(12)

Dividing numerator and denominator by (Ro+Rz), we get

$$R_{of}' = \frac{\frac{R_{o}R_{L}(1+\beta G_{m})}{R_{o}+R_{L}}}{1+\frac{\beta G_{m}R_{o}}{R_{o}+R_{L}}} = R_{o}' \cdot \frac{1+\beta G_{m}}{1+\beta G_{m}}$$
where 
$$R_{o}' = \frac{R_{o}R_{L}}{R_{o}+R_{L}}$$
 and 
$$G_{m} = \frac{G_{m}R_{o}}{R_{o}+R_{L}}$$

## 3. Current Shunt Feedback

The current shunt feedback topology is shown in figure below with the amplifier input and output circuits replaced by its Norton's model.

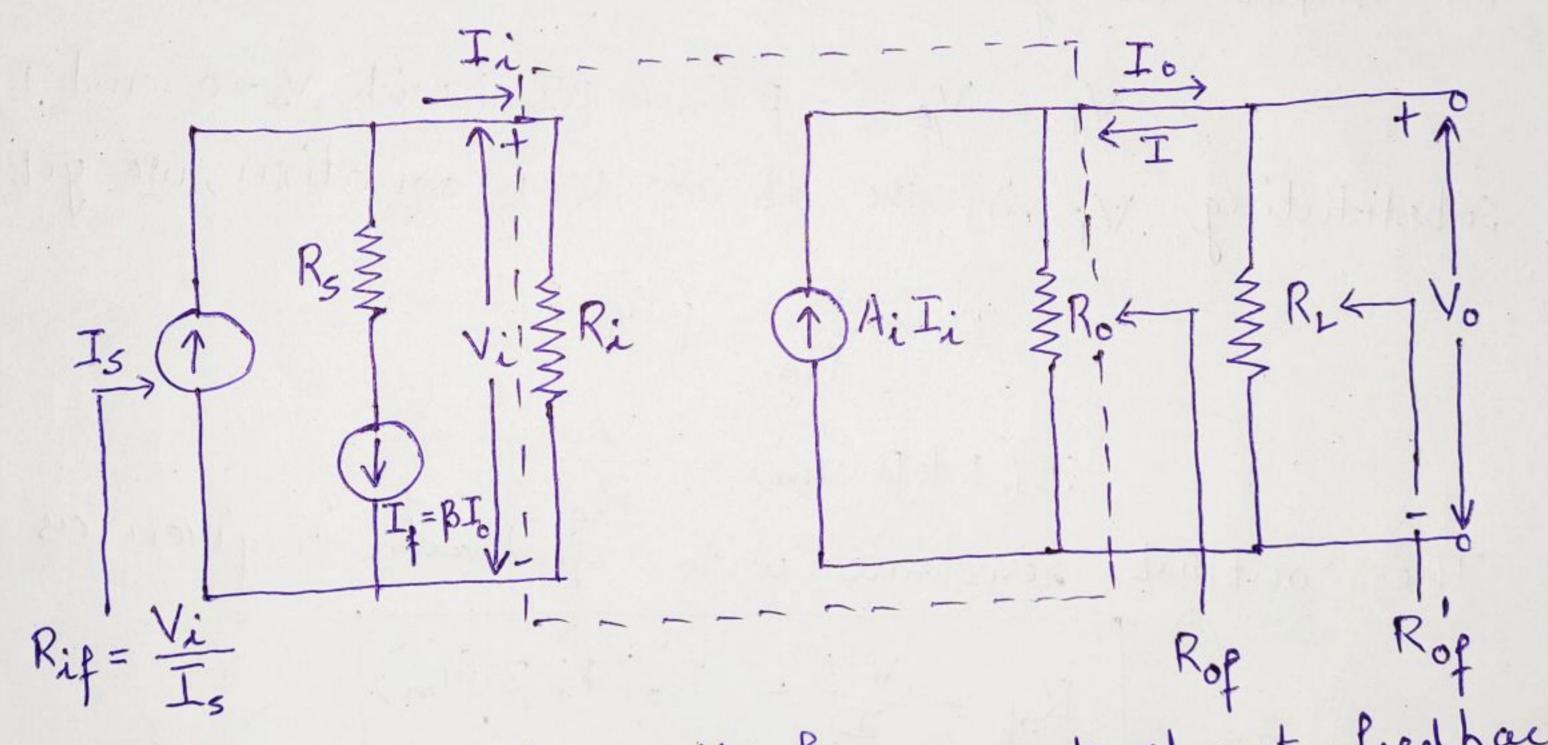


Fig. Equivalent circuit for current shunt fieldback Applying KCL to the input side, we get  $I_s = I_i + I_f = I_i + \beta I_o$ 

The output current is written as  $I_o = \frac{A_i I_i R_o}{R_o + R_L} = A_I I_i$ 

where  $A_{I} = \frac{I_{o}}{T_{i}} = \frac{A_{i}R_{o}}{R_{o}+R_{i}}$ 

Substituting the value of Io in the above KCL equation, (13)

Is = Ii + BAI Ii = (1+ BAI) Ii
The input resistance with feedback is given as,

$$Rif = \frac{Vi}{I_s} = \frac{Vi}{(1+\beta A_I)I_i} = \frac{Ri}{1+\beta A_I}$$

where Ai -> short circuit ament gain without feedback AI -> current gains without feedback taking the load R, into account.

