

# Analog and Mixed Signal Verification for Communications

-- the role of Automated Macromodelling

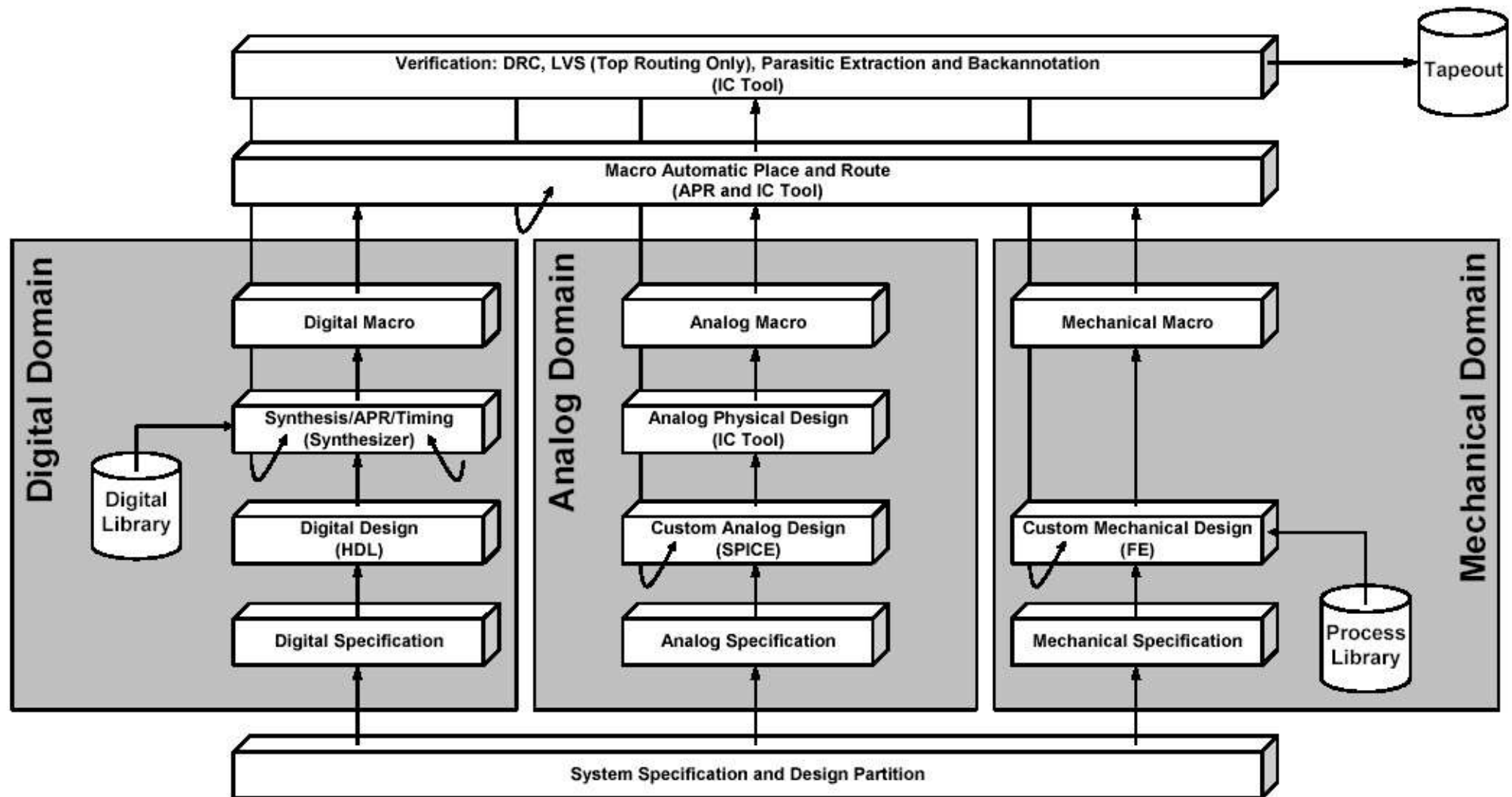
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University of Minnesota

# Automated Macromodelling

- \* What and Why
- \* Background: LTI Macromodelling
- \* Nonlinear Macromodelling
- \* Interference Noise Macromodelling
- \* Oscillator Macromodelling

# Today's bottom-up design

McCorquodale et al, U of Michigan



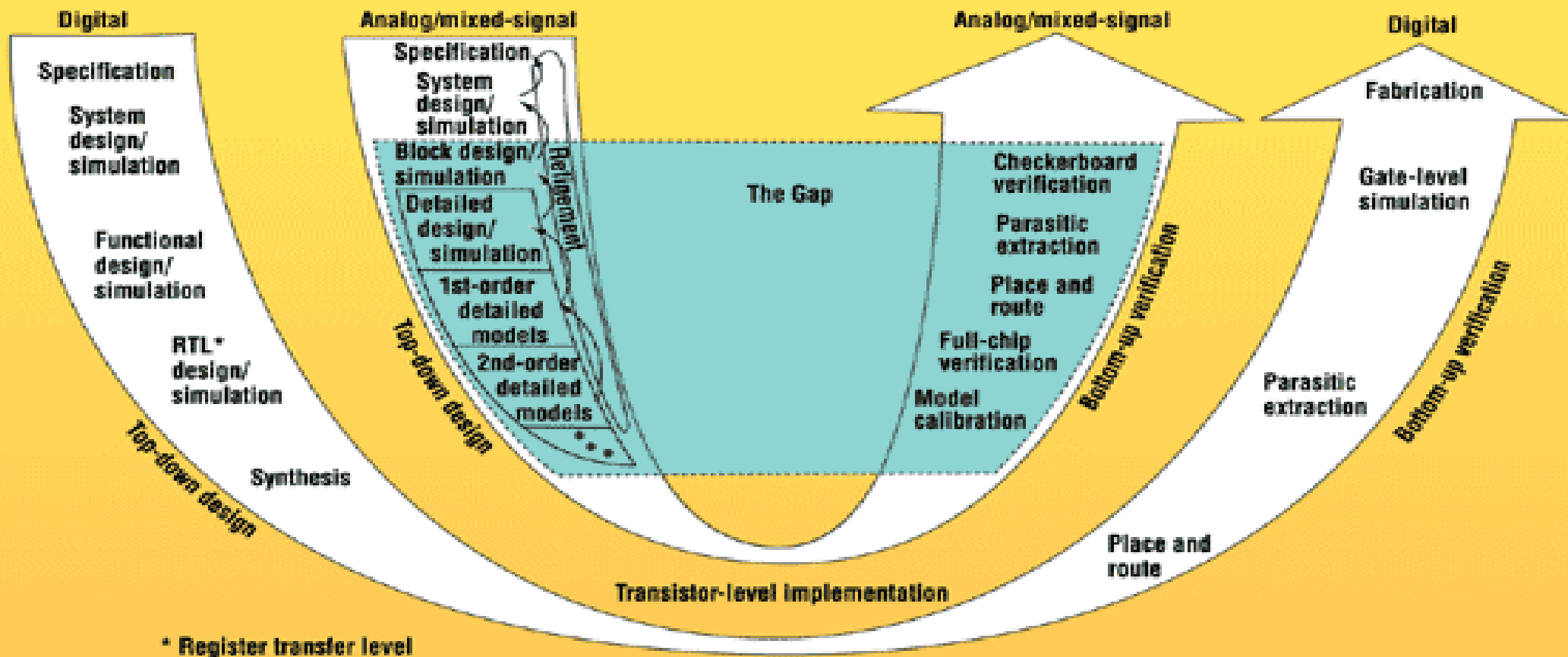
\* 3-5 spins=cutting edge; >10 at Lucent

# Mixed-Signal Verification Challenges

- \* Large entire (multi-physics) systems to verify
- \* Interactions between blocks, “second order” effects
- \* Interconnect, coupling, noise
- \* Speed with SPICE-like accuracy becoming necessity
- \* Impossible at SPICE level

# “The Gap”

EE Times

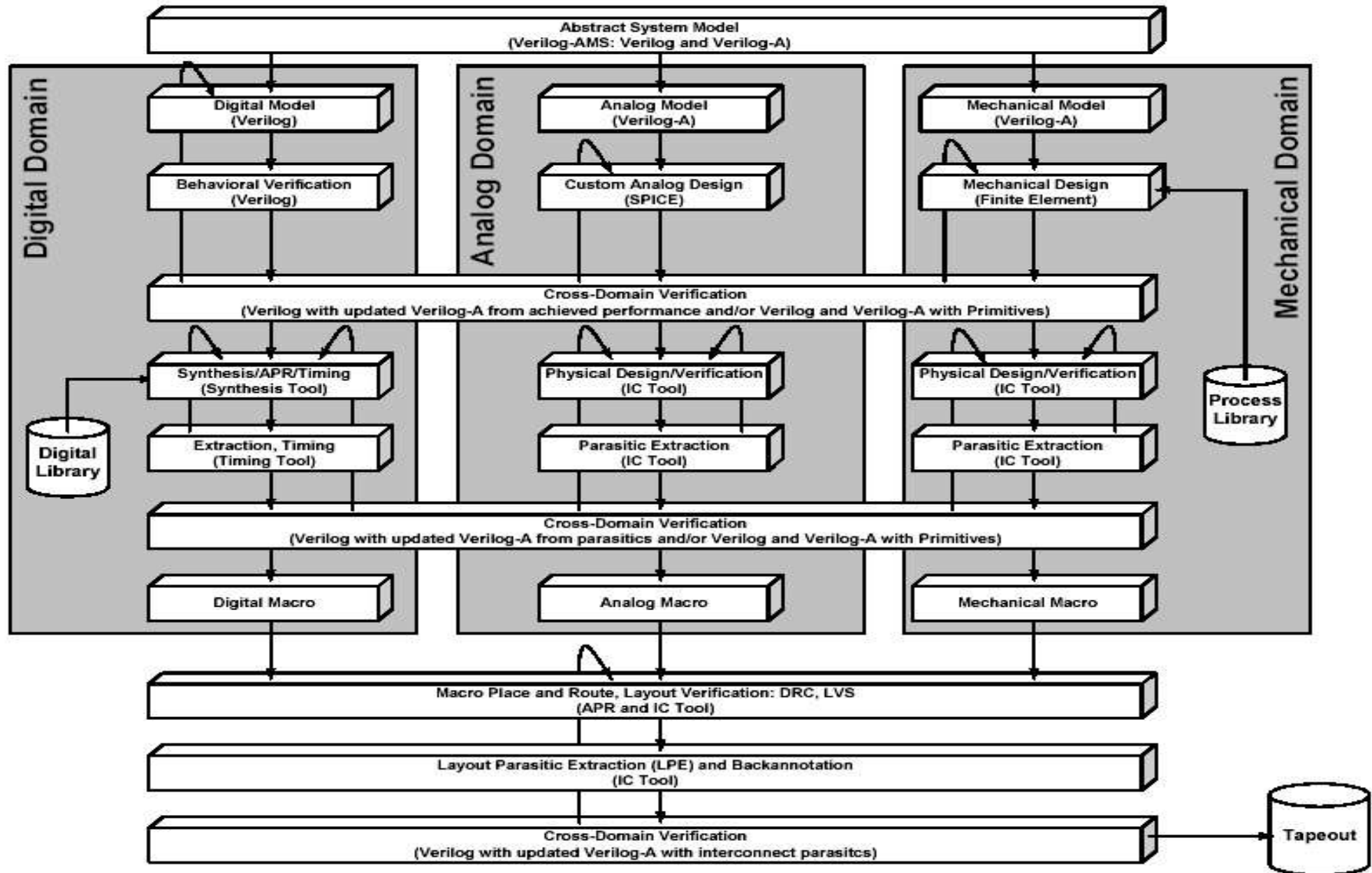


“The primary problem hindering the change to analog top-down design and bottom-up verification has been the lack of tool support for the design process between system-level specification and transistor-level implementation, as well as between transistor implementation and chip fabrication. These missing tools are commonly referred to as The Gap.” - EE Times, 2001

# “The Gap”

- \* Solution: Good bottom-up macromodels

# Top-Down Design



# Generating Macromodels

- \* Today: manually
- \* Highly skilled activity
  - \* what if designer leaves?
- \* Mistakes (especially “second order”)
- \* Time-consuming
- \* Tomorrow: myriad new technologies
  - \* carbon nanotubes, spintronics, ballistic nanotransistors, photonic crystals, ...



# Automated Macromodelling

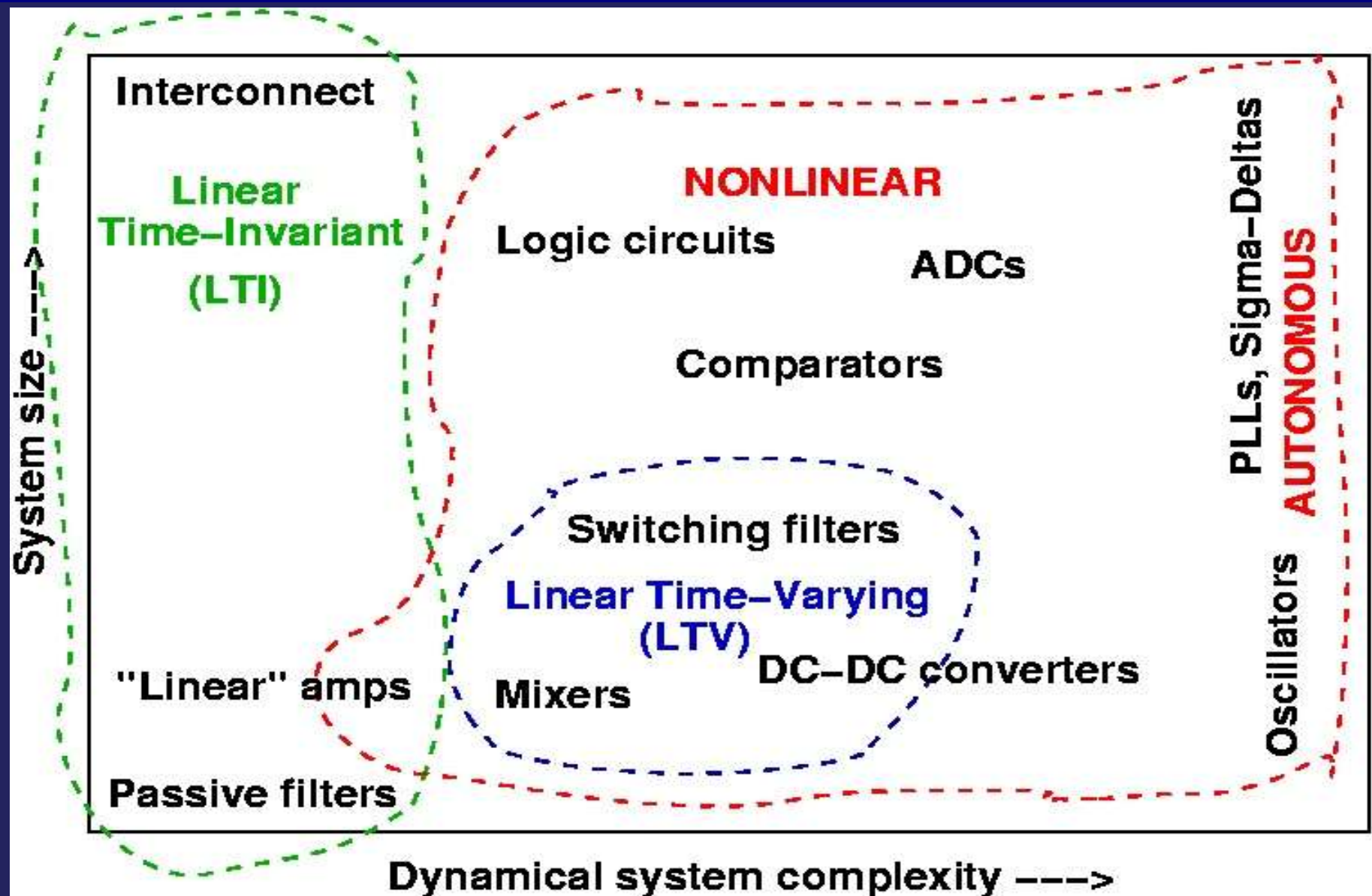
- \* The dream: push-button bottom-up model generation
  - \* a.k.a Rob's Red Button: “go make my macromodel”
  - \* prescribed accuracy guaranteed
  - \* trade off speed vs accuracy
- \* Needed for design sustainability
  - \* complexity exceeds manual ability to keep up

