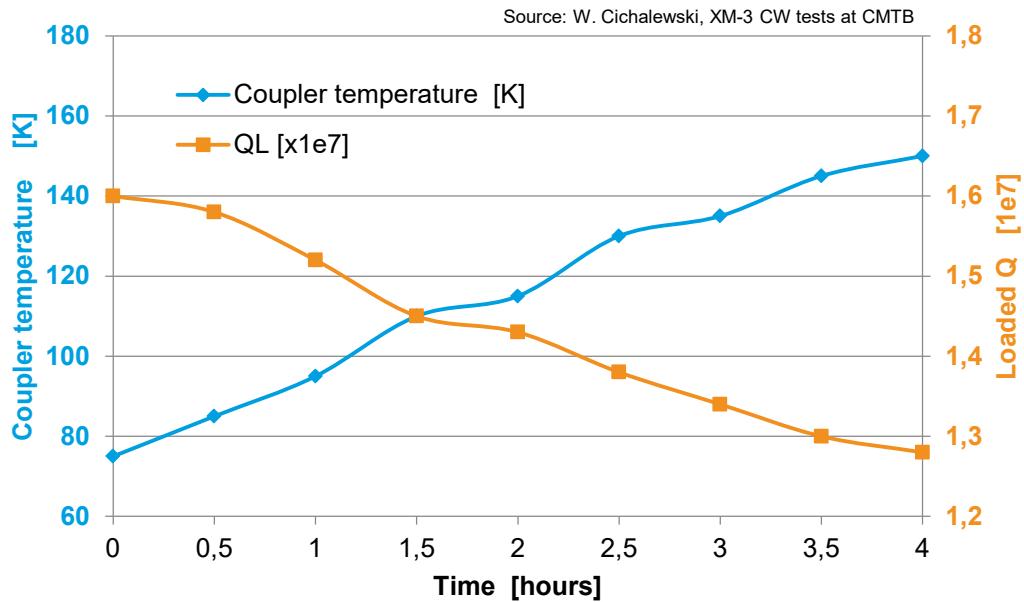
- Changing coupling has an impact on the **cavity sensitivity to microphonics**
- This translates into **RF power cost**, required for field regulation
- Example:
 - Plot shows the power required as a function of microphonic detuning for different values of Q_L



Cavity coupling

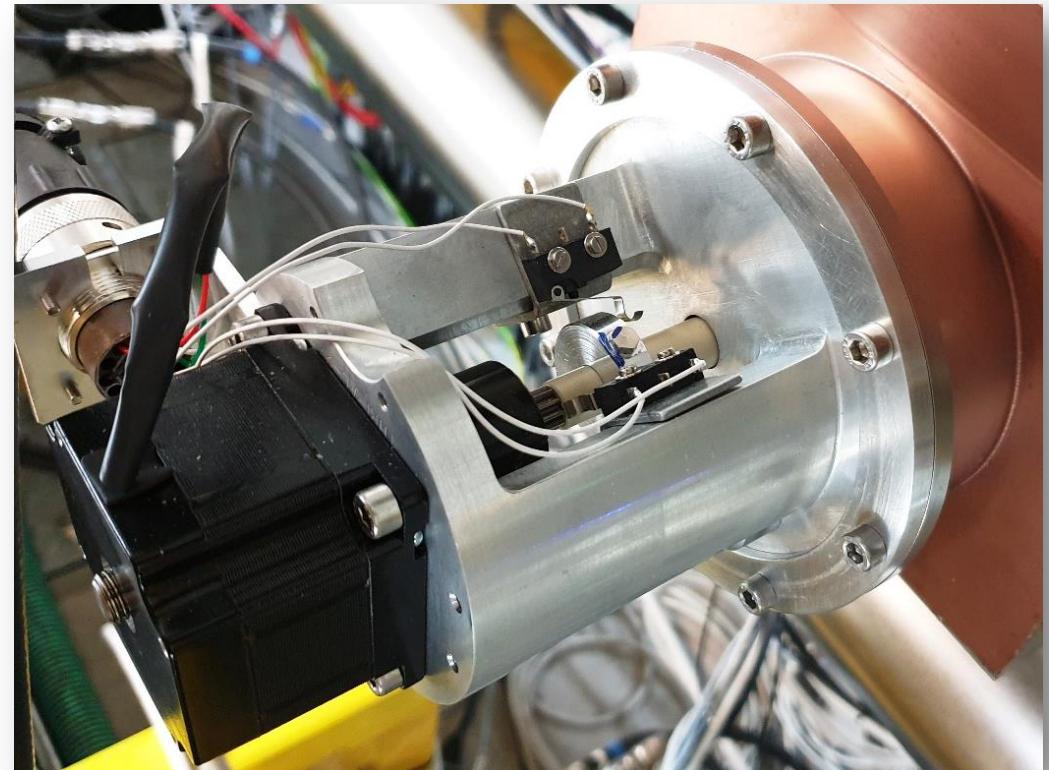
Adjustable coupler

- Coupler heating



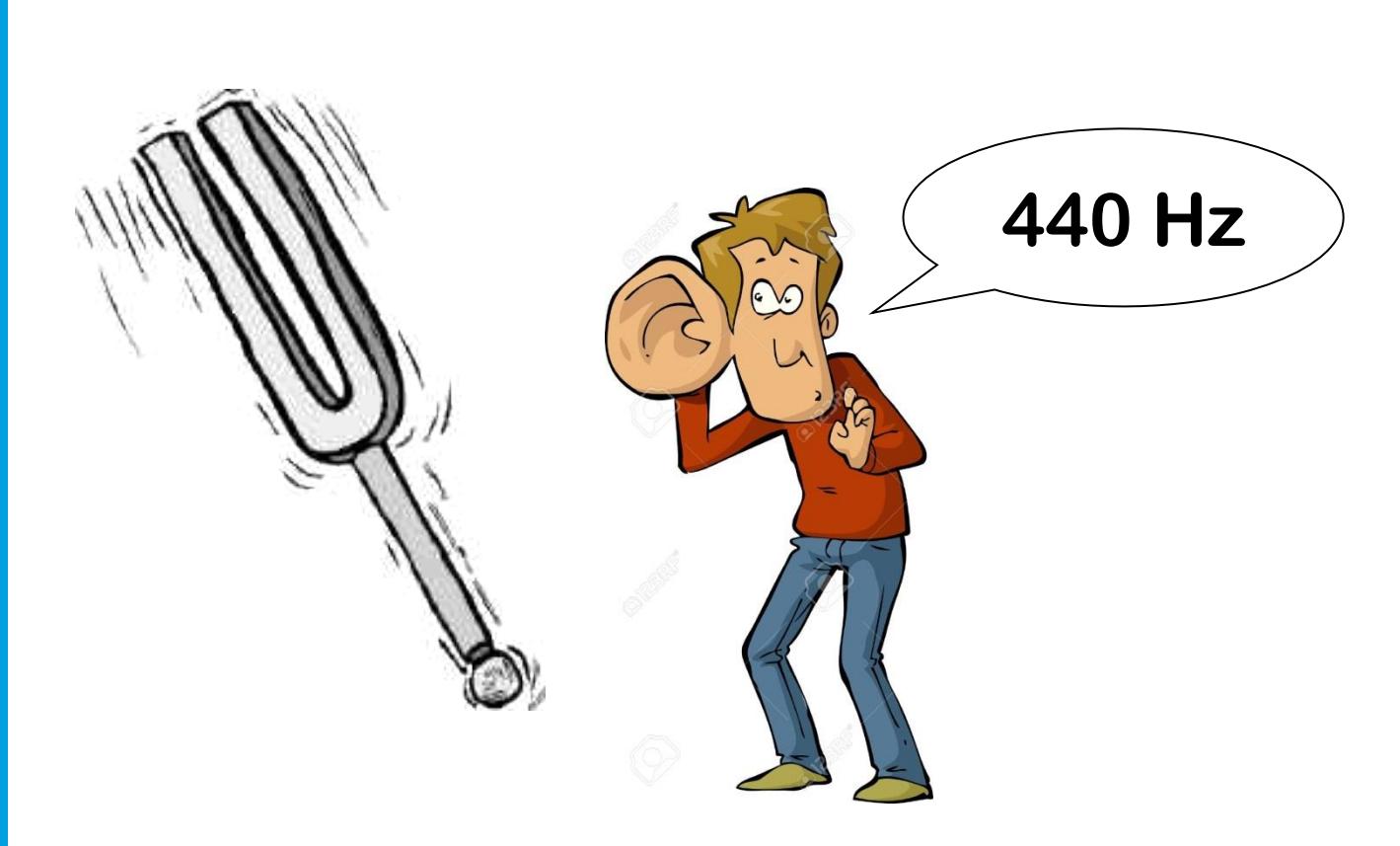
- Couplers heat up during operation, dilating the antenna, effectively lowering the loaded Q
- As a consequence, more power is needed to maintain the same gradient
- This further increases the coupler temperature, etc...

- Motor with end-switches



End switch on motorized coupler tuners on XM50.1 at DESY

Frequency tuners



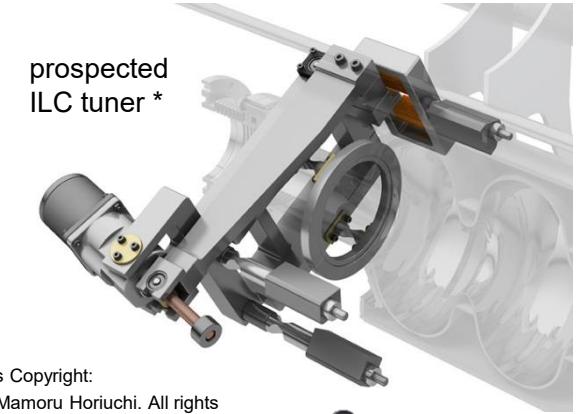
Tuning the cavity resonant frequency

- Resonance control is a fundamental part of the cavity RF system
 - Eg. copper cavities heat up and change frequency
 - Eg. synchrotrons often change frequency and cavities must track sometimes in the GHz/s rate
- SRF cavities have high loaded Qs and are thus more sensitivity to microphonics and Lorentz Force Detuning
- Tuning may be accomplished
 - by changing the cavity dimensions by temperature (i.e. water cooled normal conducting cavities)
 - by mechanical force, squeezing the cavity (i.e. motorized brackets)
 - electrically, using ferrite stubs (i.e. 3 stub tuners also affect Q_L and phase)

water cooled gun at PITZ



prospected
ILC tuner *



* Graphics Copyright:
Rey.Hori / Mamoru Horiuchi. All rights reserved. ©Rey.Hori, 1995-2017



Tuning the cavity resonant frequency

Slow and fast tuners



Slow tuners

Work over a large frequency range
Once on frequency they are only used to correct slow drifts



Note the mode dampers for the B and E strings



Fast tuners

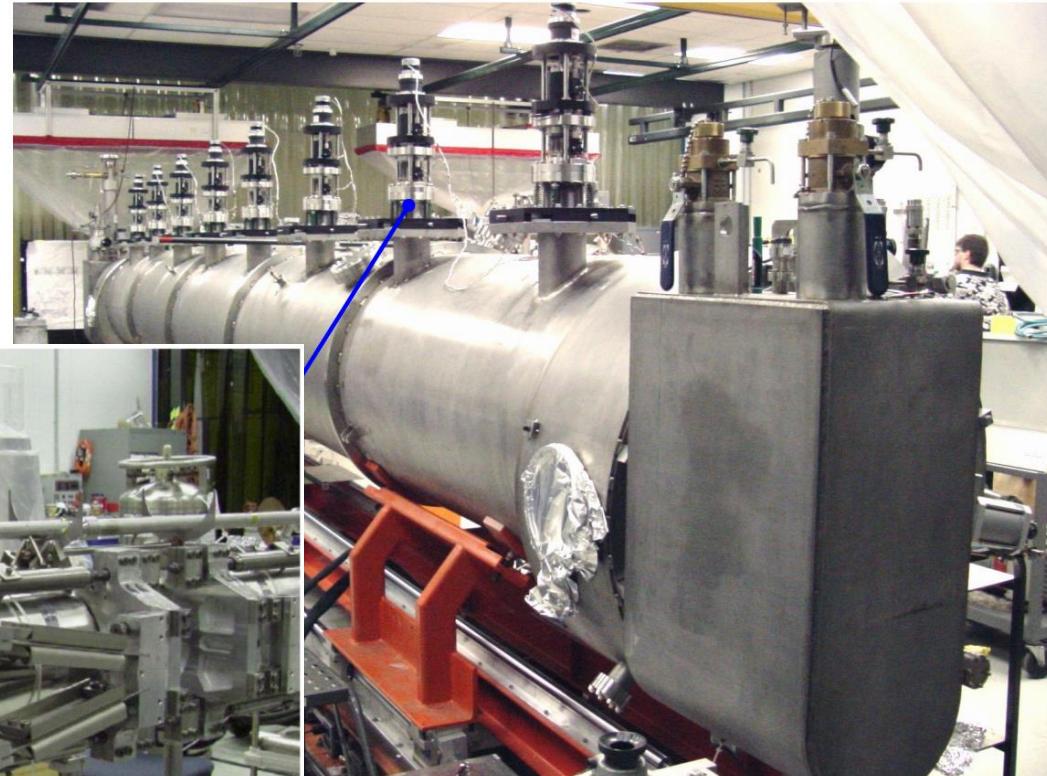
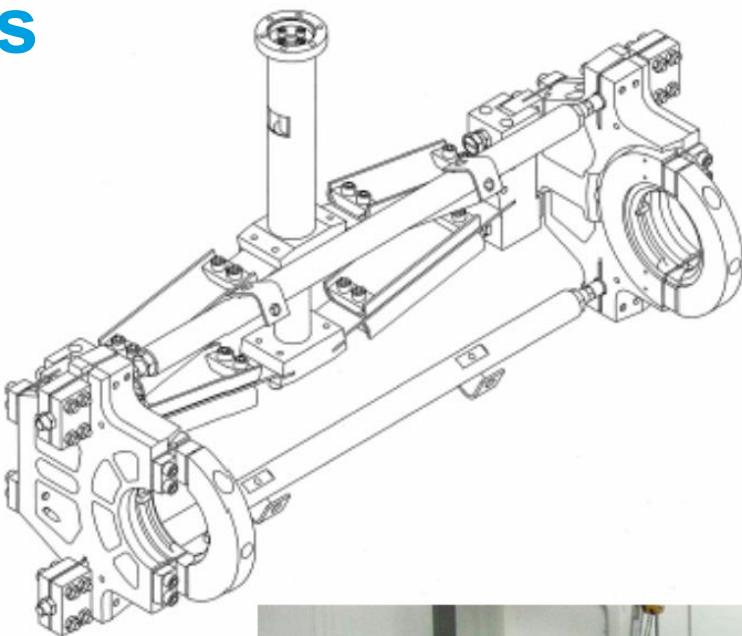
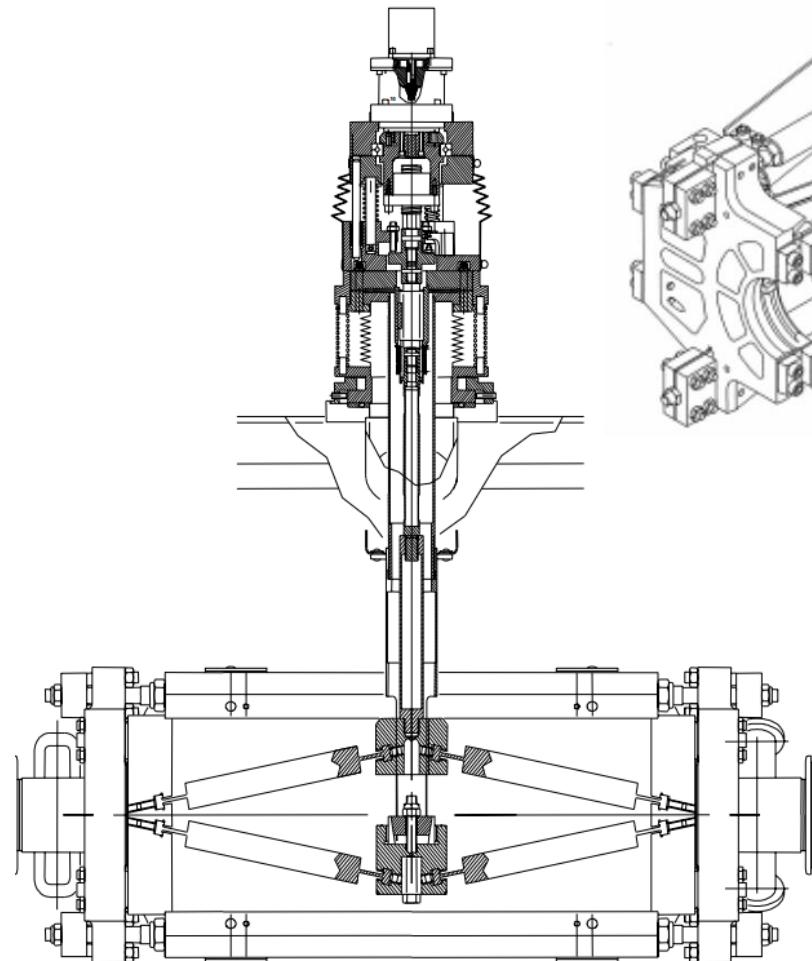
Used for fast dynamic resonance frequency modulation to compensate for microphonics or Lorentz Force Detuning



Stratocaster whammy bar

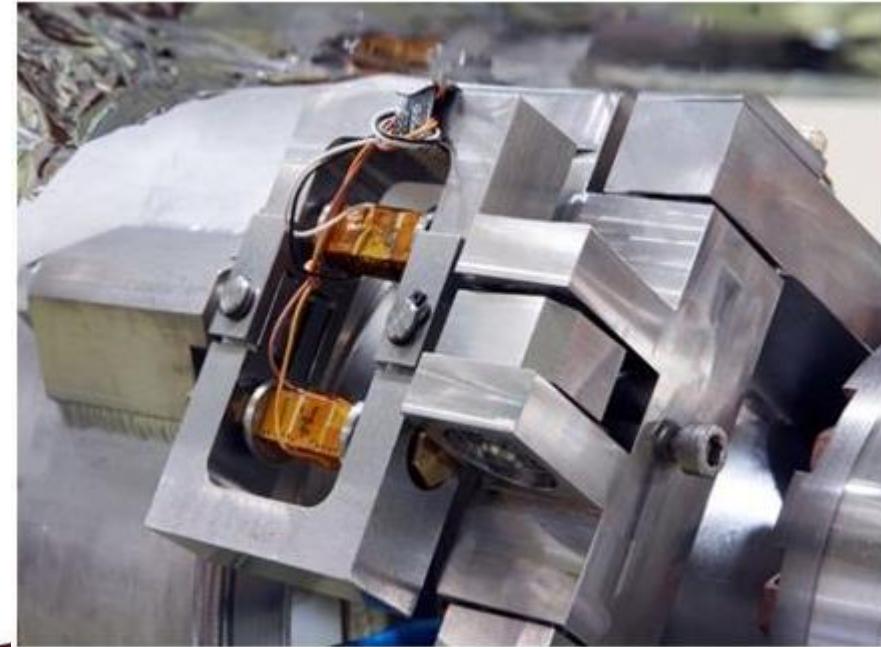
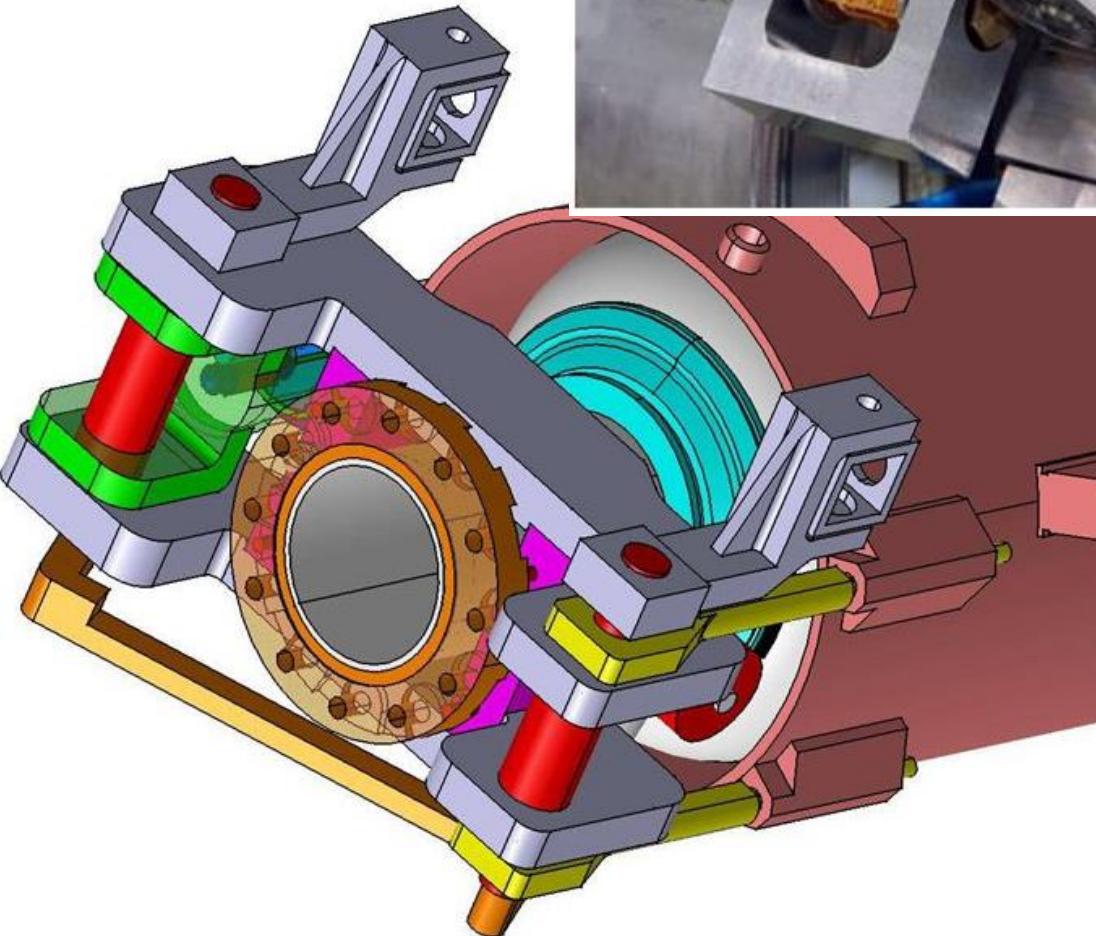
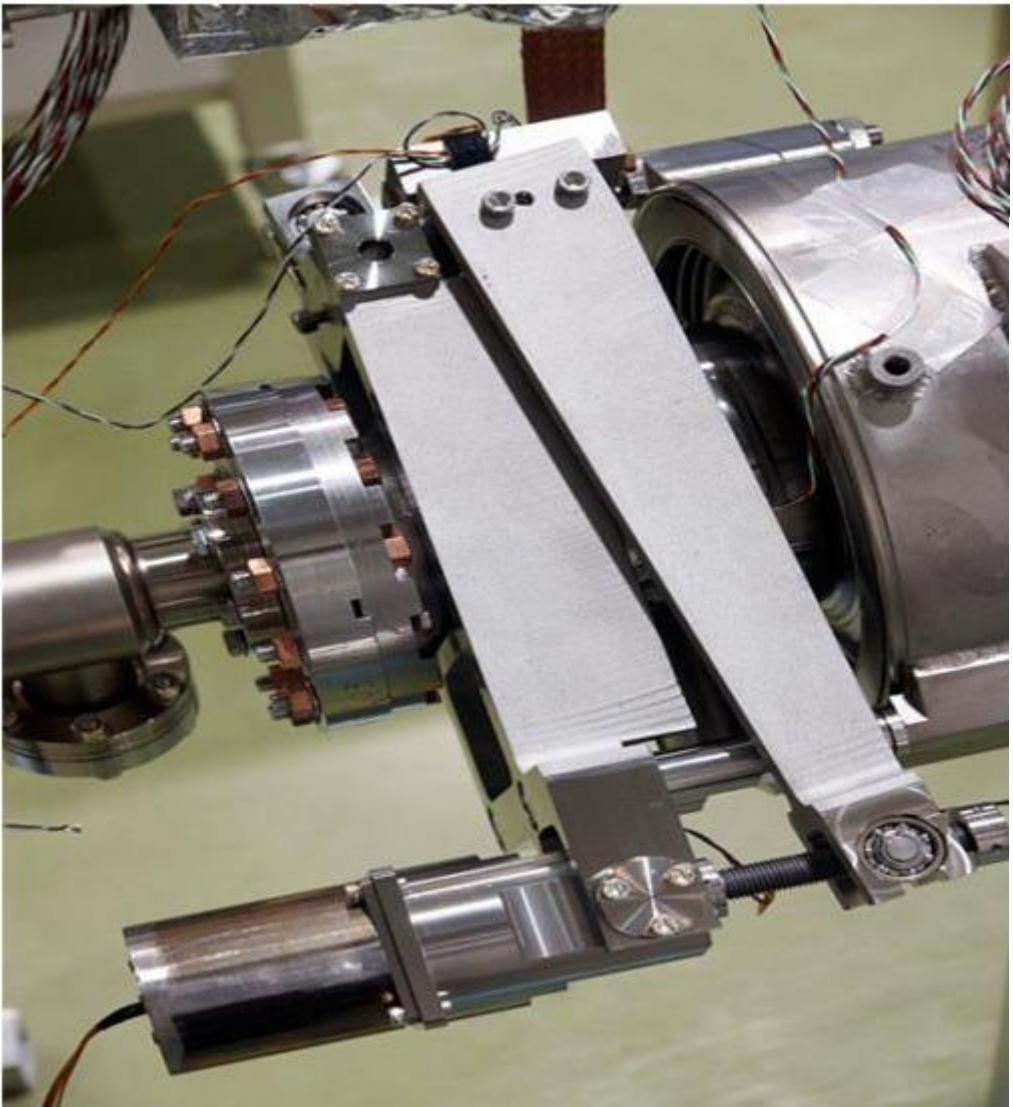
Tuner examples

