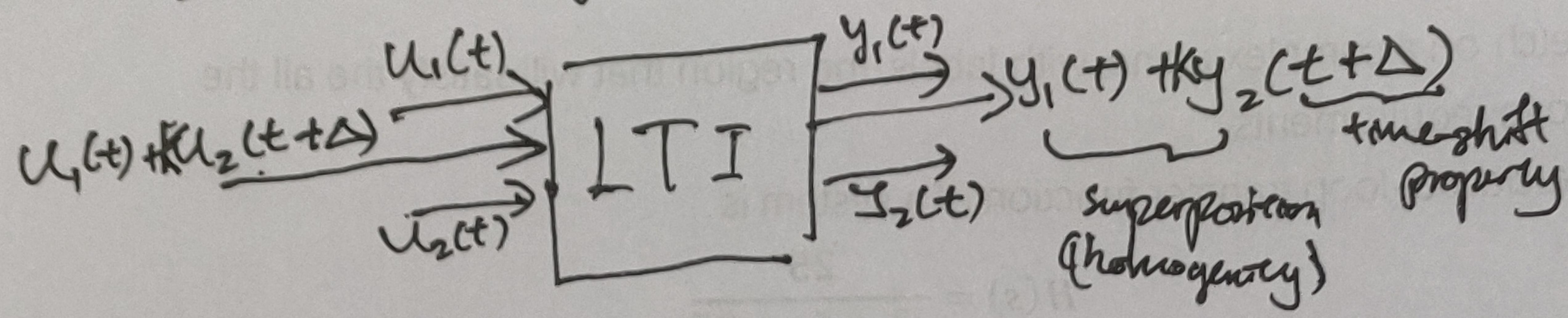


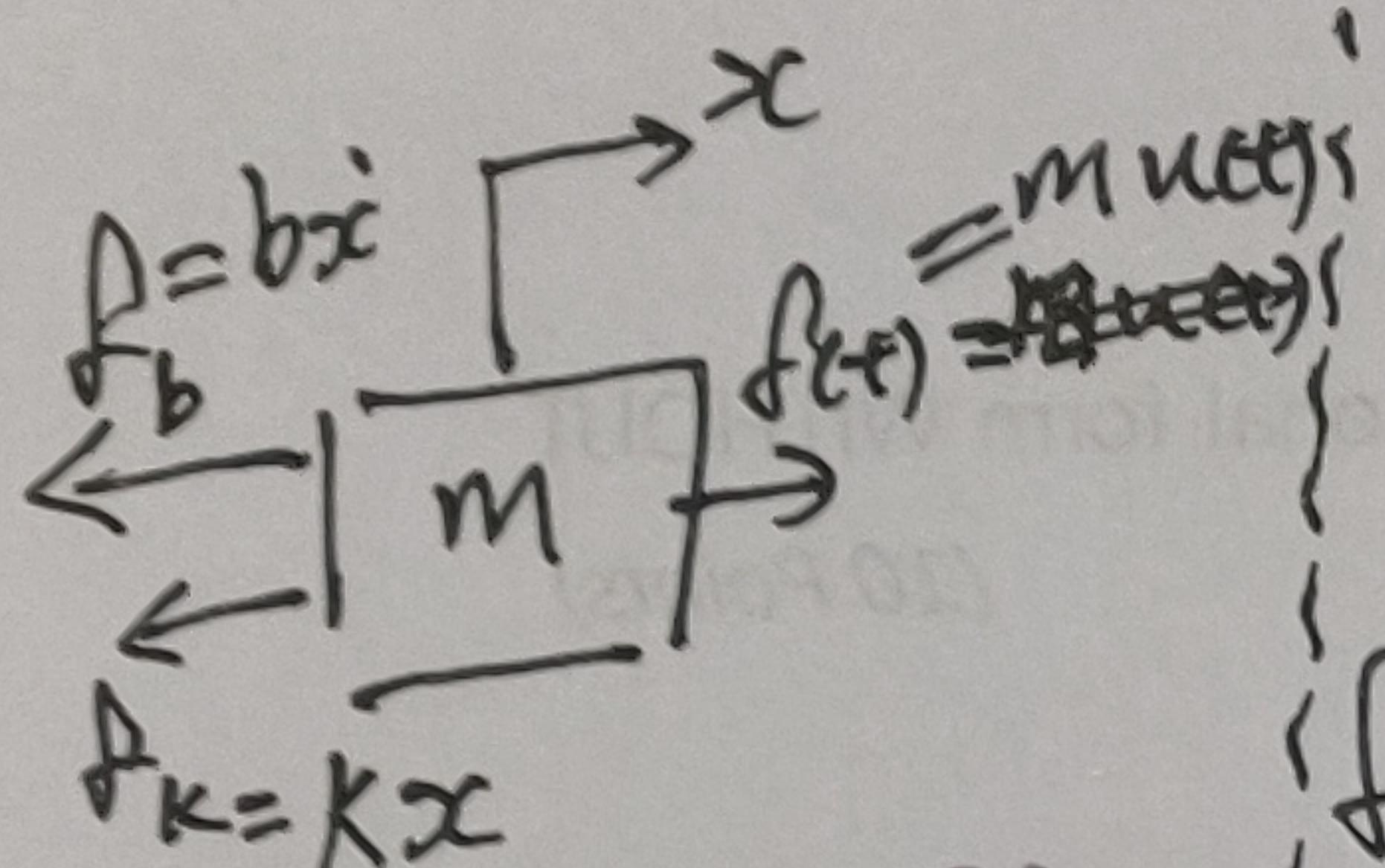
Q1

a) Linear-time-invariant systems satisfy superposition principle and time-shift property



①

b)



$$m\ddot{x} = f = f_b - f_k$$

$$m\ddot{x} = f(t) - b\dot{x} - kx$$

$$f(t) = m\ddot{x} + b\dot{x} + kx$$

$$\ddot{x} + \frac{b}{m}\dot{x} + \frac{k}{m}x = \frac{f(t)}{m}$$

$$c) \text{ Let } \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} = \begin{pmatrix} x \\ \dot{x} \end{pmatrix} \Rightarrow \begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -5x_2 - 4x_1 + u(t) \end{aligned}$$

$$\text{State-eq: } \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -4 & -5 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t) \quad ③$$

$$\text{Output-eq: } y = [1 \ 0] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad ①$$

$$h) x(t) = ? \text{ ie } x(t) = I(t) * h(t)$$

$$X(s) = \frac{1}{s} H(s) \quad ①$$

$$= \frac{1}{s} \frac{1}{s^2 + 5s + 4}$$

$$= \frac{1}{s} \frac{1}{3} \left(\frac{1}{s+1} - \frac{1}{s+4} \right) \quad ①$$

$$= \frac{1}{3} \left[\frac{1}{(s+1)(s+4)} - \frac{1}{(s+4)(s+1)} \right]$$

$$x(t) = \frac{1}{3} \left[\frac{1}{1-0} (e^0 - e^{-t}) - \frac{1}{4-0} (e^0 - e^{-4t}) \right]$$

$$x(t) = \frac{1}{3} \left[(1 - e^{-t}) - \frac{1}{4} (1 - e^{-4t}) \right] \quad ①$$

