

2011 401

$$1. (a) \lim_{x \rightarrow 0} \frac{8^x - 4^x}{2 - \sqrt{4-x}} = \lim_{x \rightarrow 0} \frac{(8^x - 4^x)(2 + \sqrt{4-x})}{(2 - \sqrt{4-x})(2 + \sqrt{4-x})} = \lim_{x \rightarrow 0} \left(\frac{8^x - 4^x}{x} \right) \times 4$$

$$= 4 \lim_{x \rightarrow 0} \left(\frac{8^x - 1}{x} - \frac{4^x - 1}{x} \right) = 4 (\ln 8 - \ln 4) = \underline{4 \ln 2}.$$

$$(b) \text{ Let } x - \frac{\pi}{2} = t, \text{ then } \lim_{x \rightarrow \frac{\pi}{2}} \frac{\tan 2x}{x - \frac{\pi}{2}} = \lim_{t \rightarrow 0} \frac{\tan(2t + \pi)}{t} = \lim_{t \rightarrow 0} \frac{\tan 2t}{t} = \underline{2}.$$

$$2. \ln y = x^n \ln x, \quad \frac{y'}{y} = nx^{n-1} \ln x + x^{n-1} = x^{n-1} (n \ln x + 1)$$

$$y' = x^{2n} \cdot x^{n-1} (n \ln x + 1) = \underline{x^{3n-1} (n \ln x + 1)}.$$

$$3. \text{ Let } \cos x = t, \quad -\sin x dx = dt.$$

$$\int \sin^3 x \cos^5 x dx = -\int (1-t^2)t^2 dt = \int t^4 - t^2 dt = \frac{1}{5}t^5 - \frac{1}{3}t^3 + C.$$

$$\therefore \int \sin^3 x \cos^5 x dx = \underline{\frac{1}{5} \cos^5 x - \frac{1}{3} \cos^3 x + C}.$$

$$4. s^2 Y - s y(0) - y'(0) + 2sY - 2y(0) + Y = \frac{1}{s+1} = s^2 Y + 2s - 3 + 2s + 4 + Y$$

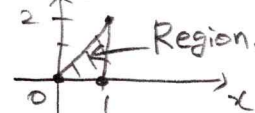
$$Y(s^2 + 2s + 1) = \frac{1}{s+1} - (2s+1) = \frac{1-2s^2-3s-1}{s+1} \Rightarrow Y(s) = \frac{-2s^2-3s}{(s+1)^3} = \frac{A}{s+1} + \frac{B}{(s+1)^2} + \frac{C}{(s+1)^3}$$

$$As^2 + 2As + A + Bs + B + C = As^2 + (2A+B)s + A+B+C = -2s^2 - 3s.$$

$$A = -2, B = 1, C = 1$$

$$Y(s) = \frac{-2}{s+1} + \frac{1}{(s+1)^2} + \frac{1}{(s+1)^3}$$

$$\therefore y(t) = \left(-2e^{-t} + te^{-t} + \frac{t^2}{2}e^{-t} \right) u(t)$$

5. $L = xy$, $M = x^2y^3$, 

$$\oint_C \{xy dx + x^2y^3 dy\} = \iint_R \{2xy^3 - x\} dx dy = \int_0^1 \int_0^{2x} 2xy^3 - x dy dx$$

$$= \int_0^1 \left[\frac{x}{2} y^4 - xy \right]_0^{2x} dx = \int_0^1 8x^5 - 2x^2 dx = \frac{8}{6} - \frac{2}{3} = \underline{\underline{\frac{2}{3}}}$$

6. (a) $\begin{bmatrix} x_1 & x_2 & x_3 \end{bmatrix} \begin{bmatrix} a & c & 1 \\ b & d & 7 \\ 0 & -2 & e \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 \\ 1 & 7 & 0 \\ 0 & 0 & 7 \end{bmatrix}$

$$\begin{bmatrix} 1 & 2 & 1 \\ 2 & 3 & 1 \\ 1 & 1 & -1 \end{bmatrix} \begin{bmatrix} a & c & 1 \\ b & d & 7 \\ 0 & -2 & e \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 \\ 1 & 7 & 0 \\ 0 & 0 & 7 \end{bmatrix} \Rightarrow \text{inverse 공해서 구해도 되지만} \\ \text{노가다가 더 빠름...}$$

(a) $\begin{pmatrix} a+2b=1 \\ 2a+3b=1 \\ a+b=0 \end{pmatrix} \quad \underline{\underline{a=-1}} \quad \underline{\underline{b=1}} \quad \begin{pmatrix} c+2d-2=1 \\ c+d+2=0 \end{pmatrix} \quad \begin{pmatrix} c+2d=3 \\ c+d=-2 \end{pmatrix} \quad \underline{\underline{c=-7}} \quad \underline{\underline{d=5}}$