... 
$$Ai = \lim_{R_L \to 0} A_I$$

For finding Rof, R<sub>L</sub> is disconnected (i.e R<sub>L</sub>=0), the enternal source signal is made zero (i.e set Is=0) and to is replaced with V.

Applying KCL to the output node, we get

$$I = \frac{V}{R_o} - A_i I_i$$

The input current is written as

Ii = - Ip = -BIo = +BI (with Is=0 and I=-Io)

Substituting Ii in the above KCL equation, we get

on I(I+BAi) = V

The output resistance with feedback is given as Rof = Y = Ro(1+BAi)

where Ai is short-circuit current gain without taking R2 into account.

The output resistance with feedback Rof including (4)
RL as part of the amplifier is given by

Therefore 
$$R_{of}^{\dagger} = \frac{R_{of}R_{L}}{R_{of}+R_{L}} = \frac{R_{o}(1+\beta Ai)R_{L}}{R_{o}(1+\beta Ai)+R_{L}}$$

$$Rof = \frac{R_0 R_L (1+BAi)}{R_0 + R_L + BAi R_0}$$

Dividing nume rator and denominator by (Ro+RL), we get 0.0 (1+BAi)

$$R_{of} = \frac{R_{o}R_{L}(1+\beta Ai)}{R_{o}+R_{L}} = R_{o} \frac{1+\beta Ai}{1+\beta A_{I}}$$

$$R_{of} = \frac{R_{o}R_{L}(1+\beta Ai)}{R_{o}+R_{L}} = R_{o} \frac{1+\beta Ai}{1+\beta A_{I}}$$

$$R_{o} = \frac{R_{o}R_{L}(1+\beta Ai)}{R_{o}+R_{L}} = R_{o} \frac{1+\beta Ai}{1+\beta A_{I}}$$

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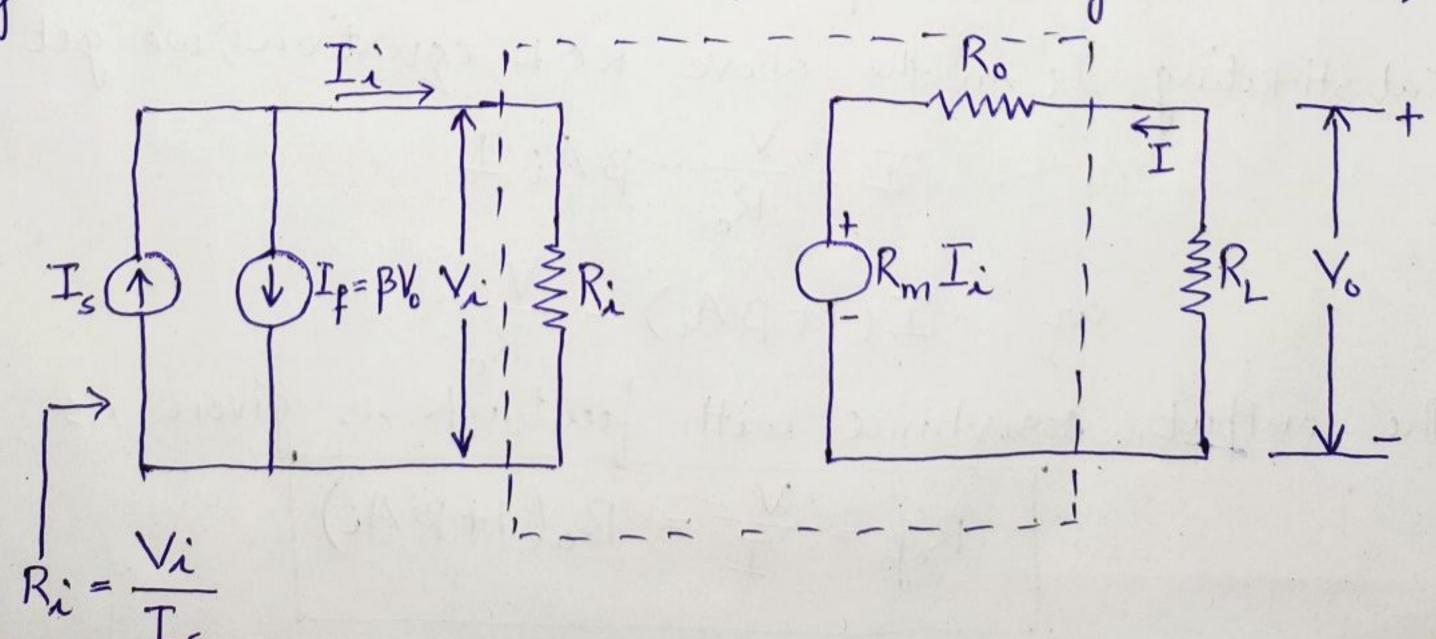
$$R_{o} = \frac{R_{o}R_{L}(1+\beta Ai)}{R_{o}+R_{L}} = R_{o} \frac{1+\beta Ai}{1+\beta A_{I}}$$

