

Exercise 1

Aug. 30, 2023

Problem 1 (LTI Systems)

Consider an LTI system whose response to the signal $x(t)$ in Figure 1(a) is the signal $y(t)$ in Figure 1(b). Sketch the response of the system to the input signal $z(t)$ shown in Figure 1(c).

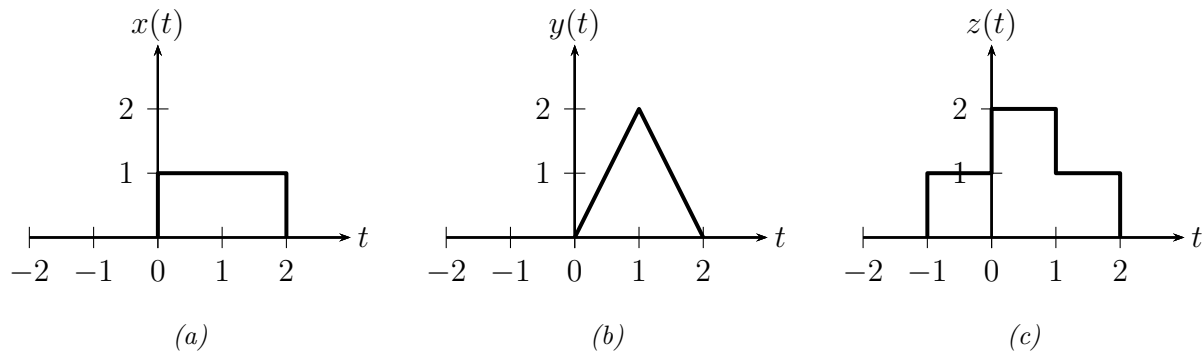


Figure 1: Problem 1.

The signal $z(t)$ can be represented as a linear combination of $x(t)$ and its shifted version $x(t+1)$, i.e.,

$$z(t) = x(t+1) + x(t).$$

Using properties of LTI systems, we conclude that the response of the LTI system to $z(t)$ is also a linear combination of $y(t)$ and its shifted version $y(t+1)$, i.e.,

$$r(t) = y(t+1) + y(t).$$

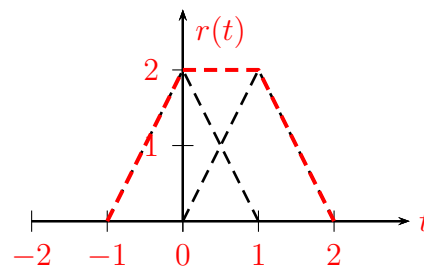


Figure 2: The output of the LTI system.

Problem 2 (Fourier Transform Properties)

Find the Fourier transform of the signal $x(t)$ shown in Figure 2, in two ways as mentioned below.

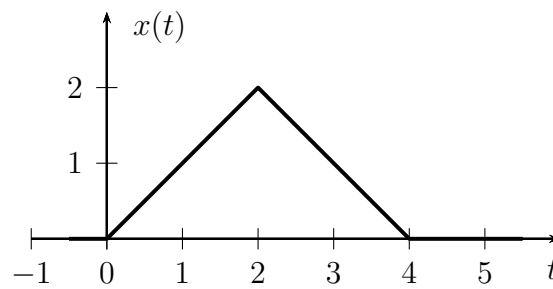


Figure 3: Problem 2.

1. Calculate the Fourier transform using the definition.

$$\begin{aligned}
 X(f) &= \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt \\
 &= \int_0^2 t e^{-j2\pi f t} dt + \int_2^4 (4-t) e^{-j2\pi f t} dt \\
 &= \int_{-2}^0 (\tau+2) e^{-j2\pi f (\tau+2)} d\tau + \int_0^2 (2-\tau) e^{-j2\pi f (\tau+2)} d\tau \\
 &= e^{-j4\pi f} \left[\int_{-2}^0 \tau e^{-j2\pi f \tau} d\tau - \int_0^2 \tau e^{-j2\pi f \tau} d\tau + 2 \int_{-2}^2 e^{-j2\pi f \tau} d\tau \right] \\
 &= e^{-j4\pi f} \left[\left. \frac{\tau e^{-j2\pi f \tau}}{-j2\pi f} \right|_{-2}^0 - \left. \frac{e^{-j2\pi f \tau}}{(j2\pi f)^2} \right|_{-2}^0 - \left. \frac{\tau e^{-j2\pi f \tau}}{-j2\pi f} \right|_0^2 + \left. \frac{e^{-j2\pi f \tau}}{(j2\pi f)^2} \right|_0^2 + 2 \left. \frac{e^{-j2\pi f \tau}}{-j2\pi f} \right|_{-2}^2 \right] \\
 &= e^{-j4\pi f} \left[\frac{e^{-j4\pi f} + e^{j4\pi f} - 2}{(j2\pi f)^2} \right] \\
 &= e^{-j4\pi f} \left[\frac{(e^{-j2\pi f} - e^{j2\pi f})^2}{(j2\pi f)^2} \right] \\
 &= e^{-j4\pi f} \left[\frac{(e^{j2\pi f} - e^{-j2\pi f})^2}{(j2\pi f)^2} \right] \\
 &= e^{-j4\pi f} \left[\frac{e^{j2\pi f} - e^{-j2\pi f}}{j2\pi f} \right]^2 \\
 &= e^{-j4\pi f} \left[\frac{\sin(2\pi f)}{\pi f} \right]^2 \\
 &= 4e^{-j4\pi f} \left[\frac{\sin(2\pi f)}{2\pi f} \right]^2 \\
 &= 4e^{-j4\pi f} \text{sinc}^2(2f).
 \end{aligned}$$

