

EXPERIMENT-2,3

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Problem 1(a,b):

Aim:

To design a single stage low pass filter with a cutoff frequency of $f_c = 64\text{kHz}$

Material Required:

Breadboard, Resistor($2.5\text{k}\Omega$), capacitor(1nF), multimeter, Oscilloscope

Theory:

A low pass filter is a circuit that attenuates all the signal above a certain frequency level and lets the frequency below that level pass through it.

The cutoff frequency is

$$f = \frac{1}{2\pi RC} \text{Hz}$$

When the voltage source of frequency w is applied as an input. The impedance of the capacitor is

$$X_c = \frac{1}{jwC}$$

From the above equation, we can say that the impedance of the capacitor decreases as the frequency w increases of the input voltage.

The gain of the circuit is

$$gain = 20 \log\left(\frac{V_{out}}{V_{in}}\right)$$

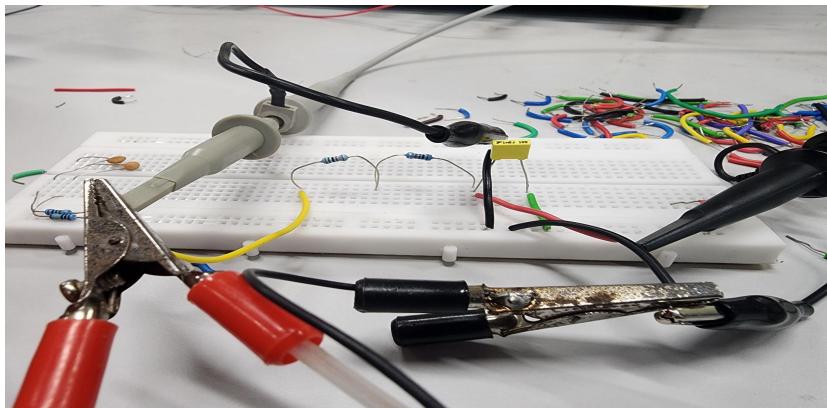
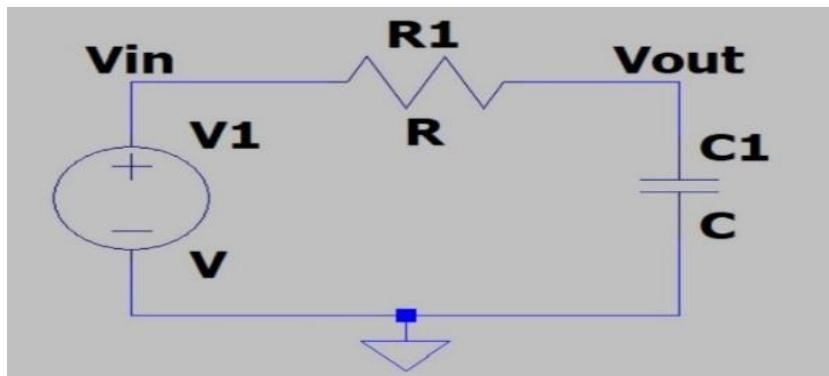
$$V_{out} = V_{in} \frac{R}{(R + \frac{1}{j\omega C})}$$

The phase is given by,

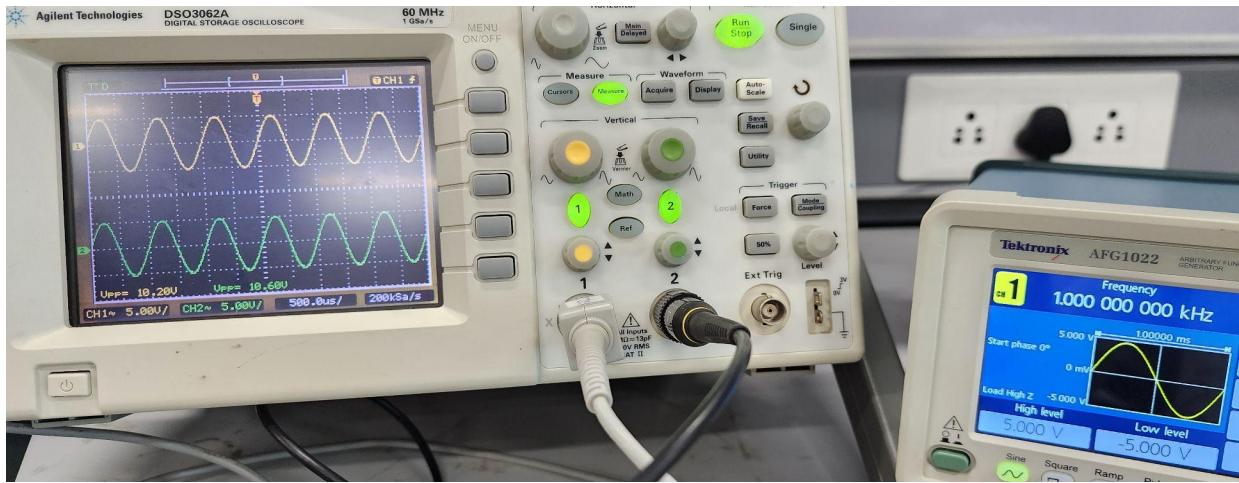
$$Phase = \tan^{-1}\left(\frac{1}{\omega RC}\right)$$

Procedure:

1. We connect the RC circuit in series, giving input from the resistor and taking output from capacitor.
2. Taking R=2.5k and C=1nF
3. We record the gain and phase difference from 10hz and 1.28Mhz

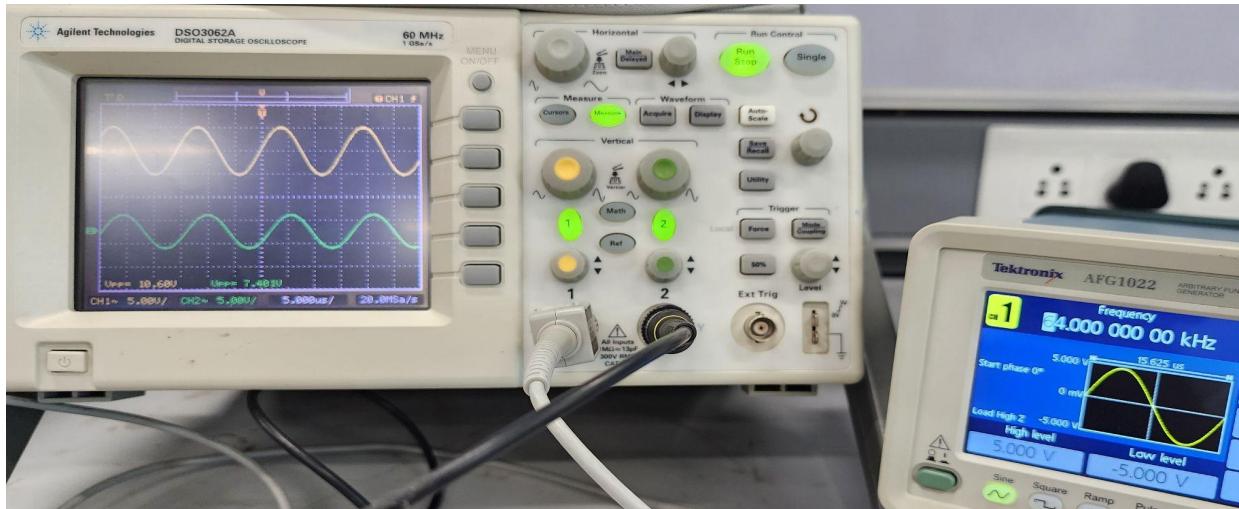


Observations:



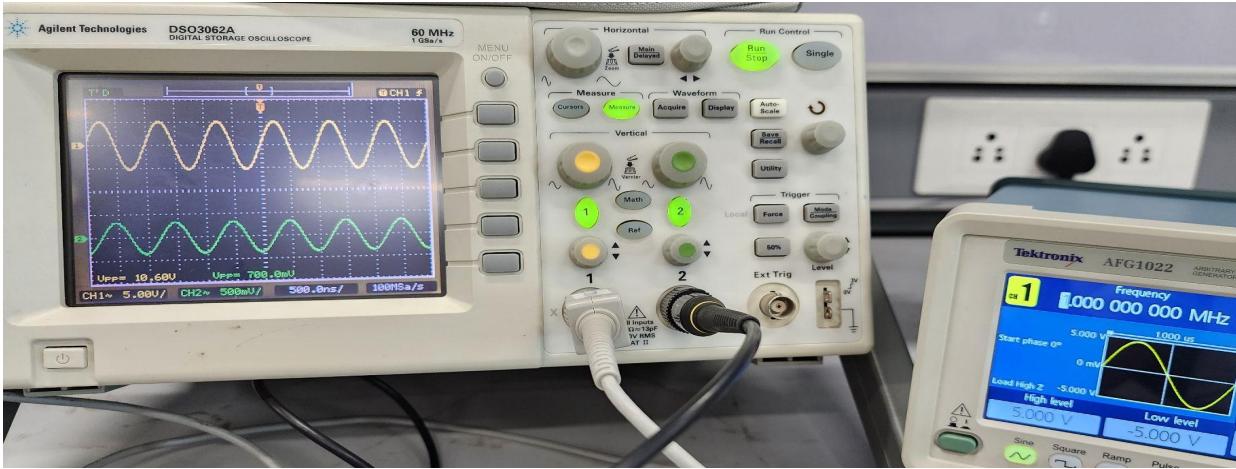
For lower frequency, we observe that

- Experimental gain = 0 and LTspice gain ≈ 0
- Experimental phase diff. = -0.36 and LTspice = -9.0E-03



For cutoff frequency, we observe that

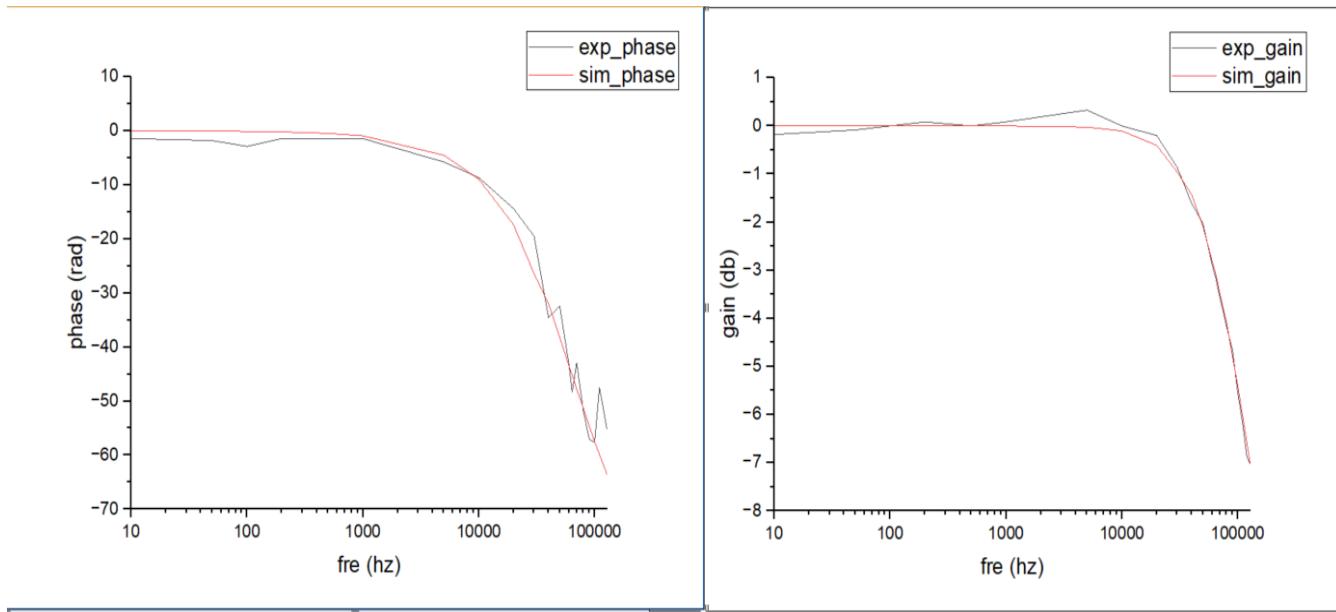
- Experimental gain = -3.12 and LTspice gain = -48.32
- Experimental phase diff. = -3.04 and LTspice = -45.15