CEBAF (JLAB)



Tuner examples

Saclay type (EuXFEL)

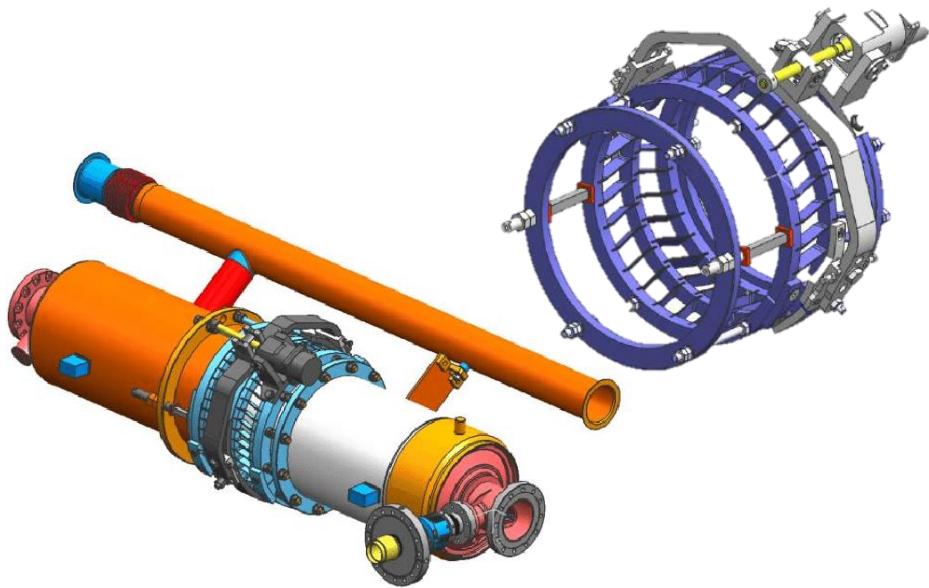


Tuner examples

(Slim) blade tuner

Blade tuner

First design



Slim blade tuner

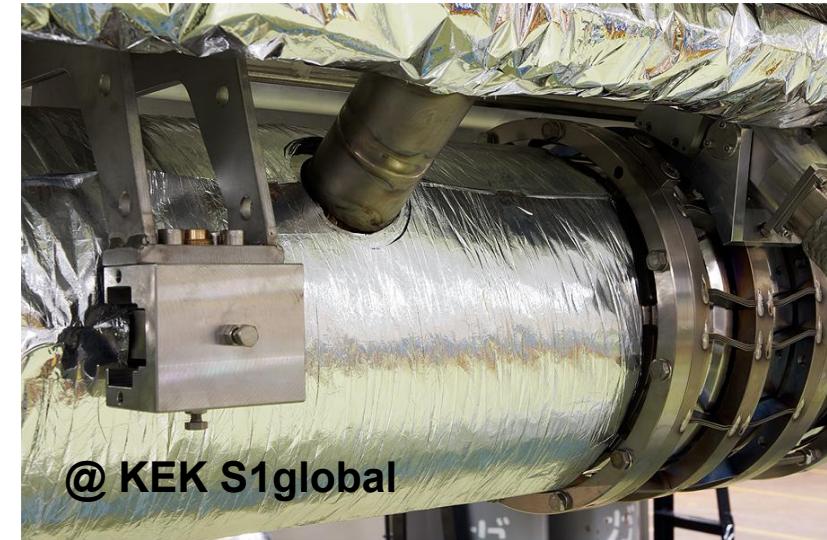
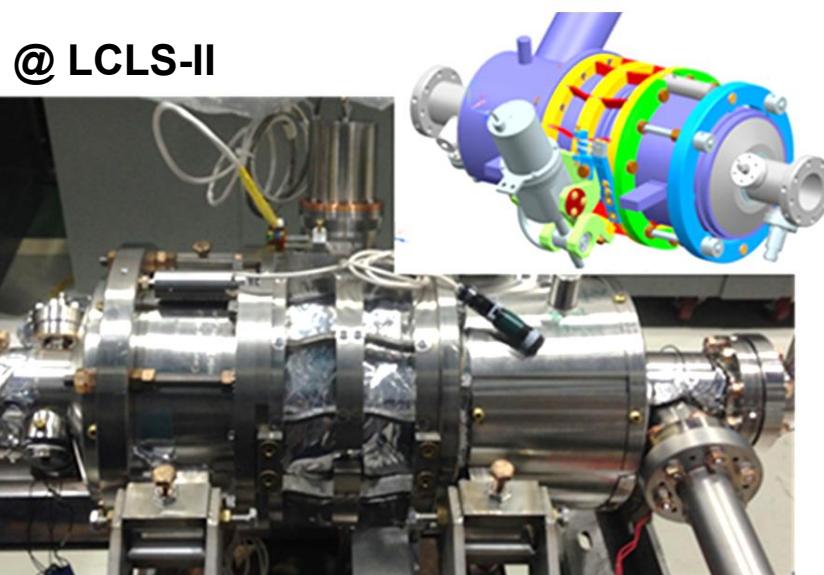
Improved design

More economical

Tested in S1Global (KEK)

Used in 3.9 GHz EuXFEL

Used in 3.9 GHz LCLS-II with piezo

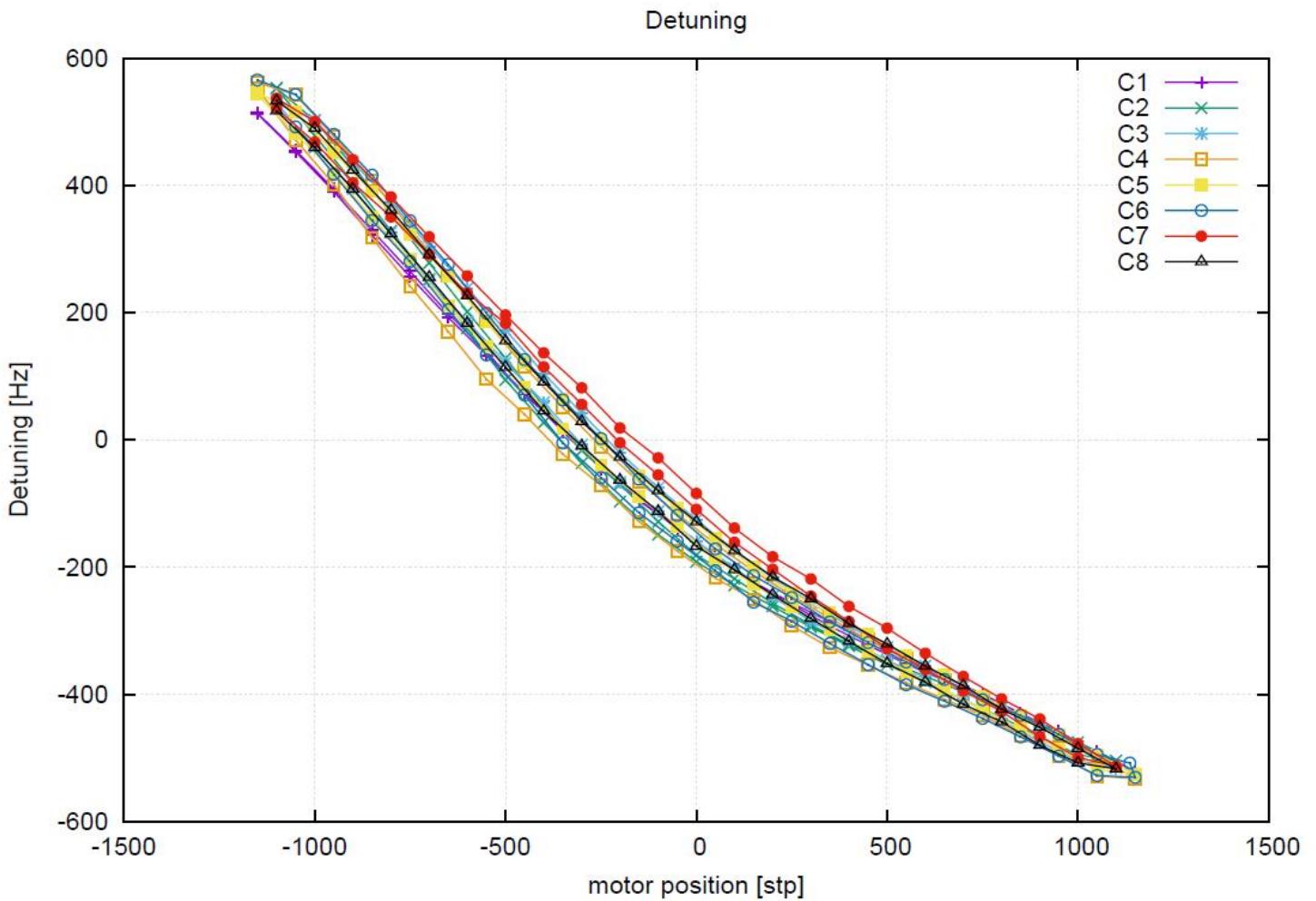


Tuners (continued)

Main characteristics

- Push / pull techniques
- Tuner pre-load
- **Motor sensitivity**
- Tuner backlash
- Motor lifetime

CAV	Detuning sensitivity [Hz/motor step]
C1	-0.45
C2	-0.49
C3	-0.49
C4	-0.48
C5	-0.48
C6	-0.48
C7	-0.49
C8	-0.49



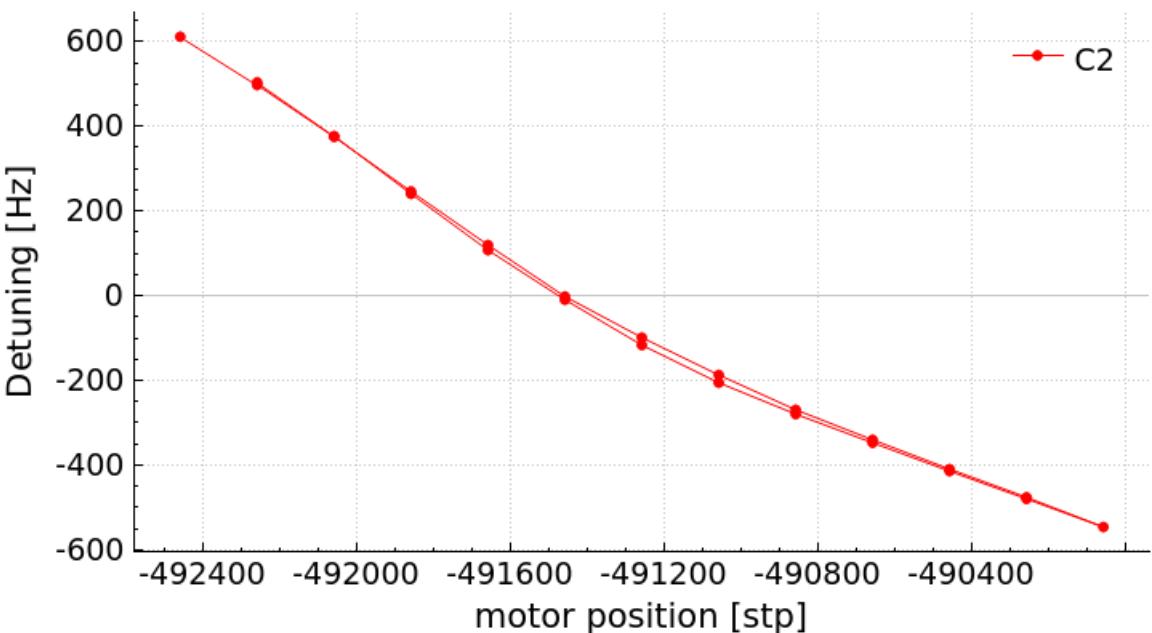
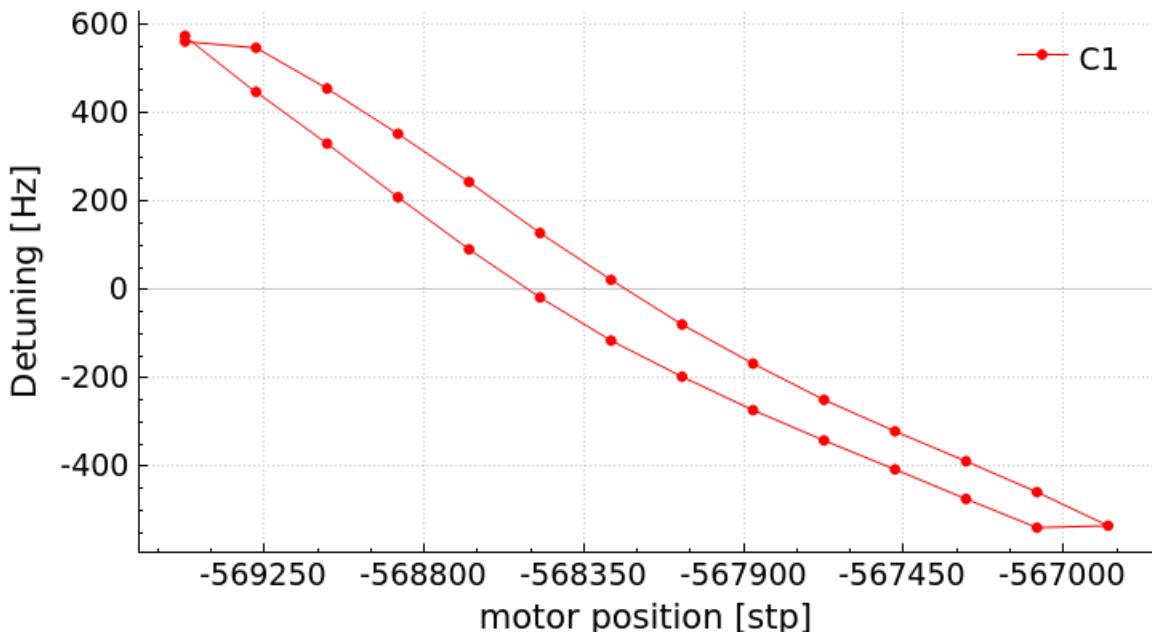
Example: EuXFEL XM36 tuner tests

Tuners (continued)

Main characteristics

- Push / pull techniques
- Tuner pre-load
- Motor sensitivity
- **Tuner backlash**
- Motor lifetime

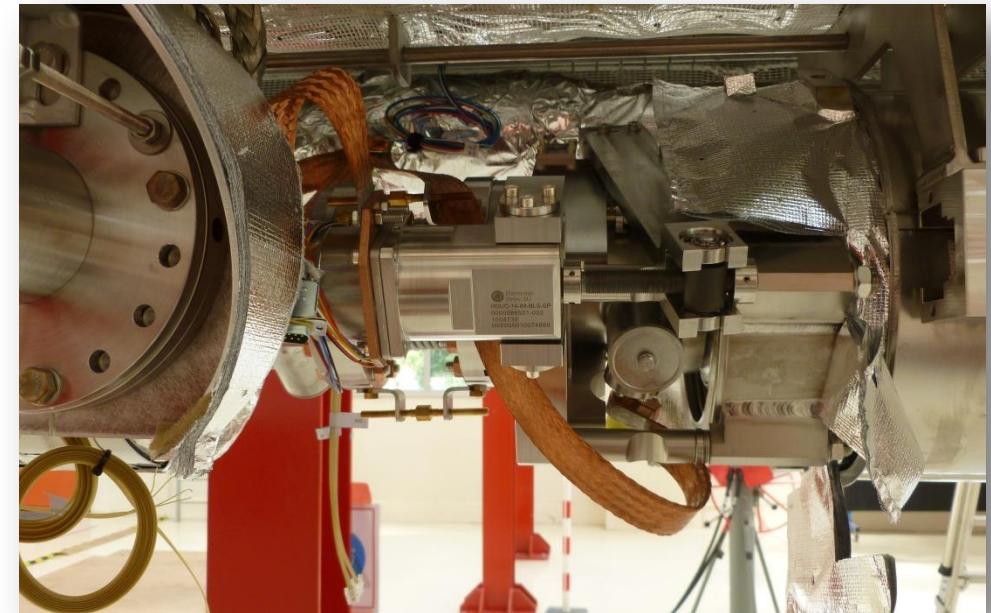
Example:
EuXFEL XM57 C1 and C2 tuner tests



Tuners (continued)

Main characteristics

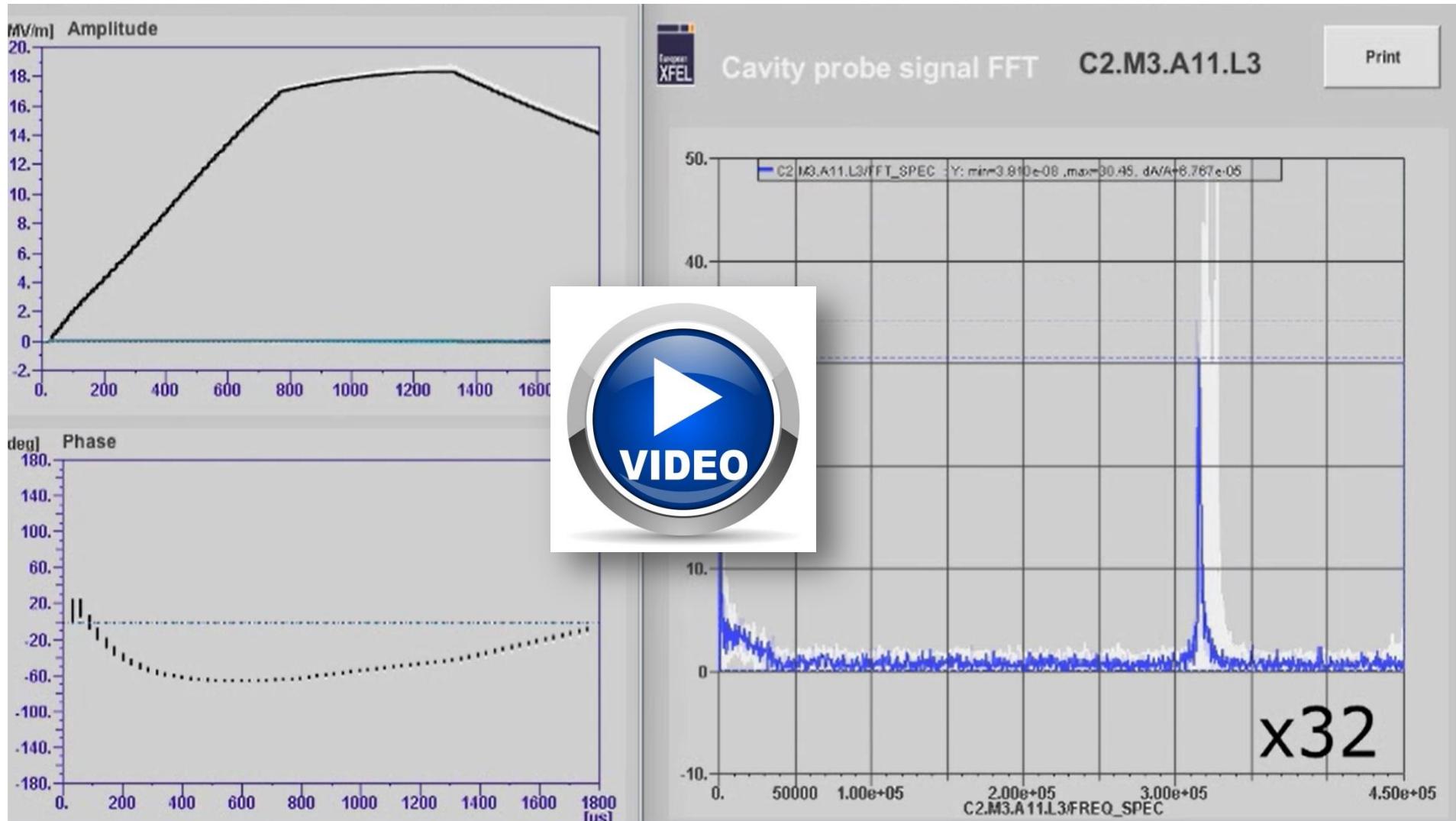
- Push / pull techniques
- Tuner pre-load
- Motor sensitivity
- Tuner backlash
- **Motor lifetime**
 - Limit motor drive (in time, in steps)
 - Avoid motor over heating (cooling with cooper braids)
- Example: Eu-XFEL tuners
 - Expected lifetime ~ 15-20 million steps
 - Moving from parking position < 1 million steps (typ. 400-600 k steps)
 - → maximum 15-20 thermal cycles (> 20 years)
 - → 1 or 2 already used during cryomodule tests



Cavity tuning

Using “slow tuner”

