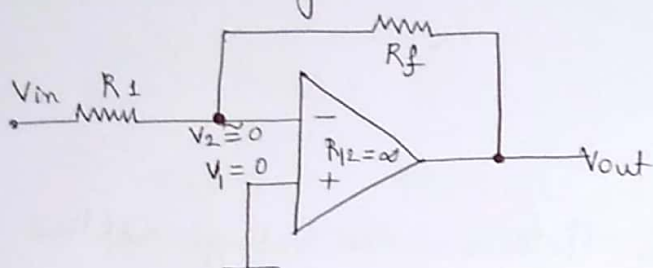


①

Virtual ground

V_{Gi} means that the voltage at that particular node is almost equal to ground voltage (0V). It is not physically connected to ground.



For ideal op-amp voltage gain = ∞ .
For real also voltage gain is high.

$$\therefore \text{Gain} = \frac{V_o}{V_{in}}$$

$$\text{Gain} = \infty, \quad V_{in} = 0.$$

$$V_{in} = V_2 - V_1$$

Here, V_1 is connected to ground, so V_2 also will be at ground potential $\therefore V_2 = 0$.

$$V_2 = \text{Virtual ground.}$$

It is used in analysis of an op-amp when negative feedback is employed.

Virtual ground

(i) Concept used in op-amp.

(ii) Voltage is approx. zero.

(iii) Not able to sink infinite current.

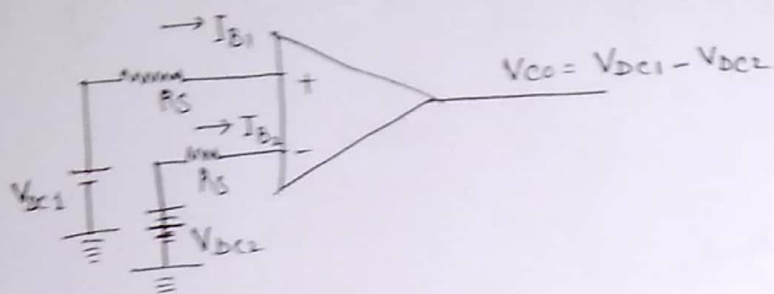
Real ground

(i) Here one terminal is connected to ground and acts as the reference point for the entire ckt.

(ii) voltage is zero.

(iii) It is an infinite current sink.

Parameters of op-amp



- ① Input offset voltage :- It is defined as the voltage applied between two input terminals of an op-amp to make the o/p zero.

V_{DC1} & V_{DC2} = DC voltages.
 R_S = source resistance

$$\boxed{V_{io} = V_{DC1} - V_{DC2}} \quad \text{It } V_{io} \text{ may be +ve or -ve.}$$

For 741C $\rightarrow \pm 6 \text{ mV}$

- ② Input offset current :-

I_{io} is the difference between the currents into inverting & non inverting terminals of a balanced amplifier.

$$\boxed{I_{io} = |I_{B1} - I_{B2}|}$$

741C $\rightarrow 200 \text{ nA}$ (max)
 $\rightarrow 6 \text{ nA}$ (precision amplifier)

- ③ Input bias current :-

I_B is the average of the current entering the input terminals of a balanced amplifier.

$$\boxed{I_B = \frac{I_{B1} + I_{B2}}{2}}$$

741C $I_{B(\text{max})} = 700 \text{ nA}$
 precision $= \pm 7 \text{ nA}$

(3)

(4) CMRR :-

It is the ratio of the differential voltage gain A_d to the common mode voltage gain A_{CM} .

$$CMRR = \frac{A_d}{A_{CM}}$$

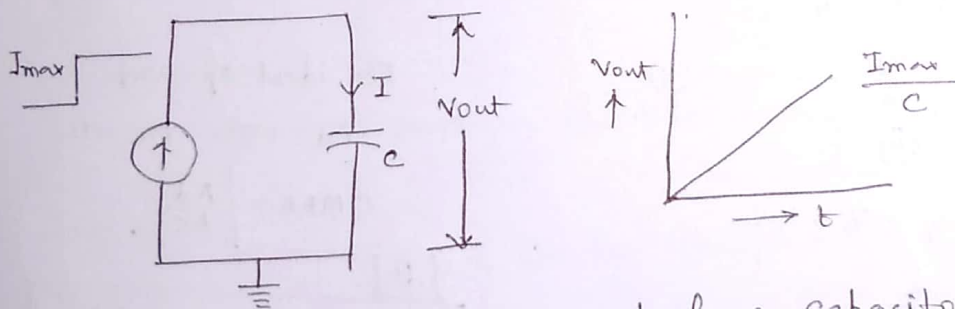
741C \rightarrow 90 dB.

Higher CMRR, better the matching b/w two input terminals and smaller is the output common mode voltage.