For higher frequency, we observe that

- Experimental gain = -26.6 and LTspice gain = -82.94
- Experimental phase diff. = -26.07 and LTspice = -87.15

| | A(X) | B(Y) | C(Y) | D(Y) | E(Y) |
|-----------|--------|----------|-----------|----------|-----------|
| Long Name | fre | gain | phase | gain | phase |
| Units | hz | db | rad | db | rad |
| Comments | | exp_gain | exp_phase | sim_gain | sim_phase |
| F(x)= | | | | | |
| 1 | 10 | -0.1791 | -1.44 | -1.07E-7 | -0.009 |
| 2 | 50 | -0.08392 | -1.8 | -2.69E-6 | -0.0451 |
| 3 | 100 | 0 | -2.88 | -1.07E-5 | -0.09 |
| 4 | 200 | 0.08312 | -1.44 | -4.27E-5 | -0.18 |
| 5 | 500 | 0 | -1.44 | -2.69E-4 | -0.451 |
| 6 | 1000 | 0.08312 | -1.44 | -0.00107 | -0.9 |
| 7 | 5000 | 0.32781 | -5.76 | -0.0268 | -4.5 |
| 8 | 10000 | 0 | -8.64 | -0.106 | -8.93 |
| 9 | 20000 | -0.20279 | -14.4 | -0.407 | -17.4 |
| 10 | 30000 | -0.85979 | -19.44 | -0.958 | -26.4 |
| 11 | 40000 | -1.61646 | -34.56 | -1.43 | -32 |
| 12 | 50000 | -2.02053 | -32.4 | -2.09 | -38.2 |
| 13 | 60000 | -2.88985 | -43.2 | -2.801 | -43.47 |
| 14 | 64000 | -3.12148 | -48.384 | -3.04 | -45.15 |
| 15 | 70000 | -3.55466 | -42.84 | -3.48 | -47.75 |
| 16 | 80000 | -4.16804 | -51.84 | -4.08 | -51.3 |
| 17 | 90000 | -4.65828 | -57.024 | -4.79 | -54.64 |
| 18 | 100000 | -5.54236 | -57.6 | -5.4 | -57.5 |
| 19 | 110000 | -6.18605 | -47.52 | -6.03 | -59.89 |
| 20 | 120000 | -6.88129 | -51.84 | -6.59 | -61.97 |
| 21 | 128000 | -7.02728 | -55.296 | -7.03 | -63.6 |

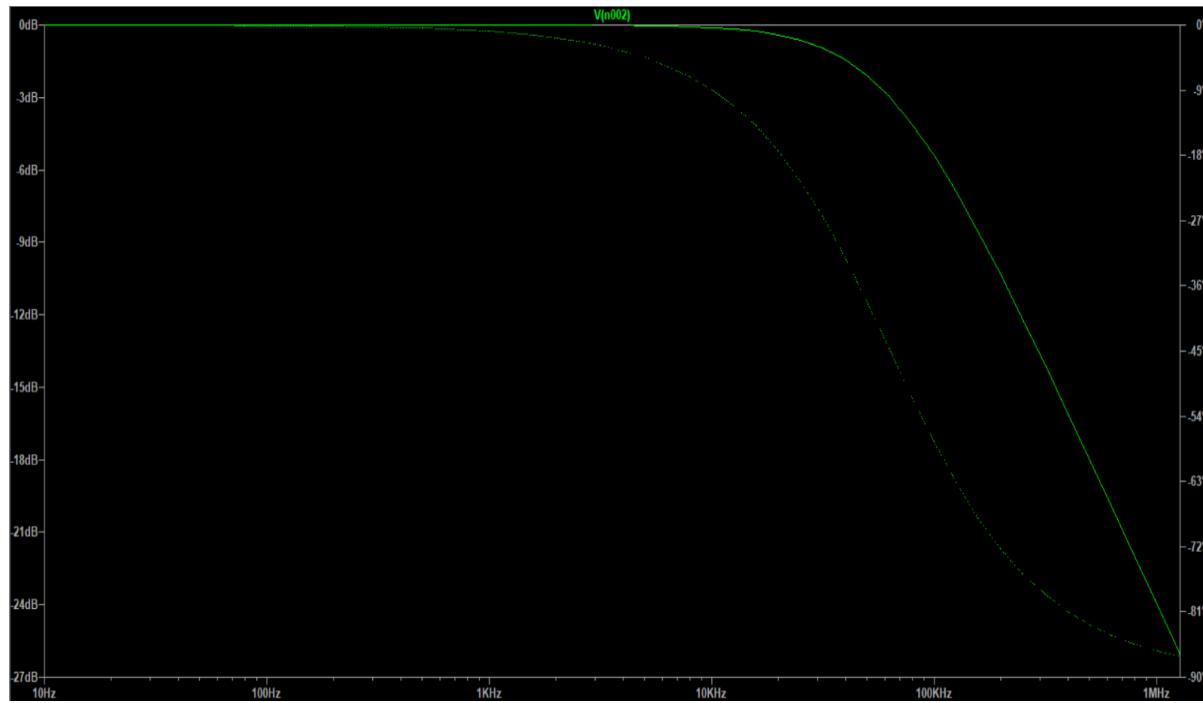


Deviation from Experimental and LTspice model,

The experimentally obtained roll-off slope is: $-22.444 - (-5.542) = -16.902 \text{ dB/decade}$

Simulation:

Graph derived from LTspice



Experimental Value of R and C are $2.5\text{k}\Omega$ and 1nF

Tweaked Value of R and C are $2.4\text{k}\Omega$ and 1nF

Problem 1(c):

Aim:

To repeat the same experiment as in 1(a,b) but with a two stage filter for same value of R,C.

Material Required:

Breadboard, Resistor($2.5\text{k}\Omega$), capacitor(1nF), multimeter, Oscilloscope and function generator

Theory:

A double-stage low-pass filter combines two single-stage filters in series. This configuration offers several advantages over a single-stage filter, including enhanced attenuation of high-frequency noise and a steeper roll-off slope.

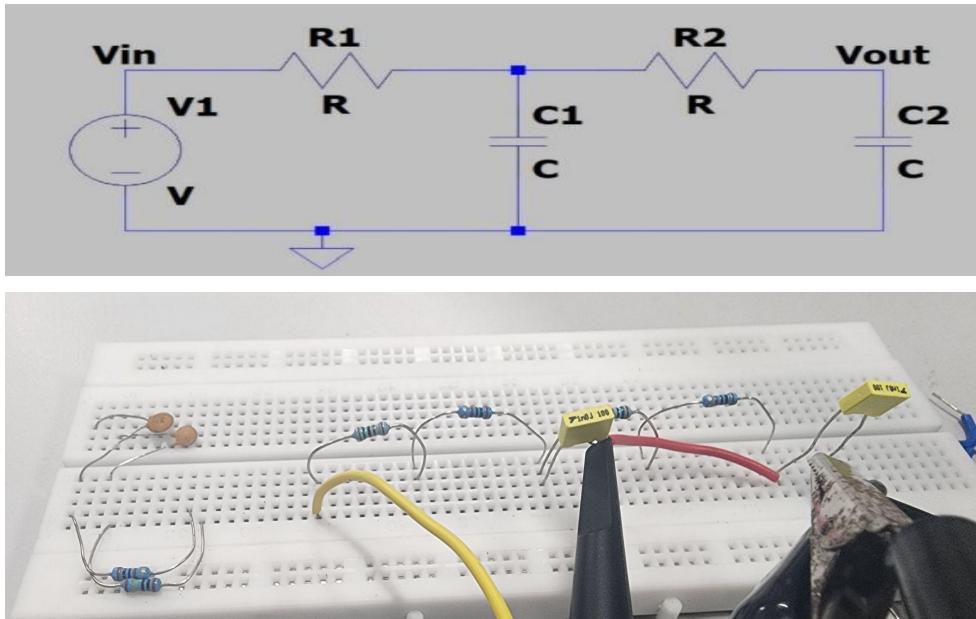
The cutoff of the double-stage filter remains a function of the individual resistor and capacitor values in each stage but is influenced by their combined effect. It is typically calculated as:

$$f_c = \frac{1}{2\pi\sqrt{R_1 C_1 R_2 C_2}}$$

The double-stage filter exhibits a steeper roll-off slope compared to a single-stage filter. The exact roll-off rate depends on the filter order. A double-stage filter has a second-order roll-off.

Procedure:

1. We connect two single stage filter in series,giving input from the resistor of first stage and taking output from capacitor of second stage.
2. Taking $R=2.5\text{k}$ and $C=1\text{n}$
3. We record the gain and phase difference from 10hz and 1.28Mhz

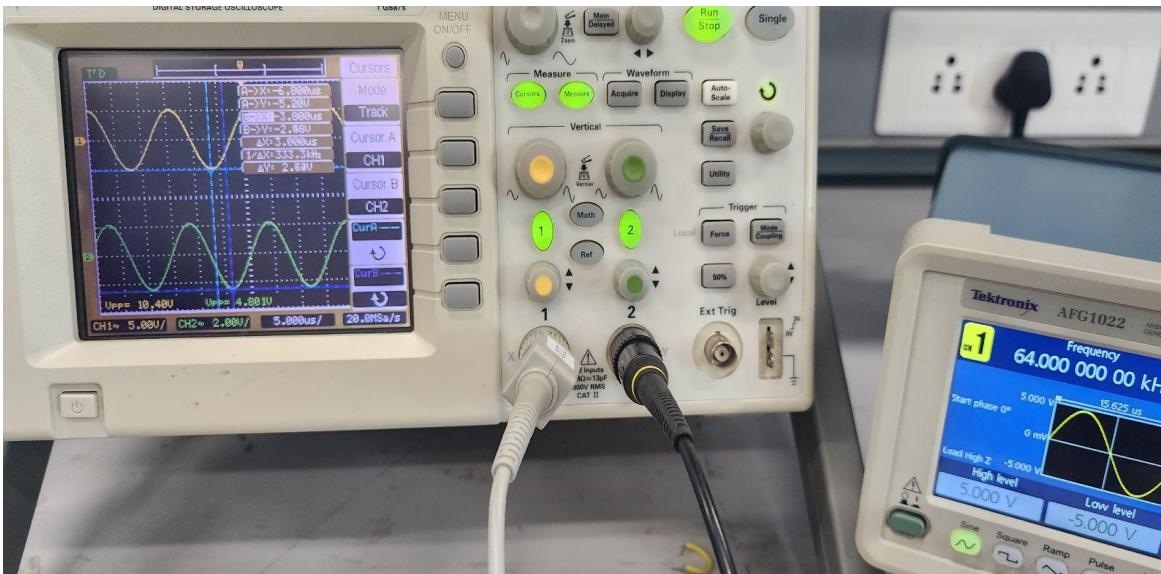


Observations:



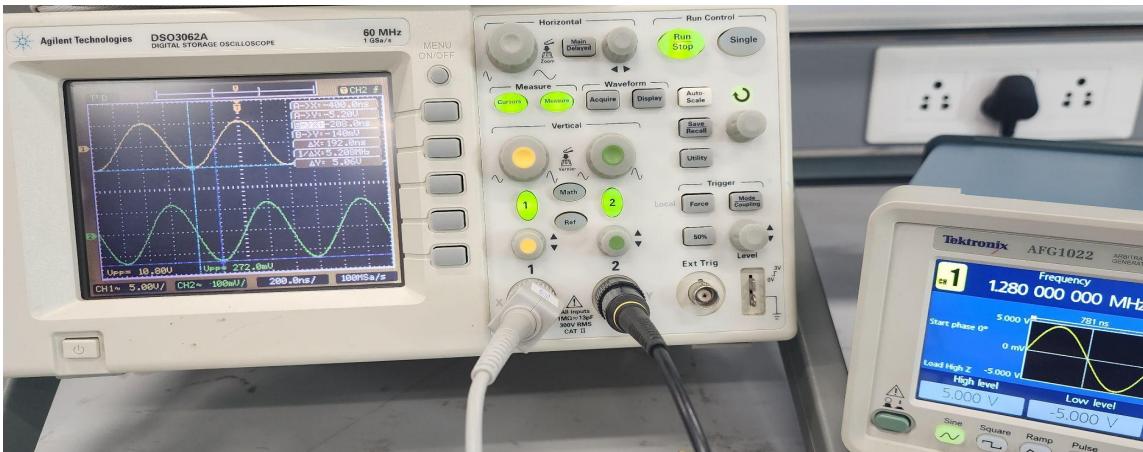
For lower frequency, we observe that

- Experimental gain = 0.3 and LTspice gain = $-134.13E-09$
- Experimental phase diff. = 0 and LTspice = $-2.7E-02$



For cutoff frequency, we observe that

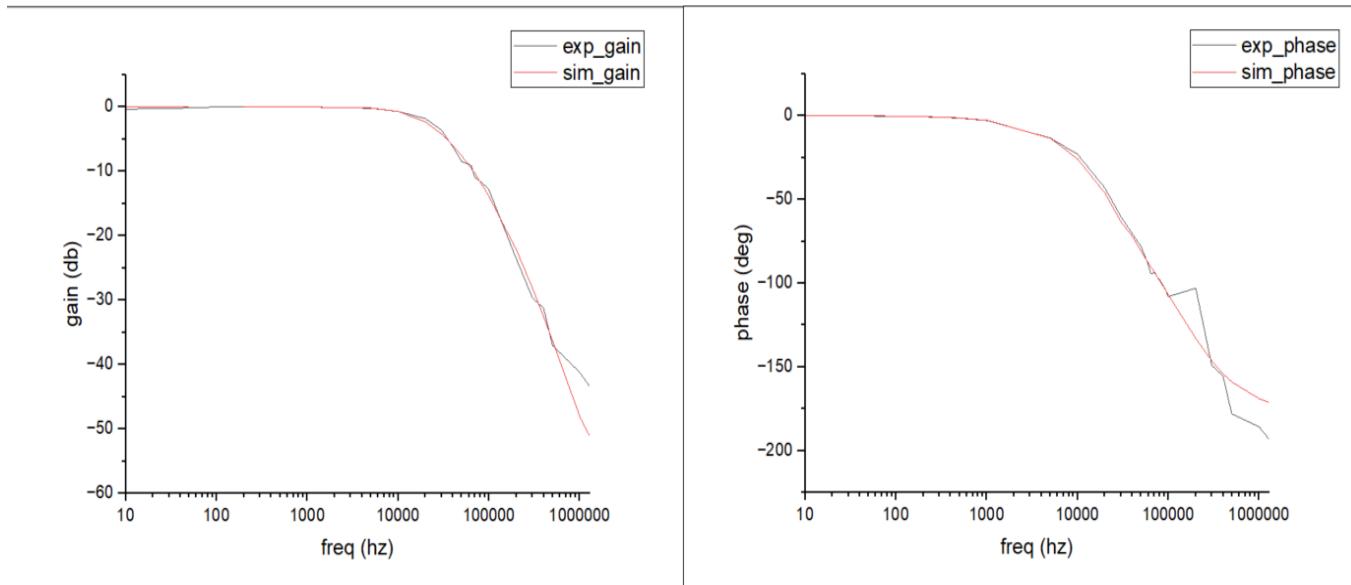
- Experimental gain = -9.18 and LTspice gain = -9.6
- Experimental phase diff. = -94.464 and LTspice = -90.38



For higher frequency, we observe that

- Experimental gain = -0.272 and LTspice gain = -51.1
- Experimental phase diff. = -193.13 and LTspice = -1.71E+02

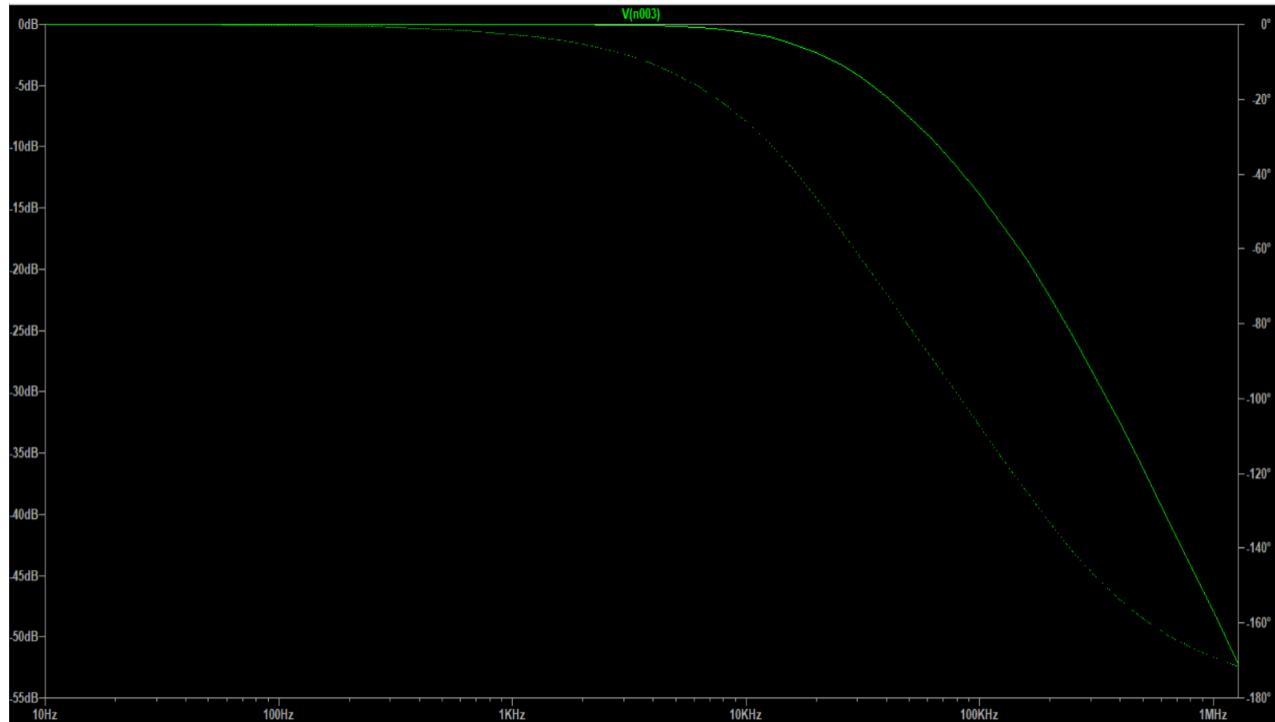
| | A(X) | B(Y) | C(Y) | D(Y) | E(Y) |
|-----------|---------|-----------|-----------|----------|-----------|
| Long Name | freq | gain | phase | gain | phase |
| Units | hz | db | deg | db | deg |
| Comments | | exp_gain | exp_phase | sim_gain | sim_phase |
| F(x)= | | | | | |
| 1 | 10 | -0.35458 | -0.02916 | 0 | -0.027 |
| 2 | 50 | -0.16866 | -0.153 | -1.8E-5 | -0.135 |
| 3 | 100 | 0 | -0.2988 | -74u | -0.27 |
| 4 | 200 | 0 | -0.5832 | -2.97E-4 | -0.539 |
| 5 | 500 | 0 | -1.44 | -0.00186 | -1.35 |
| 6 | 1000 | 0 | -2.88 | -0.00756 | -2.7 |
| 7 | 5000 | -0.19808 | -13.32 | -0.185 | -13.4 |
| 8 | 10000 | -0.69434 | -23.04 | -0.696 | -25.8 |
| 9 | 20000 | -1.81514 | -43.2 | -2.315 | -46.2 |
| 10 | 30000 | -3.60292 | -60.264 | -4.18 | -63.2 |
| 11 | 40000 | -6.18438 | -70.272 | -5.95 | -72 |
| 12 | 50000 | -8.46275 | -78.12 | -7.58 | -80.9 |
| 13 | 60000 | -8.90816 | -88.992 | -9.07 | -87.89 |
| 14 | 64000 | -9.1868 | -94.464 | -9.6 | -90.38 |
| 15 | 70000 | -10.9608 | -93.744 | -10.3 | -93.19 |
| 16 | 80000 | -11.55986 | -97.92 | -11.67 | -98.48 |
| 17 | 90000 | -12.20331 | -102.384 | -12.83 | -102.86 |
| 18 | 100000 | -12.89827 | -108 | -13.911 | -107 |
| 19 | 200000 | -23.60416 | -102.96 | -22.19 | -133 |
| 20 | 300000 | -29.62476 | -149.04 | -28.08 | -146.23 |
| 21 | 400000 | -31.25816 | -155.52 | -32.6 | -154 |
| 22 | 500000 | -36.98429 | -178 | -36.23 | -159 |
| 23 | 1000000 | -41.23036 | -185.8 | -48.09 | -169 |
| 24 | 1280000 | -43.35947 | -193.104 | -51.1 | -171 |
| 25 | | | | | |



Deviation from experimental and LTspice model,

Simulation:

Graph derived from the LTspice



Experimental Value of R_1, R_2 and C_1, C_2 are $2.5\text{k}\Omega, 2.5\text{k}\Omega$ and $1\text{nF}, 1\text{nF}$

Tweaked Value of R_1, R_2 and C_1, C_2 are $2.35\text{k}\Omega, 2.91\text{k}\Omega$ and $1\text{nF}, 1\text{nF}$

Problem 2(a,b):

Aim:

To design a better double stage filter

- Passive corrected filter
- Active corrected filter

Material Required:

capacitors(0.1nF), resistor($25\text{k}\Omega$), multimeter, Oscilloscope and function generator.

Theory:

The primary reason for the difference in cutoff frequency (f_c), in practice, between a single and a double-stage filter is due to the characteristics and constraints of the electronic devices used in the filters.

