



# **EEE 51: Second Semester 2017 - 2018**

## **Lecture 7**

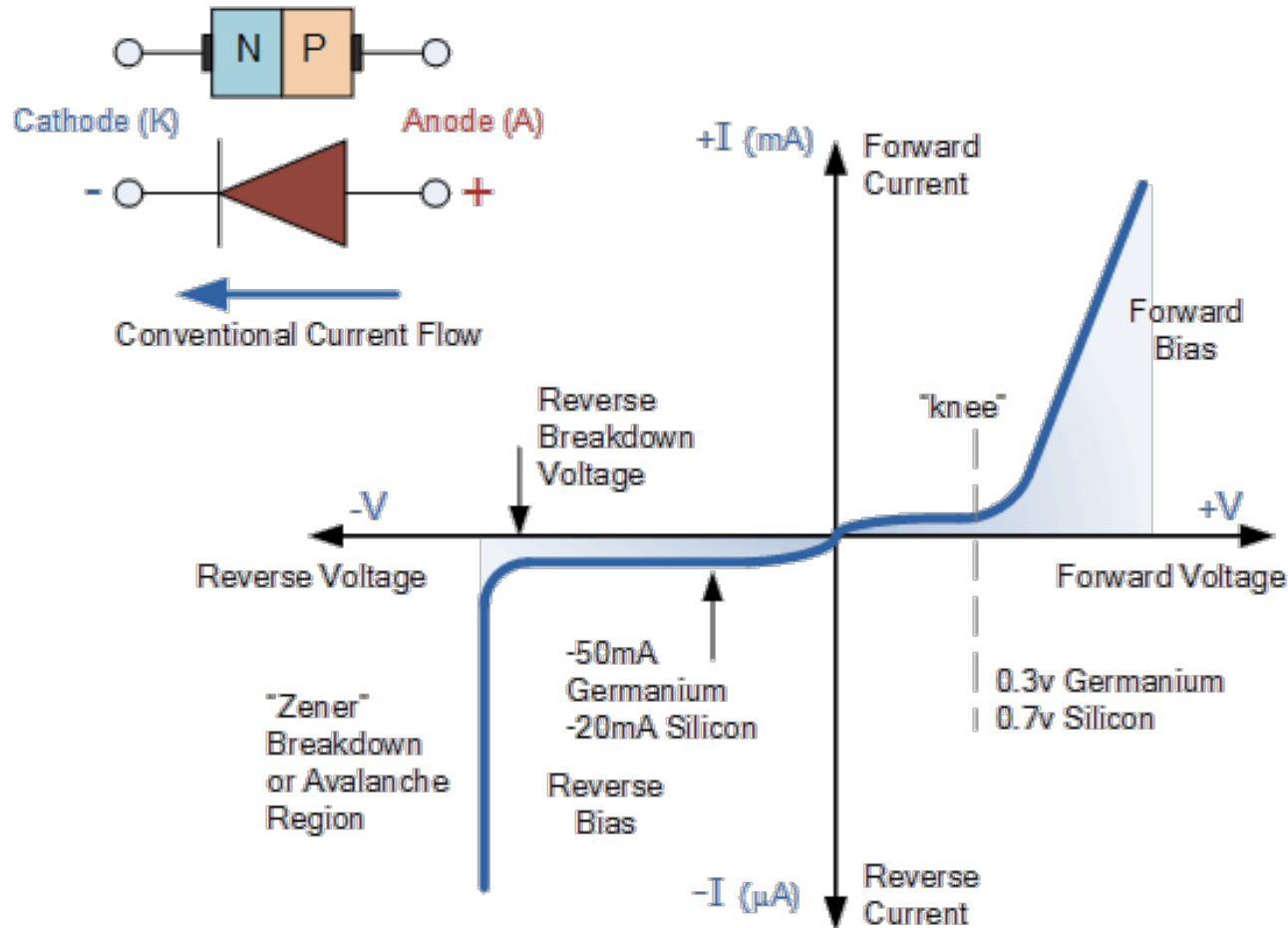
# Current Mirrors

# Today

- Current Mirrors

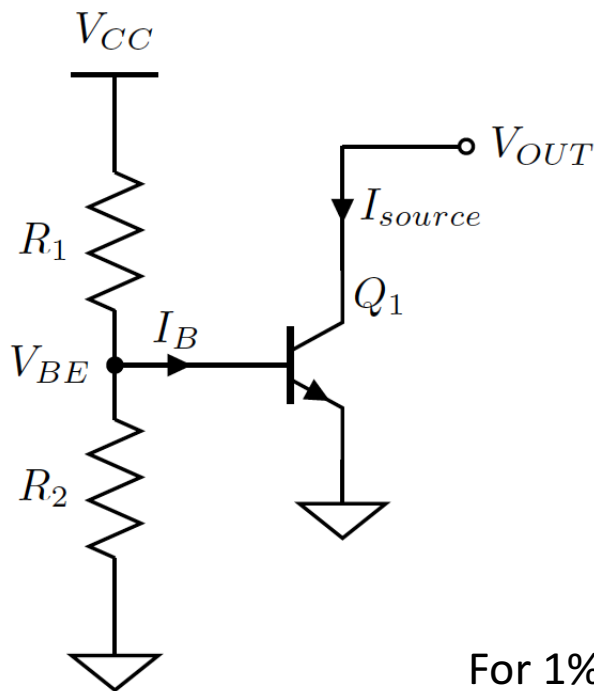


# Practical Diode Characteristics



# A Simple BJT Current Source

- Provide  $V_{BE}$  using a voltage divider



Example: what if we need 1mA?

$$V_{BE} = V_T \ln \frac{I_C}{I_S} = V_T \ln \frac{1\text{mA}}{2 \times 10^{-16} \text{A}} = 0.7603\text{V}$$

Small changes in  $V_{BE}$ :

$$V_{BE} = 0.7703\text{V} \rightarrow I_C = 1.5\text{mA}$$