$-1+e=0 \quad \underline{\underline{e=1}}$

(b) $Y = HX$, H : system matrix,

in (a), $XP=U \quad \therefore YP = HXP = HU = Y_0$

$$Y_0 = \begin{bmatrix} 1 & 3 & 2 \\ 5 & 7 & 2 \\ 4 & 4 & -4 \end{bmatrix} \begin{bmatrix} -1 & -7 & 1 \\ 1 & 5 & -1 \\ 0 & -2 & 1 \end{bmatrix}$$

$$= \begin{bmatrix} 2 & 4 & 0 \\ 2 & -4 & 0 \\ 0 & 0 & -4 \end{bmatrix}$$

(c) Gauss-Jordan method

$$\left[\begin{array}{ccc|ccc} 1 & 1 & 0 & 1 & 0 & 0 \\ 1 & 7 & 0 & 0 & 1 & 0 \\ 0 & 0 & -1 & 0 & 0 & 1 \end{array} \right] \rightarrow \left[\begin{array}{ccc|ccc} 1 & 1 & 0 & 1 & 0 & 0 \\ 0 & -2 & 0 & -1 & 1 & 0 \\ 0 & 0 & -1 & 0 & 0 & 1 \end{array} \right] \rightarrow \left[\begin{array}{ccc|ccc} 1 & 0 & 0 & \frac{1}{2} & \frac{1}{2} & 0 \\ 0 & -2 & 0 & -1 & 1 & 0 \\ 0 & 0 & -1 & 0 & 0 & 1 \end{array} \right]$$

$$\therefore U^{-1} = \begin{bmatrix} \frac{1}{2} & \frac{1}{2} & 0 \\ \frac{1}{2} & -\frac{1}{2} & 0 \\ 0 & 0 & -1 \end{bmatrix}$$

$$\left[I \mid V^{-1} \right] = \left[\begin{array}{ccc|ccc} 1 & 0 & 0 & \frac{1}{2} & \frac{1}{2} & 0 \\ 0 & 1 & 0 & \frac{1}{2} & -\frac{1}{2} & 0 \\ 0 & 0 & 1 & 0 & 0 & -1 \end{array} \right]$$

(d) $Hx = \lambda x \Rightarrow$ eigenvector 를 구하는데, 먼저 H 를 알아야 함

그러니까 $YP = \begin{bmatrix} 2 & 4 & 0 \\ 2 & -4 & 0 \\ 0 & 0 & -4 \end{bmatrix} = HV$ 이므로

$$H = \begin{bmatrix} 2 & 4 & 0 \\ 2 & -4 & 0 \\ 0 & 0 & -4 \end{bmatrix} \begin{bmatrix} \frac{1}{2} & \frac{1}{2} & 0 \\ \frac{1}{2} & -\frac{1}{2} & 0 \\ 0 & 0 & -1 \end{bmatrix} = \begin{bmatrix} 3 & -1 & 0 \\ -1 & 3 & 0 \\ 0 & 0 & 4 \end{bmatrix}$$

$$\det(H - \lambda I) = \begin{vmatrix} 3-\lambda & -1 & 0 \\ -1 & 3-\lambda & 0 \\ 0 & 0 & 4-\lambda \end{vmatrix} = (4-\lambda)(\lambda^2 - 6\lambda + 9 - 1) = 0$$

$$\Rightarrow (\lambda - 4)(\lambda - 2)(\lambda - 4) = 0$$

① $\lambda = 2$ case

$$\begin{bmatrix} 1 & -1 & 0 \\ -1 & 1 & 0 \\ 0 & 0 & 2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} \quad x_1 = x_2, \quad x_3 = 0$$

$$\therefore \begin{bmatrix} 1 \\ 1 \\ 0 \end{bmatrix} = \underline{x}_1$$

② $\lambda = 4$ case

$$\begin{bmatrix} -1 & -1 & 0 \\ -1 & -1 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} \quad \underline{x}_2 = \begin{bmatrix} 1 \\ -1 \\ 0 \end{bmatrix} \quad \underline{x}_3 = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$\therefore S = \{ \underline{x} \in \mathbb{R}^3 \mid \underline{x} = k_1 \underline{x}_2 + k_2 \underline{x}_3, k_1, k_2 \in \mathbb{R} \}$$

(e) $\underline{y} = H\underline{x}$

$$\underline{y} = \begin{bmatrix} 3 & -1 & 0 \\ -1 & 3 & 0 \\ 0 & 0 & 4 \end{bmatrix} \underline{x}$$

7.

(a) sum of two periodic signal is not a periodic signal generally.

(예제) Let $x_1(t) = \sin t$, $x_2(t) = \sin 2\pi t$.

$x_1(t)$ 의 주기는 $2\pi = T_1$, $x_2(t)$ 의 주기 $T_2 = 1$.

