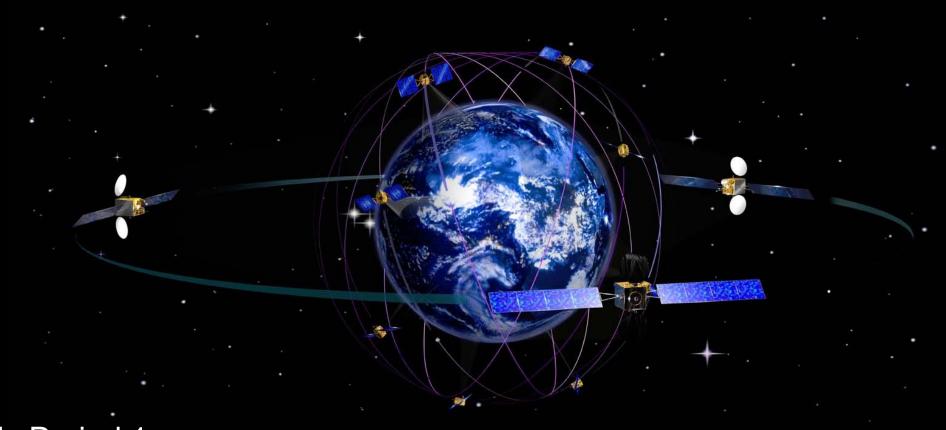
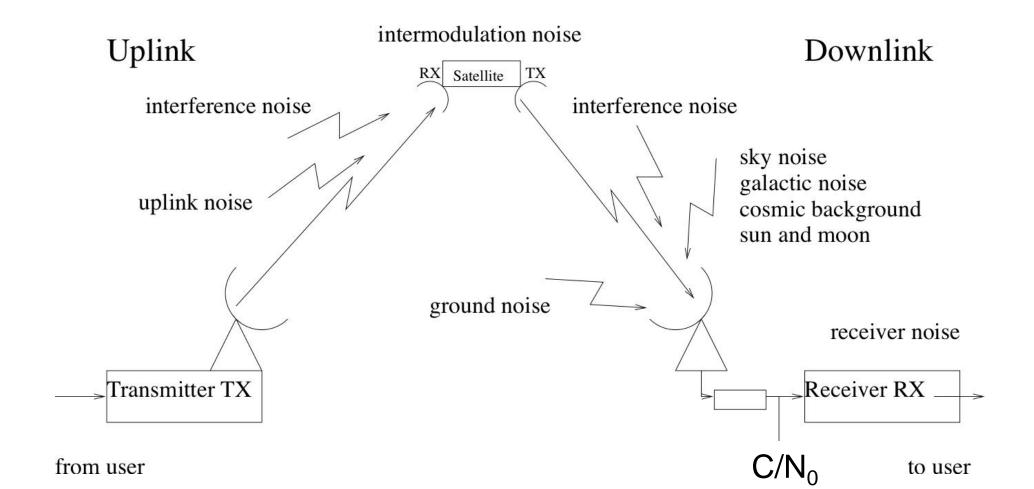
Satellite Communications - RRY100 -