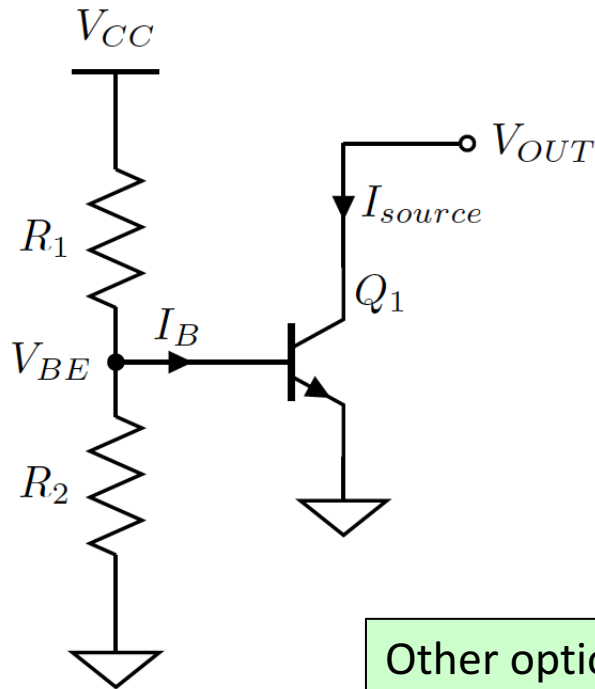
Also, for  $V_{CC} = 5\text{V}$  and  $\beta = 200$ :

$$R_1 = 10\text{k}\Omega \rightarrow R_2 = 1.8146\text{k}\Omega$$

$$\text{For 1\% error in } I_C, \Delta V_{BE} = 260\mu\text{V} \rightarrow \beta = 200 \pm 9$$

# A Simple BJT Current Source

- Provide  $V_{BE}$  using a voltage divider



## Main issues:

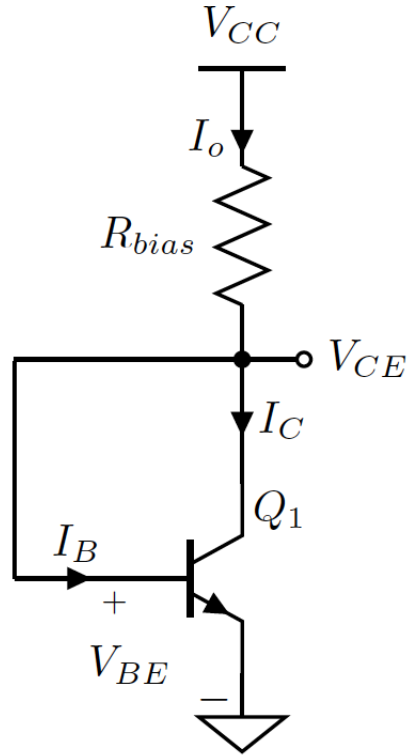
1. Need very precise resistor values
2. Need good  $\beta$  control

Expensive  
(if possible at all)

Other options in generating  $V_{BE}$ ?