# “The Gap”

“At this point, you may wonder why you should bother with behavioral libraries and calibration. Why not just submit the transistor-level design to some smart software and let it come up with a model? Unfortunately, despite some claims to the contrary, practical model synthesis is still a long way off. Attempts at this technology rely on pre-existing templates, which are unlikely to exist for leading-edge or proprietary designs. There's no pushbutton approach to analog modeling, and from all indications, this will remain the case for some time to come.” - EE Times, 2001

- \* Perhaps not quite so bleak!
- \* Automated macromodel generation **is** difficult

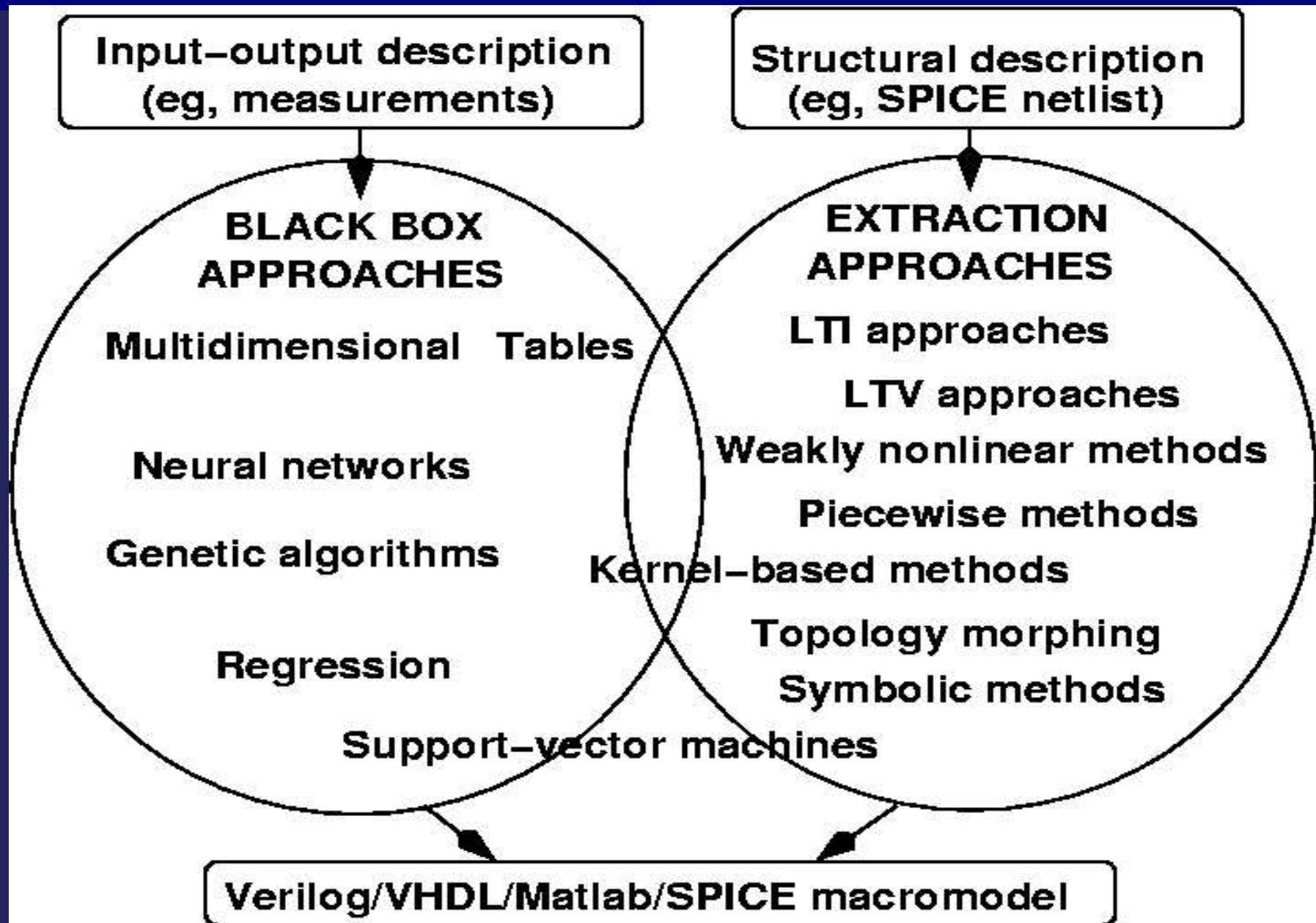
# Why Difficult?



# Approaches to Macromodelling

- \* **Black box** problems
  - \* samples of input-output pairs
  - \* from measurement and/or simulation
  - \* paucity of information
- \* **Extraction (bottom-up reduction)** problems
  - \* detailed circuit/simulation info available
  - \* eg, SPICE netlist: differential equations
  - \* surfeit of information
  - \* potential for better macromodels

# Automated MM Approaches



# “Algorithmic” Macromodelling Approaches

- \* Mathematical algorithms based on theory
- \* Provably preserve some useful property
  - \* eg, moments of transfer function
- \* AWE: first prominent method (LTI)
- \* Variety of nonlinear, LTV methods
- \* Generally applicable (eg, multi-physics)
  - \* not specific to a particular type of circuit/topology
  - \* eg: same oscillator MM technique works as well for a Colpitts oscillator as for a DFB laser or a grandfather clock

# Types of Algorithmic Macromodelling

- \* **Linear Time Invariant (LTI)**

- \* application: interconnect (delay, crosstalk)
- \* AWE, PVL, PRIMA, TBR

- \* **Linear Time Varying (LTV)**

- \* mixers, sampling/switching circuits, (oscillators)

- \* **Weakly nonlinear (Volterra)**

- \* companding ckts, amplifier/mixer gain compression

- \* **Strongly nonlinear (stable)**

- \* everything else: comparators, switching, slewing, ...

- \* **Autonomous**

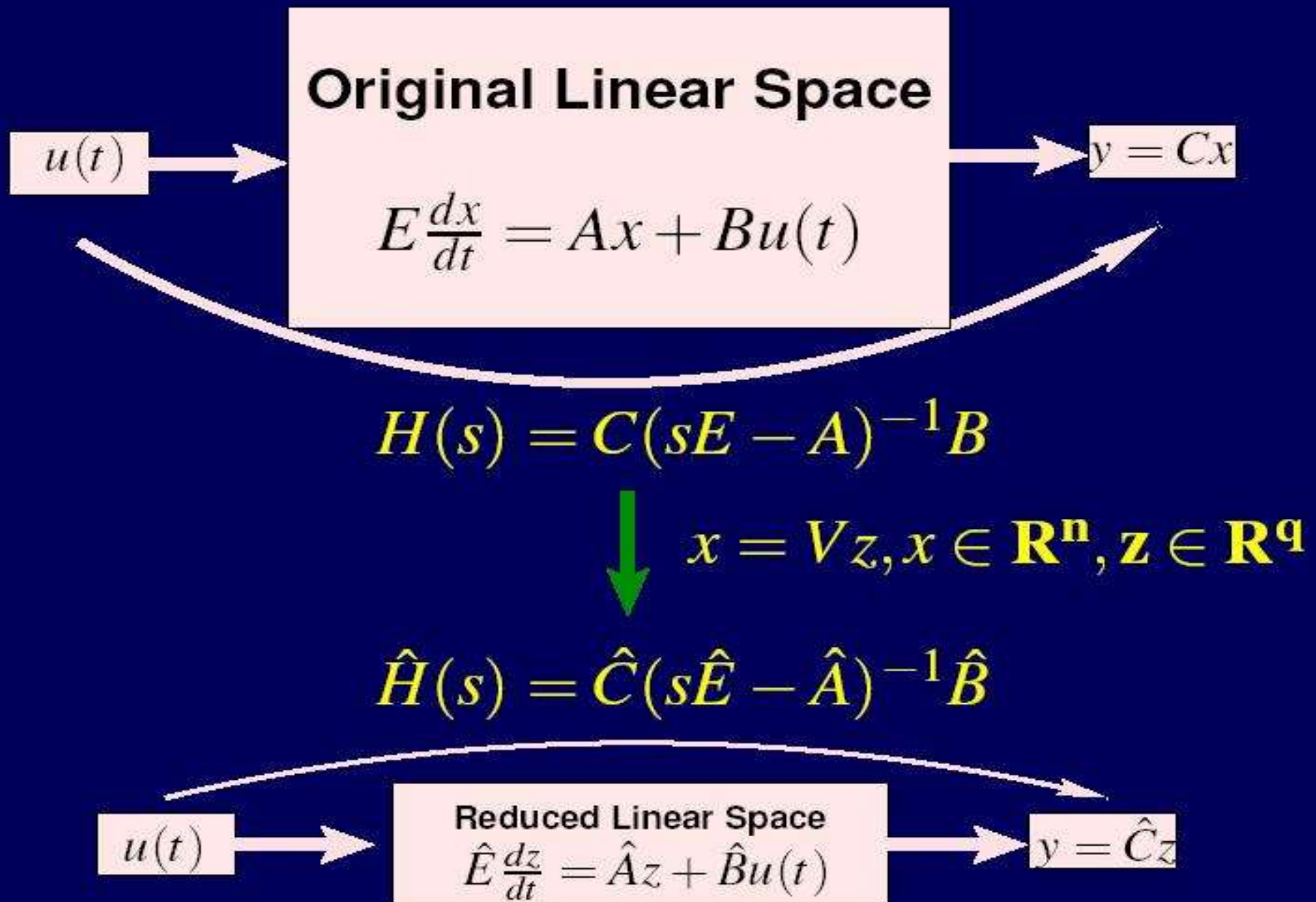
- \* oscillators, PLLs, etc.: marginally stable

# Linear Time Invariant Systems

- \* What is LTI?
  - \* Scale input waveform  $\Rightarrow$  scale output waveform
  - \* Time-shift input  $\Rightarrow$  time-shift output
- \* **Interconnect**, “linear” circuit elements
- \* **Well understood**: 50 years of theory
  - \* Laplace transforms, LTI ODEs, controllability/observability, ...
- \* Powers hand analysis by most designers



# LTI Macromodel Generation



# Asymptotic Waveform Evaluation

- \* AWE (Pillage/Rohrer ~1990)
- \* **Preserve moments** of LTI transfer function
  - \* frequency-domain xfer-fn derivatives
- \* **Explicit moment matching**
  - \* calculate moments of original system
  - \* run a Pade approximation: **small rational function**
  - \* **map to small dynamical system macromodel**

# LTI MM Accuracy/Stability

- \* **Increasing size does not increase accuracy**
  - \* explicit moment generation, Toeplitz-matrix based calculation numerically ill-conditioned
- \* **Implicit moment matching:** Krylov subspace methods
  - \* don't calculate moments: generate related Krylov subspaces robustly (Lanczos/Arnoldi methods)
  - \* generate macromodels directly - **moments matched implicitly**
- \* **Pade-via-Lanczos (PVL)**
  - \* Feldmann/Freund 1994/5

# LTI MM Summary

- \* **Important features**

- \* Accuracy vs size tradeoff
- \* MM scalability
- \* MM passivity

- \* **Computational properties**

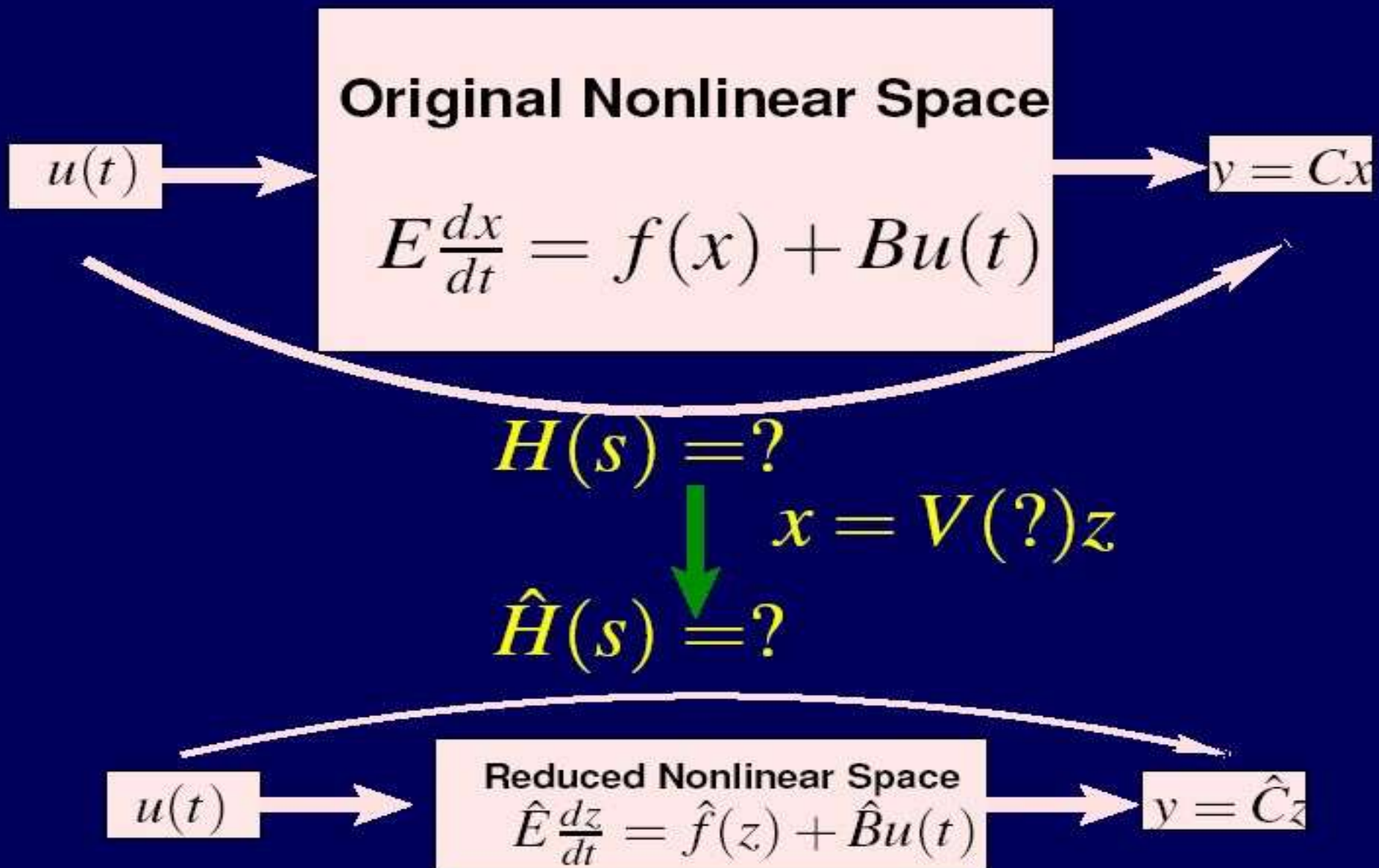
- \* AWE, Krylov methods linear with system size
- \* TBR methods cubic (but new results from Joel!)