- Tuning from parking position



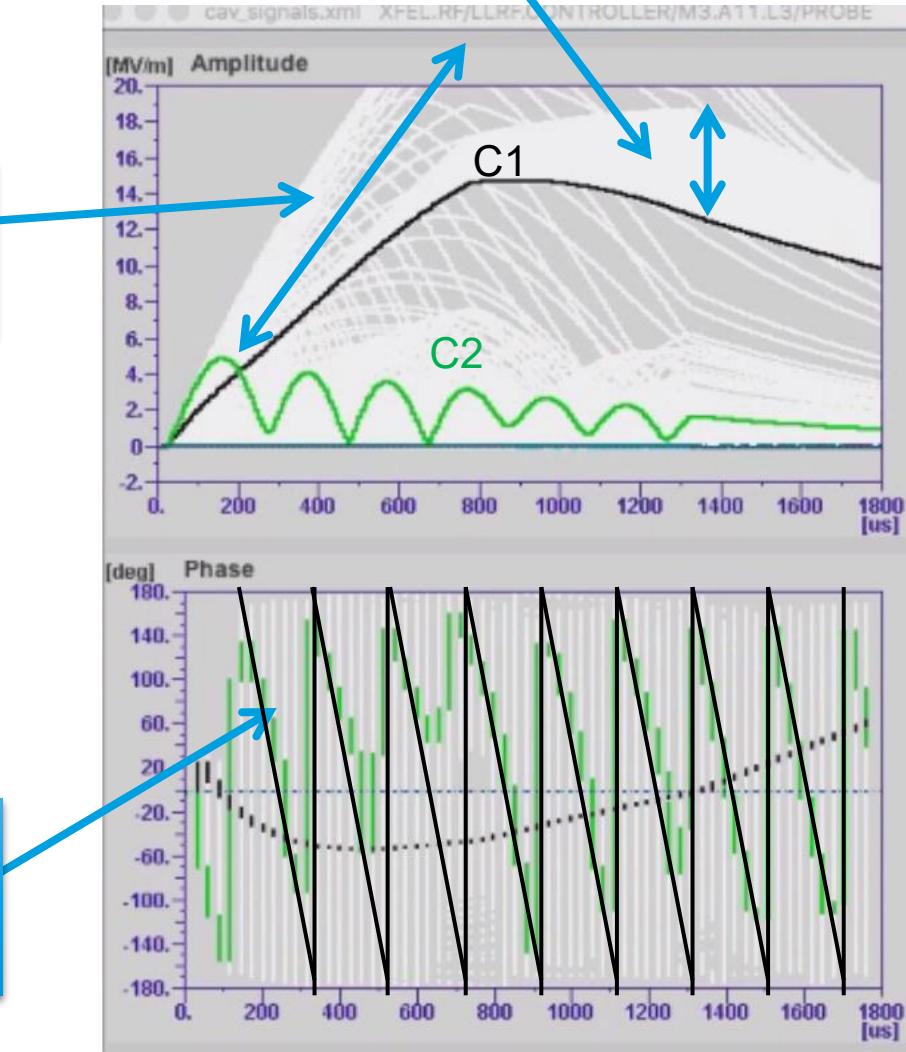
video time = (32x then 8x) actual time

Cavity tuning

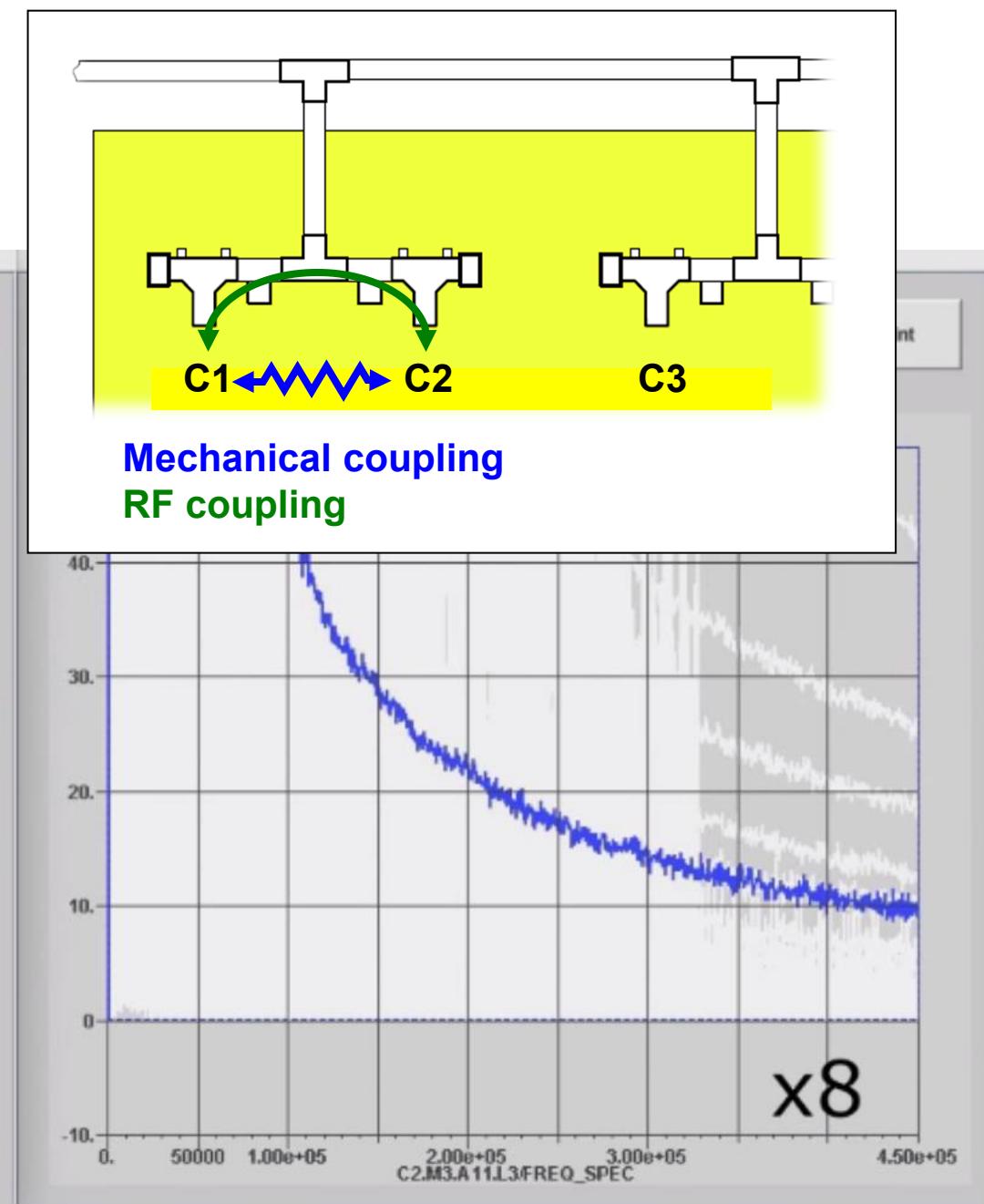
Using “slow tuner”

Detuning is visible from cavity amplitude

RF coupling to neighboring cavity is affected



Detuning is visible from cavity phase
 $\frac{d\varphi}{dt}$



Cavity tuning

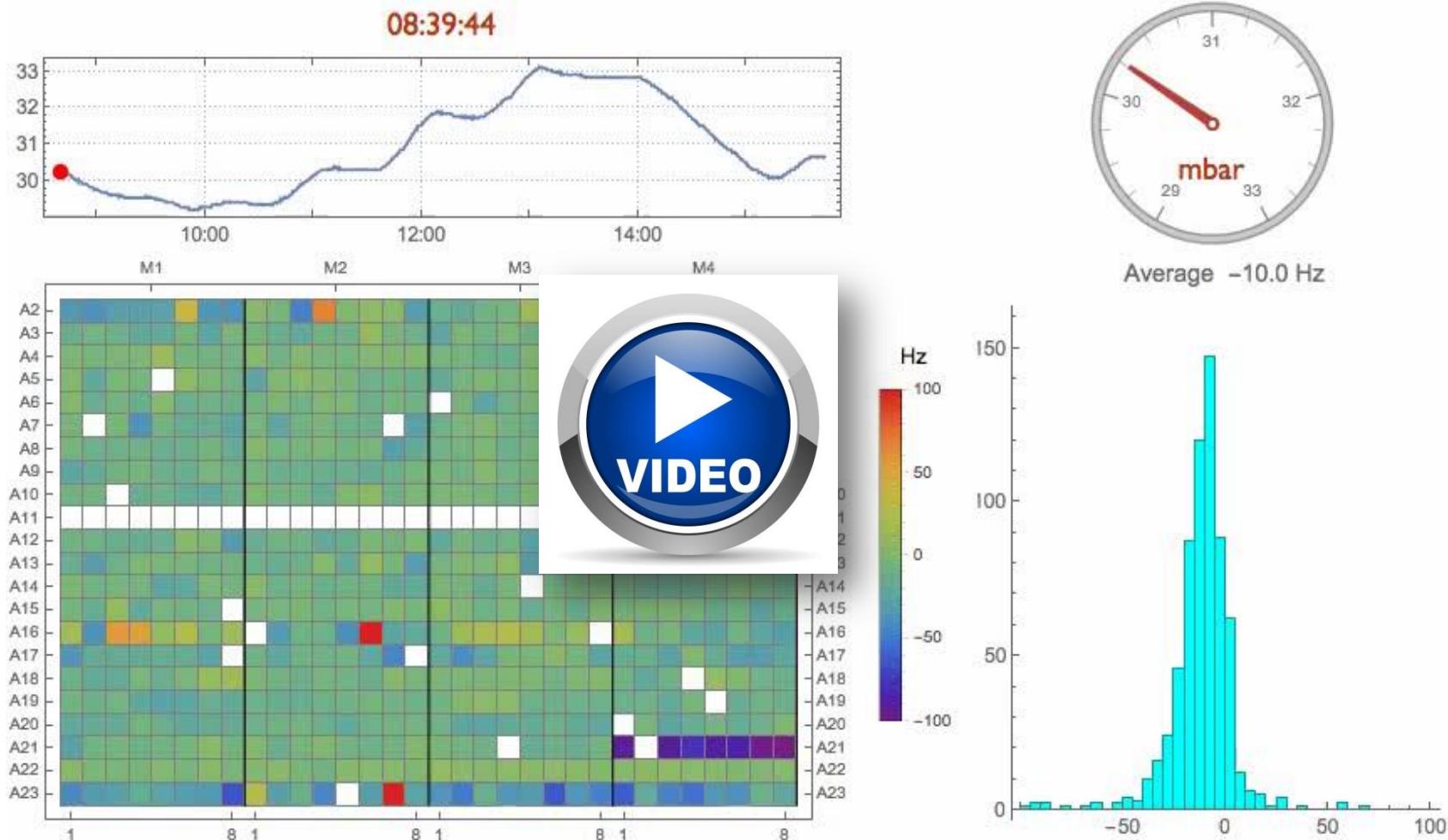
Impact of cryogenic bath

Study @ XFEL

- Vary the helium bath pressure
- Observe cavity detuning (800)

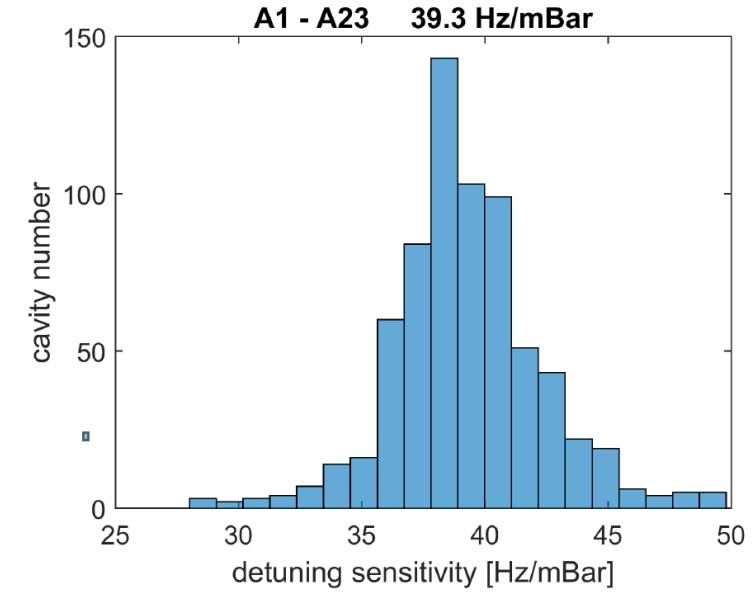
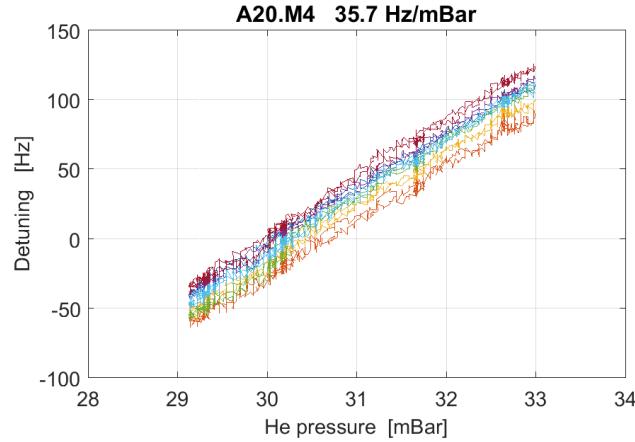
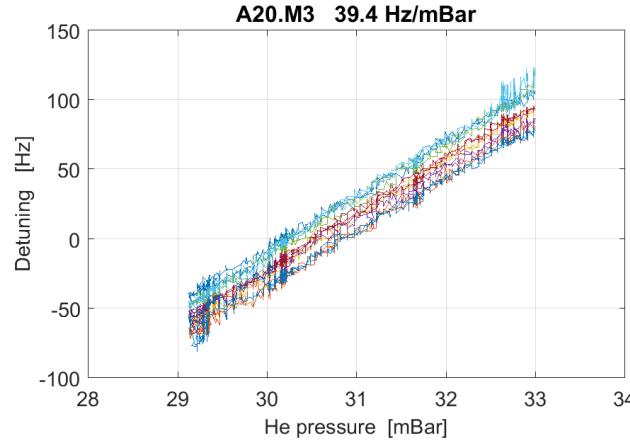
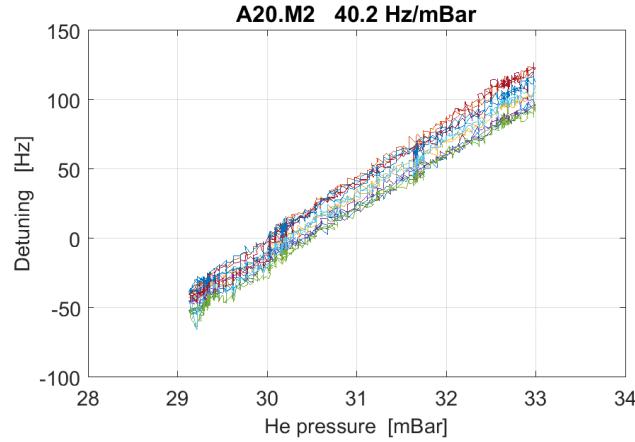
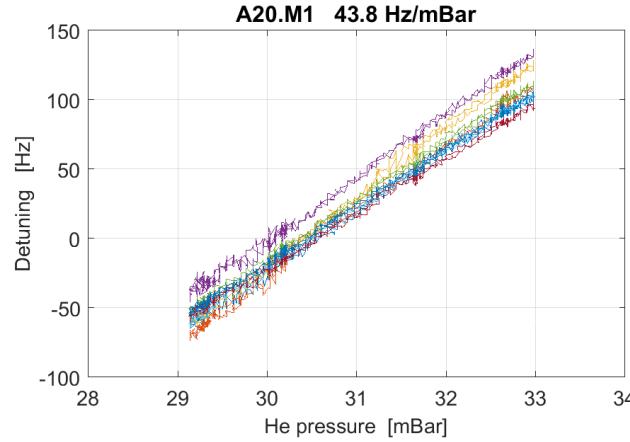
Goal

- Derive the cavity tuning sensitivity to He pressure fluctuations (i.e df/dp)
- Set safety RF operation thresholds regarding helium excursions



Cavity tuning

Impact of cryogenic bath

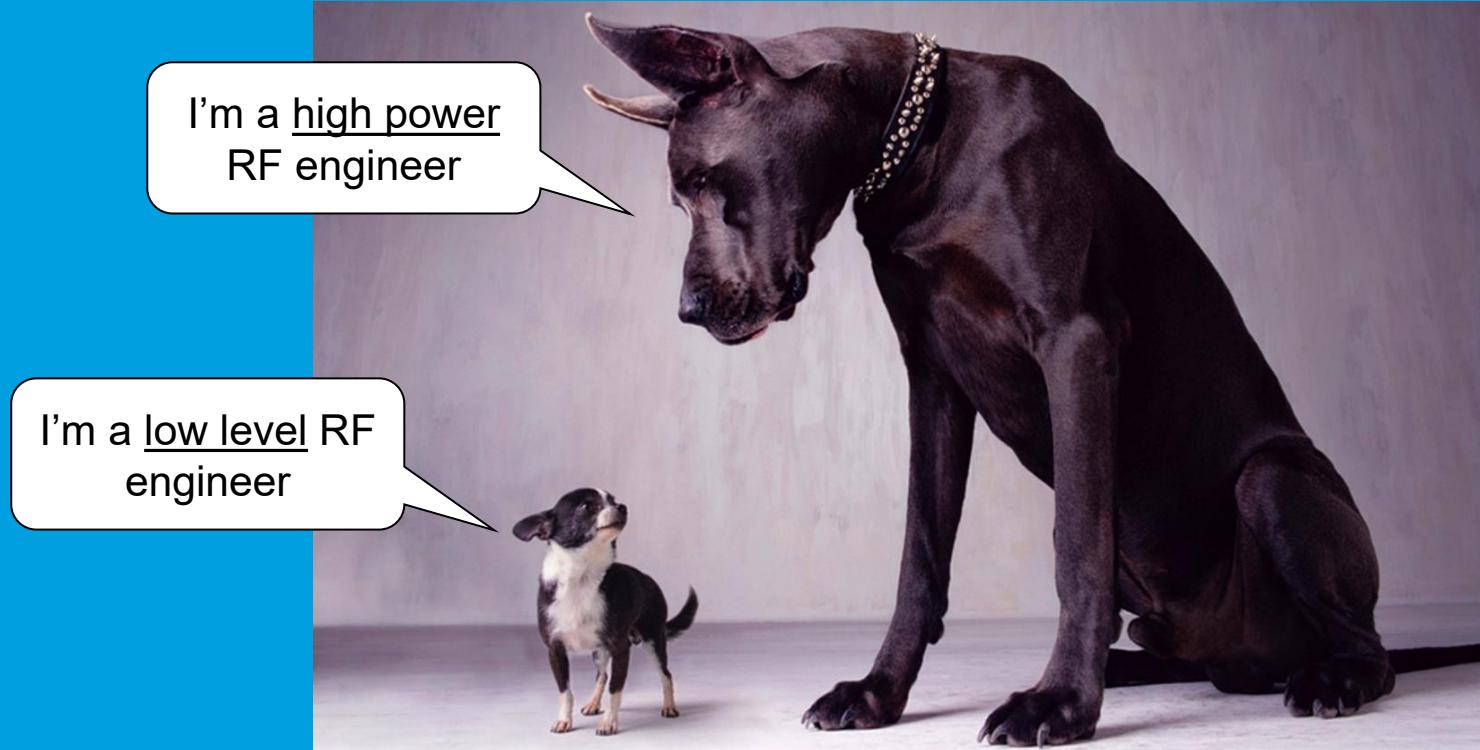


40 Hz/mBar for 1.3 GHz cavities

62 Hz/mBar for 3.9 GHz cavities

L.L.R.F.

Low Level Radio Frequency

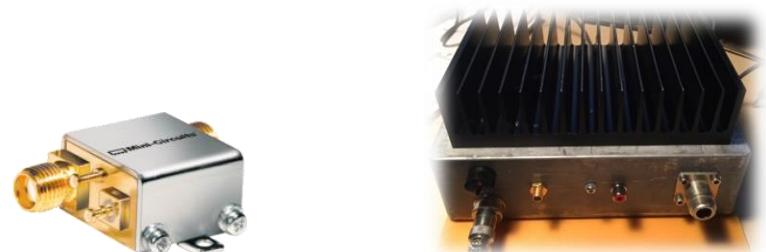


LLRF

Low level radio frequency

Low Level = low power

- mW – W



10^{-3}

mW

10^0

W

High Power

- kW – MW



10^3

kW



10^6

MW

LLRF

Low level radio frequency

Radio frequency (RF) is the oscillation rate of an alternating electric current or voltage or of a magnetic, electric or electromagnetic field or mechanical system in the frequency range from around twenty thousand times per second (**20 kHz**) to around three hundred billion times per second (**300 GHz**).

This is roughly between the **upper limit of audio frequencies** and the **lower limit of infrared frequencies**.

These are the frequencies at which energy from an oscillating current can radiate off a conductor into space as **radio waves**.

Source: Wikipedia

Frequency range	Wavelength range	ITU designation		IEEE bands ^[5]
		Full name	Abbreviation ^[6]	
3–30 Hz	10^5 – 10^4 km	Extremely low frequency	ELF	N/A
30–300 Hz	10^4 – 10^3 km	Super low frequency	SLF	N/A
300–3000 Hz	10^3 –100 km	Ultra low frequency	ULF	N/A
3–30 kHz	100–10 km	Very low frequency	VLF	N/A
30–300 kHz	10–1 km	Low frequency	LF	N/A
300 kHz – 3 MHz	1 km – 100 m	Medium frequency	MF	N/A
3–30 MHz	100–10 m	High frequency	HF	HF
30–300 MHz	10–1 m	Very high frequency	VHF	VHF
300 MHz – 3 GHz	1 m – 10 cm	Ultra high frequency	UHF	UHF, L, S
3–30 GHz	10–1 cm	Super high frequency	SHF	S, C, X, Ku, K, Ka
30–300 GHz	1 cm – 1 mm	Extremely high frequency	EHF	Ka, V, W, mm
300 GHz – 3 THz	1 mm – 0.1 mm	Tremendously high frequency	THF	N/A

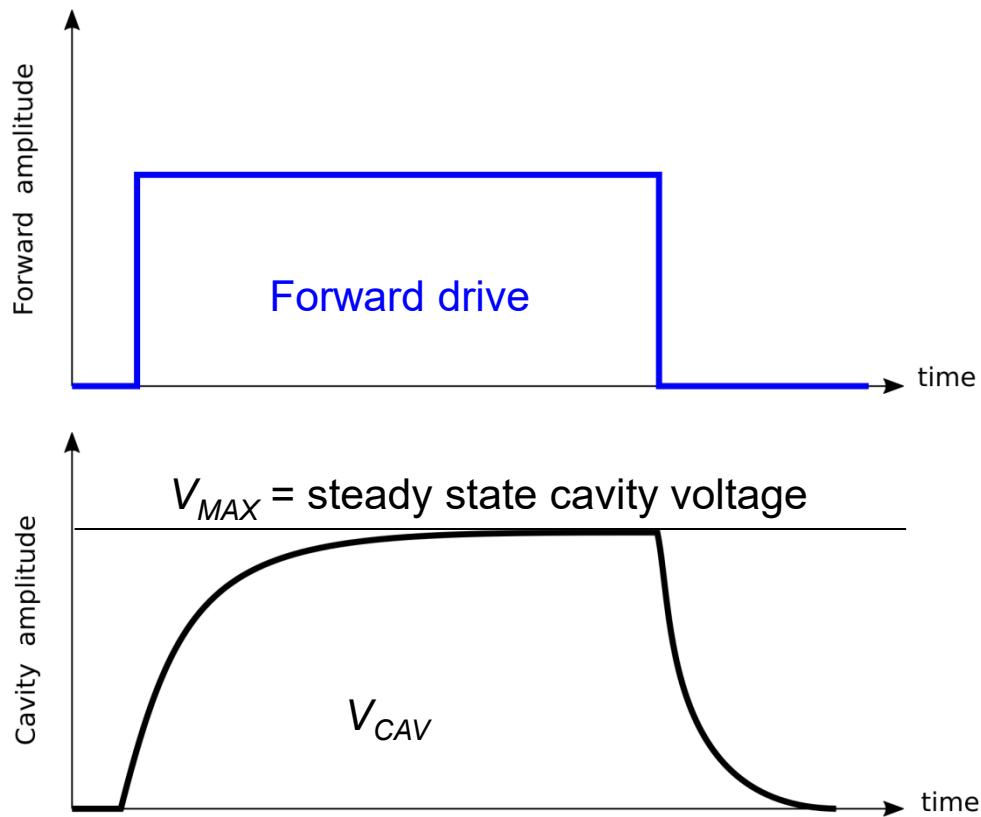


L band: 1-2 GHz
S band: 2-4 GHz
C band: 4-8 GHz
X band: 8-12 GHz

LLRF System

Feed Forward

- Cavity response to a square pulse



Q_L affects the cavity rate of filling

$$Q_L \text{ is there} \quad \tau = \frac{Q_L}{2\omega_0}$$

$$V_{CAV} = V_{MAX}(1 - e^{-t/\tau})$$

$$V_{MAX} = 2\sqrt{P_{FWD} \frac{R}{Q} Q_L}$$

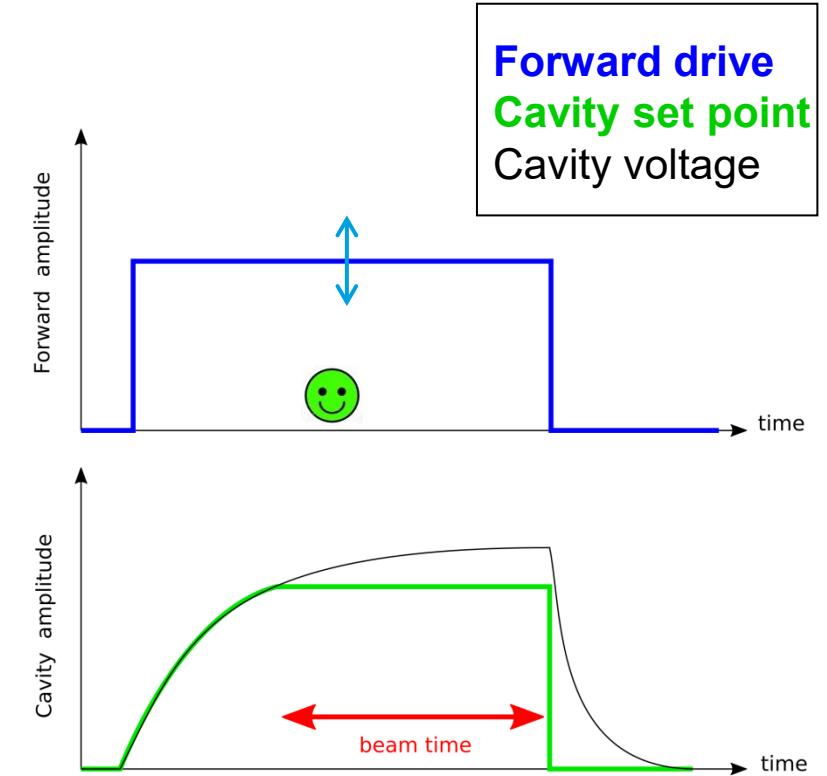
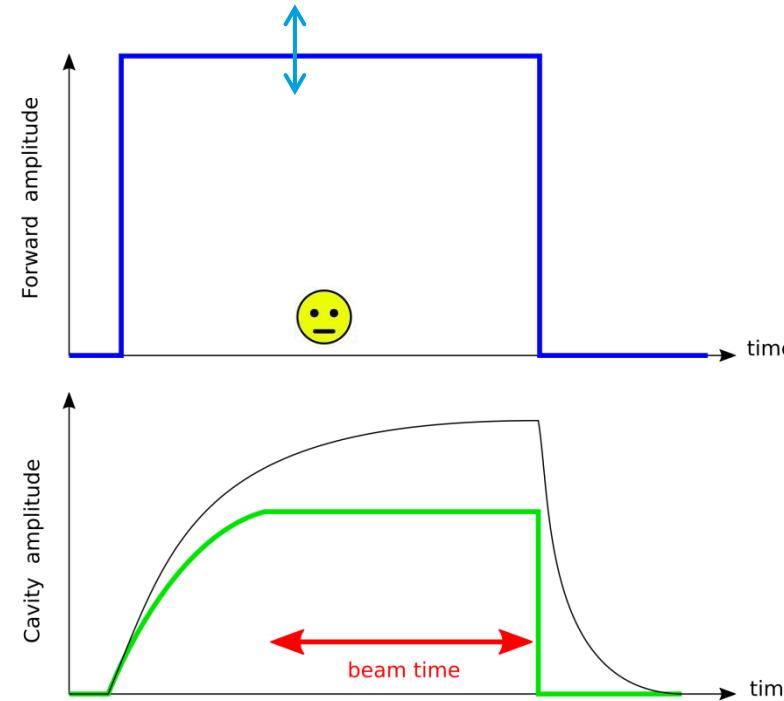
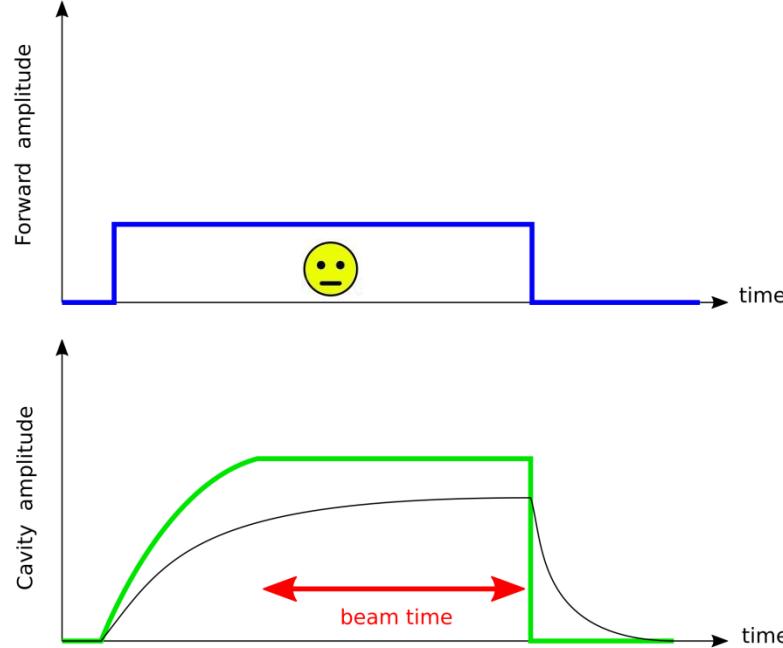
Q_L is there

