

Assignment 2 (ML for TS) - MVA

Firstname Lastname youremail1@mail.com

Firstname Lastname youremail2@mail.com

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1 Introduction

Objective. The goal is to better understand the properties of AR and MA processes and do signal denoising with sparse coding.

Warning and advice.

- Use code from the tutorials as well as from other sources. Do not code yourself well-known procedures (e.g., cross-validation or k-means); use an existing implementation.
- The associated notebook contains some hints and several helper functions.
- Be concise. Answers are not expected to be longer than a few sentences (omitting calculations).

Instructions.

- Fill in your names and emails at the top of the document.
- Hand in your report (one per pair of students) by Sunday 7th December 11:59 PM.
- Rename your report and notebook as follows:
FirstnameLastname1_FirstnameLastname1.pdf and
FirstnameLastname2_FirstnameLastname2.ipynb.
For instance, LaurentOudre_ValerioGuerrini.pdf.
- Upload your report (PDF file) and notebook (IPYNB file) using this link:
<https://forms.gle/J1pdeHspSs9zNfWAA>.

2 General questions

A time series $\{y_t\}_t$ is a single realisation of a random process $\{Y_t\}_t$ defined on the probability space (Ω, \mathcal{F}, P) , i.e. $y_t = Y_t(w)$ for a given $w \in \Omega$. In classical statistics, several independent realizations are often needed to obtain a “good” estimate (meaning consistent) of the parameters of the process. However, thanks to a stationarity hypothesis and a “short-memory” hypothesis, it is still possible to make “good” estimates. The following question illustrates this fact.

Question 1

An estimator $\hat{\theta}_n$ is consistent if it converges in probability when the number n of samples grows to ∞ to the true value $\theta \in \mathbb{R}$ of a parameter, i.e. $\hat{\theta}_n \xrightarrow{\mathcal{D}} \theta$.

- Recall the rate of convergence of the sample mean for i.i.d. random variables with finite variance.
- Let $\{Y_t\}_{t \geq 1}$ a wide-sense stationary process such that $\sum_k |\gamma(k)| < +\infty$. Show that the sample mean $\bar{Y}_n = (Y_1 + \dots + Y_n)/n$ is consistent and enjoys the same rate of convergence as the i.i.d. case. (Hint: bound $\mathbb{E}[(\bar{Y}_n - \mu)^2]$ with the $\gamma(k)$ and recall that convergence in L_2 implies convergence in probability.)

Answer 1

For i.i.d. random variables with finite variance, the variance of the sample mean is

$$\text{Var}(\bar{Y}_n) = \frac{\sigma^2}{n}$$

and the convergence rate is $1/\sqrt{n}$.

For a stationary process $\{Y_t\}$ with $\sum_k |\gamma(k)| < +\infty$,

$$\text{Var}(\bar{Y}_n) = \frac{1}{n^2} \sum_{j=1}^n \sum_{k=1}^n \text{Cov}(Y_j, Y_k) = \frac{1}{n^2} \sum_{j=1}^n \sum_{k=1}^n \gamma(j-k)$$

which can also be written as

$$\text{Var}(\bar{Y}_n) = \frac{1}{n} \sum_{h=-(n-1)}^{n-1} \left(1 - \frac{|h|}{n}\right) \gamma(h)$$

If $\sum_k |\gamma(k)| < +\infty$, then as $n \rightarrow \infty$,

$$\text{Var}(\bar{Y}_n) \rightarrow \frac{1}{n} \sum_{h=-\infty}^{\infty} \gamma(h)$$

so $\text{Var}(\bar{Y}_n) = O(1/n)$, i.e., the convergence rate is $1/\sqrt{n}$ like in the i.i.d. case.

3 AR and MA processes

Question 2 Infinite order moving average MA(∞)

Let $\{Y_t\}_{t \geq 0}$ be a random process defined by

$$Y_t = \varepsilon_t + \psi_1 \varepsilon_{t-1} + \psi_2 \varepsilon_{t-2} + \cdots = \sum_{k=0}^{\infty} \psi_k \varepsilon_{t-k} \quad (1)$$

where $(\psi_k)_{k \geq 0} \subset \mathbb{R}$ ($\psi_k = 1$) are square summable, i.e. $\sum_k \psi_k^2 < \infty$ and $\{\varepsilon_t\}_t$ is a zero mean white noise of variance σ_ε^2 . (Here, the infinite sum of random variables is the limit in L_2 of the partial sums.)