2. Use the Fourier transform properties.

$x(t)$ can be represented as a convolution of the rectangular pulse $y(t) = \text{rect}(\frac{t-1}{2}) = \mathbf{I}\{0 < t < 2\}$ with itself, i.e.,

$$x(t) = y(t) * y(t).$$

The Fourier transform of such a pulse is

$$Y(f) = 2e^{-j2\pi f} \text{sinc}(2f).$$

The Fourier transform of a convolution is the product of Fourier transforms, i.e.,

$$X(f) = \left(2e^{-j2\pi f} \text{sinc}(2f)\right)^2 = 4e^{-4j\pi f} \text{sinc}^2(2f).$$

Problem 3 (Fourier)

Let $x(t)$ be a signal that is band-limited to W .

1. Show that if $f > W$, then

$$\int_{-\infty}^{\infty} x(t) \cos(2\pi ft) \, dt = \int_{-\infty}^{\infty} x(t) \sin(2\pi ft) \, dt = 0.$$

The fact that the signal is band-limited means that

$$X(f) = 0 \text{ for } |f| > W.$$

This can be rewritten as

$$\int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} \, dt = 0 \text{ for } |f| > W.$$

Using Euler's equation for the complex exponential we get

$$\int_{-\infty}^{\infty} x(t) (\cos(2\pi ft) - j \sin(2\pi ft)) \, dt = 0 \text{ for } |f| > W$$

or

$$\int_{-\infty}^{\infty} x(t) \cos(2\pi ft) \, dt - j \int_{-\infty}^{\infty} x(t) \sin(2\pi ft) \, dt = 0 \text{ for } |f| > W.$$

A complex number is zero iff the real part and the imaginary part are both zero, which implies

$$\int_{-\infty}^{\infty} x(t) \cos(2\pi ft) \, dt = \int_{-\infty}^{\infty} x(t) \sin(2\pi ft) \, dt = 0 \text{ for } |f| > W.$$

Interpretation: the inner product of a slowly varying signal and a harmonic signal of high frequency is zero.

2. Show that if $f > W/2$, then

$$\int_{-\infty}^{\infty} x(t) \cos^2(2\pi ft) \, dt = \frac{1}{2} \int_{-\infty}^{\infty} x(t) \, dt.$$

Using trigonometric identities, we can express cosine square as

$$\cos^2(2\pi ft) = \frac{1}{2} + \frac{1}{2} \cos(4\pi ft).$$

The rest of the proof is similar to the previous item.

Problem 4 (Fourier)

Prove that

$$\text{sinc}(2Wt) \cos(2\pi Wt) = \text{sinc}(4Wt).$$

$$\begin{aligned} \text{sinc}(2Wt) \cos(2\pi Wt) &= \frac{\sin(2\pi Wt) \cos(2\pi Wt)}{2\pi Wt} \\ &= \frac{\sin(2\pi Wt + 2\pi Wt) + \sin(2\pi Wt - 2\pi Wt)}{2 \cdot 2\pi Wt} \\ &= \frac{\sin(4\pi Wt)}{4\pi Wt} = \text{sinc}(4Wt). \end{aligned}$$

Illustrate this identity in the frequency domain.

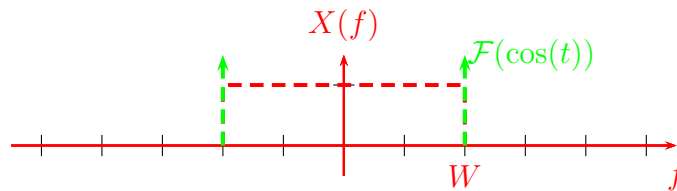


Figure 4: The signals in the frequency domain.

