

# Advanced Digital Signal Processing

## 高等數位訊號處理

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課程網頁：<http://djj.ee.ntu.edu.tw/ADSP.htm>

歡迎大家來修課，也歡迎有問題時隨時聯絡！

- 評分方式：

### Basic: 15 scores

原則上每位同學都可以拿到 12 分以上，  
另外，上課回答問題，每回答一次加1分

### Homework: 60 scores (5 times, 每 3 週一次)

請自己寫，和同學內容極高度相同，將扣 70% 的分數  
就算寫錯但好好寫也會給 40~95% 的分數，  
遲交分數打 8 折，**不交不給分**。不知道如何寫，可用 E-mail 和我  
聯絡，或於上課時發問  
**禁止 Ctrl-C Ctrl-V 的情形。**

### Term paper 25 scores

## Term paper 25 scores

方式有四種

### (1) 書面報告

10頁以上(不含封面)，中英文皆可，11或12的字體，題目可選擇和課程有關的**任何一個**主題。

格式和一般寫期刊論文或碩博士論文相同，包括 abstract, conclusion, 及 references，並且要分 sections，必要時有 subsections。References 的寫法, 可參照一般 IEEE 的論文的寫法

鼓勵**多做實驗及模擬**，**有創新更好**。

**嚴禁 Ctrl-C Ctrl-V 的情形**，否則扣 70% 的分數

### (2) Tutorial (對既有領域做淺顯易懂的整理)

限十八個名額，和書面報告格式相同，但 頁數限制為18頁以上(若為加強前人的 tutorial，則頁數為  $(2/3)N + 13$  以上， $N$  為前人 tutorial 之頁數)，題目由老師指定，以**清楚且有系統**的介紹一個主題的基本概念和應用為要求，為上課內容的進一步探討和補充，[交 Word 檔](#)。

選擇這個項目的同學，學期成績加 3分

### (3) 口頭報告

限四個人，每個人 30~45分鐘，題目可選擇和課程有關的任何一個主題。口頭報告將於 5月4日（第 9週）進行。有意願的同學，請及早告知，並且在口頭報告之前提供 PowerPoint 檔。

口頭報告時，鼓勵大家提問（包括口頭報告的同學，也可針對其他同學的報告內容提問）。曾經提問的同學，學期成績皆加 0.5 分。  
選擇這個項目的同學，學期成績加 2分

### (4) 編輯 Wikipedia

中文或英文網頁皆可，至少 2 個條目，但不可同一個條目翻成中文和英文。總計 80行以上。限和課程相關者，自由發揮，越有條理、有系統的越好

選擇編輯 Wikipedia 的同學，請於 6月22日前，向我登記並告知我要編輯的條目(2 個以上)，若有和其他同學選擇相同條目的情形，則較晚向我登記的同學將更換要編輯的條目

書面報告和編輯 Wikipedia，期限是 7月6日

## Tutorial 可供選擇的題目(共 17 個，可以略做修改)

- (1) Advanced Signal Prediction Techniques
- (2) Damaged Signal Recovery
- (3) Finger Tracking and Its Applications
- (4) Blind Audio Source Separation
- (5) Dynamic Programming for Sequence Matching
- (6) Microscopy Image Processing
- (7) Satellite Image Processing
- (8) Hidden Markov Model
- (9) Golden-section Search and Fibonacci Search
- (10) Occluded Face Recognition
- (11) Signal Processing Techniques for Healthcare

## Tutorial 可供選擇的題目(可以略做修改)

- (12) Signal Processing for Economical Data Analysis
- (13) Blockchain and Artificial Intelligence Techniques
- (14) Virtual Reality
- (15) Augmented Reality
- (16) Human Perception for Vocal Signals
- (17) Human Perception for Images

上課時間：16 週

3/2,

5/4, Oral, 交 HW2

3/9,

5/11, 出 HW3

3/16,

5/18,

3/23, 出 HW1

5/25, 交 HW3

3/30,

6/1, 出 HW4

4/13, 交 HW1

6/8,

4/20, 出 HW2

6/15, 交 HW4

4/27,

6/22, 出 HW5

7/6, 交 HW5 及 term paper

4/6 放假

原則上:  $3n+1$  週出作業,  $3n+3$  週繳交

## Matlab

Download: 請洽台大電信所

<http://comm.ntu.edu.tw/matlab/request.php>

### 參考書目

洪維恩，Matlab 7 程式設計，旗標，台北市，2010.. (合適的入門書)

張智星，Matlab 程式設計入門篇，第三版，碁峰，2011.

蒙以正，數位信號處理：應用 Matlab，旗標，台北市，2007.

繆紹綱譯，數位影像處理：運用-Matlab，東華，2005.