- Derive $\mathbb{E}(Y_t)$ and $\mathbb{E}(Y_t Y_{t-k})$. Is this process weakly stationary?
- Show that the power spectrum of $\{Y_t\}_t$ is $S(f) = \sigma_\varepsilon^2 |\phi(e^{-2\pi i f})|^2$ where $\phi(z) = \sum_j \psi_j z^j$. (Assume a sampling frequency of 1 Hz.)

The process $\{Y_t\}_t$ is a moving average of infinite order. Wold's theorem states that any weakly stationary process can be written as the sum of the deterministic process and a stochastic process which has the form (1).

Answer 2

First, since Y_t is a linear combination of zero mean white noises, we have

$$\mathbb{E}[Y_t] = \sum_{k=0}^{\infty} \psi_k \mathbb{E}[\varepsilon_{t-k}] = 0.$$

Now for the covariance:

$$\mathbb{E}[Y_t Y_{t-h}] = \mathbb{E} \left[\left(\sum_{k=0}^{\infty} \psi_k \varepsilon_{t-k} \right) \left(\sum_{l=0}^{\infty} \psi_l \varepsilon_{t-h-l} \right) \right] = \sum_{k=0}^{\infty} \sum_{l=0}^{\infty} \psi_k \psi_l \mathbb{E}[\varepsilon_{t-k} \varepsilon_{t-h-l}]$$

Since $\{\varepsilon_t\}$ are uncorrelated (and independent), the only nonzero terms are when $t - k = t - h - l$, i.e., $k = h + l$. Therefore,

$$\mathbb{E}[Y_t Y_{t-h}] = \sum_{l=0}^{\infty} \psi_{h+l} \psi_l \mathbb{E}[\varepsilon_{t-(h+l)}^2] = \sigma_\varepsilon^2 \sum_{l=0}^{\infty} \psi_{h+l} \psi_l$$

So the covariance at lag h is

$$\gamma(h) = \text{Cov}(Y_t, Y_{t-h}) = \sigma_\varepsilon^2 \sum_{l=0}^{\infty} \psi_{h+l} \psi_l$$

which does not depend on t : the process is weakly stationary.

To compute the power spectrum, recall that for a stationary process,

$$S(f) = \sum_{k=-\infty}^{+\infty} \gamma(k) e^{-2\pi i f k}.$$

Now, let's expand the modulus square in $S(f) = \sigma_\epsilon^2 |\phi(e^{-2\pi if})|^2$:

$$\phi(e^{-2\pi if}) = \sum_{k=0}^{\infty} \psi_k e^{-2\pi ifk}$$

so

$$\begin{aligned} |\phi(e^{-2\pi if})|^2 &= \phi(e^{-2\pi if}) \overline{\phi(e^{-2\pi if})} = \left(\sum_{k=0}^{\infty} \psi_k e^{-2\pi ifk} \right) \left(\sum_{j=0}^{\infty} \psi_j e^{2\pi ifj} \right) \\ &= \sum_{k=0}^{\infty} \sum_{j=0}^{\infty} \psi_k \psi_j e^{-2\pi if(k-j)} \end{aligned}$$

Therefore, the power spectrum is

$$S(f) = \sigma_\epsilon^2 \sum_{k=0}^{\infty} \sum_{j=0}^{\infty} \psi_k \psi_j e^{-2\pi if(k-j)}$$

which is an explicit expansion of the complex conjugate product.

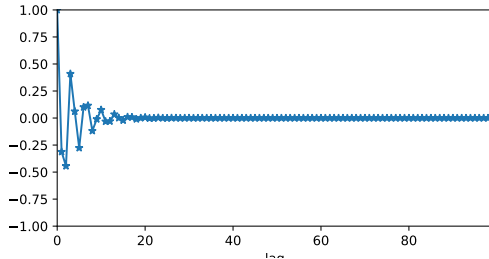
Question 3 AR(2) process

Let $\{Y_t\}_{t \geq 1}$ be an AR(2) process, i.e.

