

# GMRT transients search and related issues

Jayanti Prasad

National Center for Radio Astronomy  
Pune, India (411007)

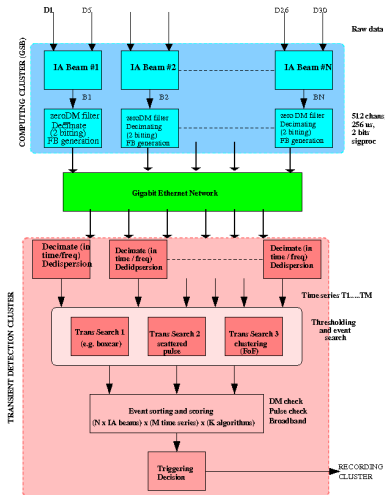
September 3, 2009

# Introduction

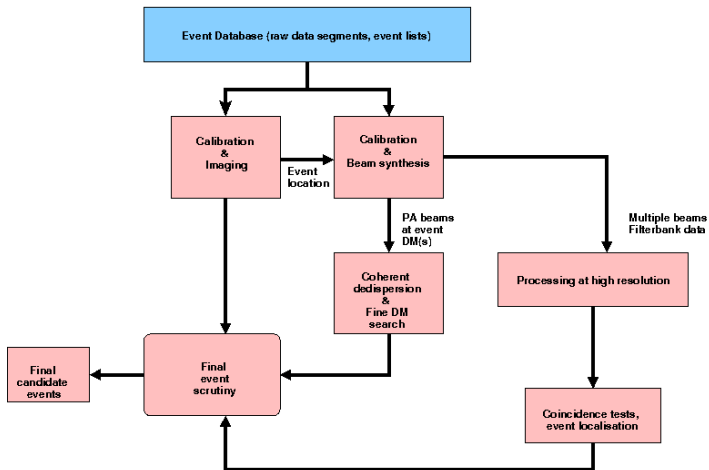
A typical transients search pipeline consists of

- 1 Dedispersion
- 2 Matched filtering
- 3 Thresholding
- 4 Diagnosis

## Transient Detection (Real-time Ops)



## Transient Analysis (offline Ops)



# False alarms

- The probability of a Gaussian noise signal with zero mean and  $\sigma$  variance, crossing a threshold  $T$  (probability of false alarms) is given by

$$P(T) = \int_T^{\infty} \frac{1}{\sqrt{2\pi}\sigma^2} \exp\left(-\frac{x^2}{2\sigma^2}\right) = \frac{1}{2} \text{Erfc}\left(\frac{T}{\sigma\sqrt{2}}\right) \quad (1)$$

- The joint probability of false alarms for  $N$  antennas is given by

$$P^A(T) = \left[ \frac{1}{2} \text{Erfc}\left(\frac{T}{\sigma\sqrt{2}}\right) \right]^N \quad (2)$$

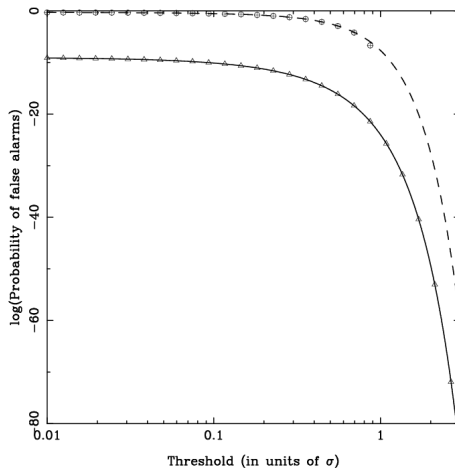
Now if we add the signals from  $N$  antennas incoherently then the joint probability of false alarms

$$P^B(T) = \frac{1}{2} \text{Erfc}\left(\frac{T}{\sigma} \sqrt{\frac{N}{2}}\right) \quad (3)$$