$$= \frac{1}{3} \left[\frac{3}{4} - e^{-t} + \frac{1}{4} e^{-4t} \right]$$

$$x(t) = \frac{1}{4} - \frac{1}{3} e^{-t} + \frac{1}{12} e^{-4t} \quad ①$$

$$d) \text{ poles at } s^2 + 5s + 4 = 0$$

$$(s+4)(s+1) = 0$$

$$\therefore p_{1,2} = -4 \text{ or } -1 \quad ②$$

$$f) \text{ since } H(s) = \frac{1}{(s+4)(s+1)} = \frac{1}{s+4} \cdot \frac{1}{s+1}$$

$$\Rightarrow (i) 4, (ii) 1 \quad ②$$

$$g) \left[U\left(\frac{1}{s+1}\right) \cancel{+} U\left(\frac{1}{s+4}\right) \right] \times \frac{1}{3} = X$$

$$\begin{aligned} \frac{X}{U} &= \frac{1}{3} \left(\frac{1}{s+1} - \frac{1}{s+4} \right) = \frac{1}{3} \left(\frac{s+4-s-1}{(s+1)(s+4)} \right) \\ &= \frac{1}{(s+4)(s+1)} = \frac{1}{s^2 + 5s + 4} \text{ (verified)} \end{aligned}$$

Q2

a) Advantage: Reject disturbance, Track reference

Disadvantage: Introduce errors, may amplify noise, costly to implement (2)

b) To reduce steady-state error (1)

c) D-term or derivative term can be noise prone and adversely affect noise suppression

$$d); H_p = \frac{1}{s^2 + 3s + 2} \quad (1), \quad H_c = K + \frac{K_I}{s}, \quad H_{CL} = \frac{H_p H_c}{1 + H_p H_c} = \frac{H_c}{\frac{1}{H_p} + H_c}$$

$$\therefore H_{CL} = \frac{K + \frac{K_I}{s}}{(s^2 + 3s + 2) + (K + \frac{K_I}{s})} = \frac{Ks + K_I}{s^3 + 3s^2 + 2s + Ks + K_I}$$