預計看書學習所花時間：3~5 天



研究所和大學以前追求知識的方法有什麼不同？

研究所：觀念的學習

大學：

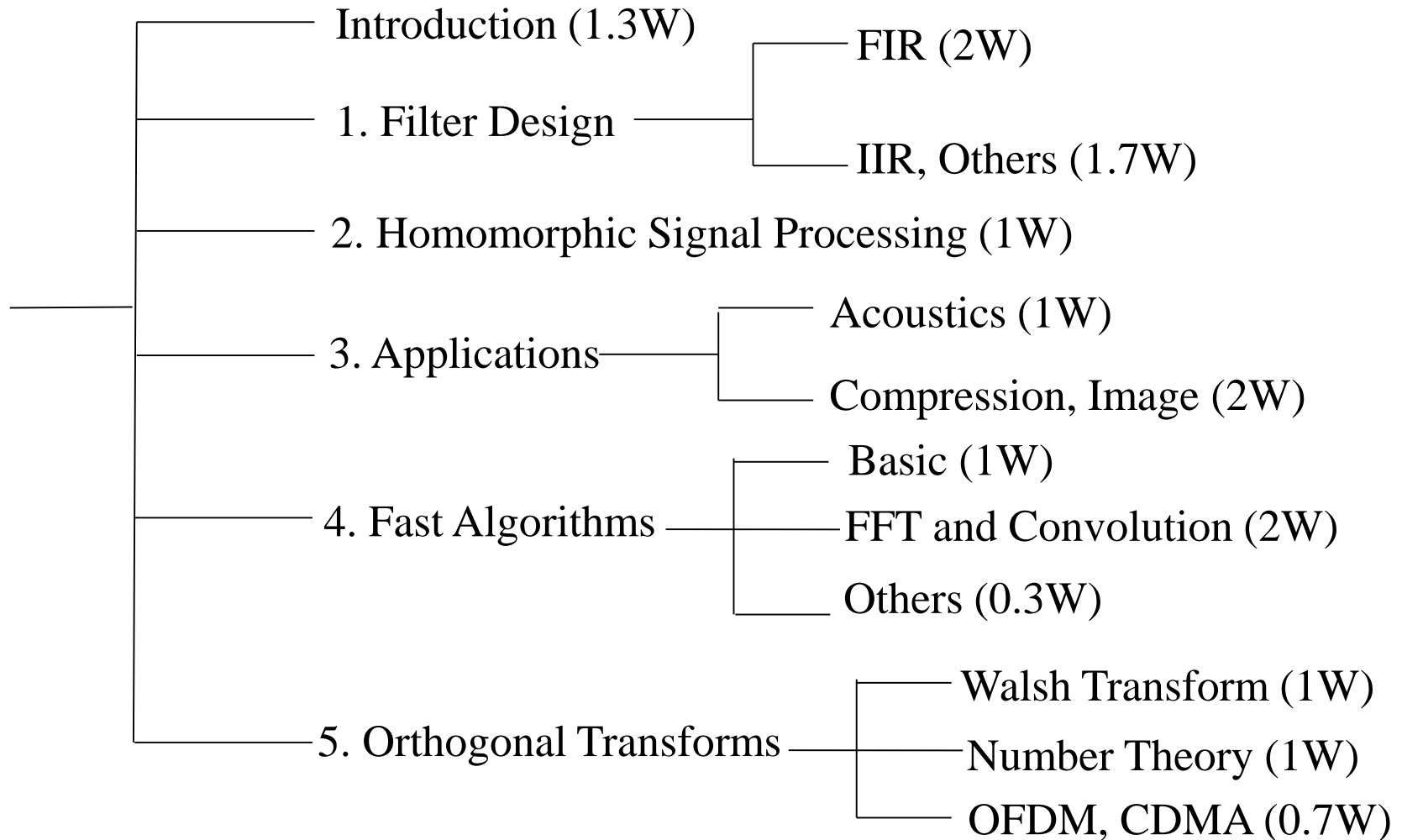
**Question:**

**Why should we use the Fourier transform?**

Is the Fourier transform the best choice in any condition?

