

# IEEE-UTD Signals Systems Exam 2 Review

Jerry Luckenbaugh

The University of Texas at Dallas: EE 3302 - Fumagalli/Khoobroo FA2020

10/26/20

# Outline

Fourier Analysis & Sound

Frequency Domain

Magnitude Spectrum

Phase Spectrum

Fourier Transform

FT Tricks

Fourier Series

FS vs FT

Common Fourier Transforms

Examples

## Fourier Analysis is Intuitive

- ▶ Fourier analysis can look scary.
- ▶ The idea becomes simple when you think of something you (probably) know:

## Fourier Analysis is Intuitive

- ▶ Fourier analysis can look scary.
- ▶ The idea becomes simple when you think of something you (probably) know:

**Sound!**

## Fourier Analysis is Intuitive

- ▶ Fourier analysis can look scary.
- ▶ The idea becomes simple when you think of something you (probably) know:

# Sound!

- ▶ Sound is what we experience when our brains interpret the vibration of our eardrum

## Fourier Analysis is Intuitive

- ▶ Fourier analysis can look scary.
- ▶ The idea becomes simple when you think of something you (probably) know:

# Sound!

- ▶ Sound is what we experience when our brains interpret the vibration of our eardrum
- ▶ We have different **qualitative** ways to describe sounds:

| Frequency  | Amplitude |
|------------|-----------|
| High Pitch | Loud      |
| Low Pitch  | Quiet     |

Table 1: Qualitative descriptors of sounds.

## Fourier Analysis is Intuitive

- ▶ Fourier analysis can look scary.
- ▶ The idea becomes simple when you think of something you (probably) know:

# Sound!

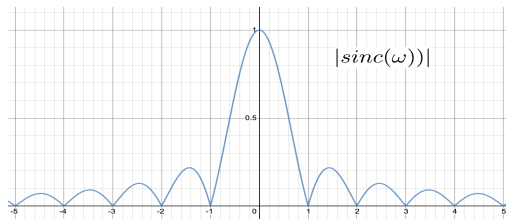
- ▶ Sound is what we experience when our brains interpret the vibration of our eardrum
- ▶ We have different **qualitative** ways to describe sounds:

| Frequency  | Amplitude |
|------------|-----------|
| High Pitch | Loud      |
| Low Pitch  | Quiet     |

Table 1: Qualitative descriptors of sounds.

- ▶ The **magnitude and phase spectrums** lets us talk **quantitatively** about the different **frequencies** of a signal and how they will interact

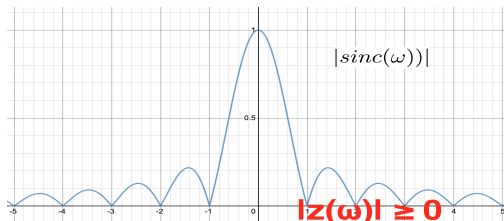
# Magnitude Spectrum



- (I) The magnitude spectrum  $|z(\omega)|$  tells us the **amplitude** of different frequencies in a signal.

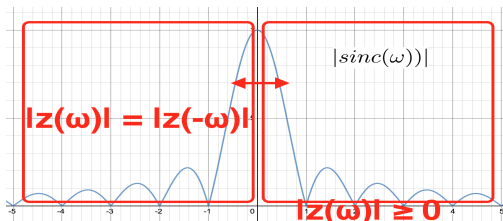


# Magnitude Spectrum



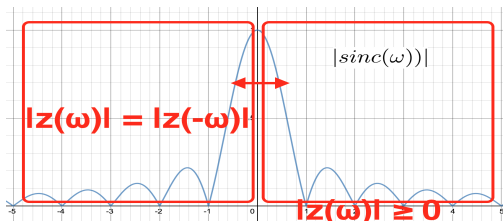
- (I) The magnitude spectrum  $|z(\omega)|$  tells us the **amplitude** of different frequencies in a signal.
- (II) Always **non-negative**.