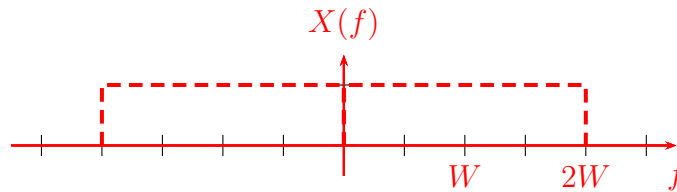


Figure 5: The result of the convolution.

Problem 5 (Nyquist pulse)

Let $v(t)$ be a continuous signal with limited energy, i.e., $\int_{-\infty}^{\infty} v^2(t) dt < \infty$ and $v(0) = 1$. Define $g(t) = v(t)\text{sinc}(t/T)$.

1. Show that $g(t)$ is a Nyquist pulse for the time interval T .

$$g(0) = v(0)\text{sinc}(0) = 1 \text{ and}$$

$$g(kT) = v(kT)\text{sinc}(k) = v(kT)0 = 0, \text{ for all } k \neq 0.$$

This means that the pulse satisfies the Nyquist criterion in the time domain.

2. Argue that the raised-cosine pulse is a Nyquist pulse.

The raised-cosine pulse is defined in the time domain as

$$r(t) = \text{sinc}\left(\frac{t}{T}\right) \frac{\cos\left(\frac{\pi\beta t}{T}\right)}{1 - \frac{4\beta^2 t^2}{T^2}}.$$

Since $r(t)$ is of form $v(t)\text{sinc}(t/T)$, it satisfies the Nyquist criterion.

3. Find the Fourier transform $G(f)$ as a function of $V(f)$ and show that it satisfies the Nyquist criterion in the frequency domain.

The Fourier transform of $g(t)$ is given by the convolution of $V(f)$ with a rectangular pulse $T \text{I}\left\{-\frac{1}{2T} < f < \frac{1}{2T}\right\}$.

$$\begin{aligned} G(f) &= \int_{-\infty}^{\infty} V(\tau) T \text{I}\left\{-\frac{1}{2T} < f - \tau < \frac{1}{2T}\right\} d\tau \\ &= T \int_{-\infty}^{\infty} V(\tau) \text{I}\left\{f - \frac{1}{2T} < \tau < f + \frac{1}{2T}\right\} d\tau \\ &= T \int_{f - \frac{1}{2T}}^{f + \frac{1}{2T}} V(\tau) d\tau. \end{aligned}$$

Let us check the Nyquist criterion in the frequency domain.

$$\begin{aligned} \sum_{k=-\infty}^{\infty} G\left(f - \frac{k}{T}\right) &= \sum_{k=-\infty}^{\infty} T \int_{f - \frac{k}{T} - \frac{1}{2T}}^{f - \frac{k}{T} + \frac{1}{2T}} V(\tau) d\tau \\ &= T \int_{-\infty}^{\infty} V(\tau) d\tau \\ &= T \int_{-\infty}^{\infty} V(\tau) e^{j2\pi t\tau} d\tau \Big|_{t=0} \\ &= T v(0) = T, \end{aligned}$$

i.e., the signal satisfies the Nyquist criterion in the frequency domain.

Problem 6 (Nyquist Pulse)

The pulses are defined in frequency domain and their spectra are shown in the below figure (frequency is in MHz).

$$X_1(f) = \begin{cases} 2 - 0.5|f| & \text{if } |f| \leq 4 \\ 0 & \text{o.w.} \end{cases} \quad X_2(f) = \begin{cases} 2 & \text{if } 0 \leq f \leq 3 \\ -2 & \text{if } -3 \leq f < 0 \end{cases}$$

1. Which pulse(s) satisfy the Nyquist criterion and for which symbol rate?

$x_1(t)$ satisfies the Nyquist criterion for symbol rate $R_1 = 4$ MHz. However, $x_2(t)$ does not satisfy the Nyquist criterion, because although for $R_2 = 3$ MHz, the sum of shifted versions of $X_2(f)$ is a constant, it is equal to zero, meaning that $x_2(kT) = 0$ for all k , while we need to have $x_2(kT) = 0$ for $k \neq 0$ in the Nyquist criterion.