# I. Introduction

## Outline

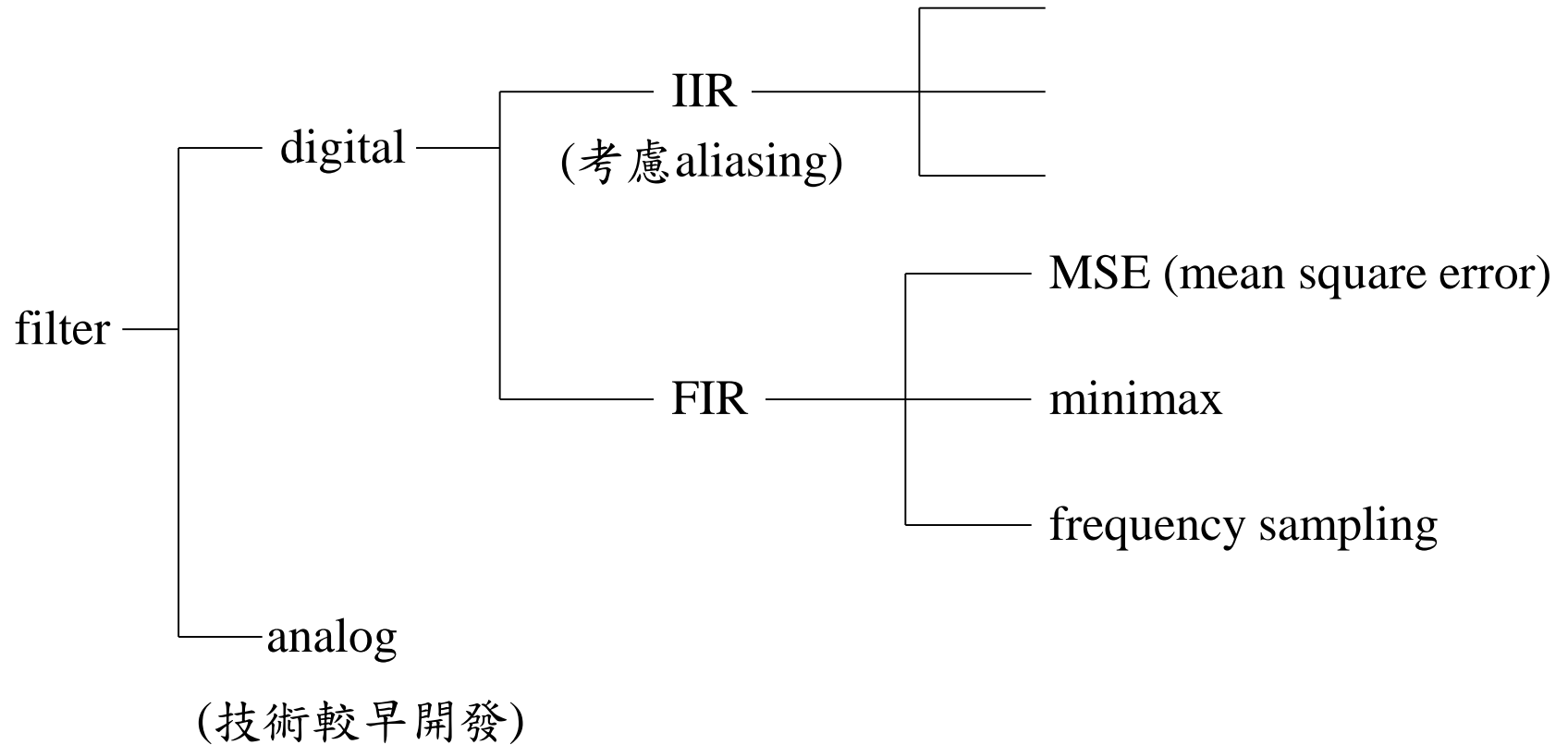


目標：

- (1) 對 Digital Signal Processing 作更有系統且深入的了解
- (2) 學習 Digital Signal Processing 幾個重要子領域的基礎知識

## Part 1: Filter

- Filter 的分類



IIR filter 的優點：(1) easy to design

(2) (sometimes) easy to implement

缺點：

FIR filter 的優點：

缺點：An FIR filter is impossible to have the ideal frequency response of



## Part 2: Homomorphic Signal Processing

- 概念：把 convolution 變成 addition

## Part 3: Applications of DSP

filter design, data compression (image, video, text), acoustics (speech, music), image analysis (structural similarity, sharpness), 3D accelerometer

- **Part 4: Fast Algorithms**
- Basic Implementation Techniques

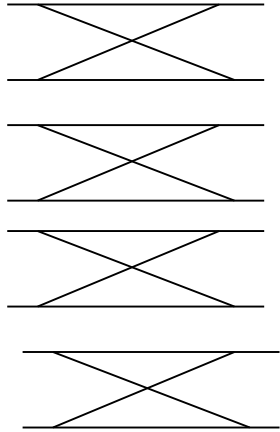
Example: one complex number multiplication  
= ? Real number multiplication.

Trade-off: “Multiplication” takes longer than “addition”



- FFT and Convolution

Due to the Cooley-Tukey algorithm (butterflies),  
the complexity of the FFT is:



The complexity of the convolution is: 3個 DFTs,  $O(N \log_2 N)$

## • Part 5: Orthogonal Transforms

DFT 的兩個主要用途:

Question: DFT 的缺點是什麼？ 
$$DFT(x[n]) = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi mn}{N}}$$

- Walsh Transform  
(CDMA)
- Number Theoretic Transform
- Orthogonal Frequency-Division Multiplexing (OFDM)
- Code Division Multiple Access (CDMA)

# Review 1: Four Types of the Fourier Transform

## (1) Fourier Transform

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt, \quad x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi f t} df$$

Alternative definitions

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt, \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

## (2) Fourier series (suitable for period function)

$$X[m] = \int_0^T x(t) e^{-j\frac{2\pi m}{T} t} dt, \quad x(t) = T^{-1} \sum_{m=-\infty}^{\infty} X[m] e^{j\frac{2\pi m}{T} t}$$