$$= \frac{Ks + K_I}{s^3 + 3s^2 + (K+2)s + K_I} \quad (3)$$

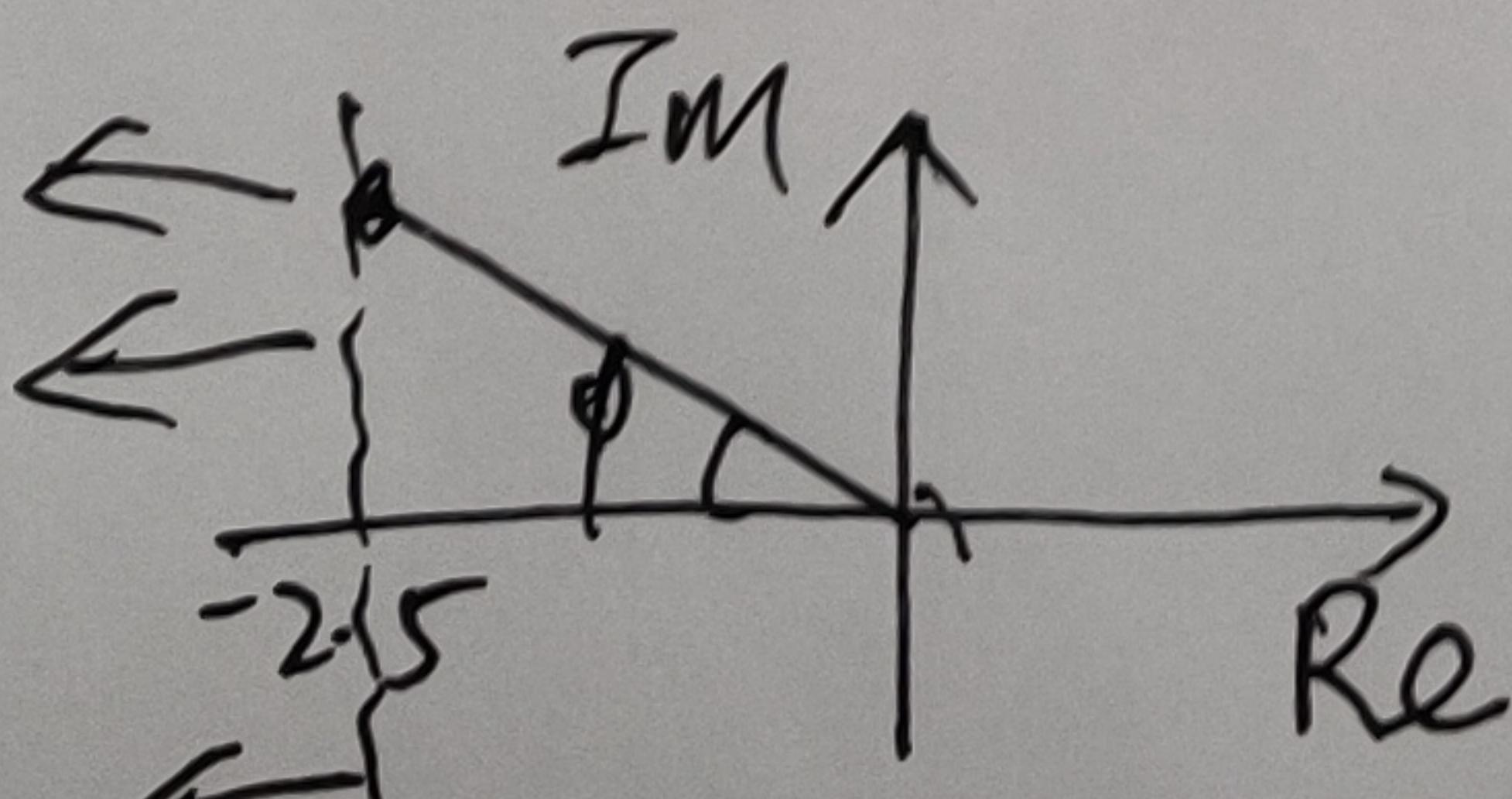
ii) Characteristic eq^y: $s^3 + 3s^2 + (K+2)s + K_I = 0$

$$\begin{array}{c} s^3 \\ \hline s^2 \left(\begin{array}{c} 1 \\ 3 \end{array} \right) \left(\begin{array}{c} K+2 \\ -K_I \end{array} \right) \quad (1) \\ \hline s^1 \left(\begin{array}{c} -[K_I - 3(K+2)] \\ 0 \end{array} \right) \quad (2) \\ \hline s^0 \left(\begin{array}{c} -3 \\ K_I \end{array} \right) \quad (2) \end{array} \quad \left. \begin{array}{l} K_I > 0 \\ -K_I + K + 2 > 0 \\ K + 2 > K_I \end{array} \right\} \text{optional}$$