Q_L affects the cavity maximum voltage
(for a given forward power)

LLRF System

Feed Forward

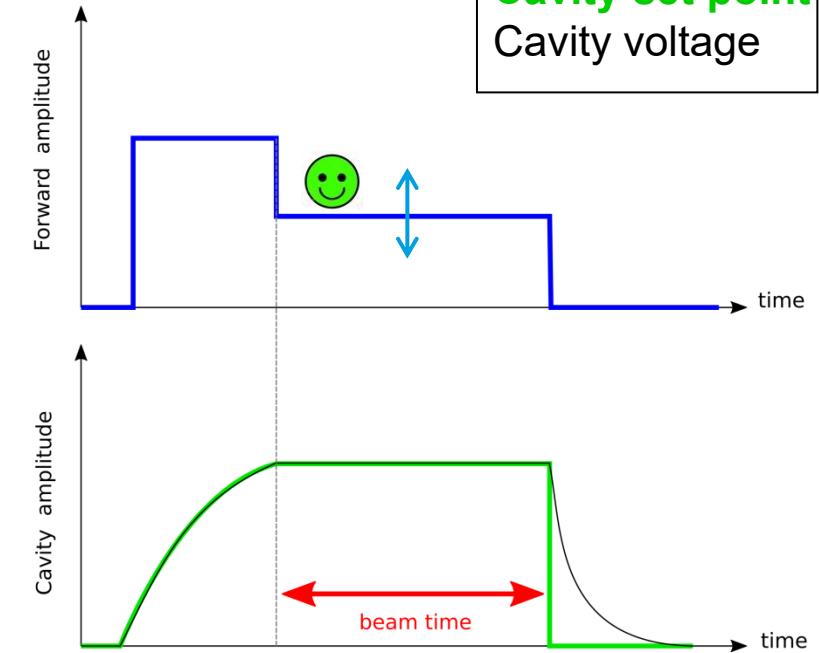
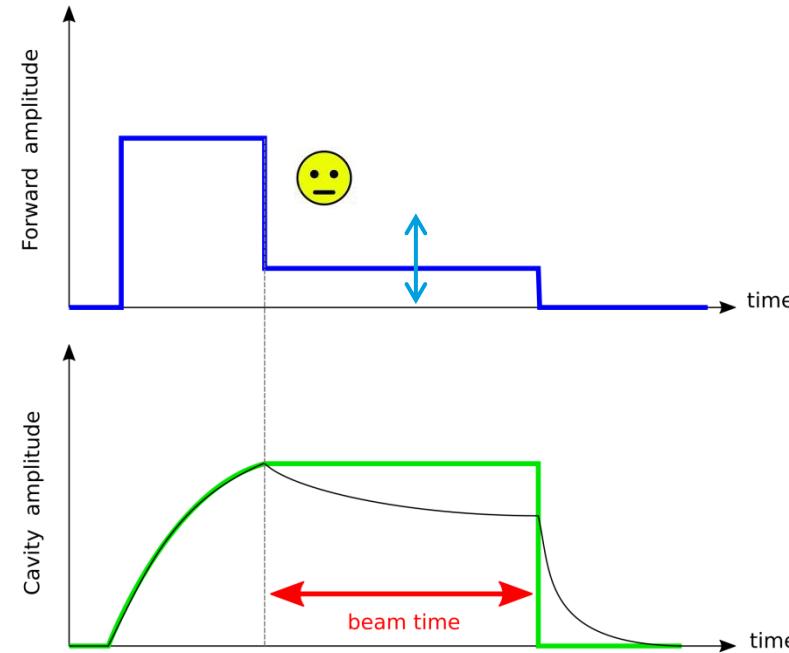
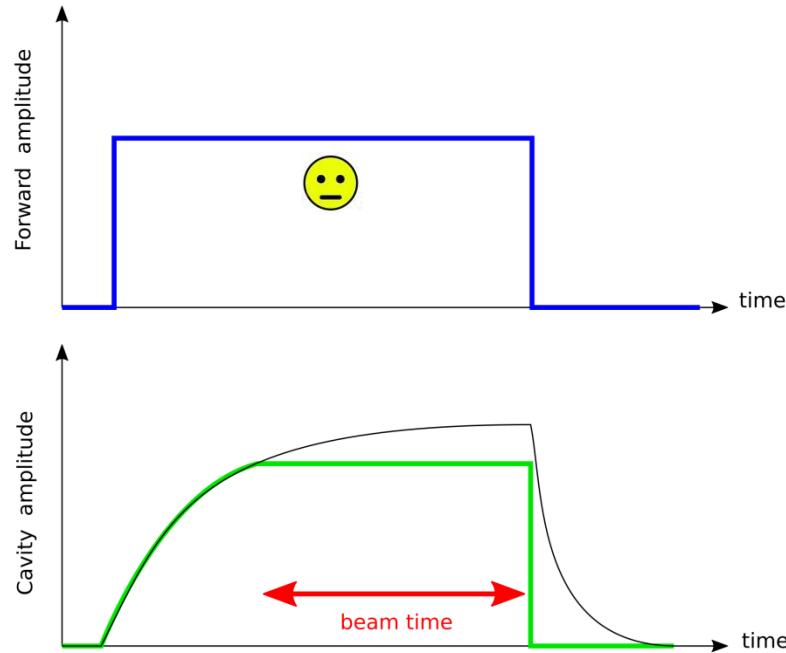
- Adjust amplitude of the **forward drive** to match the set point gradient **at the beginning of the beam time**



LLRF System

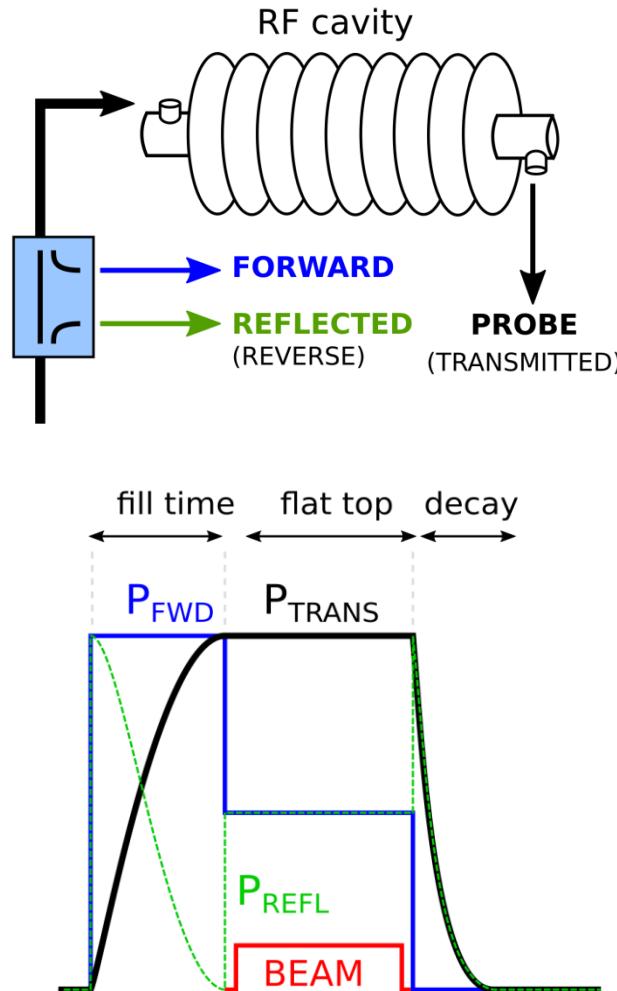
Feed Forward

- Adjust the drive **during the beam time** to **maintain a flat accelerating gradient**



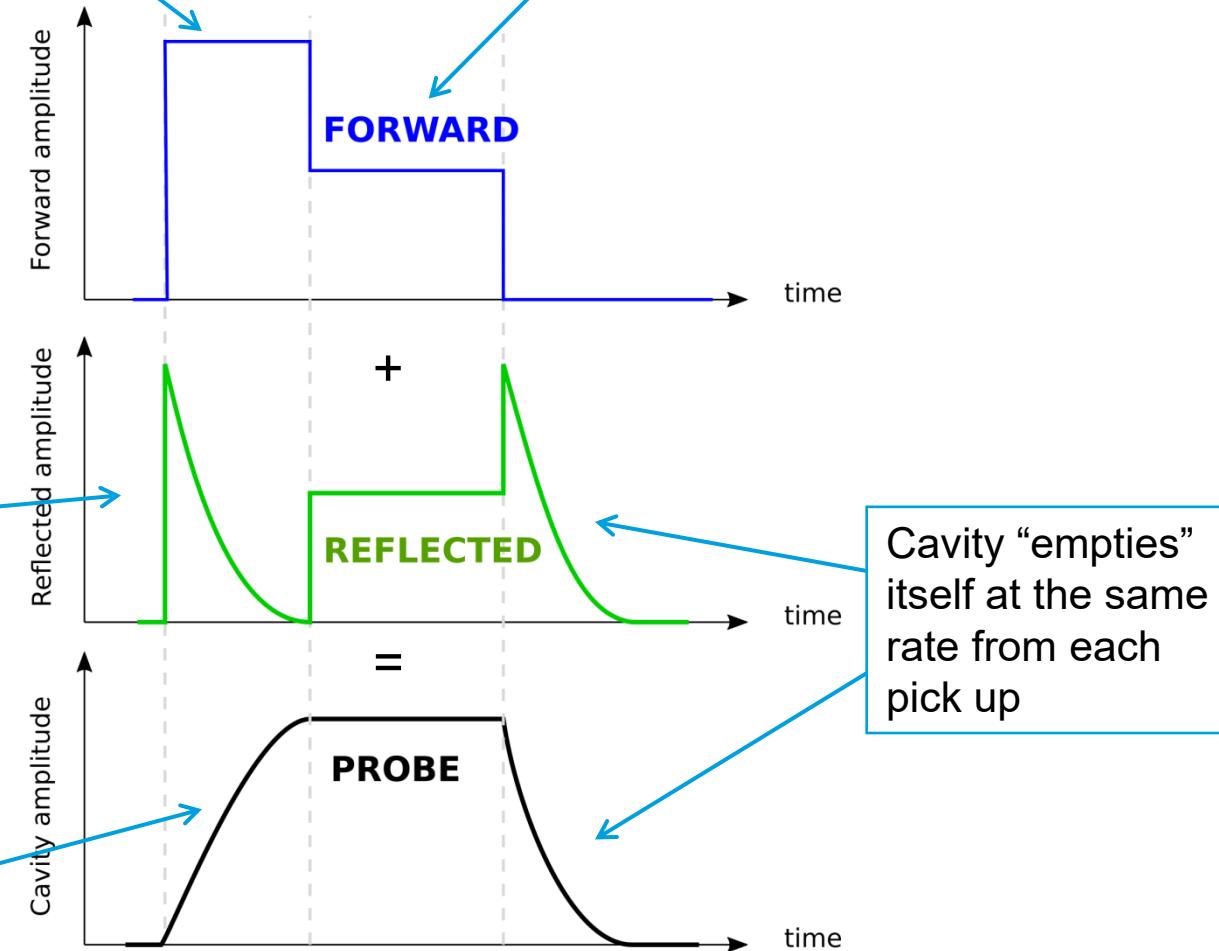
LLRF System

Forward, Reflected and Probe Signals



High forward power to fill cavity as fast as possible

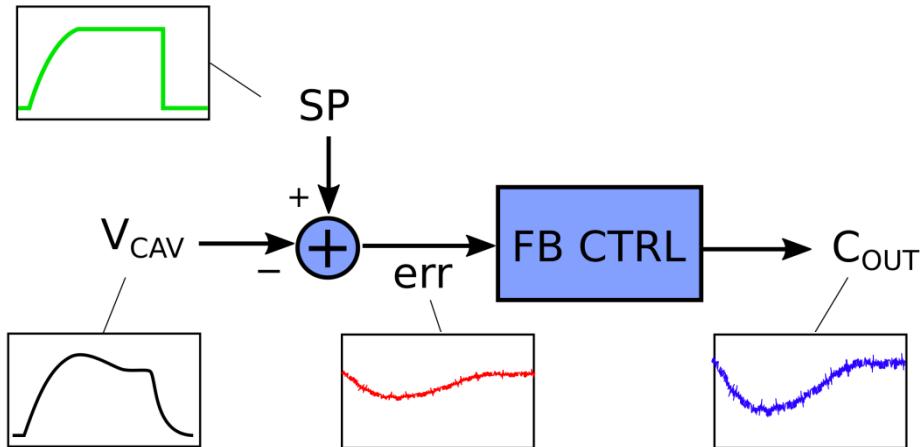
Drop forward power when target gradient is reached (i.e. don't fill till steady state)



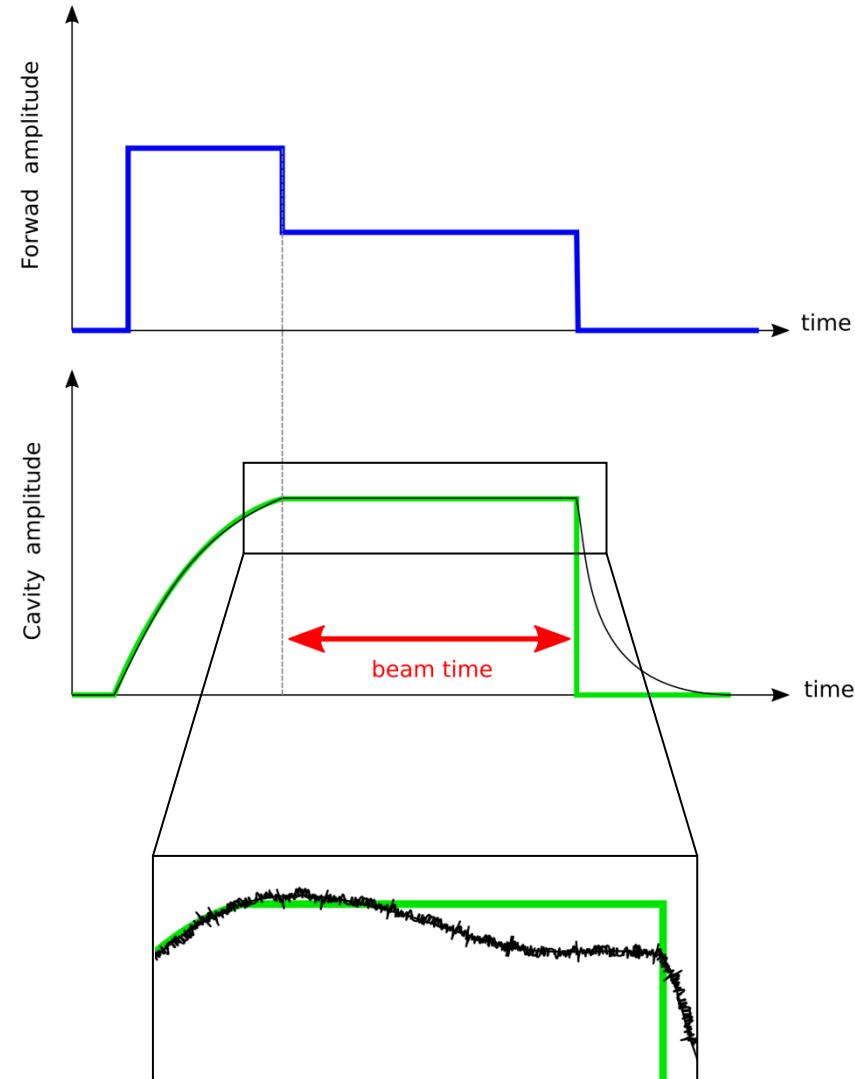
LLRF System

Why do we need feedback?

- “Feed forward only gets you in the ball park, but you need feedback to achieve the regulation required by the beam”
- Compute error: $\text{err} = \text{SP} - V_{\text{CAV}}$



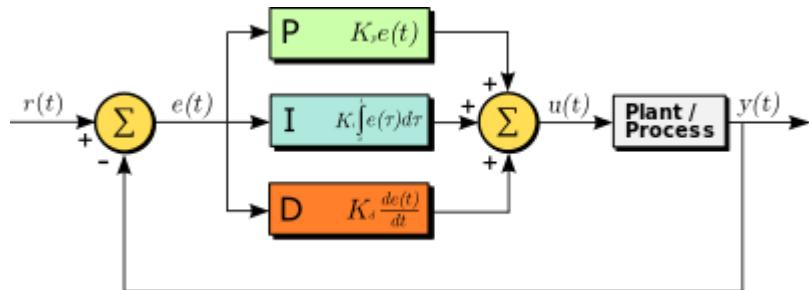
- Simplest feedback is proportional : $C_{\text{OUT}} = K_p * \text{err}$



LLRF System

Feedback

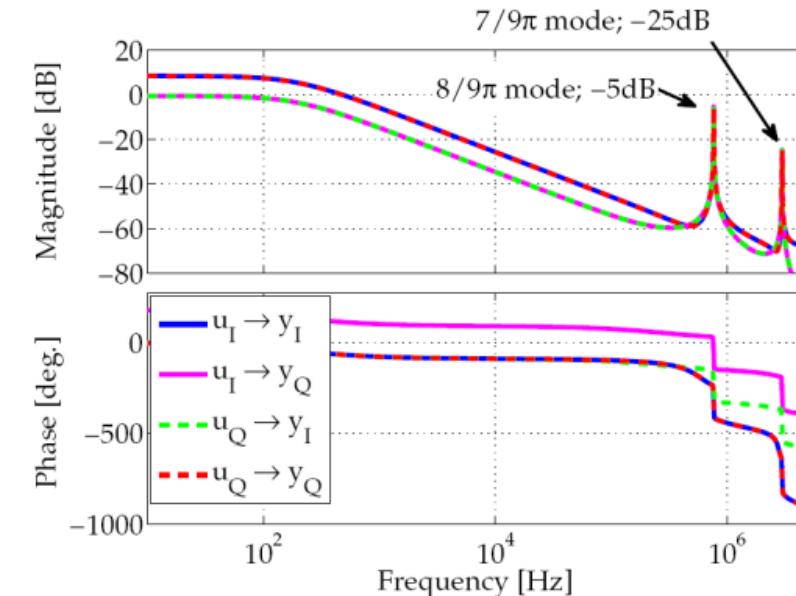
- Classic feedback controllers



- P: proportional controller output scales with the input error
- I: integral controller minimizes the steady state error left from the proportional controller correction
- D: differential controller tries to minimize rapid error changes

- Modern feedback controller

- e.g. 2x2 MIMO controller (can do PID and more...)
 - Cancellation of one pass band mode
 - Cancellation of cross coupling between inputs

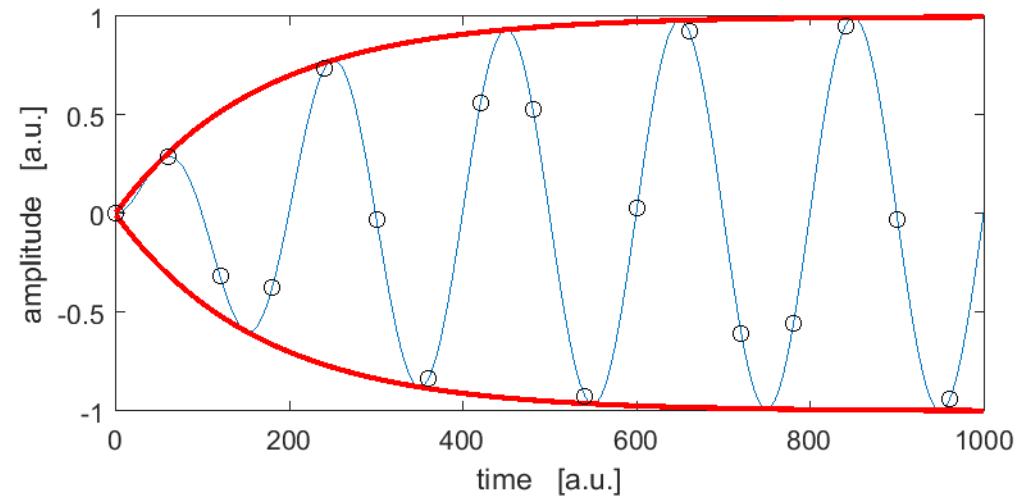
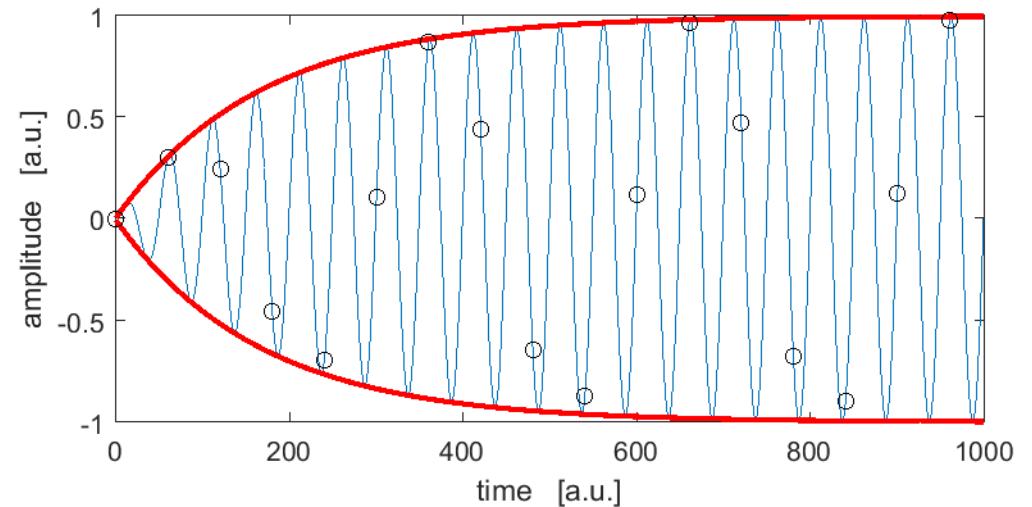