$T$ : 週期  $x(t) = x(t+T)$  possible periods:

possible frequencies:

頻率和  $m$  之間的關係： $f = \frac{m}{T}$   $\frac{1}{T}$  整數倍

### (3) Discrete-time Fourier transform (DSP 常用)

$$X(f) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi f n \Delta_t}, \quad x[n] = \Delta_t \int_0^{1/\Delta_t} X(f) e^{j2\pi f n \Delta_t} df$$

$\Delta_t$ : sampling interval

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n \Delta_t}, \quad x[n] = \frac{\Delta_t}{2\pi} \int_0^{2\pi/\Delta_t} X(\omega) e^{j\omega n \Delta_t} d\omega$$

### (4) Discrete Fourier transform (DFT) (DSP 常用)

$$X[m] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi mn}{N}}, \quad x[n] = \frac{1}{N} \sum_{m=0}^{N-1} X[m] e^{j\frac{2\pi mn}{N}}$$

頻率和  $m$  之間的關係： $f = \frac{m}{N\Delta_t} = \frac{m}{N} f_s$

where  $f_s = 1/\Delta_t$  (sampling frequency)

- 四種 Fourier transforms 的比較

	time domain	frequency domain
(1) Fourier transform	continuous, aperiodic	continuous, aperiodic
(2) Fourier series	continuous, periodic  (or continuous, only the value in a finite duration is known)	discrete, aperiodic
(3) discrete-time Fourier transform	discrete , aperiodic	continuous, periodic
(4) discrete Fourier transform	discrete, periodic  (or discrete, only the value in a finite duration is known)	discrete, periodic

## Review 2: Normalized Frequency

(1) Definition of **normalized frequency**  $F$ :

$$F = \frac{f}{f_s} = f \Delta_t = \frac{\omega \Delta_t}{2\pi} \quad \text{where } f_s = 1/\Delta_t \text{ (sampling frequency)}$$

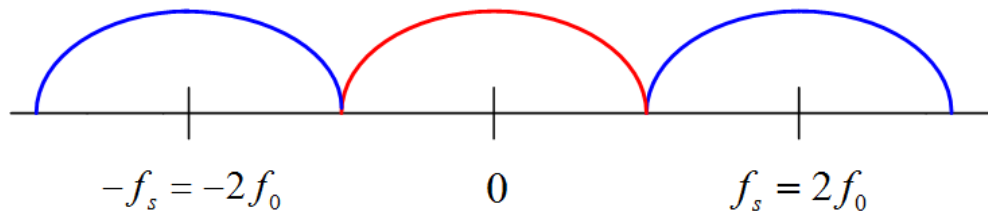
$\Delta_t$  : sampling interval

(2) folding frequency  $f_0$

$$f_0 = \frac{f_s}{2} \quad \text{若以 normalized frequency 來表示 ,}$$

folding frequency = 1/2

$H(f)$ :



For the discrete time Fourier transform

$$(1) G(f) = G(f + f_s) \longrightarrow \text{i.e., } G(F) = G(F + 1).$$

$$(2) \text{ If } g[n] \text{ is real} \longrightarrow G(F) = G^*(-F) \text{ (* means conjugation)}$$

只需知道  $G(F)$  for  $0 \leq F \leq \frac{1}{2}$  (即  $0 < f < f_0$ )

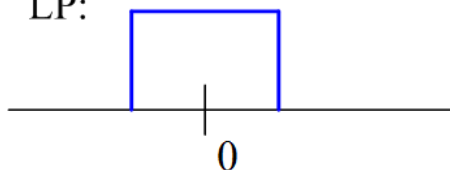
就可以知道全部的  $G(F)$

$$(3) \text{ If } g[n] = g[-n] \text{ (even)} \longrightarrow G(F) = G(-F),$$

$$g[n] = -g[-n] \text{ (odd)} \longrightarrow G(F) = -G(-F)$$

Analog  
filter:  $H(f)$

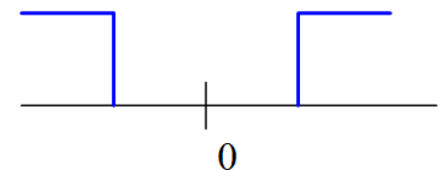
LP:



BP:

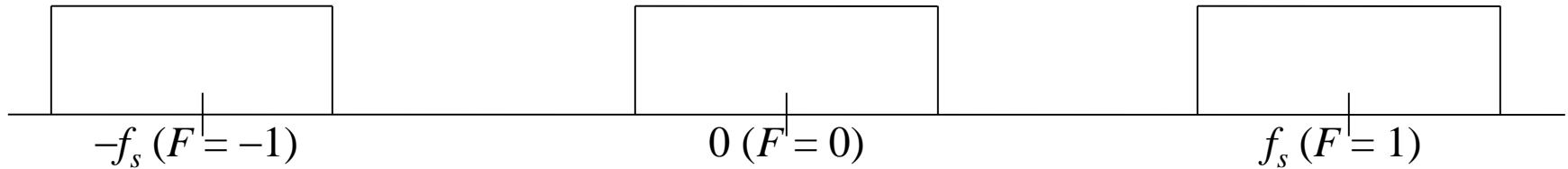


HP:

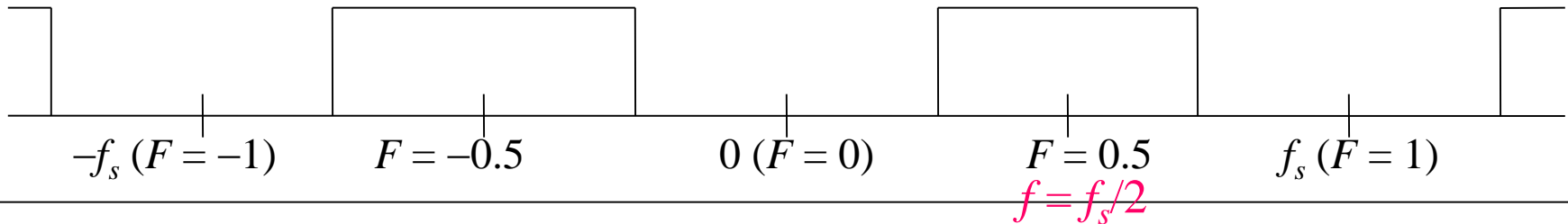


- Discrete time Fourier transform of the lowpass, highpass, and band pass filters

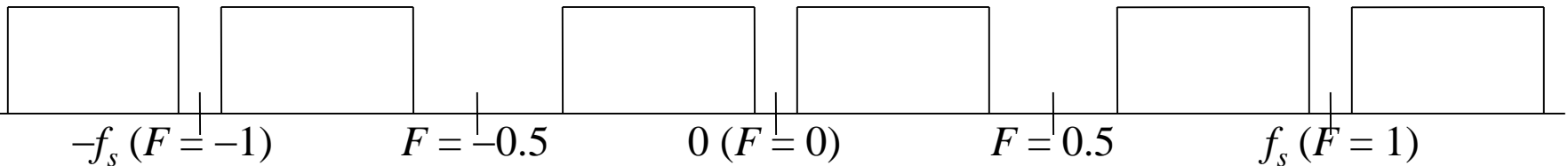
low pass filter ( **pass band** 在  $f_s$  的整數倍附近 )



high pass filter



band pass filter





# Review 3: Z Transform and Laplace Transform

- **Z-Transform**

suitable for **discrete** signals

$$G(z) = \sum_{n=-\infty}^{\infty} g[n] z^{-n}$$

Compared with the discrete time Fourier transform:

$$G(f) = \sum_{n=-\infty}^{\infty} g[n] e^{-j2\pi f n \Delta_t} \quad z = e^{j2\pi f \Delta_t}$$

## • Laplace Transform

suitable for **continuous** signals

One-sided form  $G(s) = \int_0^{\infty} g(t)e^{-st} dt$

Two-sided form  $G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$

Compared with the Fourier transform:

$$G(f) = \int_{-\infty}^{\infty} g(t)e^{-j2\pi f t} dt \qquad s = j2\pi f$$

## Review 4: IIR Filter Design

Two types of digital filter:

- (1) IIR filter (infinite impulse response filter)
- (2) FIR filter (finite impulse response filter)

There are 3 popular methods to design the IIR filter.

## Method 1: Impulse Invariance

白話一點，就是直接做 sampling

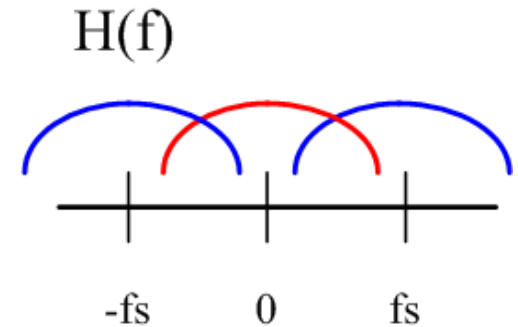
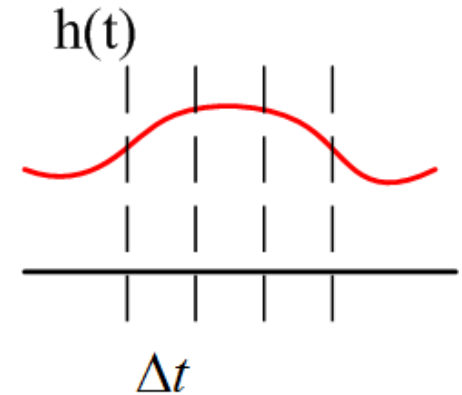
analog filter  $h_a(t)$

digital filter  $h[n]$

$$h[n] = h_a(n\Delta_t)$$

Advantage : Simple

Disadvantage : (1) infinite  
(2)



