



# **EEE 51: Second Semester 2017 - 2018**

## **Lecture 11**

# Differential Circuits

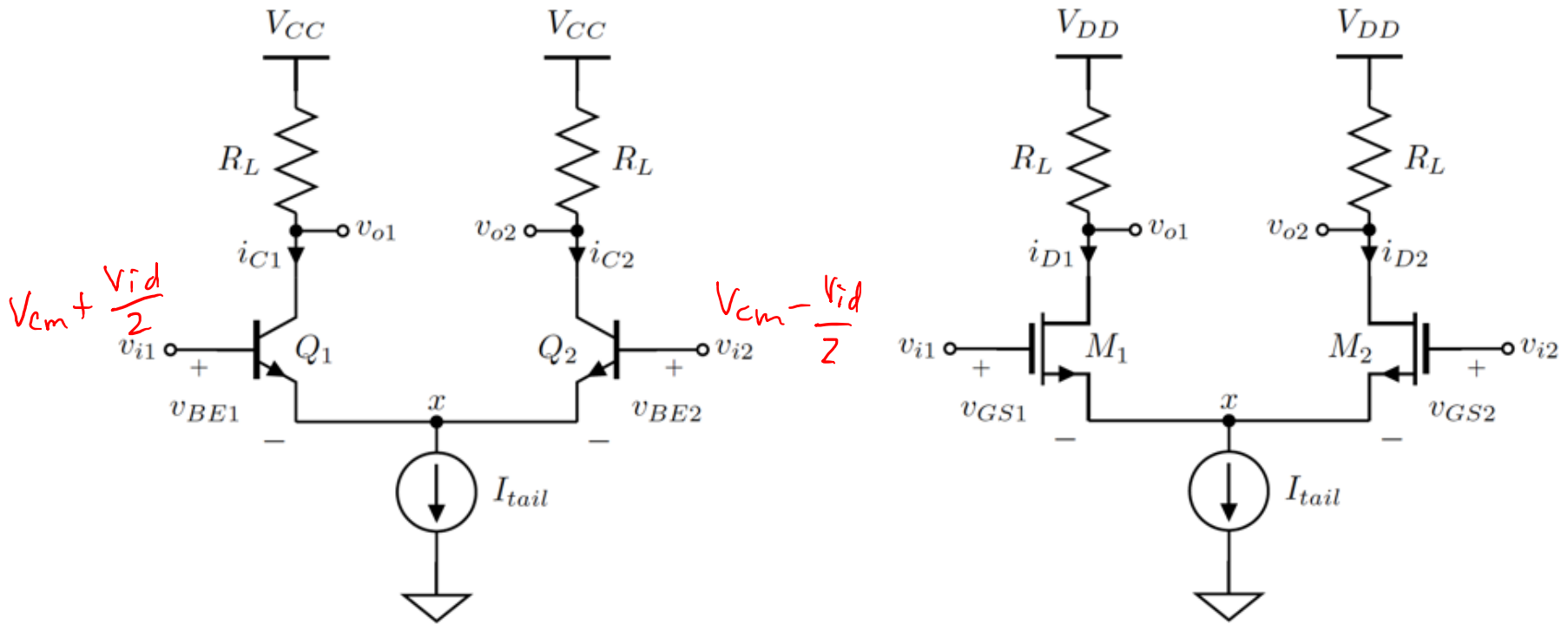
# Today

- Small signal analysis