# The Diode-Connected Transistor (1)



Use  $R_{bias}$  to apply a current  $I_o$  into the BJT:

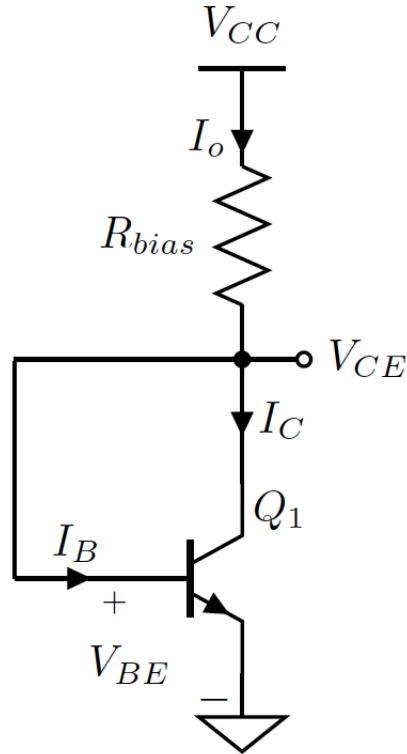
$$I_o = I_C + I_B = I_C \cdot \left(1 + \frac{1}{\beta}\right)$$

Solving for  $V_{CE} = V_{BE}$  (assuming  $V_A \rightarrow \infty$ ) :

$$V_{BE} = V_T \ln\left(\frac{I_C}{I_S}\right) = V_T \ln\left(\frac{I_o}{I_S} \cdot \frac{\beta}{\beta+1}\right)$$

How do we generate  $I_o$ ?

# The Diode-Connected Transistor (2)



KVL at base loop:

$$V_{CC} - I_o R_{bias} - V_{BE} = 0$$

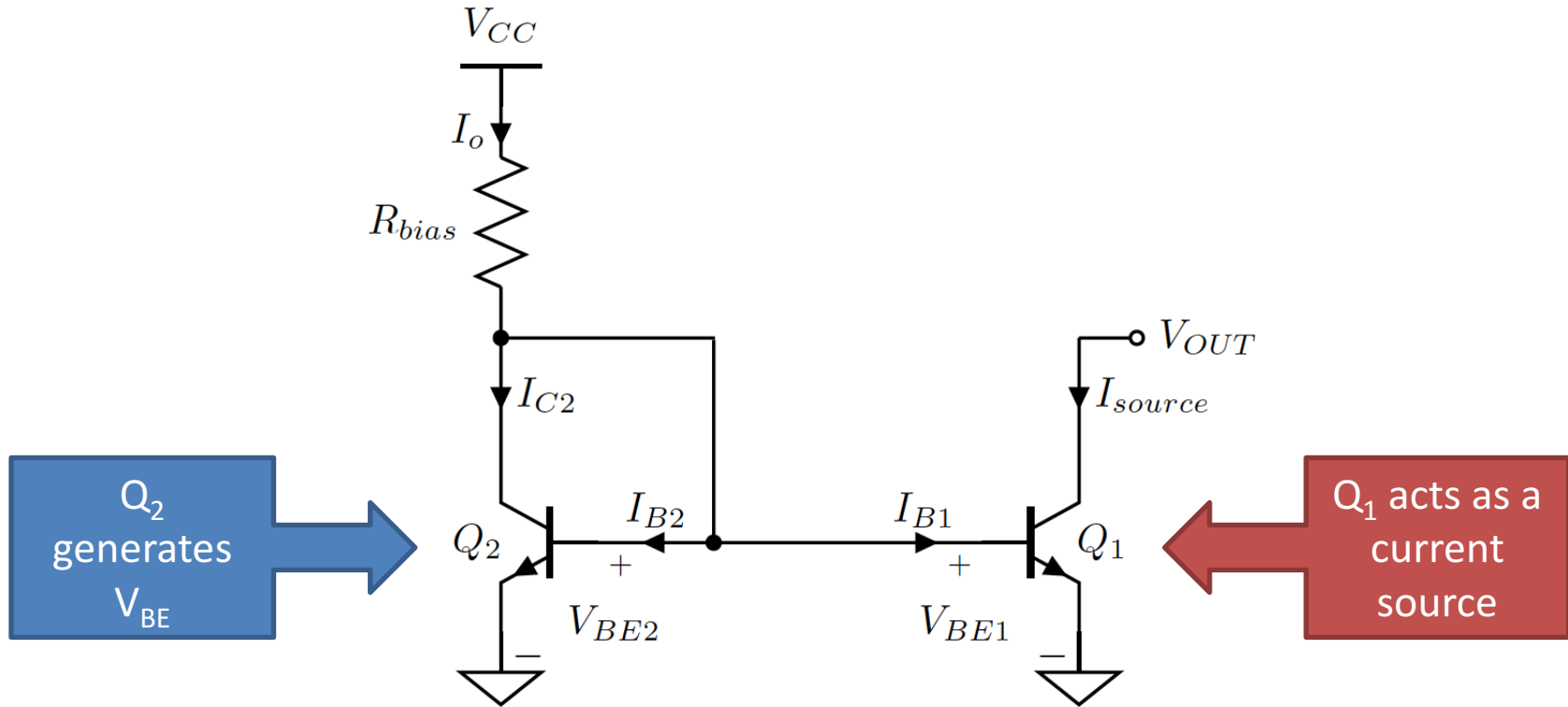
Thus,

$$R_{bias} = \frac{V_{CC} - V_{BE}}{I_o} = \frac{V_{CC} - V_{BE}}{I_S \cdot e^{\frac{V_{BE}}{V_T}} \cdot \left(1 + \frac{1}{\beta}\right)}$$