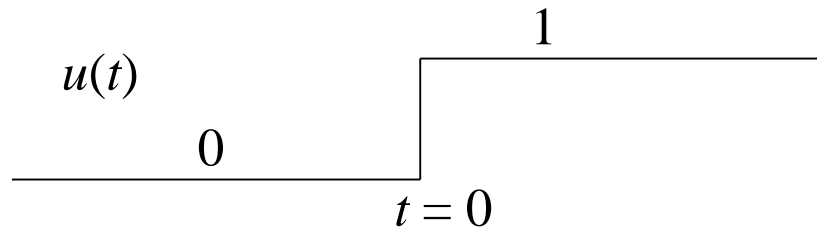
## Method 2: Step Invariance

對 step function 的 response 作 sampling

analog filter  $h_a(t)$

digital filter  $h[n]$

step function (continuous form)



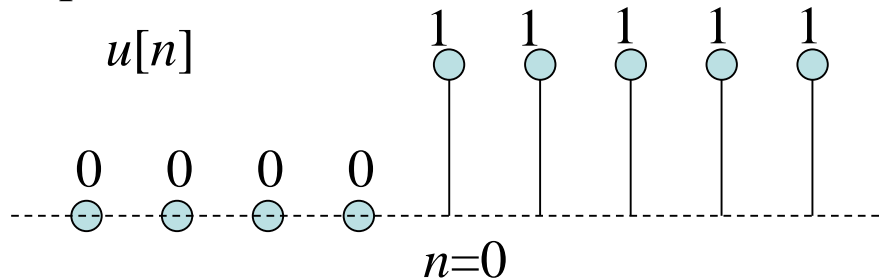
Laplace transform of  $u(t)$ :

$$\frac{1}{s}$$

Fourier transform of  $u(t)$ :

$$\frac{1}{j2\pi f}$$

step function (discrete form)



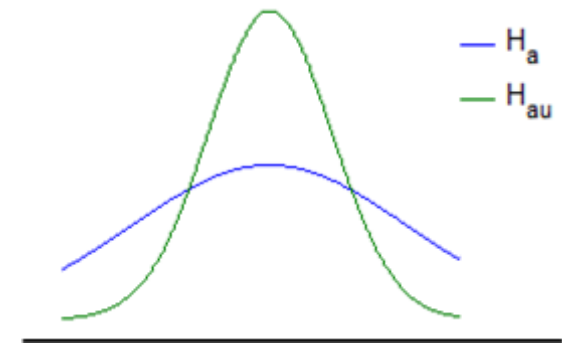
Z transform of  $u[n]$ :

$$\frac{1}{1 - z^{-1}}$$

Step 1 Calculate the convolution of  $h_a(t)$  and  $u(t)$

$$h_{a,u}(t) = h_a(t) * u(t) = \int_{-\infty}^{\infty} h_a(\tau) u(t-\tau) d\tau = \int_{-\infty}^t h_a(\tau) d\tau$$

$$H_{a,u}(f) = \frac{H_a(f)}{j2\pi f} \quad (\text{其實就是對 } h_a(t) \text{ 做積分})$$



Step 2 Perform sampling for  $h_{a,u}(t)$

$$h_u[n] = h_{a,u}(n\Delta_t)$$

Step 3 Calculate  $h[n]$  from  $h[n] = h_u[n] - h_u[n-1]$

Note: Since  $h_u[n] = h[n] * u[n]$

$$H_u(z) = \frac{1}{1-z^{-1}} H(z)$$

$$H(z) = (1-z^{-1}) H_u(z)$$

so  $h[n] = h_u[n] - h_u[n-1]$

Advantage of the step invariance method:

\* 主要 Advantage:

Disadvantage of the step invariance method:

較為間接，設計上稍微複雜

### Method 3: Bilinear Transform

Suppose that we have known an analog filter  $h_a(t)$  whose frequency response is  $H_a(f)$ .

To design the digital filter  $h[n]$  with the frequency response  $H(f)$ ,

$$H(f_{new}) = H_a(f_{old}) \quad f_{old} \in (-\infty, \infty)$$

$$f_{new} \in (-f_s/2, f_s/2)$$

$$f_s = 1/\Delta_t \text{ (sampling frequency)}$$

- The relation between  $f_{new}$  and  $f_{old}$  is determined by the mapping function