(5) Slew rate :-

It is defined as the max rate of change of o/p voltage per unit time under large signal voltage condition and expressed as V/ μ s.

$$SR = \left. \frac{dv_o}{dt} \right|_{\max} \text{ V}/\mu\text{s}$$



charging current of a capacitor $\cdot I = C \frac{dv}{dt}$
 $\frac{dv}{dt} = I/C$

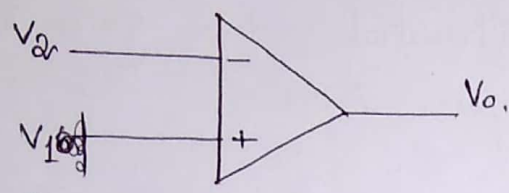
If I move capacitor charges quickly. If I is limited to I_{max} , ~~that~~ rate of change is also limited.

Slew rate indicates how rapidly the output of an op-amp can ~~the~~ change in response to changes in the i/p freq with input amplitude const.

Slew rate changes with change in voltage gain and normally specified at unity gain.

$$741C \rightarrow 0.5 \text{ V}/\mu\text{s}.$$

CMRR.



op-amp. amplifies the difference of input voltages \$V_1\$ and \$V_2\$ applied to the non inverting and inverting i/p terminals w.r to ground. The o/p voltage \$V_o\$ is measured w.r to ground.

At output difference signal $V_d = V_1 - V_2$ and common mode signal.

$$V_c = \frac{V_1 + V_2}{2} \quad \text{--- (2) both are amplified.}$$

As op-amp is a linear device, so output voltage

$$V_o = A_1 V_1 + A_2 V_2 \quad \text{--- (3)}$$

where, \$A_2, A_1\$ = voltage gain when input terminals are grounded.

Adding (i) and (ii)

$$V_d + 2V_c = V_1 - V_2 + V_1 + V_2.$$

$$\therefore V_d + 2V_c = 2V_1$$

$$\text{or, } V_1 = V_c + \frac{V_d}{2} \quad \text{--- (4)}$$

Subtracting (i) from (ii)

$$2V_c - V_d = V_1 + V_2 - V_1 + V_2$$

$$\text{or, } 2V_c - V_d = 2V_2$$

$$\therefore V_2 = V_c - \frac{V_d}{2} \quad \text{--- (5)}$$

From (3) we get,

$$V_o = A_1 \left(V_c + \frac{V_d}{2} \right) + A_2 \left(V_c - \frac{V_d}{2} \right)$$

$$\text{or, } V_o = V_c (A_1 + A_2) + \frac{V_d}{2} (A_1 - A_2)$$

$$= A_c V_c + A_d V_d.$$

where, \$A_c\$ = common mode gain = \$(A_1 + A_2)\$
 \$A_d\$ = differential mode gain = \$\frac{1}{2}(A_1 - A_2)\$

For ideal op-amp.

$$A_d \rightarrow \infty, A_c = 0.$$

$$\therefore \text{CMRR} = \left| \frac{A_d}{A_c} \right|.$$

$$\text{dB} \quad \text{CMRR} = 20 \log_{10} \left| \frac{A_d}{A_c} \right|$$

op-amp characteristics

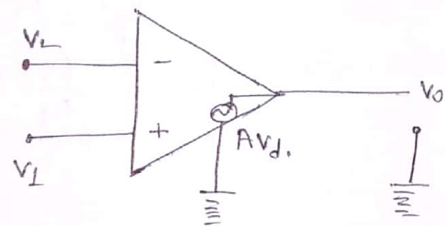
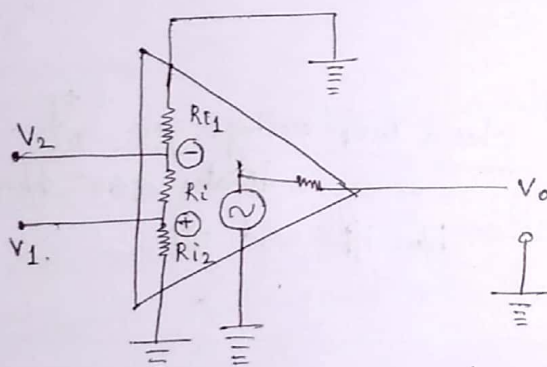
Ideal op-amp

- (i) Voltage gain $= \infty$
- (ii) Input impedance $= \infty$
- (iii) Output impedance $= 0$.
- (iv) Bandwidth $= \infty$
- (v) Perfect balance, when ~~iff~~ both i/p voltages are equal, $V_o = 0$.
- (vi) Characteristics do not drift with temp.

Practical op-amp

- (i) Voltage gain 10^3 to 10^6
- (ii) Input impedance $= 150 \text{ k}\Omega$ to $100 \text{ M}\Omega$
- (iii) 0.75 to 100Ω
- (iv) finite $\rightarrow 100 \text{ k}\Omega$
- (v) do not have perfect balance.
- (vi) Characteristics drifted with temp.

