



Principles of Digital Signal Processing

Dr. B. Niranjana Krupa

Department of Electronics and Communication.

DSP



Discrete Fourier Transform

Dr. B. Niranjana Krupa

Department of Electronics and Communication.

Recovering $x(n)$:

When $x(n)$ has infinite duration, the equally spaced frequency samples do not represent $x(n)$.

They (frequency samples) only correspond to a periodic sequence $x_p(n)$, the aliased version of $x(n)$, as in

$$x_p(n) = \sum_{l=-\infty}^{\infty} x(n - lN)$$

When $x(n)$ has finite duration of length $L \leq N$,
then

$$x_p(n) = \sum_{l=-\infty}^{\infty} x(n - lN)$$

$$x_p(n) = \begin{cases} x(n), & 0 \leq n \leq L - 1 \\ 0, & L \leq n \leq N - 1 \end{cases}$$

Frequency domain sampling

DFT

It is important to note that zero padding doesn't provide any additional information about the spectrum of the sequence. However padding zeros and computing N point DFT only results in better display.

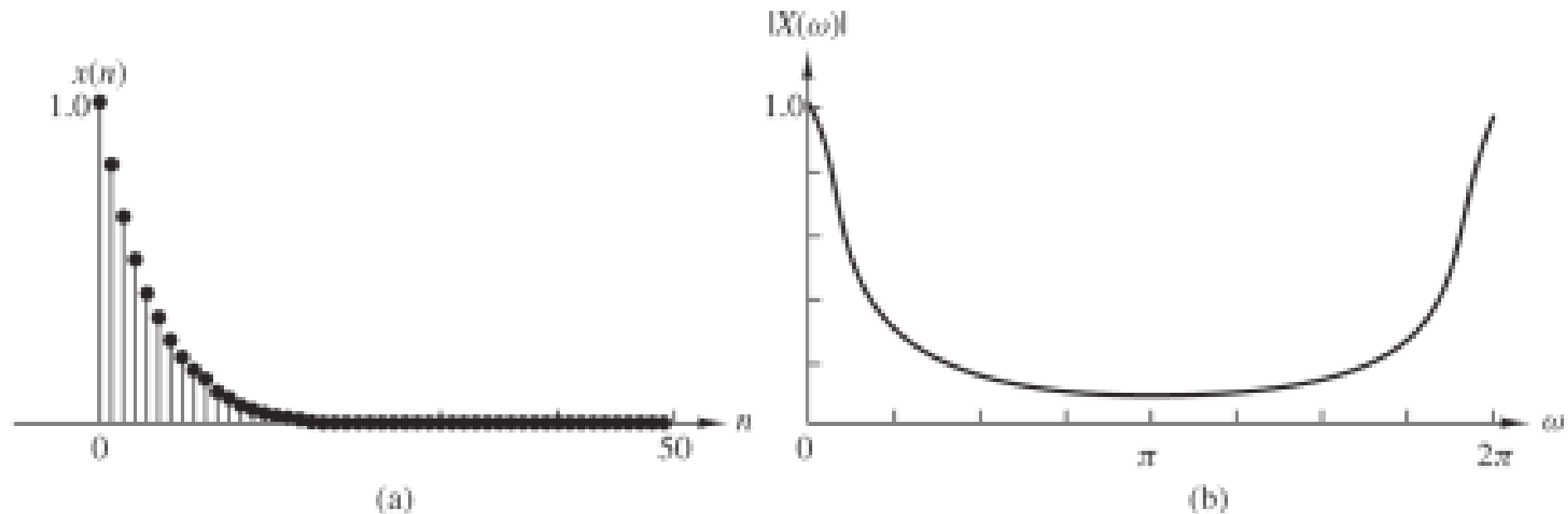


Figure 7.1.4 (a) Plot of sequence $x(n) = (0.8)^n u(n)$; (b) its Fourier transform