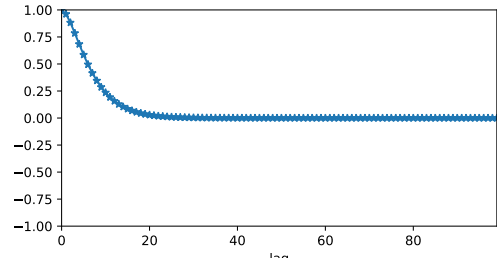
$$Y_t = \phi_1 Y_{t-1} + \phi_2 Y_{t-2} + \varepsilon_t \quad (2)$$

with $\phi_1, \phi_2 \in \mathbb{R}$. The associated characteristic polynomial is $\phi(z) := 1 - \phi_1 z - \phi_2 z^2$. Assume that ϕ has two distinct roots (possibly complex) r_1 and r_2 such that $|r_i| > 1$. Properties on the roots of this polynomial drive the behavior of this process.

- Express the autocovariance coefficients $\gamma(\tau)$ using the roots r_1 and r_2 .
- Figure 1 shows the correlograms of two different AR(2) processes. Can you tell which one has complex roots and which one has real roots?
- Express the power spectrum $S(f)$ (assume the sampling frequency is 1 Hz) using $\phi(\cdot)$.
- Choose ϕ_1 and ϕ_2 such that the characteristic polynomial has two complex conjugate roots of norm $r = 1.05$ and phase $\theta = 2\pi/6$. Simulate the process $\{Y_t\}_t$ (with $n = 2000$) and display the signal and the periodogram (use a smooth estimator) on Figure 2. What do you observe?



Correlogram of the first AR(2)



Correlogram of the second AR(2)

Figure 1: Two AR(2) processes

Answer 3

First, let us establish the stationarity and the form of the solution. Let $\nu_1 = 1/r_1$ and $\nu_2 = 1/r_2$. Since $|r_i| > 1$, we have $|\nu_i| < 1$. Using the state-space representation, let $X_t = [Y_t, Y_{t-1}]^\top$. We can write:

$$X_t = \Phi X_{t-1} + \xi_t, \quad \text{with} \quad \Phi = \begin{pmatrix} \phi_1 & \phi_2 \\ 1 & 0 \end{pmatrix}, \quad \xi_t = \begin{pmatrix} \varepsilon_t \\ 0 \end{pmatrix}. \quad (3)$$

By iterating this equation, we obtain $X_t = \sum_{k=0}^{\infty} \Phi^k \xi_{t-k}$. Since the eigenvalues of Φ are exactly ν_1 and ν_2 (roots of $z^2 - \phi_1 z - \phi_2 = 0$), and $|\nu_i| < 1$, the spectral radius $\rho(\Phi) < 1$. Thus, the series converges in L^2 , defined as a sum of independent Gaussian variables. The variance of the process converges to a finite limit $\gamma(0)$, confirming stationarity.

We now derive $\gamma(0)$ and $\gamma(1)$ algebraically using the Yule-Walker equations. Multiplying the AR(2) equation by $Y_{t-\tau}$ and taking expectations yields the recursion:

$$\gamma(\tau) = \phi_1 \gamma(\tau - 1) + \phi_2 \gamma(\tau - 2), \quad \forall \tau \geq 1. \quad (4)$$

For $\tau = 0$, taking the expectation with Y_t (noting $E[\varepsilon_t Y_t] = \sigma^2$):

$$\gamma(0) = \phi_1 \gamma(1) + \phi_2 \gamma(2) + \sigma^2. \quad (5)$$

For $\tau = 1$:

$$\gamma(1) = \phi_1\gamma(0) + \phi_2\gamma(1) \implies \gamma(1)(1 - \phi_2) = \phi_1\gamma(0) \implies \gamma(1) = \frac{\phi_1}{1 - \phi_2}\gamma(0). \quad (6)$$