- \* **Relatively mature and practically usable**

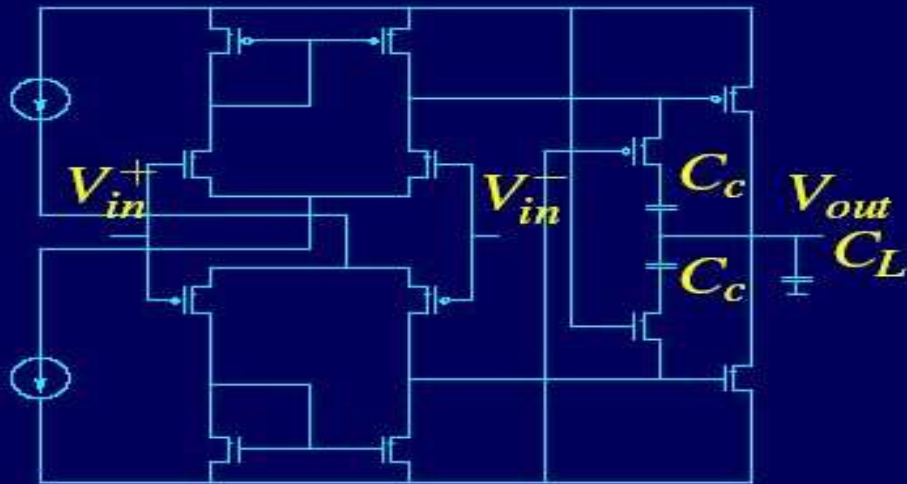
- \* **Basis for nonlinear approaches**

# Macromodelling Nonlinear Systems

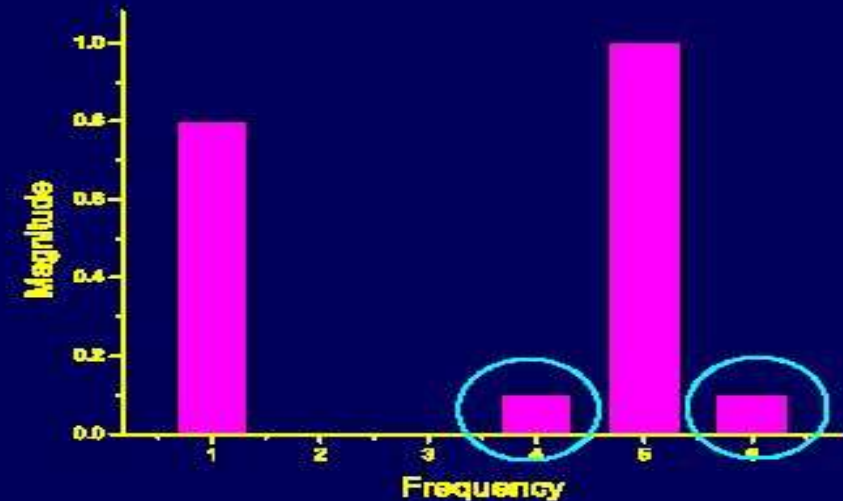
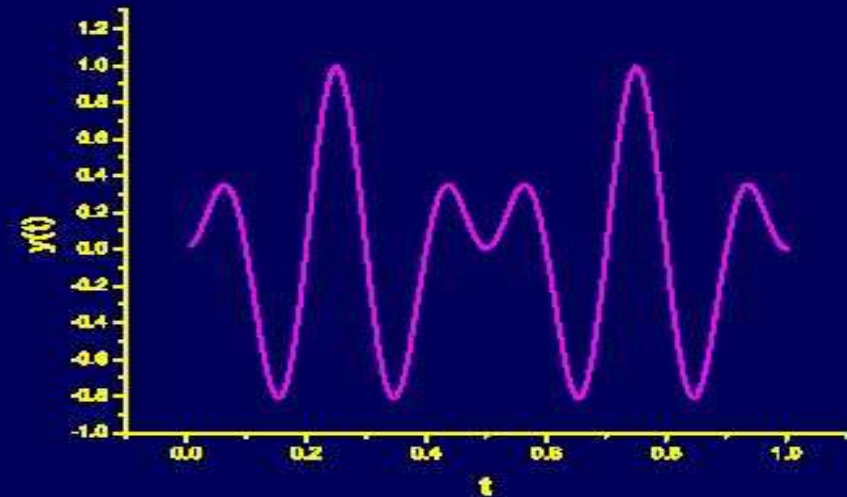
# Nonlinear Macromodel Generation



# Weakly Nonlinear Systems



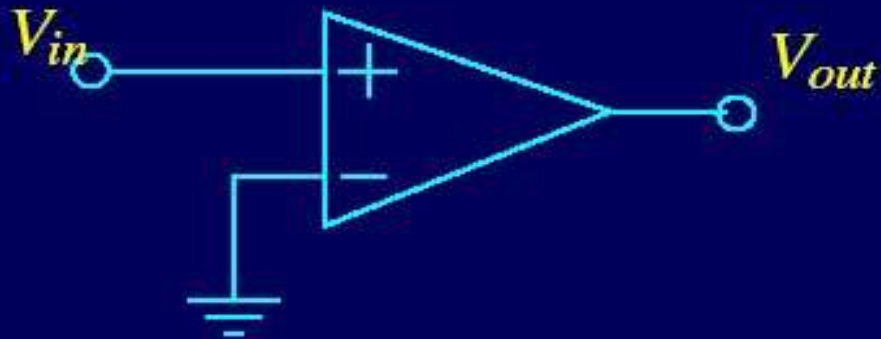
- Distortion, IM important!
- Must capture small distortion/IM