of differential circuits



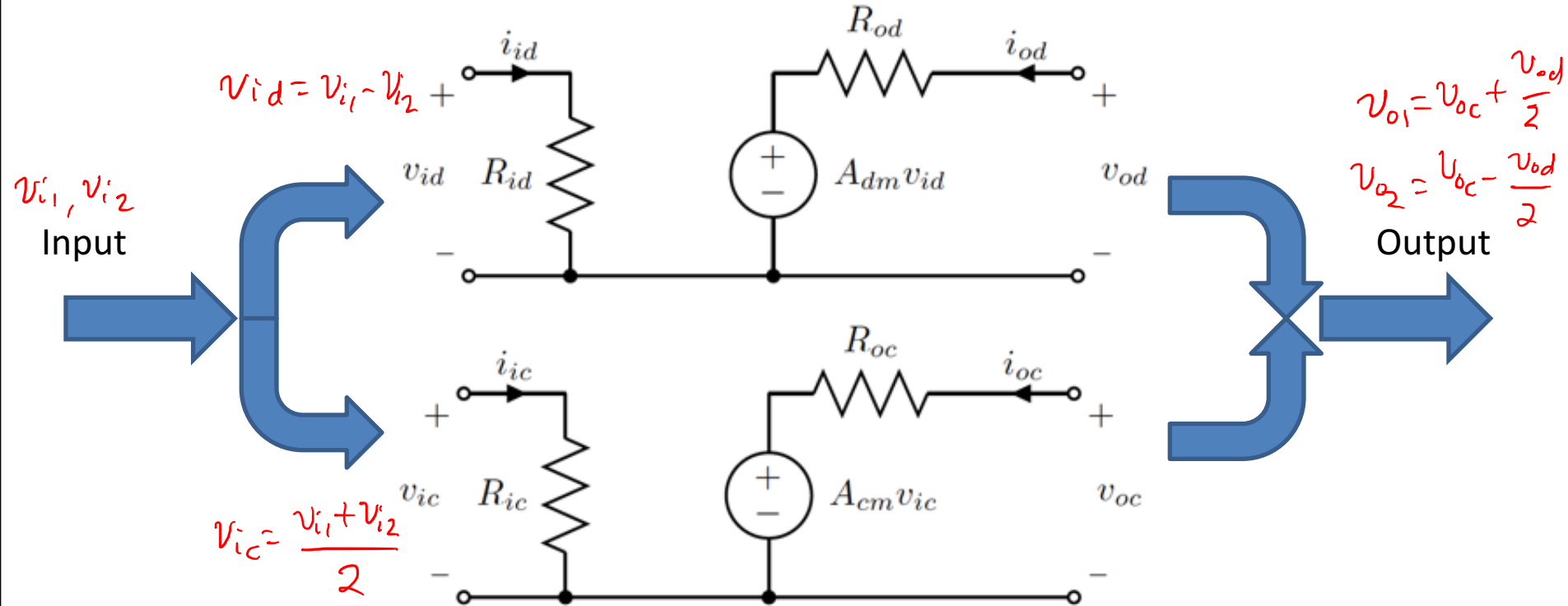
# Differential Amplifiers

- After DC analysis  $\rightarrow$  same small signal model