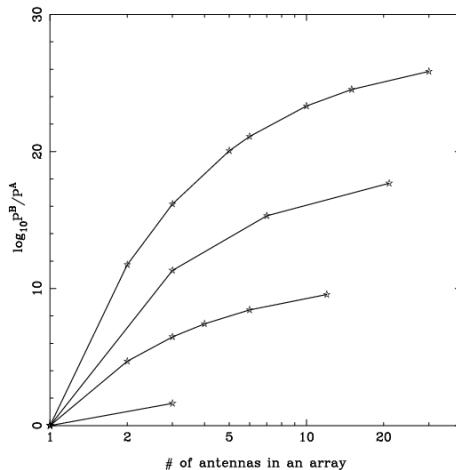
If we break  $N$  antennas into  $p$  incoherent sub-arrays each with  $N/p$  antennas

$$P^c(T) = \left[ \frac{1}{2} \text{Erfc} \left( \frac{T}{\sigma} \sqrt{\frac{N}{2p}} \right) \right]^p \quad (4)$$

What is the best combination for a given number of antennas  $N$  and threshold  $T$  ?



**Figure: 1.** Probabilities of false alarms  $P^A(T)$  (solid line) and  $P^B(T)$  (dotted line). Symbols represent the fraction of events observed above the threshold in a simulated Gaussian noise.



**Figure:** 2. Probabilities of false alarms  $P^B$  and  $P^A$  as a function of the number of antennas in a beam for  $T = 3\sigma$ , for a given number of total antennas. The curves from the top are for the total number of antennas  $N = 30, 21, 12$  and  $3$  respectively.



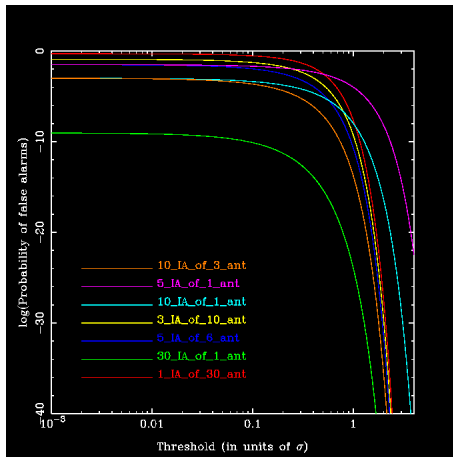


Figure: 3. Probability of false alarms  $P^c(T)$  for  $p$  incoherent sub-arrays of  $N$  antennas. Note that the number of antennas in an IA array is important for high value of threshold.

# Single pulse simulations

- How the difference between the actual DM and trial DM changes the
  - ① location
  - ② width and
  - ③ Amplitude (SNR)of a pulse for a given set of parameters  $W, \nu, n_{chan}, t_{samp}$  etc.
- How the amplitude of a given pulse changes with the size of smoothing window.
- How zero-dm filtering change the shape and size of a pulse.

- Dispersion delay:

$$\tau_f(\text{DM}) \approx K \left[ \left( \frac{\text{MHz}}{f} \right)^2 - \left( \frac{\text{MHz}}{f_h} \right)^2 \right] \left( \frac{\text{DM}}{\text{pc cm}^{-3}} \right) \text{sec} \quad (5)$$

where  $K = 4.154553 \times 10^3$

- Pulse location:

$$\bar{\tau}(\text{DM}) = \frac{\tau_{f_h} + \tau_{f_l}}{2} = \frac{\tau_{f_l}}{2} \quad (6)$$

If the true dispersion measure of a pulse is DM and we dedisperse it with a trial dispersion measure DM+dDM then shift in the location of the pulse is given by

$$\Delta \bar{\tau} = \frac{K}{2} \left( \frac{\text{MHz}}{f_c} \right)^3 \left( \frac{\Delta f}{\text{MHz}} \right) \left( \frac{\text{dDM}}{\text{pc cm}^{-3}} \right) \text{sec} \quad (7)$$

If in place of considering the highest frequency channel as the reference i.e., there is no delay for this channel, we consider any other channel  $f_{ref}$  as reference

$$\tau_f(\text{DM}) \approx K \left[ \left( \frac{\text{MHz}}{f} \right)^2 - \left( \frac{\text{MHz}}{f_{ref}} \right)^2 \right] \left( \frac{\text{DM}}{\text{pc cm}^{-3}} \right) \text{sec} \quad (8)$$

if  $f_{ref} = f_c$ ,  $f_l = f_c - \Delta f/2$ ,  $f_h = f_c + \Delta f/2$  we get

$$\bar{\tau}(\text{DM}) = \frac{\tau_{f_h} + \tau_{f_l}}{2} = \frac{K}{2} \frac{3\Delta f^2}{2(f_c^2 - \Delta f^2/4)^2} \text{MHz}^2 \left( \frac{\text{DM}}{\text{pc cm}^{-3}} \right) \quad (9)$$

then the shift is of the order of  $\Delta f^2$ .