$y(t) = x_1(t) + x_2(t)$ 일때 $y(t)$ 의 주기 T_y 가 존재하기 위해서는

$k_1 T_1 = k_2 T_2$ 인 $k_1 \in \mathbb{N}$, $k_2 \in \mathbb{N}$ 가 존재해야 함.

But, $2\pi k_1 = k_2$, $2\pi = \frac{k_2}{k_1}$ 는 rational number가 아니므로

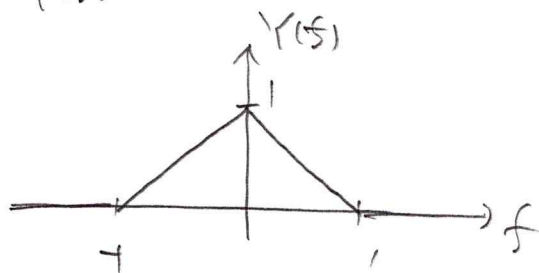
$k_1 \in \mathbb{N}$ and $k_2 \in \mathbb{N}$ 존재하지 않으므로 $y(t)$ is not a periodic signal.

$$\begin{aligned}
 (b) \quad h(t) * \left\{ \sum_{k=-K}^K a_k e^{j \frac{2\pi k t}{T}} \right\} &= \int_{-\infty}^{\infty} h(\tau) \sum_{k=-K}^K a_k e^{j \frac{2\pi k (t-\tau)}{T}} d\tau \\
 &= \sum_{k=-K}^K a_k e^{j \frac{2\pi k t}{T}} \int_{-\infty}^{\infty} h(\tau) e^{-j \frac{2\pi k \tau}{T}} d\tau = \sum_{k=-K}^K a_k e^{j \frac{2\pi k t}{T}} H\left(\frac{k}{T}\right) \\
 &= \sum_{k=-K}^K a_k H\left(\frac{k}{T}\right) e^{j \frac{2\pi k t}{T}} = \sum_{k=-K}^K b_k e^{j \frac{2\pi k t}{T}}.
 \end{aligned}$$

\therefore periodic function with Fourier coefficient $b_k = a_k H\left(\frac{k}{T}\right)$

(c) $x(t) = \text{sinc } t$.

$$Y(f) = X(f) * X(f) = \text{rect}(f) * \text{rect}(f) = \Lambda(f)$$



$$Y(f) = \begin{cases} 1 - |f|, & \text{for } -1 \leq f \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

2011 통신

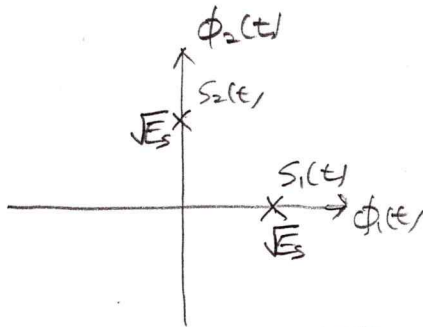
1. 2014 통신 참고!

2. a) $s_1(t) = A \cos 2\pi f_1 t$, $\int_0^T A^2 \cos^2(2\pi f_1 t) dt = \frac{A^2}{2} T = E_s$, $A = \sqrt{\frac{2E_s}{T}}$

$$s_1(t) = \sqrt{\frac{2E_s}{T}} \cos 2\pi f_1 t, \quad s_2(t) = \sqrt{\frac{2E_s}{T}} \cos 2\pi f_2 t \quad (f_1 \neq f_2)$$
$$(0 \leq t < T) \quad (0 \leq t < T)$$

b) $\phi_1(t) = \sqrt{\frac{2}{T}} \cos 2\pi f_1 t \quad (0 \leq t < T)$

$$\phi_2(t) = \sqrt{\frac{2}{T}} \cos 2\pi f_2 t \quad (0 \leq t < T)$$



c) $P_e = \frac{1}{2} \times Q\left(\frac{\sqrt{2E_s}}{\sqrt{2N_0}}\right) + \frac{1}{2} \times Q\left(\frac{\sqrt{2E_s}}{\sqrt{2N_0}}\right) = \underline{Q\left(\sqrt{\frac{E_s}{N_0}}\right)}$

d) $\{\phi_i(t)\}_{i=1}^M$ are all orthonormal basis, so

minimum distance between two symbols are the same as $\sqrt{2E_s}$.

$$\therefore \underline{P_e \leq (M-1) Q\left(\sqrt{\frac{E_s}{N_0}}\right)}$$

2011 제1회

제어공학

1. (a)

$$u(t)\cos\theta - Mg\sin\theta = ML\ddot{\theta}$$

$$\ddot{\theta} = \frac{1}{ML}u(t)\cos\theta - \frac{g}{L}\sin\theta$$

(b) $x_1(t) = \theta(t)$

$x_2(t) = \dot{\theta}(t)$

$$\begin{cases} \dot{x}_1(t) = x_2(t) \\ \dot{x}_2(t) = -\frac{g}{L}\sin x_1(t) + \frac{u(t)}{ML}\cos x_1(t) \end{cases}$$

(평형상태일 경우)

$$0 = \frac{1}{ML}u(t)\cos\theta - \frac{g}{L}\sin\theta$$

$$\rightarrow u(t) = Mg\tan\theta$$

(c) Linear 한 시스템이 아니기 때문에 input-output 간의 linear relationship을

정의할 수 없으므로 transfer function can't be found.