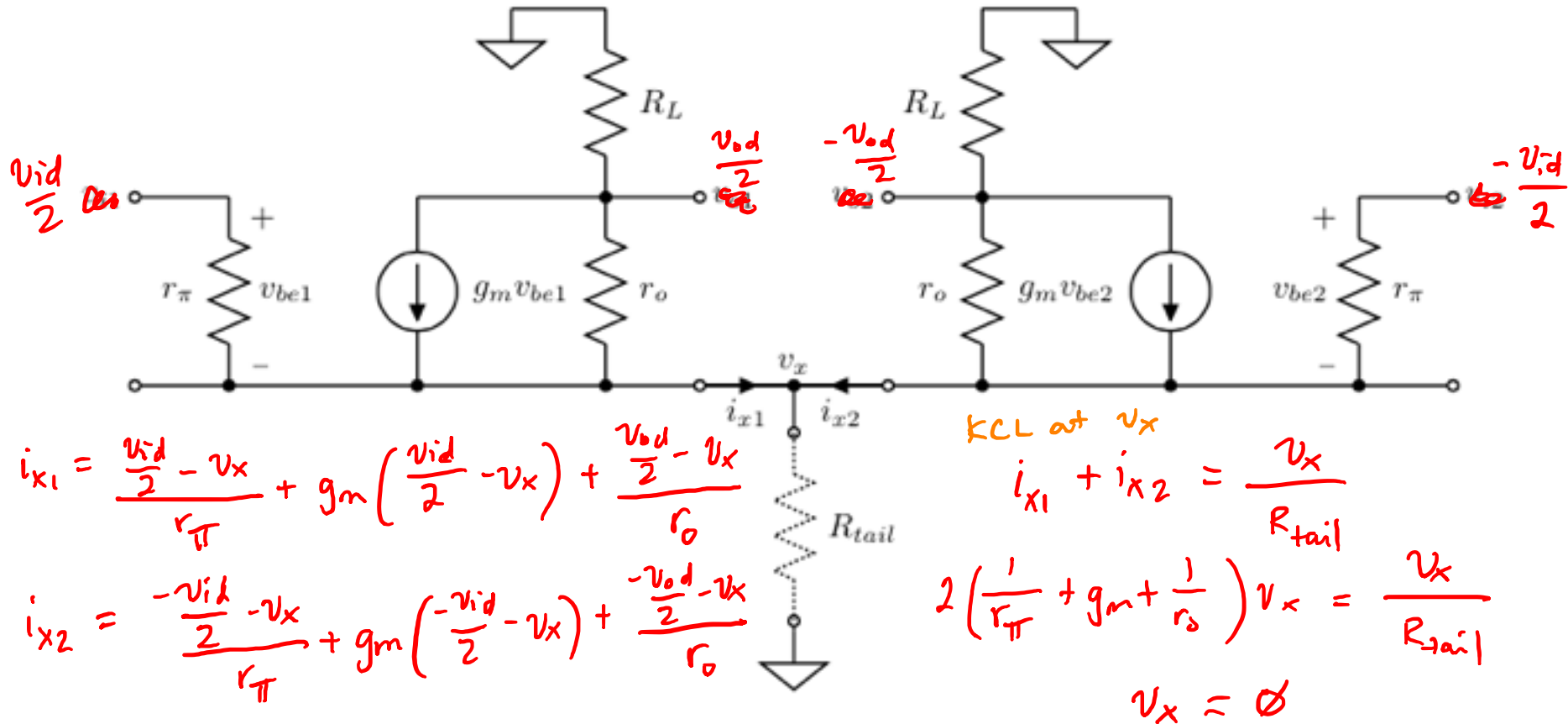


# Differential vs. Common-Mode

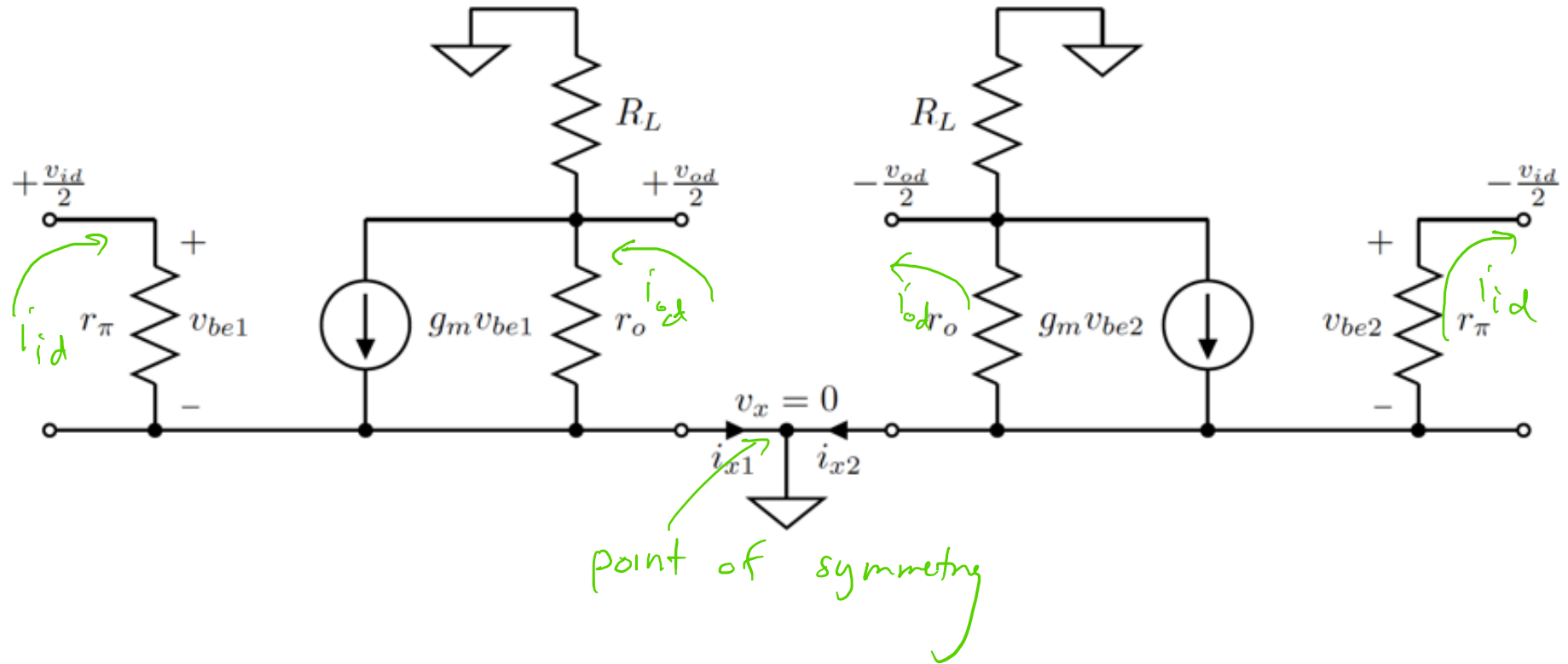
- The amplifier processes the two signals differently!