- **Pulse width:** If the intrinsic pulse width is  $W$  then the observed pulse width can be defined as

$$\begin{aligned}
 w &= W + (\tau_{f_l} - \tau_{f_h}) \\
 &= K \left( \frac{\text{MHz}}{f_c} \right)^3 \left( \frac{\Delta f}{\text{MHz}} \right) \left( \frac{\text{DM}}{\text{pc cm}^{-3}} \right) \text{sec}
 \end{aligned} \tag{10}$$

If the true dispersion measure of a pulse is  $DM$  and we dedisperse it with a trial dispersion measure  $DM+dDM$  then the change in the width of the pulse is given by

$$\Delta w = K \left( \frac{\text{MHz}}{f_c} \right)^3 \left( \frac{\Delta f}{\text{MHz}} \right) \left( \frac{dDM}{\text{pc cm}^{-3}} \right) \text{sec} \tag{11}$$

- **Pulse amplitude:** For a rectangular bandpass function and for a Gaussian shape pulse with width  $W$  c the ratio of measured peak flux  $S(\delta DM)$  to true flux  $S$  to for a  $DM$  error  $\delta DM$  is

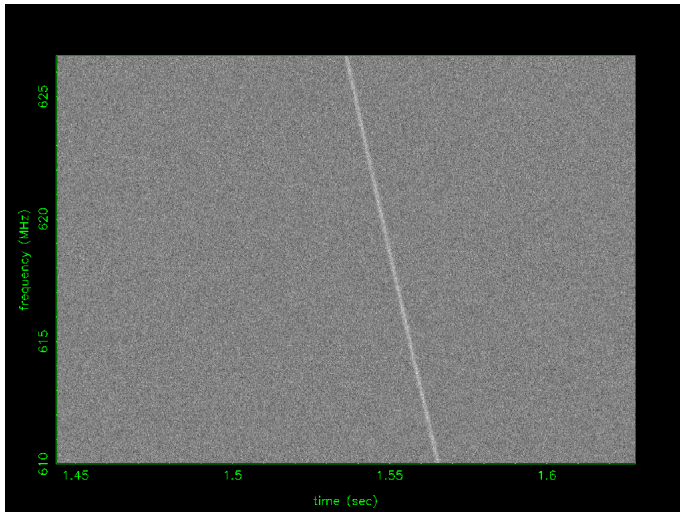
$$\frac{S(\delta DM)}{S} = \frac{\sqrt{\pi}}{2\xi} \text{erfc}(\xi) \quad (12)$$

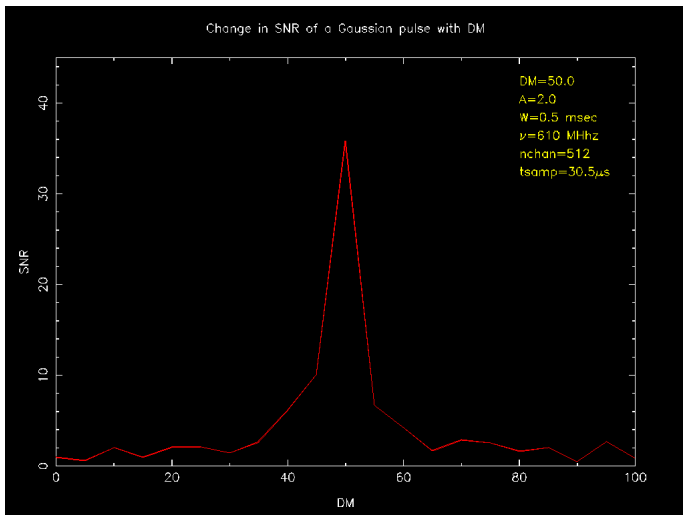
where

$$\xi = 6.91 \times 10^{-3} \delta DM \frac{\delta \nu}{\text{MHz}} \frac{\text{msec}}{W} \left( \frac{\text{Ghz}}{\nu} \right)^2 \quad (13)$$