2024 Study Period 1 Lecturer: Rüdiger Haas

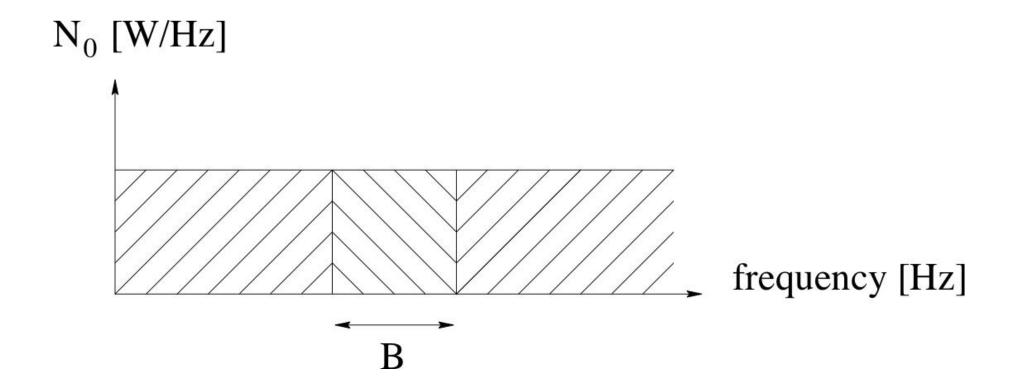
Lecture-3: Noise on satellite links

Noise in satellite communications

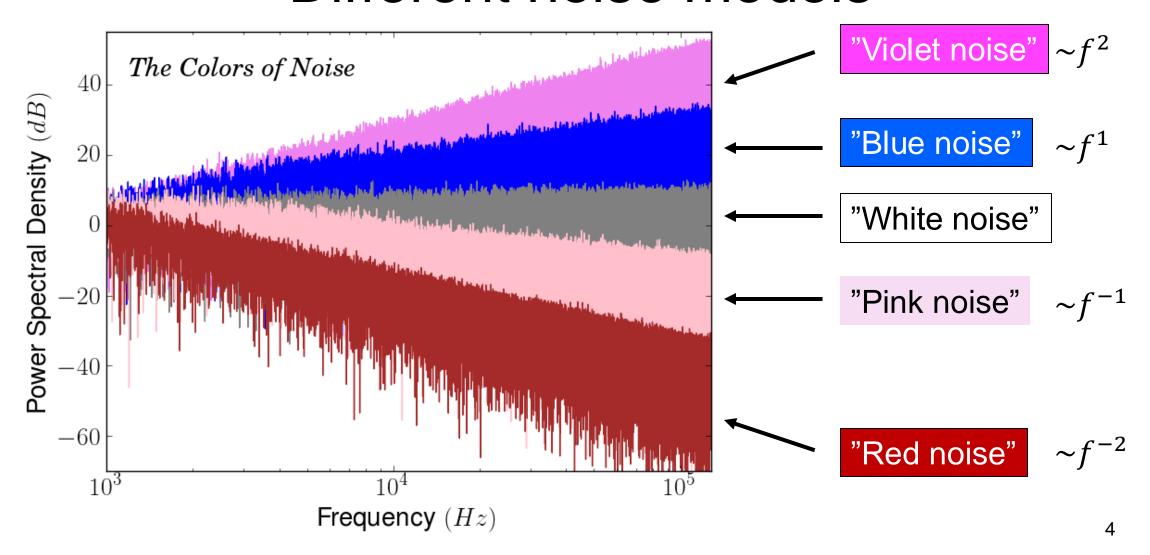


Noise model:

- We usually work with limited bandwidth only
- An appropriate model is 'white noise'
- Real noise over larger bandwidth usually is 'coloured', i.e. not constant over frequency

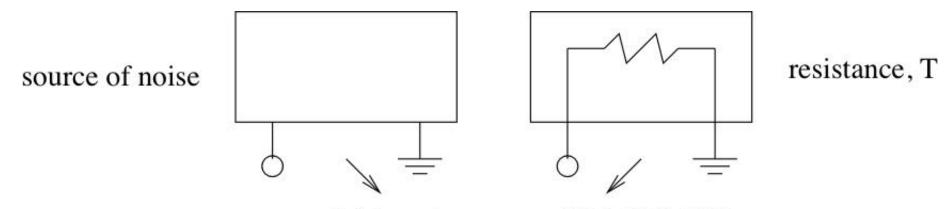


Different noise models



Concept of 'Noise Temperature':

- All objects with a physical temperature T generate electromagnetic radiation
- We compare the noise power of the source of noise (its physical temperature may be different from T) with an object that has a physical temperature T
- If they have the same noise power N we identify the amount of noise by the 'noise temperature T' in Kelvin



available noise power: N=k T B [W]

Ludwig Boltzmann (1844-1906)

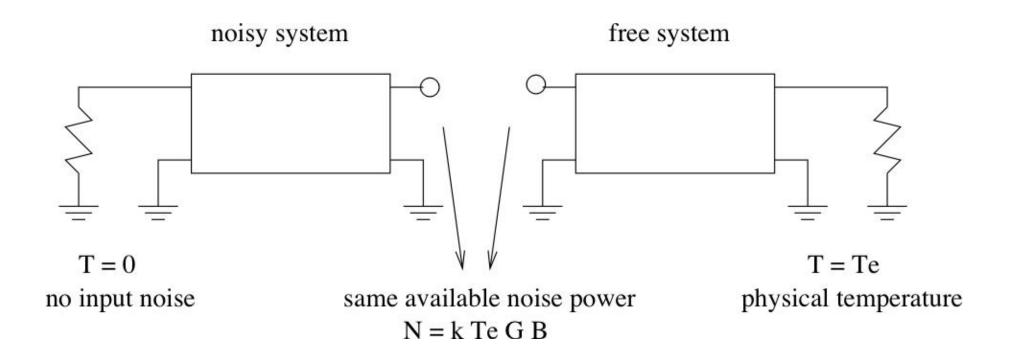
$$N_0 = k \cdot T$$
 [Ws] or [W Hz⁻¹]
 $N = k \cdot B \cdot T$ [W]

Boltzmann's constant:

$$k = 1.3806 \times 10^{-23}$$
 [W s K⁻¹]
 $k = -228.599$ [dB W Hz⁻¹ K⁻¹]

Concept of 'Effective Input Noise Temperature':

- Is a measure of the noise generated by internal components of a two-port element
- Thermodynamic temperature of resistance connected at the input to the four-port element (assumed to be noiseless) that gives the same noise power at the output of the element



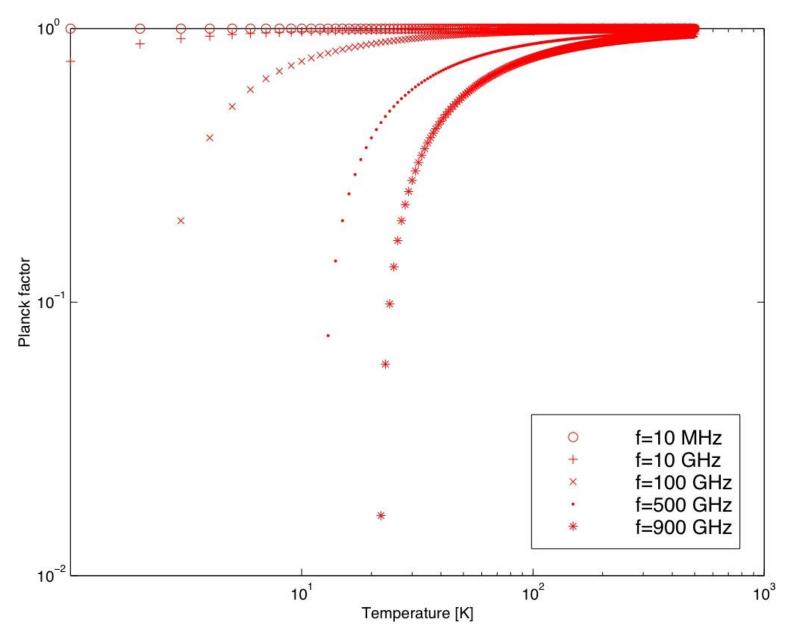
Planck's radiation law:

 The noise power N of an electric circuit in thermodynamical equilibrium at a temperature T within a given bandwidth B is:

$$N = k \times T \times B \times p(f)$$

- With a frequency dependent Planck's factor p(f)
- For satellite communications with frequencies f < 100 GHz the simple formula for noise is good enough

$$N = k \times T \times B$$

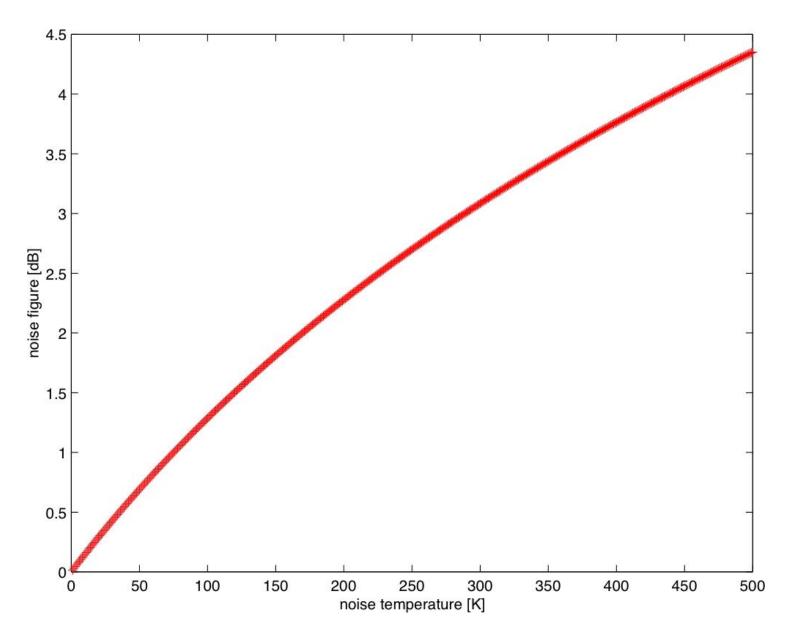


Planck's factor as function of temperature and frequency.

Noise figure:

- The ratio of total noise at the output of a 4-port element to the contribution of the input noise with reference temperature T₀
- Reference temperature is $T_0=290$ K, (in Japan $T_0=293$ K)
- For electronic equipment often given in dB

$$F = \frac{G \cdot k \cdot (T_e + T_0) \cdot B}{G \cdot k \cdot T_0 \cdot B} = \frac{T_e + T_0}{T_0}$$
$$= 1 + \frac{T_e}{T_0} \quad [/]$$



Noise figure *F* versus noise temperature *T*.

- System noise temperature,
 - For example at receiving earth station or satellite
 - At the receiver input (useful convention)
 - Contributions from antenna, feeder and receiving system

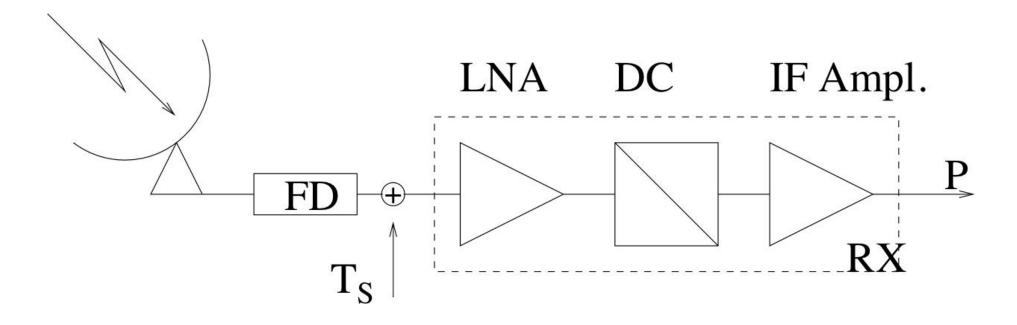


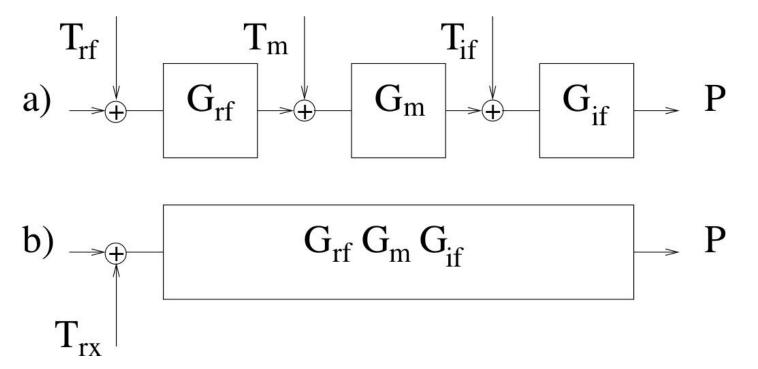
Figure: Schematic drawing of an earth station receiver.