# Small-signal equivalent circuit



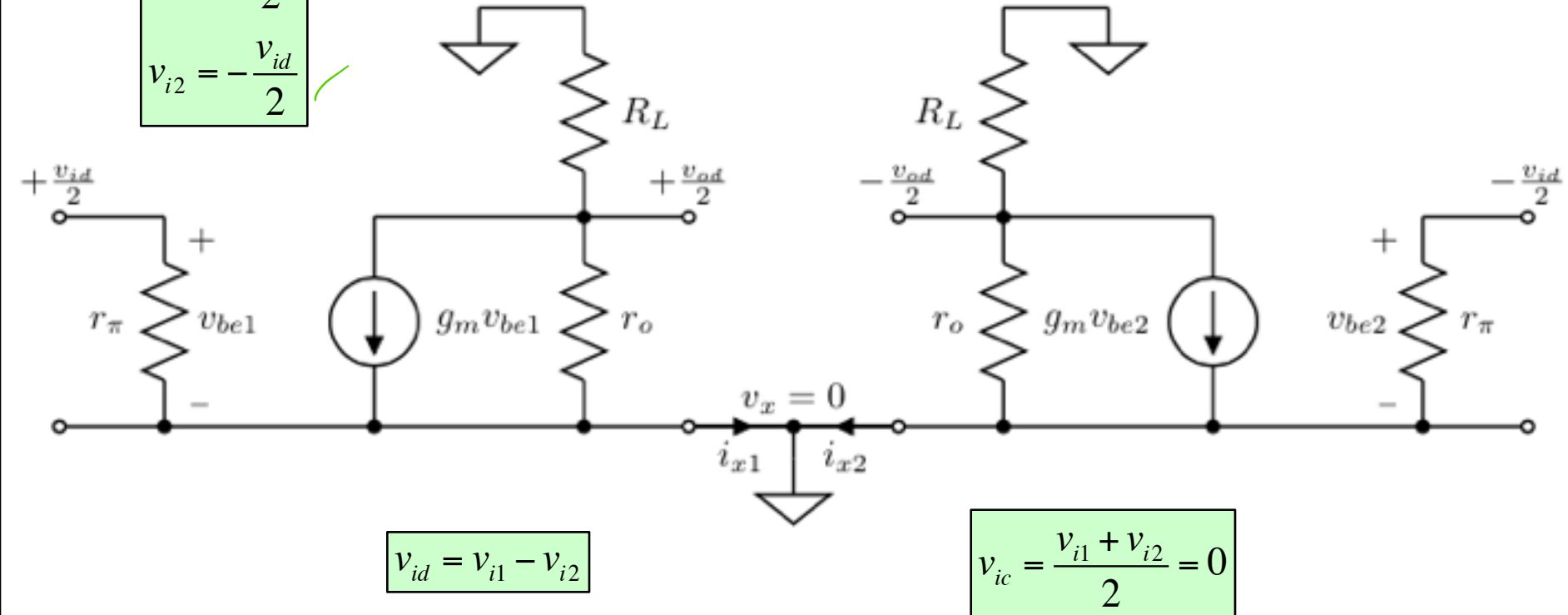
# The Differential Mode Half Circuit



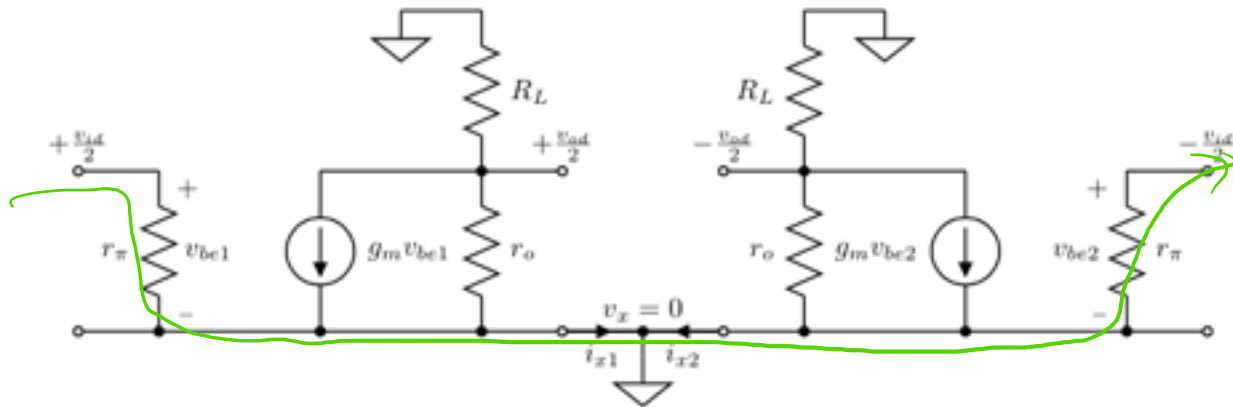
# Half-circuit analysis (DM)

$$v_{i1} = +\frac{v_{id}}{2}$$

$$v_{i2} = -\frac{v_{id}}{2}$$



# Half-circuit analysis (DM)



$$v_{i1} - v_{be1} + v_{be2} - v_{i2} = \left(+\frac{v_{id}}{2}\right) - v_{be1} + v_{be2} - \left(-\frac{v_{id}}{2}\right) = 0$$

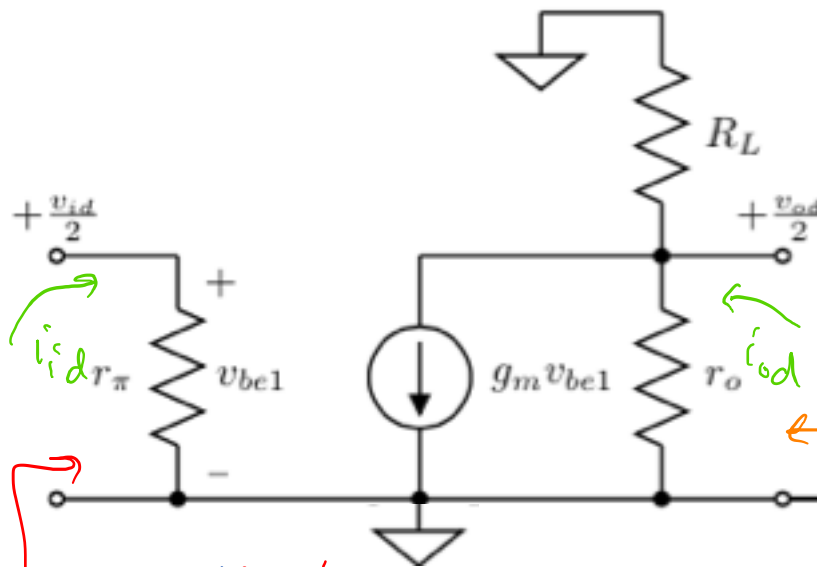
$$v_{id} = v_{i1} - v_{i2} = v_{be1} - v_{be2}$$

$$i_{x1} = -i_{x2} \quad \rightarrow \quad i_{x1} = i_{x2} = 0$$





# Half-circuit analysis (DM)



$$A_{dm} = \frac{+\frac{v_{od}}{2}}{+\frac{v_{id}}{2}} = \frac{v_{od}}{v_{id}} = -g_m \cdot (r_o \parallel R_L)$$