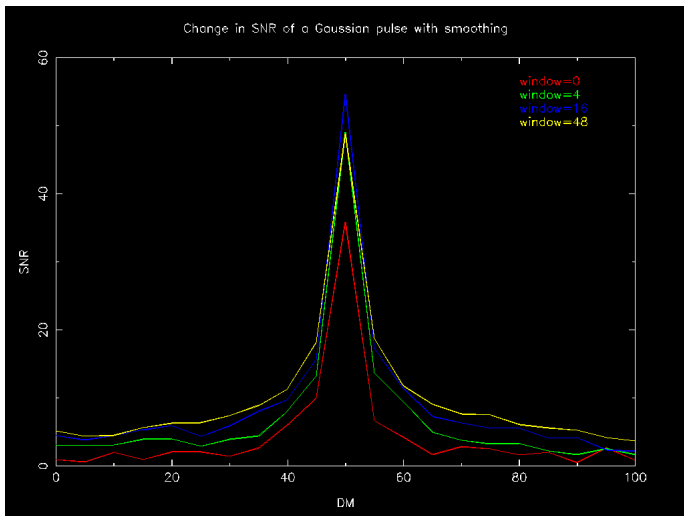
( Cordes & McLaughlin 2003)

# Simulations

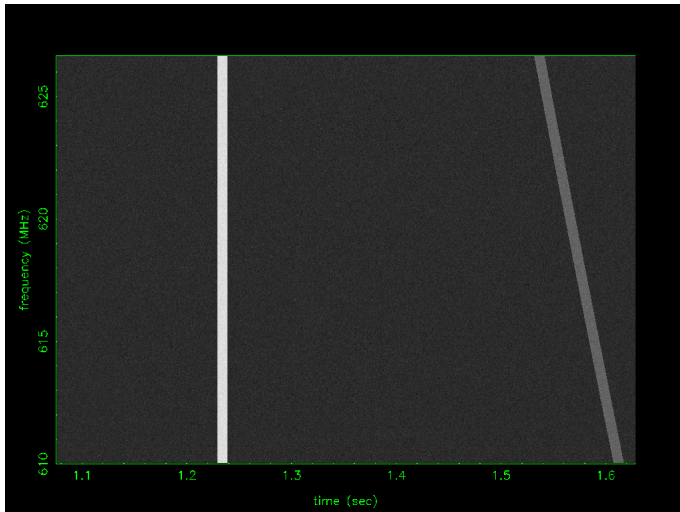




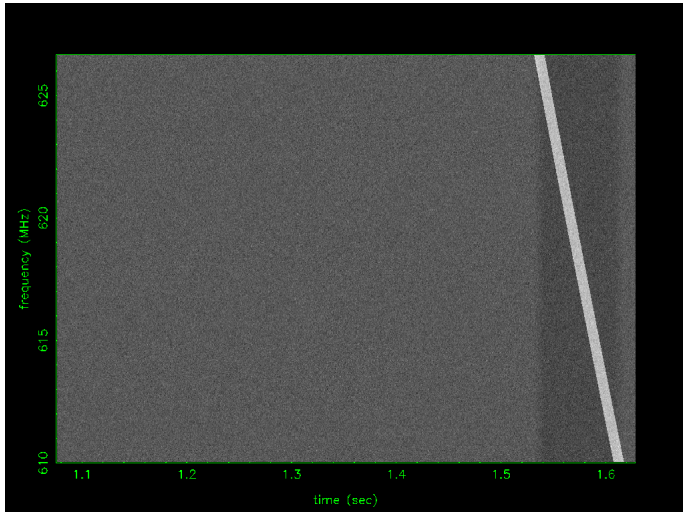




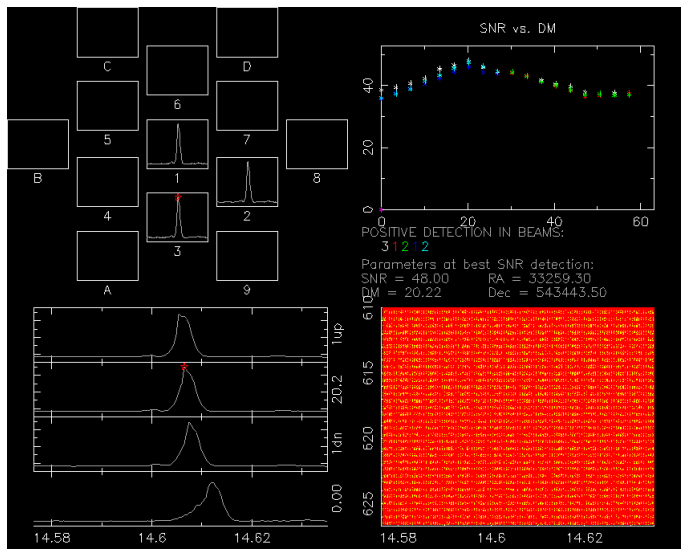