# Magnitude Spectrum



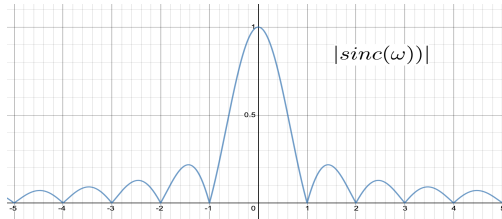
- (I) The magnitude spectrum  $|z(\omega)|$  tells us the **amplitude** of different frequencies in a signal.
- (II) Always **non-negative**.
- (III) For **purely real** signals, **conjugate symmetry** occurs in freq. domain, i.e.  $z(\omega) = z^*(-\omega)$

# Magnitude Spectrum



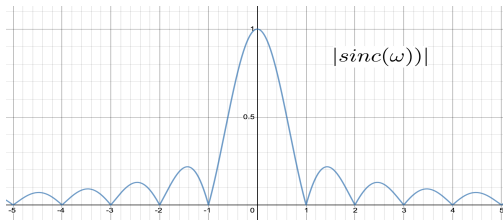
- (I) The magnitude spectrum  $|z(\omega)|$  tells us the **amplitude** of different frequencies in a signal.
- (II) Always **non-negative**.
- (III) For **purely real** signals, **conjugate symmetry** occurs in freq. domain, i.e.  $z(\omega) = z^*(-\omega)$ 
  - Magnitude spectrum will be an even function

# Magnitude Spectrum and Bandwidth



Mag. Spec. is typically the important domain for designing systems

# Magnitude Spectrum and Bandwidth

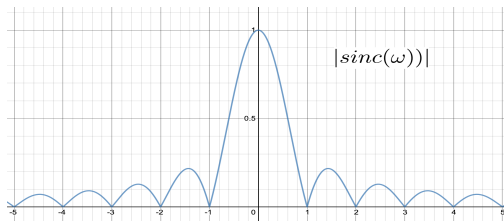


Mag. Spec. is typically the important domain for designing systems  
Recall that LTI systems are characterized completely by their  
impulse response

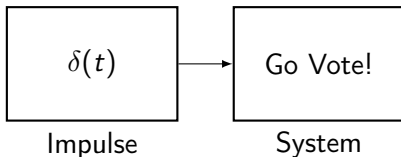
Go Vote!

System

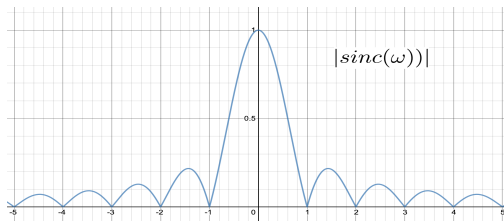
# Magnitude Spectrum and Bandwidth



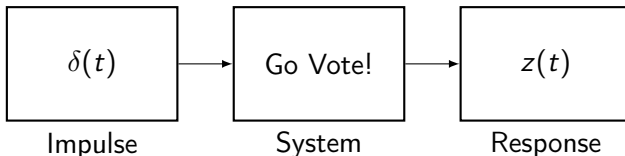
Mag. Spec. is typically the important domain for designing systems  
Recall that LTI systems are characterized completely by their  
impulse response



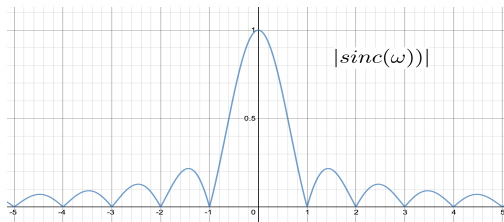
# Magnitude Spectrum and Bandwidth



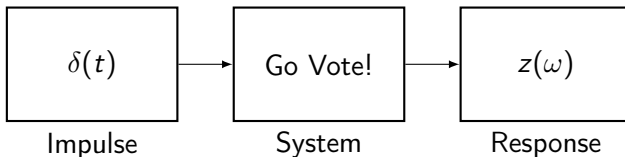
Mag. Spec. is typically the important domain for designing systems  
Recall that LTI systems are characterized completely by their impulse response