$$R_{id} = \frac{v_{id}}{i_{id}} = 2 \cdot r_{\pi}$$

$$R_{od} = \frac{v_{od}}{i_{od}} = 2 \cdot (r_o \parallel R_L)$$

$$G_{md} = \frac{-A_{dm}}{R_{od}} = \frac{g_m}{2}$$

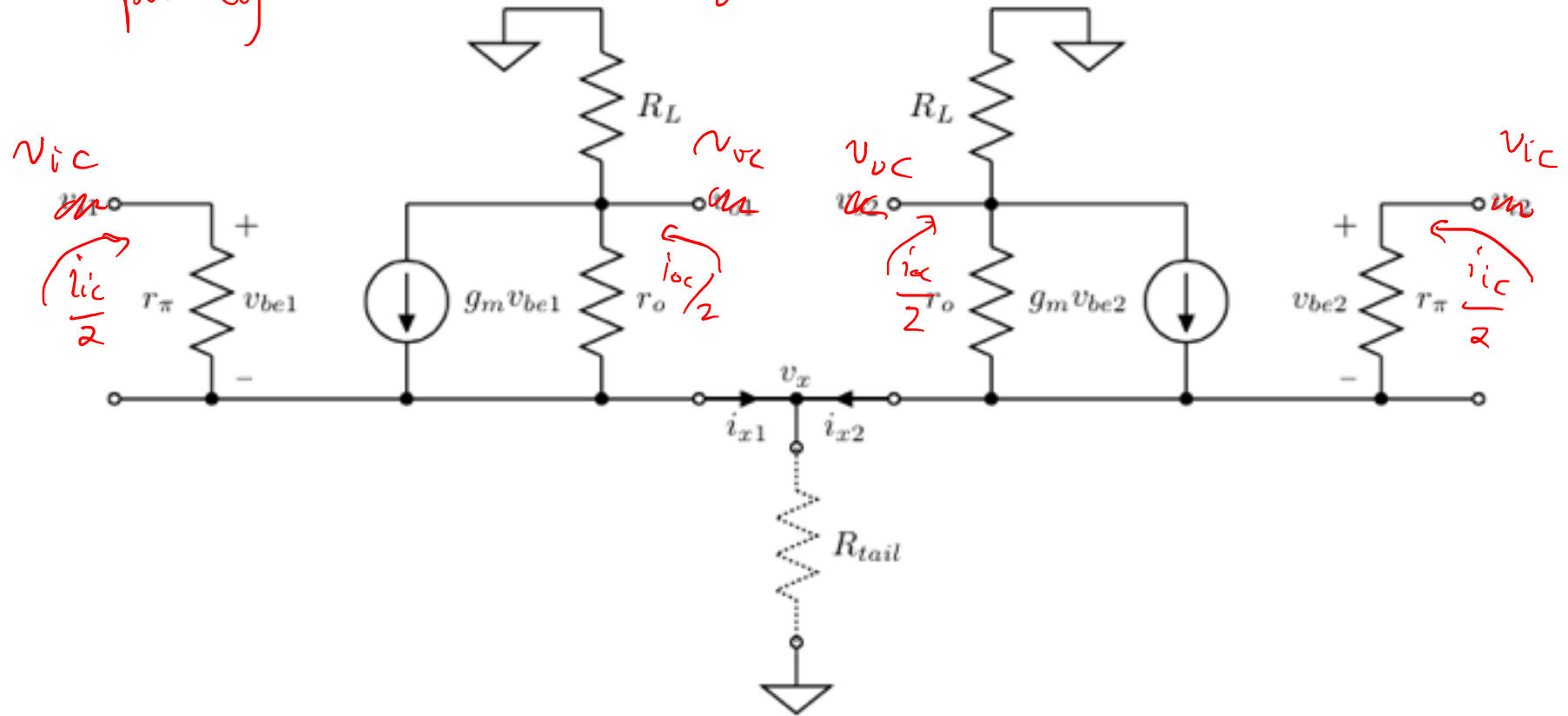
$$R_{out} = \left( \frac{v_{od}/2}{i_{od}} = r_o \parallel R_L \right) 2$$