We can pick a resistor to generate  $V_{BE}$

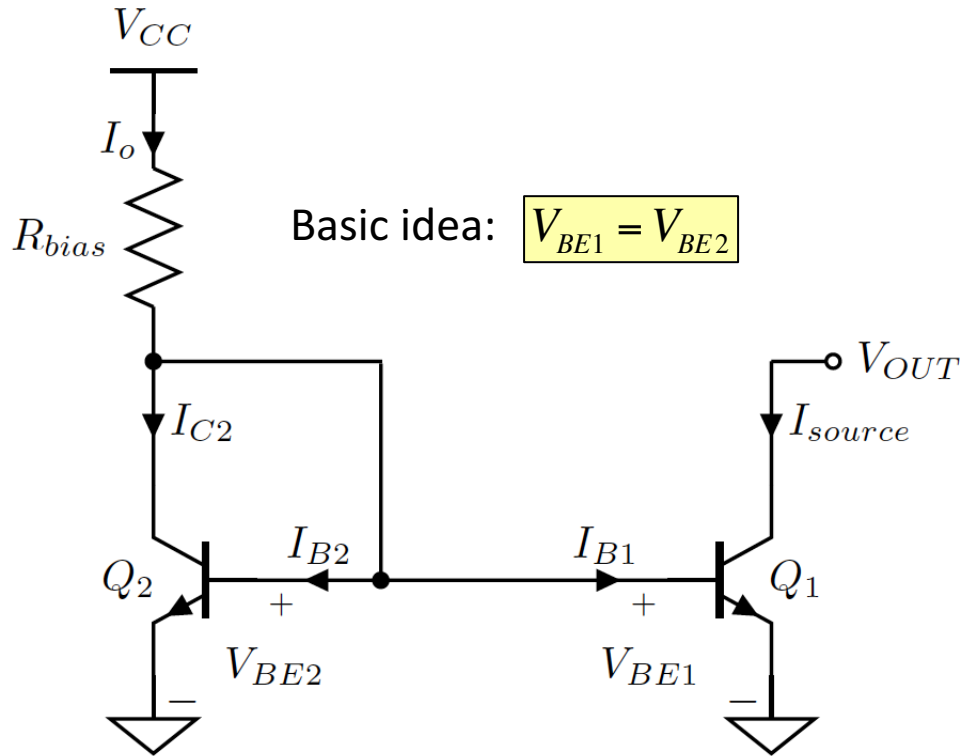


# A Simple BJT Current Mirror (1)





# A Simple BJT Current Mirror (2)

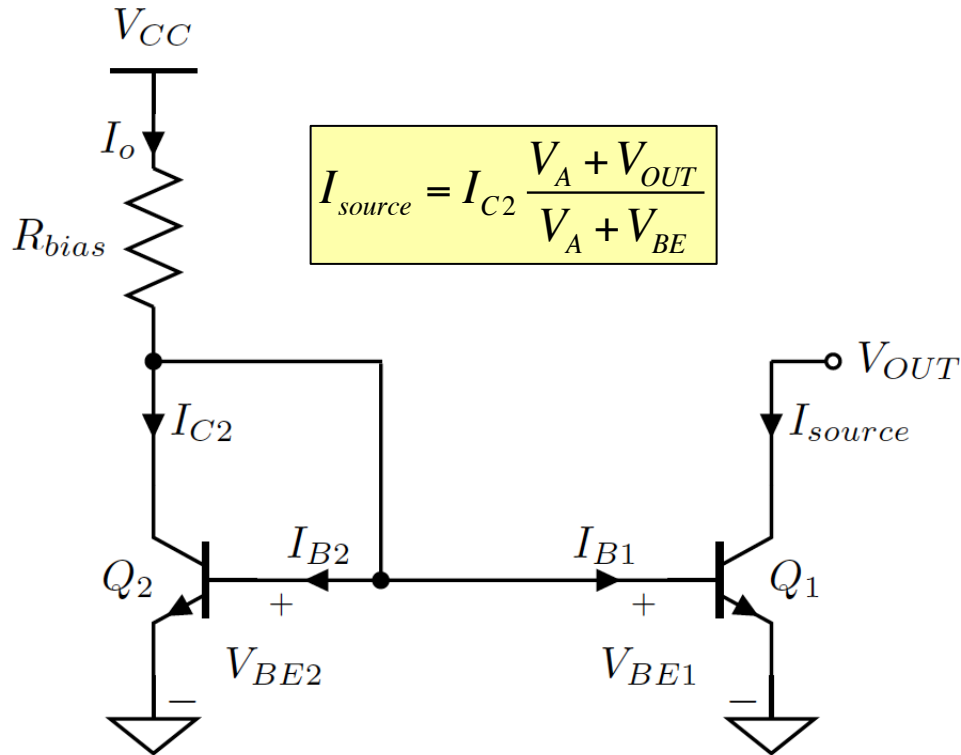


Recall:  $V_{BE} = V_T \ln \left( \frac{I_C}{I_S \left( 1 + \frac{V_{CE}}{V_A} \right)} \right)$

➔  $\frac{I_{C1}}{\left( 1 + \frac{V_{CE1}}{V_A} \right)} = \frac{I_{C2}}{\left( 1 + \frac{V_{CE2}}{V_A} \right)}$

$$I_{source} = I_{C2} \frac{\left( 1 + \frac{V_{CE1}}{V_A} \right)}{\left( 1 + \frac{V_{CE2}}{V_A} \right)} = I_{C2} \frac{V_A + V_{OUT}}{V_A + V_{BE}}$$