# Magnitude Spectrum and Bandwidth



Mag. Spec. is typically the important domain for designing systems  
Recall that LTI systems are characterized completely by their impulse response

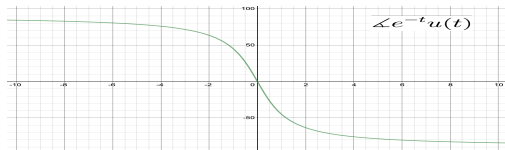


Systems typically specified by their **bandwidth** (BW) or gain

- (I) For baseband signals: BW is the **half** width
- (II) Otherwise: BW is the **full** width

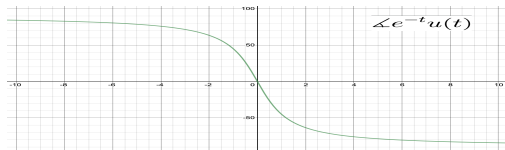


# Phase Spectrum



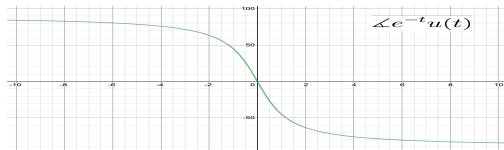
- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another

# Phase Spectrum



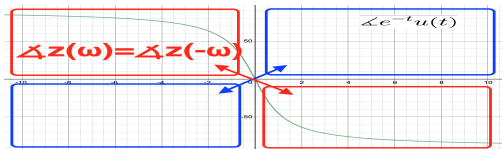
- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another
- (II) **Bounded** range from  $[-180^\circ, 180^\circ)$  or  $[-\pi, \pi)$

# Phase Spectrum



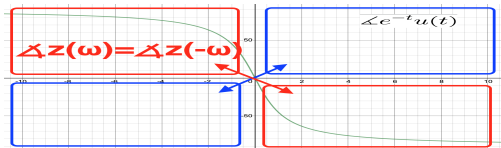
- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another
- (II) **Bounded** range from  $[-180^\circ, 180^\circ)$  or  $[-\pi, \pi)$
- (III) Gives **how each frequency will interact** with another signal/system

# Phase Spectrum



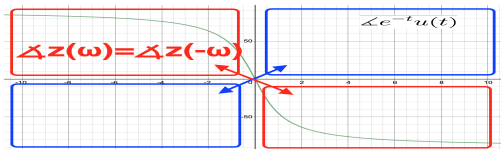
- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another
- (II) **Bounded** range from  $[-180^\circ, 180^\circ)$  or  $[-\pi, \pi)$
- (III) Gives **how each frequency will interact** with another signal/system
- (IV) For **purely real** signals, **conjugate symmetry** occurs in freq. domain, i.e.  $z(\omega) = z^*(-\omega)$

# Phase Spectrum



- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another
- (II) **Bounded** range from  $[-180^\circ, 180^\circ)$  or  $[-\pi, \pi)$
- (III) Gives **how each frequency will interact** with another signal/system
- (IV) For **purely real** signals, **conjugate symmetry** occurs in freq. domain, i.e.  $z(\omega) = z^*(-\omega)$ 
  - Phase spectrum will be an odd function.