# Broad Band RFI



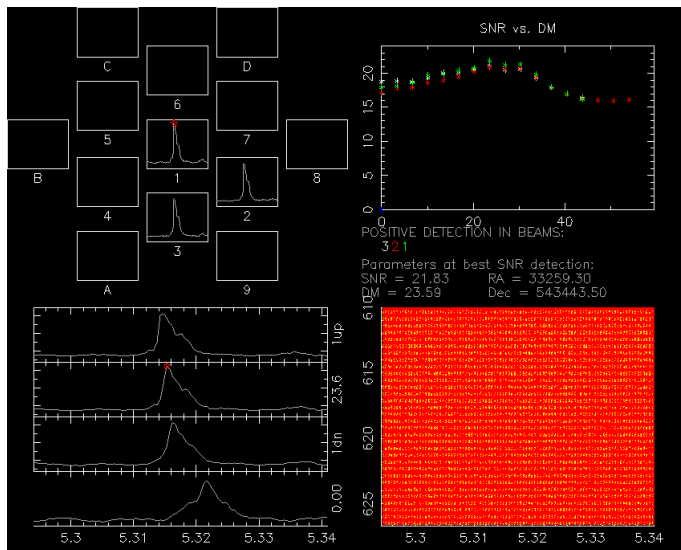
# Broad Band RFI removed with zeron DM filtering



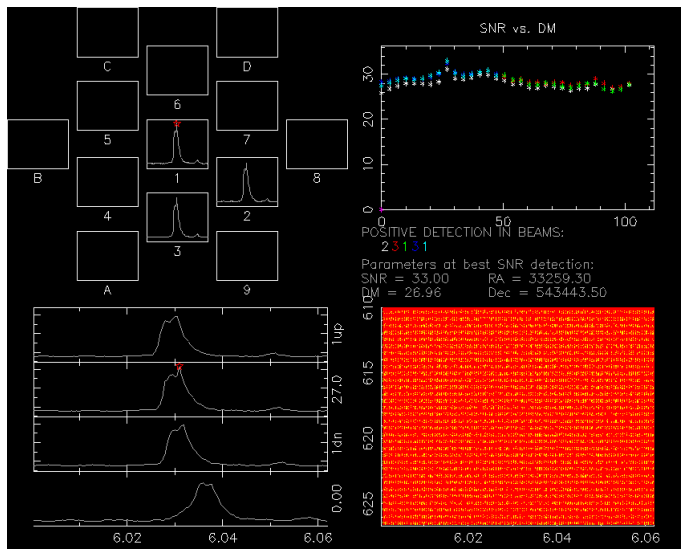
# Giant pulse search:



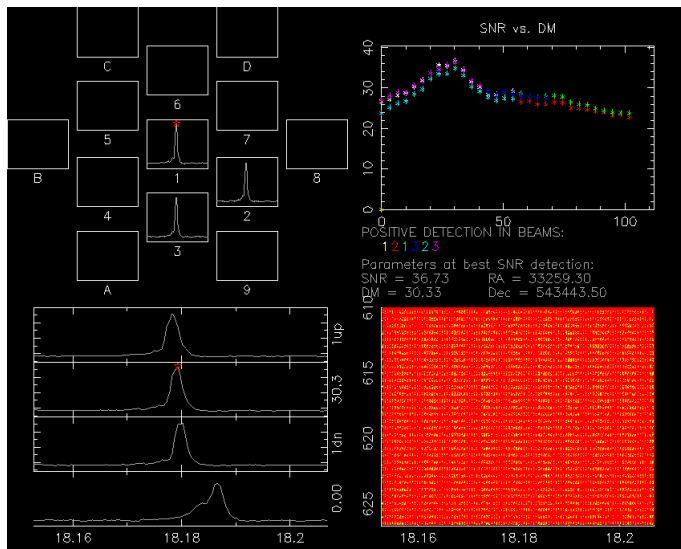
# Giant pulse search..