Frequency domain sampling

DFT

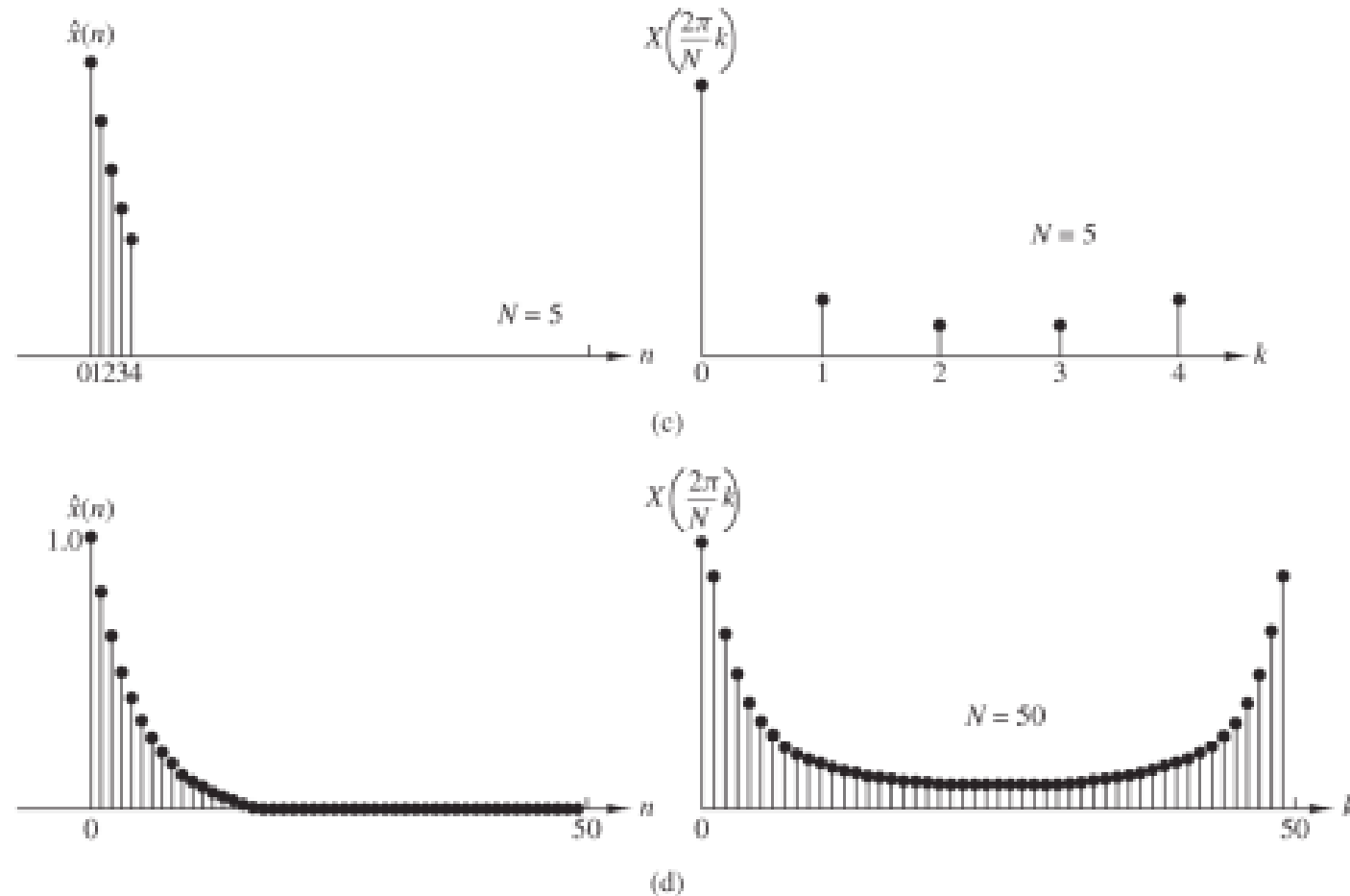


Figure 7.1.4 (a) Plot of sequence $x(n) = (0.8)^n u(n)$; (b) its Fourier transform (magnitude only); (c) effect of aliasing with $N = 5$; (d) reduced effect of aliasing with $N = 50$.