Reference : S. Pfeiffer: "LLRF controls and Feedback", CERN Accel. School on FELs & ERLs, Hamburg, 2016
S. Pfeiffer: "Advanced LLRF system setup tool for RF field regulation of SRF cavities", SRF, 2019

LLRF System

Down Conversion

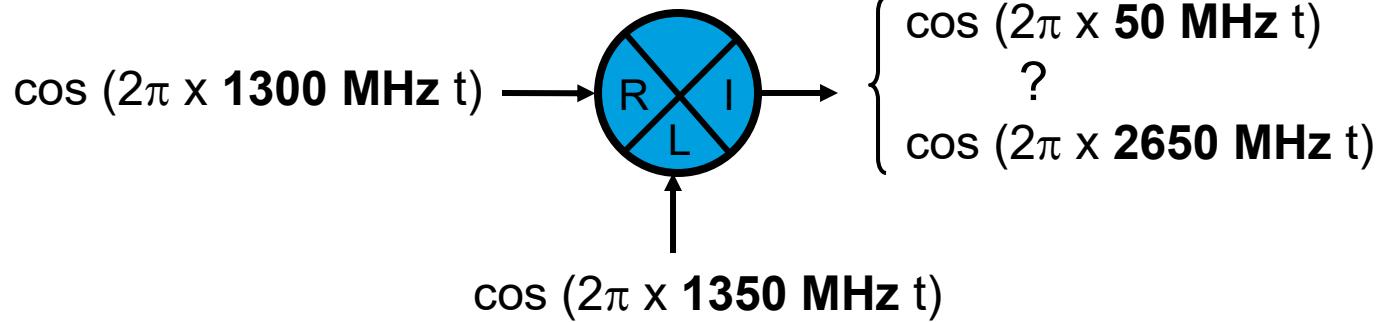
- Transfer envelop information of the cavity signals to a lower carrier frequency, easier to digitize
 - RF frequency: 1.3 GHz
 - Typical ADC sample frequencies 60-150 MSPS
 - Typical down converted frequencies: 10-60 MHz



LLRF System

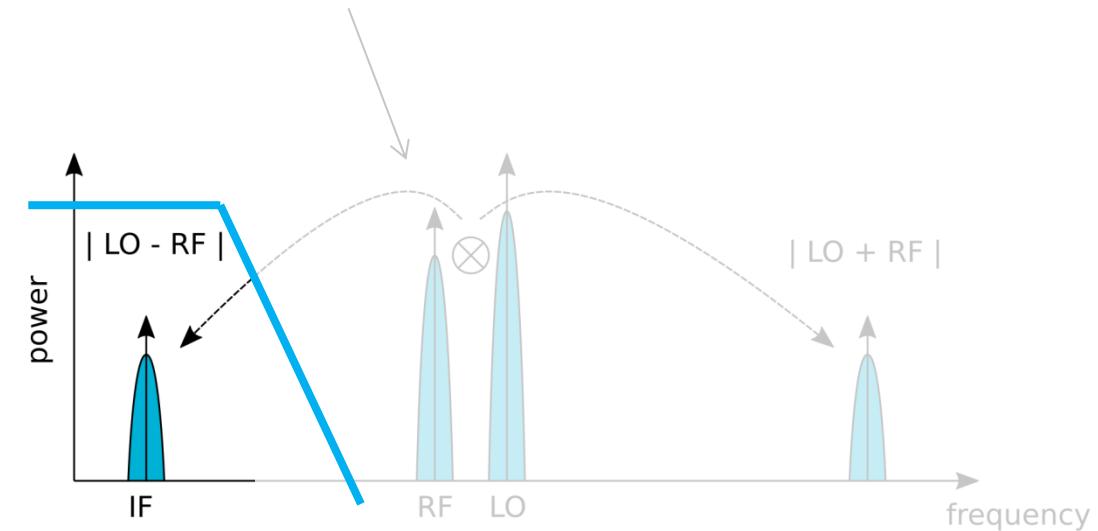
Down Conversion

- Frequency translation
from Radio Frequency **RF**
to Intermediate Frequency **IF**
via a Local Oscillator **LO** signal
- The fundamental component allowing
for this operation is the **RF mixer**
- The corresponding operation is a **frequency
multiplication**



Source: RF microwave

Cavity signal with relevant information
(probe, forward, reflected) at RF frequency

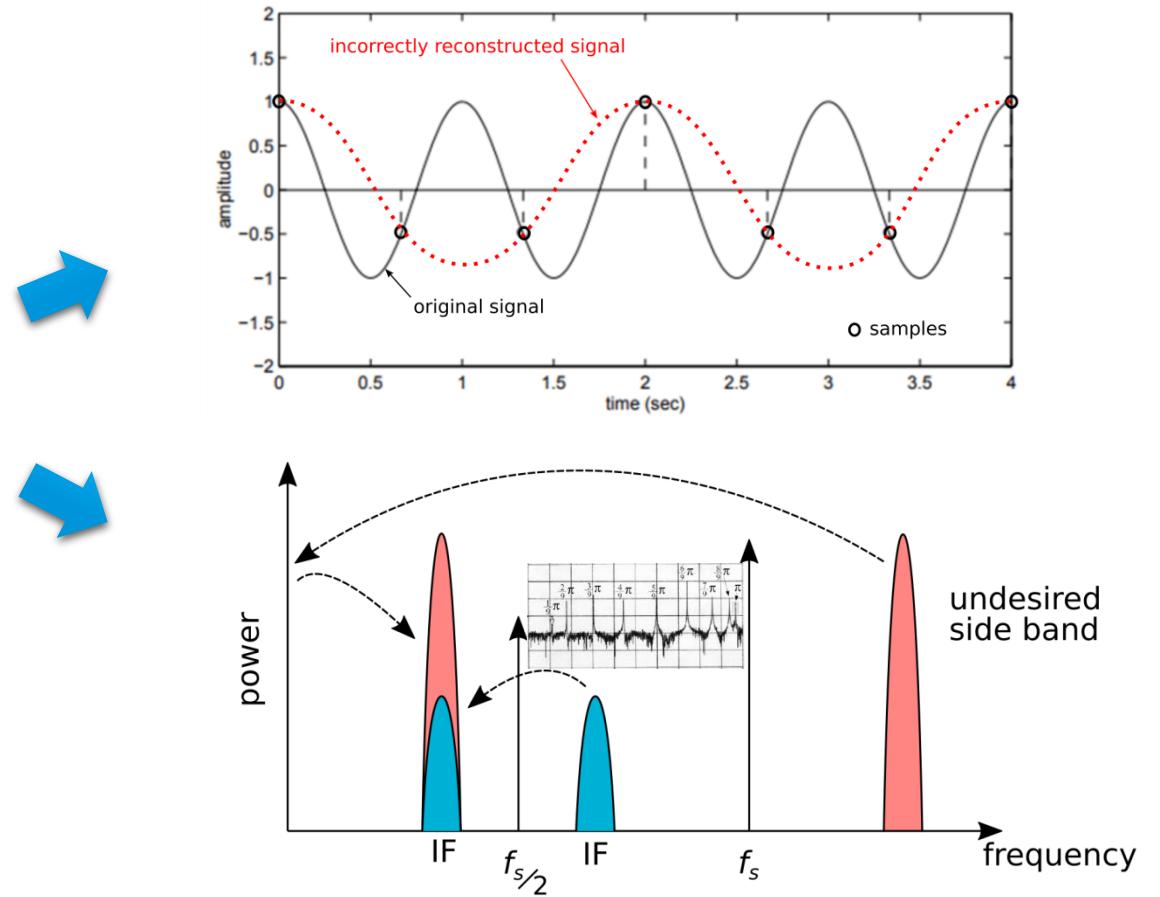


$$\begin{aligned} \cos a \cos b &= \frac{1}{2} \cos(a+b) + \frac{1}{2} \cos(a-b) \\ \sin a \sin b &= \frac{1}{2} \cos(a-b) - \frac{1}{2} \cos(a+b) \\ \sin a \cos b &= \frac{1}{2} \sin(a+b) + \frac{1}{2} \sin(a-b) \\ \cos a \sin b &= \frac{1}{2} \sin(a+b) - \frac{1}{2} \sin(a-b) \end{aligned}$$

LLRF

Sampling

- Sampling frequency
 - Under-sampling looses information
 - avoid undesired information (noise, side bands), folding onto the carrier information
- Nyquist frequency
 - *"the minimum rate at which a signal can be sampled without introducing errors, which is twice the highest frequency present in the signal."*
 - In practice, under sampling is OK if the signal bandwidth is within 1 Nyquist zone
- ADC range
 - ADC saturation loses amplitude information
 - ADC range under usage deteriorates the signal to noise ratio



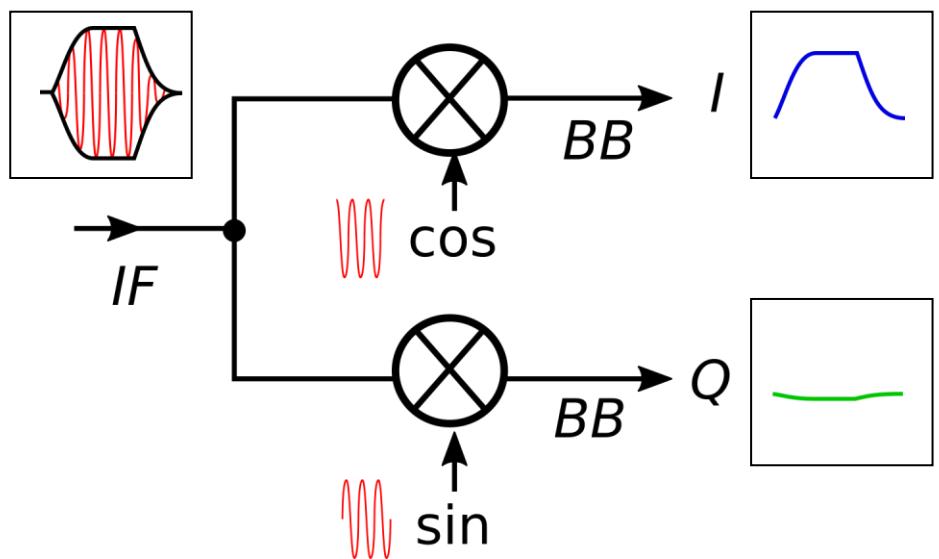
optimization of ADC dynamic range

Typically condition signal to stay at 70-80% of max ADC range

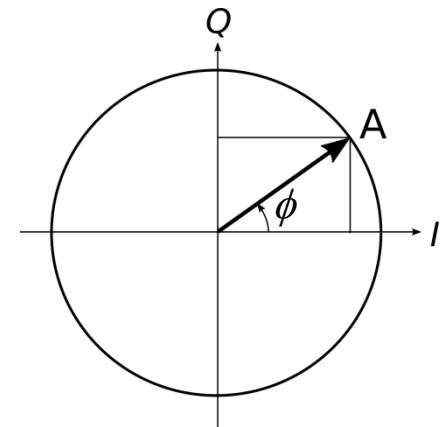
LLRF System

I Q detection

- I Q detection
 - At IF sampling, **multiplication by IF sine and cosine tables**



- Phasor diagram



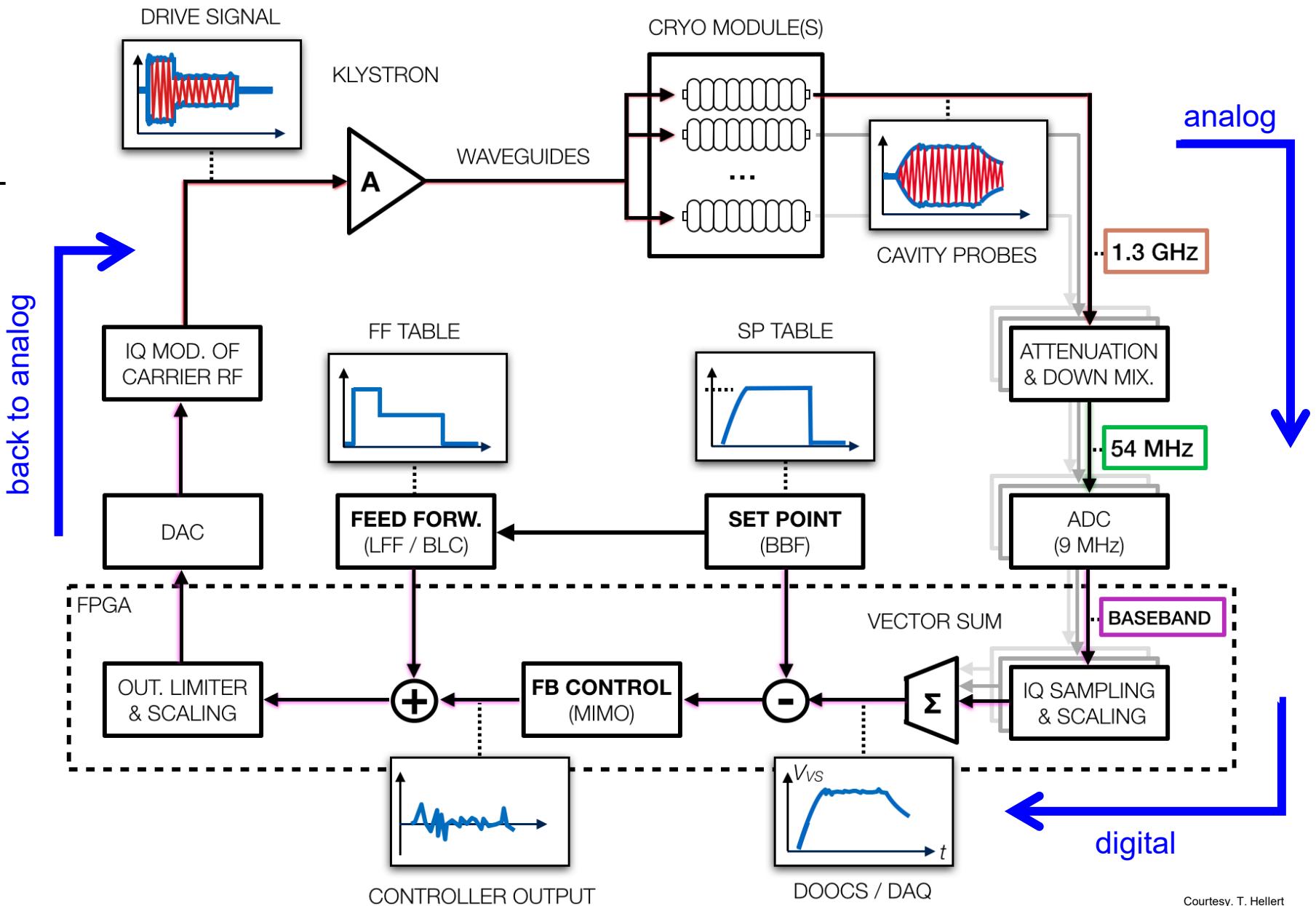
$$\begin{aligned} I &= A \cdot \cos \phi \\ Q &= A \cdot \sin \phi \\ A &= \sqrt{I^2 + Q^2} \\ \phi &= \text{atan2}(Q, I) \end{aligned}$$

- In practice, digital scaling and rotation is also taking place at this stage → **digital signal calibration**

LLRF system

The complete loop

- Example from EuXFEL

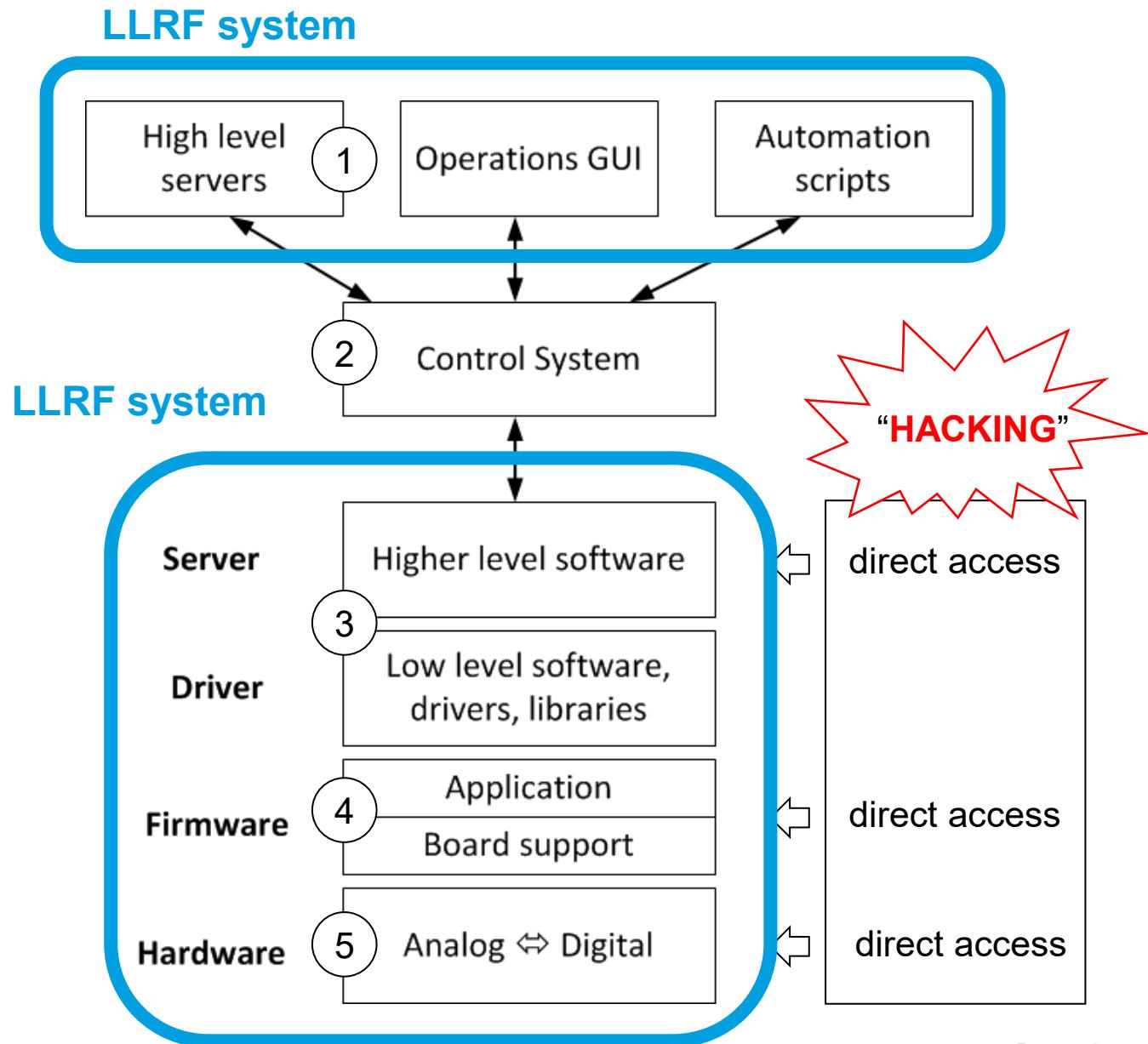


Courtesy: T. Hellert

LLRF system

The bird's eye overview

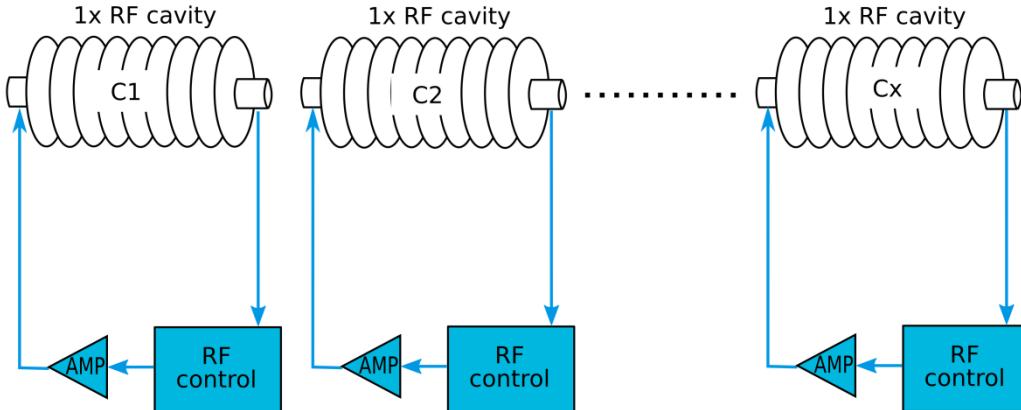
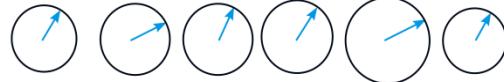
- Typical layers in system architecture
- The functionalities described earlier are spread over several layers
- E.g. set point
 - 1 User can change the SP via GUI (new value, slope etc...) OR energy server can request a SP change
 - 2 The request is sent over the control system to the LLRF controller server
 - 3 The LLRF controller server computes the new set point tables and writes them to firmware registers via the driver
 - 4 The firmware feedback mechanisms adapt the drive to this new request
 - 5 The hardware drive signal is modified accordingly



Single Cavity versus Vector Sum RF regulation

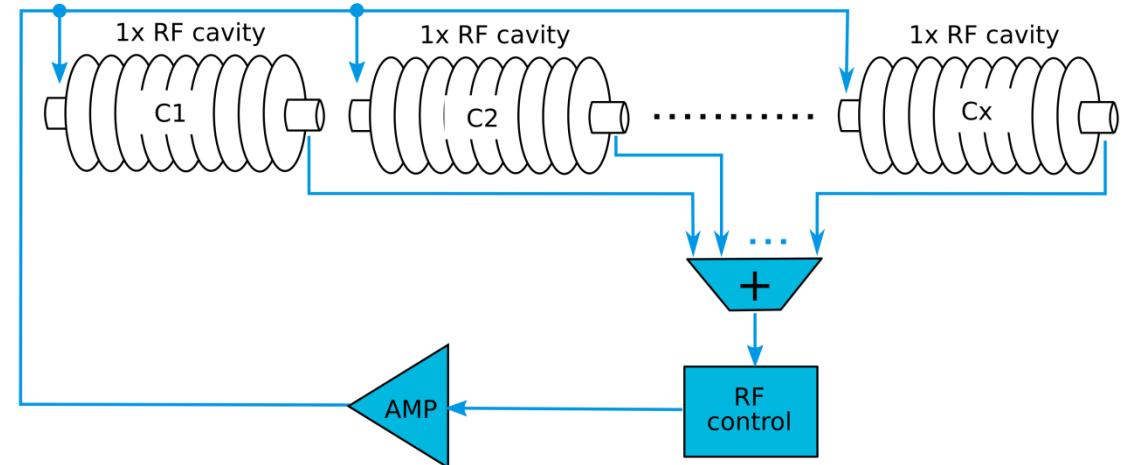
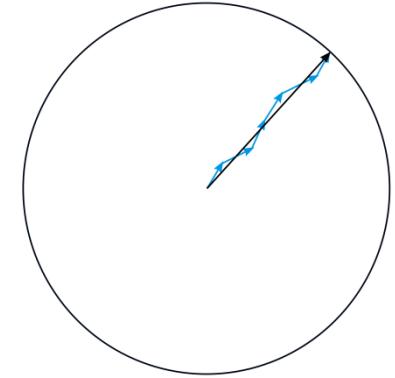
Single Cavity

- One high power actuator per cavity
- i.e. Solid State Amplifier (SSA)
- Example: LCLS-II
- Pros
- Simpler regulation



Vector Sum

- One high power actuator per many cavities
- i.e. pulsed klystron
- Example: Eu-XFEL
- Pros
- Cost reduction