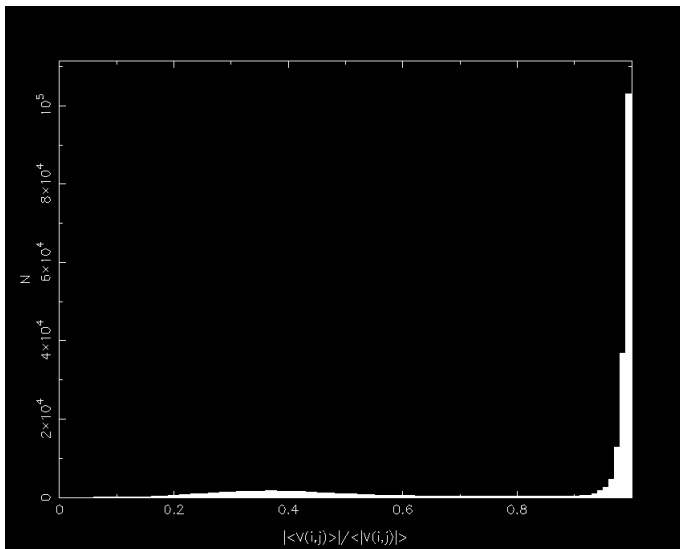
# Giant pulse search..



# Giant pulse search..

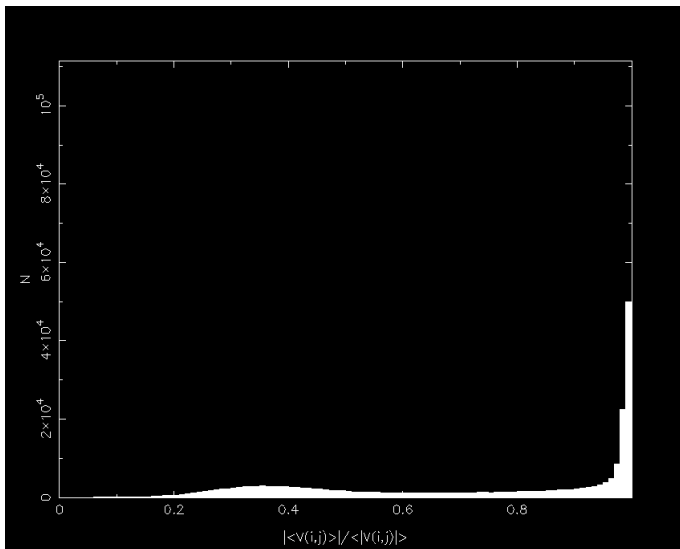


# Time averaging visibilities (Flux calibrator)

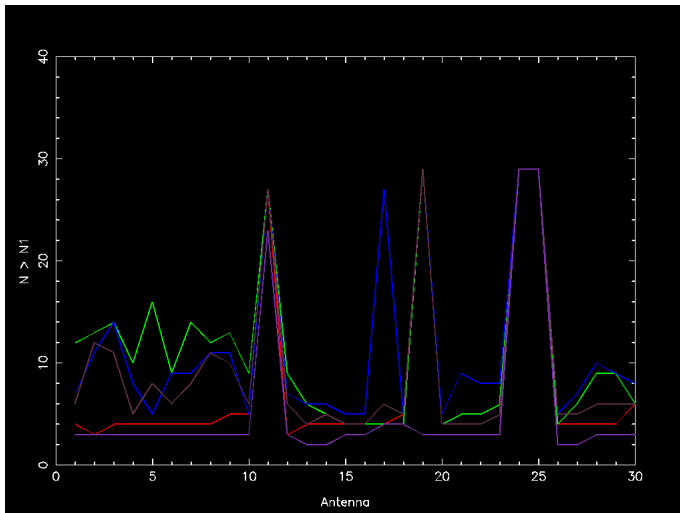




# Time averaging visibilities (Phase calibrator)



# Bad Antennas



*Thank You !*