Cascaded systems:

- Compare the power at the output of
 - a) the series of individual elements

and

b) one combined element



- First element should have high gain and low effective input noise temperature (==> low noise amplifier – LNA)
- The temperatures and gains of the following elements are less important

Noise temperature of an attenuator:

- Attenuator contains passive elements at ambient physical temperature T_p
- Attenuation L_{att} (gain $G_{att} = 1/L_{att}$)
- Effective input noise temperature

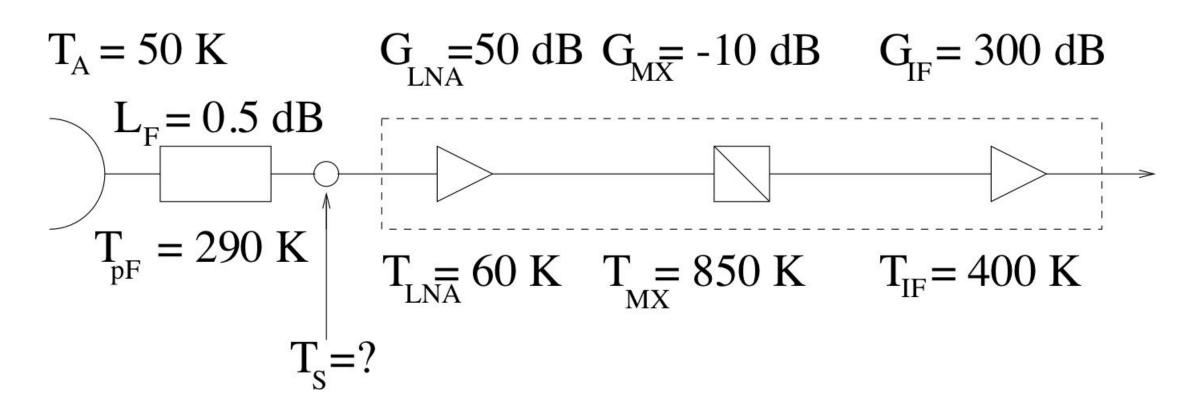
$$T_{e,att} = (L_{att} - 1) \times T_P$$

- An antenna picks up noise from all radiating bodies within its radiation pattern
- Brightness temperature of a radiating body in a specific direction $T_b(\theta, \phi)$
- Antenna gain in this direction $G(\theta, \phi)$
- Antenna noise temperature T_A

$$T_A = \frac{1}{4 \times D} \grave{0} \grave{0} T_b(q, f) \times G(q, f) \sin(q) dq df$$

Example-1:

Calculate the system noise temperature for the following receiving system



- Figure of merit (G/T):
 - Relation of gain to system noise temperature (G/T)
 - Useful as a measure for the quality of a receiving station
 - Quality of the overall link (C/N₀) can be split up in contributions from (equation in dB):
 - sending side (EIRP, losses)
 - path loss (L_{FS}, L_A)
 - receiving side (Figure of merit G/T, losses)

$$\frac{C}{N_0} = \frac{(P_{tx} + G_{t.max} - L_t - L_{f.tx})}{(-L_{FS} - L_A)}$$
 sending side path loss
$$\frac{(+G_{r.max} - L_r - L_{pol} - L_{f.rx} - k - T_s)}{(+G_{r.max} - L_r - L_{pol} - L_{f.rx} - k - T_s)}$$
 receiving side

Some examples:

Туре	f [GHz]	Te [K]	G [dB]	cost
Cryogenic	4	15	30	high
Parametric	20	<100	30	
Cooled	4	35	30	medium
Parametric	12	85	30	
(Peltier)	20	150	30	
Ambient	4	55	30	medium
Parametric	12	150	30	
FET cryogenic	20	200	30	
FET cooled	4	40	60	low
(Peltier)	12	120	60	
	12	160	60	
	20	180	45	
FET ambient	4	70	60	low
	12	130	60	
	12	180	60	
	20	350	22	
HEMT	22	300	16	low

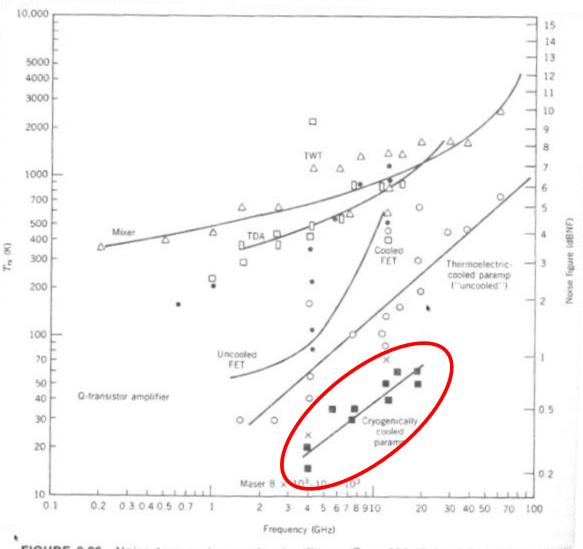
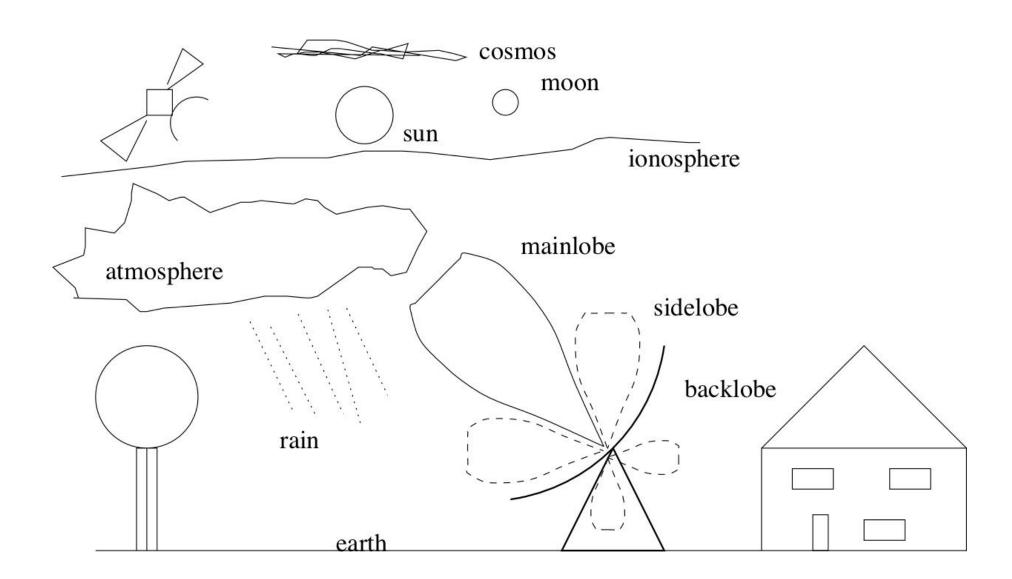
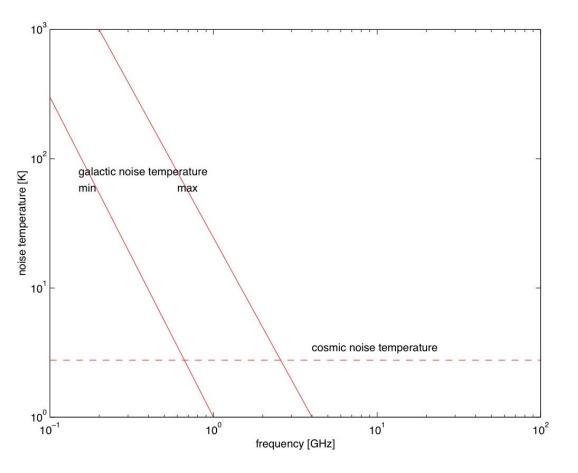