Resistors, capacitors, inductors, etc. have tolerances, and parasitic effects that cause variations in their true values, which affect the behavior of the filter.

When the filter is in the second stage, the first stage that precedes it actually loads the second stage.

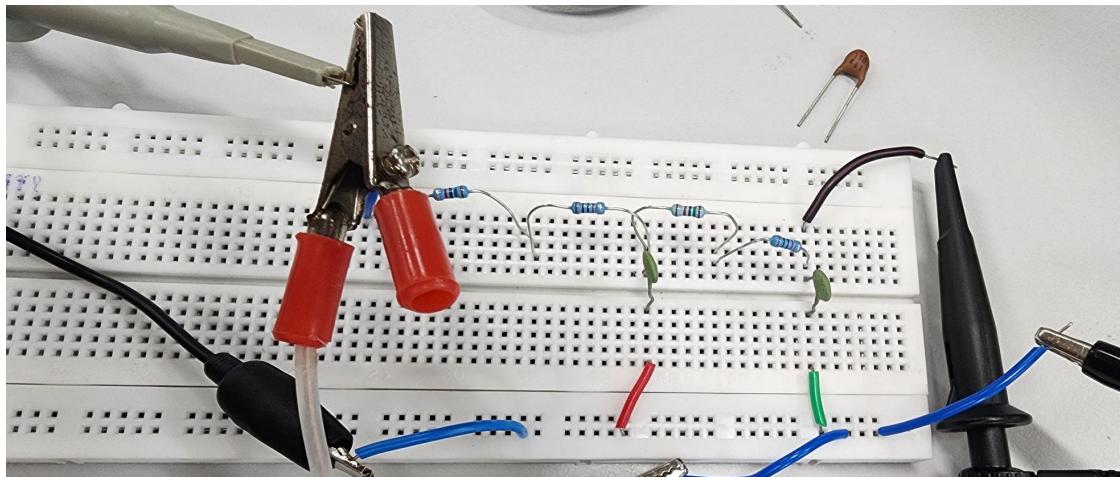
This load introduces resistance and influences the overall performance of a cascaded filter. Because the first stage changes the input impedance and interactions between the stages, it is difficult to isolate the behaviour of the second stage itself.

Measurement of the precise phase and gain at the output becomes difficult.

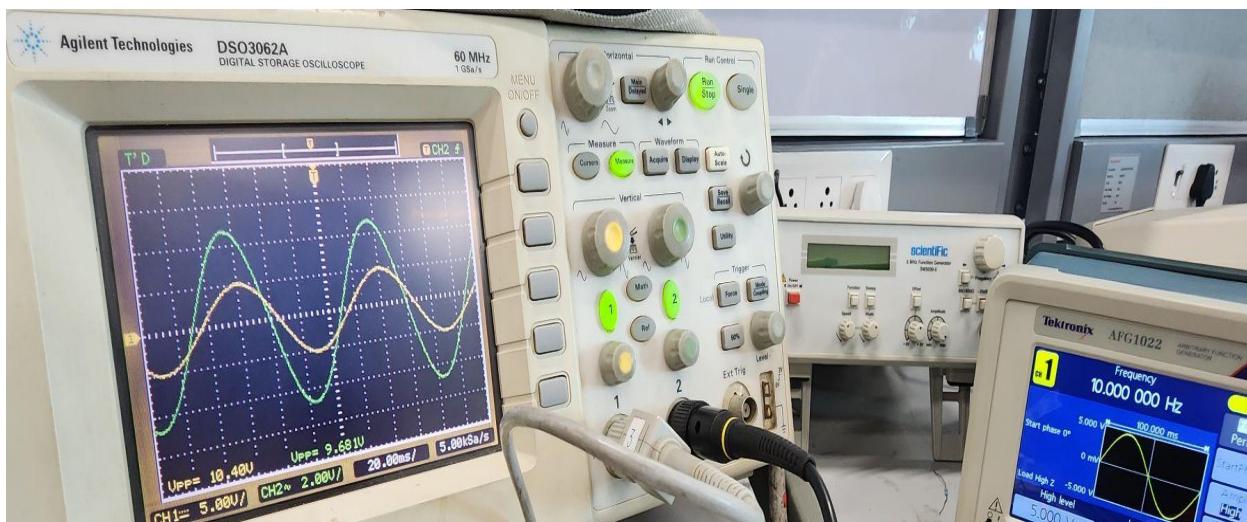
To solve this issue, we design the second stage passive high pass filter. The basic idea behind this filter is to increase the impedance within a cascaded filter configuration. By increasing the output impedance we are effectively increasing the output impedance so that we can measure a more magnified representation of the initial signal. As a result, this improves the gain and

allows for more accurate measurements of the cut-off frequency (f_c) so that we can better characterize the filter's performance.

Procedure:



Observations:



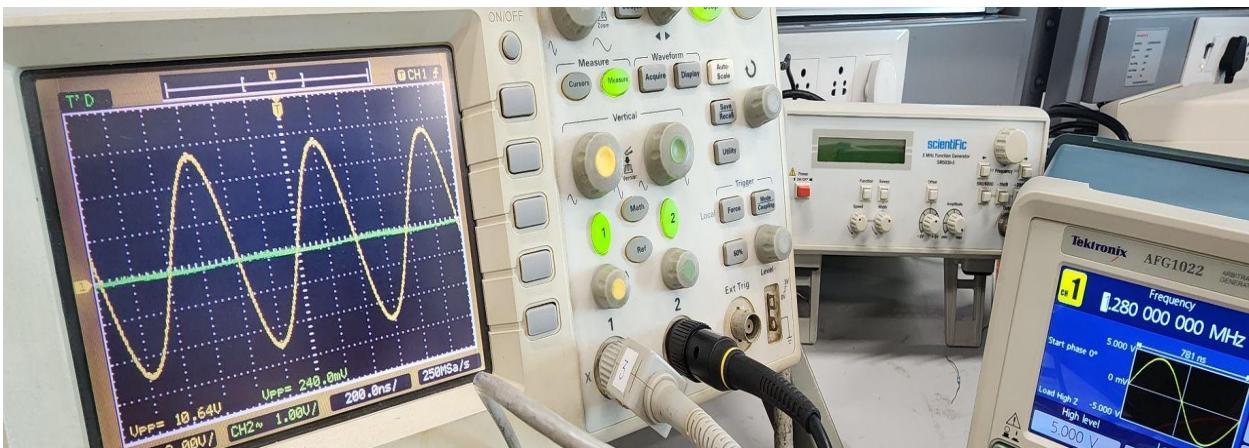
For lower frequency, we observe that

- Experimental gain = 0.3 and LTspice gain = -258E-09
- Experimental phase diff. = 0 and LTspice = -1.89E-02



For cutoff frequency, we observe that

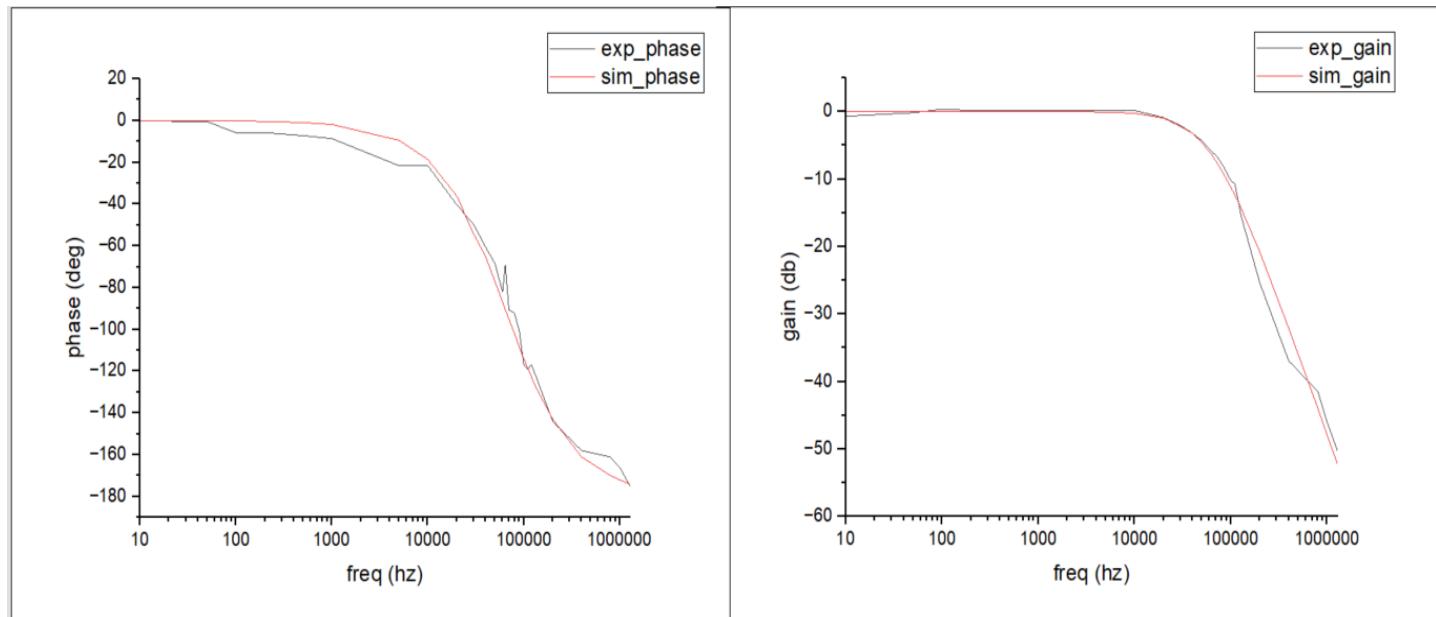
- Experimental gain = -6.429 and LTspice gain = -7.32
- Experimental phase diff. = -9.07E+01 and LTspice = -94.98



For higher frequency, we observe that

- Experimental gain = -43.69 and LTspice gain = -4.39E+01
- Experimental phase diff. = -1.84E+02 and LTspice = -1.70E+02