# A Simple BJT Current Mirror (3)

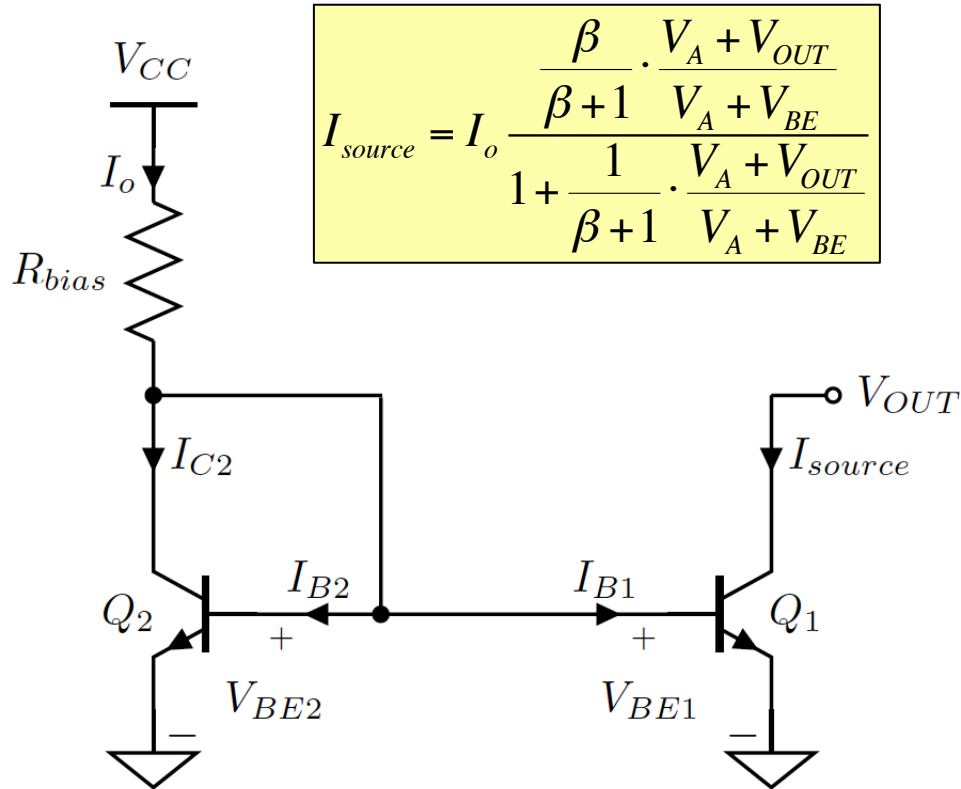


$$I_{C2} = I_o - I_{B1} - I_{B2} = I_o - \frac{I_{source}}{\beta} - \frac{I_{C2}}{\beta}$$

$$= \frac{\beta}{\beta+1} I_o - \frac{I_{source}}{\beta+1}$$

$$I_{source} = I_o \frac{\frac{\beta}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}{1 + \frac{1}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}$$

# A Simple BJT Current Mirror (4)



$$I_{source} = I_o \frac{\frac{\beta}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}{1 + \frac{1}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}$$

Assume  $V_A \rightarrow \infty$

$$I_{source} \approx I_o \frac{\frac{\beta}{\beta+1}}{1 + \frac{1}{\beta+1}} = \frac{I_o}{1 + \frac{2}{\beta}}$$