Substituting $\gamma(2) = \phi_1\gamma(1) + \phi_2\gamma(0)$ into (5):

$$\gamma(0) = \phi_1\gamma(1) + \phi_2(\phi_1\gamma(1) + \phi_2\gamma(0)) + \sigma^2 = (\phi_1^2 + \phi_2)\gamma(1) + \phi_2^2\gamma(0) + \sigma^2. \quad (7)$$

Substituting $\gamma(1)$ from (6):

$$\gamma(0)(1 - \phi_2^2) - \gamma(0)\frac{\phi_1(\phi_1 + \phi_1\phi_2)}{1 - \phi_2} = \sigma^2. \quad (8)$$

Factorizing and solving for $\gamma(0)$:

$$\gamma(0) \left[\frac{(1 - \phi_2)(1 - \phi_2^2) - \phi_1^2(1 + \phi_2)}{1 - \phi_2} \right] = \sigma^2 \implies \gamma(0) = \frac{(1 - \phi_2)\sigma^2}{(1 + \phi_2)((1 - \phi_2)^2 - \phi_1^2)}. \quad (9)$$

Using Vieta's formulas, $\phi_1 = \nu_1 + \nu_2$ and $\phi_2 = -\nu_1\nu_2$. The denominator term $(1 - \phi_2)^2 - \phi_1^2$ simplifies to $(1 + \nu_1\nu_2)^2 - (\nu_1 + \nu_2)^2 = (1 - \nu_1^2)(1 - \nu_2^2)$. Thus:

$$\gamma(0) = \frac{(1 + \nu_1\nu_2)\sigma^2}{(1 - \nu_1\nu_2)(1 - \nu_1^2)(1 - \nu_2^2)}. \quad (10)$$

The autocovariance satisfies the homogeneous difference equation for $\tau \geq 1$, so its solution is of the form $\gamma(\tau) = c_1\nu_1^\tau + c_2\nu_2^\tau$. We solve for c_1, c_2 using the boundary conditions at $\tau = 0, 1$:

$$\begin{pmatrix} 1 & 1 \\ \nu_1 & \nu_2 \end{pmatrix} \begin{pmatrix} c_1 \\ c_2 \end{pmatrix} = \begin{pmatrix} \gamma(0) \\ \gamma(1) \end{pmatrix}. \quad (11)$$

Inverting the matrix:

$$\begin{pmatrix} c_1 \\ c_2 \end{pmatrix} = \frac{1}{\nu_2 - \nu_1} \begin{pmatrix} \nu_2 & -1 \\ -\nu_1 & 1 \end{pmatrix} \begin{pmatrix} \gamma(0) \\ \gamma(1) \end{pmatrix}. \quad (12)$$

Using $\gamma(1) = \frac{\nu_1 + \nu_2}{1 + \nu_1\nu_2}\gamma(0)$, we calculate c_1 :

$$c_1 = \frac{\nu_2\gamma(0) - \gamma(1)}{\nu_2 - \nu_1} = \frac{\gamma(0)}{\nu_2 - \nu_1} \left(\nu_2 - \frac{\nu_1 + \nu_2}{1 + \nu_1\nu_2} \right) = \frac{\gamma(0)}{\nu_2 - \nu_1} \frac{\nu_2 + \nu_1\nu_2^2 - \nu_1 - \nu_2}{1 + \nu_1\nu_2} = \gamma(0) \frac{\nu_1\nu_2^2 - \nu_1}{(\nu_2 - \nu_1)(1 + \nu_1\nu_2)}. \quad (13)$$

This simplifies to:

$$c_1 = \frac{\sigma^2\nu_1(1 - \nu_2^2)}{(\nu_1 - \nu_2)(1 - \nu_1\nu_2)(1 - \nu_1^2)(1 - \nu_2^2)} = \frac{\sigma^2\nu_1}{(\nu_1 - \nu_2)(1 - \nu_1\nu_2)(1 - \nu_1^2)}. \quad (14)$$

By symmetry for c_2 and expressing in terms of roots $r_i = 1/\nu_i$:

$$\gamma(\tau) = \sigma^2 \sum_{i=1}^2 \frac{r_i^{-|\tau|}}{(1 - r_i^{-2})(1 - r_1^{-1}r_2^{-1})(r_i^{-1} - r_{3-i}^{-1})} r_i^{-1}. \quad (15)$$

The left correlogram exhibits oscillating behavior characteristic of complex conjugate roots in the characteristic polynomial, whereas the right correlogram displays a monotonic exponential

decay typical of real roots, allowing us to identify the left as the complex case and the right as the real case.

