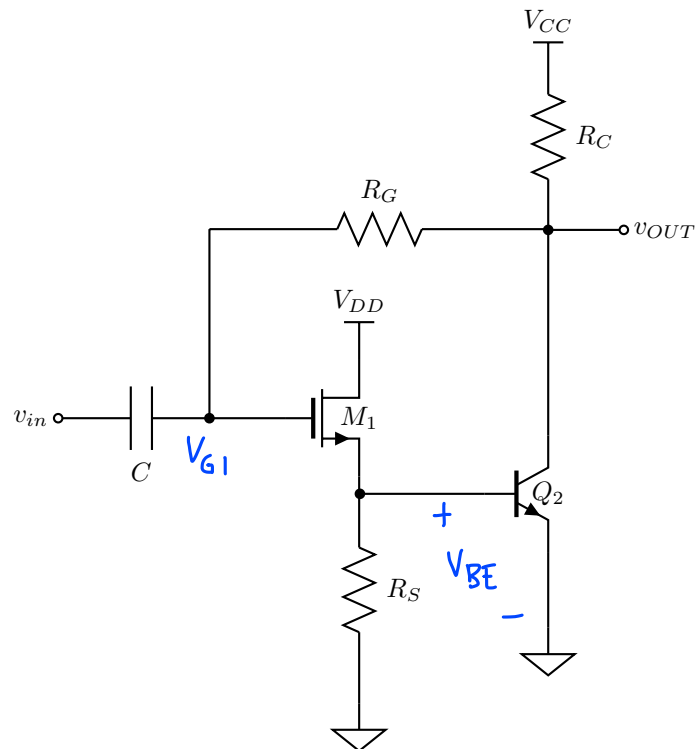


## Part I:

(20 points) In the BiCMOS amplifier circuit below,  $V_{DD} = V_{CC} = 5\text{ V}$ ,  $R_G = 10\text{ M}\Omega$ , and  $C \rightarrow \infty$ . The BJT has  $V_{BE} = 0.7\text{ V}$ ,  $V_{CE,SAT} = 0.2\text{ V}$ ,  $V_A \rightarrow \infty$ , and  $\beta \rightarrow \infty$  while the NMOS has  $V_{TH} = 1\text{ V}$ ,  $k = 2 \frac{\text{mA}}{\text{V}^2}$ , and  $\lambda \rightarrow 0$ .



- Given  $R_G = 10\text{ M}\Omega$ ,  $R_S = 6.8\text{ k}\Omega$ , and  $R_C = 3\text{ k}\Omega$ , determine the quiescent currents and voltages for both transistors. (8 points)

$$\text{since } V_{BE} = 0.7\text{ V}, \quad I_{O1} = \frac{0.7}{R_S} = \underline{102.94\text{ }\mu\text{A}}$$

$$V_{DS1} = V_{DD} - 0.7 = \underline{4.3\text{ V}}$$

$$V_{GS1} = V_{TH} + \sqrt{\frac{I_{O1}}{k}} = 1.23\text{ V}$$

$$\text{since } I_G = 0; \quad V_{OUT} = V_{GS1} + 0.7 = 1.93\text{ V}$$

$$I_{C2} = \frac{V_{CC} - V_{OUT}}{R_C} = \underline{1.02\text{ mA}}$$

$$V_{CE2} = V_{OUT} = \underline{1.93\text{ V}}$$

$I_{D1} = 102.94\text{ }\mu\text{A}$	$V_{DS1} = 4.3\text{ V}$	$I_{C2} = 1.02\text{ mA}$	$V_{CE2} = 1.93\text{ V}$
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2. If you are to design this circuit for **maximum** output swing, what should the value of  $R_S$  be? (4 points)

$$V_{out,min} = V_{GS1,min} + V_{BE} = V_{TH} + V_{BE} \quad \left( \begin{array}{l} \text{since } V_{out} \text{ is} \\ \text{same as } V_{G1} \end{array} \right)$$