$$s = c \frac{1 - z^{-1}}{1 + z^{-1}}$$

$s$ : index of the Laplace transform

$z$ : index of the Z transform

$c$ : some constant



$$s = c \frac{1 - z^{-1}}{1 + z^{-1}}$$

$$s = j2\pi f_{old}$$

$$z = e^{j2\pi f_{new} \Delta_t}$$

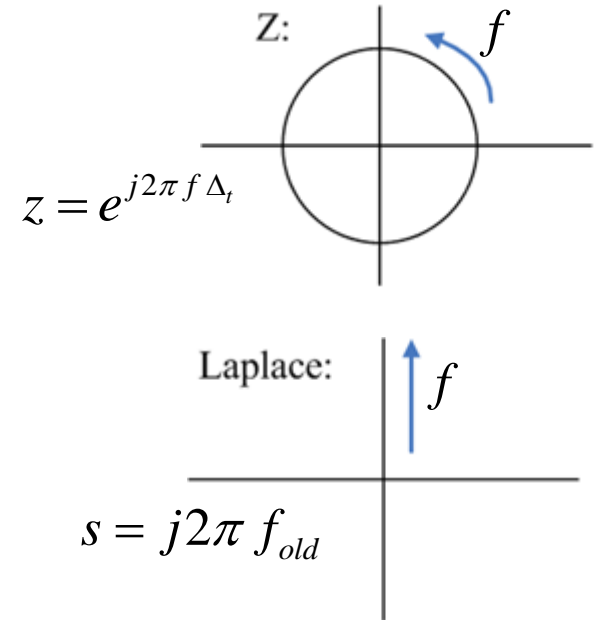
代入

参考 page 25 、 page 26

$$\begin{aligned} j2\pi f_{old} &= c \frac{1 - e^{-j2\pi f_{new} \Delta_t}}{1 + e^{-j2\pi f_{new} \Delta_t}} = c \frac{e^{j\pi f_{new} \Delta_t} - e^{-j\pi f_{new} \Delta_t}}{e^{j\pi f_{new} \Delta_t} + e^{-j\pi f_{new} \Delta_t}} \\ &= c \frac{j \sin(\pi f_{new} \Delta_t)}{\cos(\pi f_{new} \Delta_t)} \end{aligned}$$

$$2\pi f_{old} = c \tan(\pi f_{new} \Delta_t)$$

$$f_{new} = \frac{1}{\pi \Delta_t} \operatorname{atan}\left(\frac{2\pi}{c} f_{old}\right) = \frac{f_s}{\pi} \operatorname{atan}\left(\frac{2\pi}{c} f_{old}\right)$$



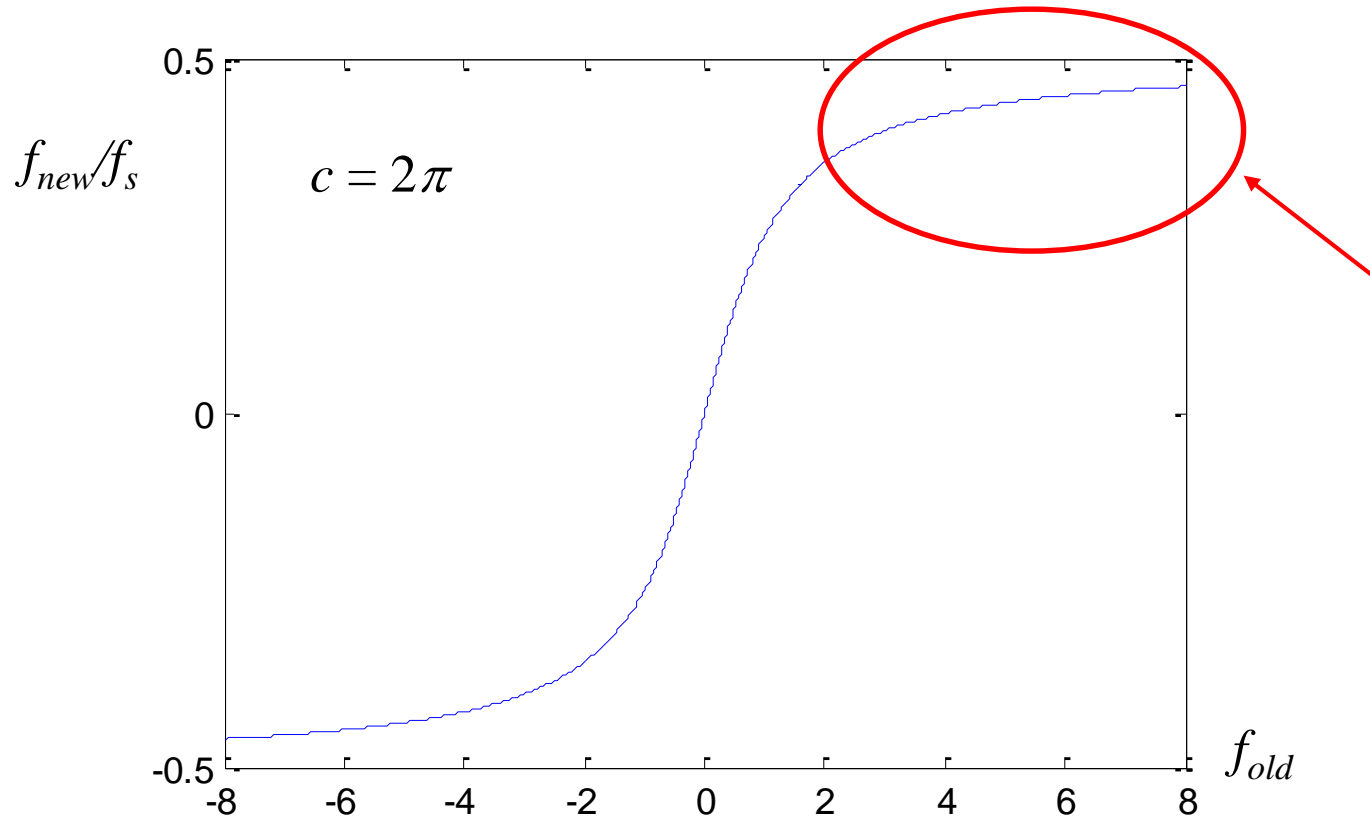
- Suppose that the Laplace transform of the analog filter  $h_a(t)$  is  $H_{a,L}(s)$