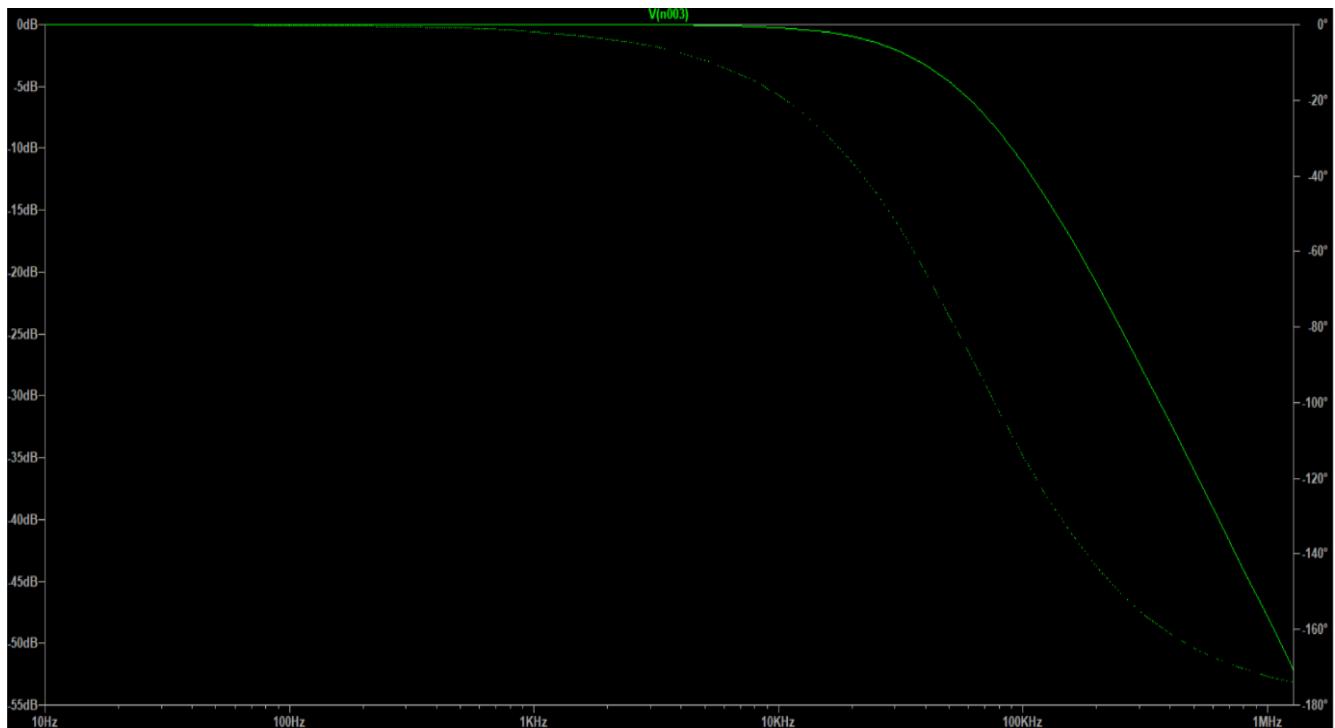
| | A(X) | B(Y) | C(Y) | D(Y) | E(Y) |
|-----------|---------|-----------|-----------|----------|------------|
| Long Name | freq | gain | phase | gain | phase |
| Units | hz | db | deg | db | deg |
| Comments | | exp_gain | exp_phase | sim_gain | sim_phase |
| F(x)= | | | | | |
| 1 | 10 | -0.69524 | -0.144 | -2.58E-7 | -0.0189 |
| 2 | 50 | -0.16866 | -0.54 | -6.49E-6 | -0.0947 |
| 3 | 100 | 0.32781 | -5.76 | -2.58E-5 | -0.189 |
| 4 | 200 | 0.16545 | -5.76 | -1.03E-4 | -0.377 |
| 5 | 500 | 0.16545 | -7.2 | -6.49E-4 | -0.947 |
| 6 | 1000 | 0.16545 | -8.64 | -0.00258 | -1.89 |
| 7 | 5000 | 0.16545 | -21.6 | -0.0646 | -9.45 |
| 8 | 10000 | 0.16545 | -21.6 | -0.253 | -18.7 |
| 9 | 20000 | -0.87719 | -40.3 | -0.957 | -36.1 |
| 10 | 30000 | -2.06333 | -49.7 | -2.19 | -54.2 |
| 11 | 40000 | -3.19281 | -60.5 | -3.21 | -65.1 |
| 12 | 50000 | -4.32495 | -68.4 | -4.59 | -77 |
| 13 | 60000 | -5.50031 | -82.1 | -5.998 | -86.92785 |
| 14 | 64000 | -6.01893 | -69.1 | -6.492 | -90.5082 |
| 15 | 70000 | -6.42928 | -90.7 | -7.32 | -94.98919 |
| 16 | 80000 | -7.62898 | -92.2 | -8.69 | -102 |
| 17 | 90000 | -8.92747 | -101 | -9.987 | -108.60398 |
| 18 | 100000 | -10.23495 | -117 | -11.2 | -114 |
| 19 | 110000 | -10.68034 | -119 | -12.368 | -118.5388 |
| 20 | 120000 | -13.33183 | -117 | -13.433 | -122.50248 |
| 21 | 128000 | -15.42548 | -120 | -14.309 | -125.95298 |
| 22 | 200000 | -25.37691 | -144 | -20.8 | -143 |
| 23 | 400000 | -36.93543 | -158 | -32.1 | -161 |
| 24 | 800000 | -41.45101 | -161 | -43.9 | -170 |
| 25 | 1000000 | -46.0206 | -166 | -47.9 | -172 |
| 26 | 1280000 | -50.23767 | -175 | -52.2 | -174 |



Deviation from experimental and LTspice model,

Simulation:

Graph derived from LTspice model



Problem 2(c):

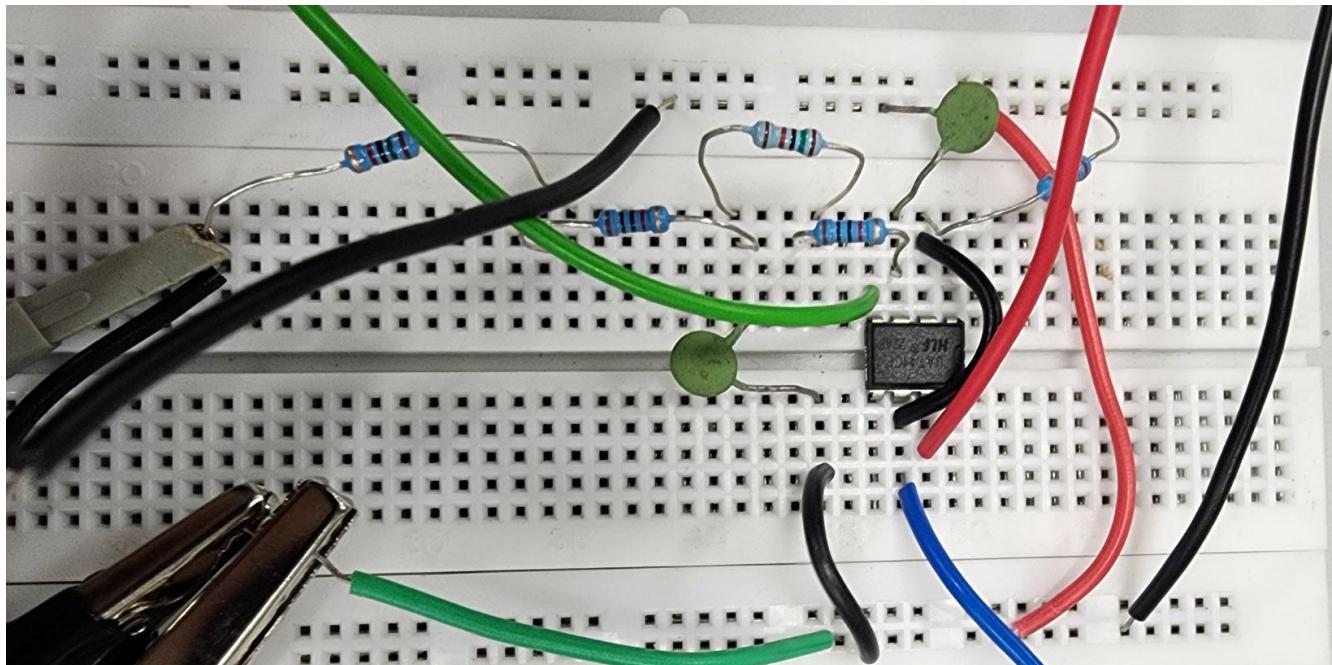
Aim:

To design an active low pass filter.

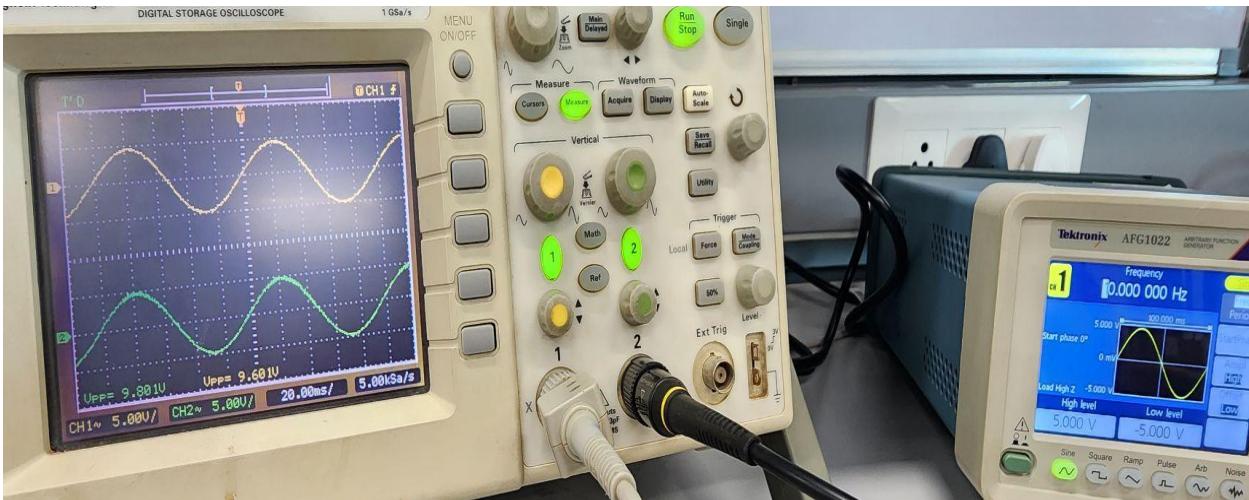
Material Required:

25k Ω resistor ,0.1nF capacitor, multimeter, breadboard, wires, function generator, oscilloscope.

Procedure:

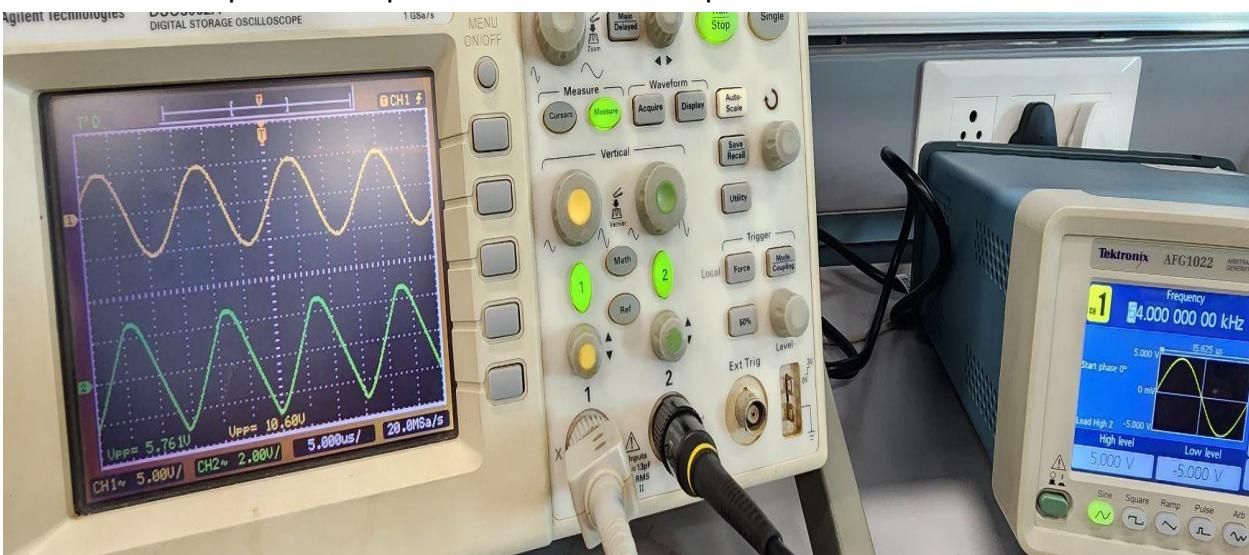


Observations



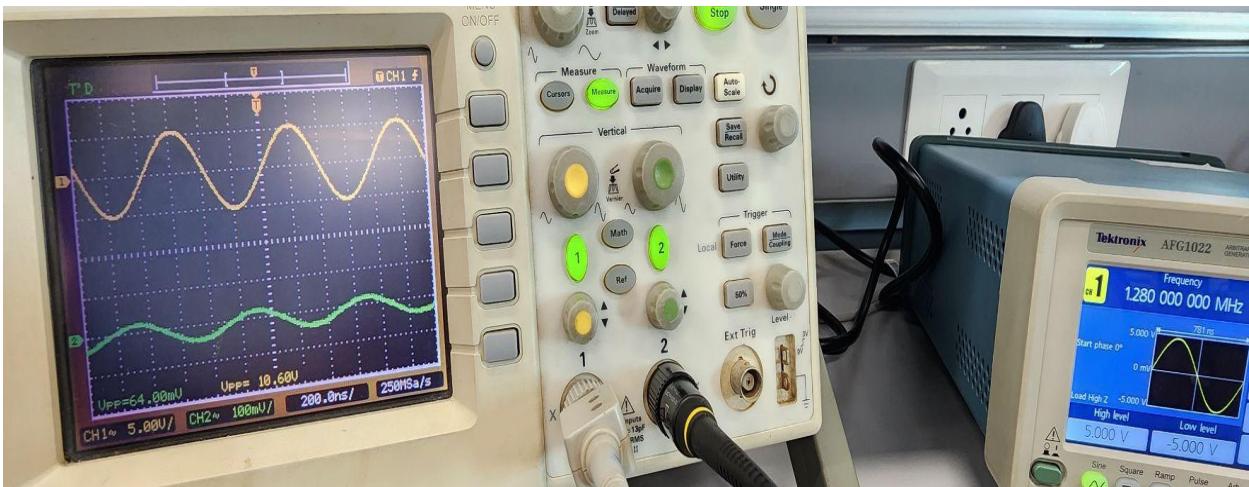
For lower frequency, we observe that

- Experimental gain = -0.179 and LTspice gain = 2.50E-04
- Experimental phase diff. = 0 and LTspice = -1.86E-02



For higher frequency, we observe that

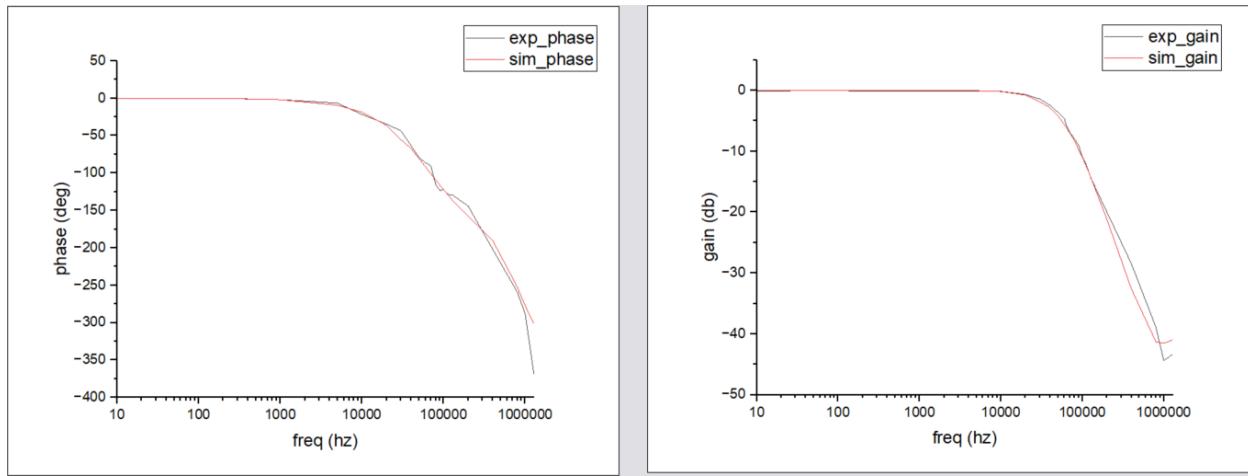
- Experimental gain = -5.858 and LTspice gain = -6.182
- Experimental phase diff. = -87.552 and LTspice = -95.647



For higher frequency, we observe that

- Experimental gain = -44.382 and LTspice gain = -4.10E+01
- Experimental phase diff. = -368.64 and LTspice = -301.293

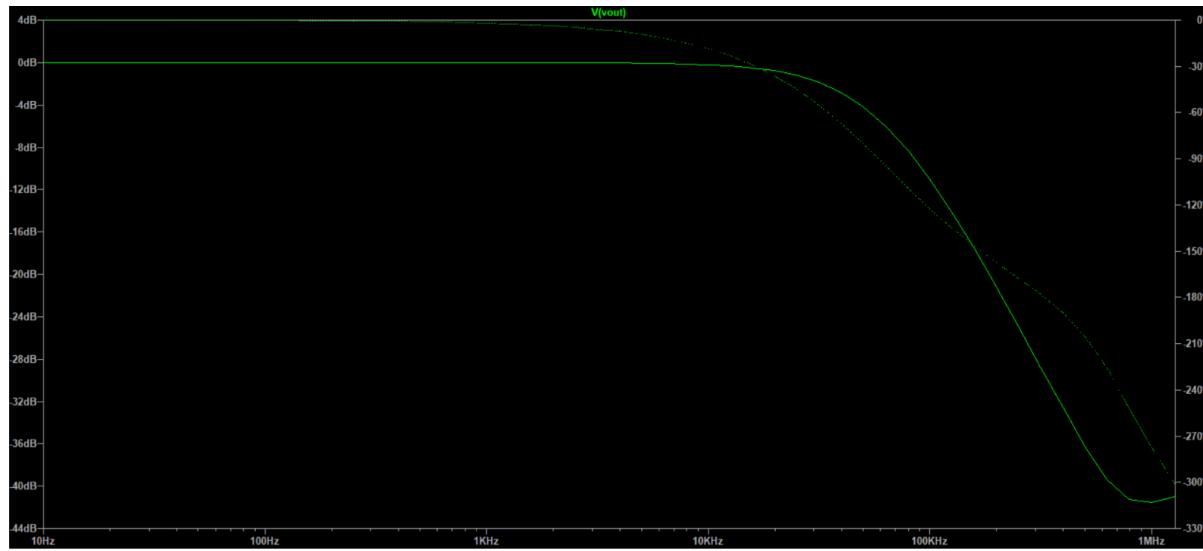
| | A(X) | B(Y) | C(Y) | D(Y) | E(Y) |
|-----------|---------|-----------|-----------|----------|-----------|
| Long Name | freq | gain | phase | gain | phase |
| Units | hz | db | deg | db | deg |
| Comments | | exp_gain | exp_phase | sim_gain | sim_phase |
| F(x)= | | | | | |
| 1 | 10 | -0.1791 | 0 | 2.5E-4 | -0.0186 |
| 2 | 50 | 0 | -0.036 | 2.45E-4 | -0.0933 |
| 3 | 100 | 0 | -0.1152 | 2.3E-4 | -0.186 |
| 4 | 200 | -0.16545 | -0.2448 | 1.69E-4 | -0.371 |
| 5 | 500 | -0.16545 | -1.08 | -2.66E-4 | -0.933 |
| 6 | 1000 | -0.16545 | -1.872 | -0.00181 | -1.86 |
| 7 | 5000 | -0.16545 | -6.48 | -0.0513 | -9.31 |
| 8 | 10000 | -0.16545 | -21.6 | -0.204 | -18.5 |
| 9 | 20000 | -0.6816 | -34.56 | -0.79 | -36.2 |
| 10 | 30000 | -1.42127 | -43.2 | -1.88 | -55.2 |
| 11 | 40000 | -2.44432 | -63.36 | -2.84 | -67.1 |
| 12 | 50000 | -3.60416 | -79.2 | -4.19 | -80.4 |
| 13 | 60000 | -4.60243 | -86.4 | -5.624 | -91.588 |
| 14 | 64000 | -5.85824 | -87.552 | -6.182 | -95.647 |
| 15 | 70000 | -6.88129 | -90.72 | -7.003 | -100.748 |
| 16 | 80000 | -7.9588 | -115.2 | -8.29 | -109 |
| 17 | 90000 | -9.00236 | -123.12 | -9.789 | -116.23 |
| 18 | 100000 | -10.9608 | -122.4 | -11 | -122 |
| 19 | 110000 | -12.00522 | -126.72 | -12.31 | -127.611 |
| 20 | 120000 | -13.42011 | -129.6 | -13.43 | -132.613 |
| 21 | 128000 | -14.48118 | -129.024 | -14.46 | -135.731 |
| 22 | 200000 | -19.99181 | -144 | -21.2 | -157 |
| 23 | 400000 | -28.46492 | -201.6 | -32.6 | -190.049 |
| 24 | 800000 | -38.92249 | -259.2 | -41.3 | -252.306 |
| 25 | 1000000 | -44.38252 | -288 | -41.5 | -277.432 |
| 26 | 1280000 | -43.35947 | -368.64 | -41 | -301.293 |