# Phase Spectrum



- (I) The phase spectrum  $\angle z(\omega)$  tells us how frequencies are **offset (in time)** relative to one another
- (II) **Bounded** range from  $[-180^\circ, 180^\circ)$  or  $[-\pi, \pi)$
- (III) Gives **how each frequency will interact** with another signal/system
- (IV) For **purely real** signals, **conjugate symmetry** occurs in freq. domain, i.e.  $z(\omega) = z^*(-\omega)$ 
  - ▶ Phase spectrum will be an odd function.
  - ▶ Sometimes omitted if  $f(t)$  is real & odd or real & even

## Misc. Frequency Domain Facts

- ▶  $f(t)$  odd & purely real  $\iff z(\omega)$  odd & purely imaginary
- ▶  $f(t)$  even & purely real  $\iff z(\omega)$  even & purely real

# Fourier Transform

The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(t)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$



# Fourier Transform

The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

# Fourier Transform

The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(t)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:

$f(t)$

Real Function

# Fourier Transform

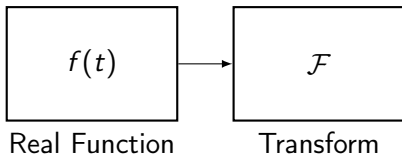
The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:



# Fourier Transform

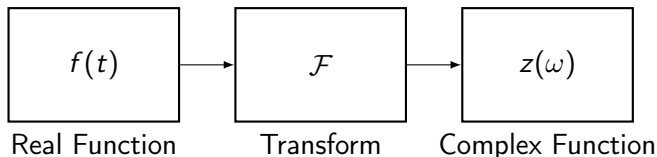
The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:



# Fourier Transform

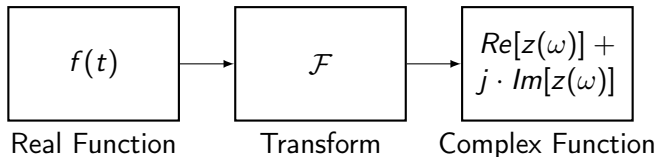
The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:



# Fourier Transform

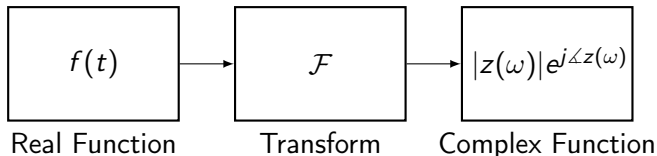
The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:



## Fourier Transform

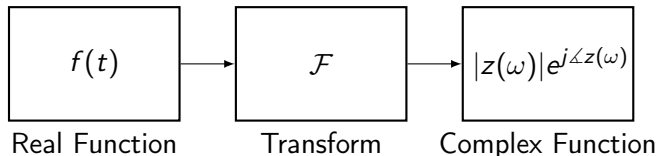
The Fourier Transform is an invertible, integral transform  $\mathcal{F} : \mathbb{C}(t) \mapsto \mathbb{C}(\omega)$ , defined by,

$$\mathcal{F}[f(t)] = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$$

with inverse,

$$\mathcal{F}^{-1}[z(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} z(\omega)e^{j\omega t} d\omega$$

In practice:



### Why Magnitude and Phase Spectrums?

Plotting the magnitude ( $|z(\omega)|$ ) and argument ( $\angle z(\omega)$ ) across all frequencies allow us to easily interpret our transformed data.

# Duality & The Convolution Theorem

Recall the convolution theorem: convolution in TD is multiplication in FD

## Theorem

*Given signals  $x(t), y(t) \in \mathbb{C}(t)$  then,*

$$\mathcal{F}[x(t) * y(t)] = X(\omega)Y(\omega)$$

This is subject to duality

$$\mathcal{F}^{-1}[X(\omega) * Y(\omega)] = x(t)y(t)$$



## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time shifting

$$x(t + \tau) \longrightarrow e^{-j\omega\tau} X(\omega)$$

## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time shifting

$$x(t + \tau) \longrightarrow e^{-j\omega\tau} X(\omega)$$

- ▶ Time scaling

$$x(at) \longrightarrow \frac{1}{a} X\left(\frac{\omega}{a}\right)$$

## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time shifting

$$x(t + \tau) \longrightarrow e^{-j\omega\tau} X(\omega)$$

- ▶ Time scaling

$$x(at) \longrightarrow \frac{1}{a} X\left(\frac{\omega}{a}\right)$$

- ▶ Linearity

$$ax(t) + by(t) \longrightarrow aX(\omega) + bY(\omega)$$

Useful for drawing block diagrams. Makes them the same whether they are in TD or FD

## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time shifting

$$x(t + \tau) \longrightarrow e^{-j\omega\tau} X(\omega)$$

- ▶ Time scaling

$$x(at) \longrightarrow \frac{1}{a} X\left(\frac{\omega}{a}\right)$$

- ▶ Linearity

$$ax(t) + by(t) \longrightarrow aX(\omega) + bY(\omega)$$

Useful for drawing block diagrams. Makes them the same whether they are in TD or FD

- ▶ Parseval's Theorem: Energy is unitary

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)|^2 d\omega$$

## FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

# FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time derivatives

$$\frac{d^n}{dt^n}x(t) \longrightarrow (j\omega)^n X(j\omega)$$

# FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time derivatives

$$\frac{d^n}{dt^n}x(t) \longrightarrow (j\omega)^n X(j\omega)$$

- ▶ Time integrals

$$\int_{-\infty}^t x(\tau) d\tau \longrightarrow \frac{1}{j\omega} X(j\omega) + \pi X(0) \delta(j\omega)$$



# FT Properties

If we have some signal  $x(t)$  with  $X(\omega) = \mathcal{F}[x(t)]$

- ▶ Time derivatives

$$\frac{d^n}{dt^n}x(t) \longrightarrow (j\omega)^n X(j\omega)$$

- ▶ Time integrals

$$\int_{-\infty}^t x(\tau) d\tau \longrightarrow \frac{1}{j\omega} X(j\omega) + \pi X(0) \delta(j\omega)$$

- ▶ Frequency derivatives

$$tx(t) \longrightarrow j \frac{d}{d\omega} X(j\omega)$$

# Fourier Series

Recall Euler's formula

$$\cos(\theta) + j\sin(\theta) = e^{j\theta}$$

## Fourier Series

Recall Euler's formula

$$\cos(\theta) + j\sin(\theta) = e^{j\theta}$$

This gives several equivalent ways of writing Fourier Series

# Fourier Series

Recall Euler's formula

$$\cos(\theta) + j\sin(\theta) = e^{j\theta}$$

This gives several equivalent ways of writing Fourier Series

► Exponential form

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega t}$$

# Fourier Series

Recall Euler's formula

$$\cos(\theta) + j\sin(\theta) = e^{j\theta}$$

This gives several equivalent ways of writing Fourier Series

► Exponential form

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega t}$$

► Trigonometric form

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t)$$

# Fourier Series

Recall Euler's formula

$$\cos(\theta) + j\sin(\theta) = e^{j\theta}$$

This gives several equivalent ways of writing Fourier Series

- ▶ Exponential form

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega t}$$

- ▶ Trigonometric form

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t)$$

- ▶ For purely real signals: Compact form

$$x(t) = C_0 + \sum_{n=1}^{\infty} C_n \cos(n\omega_0 t + \theta_n)$$

## Getting FS coefficients

DON'T do integrals if possible. Try by inspection first. Use these:

$$\cos(\theta) = \frac{1}{2}e^{j\theta} + \frac{1}{2}e^{-j\theta} \quad \& \quad \sin(\theta) = \frac{1}{2j}e^{j\theta} + \frac{1}{2j}e^{-j\theta}$$

I usually turn it into **exponential form first**

## Getting FS coefficients

DON'T do integrals if possible. Try by inspection first. Use these:

$$\cos(\theta) = \frac{1}{2}e^{j\theta} + \frac{1}{2}e^{-j\theta} \quad \& \quad \sin(\theta) = \frac{1}{2j}e^{j\theta} - \frac{1}{2j}e^{-j\theta}$$