The power spectrum density $S(f)$ is the Fourier transform of the autocovariance function:

$$S(f) = \sum_{\tau=-\infty}^{\infty} \gamma(\tau) e^{-i2\pi f \tau}. \quad (16)$$

Using the expression derived previously:

$$\gamma(\tau) = c_1 v_1^{|\tau|} + c_2 v_2^{|\tau|}, \quad \text{where } v_i = 1/r_i. \quad (17)$$

We can split the sum for each component $k \in \{1, 2\}$:

$$S_k(f) = \sum_{\tau=-\infty}^{\infty} v_k^{|\tau|} e^{-i2\pi f \tau} = 1 + \sum_{\tau=1}^{\infty} (v_k e^{-i2\pi f})^{\tau} + \sum_{\tau=1}^{\infty} (v_k e^{i2\pi f})^{\tau}. \quad (18)$$

Since $|v_k| < 1$, these are convergent geometric series:

$$S_k(f) = 1 + \frac{v_k e^{-i2\pi f}}{1 - v_k e^{-i2\pi f}} + \frac{v_k e^{i2\pi f}}{1 - v_k e^{i2\pi f}} = 1 + \frac{v_k e^{-i\omega}}{1 - v_k e^{-i\omega}} + \frac{v_k e^{i\omega}}{1 - v_k e^{i\omega}}, \quad (19)$$

where $\omega = 2\pi f$. Combining terms:

$$S_k(f) = \frac{(1 - v_k e^{-i\omega})(1 - v_k e^{i\omega}) + v_k e^{-i\omega}(1 - v_k e^{i\omega}) + v_k e^{i\omega}(1 - v_k e^{-i\omega})}{|1 - v_k e^{-i\omega}|^2}. \quad (20)$$

The numerator simplifies to:

$$1 - v_k e^{i\omega} - v_k e^{-i\omega} + v_k^2 + v_k e^{-i\omega} - v_k^2 + v_k e^{i\omega} - v_k^2 = 1 - v_k^2. \quad (21)$$

Thus:

$$S_k(f) = \frac{1 - v_k^2}{|1 - v_k e^{-i2\pi f}|^2}. \quad (22)$$

Substituting c_k :

$$S(f) = c_1 \frac{1 - v_1^2}{|1 - v_1 z^{-1}|^2} + c_2 \frac{1 - v_2^2}{|1 - v_2 z^{-1}|^2}, \quad \text{with } z = e^{i2\pi f}. \quad (23)$$

Recall from the derivation of $\gamma(0)$:

$$c_1(1 - v_1^2) = \frac{\sigma^2 v_1}{(v_1 - v_2)(1 - v_1 v_2)}, \quad c_2(1 - v_2^2) = \frac{\sigma^2 v_2}{(v_2 - v_1)(1 - v_1 v_2)}. \quad (24)$$

So:

$$S(f) = \frac{\sigma^2}{1 - v_1 v_2} \left(\frac{v_1}{(v_1 - v_2)|1 - v_1 z^{-1}|^2} + \frac{v_2}{(v_2 - v_1)|1 - v_2 z^{-1}|^2} \right). \quad (25)$$

Finding a common denominator $D = |1 - v_1 z^{-1}|^2 |1 - v_2 z^{-1}|^2 = |(1 - v_1 z^{-1})(1 - v_2 z^{-1})|^2 = |1 - (v_1 + v_2)z^{-1} + v_1 v_2 z^{-2}|^2$. This matches $|\phi(e^{-i2\pi f})|^2$ since $\phi_1 = v_1 + v_2$ and $\phi_2 = -v_1 v_2$.