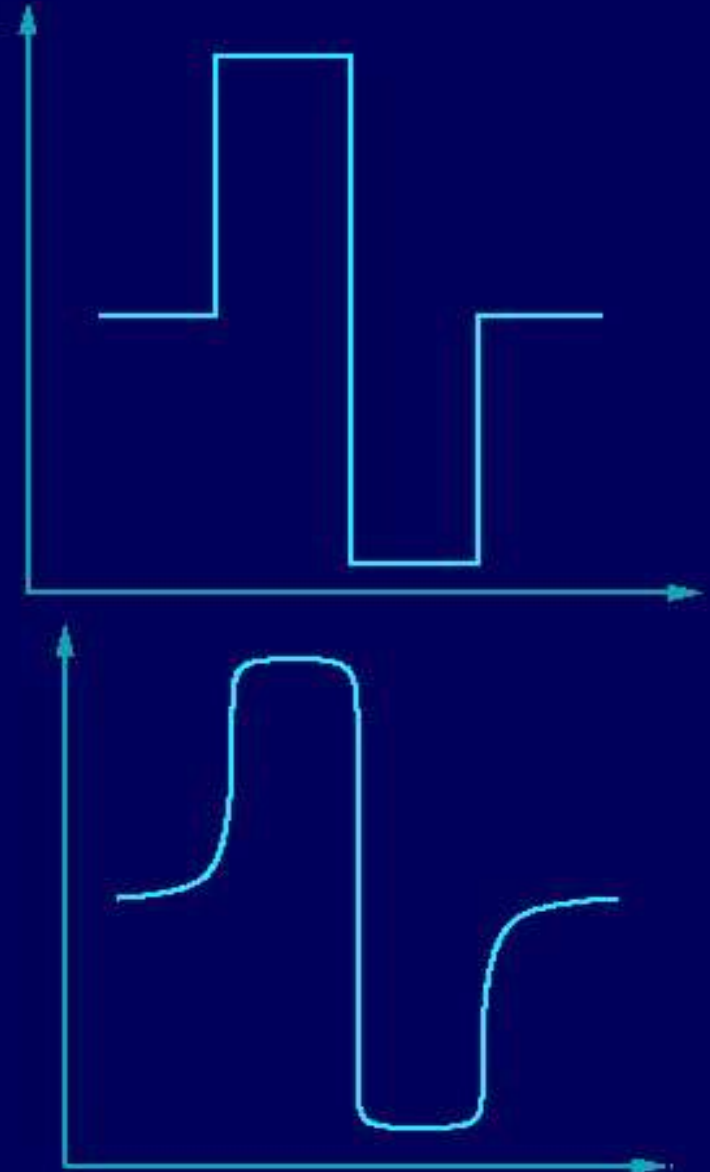


# Strong Nonlinearities

Comparators, switching mixers

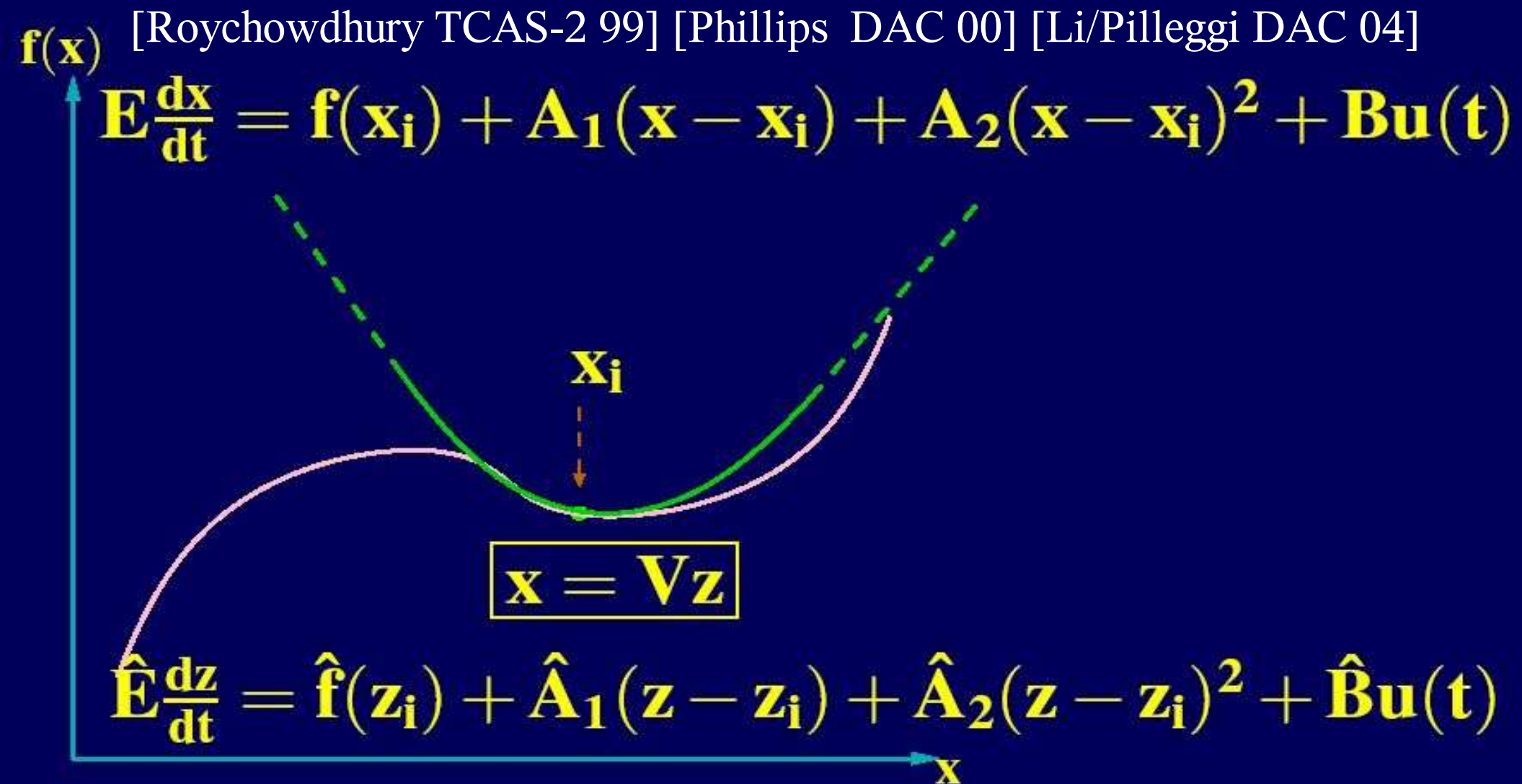


- Large signal clipping
- Must capture strong nonlinearities



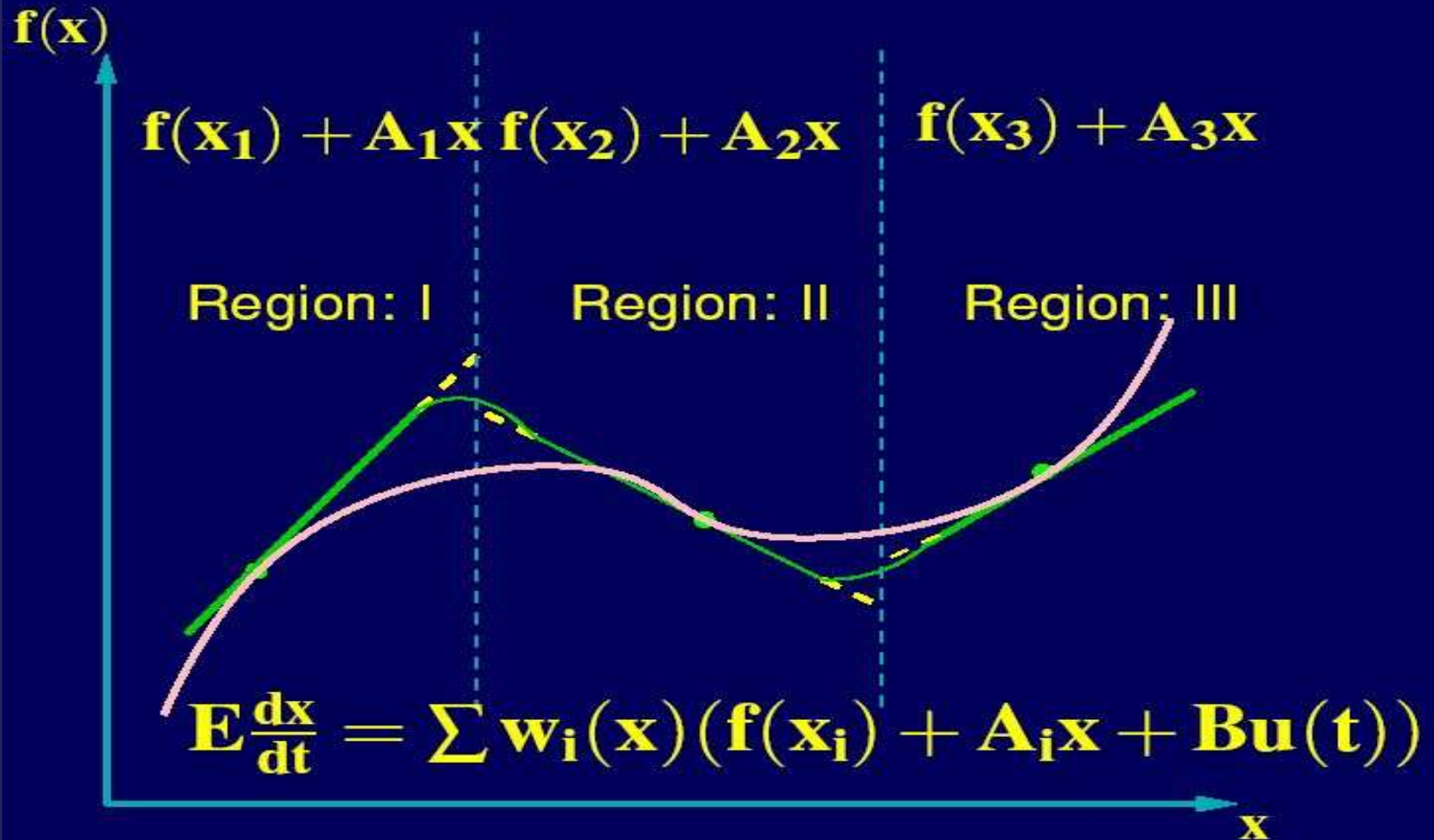


# Polynomial (Volterra) Reduction



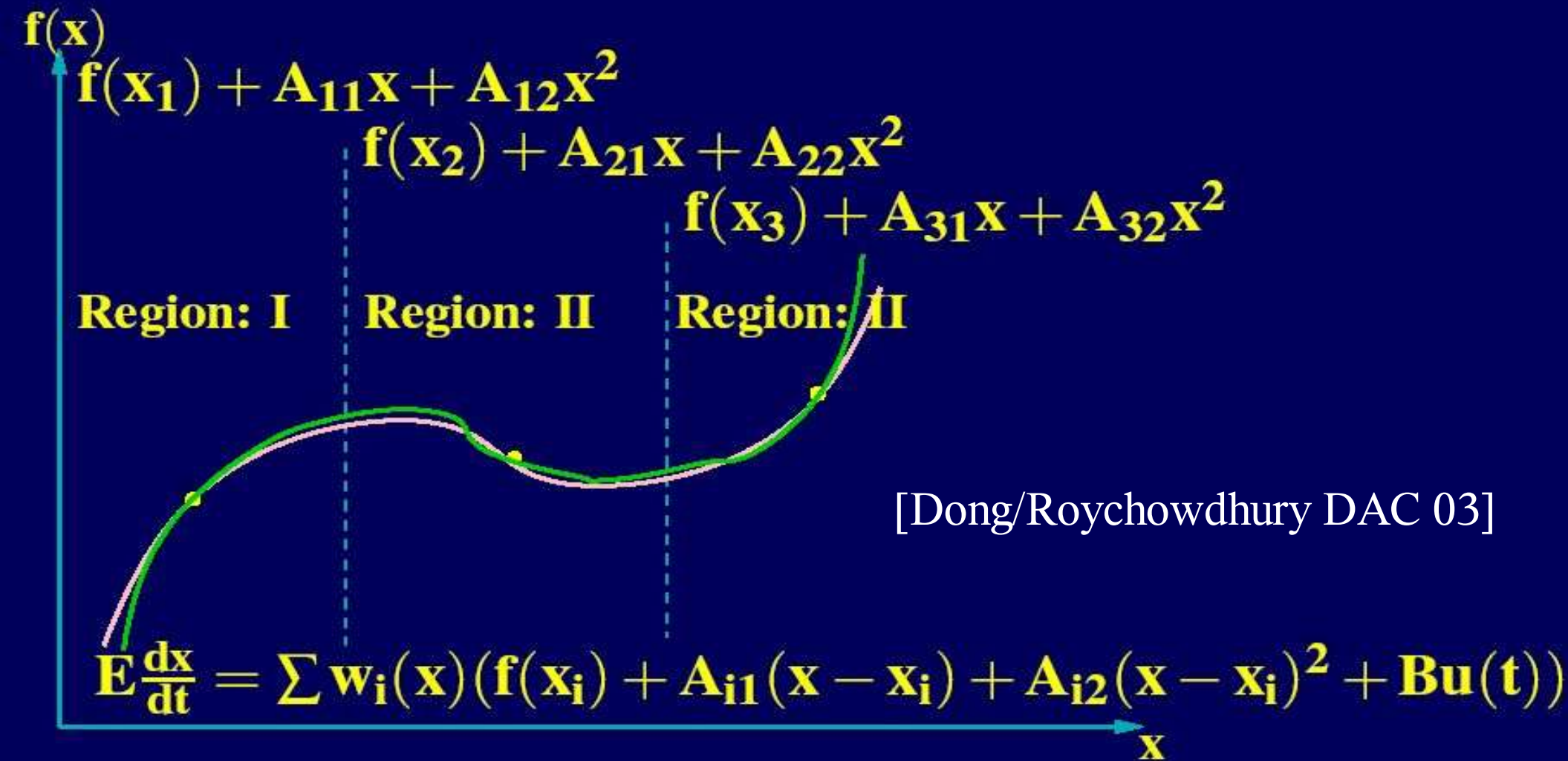
- Good for small distortion, Poor for large swing

# Trajectory Piecewise Linear MM



[Rewiński/White ICCAD 01]

# Piecewise Polynomial Macromodelling

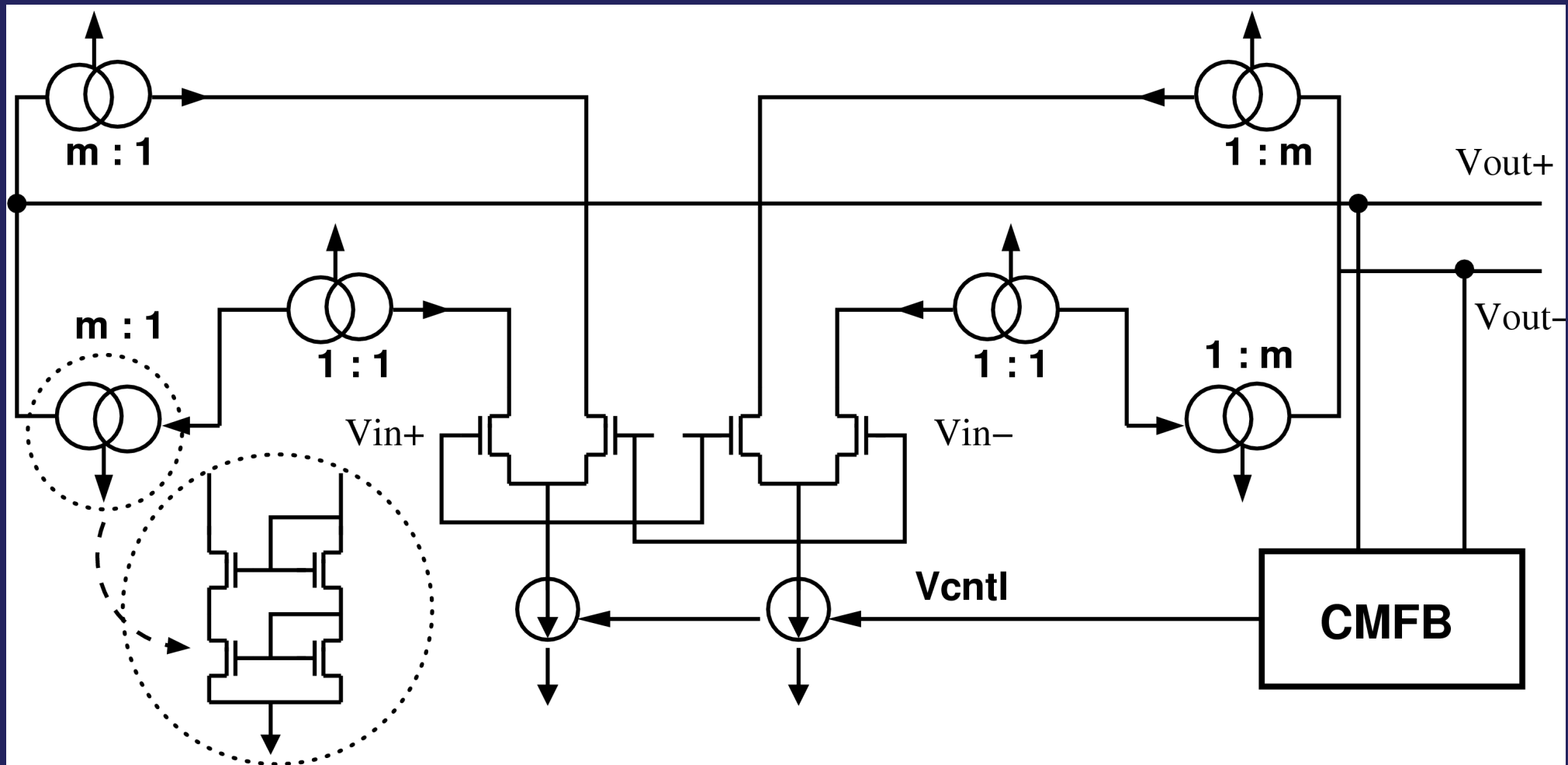


- Good for small distortion, also good for large swing

# Drop-in Replacement Macromodels

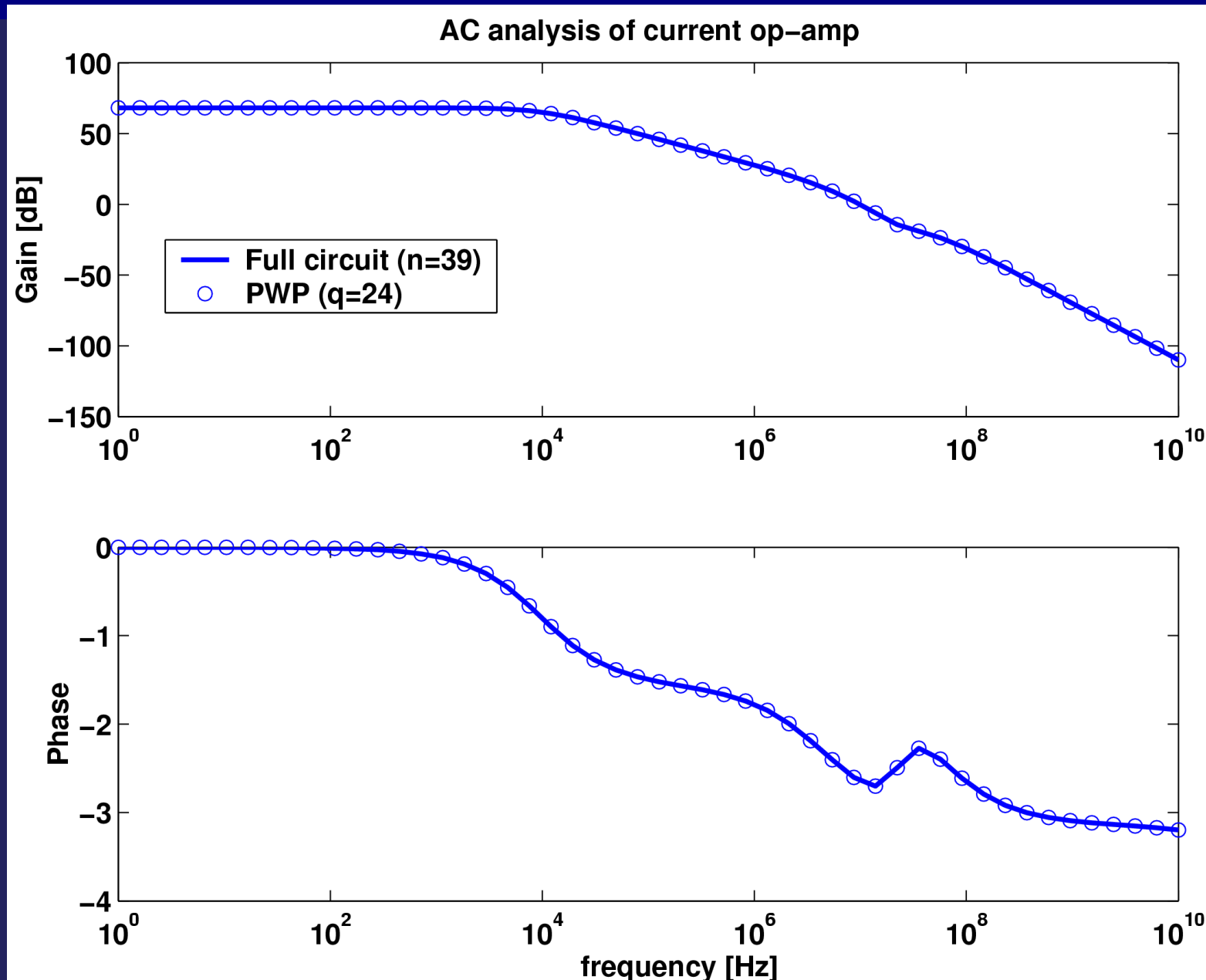
- \* Design process typically runs many simulations
  - \* DC (sweep), AC, small-signal distortion, transient
  - \* time- and frequency-domain analyses
- \* Would like one extracted macromodel to work for all analyses
  - \* ie, a drop-in replacement for the original
  - \* PWP-generated macromodels: good candidates