Assume  $V_A \rightarrow \infty$  and  $\beta \rightarrow \infty$

$$I_{source} \approx I_o$$

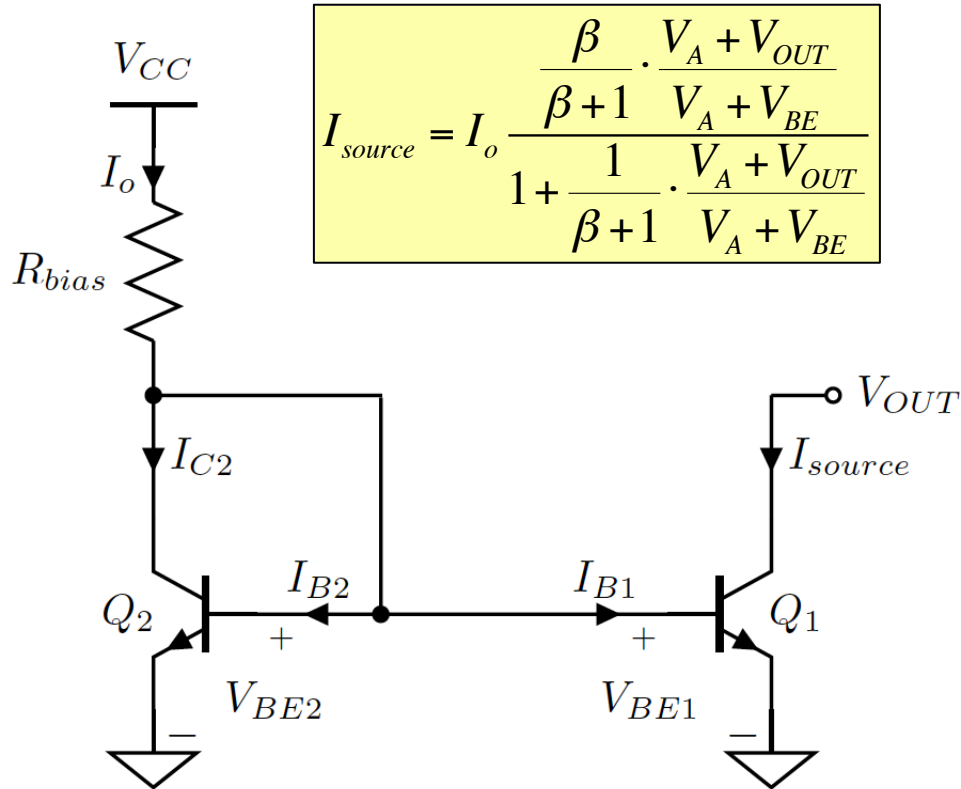
**Mirror!**

## Mirroring Error:

1. Due to the base currents
2. Due to  $V_{CE}$  "mismatch"



# A Simple BJT Current Mirror (4)



$$I_{source} = I_o \frac{\frac{\beta}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}{1 + \frac{1}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}$$

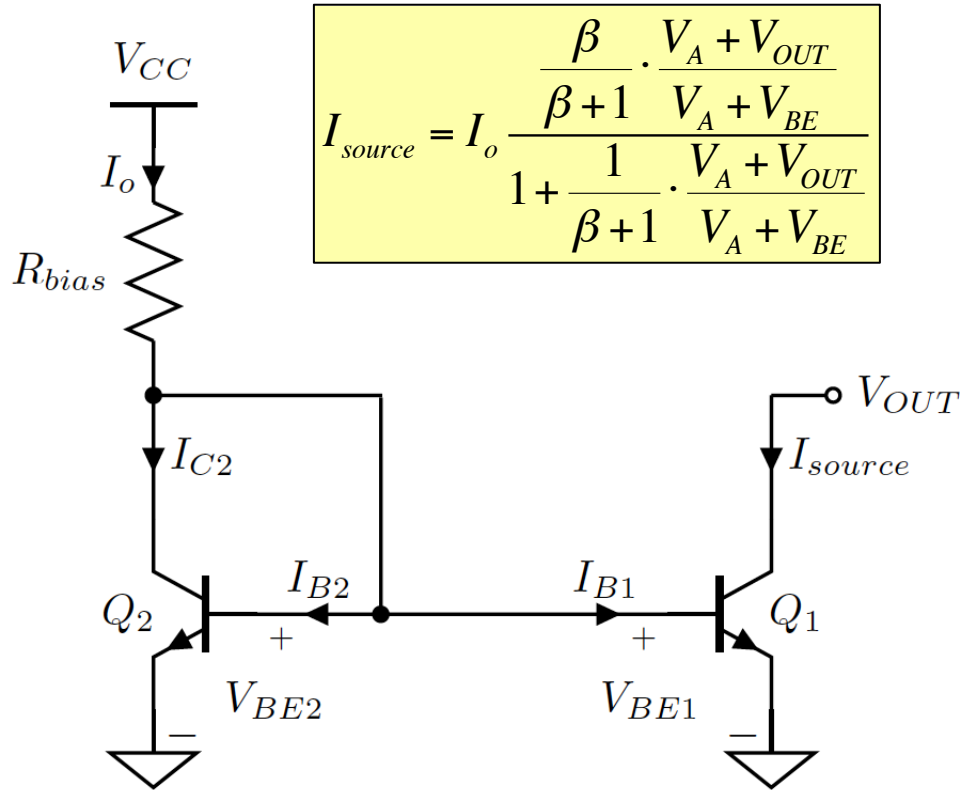
What about  $R_{bias}$ ?

$$R_{bias} = \frac{V_{CC} - V_{BE}}{I_o}$$

**R is linearly**  
related to  $I_o$  and  $I_{source}$   
(not exponentially!)



# A Simple BJT Current Mirror (5)



$$I_{source} = I_o \frac{\frac{\beta}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}{1 + \frac{1}{\beta+1} \cdot \frac{V_A + V_{OUT}}{V_A + V_{BE}}}$$

Assume  $V_A \rightarrow \infty$

$$I_{source} \approx \frac{I_o}{1 + \frac{2}{\beta}}$$

Again, what if we need 1mA?

- We can tolerate  $\beta$  as low as 99 and still get only 1% error!



# Next Meeting

- Biasing Amplifiers Using Current Sources
- Differential Circuits