The numerator term is:

$$N = \frac{\sigma^2}{(v_1 - v_2)(1 - v_1 v_2)} \left[v_1(1 - v_2 z)(1 - v_2 z^{-1}) - v_2(1 - v_1 z)(1 - v_1 z^{-1}) \right]. \quad (26)$$

Expanding the bracket:

$$\nu_1(1 - \nu_2(z + z^{-1}) + \nu_2^2) - \nu_2(1 - \nu_1(z + z^{-1}) + \nu_1^2) \quad (27)$$

$$= \nu_1 - \nu_1\nu_2(z + z^{-1}) + \nu_1\nu_2^2 - \nu_2 + \nu_1\nu_2(z + z^{-1}) - \nu_1^2\nu_2 \quad (28)$$

$$= (\nu_1 - \nu_2) + \nu_1\nu_2(\nu_2 - \nu_1) = (\nu_1 - \nu_2)(1 - \nu_1\nu_2). \quad (29)$$

Thus, the entire numerator N simplifies to just σ^2 .

$$S(f) = \frac{\sigma^2}{|\phi(e^{-i2\pi f})|^2} = \frac{\sigma^2}{|1 - \phi_1 e^{-i2\pi f} - \phi_2 e^{-i4\pi f}|^2}. \quad (30)$$

To analyze the peak of the spectrum, let us factorize the characteristic polynomial as $\phi(z) = (1 - z/r_1)(1 - z/r_2)$. Assuming complex conjugate roots $r_{1,2} = \rho e^{\pm i\theta}$ with $\rho > 1$, the power spectrum is given by:

$$S(f) = \frac{\sigma^2}{|1 - \rho^{-1} e^{i(2\pi f - \theta)}|^2 |1 - \rho^{-1} e^{i(2\pi f + \theta)}|^2}. \quad (31)$$

For complex conjugate roots $r_{1,2} = r e^{\pm i\theta}$ with $r = 1.05$ and $\theta = 2\pi/6$, we have $\nu_{1,2} = r^{-1} e^{\pm i\theta}$ where $r^{-1} \approx 0.952$. The power spectrum is:

$$S(f) = \frac{\sigma^2}{|1 - \phi_1 e^{-i2\pi f} - \phi_2 e^{-i4\pi f}|^2} = \frac{\sigma^2}{|(1 - \nu_1 e^{-i2\pi f})(1 - \nu_2 e^{-i2\pi f})|^2}. \quad (32)$$

For frequency $f = f_0 := \theta/(2\pi) = 1/6$ Hz, we have $e^{-i2\pi f_0} = e^{-i\theta}$. Thus:

$$|1 - \nu_1 e^{-i\theta}|^2 = |1 - r^{-1} e^{i\theta} e^{-i\theta}|^2 = |1 - r^{-1}|^2 = (1 - r^{-1})^2. \quad (33)$$

Similarly, $|1 - \nu_2 e^{-i\theta}|^2 = |1 - r^{-1} e^{-i\theta} e^{-i\theta}|^2 = |1 - r^{-1} e^{-i2\theta}|^2$. Since $r \approx 1$, the denominator $|1 - r^{-1}|^2 = (r - 1)^2/r^2 \approx (r - 1)^2$ is very small. For $r = 1.05$, we have $(r - 1)^2 = 0.05^2 = 0.0025$, yielding:

$$S(f_0) \approx \frac{\sigma^2}{(r - 1)^2 \cdot |1 - r^{-1} e^{-i2\theta}|^2} \approx \frac{\sigma^2}{0.0025 \cdot C} \sim 400\sigma^2/C, \quad (34)$$

where $C = |1 - r^{-1} e^{-i2\theta}|^2 \approx O(1)$. This produces a sharp spectral peak at frequency $f = \theta/(2\pi) = 1/6$ Hz, corresponding to the imaginary part (phase) of the complex roots, with amplitude scaling as $(r - 1)^{-2}$, explaining the observed peak around magnitude ~ 350 in the periodogram.

We can see that the process is stationary non diverging and the fourier spectrum shows a peak at the frequency of the roots.