I usually turn it into **exponential form first**

Use these relations for changing between forms

► Exponential form

$$c_n = \frac{1}{2}(a_n - jb_n) \quad \& \quad c_n = \frac{1}{2}(a_n + jb_n)$$



## Getting FS coefficients

DON'T do integrals if possible. Try by inspection first. Use these:

$$\cos(\theta) = \frac{1}{2}e^{j\theta} + \frac{1}{2}e^{-j\theta} \quad \& \quad \sin(\theta) = \frac{1}{2j}e^{j\theta} - \frac{1}{2j}e^{-j\theta}$$

I usually turn it into **exponential form first**

Use these relations for changing between forms

► Exponential form

$$c_n = \frac{1}{2}(a_n - jb_n) \quad \& \quad c_n = \frac{1}{2}(a_n + jb_n)$$

► Trigonometric form

$$a_0 = c_0, a_{n \neq 0} = 2\operatorname{Re}(c_n) \quad \& \quad b_n = -2\operatorname{Im}(c_n)$$

## Getting FS coefficients

DON'T do integrals if possible. Try by inspection first. Use these:

$$\cos(\theta) = \frac{1}{2}e^{j\theta} + \frac{1}{2}e^{-j\theta} \quad \& \quad \sin(\theta) = \frac{1}{2j}e^{j\theta} - \frac{1}{2j}e^{-j\theta}$$

I usually turn it into **exponential form first**

Use these relations for changing between forms

- ▶ Exponential form

$$c_n = \frac{1}{2}(a_n - jb_n) \quad \& \quad c_n = \frac{1}{2}(a_n + jb_n)$$

- ▶ Trigonometric form

$$a_0 = c_0, a_{n \neq 0} = 2\operatorname{Re}(c_n) \quad \& \quad b_n = -2\operatorname{Im}(c_n)$$

- ▶ For purely real signals: Compact form

$$C_0 = c_0, C_{n \neq 0} = 2|c_n| \quad \& \quad \theta_n = \angle c_n$$

## Getting FS coefficients

DON'T do integrals if possible. Try by inspection first. Use these:

$$\cos(\theta) = \frac{1}{2}e^{j\theta} + \frac{1}{2}e^{-j\theta} \quad \& \quad \sin(\theta) = \frac{1}{2j}e^{j\theta} - \frac{1}{2j}e^{-j\theta}$$

I usually turn it into **exponential form first**

Use these relations for changing between forms

- ▶ Exponential form

$$c_n = \frac{1}{2}(a_n - jb_n) \quad \& \quad c_n = \frac{1}{2}(a_n + jb_n)$$

- ▶ Trigonometric form

$$a_0 = c_0, a_{n \neq 0} = 2\operatorname{Re}(c_n) \quad \& \quad b_n = -2\operatorname{Im}(c_n)$$

- ▶ For purely real signals: Compact form

$$C_0 = c_0, C_{n \neq 0} = 2|c_n| \quad \& \quad \theta_n = \angle c_n$$

## Getting FS coefficients

If necessary, here are the integrals for the coefficients.

- ▶ Important if you are given a periodic function with discontinuities. e.g. triangle, square, sawtooth waves
- ▶ These can be useful

$$\cos^2(t) = \frac{1 + \cos(2t)}{2} \quad \& \quad \sin^2(t) = \frac{1 - \cos(2t)}{2}$$

- ▶ Exponential form

$$c_n = \frac{1}{T} \int_0^T x(t) e^{-jn\omega_0 t} dt$$

- ▶ Trigonometric form

$$a_n = \begin{cases} \frac{1}{T} \int_0^T x(t) dt; & n = 0 \\ \frac{2}{T} \int_0^T x(t) \cos(n\omega_0 t) dt & \end{cases} \quad \& \quad b_n = \frac{2}{T} \int_0^T x(t) \sin(n\omega_0 t) dt$$