Frequency domain sampling

DFT

In summary, a finite-duration sequence $x(n)$ of length L [i.e., $x(n) = 0$ for $n < 0$ and $n \geq L$] has a Fourier transform

$$X(\omega) = \sum_{n=0}^{L-1} x(n)e^{-j\omega n} \quad 0 \leq \omega \leq 2\pi$$

where the upper and lower indices in the summation reflect the fact that $x(n) = 0$ outside the range $0 \leq n \leq L-1$. When we sample $X(\omega)$ at equally spaced frequencies $\omega_k = 2\pi k/N$, $k = 0, 1, 2, \dots, N-1$, where $N \geq L$, the resultant samples are

$$X(k) \equiv X\left(\frac{2\pi k}{N}\right) = \sum_{n=0}^{L-1} x(n)e^{-j2\pi kn/N}$$

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, 2, \dots, N-1$$

Frequency domain sampling

DFT

To summarise DFT and IDFT

$$\text{DFT: } X(k) = \sum_{n=0}^{N-1} x(n) e^{-j2\pi kn/N}, \quad k = 0, 1, 2, \dots, N-1$$

$$\text{IDFT: } x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j2\pi kn/N}, \quad n = 0, 1, 2, \dots, N-1$$

Frequency domain sampling

DFT as linear transform

DFT and IDFT expressed as

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn} \quad k = 0, 1, \dots, N-1$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} \quad n = 0, 1, \dots, N-1$$

Where, $W_N = e^{-j2\pi/N}$

The N th root of unity

- Computational complexity of DFT
 - Each point: N complex multiplications and $(N-1)$ complex additions
 - Hence, N -point DFT:
 - $N*N$ complex multiplications and
 - $N*(N-1)$ complex additions
- DFT and IDFT as linear transformations on $x(n)$ and $X(k)$
 - Consider N -point vectors x_N of $x(n)$ and X_N of $X(k)$ and an $N*N$ matrix W_N as:

$$\mathbf{x}_N = \begin{bmatrix} x(0) \\ x(1) \\ \vdots \\ x(N-1) \end{bmatrix}, \quad \mathbf{X}_N = \begin{bmatrix} X(0) \\ X(1) \\ \vdots \\ X(N-1) \end{bmatrix}$$

Frequency domain sampling

DFT as linear transform

$$\mathbf{W}_N = \begin{bmatrix} 1 & 1 & 1 \dots & 1 \\ 1 & W_N & W_N^2 & \dots & W_N^{N-1} \\ & W_N^2 & W_N^4 & \dots & W_N^{2(N-1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 1 & W_N^{N-1} & W_N^{2(N-1)} & \dots & W_N^{(N-1)(N-1)} \end{bmatrix}$$

N-point DFT expressed in matrix form as:

$$\mathbf{X}_N = \mathbf{W}_N \mathbf{x}_N$$

W_N is the matrix of linear transformation and is a symmetric matrix

If we assume that the inverse of W_N exists then,

$$\text{IDFT} \rightarrow \mathbf{x}_N = \mathbf{W}_N^{-1} \mathbf{X}_N \qquad \mathbf{x}_N = \frac{1}{N} \mathbf{W}_N^* \mathbf{X}_N$$

Frequency domain sampling

DFT as linear transform

Comparing

$$\mathbf{W}_N^{-1} = \frac{1}{N} \mathbf{W}_N^*$$

$$\mathbf{W}_N \mathbf{W}_N^* = N \mathbf{I}_N$$

\mathbf{W}_N is an orthogonal (unitary) matrix

And its inverse exists, given as shown above



THANK YOU

Dr. B. Niranjana Krupa

Department of Electronics and Communication

bnkrupa@pes.edu

+91 80 6666 3333 Ext 777