$$R_{in} = \left( \frac{v_{id}/2}{i_{id}} = r_{\pi} \right) 2$$

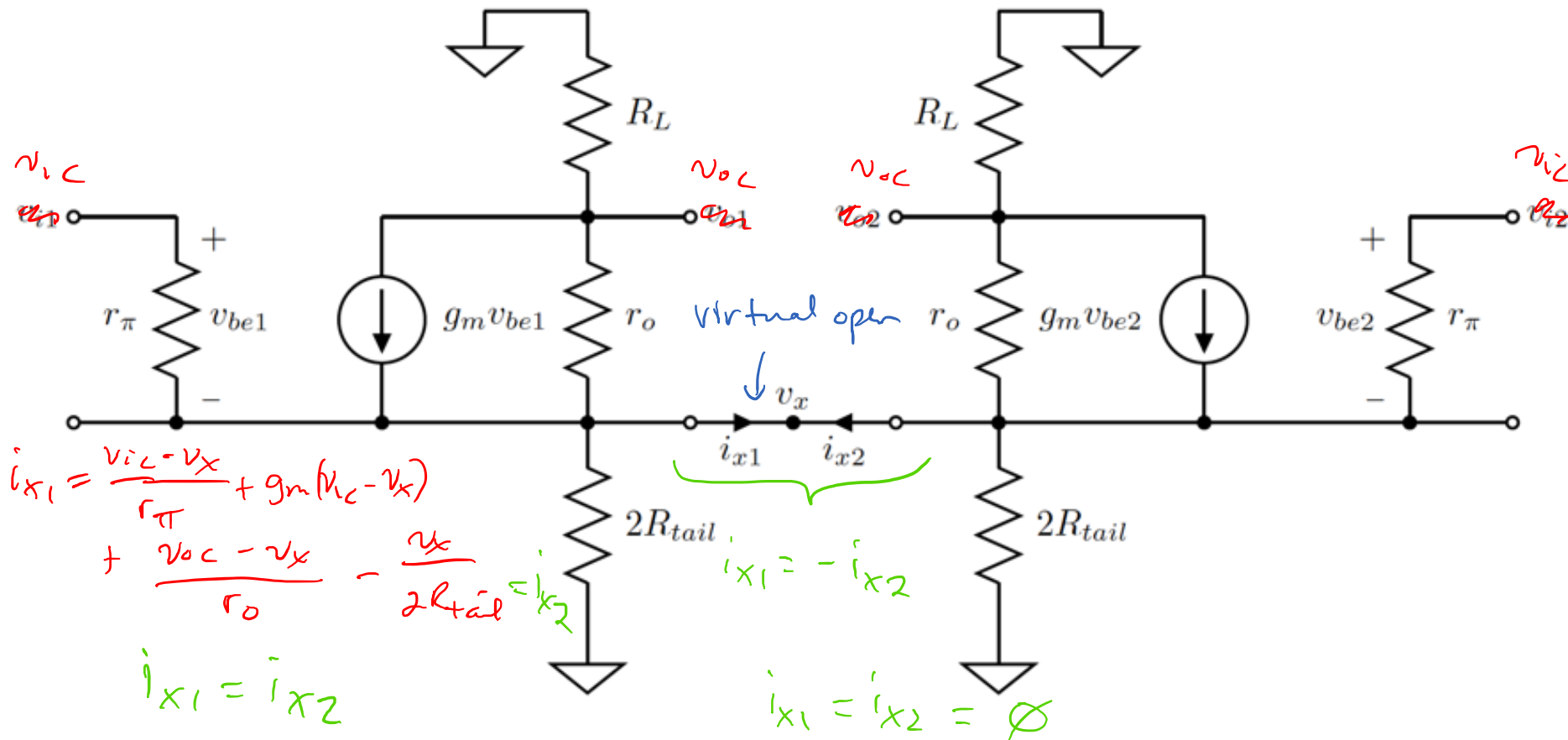


# Small-signal equivalent circuit

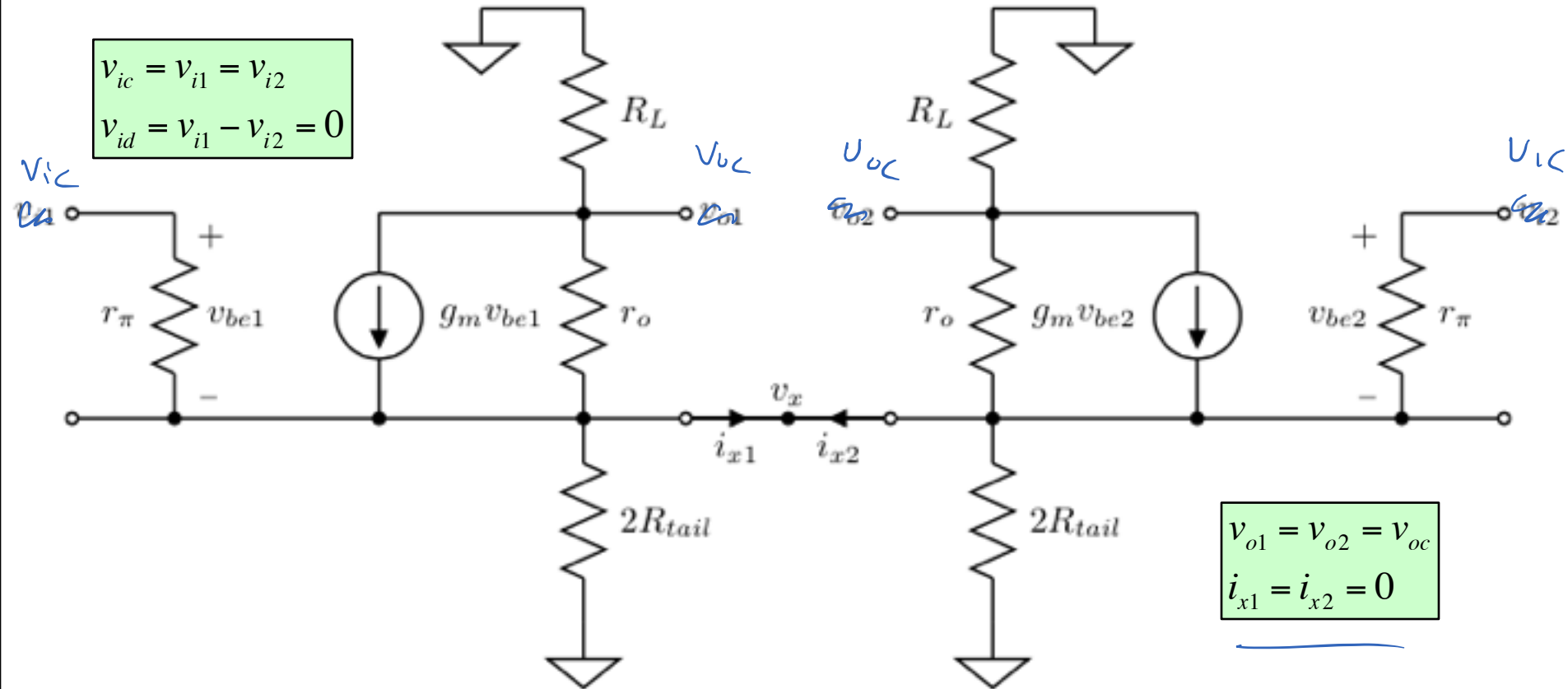
purely common mode signals ( $v_{id} = 0$ )