# Current Mirror Op-Amp

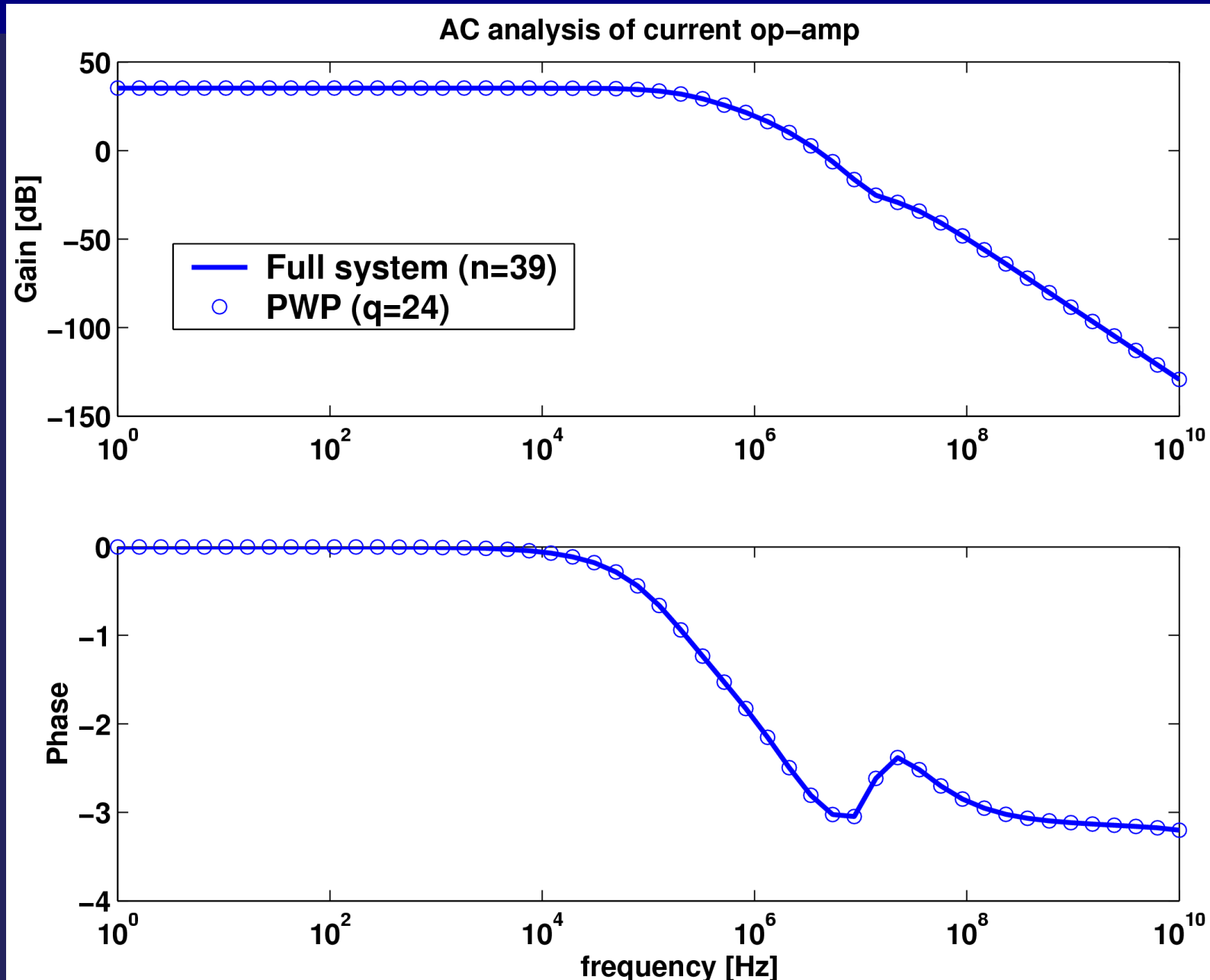


Original: ~50 MOSFETs. Macromodel size: 19, 27 regions

# Macromodel vs Original: AC Sweep 1

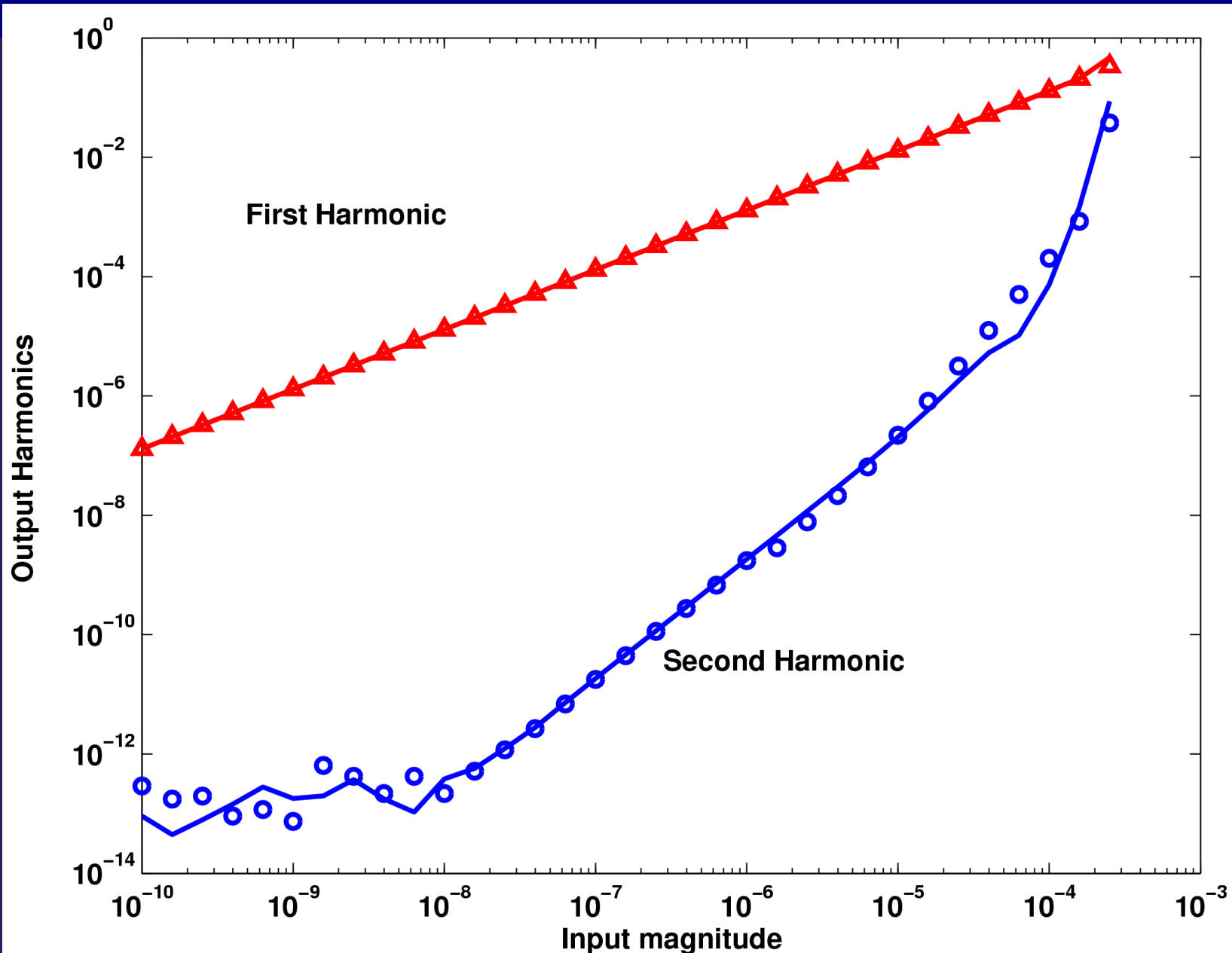


# Macromodel vs Original: AC Sweep 2



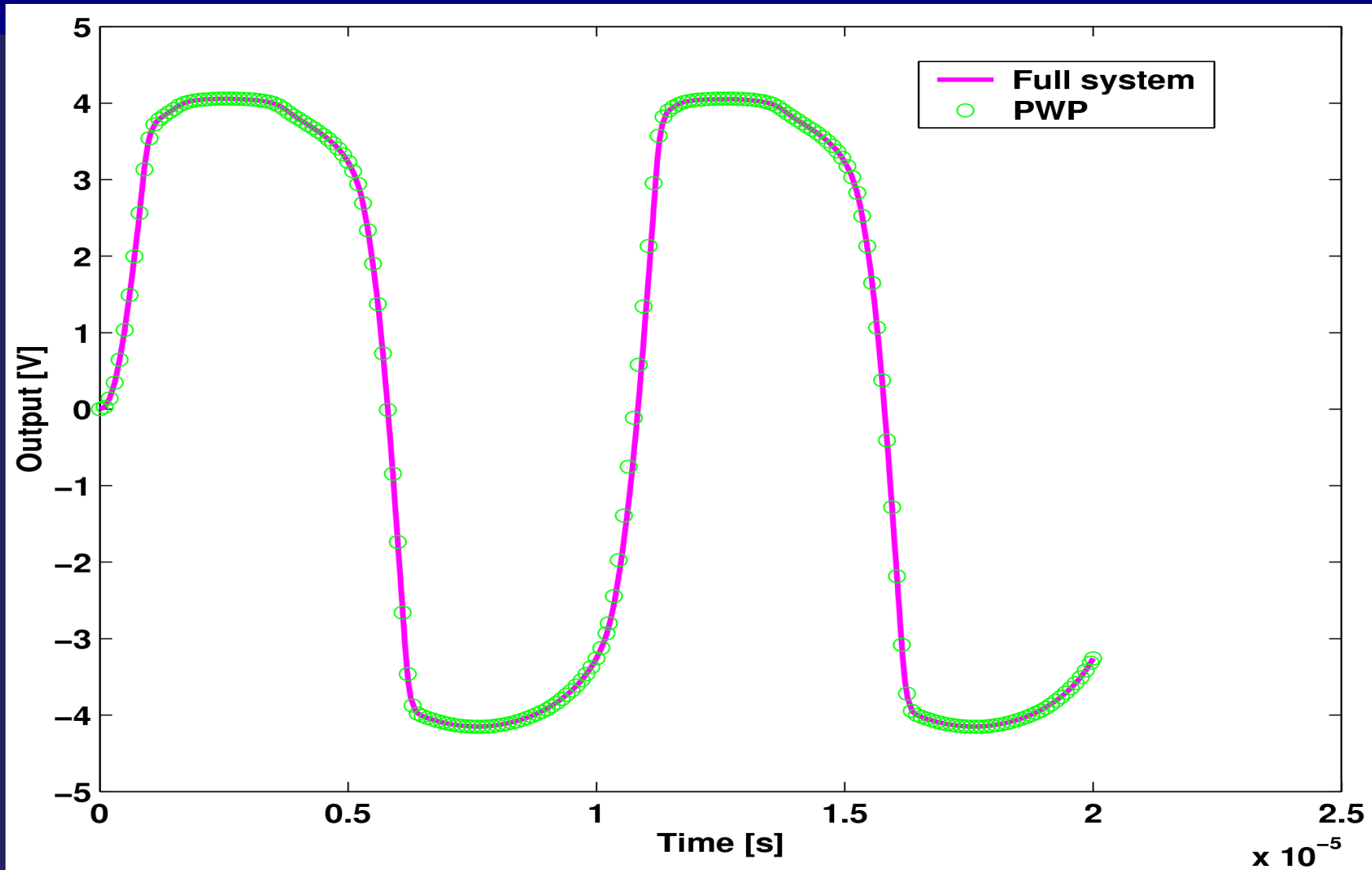


# Macromodel vs Original: Distortion



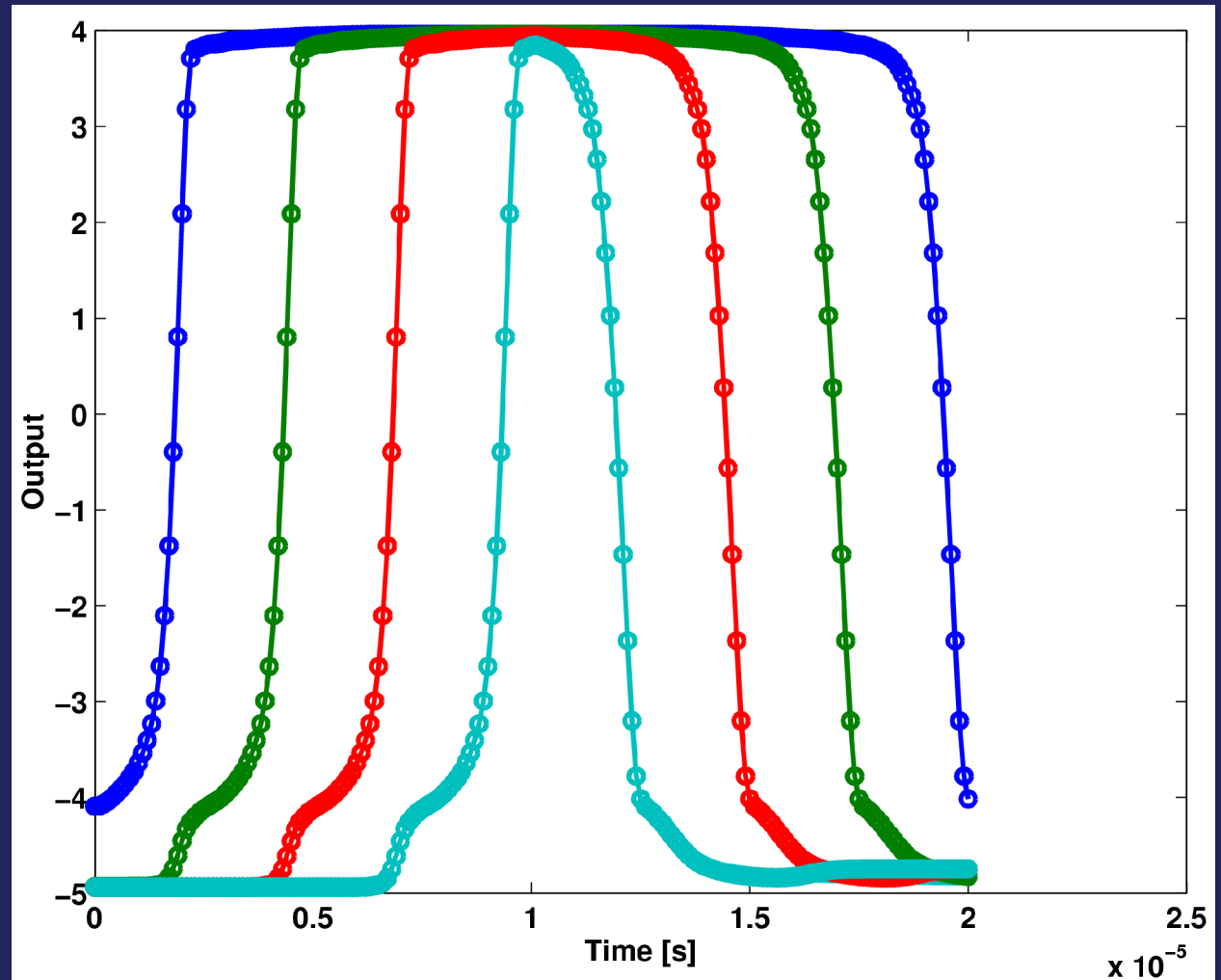
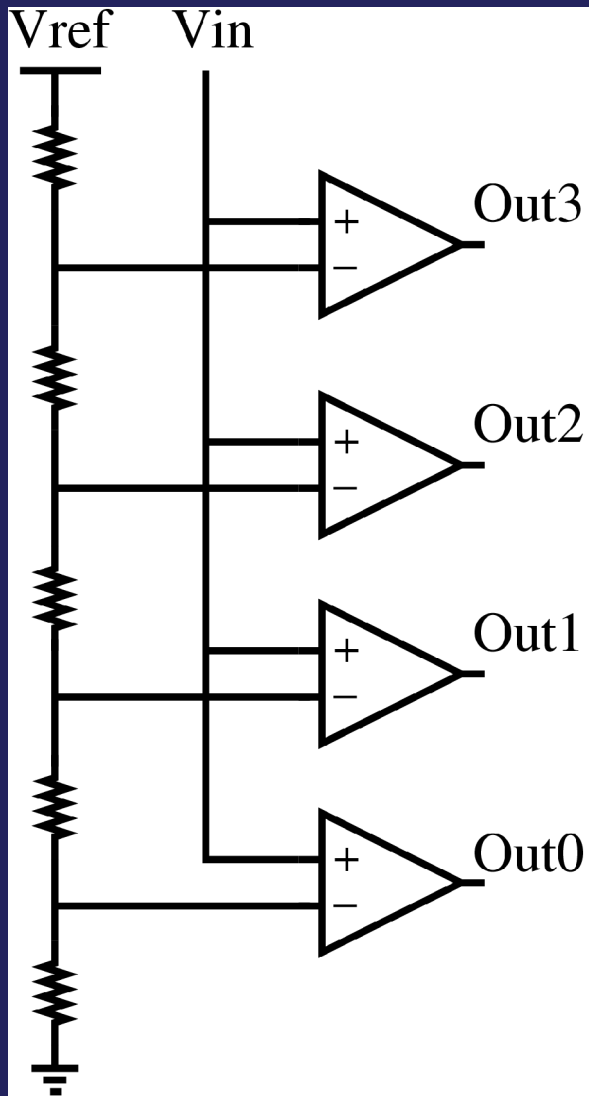