Deviation from experimental and LTspice model

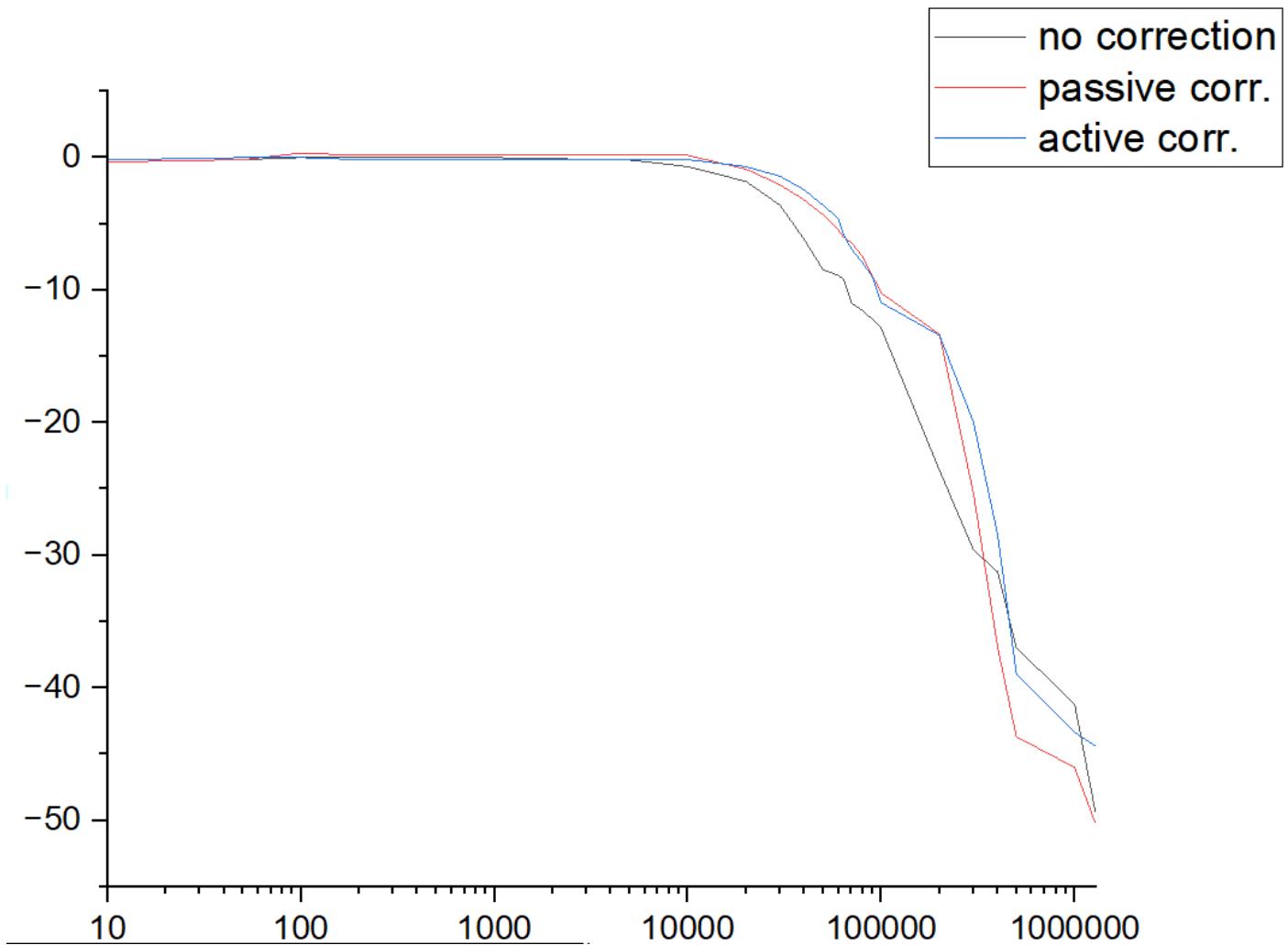
Simulation:

Graph derived from LTspice model



Problem 2d:

| | A(X) | B(Y) | C(Y) | D(Y) |
|-----------|---------|--------------|--------------|--------------|
| Long Name | | | | |
| Units | | | | |
| Comments | | no correctio | passive corr | active corr. |
| F(x)= | | | | |
| 1 | 10 | -0.35458 | -0.35458 | -0.1791 |
| 2 | 50 | -0.16866 | -0.16866 | 0 |
| 3 | 100 | 0 | 0.32781 | 0 |
| 4 | 200 | 0 | 0.16545 | -0.16545 |
| 5 | 500 | 0 | 0.16545 | -0.16545 |
| 6 | 1000 | 0 | 0.16545 | -0.16545 |
| 7 | 5000 | -0.19808 | 0.16545 | -0.16545 |
| 8 | 10000 | -0.69434 | 0.16545 | -0.16545 |
| 9 | 20000 | -1.81514 | -0.87719 | -0.6816 |
| 10 | 30000 | -3.60292 | -2.06333 | -1.42127 |
| 11 | 40000 | -6.18438 | -3.19281 | -2.44432 |
| 12 | 50000 | -8.46275 | -4.32495 | -3.60416 |
| 13 | 60000 | -8.90816 | -5.50031 | -4.60243 |
| 14 | 64000 | -9.1868 | -6.01893 | -5.85824 |
| 15 | 70000 | -10.9608 | -6.42928 | -6.88129 |
| 16 | 80000 | -11.55986 | -7.46964 | -7.9588 |
| 17 | 90000 | -12.20331 | -8.92747 | -9.00236 |
| 18 | 100000 | -12.89827 | -10.23495 | -10.9608 |
| 19 | 200000 | -23.60416 | -13.33183 | -13.42011 |
| 20 | 300000 | -29.62476 | -25.37691 | -19.99181 |
| 21 | 400000 | -31.25816 | -36.93543 | -28.46492 |
| 22 | 500000 | -36.98429 | -43.69049 | -38.92249 |
| 23 | 1000000 | -41.23036 | -46.0206 | -43.35947 |
| 24 | 1280000 | -49.38007 | -50.23767 | -44.38252 |
| 25 | | | | |



Problem 3:

Aim:

To design band pass LC filter with lower cut off frequency equal 5khz and higher cut off frequency equal 250khz.

Material Required:

Breadboard, Resistor($19.6\text{k}\Omega$), capacitor(2.2nF), inductor(4mH), multimeter, Oscilloscope and function generator

Theory:

A bandpass filter is a type of electronic circuit that uses a combination of capacitors and inductors to selectively filter out certain frequencies while allowing others to pass through. In this experiment, we have to create a bandpass filter with a lower cutoff frequency (f_l) of 5 kHz and a higher cutoff frequency (f_h) of 250 kHz.

$f_l = 7 \text{ kHz}$ and $f_h = 350 \text{ kHz}$.

To determine the necessary values for the resistance (R), capacitance (C), and inductance (L) in the circuit, we need to solve two equations,

$$2\pi f_l = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

$$2\pi f_h = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

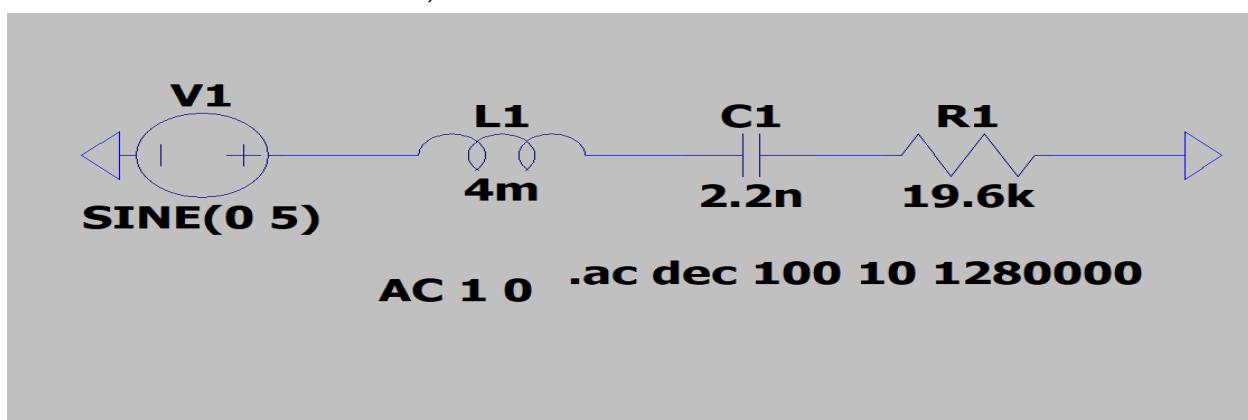
By solving the above two equations, we get,

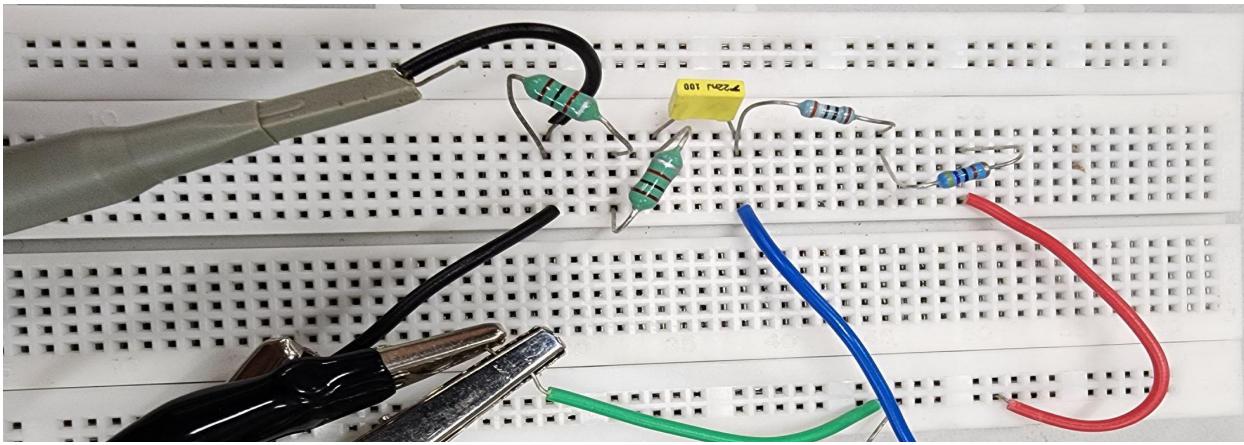
$L = 4 \text{ mH}$, $C = 0.5 \text{ nF}$ and $R = 19.6 \text{k}\Omega$.

And, centre frequency = $\sqrt{f_l * f_h} = 35 \text{ kHz}$

Procedure:

We connect the RLC in series,



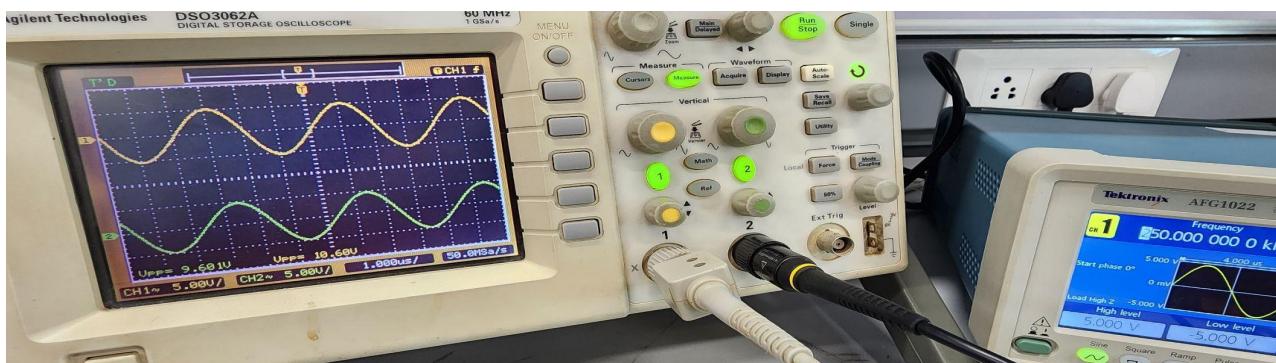


Observations:



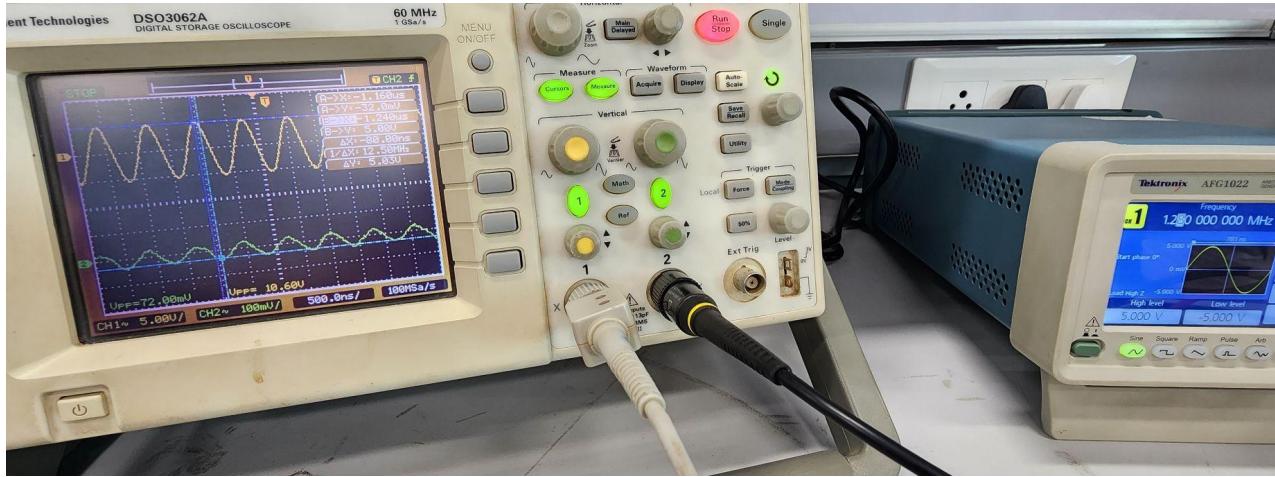
For lower frequency, we observe that

- Experimental gain = -0.368 and LTspice gain = -134.13E-09
- Experimental phase diff. = 0 and LTspice = -2.7E-02



For cutoff frequency, we observe that

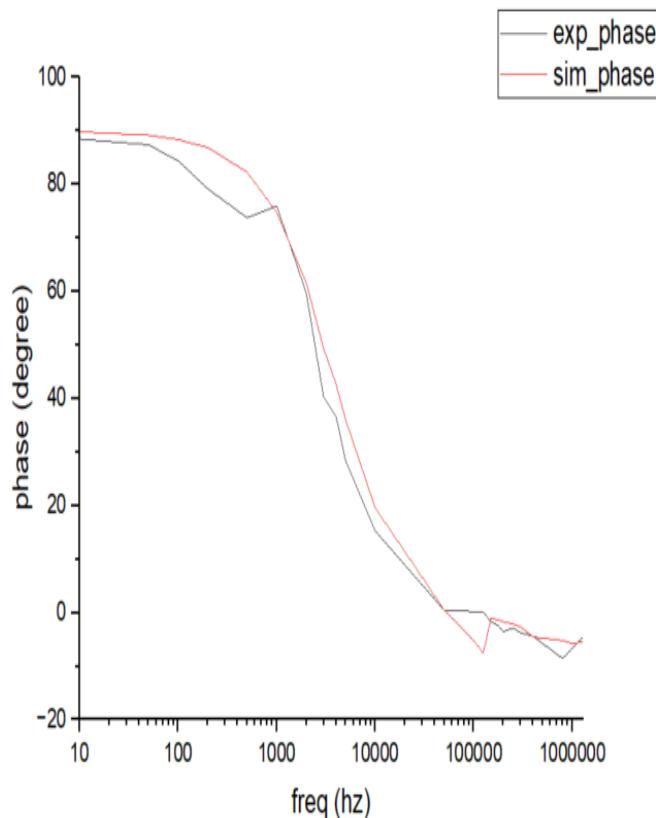
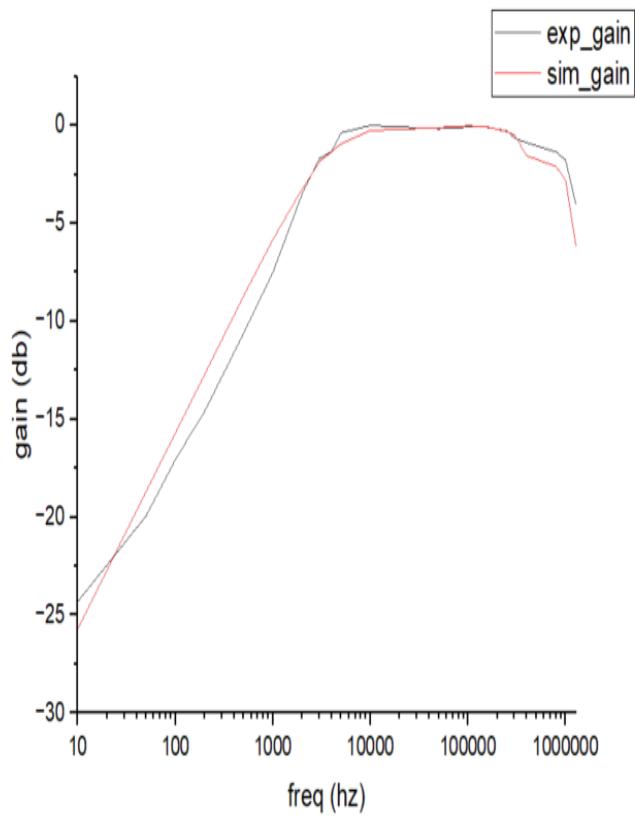
- Experimental gain = -0.137 and LTspice gain = -0.102
- Experimental phase diff. = 0.41 and LTspice = -0.357

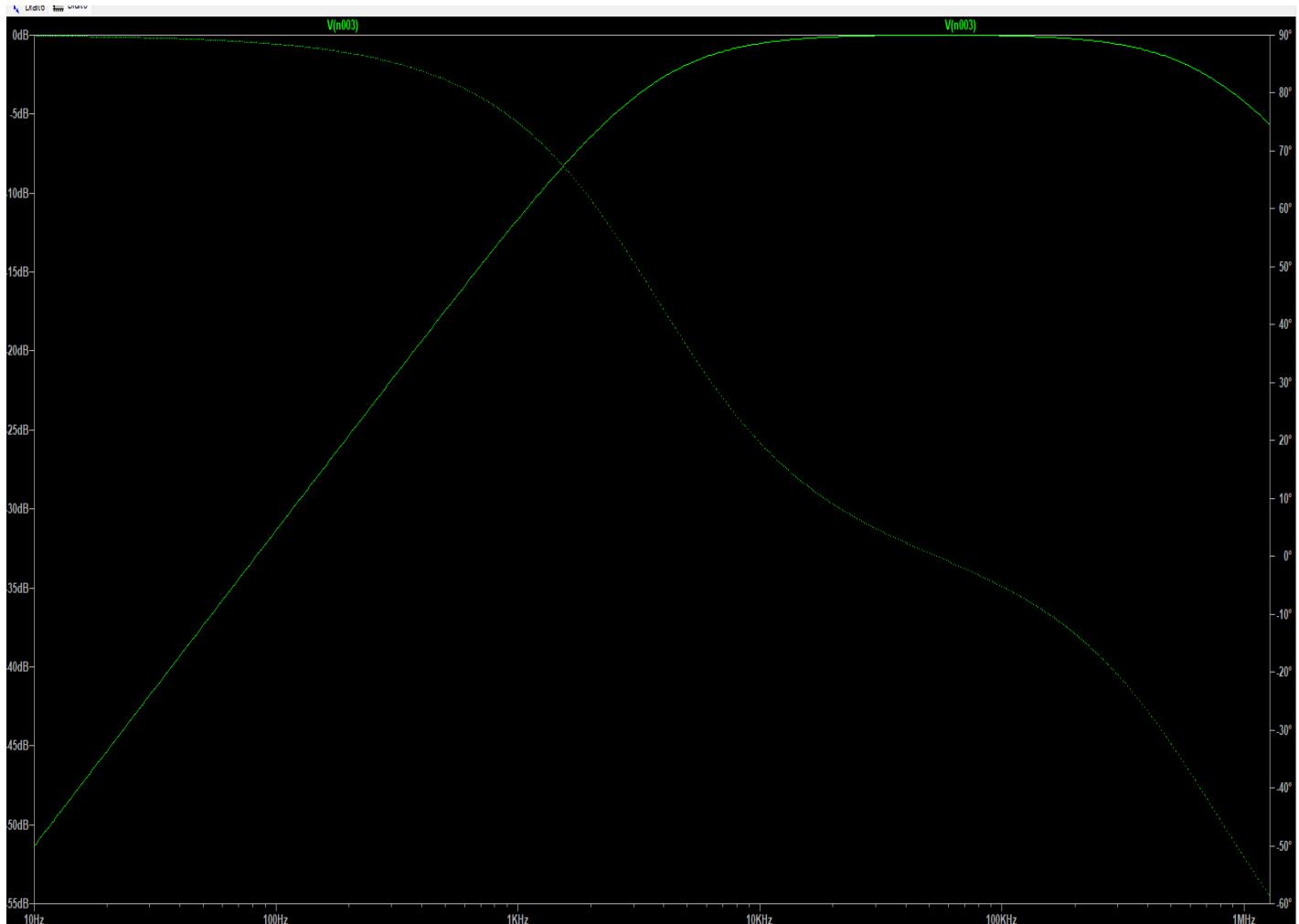


For higher frequency, we observe that

- Experimental gain = -4.03 and LTspice gain = -6.16
- Experimental phase diff. = -4.608 and LTspice = -5.43

| | A(X) | B(Y) | C(Y) | D(Y) | E(Y) |
|-----------|---------|-----------|-----------|----------|-----------|
| Long Name | freq | gain | phase | gain | phase |
| Units | hz | db | degree | db | deg |
| Comments | | exp_gain | exp_phase | sim_gain | sim_phase |
| F(x)= | | | | | |
| 1 | 10 | -24.29947 | 88.5 | -25.7 | 89.8 |
| 2 | 50 | -19.94309 | 87.4 | -18.7 | 89.2 |
| 3 | 100 | -17.0522 | 84.4 | -15.7 | 88.4 |
| 4 | 200 | -14.58356 | 79.2 | -12.7 | 86.9 |
| 5 | 500 | -10.6095 | 73.8 | -8.71 | 82.3 |
| 6 | 1000 | -7.49283 | 76 | -5.82 | 74.8 |
| 7 | 2000 | -3.48203 | 59.71 | -3.22 | 61.6 |
| 8 | 3000 | -1.69524 | 40.24 | -1.86 | 49.3 |
| 9 | 4000 | -1.31615 | 36.67 | -1.34 | 42.7 |
| 10 | 5000 | -0.36866 | 28.34 | -0.928 | 36.1 |
| 11 | 10000 | 0 | 15.28 | -0.259 | 19.6 |
| 12 | 50000 | -0.16545 | 0.46 | -0.116 | 0.537 |
| 13 | 100000 | -0.08273 | 0.22 | -0.018 | -5.22 |
| 14 | 125000 | -0.03341 | 0.17 | -0.0376 | -7.53 |
| 15 | 150000 | -0.08433 | -1.73 | -0.0692 | -1.02 |
| 16 | 175000 | -0.11735 | -2.34 | -0.119 | -1.34 |
| 17 | 200000 | -0.20157 | -3.52 | -0.196 | -1.71 |
| 18 | 250000 | -0.27059 | -2.89 | -0.313 | -2.15 |
| 19 | 300000 | -0.64859 | -3.76 | -0.487 | -2.66 |
| 20 | 400000 | -0.87641 | -4.43 | -1.54 | -4.54 |
| 21 | 800000 | -1.3575 | -8.56 | -2.1 | -5.2 |
| 22 | 1000000 | -1.74961 | -6.61 | -2.78 | -5.82 |
| 23 | 1280000 | -4.03158 | -4.608 | -6.16 | -5.43 |





Experimental Value of R,L and C are $k\Omega$, $0.5nF$ and $4mH$
Tweaked Value of R, L and C are $19.6k\Omega$, $2.2nF$ and $4mH$