$$2 e) i) \varphi = 45^\circ (= \cos^{-1} \frac{1}{\sqrt{2}})$$

$$(ii) t_{ss5\%} = \frac{3}{2} \Rightarrow \delta \geq \frac{3}{1.2}$$

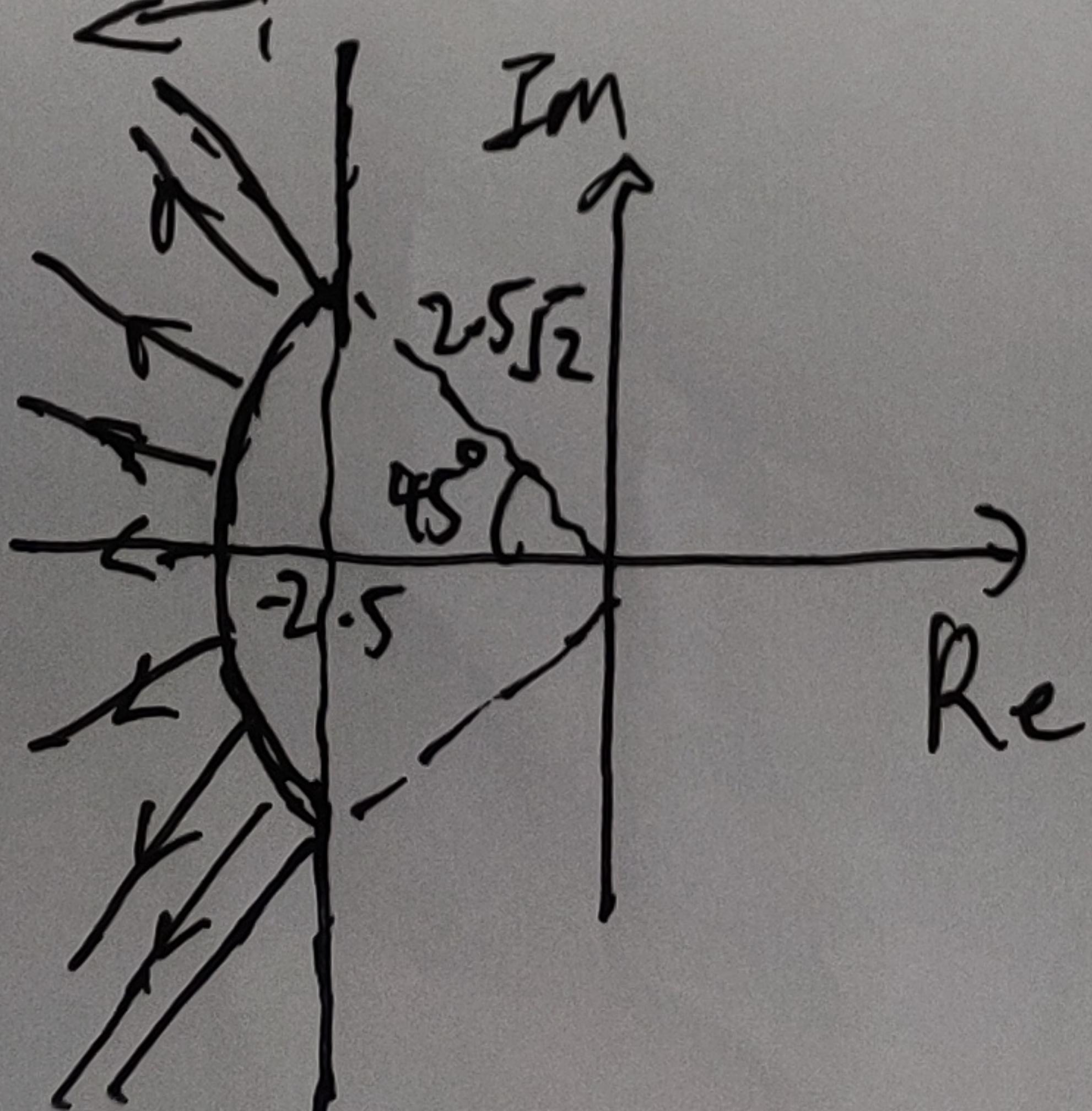
$$\therefore \delta = 2.5$$



$$(iii) \omega_n \bar{z} = 3$$

$$\omega_n > 2.5\sqrt{2}$$

iv)



$$(v) H(s) = \frac{25}{s^2 + 6s + 25} = \frac{25}{(s + 3 + 4j)(s + 3 - 4j)}$$

P_1 and P_2 are out of the satisfactory region.

$$\begin{aligned} P_{1,2} &= \frac{-6 \pm \sqrt{36 - 100}}{2} \\ &= -3 \pm \sqrt{6(-1)} \\ P_{1,2} &= -3 \pm 4j \end{aligned}$$