FIGURE 9.20 Noise temperature and noise figure ($T_R = 290 \text{ K}$) for RF amplifiers and mixers.

- Cryogenically cooled paramp
- O = Thermoelectrically cooled paramp
- ☐ = Tunnel diode amplifier
- $\triangle = Mixer$
- Uncooled paramp
- ☐ = Travelling wave tube amplifier
- × = High electron mobility transistor (HEMT) low noise converter (1993)

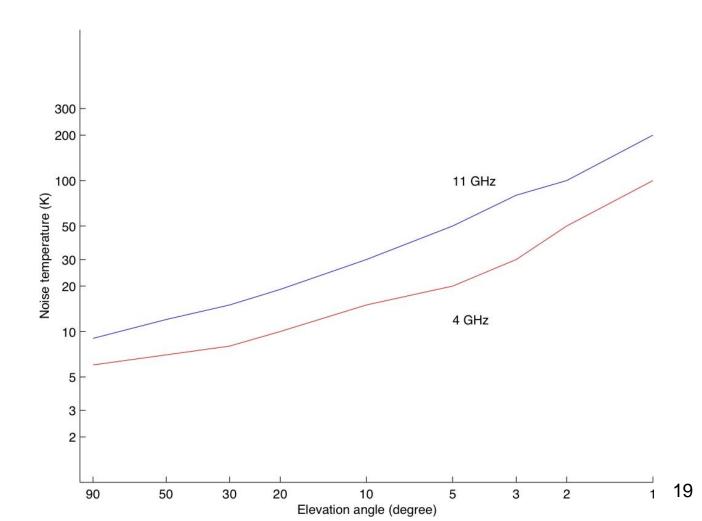
• Contributions to the antenna noise temperature of an earth station:





Example for cosmic and galactic noise temperatures

Example for atmospheric noise temperature ("clear-sky") as function of frequency and elevation



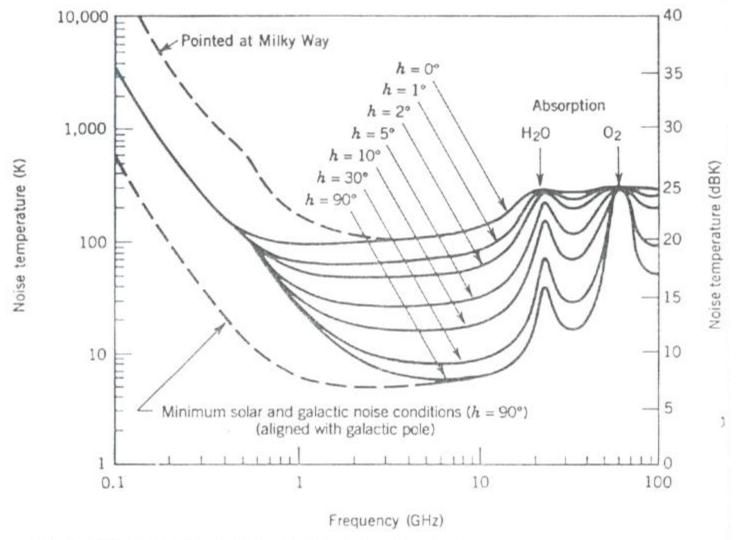


FIGURE 9.8 Earth station antenna noise temperature.