Pulsed versus Continuous Wave (CW) operation

Note: the beam is always pulsed

Benefit of Short Pulse (SP) operation

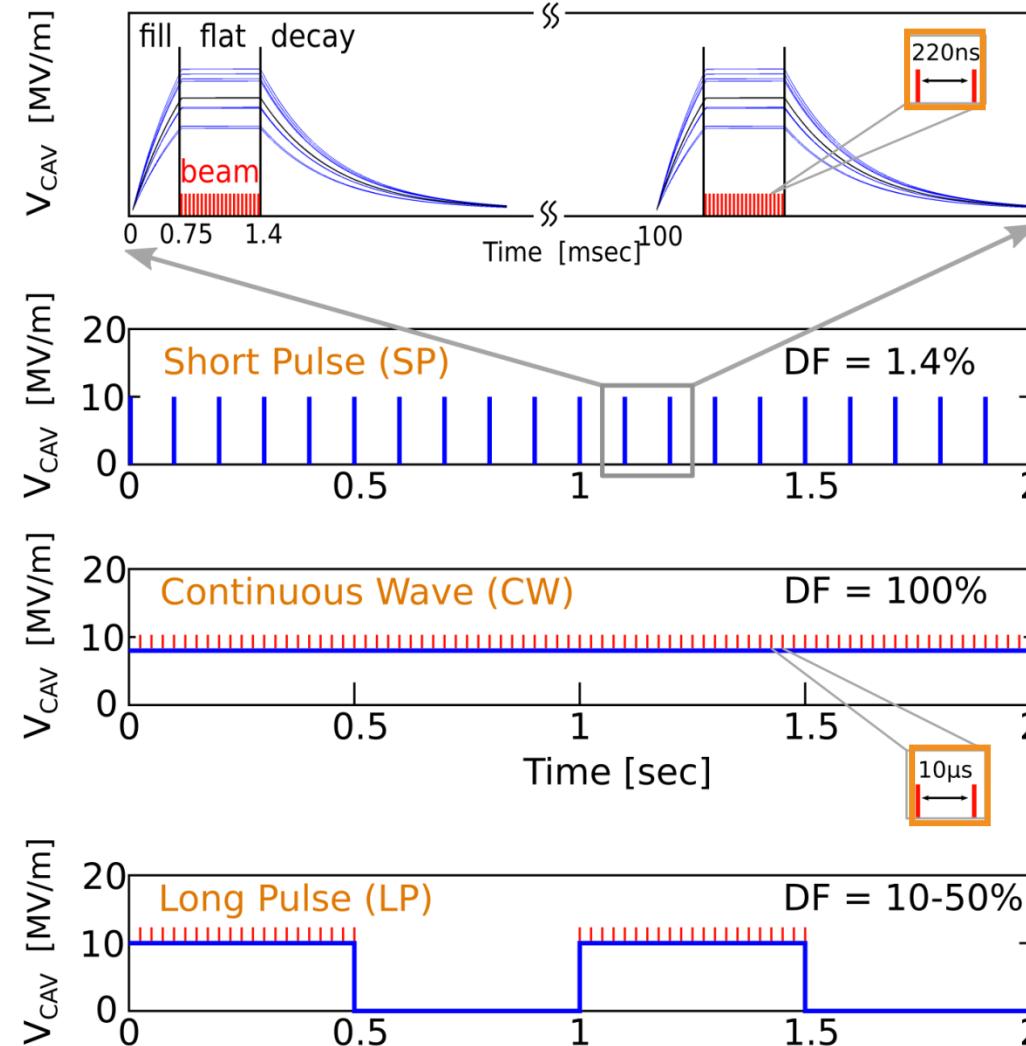
- Lower dynamic heat load → cryo ☺
- Higher energies → eg. shorter wavelength for FEL

Benefits of Continuous Wave (CW) operation

- Flexible beam patterns for detectors
- Slower repetition rate lasers
- Fill-transients no longer an issue

Benefits of Long Pulse (LP) operation

- Still high duty factor (DF = 10-50%)
- Higher gradients than CW with same heat load



FLASH.
Free-Electron Laser FLASH

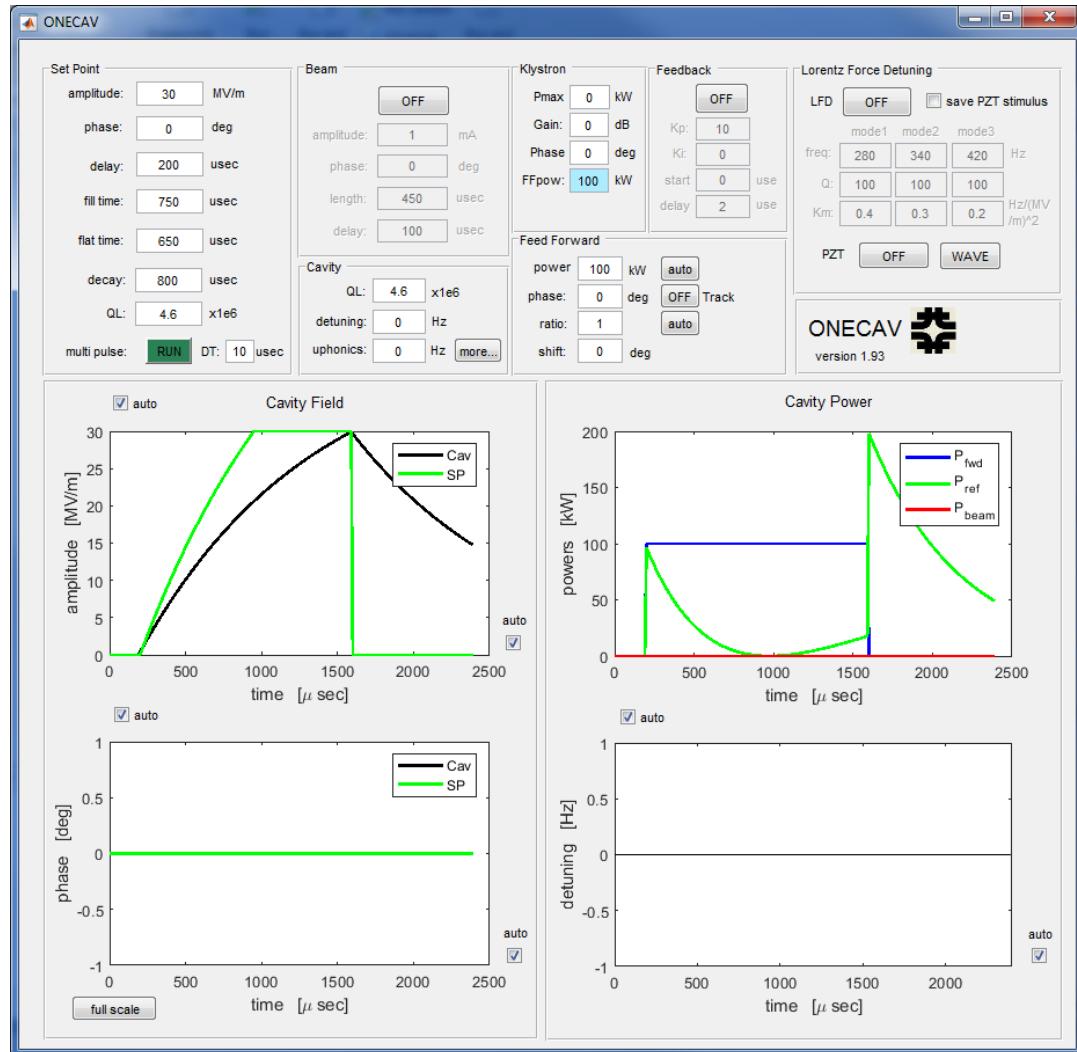
$E = 17.5 \text{ GeV}$
 $N_{\text{bunch}}/\text{s} = 27k$

$E = 8 \text{ GeV}$
 $N_{\text{bunch}}/\text{s} = 100k$

$E = 10 \text{ GeV}$
 $N_{\text{bunch}}/\text{s} = 50k$

LLRF System

DEMO: ONECAV simulation



- MATLAB simulator available for download
 - DESY intranet: <http://www.desy.de/~branlard/>
 - On request: julien.branlard@desy.de
 - Useful to understand cavity behavior under RF control
 - Comments, bug fixes are welcome!
- Demo covers:
 - FF control
 - Probe, forward and reflected signals
 - Impact of detuning
 - Feedback
 - Proportional and integral gains actions
 - Beam and beam loading
 - Impact of changing Q_L
 - Long pulse and power overhead

RF operations



Source: gifsec.com

RF operations

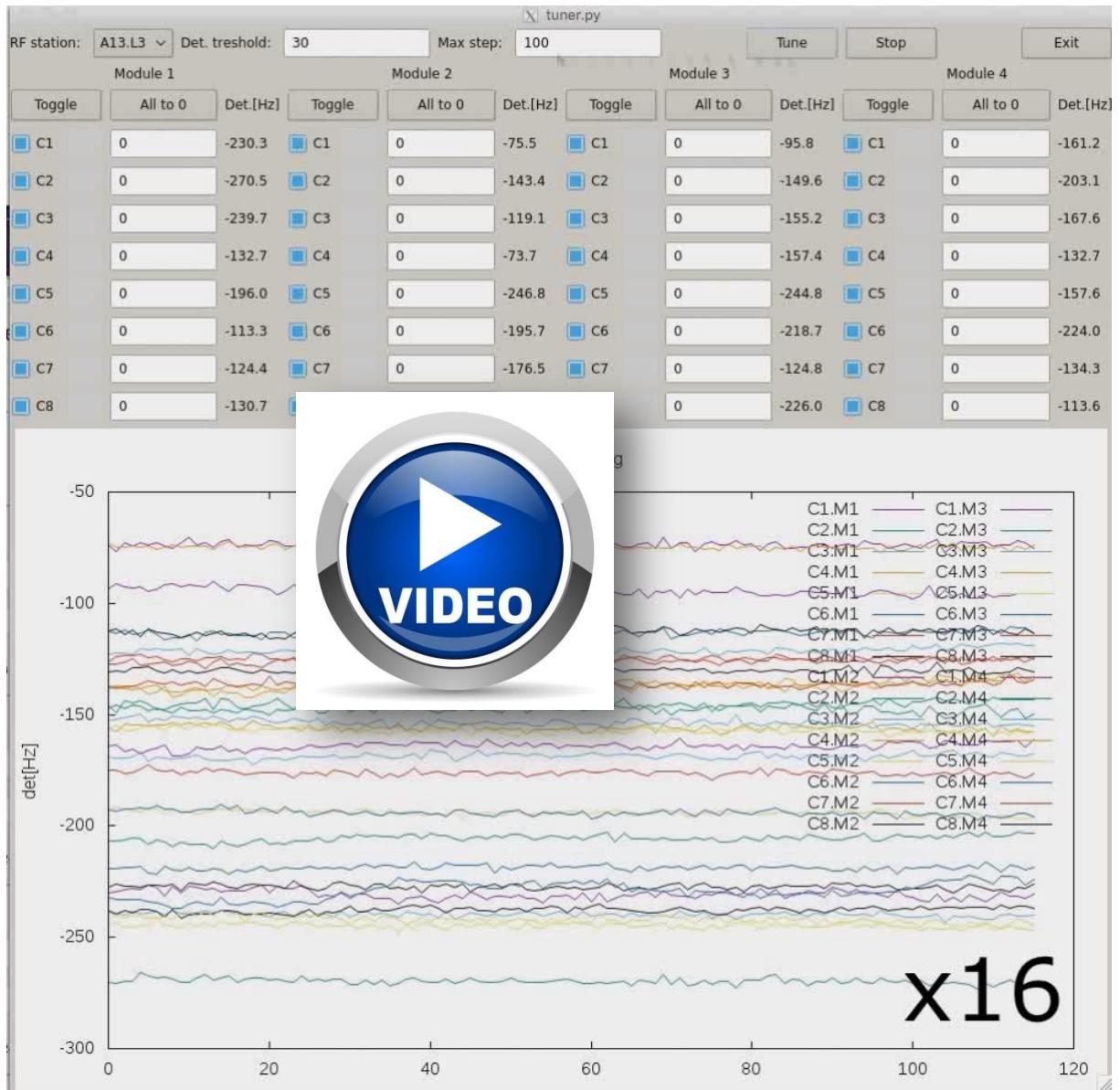
Routine operation

- RF operation covers many aspects of what was presented until now
 - An RF station ramping up / ramping down
 - Cavity fine tuning
 - Energy feedback server adjusting set point
 - Calibration procedures using beam
 - RF trips investigation (post mortem DAQ analysis, finding the root cause)
 - Dedicated study (e.g. additive beam arrival time jitter induced by individual components)
 - Etc...
- Only present a couple of examples here

Cavity coarse tuning

Using tuner motor

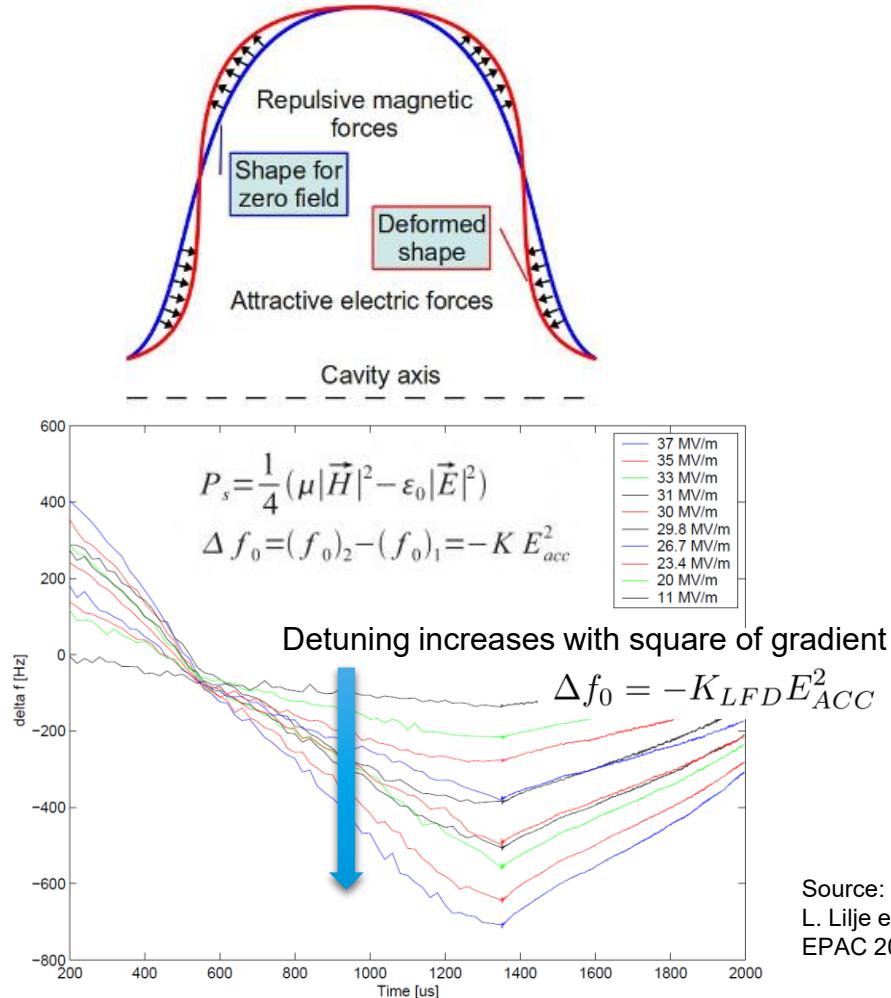
- Example taken from EuXFEL
- 32 cavities to be tuned
- Initial detuning ranging from -50 to -300 Hz
- Goal is: $| \text{detuning} | < 30 \text{ Hz}$
- Script adjusted 32 cavities in < 1 min



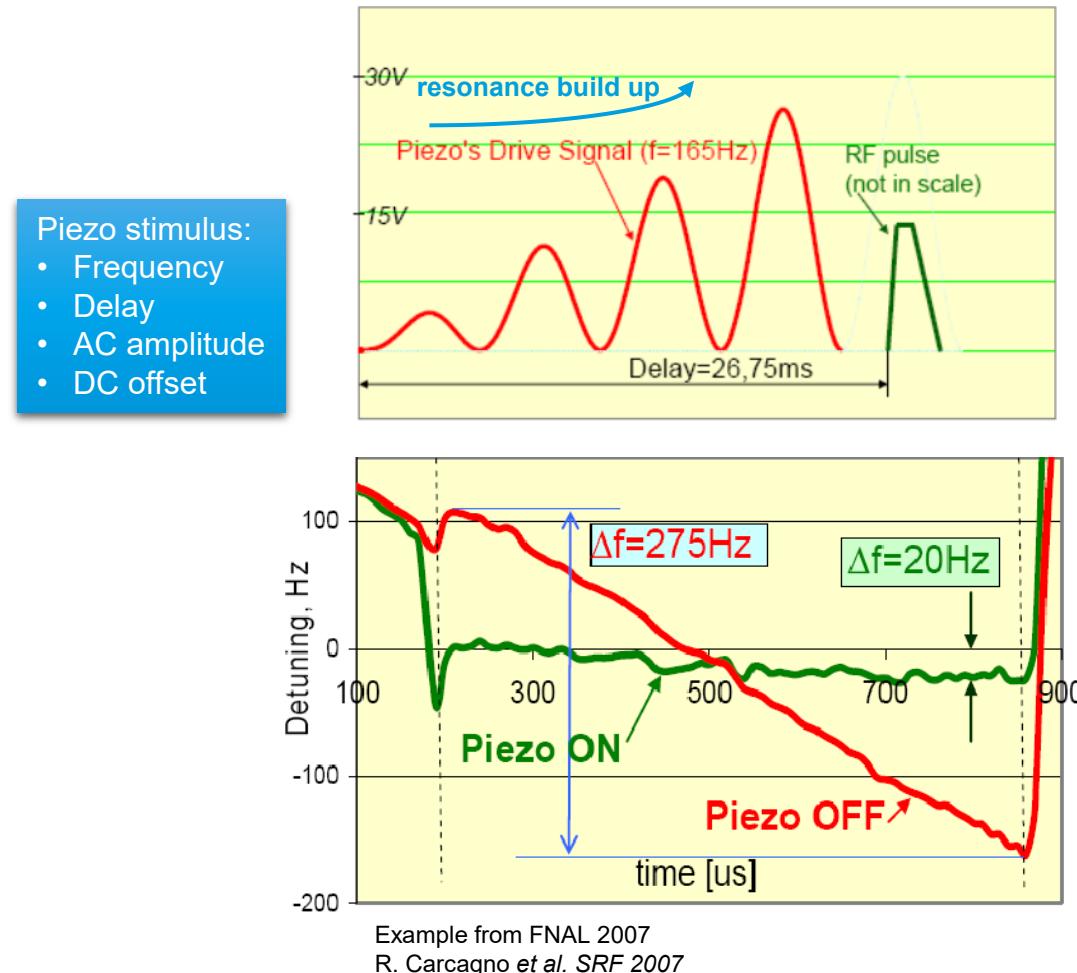
Cavity “fine” tuning

Use of piezo in pulsed mode

- Lorentz Force detuning



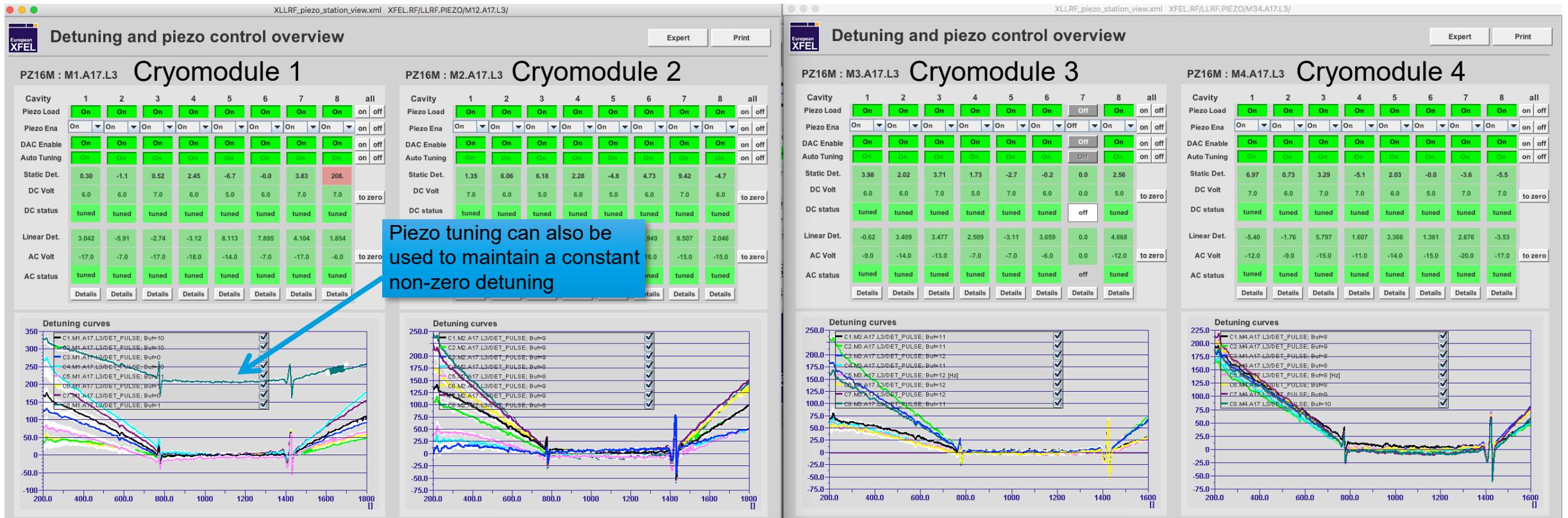
- Lorentz force detuning compensation using piezo



Cavity “fine” tuning

Another example of Lorentz force detuning compensation

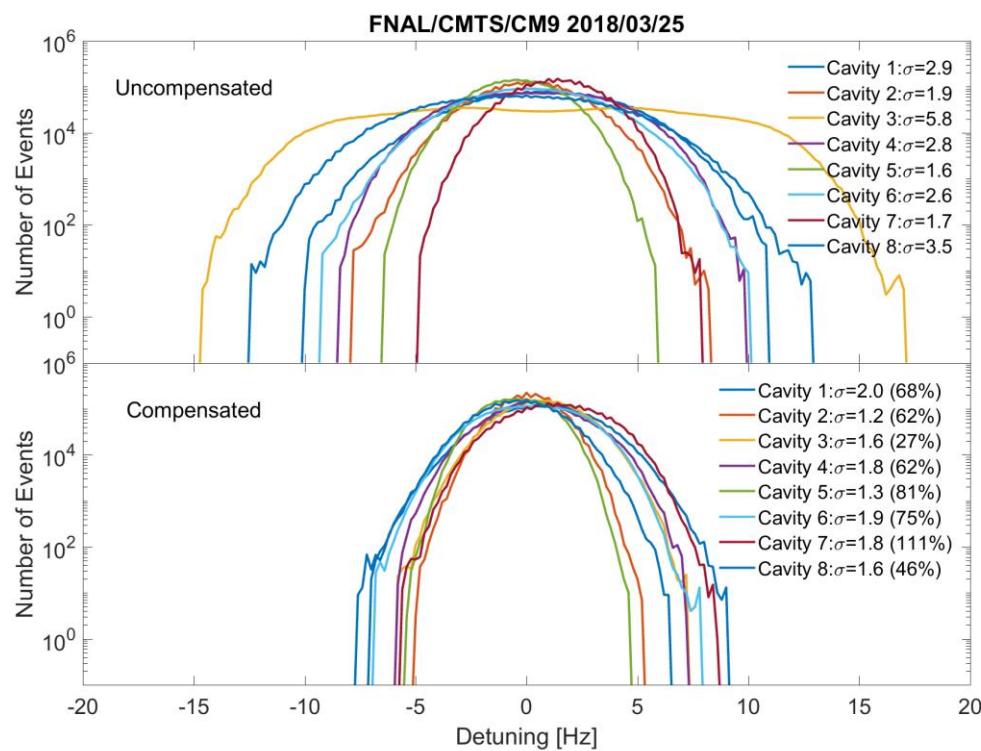
- Example from EuXFEL: (station A17)
 - 32 cavities, LFD compensated



Cavity “fine” tuning

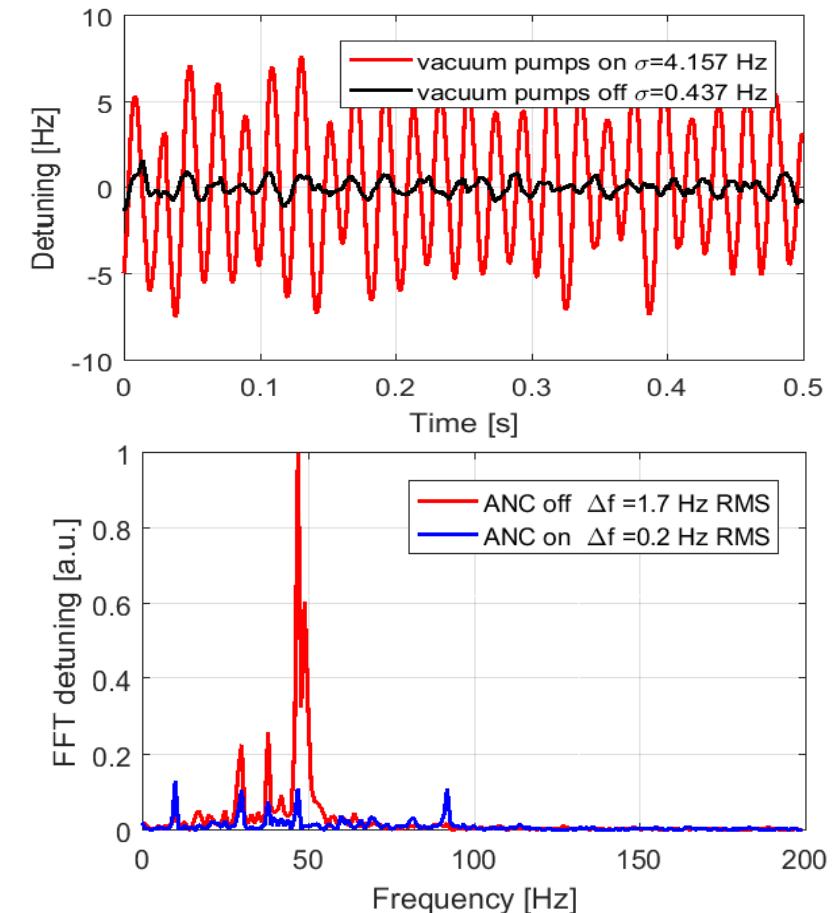
Microphonics and microphonics compensation