The Z transform of the digital filter  $h[n]$  is  $H_z(z)$

$$H_z(z) = H_{a,L}\left(c \frac{1 - z^{-1}}{1 + z^{-1}}\right)$$

$$f_{new} = \frac{f_s}{\pi} \operatorname{atan} \left( \frac{2\pi}{c} f_{old} \right)$$

$f_{old}$	$-\infty$	0	$\infty$	1
$f_{new}$				

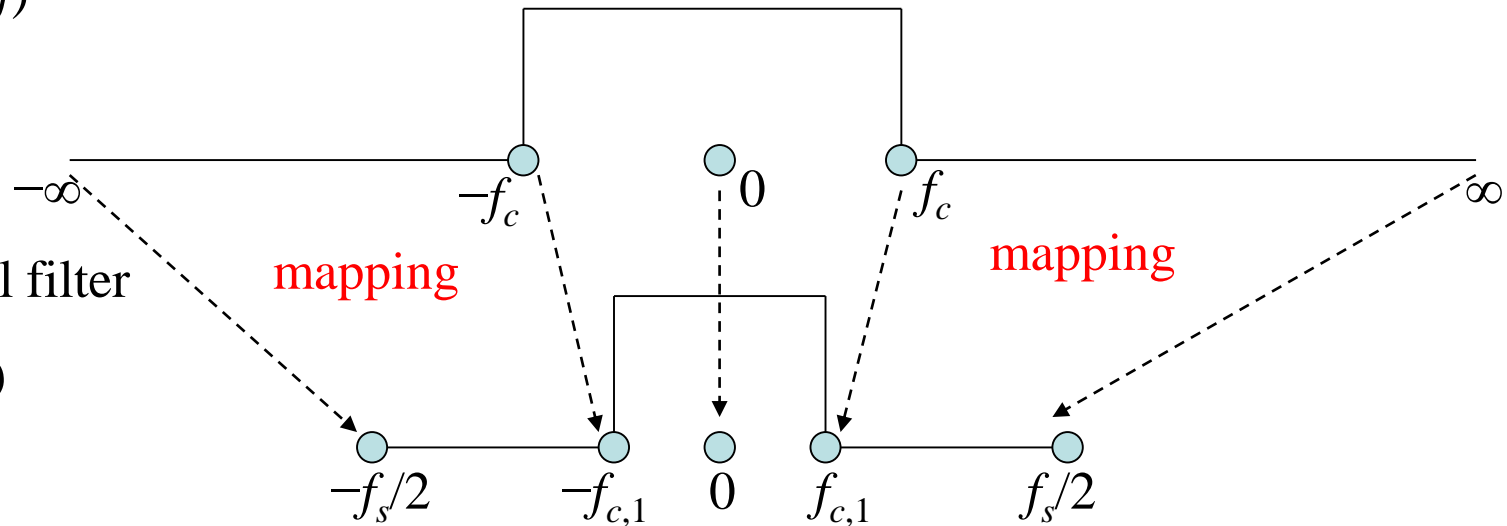


analog filter

$$H_a(f)$$

digital filter

$$H(f)$$



$$f_{c,1} = \frac{f_s}{\pi} \operatorname{atan} \left( \frac{2\pi}{c} f_c \right)$$

Advantage of the bilinear transform

Disadvantage of the bilinear transform

## 附錄一：學習 DSP 知識把握的要點

- (1) **Concepts**: 這個方法的核心概念、基本精神是什麼
- (2) **Comparison**: 這方法和其他方法之間，有什麼相同的地方？  
有什麼相異的地方
- (3) **Advantages**: 這方法的優點是什麼  
(3-1) Why? 造成這些優點的原因是什麼
- (4) **Disadvantages**: 這方法的缺點是什麼  
(4-1) Why? 造成這些缺點的原因是什麼
- (5) **Applications**: 這個方法要用來處理什麼問題，有什麼應用
- (6) **Innovations**: 這方法有什麼可以改進的地方  
或是可以推廣到什麼地方