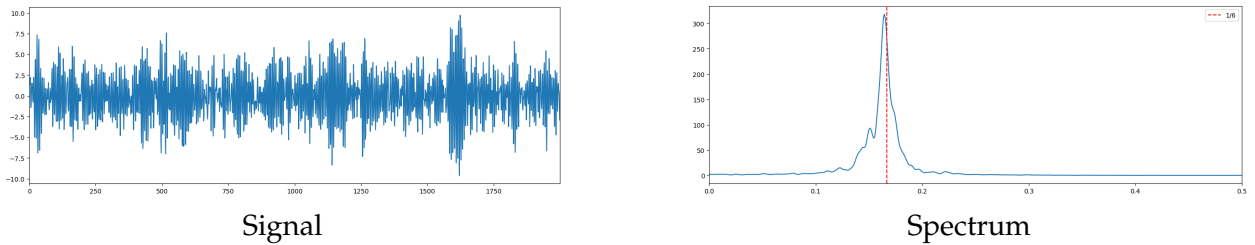


Figure 2: AR(2) process

4 Sparse coding

The modulated discrete cosine transform (MDCT) is a signal transformation often used in sound processing applications (for instance, to encode an MP3 file). A MDCT atom $\phi_{L,k}$ is defined for a length $2L$ and a frequency localisation k ($k = 0, \dots, L - 1$) by

$$\forall u = 0, \dots, 2L - 1, \quad \phi_{L,k}[u] = w_L[u] \sqrt{\frac{2}{L}} \cos\left[\frac{\pi}{L} \left(u + \frac{L+1}{2}\right) \left(k + \frac{1}{2}\right)\right] \quad (35)$$

where w_L is a modulating window given by

$$w_L[u] = \sin\left[\frac{\pi}{2L} \left(u + \frac{1}{2}\right)\right]. \quad (36)$$

Question 4 *Sparse coding with OMP*

For the signal provided in the notebook, learn a sparse representation with MDCT atoms. The dictionary is defined as the concatenation of all shifted MDCT atoms for scales L in $[32, 64, 128, 256, 512, 1024]$.

- For the sparse coding, implement the Orthogonal Matching Pursuit (OMP). (Use convolutions to compute the correlation coefficients.)
- Display the norm of the successive residuals and the reconstructed signal with 10 atoms.

Answer 4

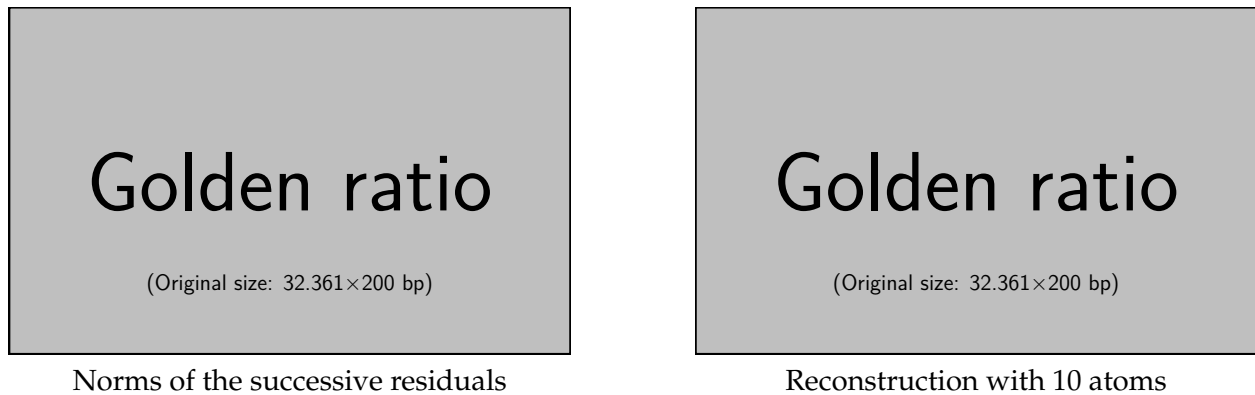


Figure 3: Question 4