- Transfer function / noise spectrum techniques



Source: W. Schappert “Active Resonance Control Algorithm Development for LCLS-II”, 2nd microphonics workshop, 2018

- Active noise compensation (notch filter) techniques

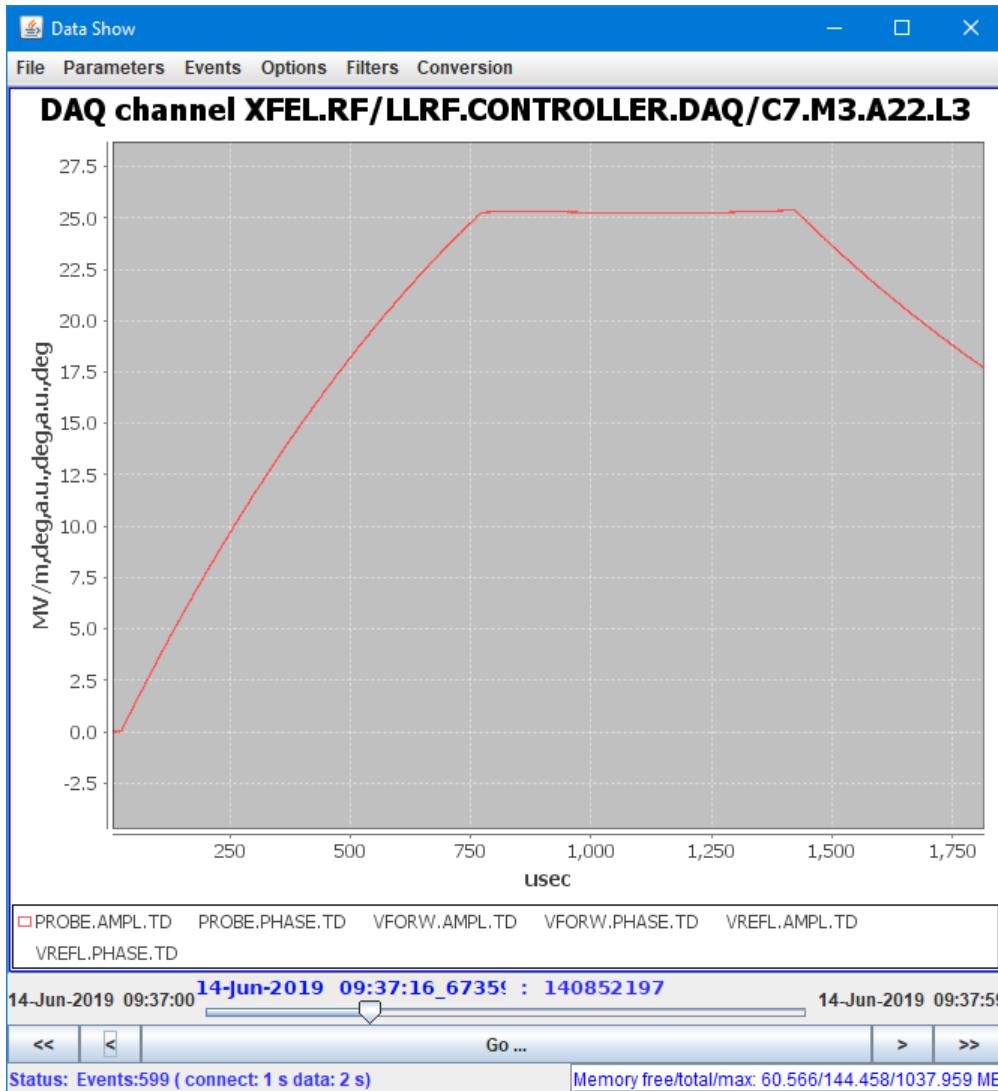


RF operation

Experience of a cavity quench

Data retrieved from the DAQ system at EuXFEL on 14 Jun. 2019

Quench triggered (by slowly increasing the gradient) to test the quench detection / reaction mechanism