# Macromodel vs Original: Transient



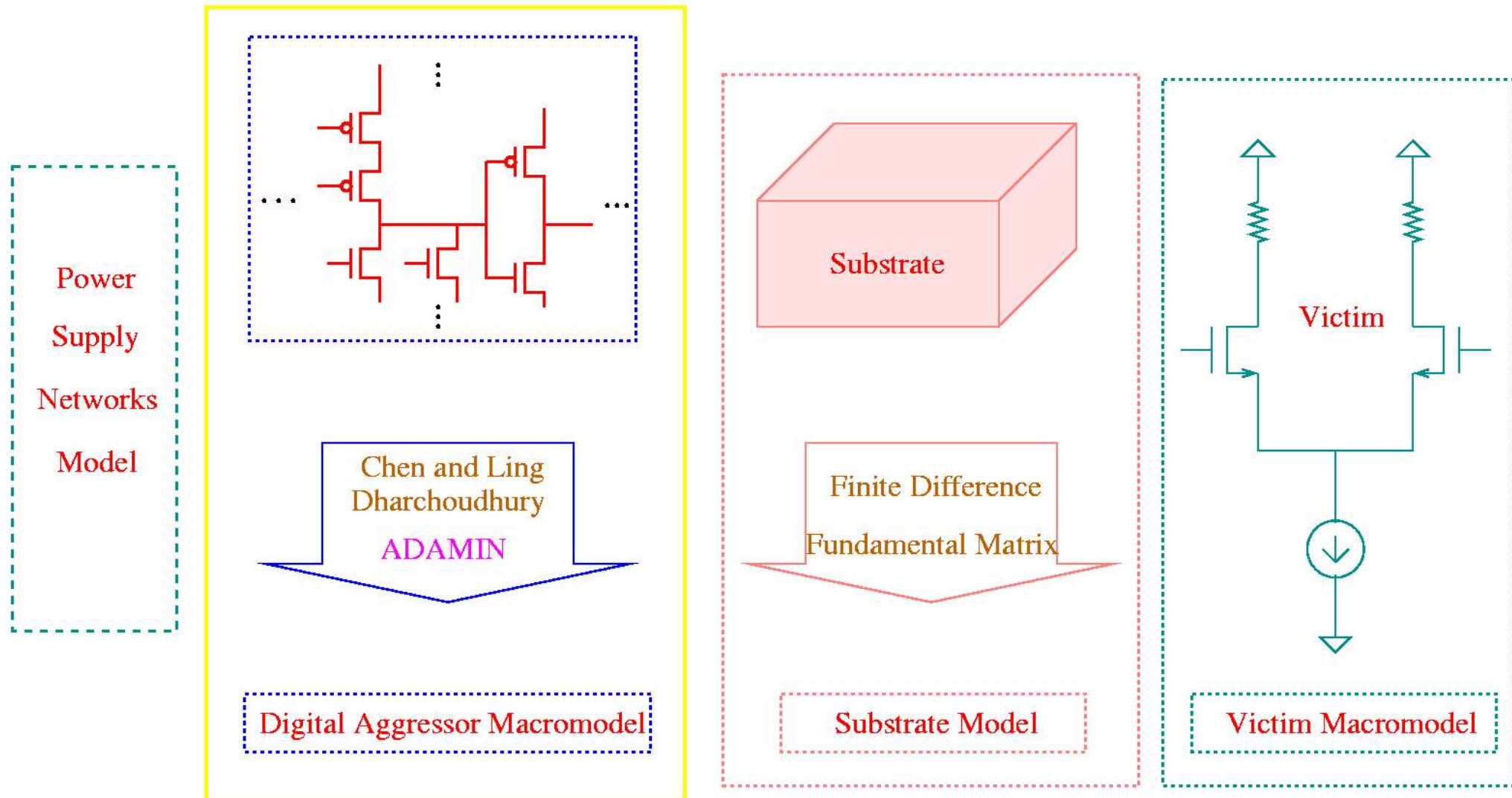
- *speedup: 41x over original*

# Small ADC: System-level Simulation

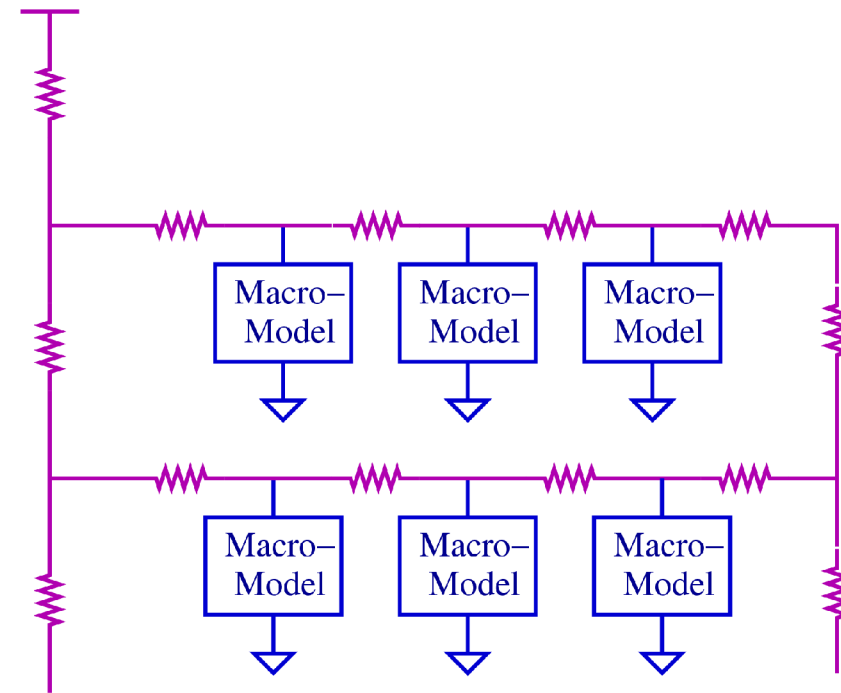
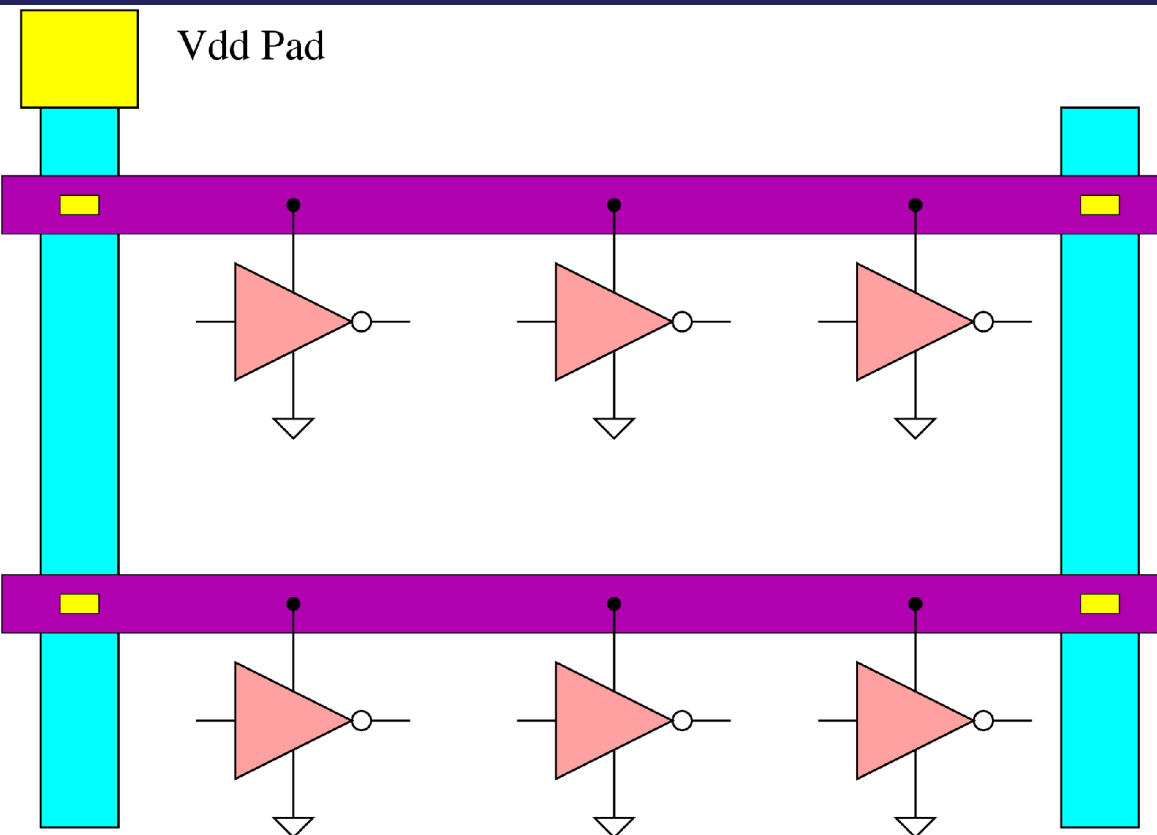


Excellent match between original and macromodel

# The Mixed Signal Interference Noise Problem



# Macromodelling Digital Noise Injection



- \* Interested **only in interference injection characteristics**
  - \* supply and substrate injections: analog waveforms
  - \* digital signals: system time variation
  - \* **LTV model captures switching behaviour well!**

[Wang/Murgai/Roychowdhury DATE04]

# Linear Time-Varying Macromodelling

- \* Useful abstraction for some nonlinear systems
  - \* mixers, switching filters, samplers, DC/DC converters
  - \* leverage LTI methods
  - \* frequency translation, sampling captured
  - \* signal-path nonlinearities **not** captured
- \* Input-output relationship linear
  - \* but **not** time-invariant

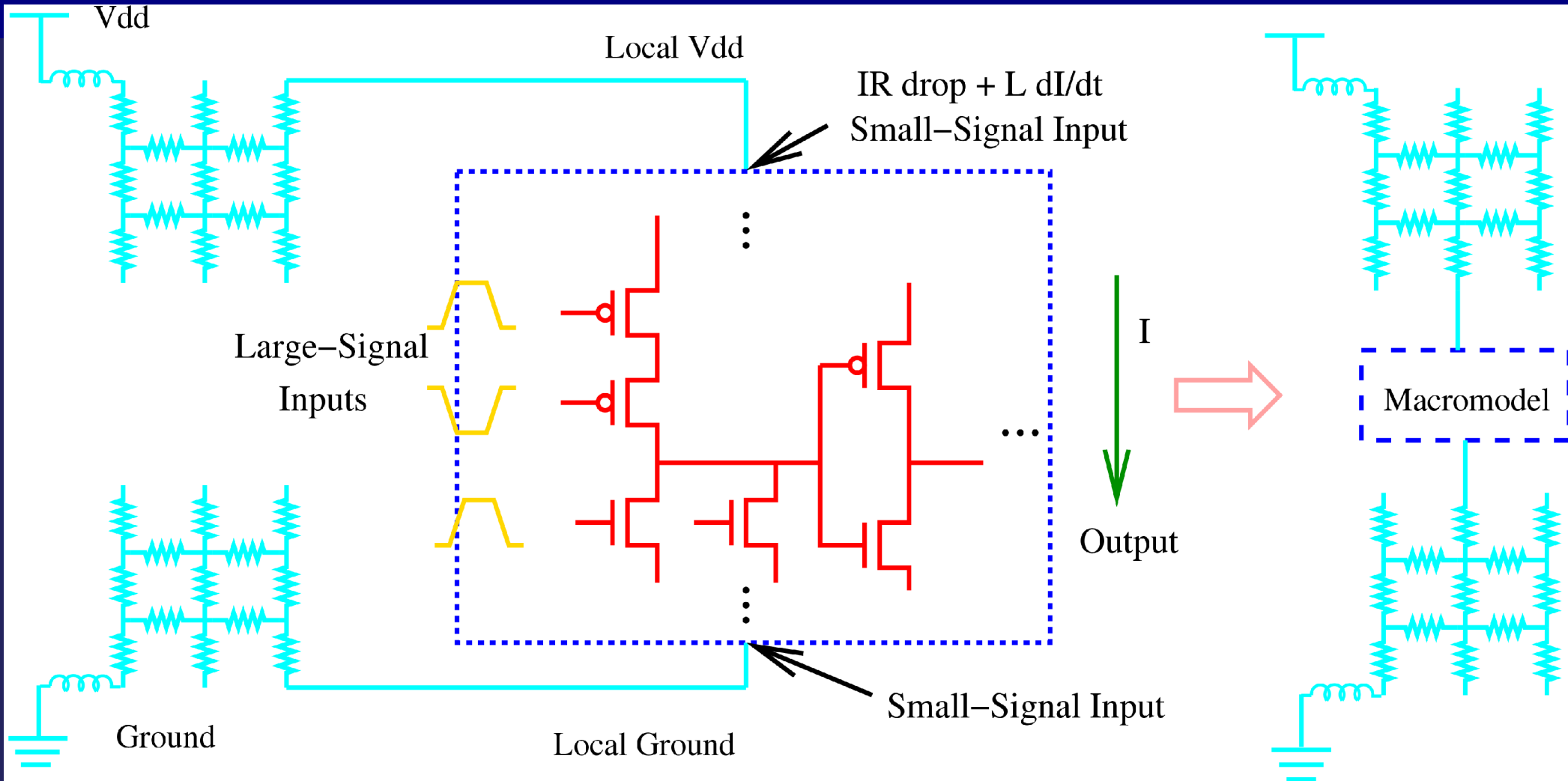
# Basic Principles of LTV Macromodelling

- \* LTV systems have time-varying (but linear) transfer functions
  - \* LTI transfer function  $H(s)$  -->  $H(t,s)$ 
    - \*  $t$  captures **system time** variation
    - \*  $s$  captures **input/output** time variations
  - \* computationally useful form of  $H(t,s)$ :
    - \* linear matrices  $C(t)$  and  $G(t)$ : from transient simulation/steady state calculations

$$H(t_1, s) = d^T \left( \frac{\partial}{\partial t_1} C(t_1) + sC(t_1) + G(t_1) \right)^{-1} [b]$$