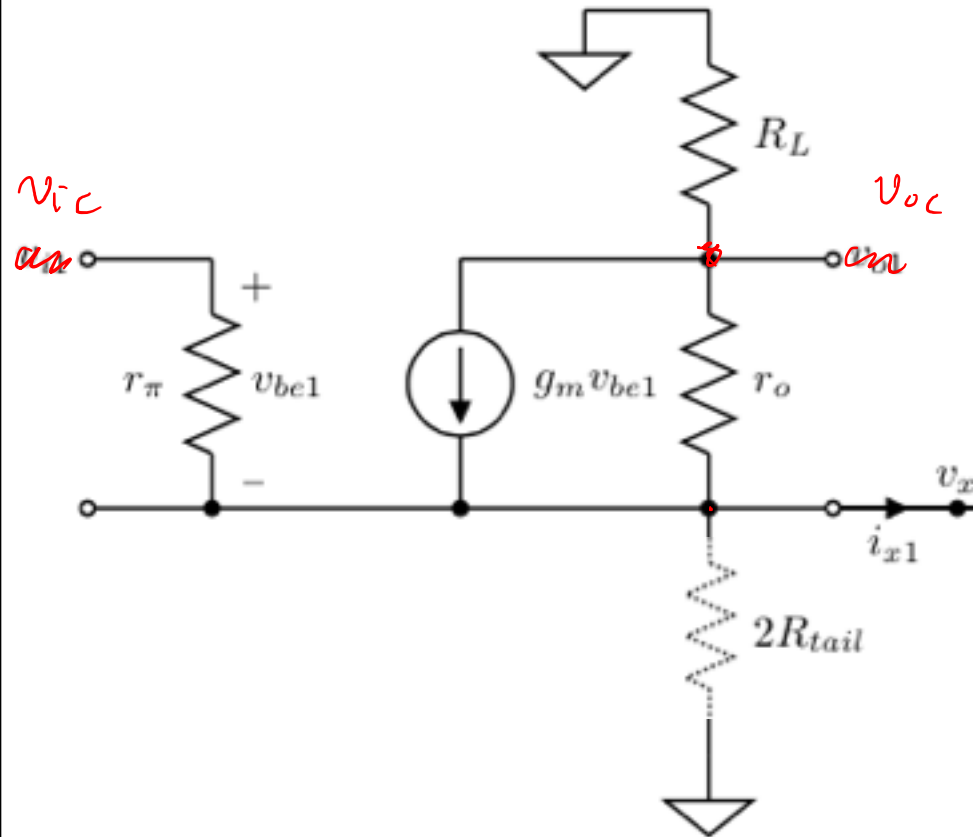
# The Common-Mode Half Circuit



# Half-circuit analysis (CM)



# Half-circuit analysis (CM)



if  $I_{tail}$  is ideal ( $2R_{tail} \rightarrow \infty$ )