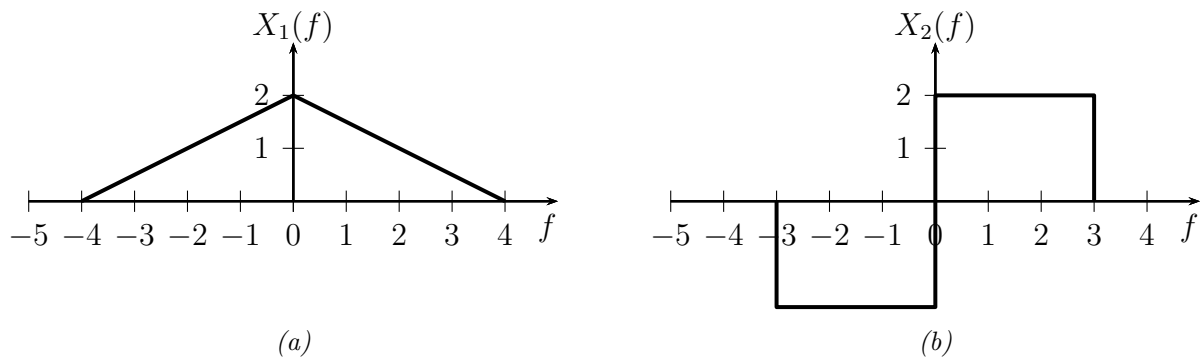


Figure 6: Problem 6.

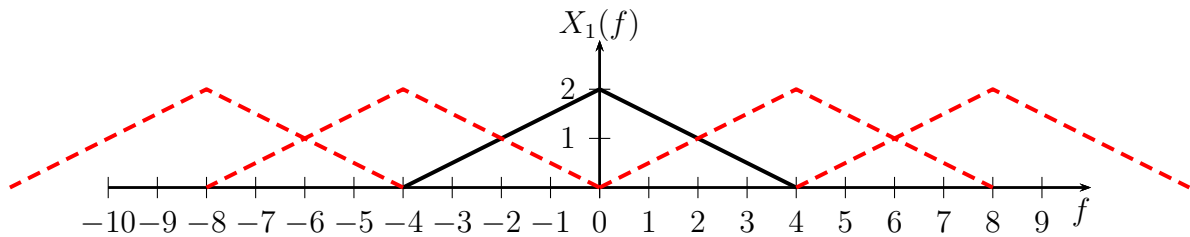


Figure 7

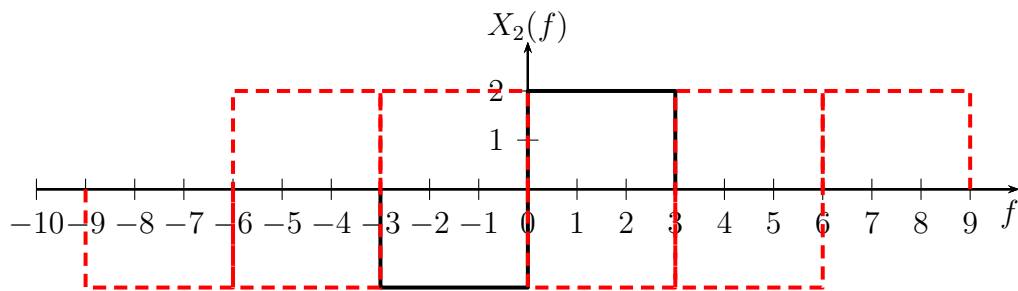


Figure 8

2. Find the value at $t = 0$ and the energy for these signals.

- For the Nyquist pulse $x_1(t)$, one can write

$$\sum_{k=-\infty}^{\infty} X_1\left(f - \frac{k}{T_1}\right) = T_1 x_1(0).$$

Therefore,

$$x_1(0) = \frac{\sum_{k=-\infty}^{\infty} X_1\left(f - \frac{k}{T_1}\right)}{T_1} = R_1 \sum_{k=-\infty}^{\infty} X_1\left(f - \frac{k}{T_1}\right).$$

To obtain the energy, we use the parseval's theorem:

$$E_1 = \int_{-\infty}^{\infty} x_1^2(t) dt = \int_{-\infty}^{\infty} X_1^2(f) df$$

- Since $x_2(t)$ is not a Nyquist pulse, we need to find the value of the signal at $t = 0$

in another way. We know that

$$x_2(t) = \int_{-\infty}^{\infty} X_2(f)e^{j2\pi ft}df.$$

Therefore

$$x_2(0) = \int_{-\infty}^{\infty} X_2(f)df = 0.$$

For the energy, we can still use parseval's theorem write

$$E_2 = \int_{-\infty}^{\infty} X_2^2(f)df.$$