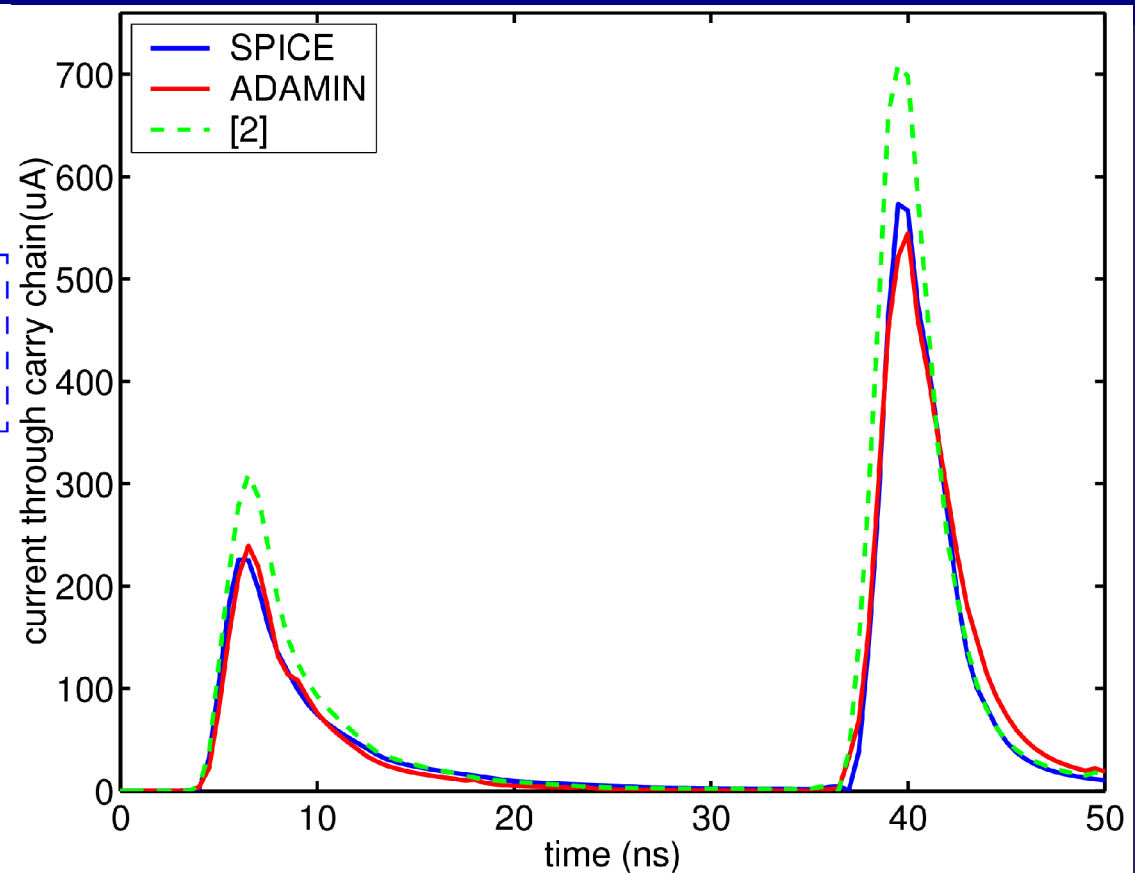
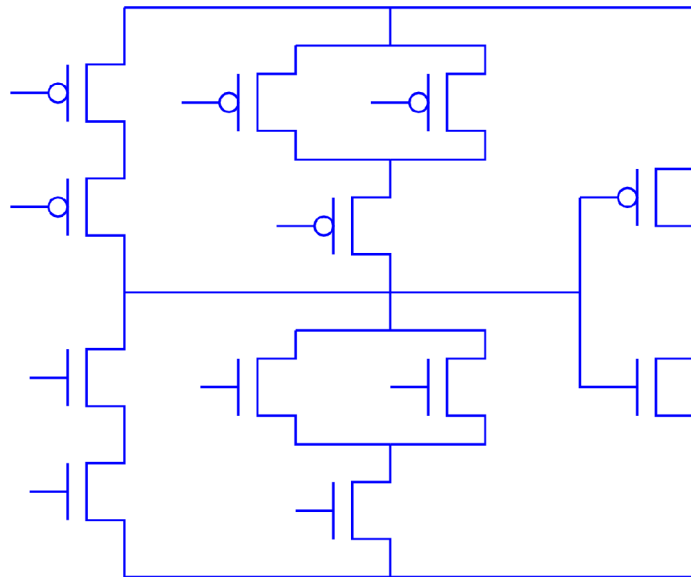
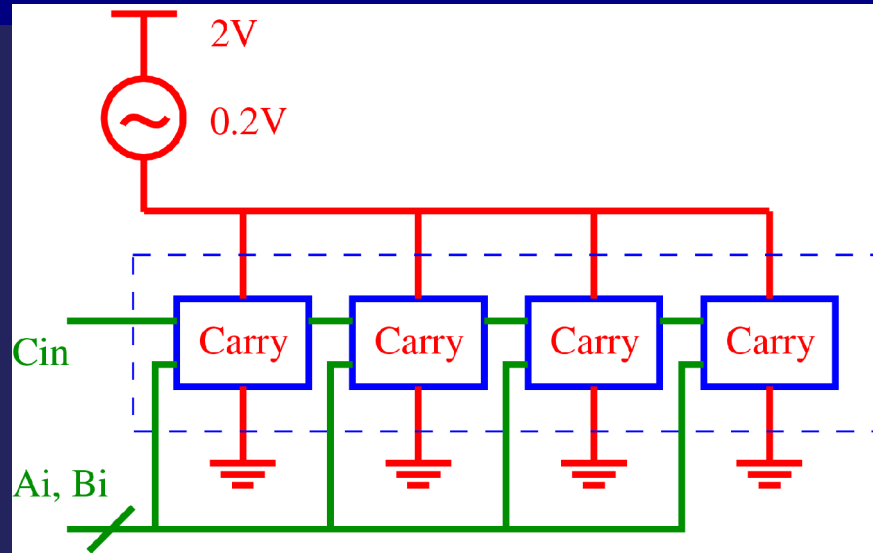
[Roychowdhury, Phillips ICCAD 98] [Roychowdhury TCAS-2 99]

# LTV Digital Aggressor Macromodels



- \* Input: (eg) power supply voltage variation
- \* Output: resulting current variation
- \* MM: small time-varying system relating input to output

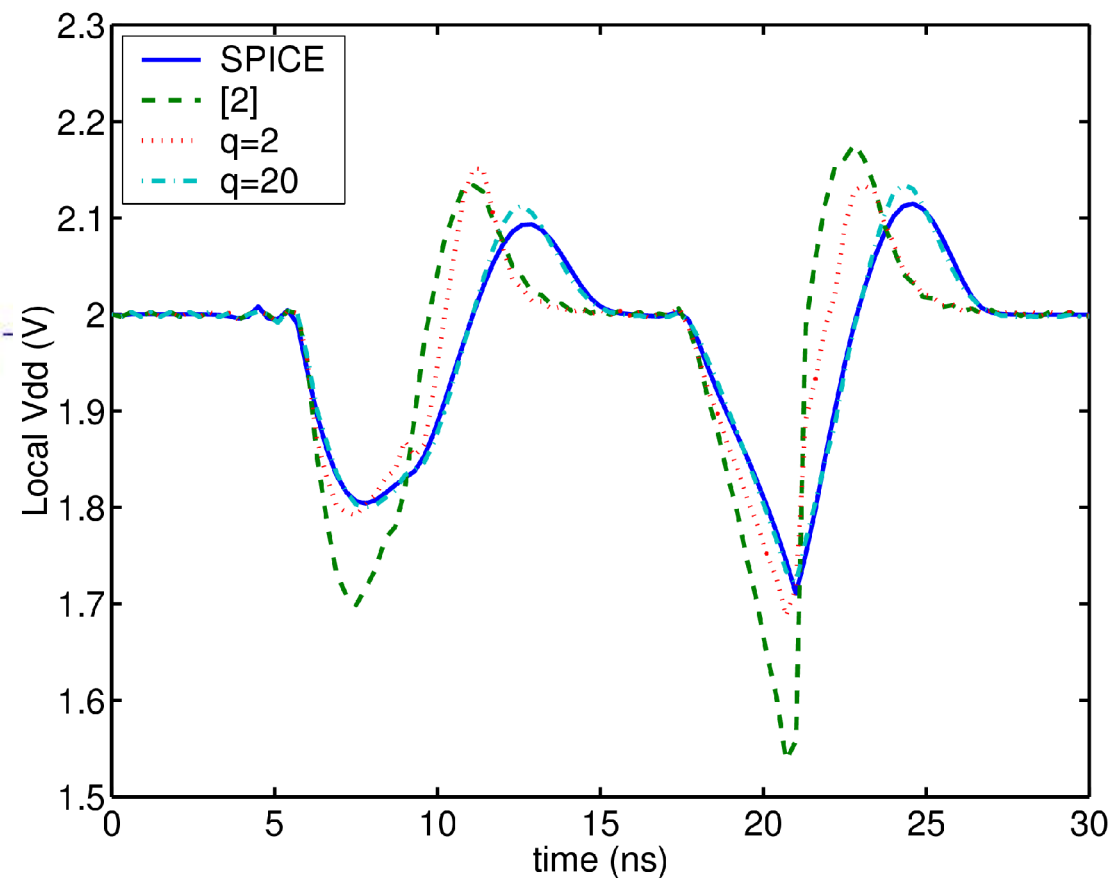
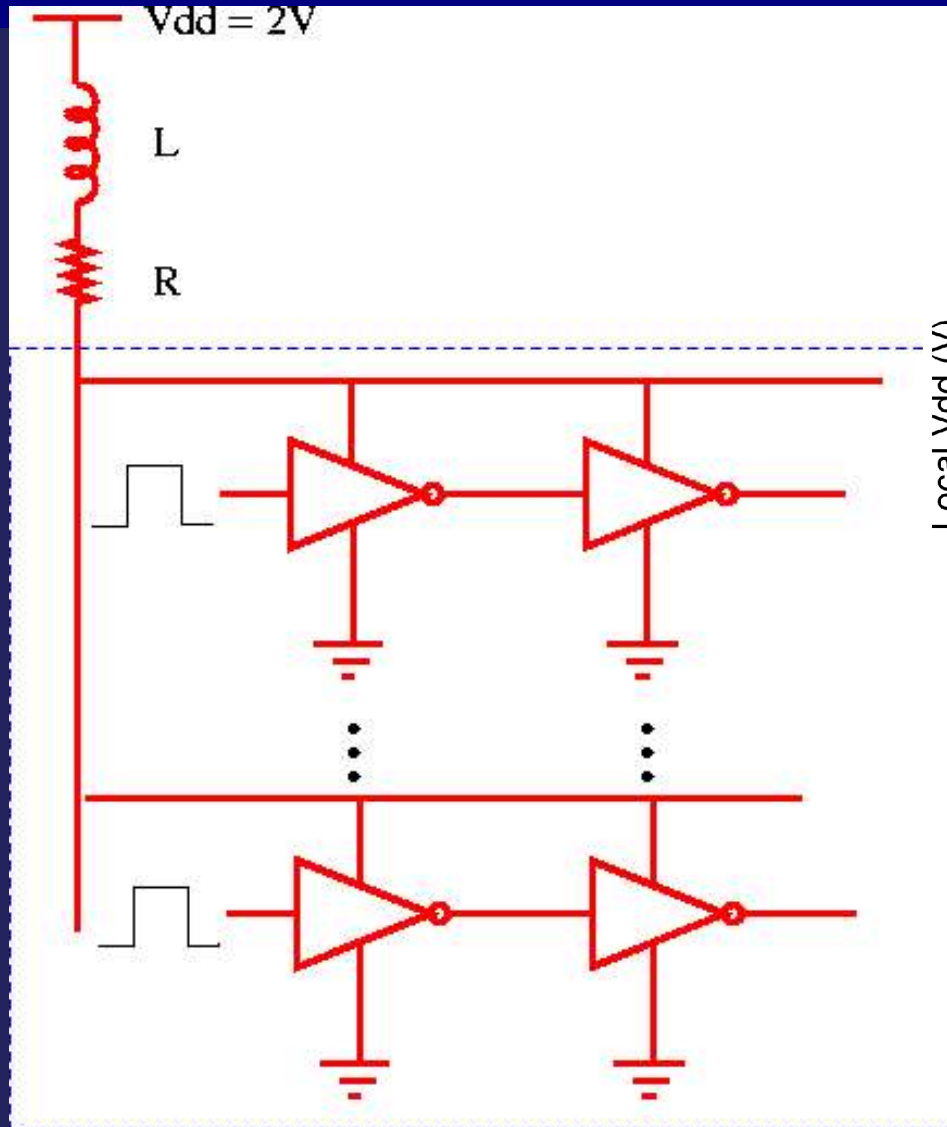
# Supply-noise Induced Currents: Carry Chain



- \* Considerable accuracy improvement over current-source macromodel [Dharchoudhury 98]



# “System” Simulation with Inductive Supply



- \* size 2 MM: already better
- \* size 20: much more accurate
- \* 17000x speedup for 8000 gates

# Macromodelling Oscillatory Systems

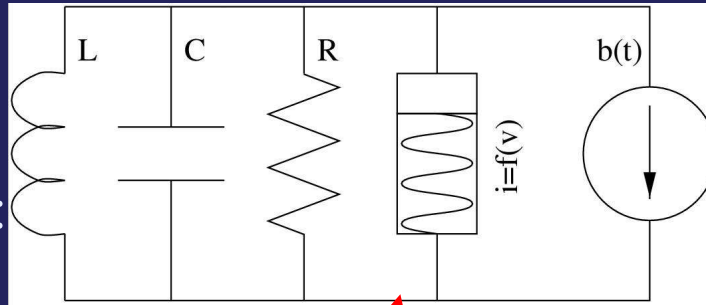
# Oscillators

- Oscillators are critical in communication systems:

- LC oscillators
- Ring oscillators

- Used everywhere:

- VCOs, PLLs
- CDR ckts
- synchronization loops

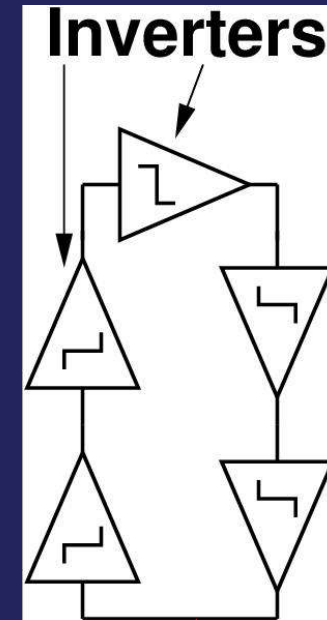


**-ve feedback  
LC oscillator**

- **Very slow** to simulate
- **Noise prediction problematic**

- Needed:

- Accurate/fast oscillator macromodelling capability
- Accurate oscillator jitter/phase noise prediction