AC equivalent ckt of op-amp



R_{i1} = i/p resistance b/w ~~inter~~ inverting & ground.

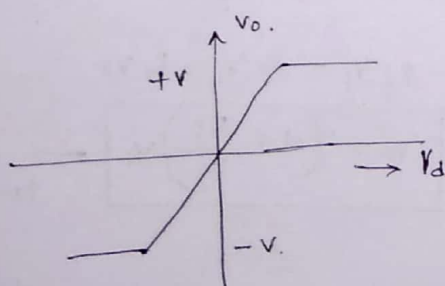
R_{i2} = i/p " " " non " " "

R_i = Resistance b/w two i/p terminals.

Ideal R_{i1} , R_{i2} & $R_i \rightarrow \infty$.

The diff of i/p voltage $V_d = (V_1 - V_2)$ is amplified at the o/p and taken through o/p resistance R_o which is $R_o \approx 0$.

open loop characteristics



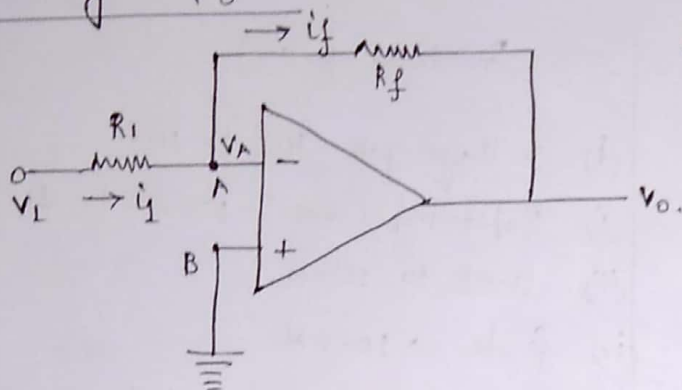
A_d is very high. $A = \frac{V_o}{V_d}$.

V_d slightly +ve, V_o goes to nearly +V.
If $V_d \uparrow$, V_o remains saturated.

V_d slightly -ve, V_o attains nearly -V
and remains saturated.

-V and +V is called +ve & -ve saturated voltages.

(6)

Inverting amplifier

R_1 & R_f external resistances at i/p and feedback path.

V_1 is i/p voltage applied to inverting terminal through R_1

\neq \oplus +ve is grounded.

\therefore -ve is virtually grounded.

$\therefore V_A = 0$.

Again i/p impedance is very high ($\approx \infty$) so no current enters into the op-amp.

At point A, applying Kirchhoff's current law (KCL)

$$i_1 = i_f$$

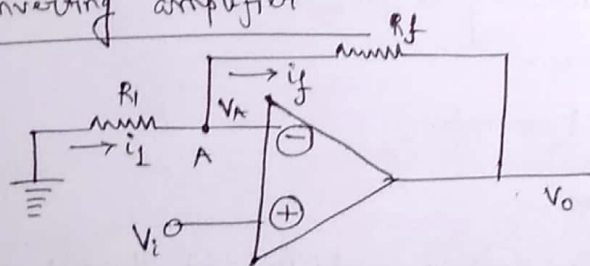
$$\text{or } \frac{V_1 - 0}{R_1} = \frac{0 - V_0}{R_f}$$

$$\text{or } -\frac{V_0}{R_f} = \frac{V_1}{R_1}$$

$$\therefore \boxed{V_0 = -\frac{R_f}{R_1} V_1}$$

closed loop voltage gain $\frac{R_f}{R_1}$.

-ve sign indicates $\pm 180^\circ$ phase shift b/w i/p and o/p.

Non inverting amplifier

Input applied to +ve terminal, R_f provides negative feedback.

Voltage gain of op-amp = ∞ .

KCL at A

$$i_1 = i_f$$

$$\text{or } \frac{0 - V_1}{R_1} = \frac{V_1 - V_0}{R_f}$$

$$\text{or } -R_f V_1 = R_1 V_1 - R_1 V_0$$

$$\text{or } \boxed{V_0 = \left(1 + \frac{R_f}{R_1}\right) V_1} \rightarrow \text{+ve & -ve terminal phase shift } 0^\circ$$

Operational Amplifier :- (OP AMP)

OPAMP is general purpose linear integrated circuit. It was developed by Robert Widlar in 1964. It is a direct coupled, high gain, differential input amplifier.

The term operational signifies that various mathematical operations such as addition, subtraction, integration, differentiation can be performed by using op-amp.

It can be used in →

waveform forming

active filters

oscillators

A/D and D/A converters.

Advantages :-

→ Here -ve feedback is applied so performance of the op-amp with negative feedback is controlled by the feedback elements independent of the characteristics of the transistors and other elements that constitute of the ~~amp~~ op-amp.

→ Feedback elements are usually passive so ckt operation is very stable and predictable.

→ IC of ~~op~~ op-amp is inexpensive and have temperature stabilization.

Block diagram of OP-AMP :-