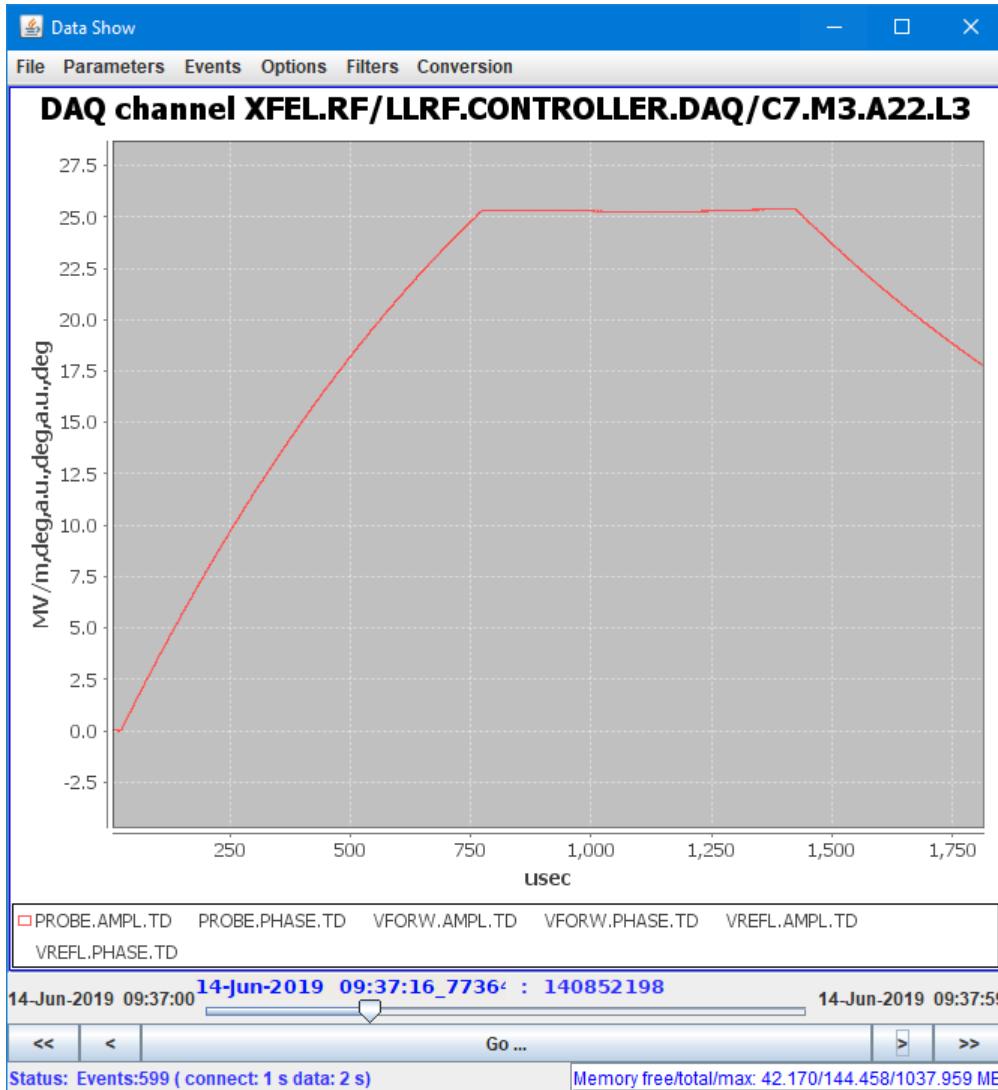
Pulse 1

$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

Gradient is slowly increased



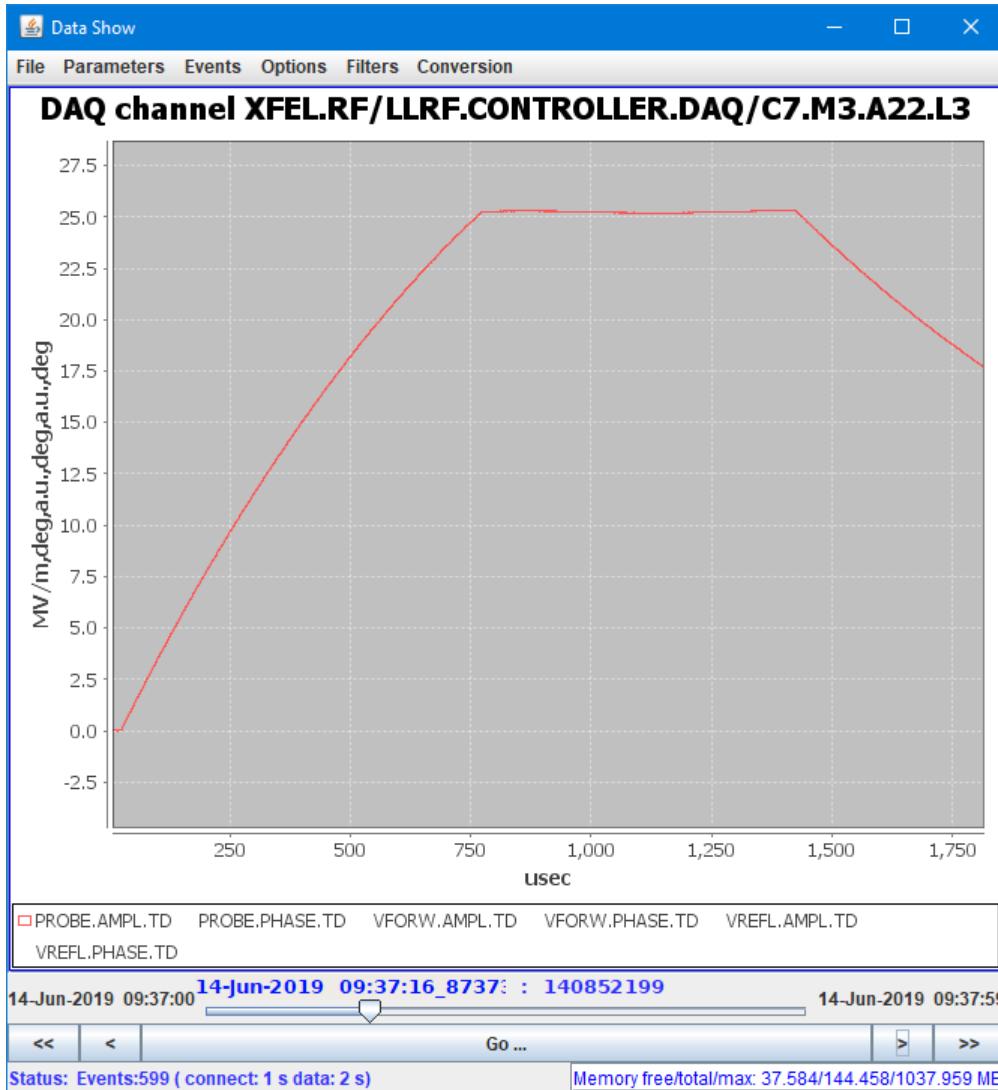
Pulse 2

$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

Gradient is slowly increased



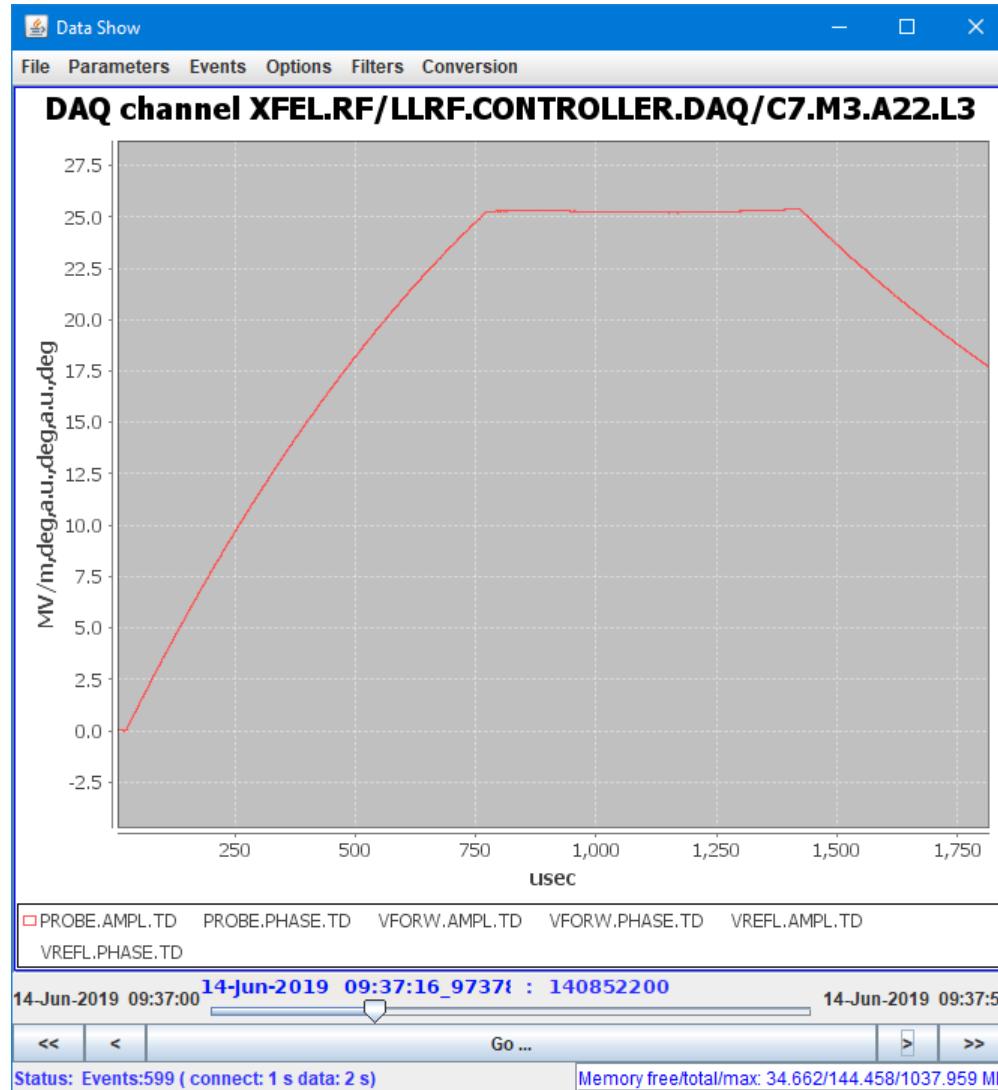
Pulse 3

$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

Gradient is slowly increased



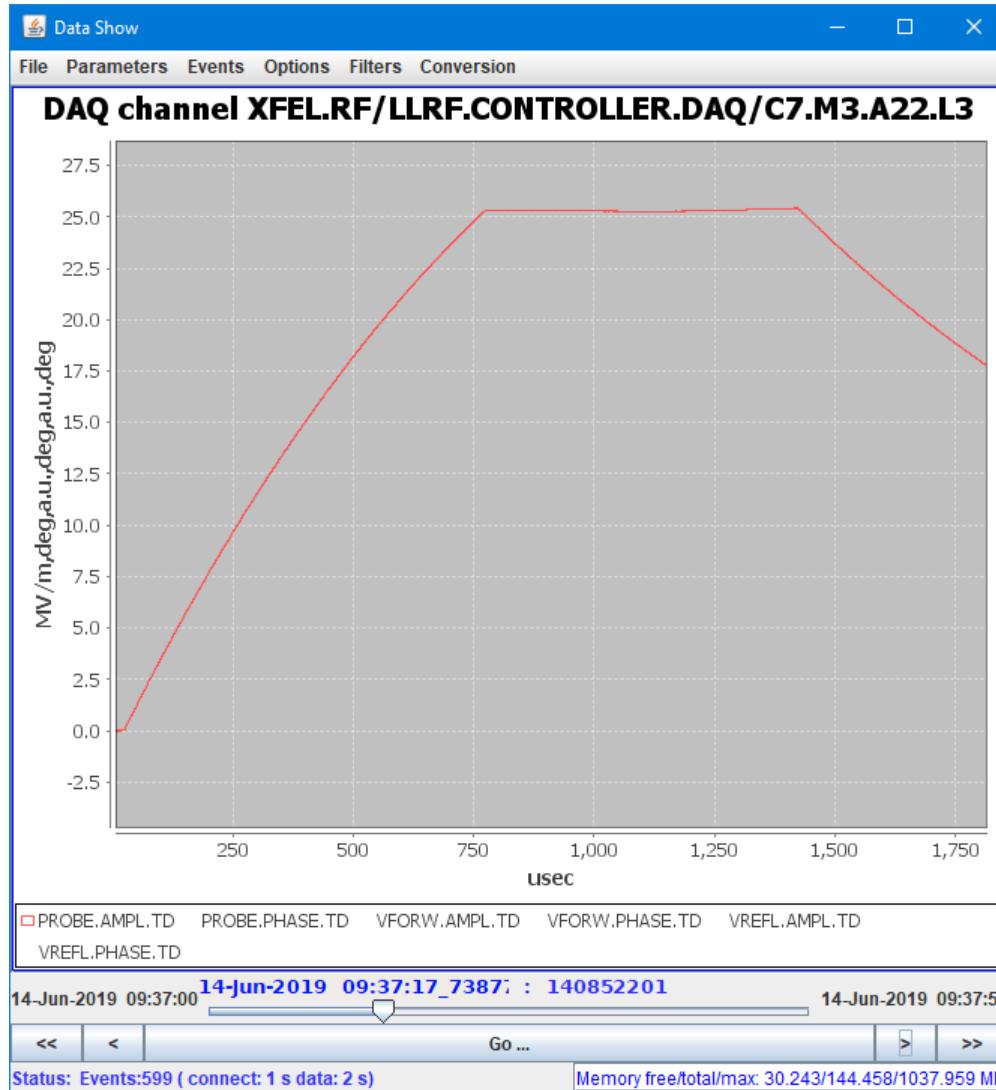
Pulse 4

$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

Gradient is slowly increased



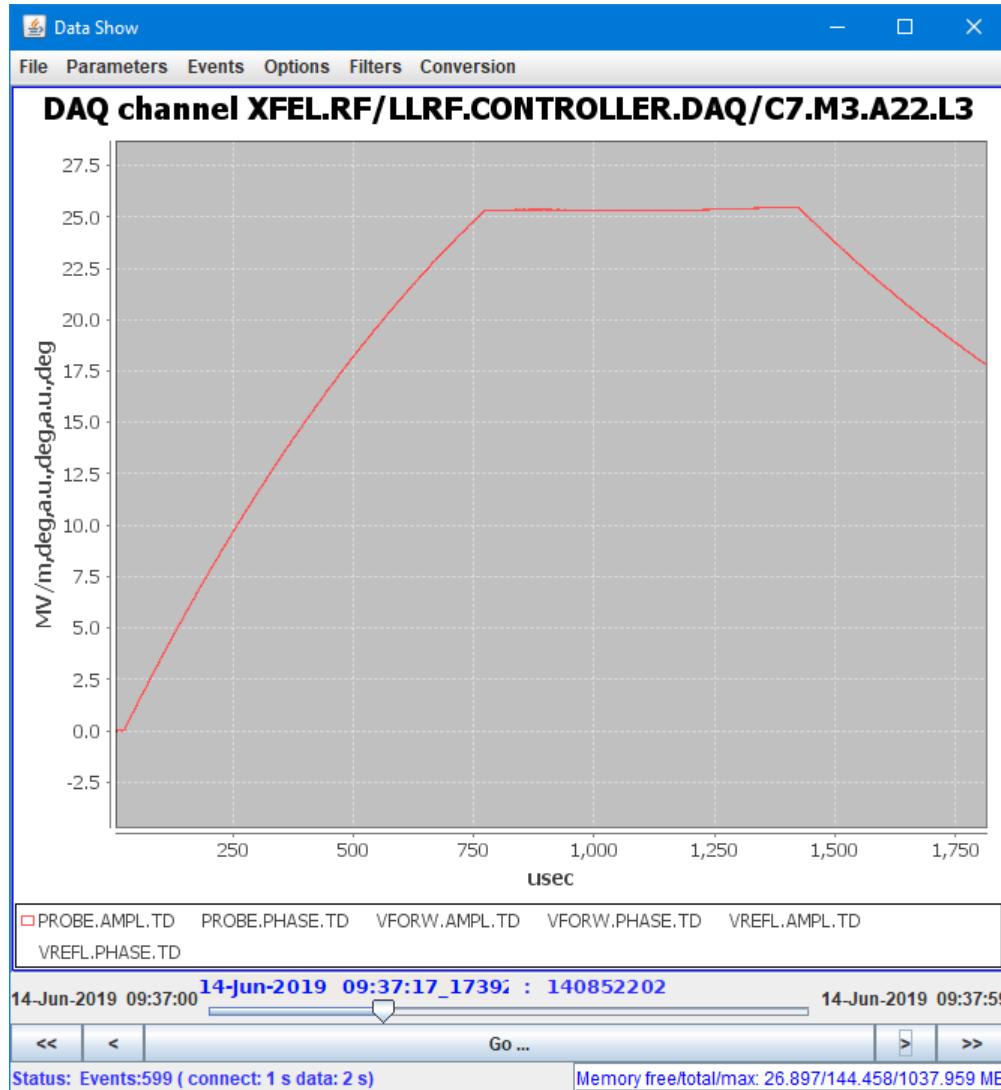
Pulse 5

$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

Gradient is slowly increased



Pulse 6

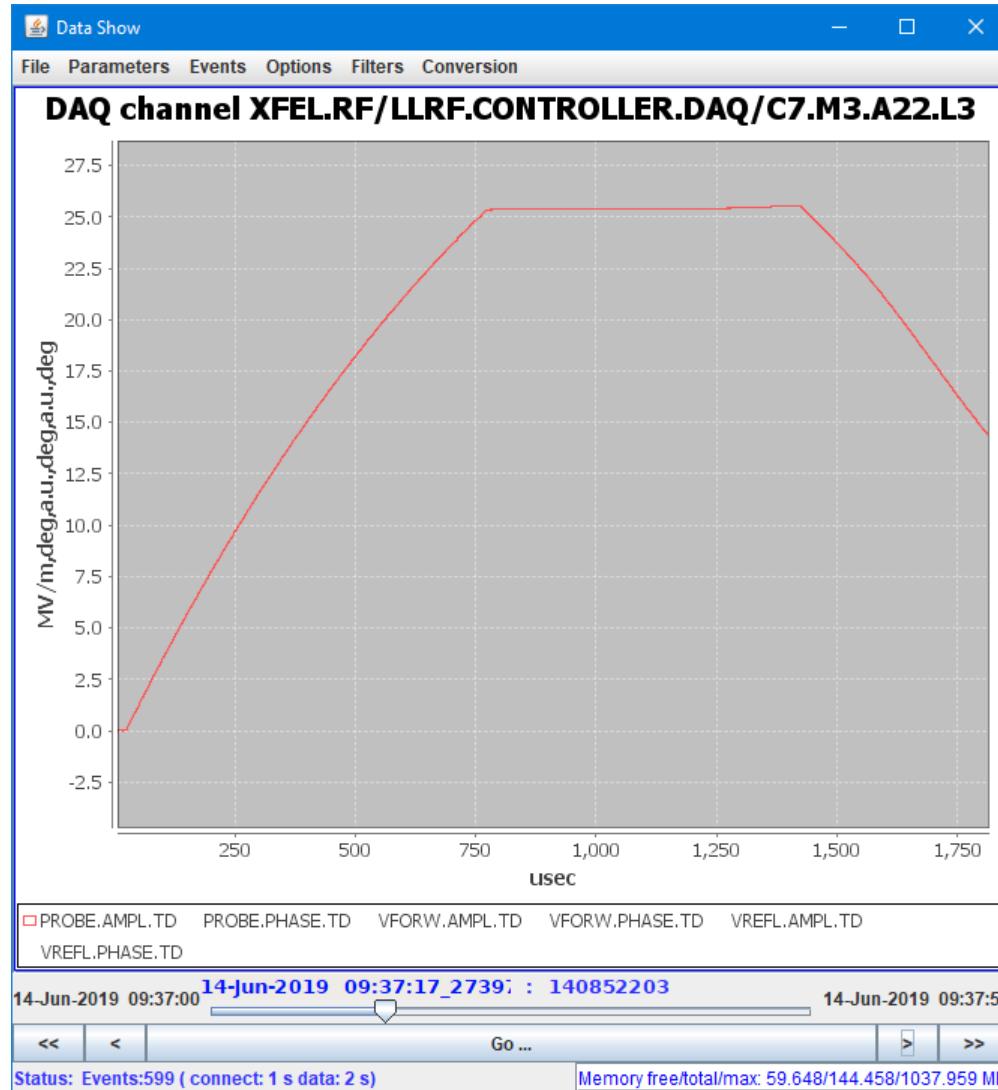
$$Q_L = 4.6 \times 10^6$$

RF operation

Experience of a cavity quench

First indication of a quench visible during the cavity decay

Q_L value drops but flat top gradient is still preserved



Pulse 7

$$Q_L = 4.3e6$$

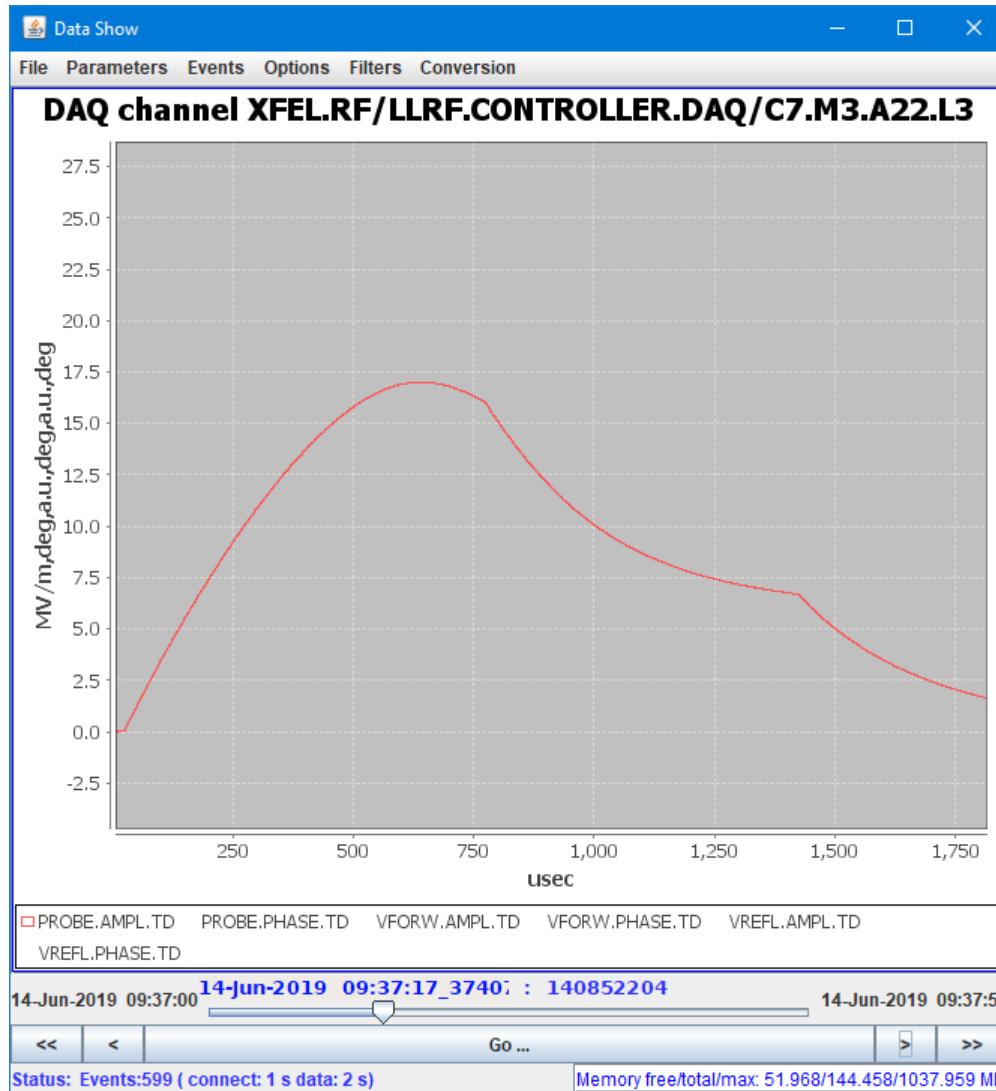
RF operation

Experience of a cavity quench

Gradient collapses: “**break down**” to normal conducting conditions

Note: this exercise is done in **open loop**

Closed loop operation would generate a sudden increase in forward power to compensate for the gradient loss
→ possible sparks at coupler



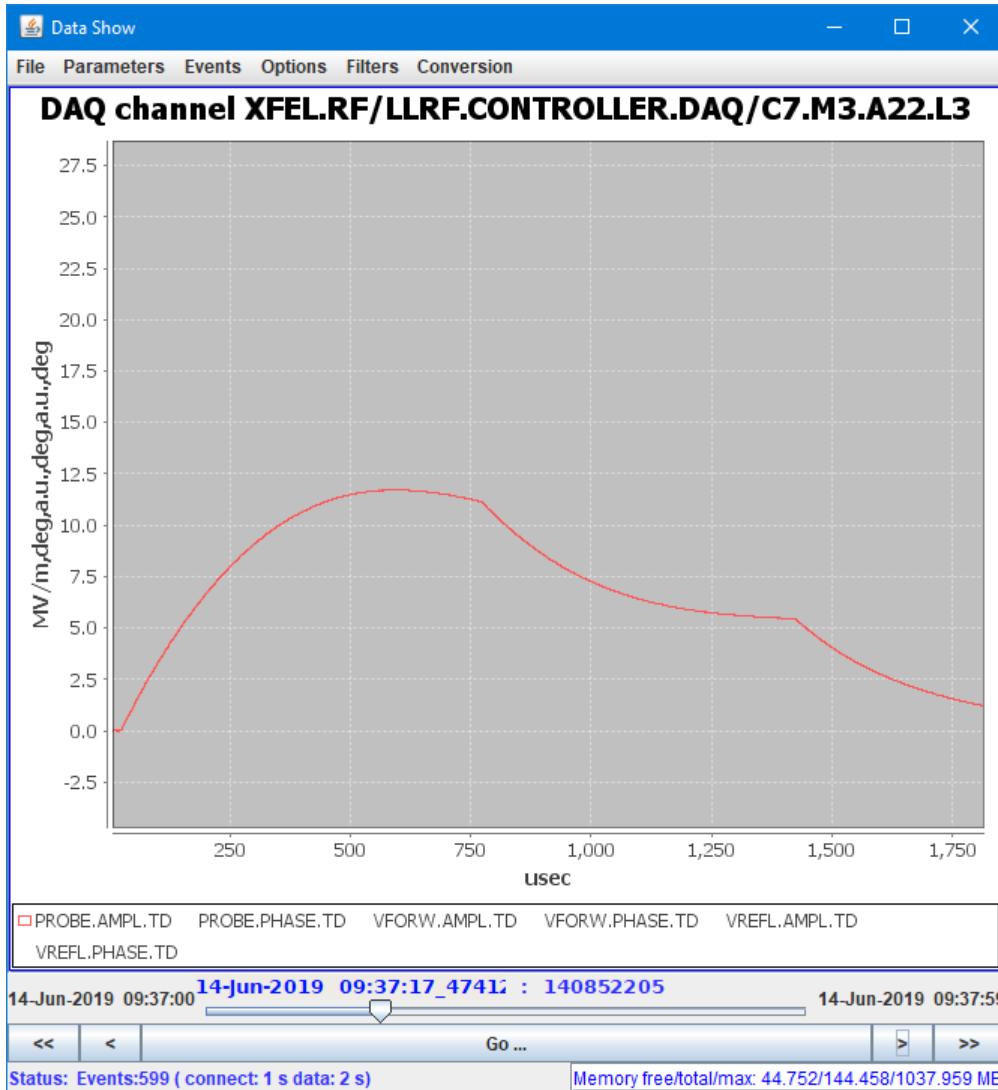
Pulse 8

$Q_L = 1.0e6$

RF operation

Experience of a cavity quench

Gradient collapses further



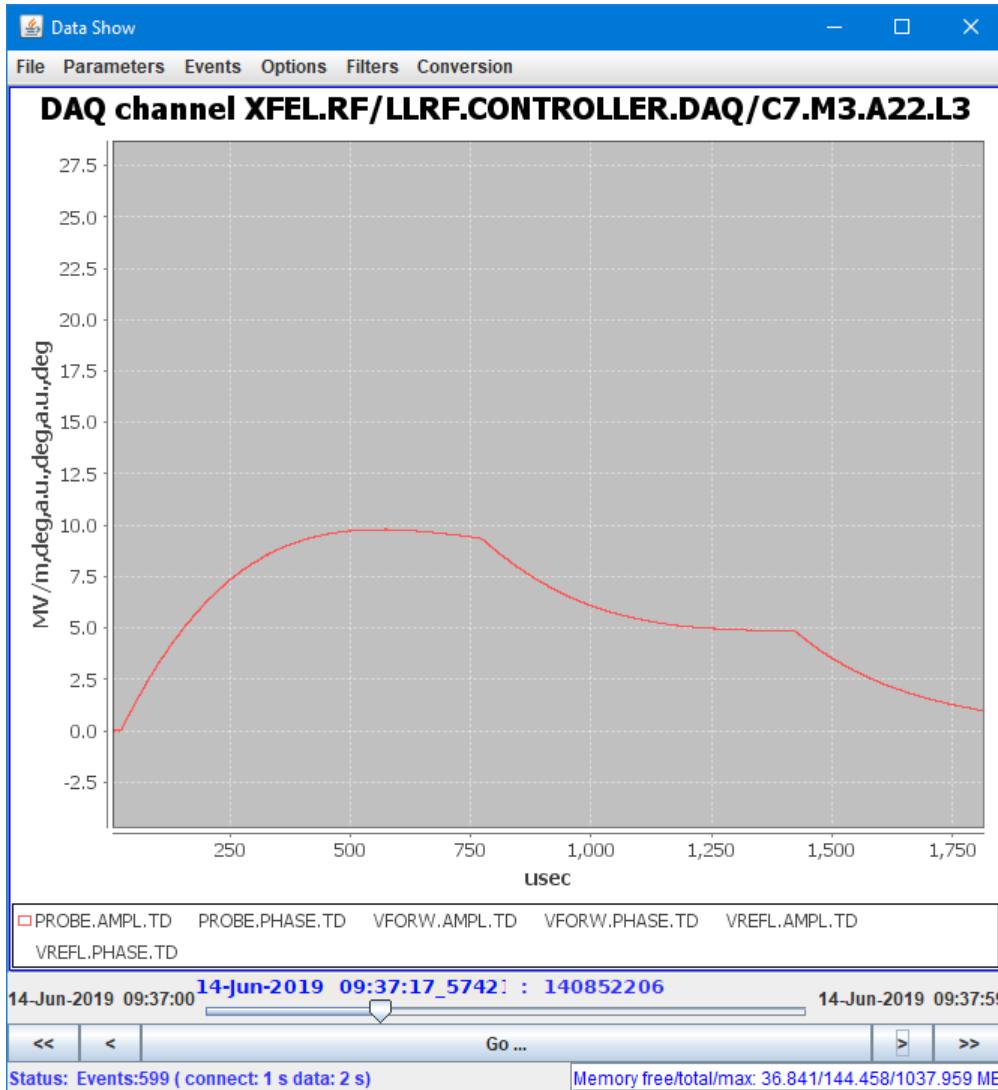
Pulse 9

$$Q_L = 1.0e6$$

RF operation

Experience of a cavity quench

Gradient collapses further



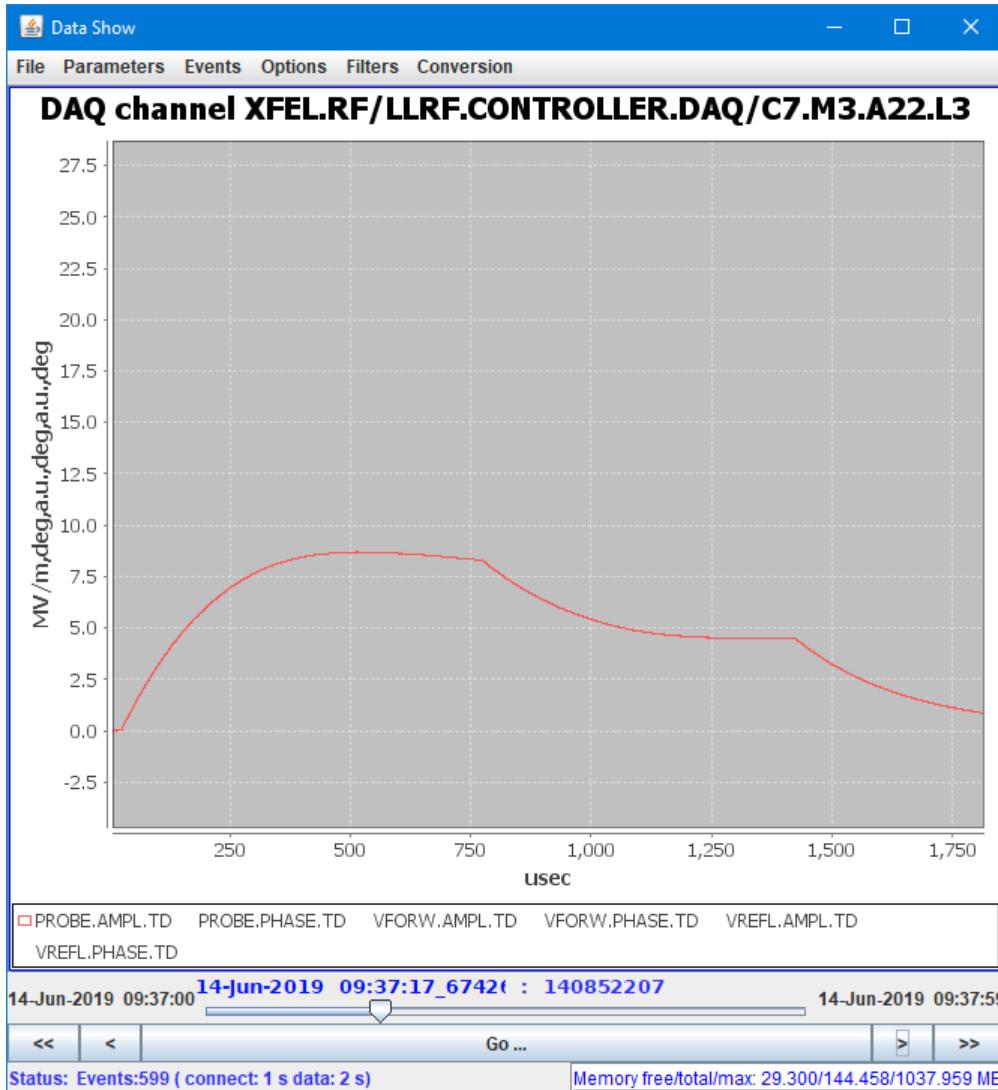
Pulse 10

$$Q_L = 1.0e6$$

RF operation

Experience of a cavity quench

Gradient collapses further



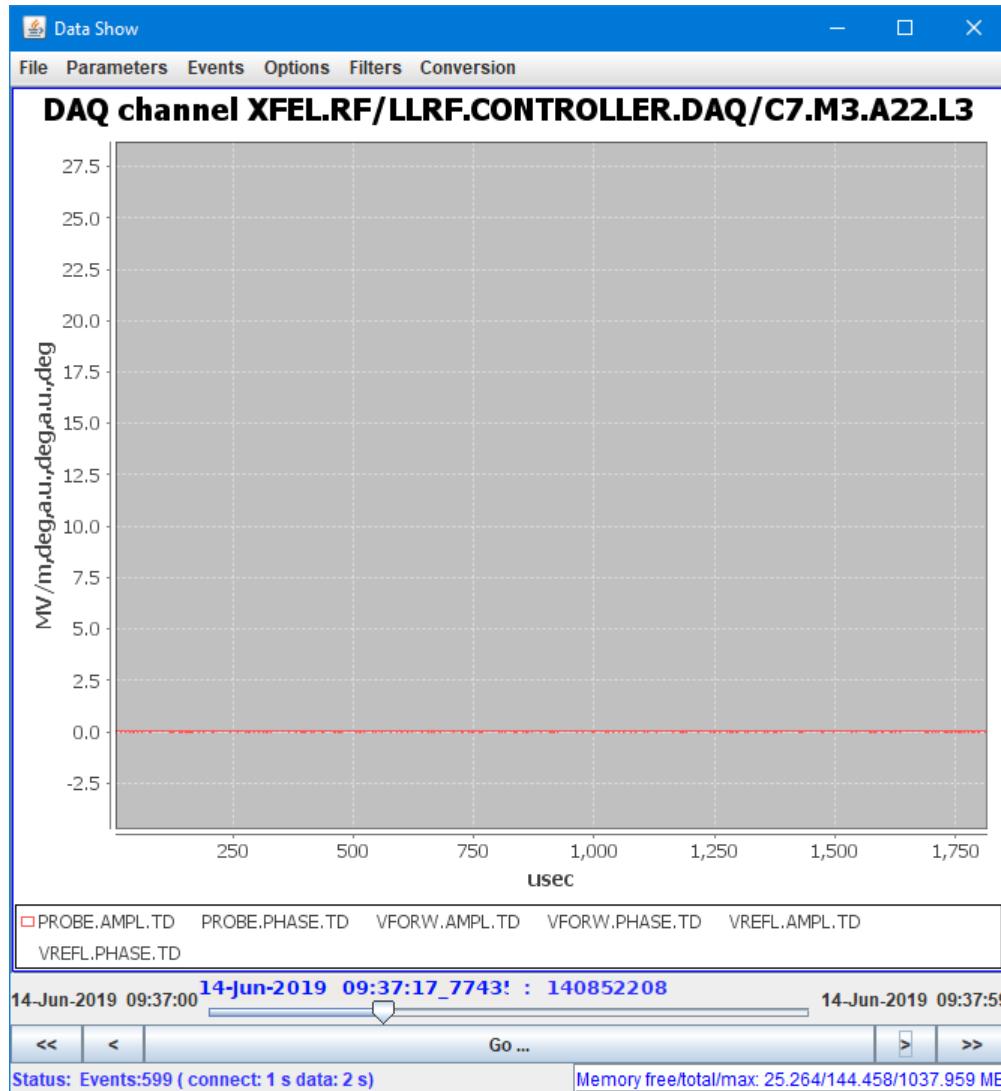
Pulse 11

$$Q_L = 1.0e6$$

RF operation

Experience of a cavity quench

RF operation is interrupted



Pulse 12

$Q_L = \dots$

RF operation

Experience of a cavity quench

- What does a **quench** mean in terms of **additional dynamic heat load** ?

- Nominal operation dynamic heat load
 $HL = 0.3 \text{ W / cavity}$
- Quench:** Q_0 drops by 4 orders of magnitude
- Instantaneous heat load increases by > 3 orders of magnitude!

$$Q_L = \left(\frac{1}{Q_{ext}} + \frac{1}{Q_0} \right)^{-1}$$

observed Q_L change due to quench:
 $4.6e6 \rightarrow 1e6$ ($= 3.6e6$)

Set to 4.6e6

Q_0 must have been in the $1.3e6$ range (from $2e10$) **→ 4 orders of magnitude change!**

Duty Factor:
1 msec @ 10 Hz = 1%

$$HL = DF \times \frac{(V_{CAV})^2}{\frac{R}{Q} Q_0}$$

$\approx 0.3 \text{ W}$

Quench → 490 W !!!

$25 \text{ MV/m} \times l_{cav} \approx 26 \text{ MV}$

Quench → 8 MV !!!

1012Ω

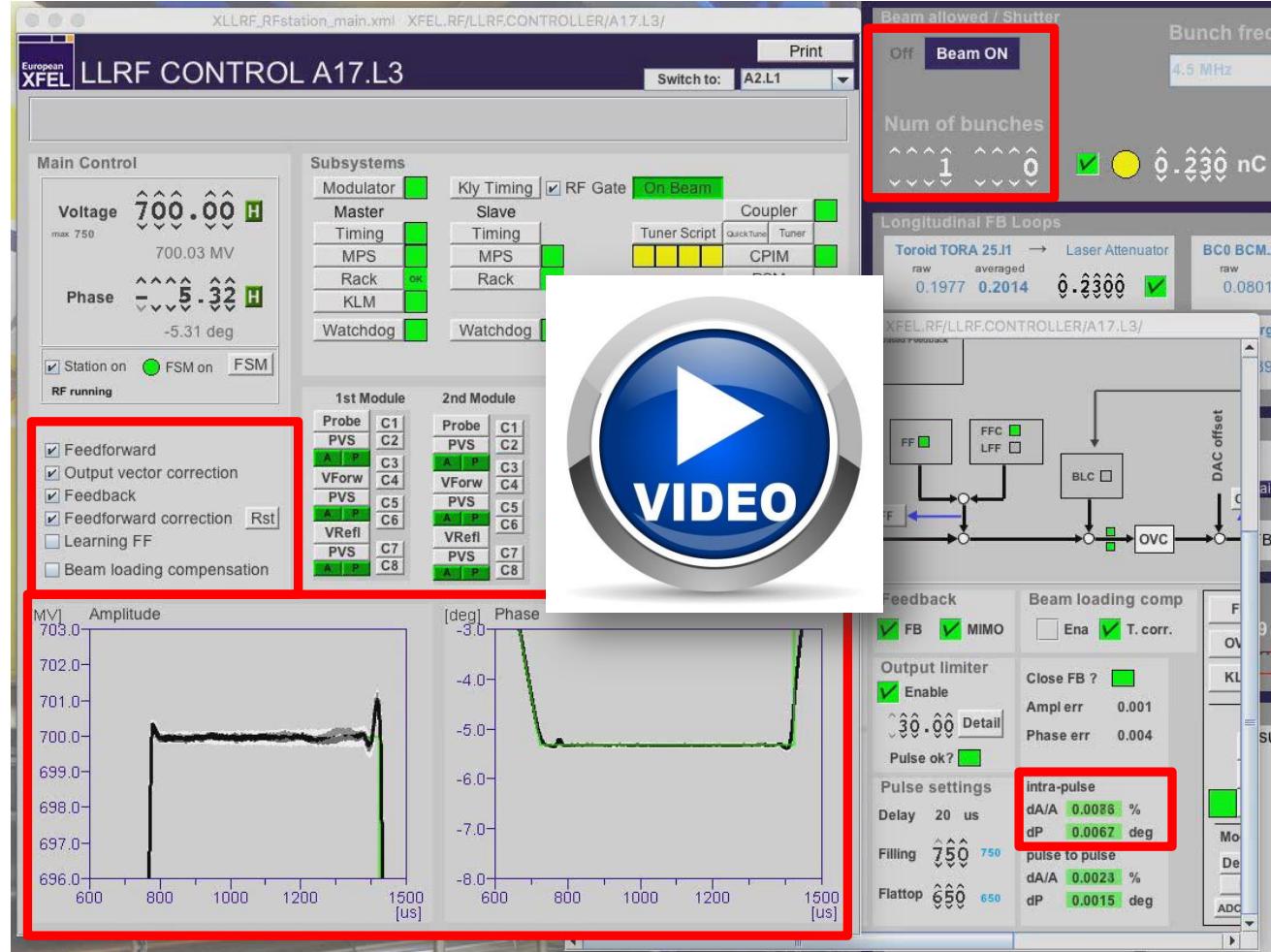
$2e10$

Quench → 1.3e6 !!!

RF operation

Beam loading compensation

- Compensates beam loading, assuring same energy gain for all electrons along bunch trains



Requirements
 $dA/A \leq 0.01\%$
 $dP \leq 0.01 \text{ deg.}$

Summary

Anybody awake ?

RF cavities fundamentals

- RLC model, envelope equation, detuning, sub harmonics, mechanical model, microphonics

RF power couplers

- Input power coupler, why and how changing Q_{ext} , impact on bandwidth, on power, heating of couplers

Frequency tuners

- Why tuning is important, slow and fast tuners, tuner figures of merit, “real world” examples, impact of Helium pressure

LLRF system

- LLRF versus HPRF, feed forward, feedback, down conversion, sampling and ADCs, IQ detection, system level description, single regulation, vector sum, pulsed and CW operation

Demo

- Simple examples using simulator

RF operations

- RF tuning, LFD compensation, quench, beam loading compensation

Thank you!

I wish everyone a fruitful conference
and a pleasant stay in Dresden.



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