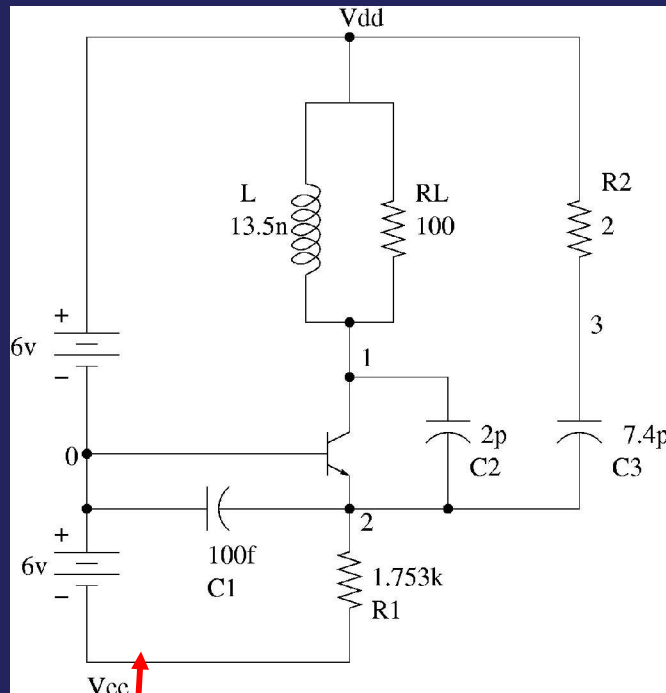


**Ring  
oscillator**

# Why Oscillators are a Special Simulation Challenge

- **Computation/size/accuracy**: much greater than amps/mixers
- Even 1-transistor oscillators (eg, UHF oscs, >100GHz)
  - long startups, tiny timesteps needed
- On-chip RF: 100s to 1000s of transistors
  - VERY challenging to simulate
- Macromodelling offers dramatic speedup
  - Even for 1-transistor oscillator
- Oscillators feature complex phenomena: **injection locking**
  - oscillator's frequency “locks” to frequency of external input
  - if frequencies close enough, even if input is very small
  - can take extremely long to simulate
  - universal phenomenon: grandfather clocks, fireflies flashing, etc

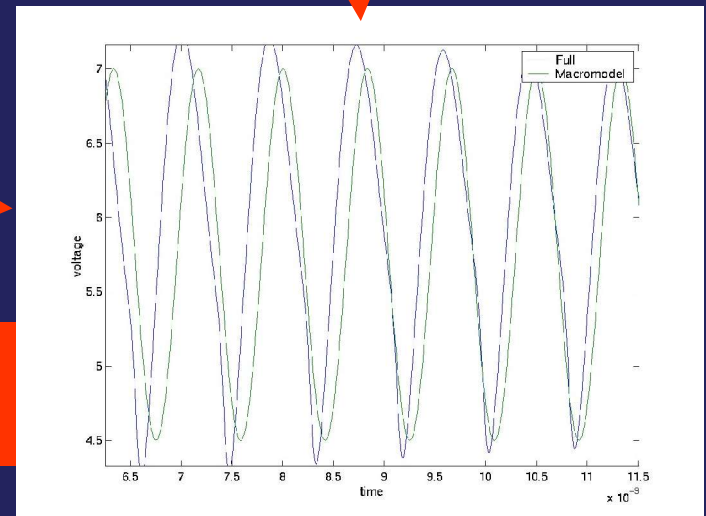
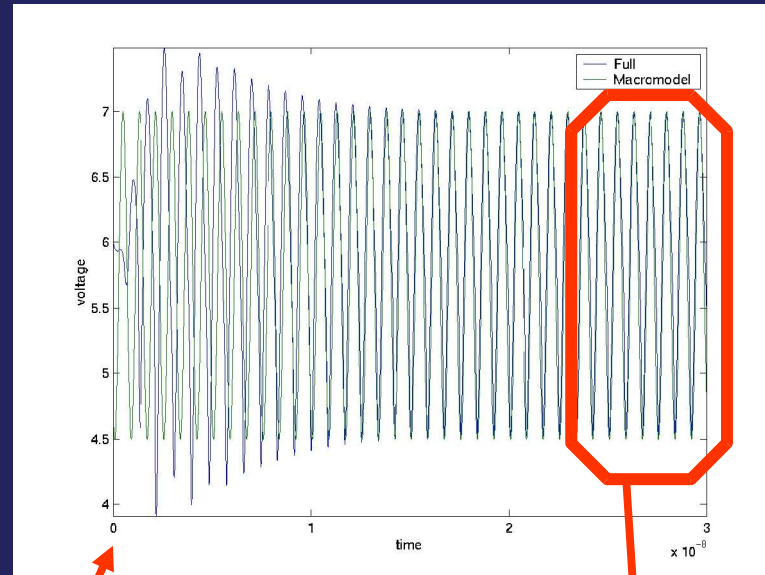
# Capturing Injection Locking in a Colpitts (LC) Oscillator



**Colpitts  
oscillator**

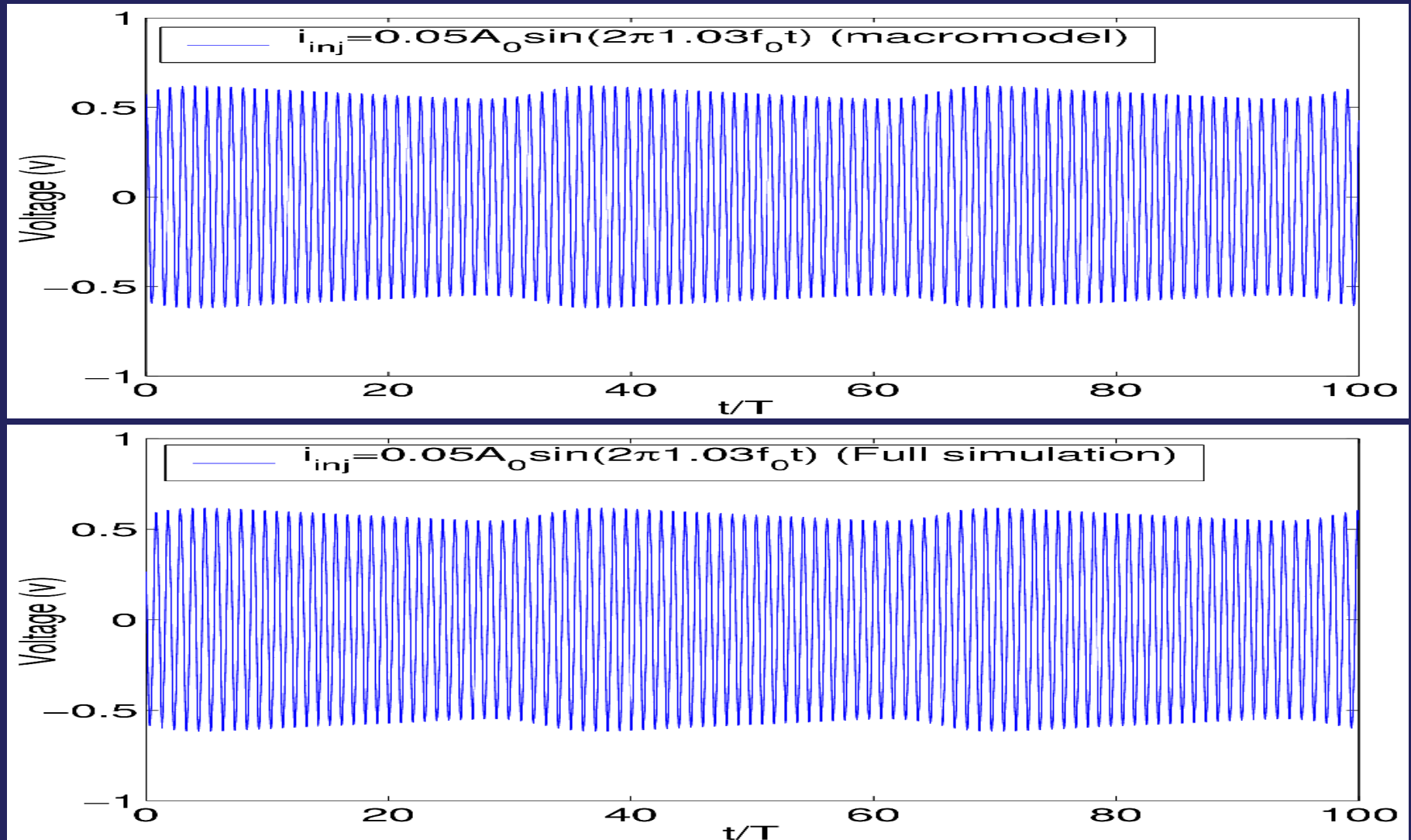
**Full simulation  
vs macromodel**

- 89x speedup over original



# Capturing Amplitude Changes

- nonlinear phase + amplitude components (via LTV reduction)*

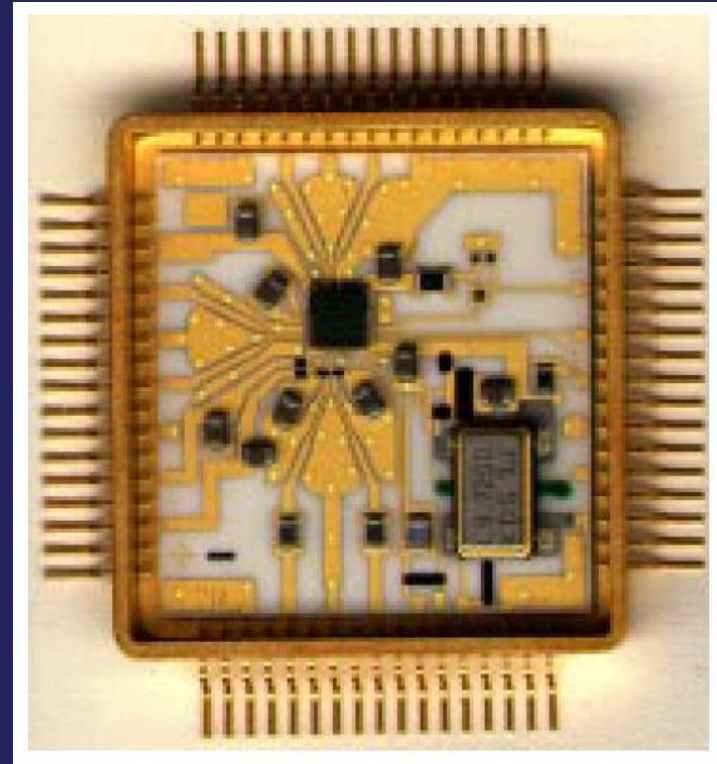


# PLLs: Commodity, High-Margin



Rick Walker, HP/Agilent

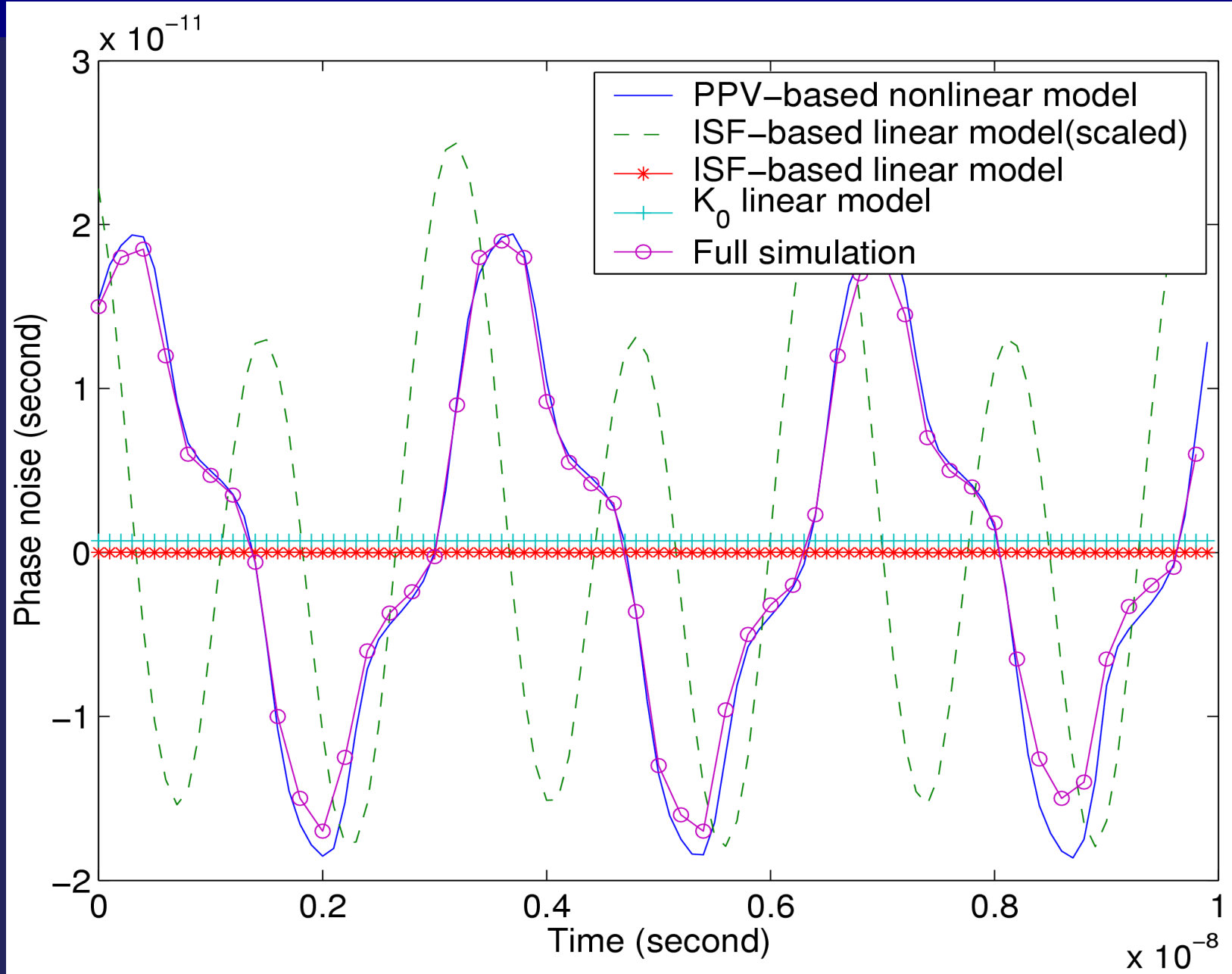
1.25Gbps Ethernet xcvr  
<< \$10



Rick Walker, HP/Agilent

2.44Gbps SONET CDR  
\$500

# PLL Phase Response to Periodic Supply Noise





# Conclusion

- Coming “soon”: automated nonlinear macromodelling for...
  - op-amps, mixers, switching filters, comparators
  - oscillators, VCOs (any kind: LC/ring/relaxation/etc)
  - large digital aggressor blocks: interference macromodels
  - bottom-up “extracted” macromodels: much more accurate, second-order effects, ...
- Use of core macromodels for system simulation
  - ADCs/DACs/Sigma-Deltas
  - PLLs/Sigma-Deltas (incl jitter and noise)
  - SOC
  - MATLAB/Simulink/Verilog-A/VHDL-AMS/etc
- Automated Macromodelling: the ONLY sustainable methodology for effective CAD support of nano-era analog, mixed-signal and RF design