$$R_{o} = \frac{R_{o}R_{L}(1+\beta Ai)}{R_{o}+R_{L}} = R_{o} \frac{1+\beta Ai}{1+\beta A_{I}}$$

where 
$$R_0' = \frac{R_0 R_L}{R_0 + R_L}$$
 and  $A_I = \frac{A_i R_0}{R_0 + R_L}$ 

4. Voltage Shunt Feedback

The voltage shunt feedback to pology is shown in figure below with the amplifier input circuit represented by Norton's model and output circuit by The venis's equivalent.



4 .

Applying KCL to the input side, we get

The output voltage is written as

where 
$$R_{M} = \frac{V_0}{T_i} = \frac{R_m R_L}{R_0 + R_L}$$

Substituting the value of Vo in the above KCL equation, we get

Is = Ii + B Rm Ii = (1+BRm) Ii

The input resistance with feedback is given as

$$Rif = \frac{Vi}{I_s} = \frac{Vi}{(1+\beta R_M)Ii} = \frac{Ri}{1+\beta R_M}$$

where Rm represents the open circuit translistance without feedback and Rm is the transresistance without feedback taking the load RL into account. There fore

$$R_m = \lim_{R_L \to \infty} R_M$$

Now, for finding Rof,  $R_L$  is disconnected (i.e  $R_L = \infty$ ), the enternal source signal is made zero (i.e set  $T_S = 0$ ) and  $V_0$  is replaced with V.

Applying KVL to the output side, we get  $I = \frac{V - R_m V_i}{2}$ 

The imput current is written as

Substituting Ii in the above KVL equation, we get

$$I = \frac{V + \beta R_m V}{R_o} = \frac{V(1 + \beta R_m)}{R_o}$$

The output resistance with feedback is given as

$$Rof = \frac{V}{I} = \frac{Ro}{1+\beta Rm}$$

where Rm represents the open circuit transperistance without taking the load R\_ into auount.

The output resistance with feedback Rof including R<sub>2</sub> as part of the amplifier is given by

Therefore, 
$$R_{of} = \frac{R_{of} || R_L}{R_{of} + R_L} = \frac{\frac{R_o}{1 + \beta R_m} \cdot R_L}{\frac{R_o}{1 + \beta R_m} + R_L}$$

Ro + RL + BRm RL emd denominator by (Ro + RL), Dividing nume rator we get

$$Rof = \frac{R_0 R_L / (R_0 + R_L)}{1 + \left[ \beta R_m R_L / (R_0 + R_L) \right]}$$

or 
$$Rop' = \frac{Ro'}{1+\beta R_M}$$

where 
$$R_0' = \frac{R_0 R_L}{R_0 + R_L}$$
 and  $R_M = \frac{R_m R_L}{R_0 + R_L}$ 

indicates the open-circuit transresistance taking the load R<sub>L</sub> into account

Due to non-linear characteristics of an active device, it is not possible to construct an ideal amplifier. Therefore, the output of an amplifier will differ from the input either in its waveform or frequency content. The difference between the output waveform and the input waveform in an amplifier is called distortion.

#### Types of Distortion

- 1. Harmonic Distortion
- 2. Frequency Distortion
- 3. Phase or Delay Distortion

### 1. Harmonie Distortion

In this type of distortion, the new frequencies are produced in the output, which are not present in the imput signal. This harmonic distortion is sometimes called amplitude distortion.

\* The intermodulation distortion is also a type of non-linear distortion which occurs when the input signal consists of more than one frequency. If an input signal contains two frequencies franch of the sum (fift) and difference (fi-fi)

are called intermodulation frequencies.

## 2. Frequency Distortion

In this type of distortion, the signal components at different frequencies are amplified by different amounts. Due to the capacitive and inductive components in the amplifier circuits, and the active device, there is a loss in gain at the lower and higher entremes of the frequency range.

# 3. Phase Distortion or Delay Distortion

In this type of distortion, the phase shift between input and output waveforms depends upon the signals of different frequencies.