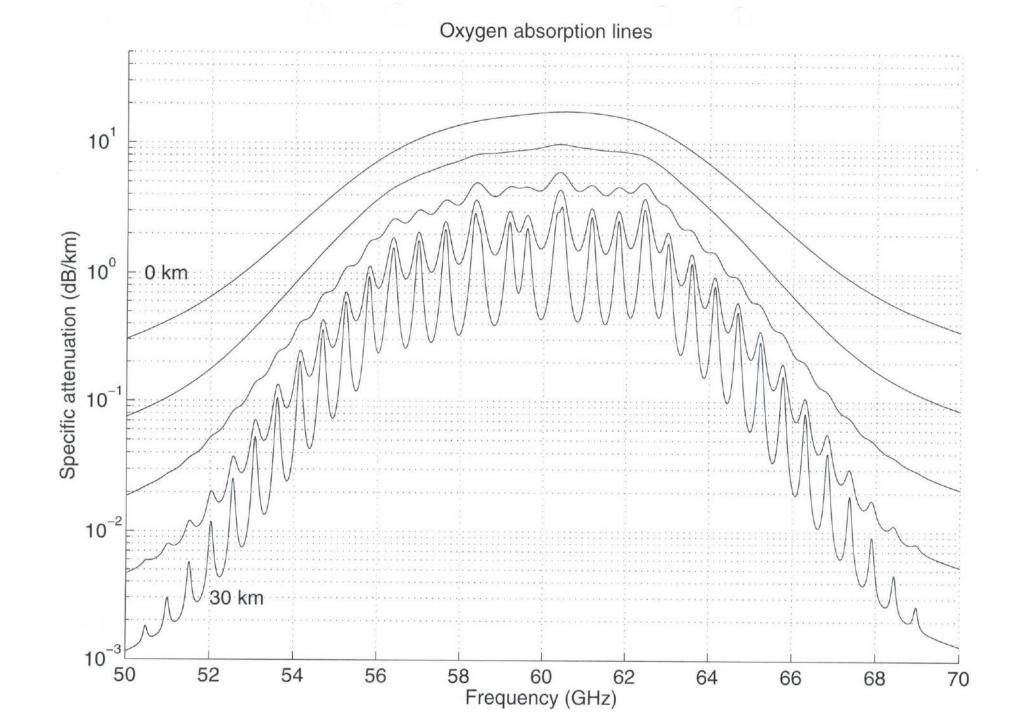
Assumptions:

Antenna has no earth-looking sidelobes or a backlobe (zero ground noise)
Antenna is lossless

h is antenna elevation angle (deg)

Sun not considered

Cool, temperate-zone troposphere



Rain has two effects:

- Rain acts as an attenuator with temperature $T_m = 275$ K and contributes with a noise temperature:

$$T_{Rain} = T_m \cdot (1 - 1/A_{Rain})$$

Rain increases the sky noise (for downlink)

$$T_{skyrain} = T_{sky}/A_{Rain}$$

- Noise from the ground T_G:
 - Picked up from mainlobe in low elevations
 - Picked up from back- and sidelobes
 - Approximate calculation:

$$T_{Ground} = \sum_{i}^{lobes} G_i \cdot \frac{\Omega_i}{4 \cdot \pi} \cdot T_G$$

– Where G_i and Ω_i are the mean gain and the solid angle of the lobe, respectively, T_G the ground contribution is depending on elevation

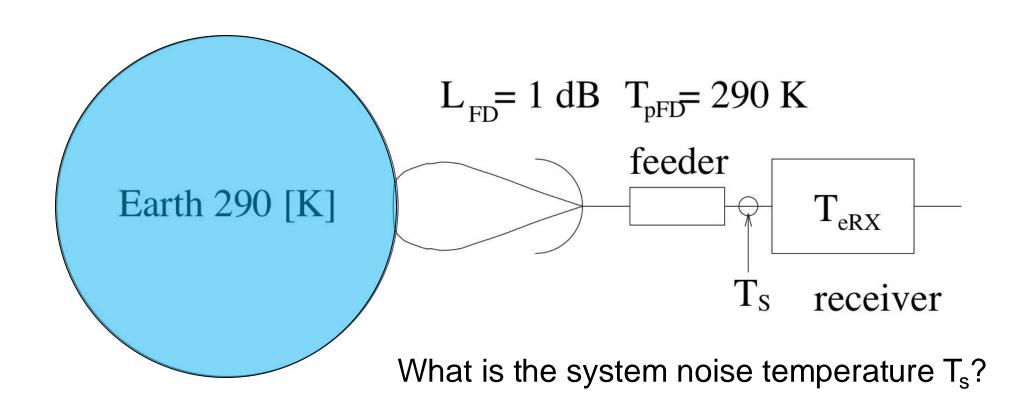
$$T_G = 290 \text{ K}$$
 for $\varepsilon < -10^\circ$

$$T_{\rm G} = 150 \text{ K for } -10^{\circ} < \varepsilon < 0^{\circ}$$

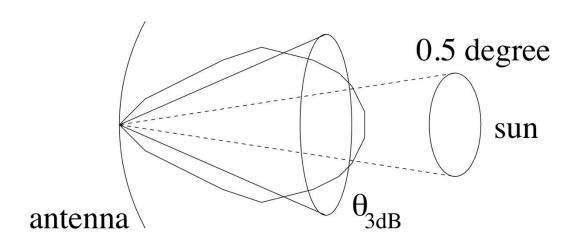
$$T_G = 50 \text{ K}$$
 for $0^{\circ} < \varepsilon 10^{\circ}$

$$T_G = 10 \text{ K for} \quad 10^{\circ} < \varepsilon < 90^{\circ}$$

- Satellite antenna noise temperature:
 - Satellite captures noise from the earth and space
 - Satellites "sees" the earth approximately as "black body" of about 290 K
 - Thus there is always high antenna noise temperature onboard satellites
 - Not much can be gained by using LNAs onboard satellites