But, by using linearization method, we can approximately find the transfer function.

$$\Delta \dot{x}_2(t) = \frac{\Delta u(t)}{ML}\cos\theta_0 - \frac{u_0}{ML}\Delta x_1(t)\sin\theta_0 - \frac{g}{L}\Delta x_1(t)\cos\theta_0$$

$$= \frac{1}{ML}\cos\theta_0 \cdot \Delta u(t) - \left(\frac{1}{ML}\sin\theta_0 \cdot Mg\tan\theta_0 + \frac{g}{L}\cos\theta_0 \right) \Delta x_1(t)$$

$$= \frac{1}{ML}\cos\theta_0 \cdot \Delta u(t) - \frac{g}{L}\sec\theta_0 \cdot \Delta x_1(t)$$

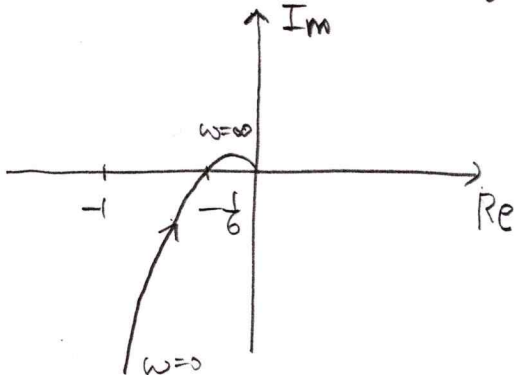
$$\Delta \dot{x}(t) = \begin{bmatrix} 0 & 1 \\ -\frac{g}{L}\sec\theta_0 & 0 \end{bmatrix} \Delta x(t) + \begin{bmatrix} 0 \\ \frac{1}{ML}\cos\theta_0 \end{bmatrix} \Delta u(t)$$

2.

$$a) L(s)|_{T=0} = \frac{1}{s(s+1)(s+2)} \Rightarrow L(j\omega) = \frac{1}{j\omega(j\omega+1)(j\omega+2)}$$

$$\textcircled{1} \omega=0 \Rightarrow L(j\omega) = \infty \angle -90^\circ$$

$$\textcircled{2} \omega=\infty \Rightarrow L(j\omega) = 0 \angle -270^\circ$$



③ Gain Margin

$$L(j\omega) = \frac{1}{(j\omega-j\omega^2)(j\omega+2)} = \frac{1}{-\omega^2+2j\omega-j\omega^3-2\omega^2} = \frac{1}{-3\omega^2+j(2\omega-\omega^3)}$$

$$\text{Im}\{L(j\omega)\} = 0 \text{ 이 되는 } \omega_p = \sqrt{2}.$$

$$\therefore \text{이 때 } L(j\omega) = -\frac{1}{6}.$$

$$\text{Gain margin} : -20 \log_{10}\left(\frac{5}{6}\right) = 20(\log_{10} 6 - \log_{10} 5) = \underline{\underline{2.33 \text{ dB}}}$$

$$b) L(j\omega) = \frac{e^{-j\omega T}}{j\omega(j\omega+1)(j\omega+2)}$$

$$\textcircled{1} |L(j\omega)| = 1 \text{ 이 되는 } \omega_g \text{ 찾는 것.$$

$$|L(j\omega)| = \frac{1}{\sqrt{9\omega^4 + (2\omega - \omega^3)^2}} = 1 \Rightarrow \omega_g = 0.447 \dots$$

$$\textcircled{2} \omega_g = 0.447 \text{ 일 때, phase } \rightarrow \text{찾는 것.}$$

$$\begin{aligned} \angle L(j\omega_g) &= -90^\circ - \tan^{-1}(\omega) - \tan^{-1}\left(\frac{\omega}{2}\right) - \omega T \\ &= -126.7^\circ - 0.447T \end{aligned}$$

$$\textcircled{3} \text{ 이 때 phase } > -180^\circ \text{ 일 때 marginally stable 이다.}$$

$$0.447T = 53.3^\circ, \quad 53.3^\circ = 53.3 \times \frac{\pi}{180} = 0.3\pi \text{ rad.}$$

$$\therefore \underline{\underline{T = 0.67 \text{ s}}}$$