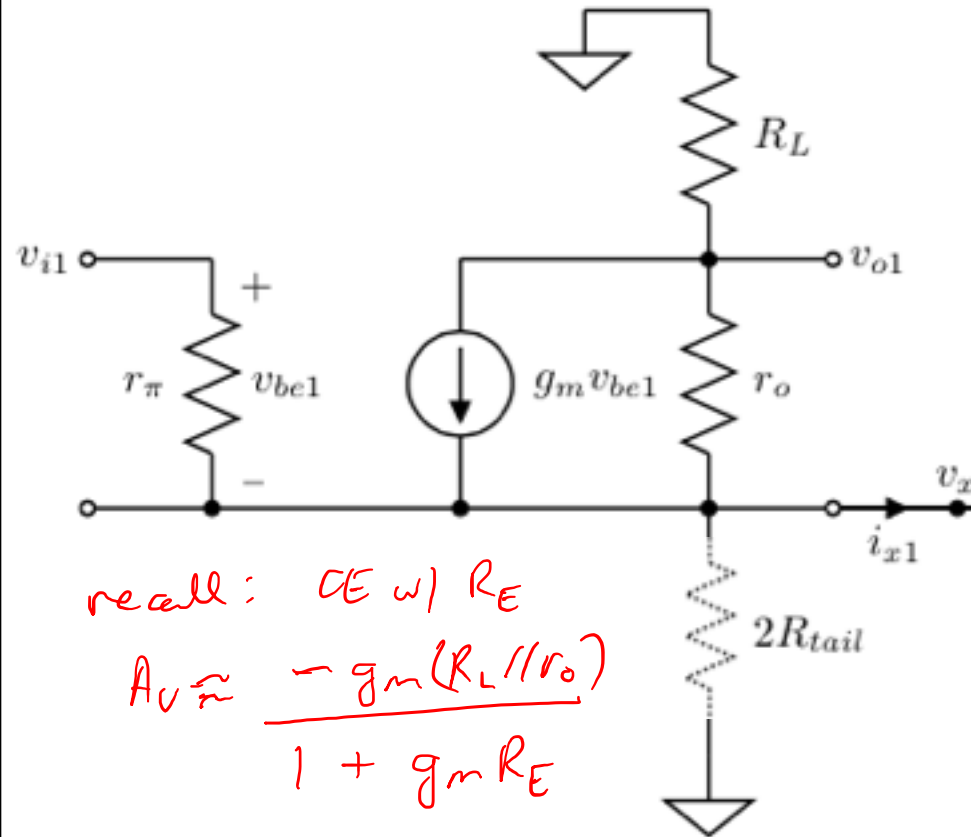
$$\frac{v_x - v_{ic}}{r_\pi} + \frac{v_x - v_{oc}}{r_o} - g_m (v_{ic} - v_x) = 0$$

$$\frac{v_{oc}}{R_L} = \frac{v_{ic} - v_x}{r_\pi}$$

$$v_{oc} = v_{ic} \cdot \frac{R_L}{R_L + r_o + r_\pi (1 + g_m r_o)}$$



# Half-circuit analysis (CM)



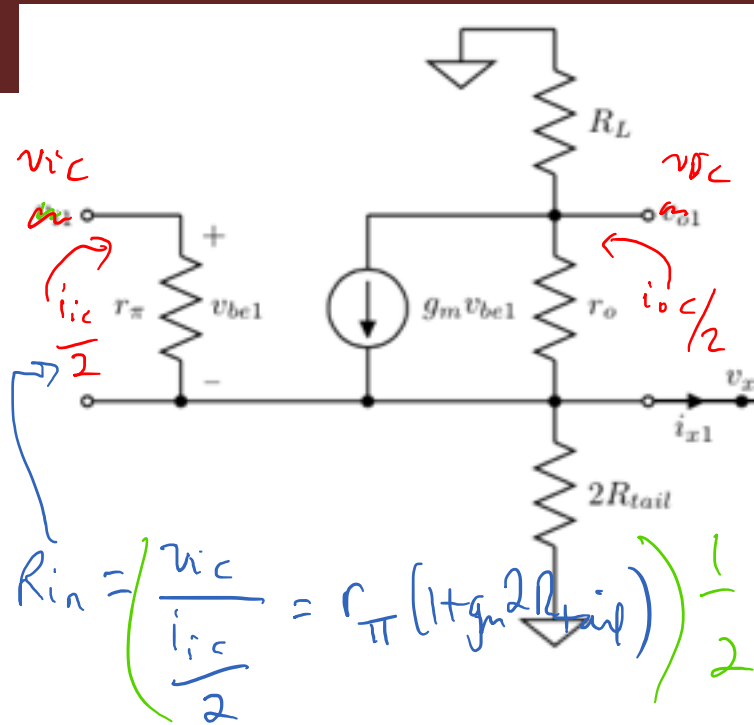
$$A_{cm} = \frac{v_{oc}}{v_{ic}} = \frac{R_L}{R_L + r_o + r_\pi (1 + g_m r_o)}$$

$$A_{cm} \approx \frac{R_L}{r_\pi g_m r_o} \approx \frac{R_L}{\frac{\beta}{g_m} g_m r_o} \approx \frac{R_L}{\beta \cdot r_o} \approx 0$$

If current source is not ideal:

$$A_{cm} = -\frac{g_m R_L}{1 + 2 \cdot g_m R_{tail}}$$





$$R_{in} = \left( \frac{v_{ic}}{\frac{i_{ic}}{2}} = r_{\pi} (1 + g_m 2R_{tail}) \right) \frac{1}{2}$$

$$R_{out} = \frac{v_{oc}}{\frac{i_{oc}}{2}} = r_o (1 + g_m 2R_{tail}) // R_L$$

$$A_{cm} = \frac{v_{oc}}{v_{ic}} = \frac{-g_m (r_o // R_L)}{1 + g_m (2R_{tail})}$$

$$R_{ic} = \frac{v_{ic}}{i_{ic}} = \frac{r_{\pi} (1 + g_m (2R_{tail}))}{2}$$

$$R_{oc} = \frac{v_{oc}}{i_{oc}} = \frac{r_o (1 + g_m (2R_{tail})) // R_L}{2}$$

$$G_{mc} = \frac{-A_{cm}}{R_{oc}} \approx \frac{2g_m}{1 + 2g_m R_{tail}}$$



# Next Meeting

- Finish up discussion of Differential Circuits
- Compound Amplifiers