Additional noise due to sun interference



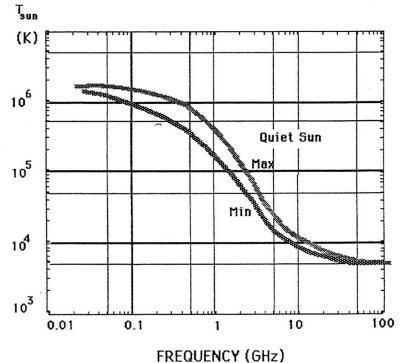
0.5 degree
$$\Delta T_A = T_{sun} \left(\frac{0.5}{\theta_{3dB}} \right)^2 \text{ if } \theta_{3dB} > 0.5^{\circ}$$

$$\Delta T_A = T_{sun} \qquad \text{if } \theta_{3dB} \le 0.5^{\circ}$$

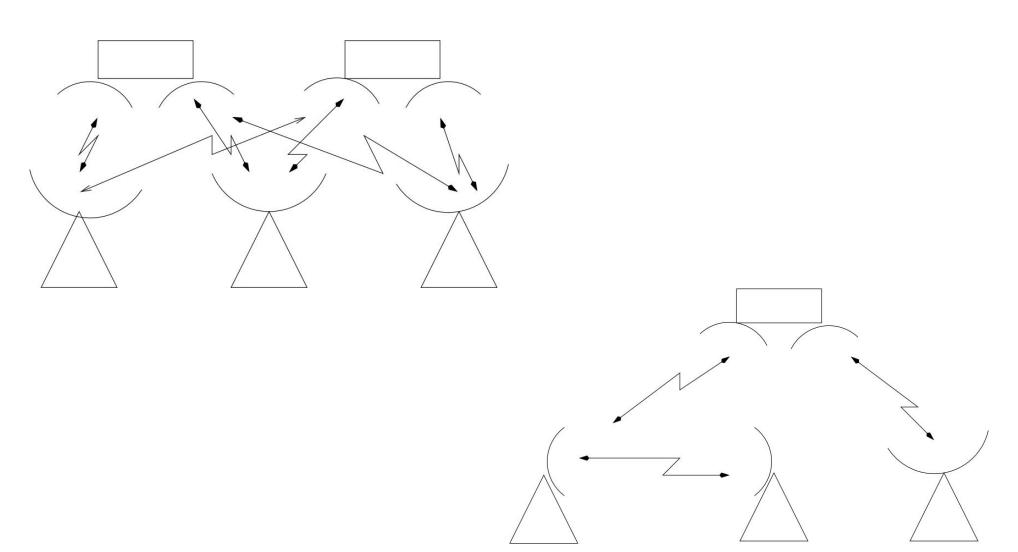
$$\Delta T_A = T_{sun}$$
 if $\theta_{3dB} \le 0.5^\circ$

The mean brightess temperature of the sun is frequency dependent, e.g.:

$$T_{Sun} = \frac{1.9610^5}{f} \left[1 + \frac{\sin 2\pi \left(\frac{\log 6(f - 0.1)}{2.3} \right)}{2.3} \right]$$



- Interference noise:
 - Between different satellite systems
 - Between satellite and terrestrial systems



Intermodulation noise:

- Due to non-linearity of satellite transponders
- Transponders have a saturation point

Intermodulation products are generated when the transponder is operated in the

saturation region

output power / saturated output power with 1 carrier output saturated power 0 dBsingle carrier mode $P_{1,out}$ - 5 dB one out of two carriers $P_{IMP.3rd}$ -10 dB one out of the 3rd order intermodulation products -15 dB -10 dB -5 dB 0 dB-15 dB input power 1(2) / input power at saturation with 1 carrier

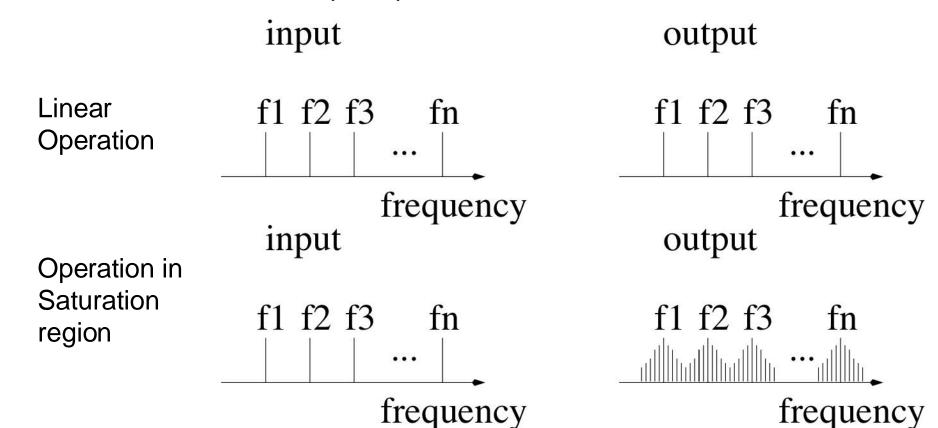
3rd order interception point

Simplified model of a satellite channel:

- Output in linear region is the amplified carriers
- Output in the saturation region is amplified carriers plus intermodulation products f_{IM}:

$$f_{IM} = m_1 \cdot f_1 + m_2 \cdot f_2 + \cdots + m_n \cdot f_n$$

- Only odd products of intermodulation noise are dangerous
- 3rd and 5th order intermodulation products are most important
- Of interest is the third order interception point



- Repeater as non-linear device:
 - Input voltage of e.g. two carriers with $f_i=\omega_i$

$$V_i = A \times \cos(W_1 \times t) + B \times \cos(W_2 \times t)$$

Output voltage includes higher order products

$$V_{o} = a \times \left(A \times \cos(W_{1} \times t) + B \times \cos(W_{2} \times t) \right)$$

$$+b \times \left(A \times \cos(W_{1} \times t) + B \times \cos(W_{2} \times t) \right)^{2}$$