$$V_{out,max} = V_{CC} = 5V$$

$$V_{out,DC} = \frac{V_{out,min} + V_{out,max}}{2} = 3.35V$$

$$V_{G1} = V_{out} = V_{GS1} + 0.7 = 3.35$$

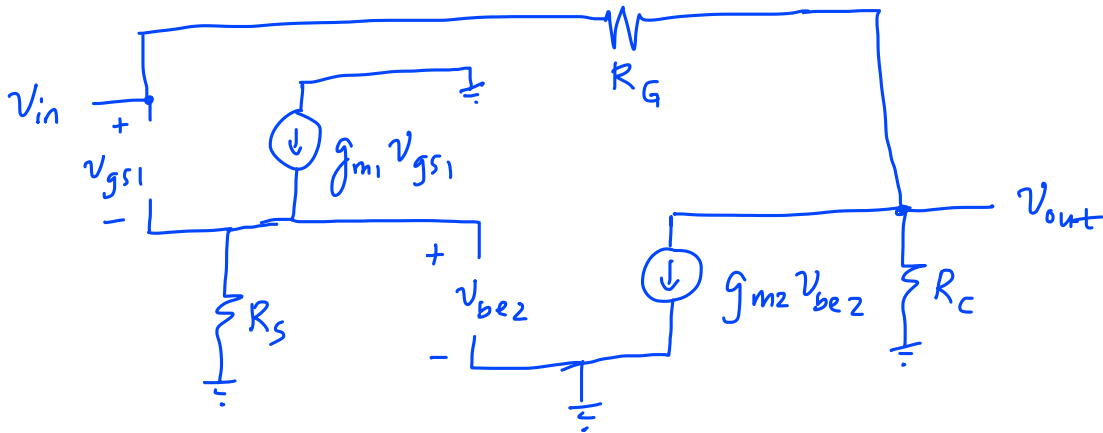
$$V_{GS1} = 2.65$$

$$I_{D1} = K(V_{GS1} - V_{TH})^2 = 5.45 \text{ mA} = \frac{0.7}{R_S}$$

$$R_S = 128 \Omega$$

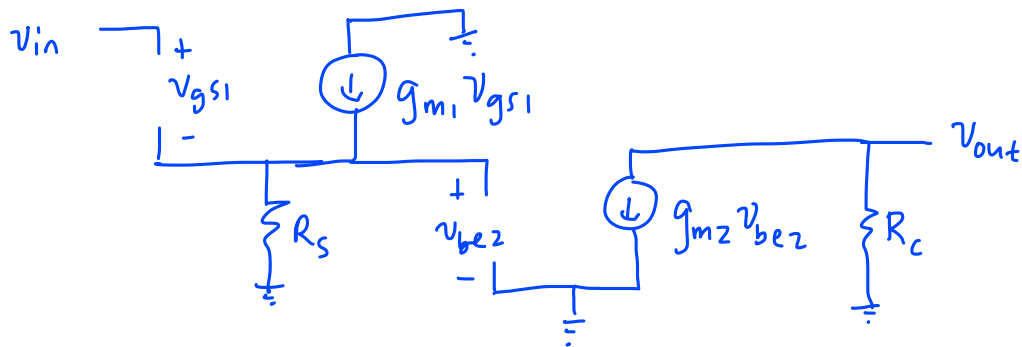
$$R_S = 128 \Omega$$

3. Draw the small-signal equivalent circuit of the given BiCMOS amplifier. Label completely. (3 points)



4. Suppose the transistors were biased so that  $g_{m1} = 1 \text{ mS}$  and  $g_{m2} = 50 \text{ mS}$ . Determine the voltage gain of the BiCMOS amplifier given  $R_S = 5.6 \text{ k}\Omega$ ,  $R_C = 2.9 \text{ k}\Omega$ , and  $R_G \rightarrow \infty$ . (5 points)

Since  $R_G \rightarrow \infty$ , small-signal equivalent circuit simplifies to:



KCL at  $v_{out}$  :  $v_{out} = -g_{m2} v_{be2} R_C$

KCL at  $v_{be2}$  :  $v_{be2} = g_{m1} v_{gs1} R_S$

Note that  $v_{gs1} = v_{in} - v_{be2}$

$$v_{be2} = g_{m1} (v_{in} - v_{be2}) R_S$$

$$v_{be2} = \frac{g_{m1} R_S}{1 + g_{m1} R_S} \cdot v_{in}$$

$$v_{out} = -g_{m2} \left( \frac{g_{m1} R_S}{1 + g_{m1} R_S} \right) R_C v_{in}$$

$$\frac{v_{out}}{v_{in}} = -(50 \text{ mS}) \left( \frac{(1 \text{ mS})(5600)}{1 + (1 \text{ mS})(5600)} \right) (2900)$$

$$\frac{v_{out}}{v_{in}} = -123$$

$$= \underline{-123}$$