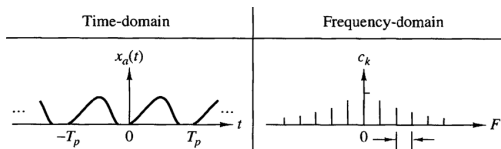
# Fourier Series vs Fourier Transform

Fourier Series and Fourier Transform capture the same idea

# Fourier Series vs Fourier Transform

Fourier Series and Fourier Transform capture the same idea  
FS is for periodic signals (i.e.  $x(t) = x(t + \tau)$ )

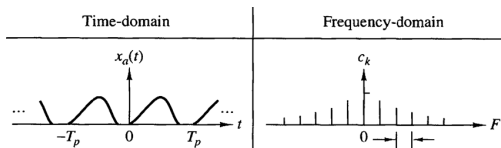
- ▶ Will be discrete in the frequency domain



# Fourier Series vs Fourier Transform

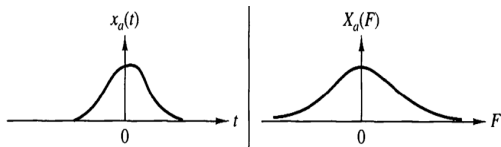
Fourier Series and Fourier Transform capture the same idea  
FS is for periodic signals (i.e.  $x(t) = x(t + \tau)$ )

- ▶ Will be discrete in the frequency domain



FT is for aperiodic signals

- ▶ Will be continuous in the frequency domain



# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$



# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$

$$(II) \quad \cos(\omega_0 t) \rightleftharpoons [\delta(\omega - \omega_0) + [\delta(\omega + \omega_0)]]$$

# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$

$$(II) \quad \cos(\omega_0 t) \rightleftharpoons [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$(III) \quad \sin(\omega_0 t) \rightleftharpoons \frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$

$$(II) \quad \cos(\omega_0 t) \rightleftharpoons [\delta(\omega - \omega_0) + [\delta(\omega + \omega_0)]]$$

$$(III) \quad \sin(\omega_0 t) \rightleftharpoons \frac{\pi}{j} [\delta(\omega - \omega_0) + [\delta(\omega + \omega_0)]]$$

$$(IV) \quad \Pi(t) = \begin{cases} 1 & |t| < \frac{1}{2} \\ 0 & \text{otherwise} \end{cases} \rightleftharpoons \text{sinc}\left(\frac{\omega}{2\pi}\right)$$

# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$

$$(II) \quad \cos(\omega_0 t) \rightleftharpoons [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$(III) \quad \sin(\omega_0 t) \rightleftharpoons \frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

$$(IV) \quad \Pi(t) = \begin{cases} 1 & |t| < \frac{1}{2} \\ 0 & \text{otherwise} \end{cases} \rightleftharpoons \text{sinc}\left(\frac{\omega}{2\pi}\right)$$

$$(V) \quad \delta(t) \rightleftharpoons 1(\omega)$$

# Common Fourier Transforms

Several Fourier Transforms / Fourier Series are useful for solving problems:

$$(I) \quad e^{j\omega_0 t} \rightleftharpoons \delta(\omega - \omega_0)$$

$$(II) \quad \cos(\omega_0 t) \rightleftharpoons [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$(III) \quad \sin(\omega_0 t) \rightleftharpoons \frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

$$(IV) \quad \Pi(t) = \begin{cases} 1 & |t| < \frac{1}{2} \\ 0 & \text{otherwise} \end{cases} \rightleftharpoons \text{sinc}\left(\frac{\omega}{2\pi}\right)$$

$$(V) \quad \delta(t) \rightleftharpoons 1(\omega)$$

$$(VI) \quad \sum_{i=-\infty}^{\infty} \delta(t - n\tau) \rightleftharpoons \frac{\sqrt{2\pi}}{\tau} \sum_{i=-\infty}^{\infty} \delta\left(\frac{\omega}{2\pi} - \frac{n}{\tau}\right)$$

## Examples