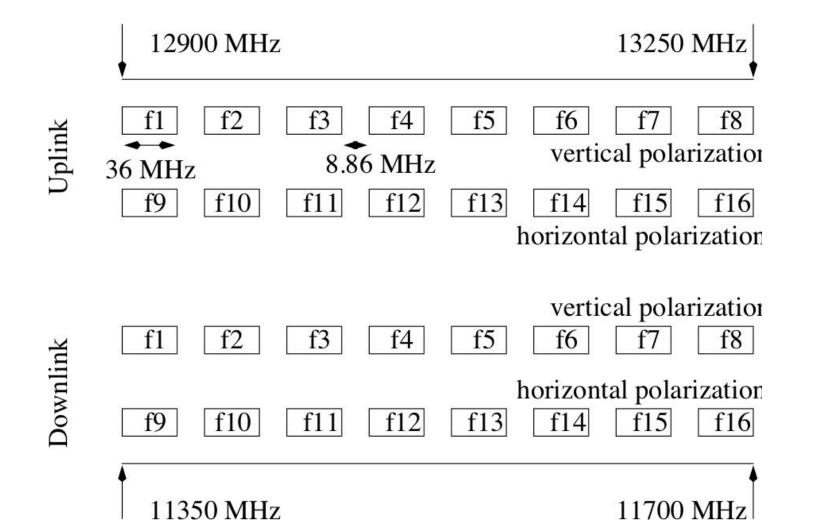
$$+c \times \left(A \times \cos(W_{1} \times t) + B \times \cos(W_{2} \times t) \right)^{3}$$

$$+ \dots + \frac{3}{4} c \times A^{2} \times B \times \cos\left((2 \times W_{1} - W_{2}) \times t\right)$$

$$+ \dots + \frac{3}{4} c \times B^{2} \times A \times \cos\left((2 \times W_{1} - W_{2}) \times t\right)$$

Example for a satellite payload:

16 transponders with 36 MHz each, uplink frequencies 12.90–13.25 GHz and downlink frequencies 11.25–11.70 GHz



29

```
f1=12918.00 MHz f5=13097.43 MHz
f2=12962.86 MHz f6=13142.29 MHz
f3=13007.71 MHz f7=13187.14 MHz
f4=13052.57 MHz f8=13232.00 MHz
```

- 2nd order intermodulation products:=> outside carriers
- 3rd order intermodulation products: => affect f2 and f5
- 4th order intermodulation products:=> outside carriers
- 5th order intermodulation products: => affect f6 and f1

Example-2

A satellite transponder with power $P_{tx} = 10$ W is connected via a feeder (feeder loss $L_{f,tx} = 1$ dB) to a transmitting antenna with maximal gain $G_{max} = 33$ dBi and depointing loss $L_t = 3$ dB. An earth station with a reflector of diameter D = 3.6 m and aperture efficiency of 0.6 is connected via a feeder of loss L_{frx} =0.5 dB with physical temperature $T_{ph.fd}$ = 290 K to a receiver with effective input noise temperature of T_{erx} = 60 K. The depointing loss of the antenna is $L_t = 1$ dB and the half-power beam-width of the reflector is $\theta_{3dB} = 0.5$ degrees. The sky temperature is $T_{skv} = 8$ K and the ground noise temperature is $T_{Ground} = 20 \text{ K}$. The free space loss is $L_{FS} = 206 \text{ dB}$.

Questions:

- 1. What is the system noise temperature T_{sys} of the earth station under clear-sky conditions?
- 2. What is the system noise temperature of the earth station under rain conditions, assuming a rain attenuation of $A_{rain} = 6 \, dB$ and thermodynamic temperature of $T_{ph} = 275 \, K$?
- 3. What does the the rain mean for the link performance of this downlink?

Antenna temperature in clear sky: $T_A = T_{sky} + T_{ground} = 28 \text{ K}$ $T_{A.clear-skv} = 28 \text{ K}$

Antenna temperature in rain: $T_A = T_{sky}/A_{rain} + T_m (1-1/A_{rain}) + T_{ground}$ $T_{A rain} = 228 \text{ K}$

System noise temperature: $T_S = T_A/L_{f.rx} + T_{ph.frx} (1-1/L_{ph.frx}) + T_{e.RX}$

- 1) T_S in clear sky: $T_S = 117 \text{ K}$
- 2) T_S in rain: $T_S = 295 \text{ K}$

Satellite EIRP: $EIRP = P_{tx} + G_{t.max} - L_{ftx} - L_t$

EIRP = 39 dBW

Gain of earth station: $G_{r.max} = \eta \cdot ((\pi \cdot 70^{\circ})/\theta_{3dB})^{2}$ [/]

 $G_{r,max} = 51 \text{ dBi}$

$$G_r = G_{r,max} - L_r - L_{frx} = 49.5 \text{ dBi}$$

Figure of merit of earth station: $(G/T)_{es} = G_r - T_S$

 $(G/T)_{es}$ in clear sky: $(G/T)_{es} = 28.5 dBi K^{-1}$

 $(G/T)_{es}$ in rain: $(G/T)_{es} = 24.5 dBi K^{-1}$

Path loss:

 L_{path} in clear sky: $L_{path} = L_{FS}$ = 206 dB

 L_{path} in rain: $L_{path} = L_{FS} + A_{rain} = 212 \text{ dB}$

Downlink performance: $(C/N_0)_{down} = EIRP - L_{path} + (G/T_S)_{es} - k$

 $(C/N_0)_{down}$ in clear sky: $(C/N_0)_{down} = 90.1 dBHz$

 $(C/N_0)_{down}$ in rain: $(C/N_0)_{down} = 80.1 dBHz$

3) Rain reduces the system performance by 10 dB.

Short summary of today's topics

- Noise models
- Noise temperature, effective input noise temperature
- Noise figure
- System noise temperature T_{sys}
- Cascades systems
- Figure of merit
- Gases and rain
- Non-linear devices, intermodulation noise