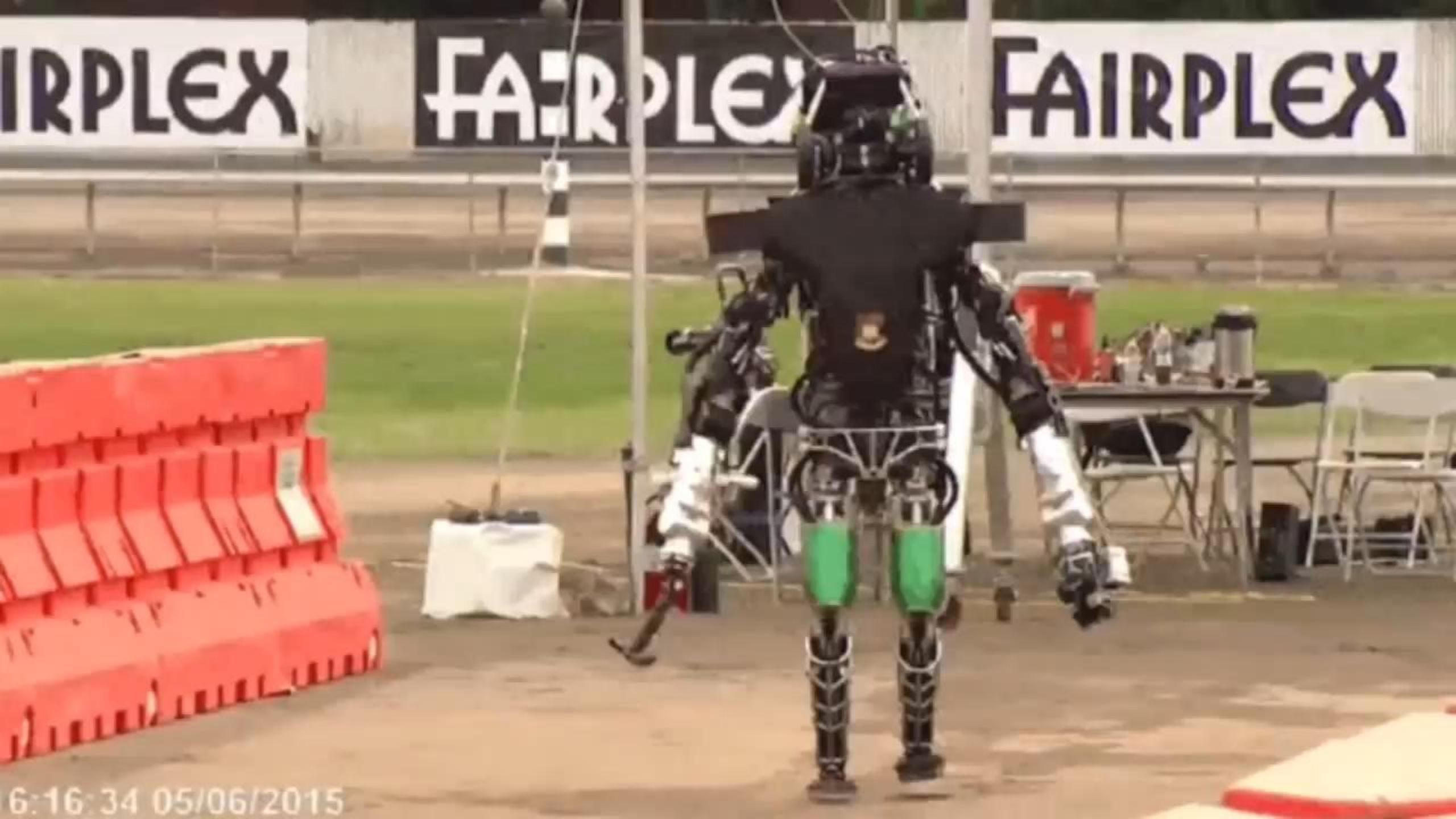


A decorative graphic on the left side of the slide, consisting of a network of thin, light-blue lines and small circles, resembling a circuit board or a neural network, extending vertically from the top to the bottom.

# FUNDAMENTALS OF FEEDBACK CONTROL

LAB 3



6:16:34 05/06/2015

# FEEDBACK CONTROL

Often the inherent system behavior is unsatisfactory

- Slow response
- Unstable response
- External influences

*No control:*



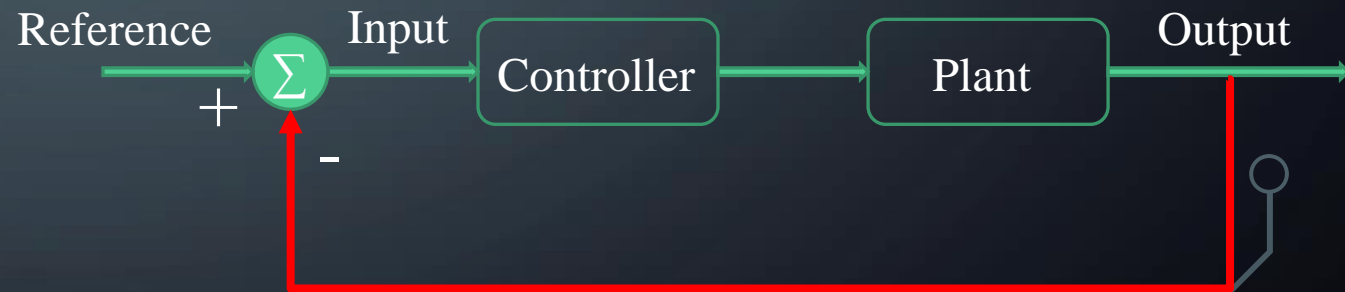
Output is the natural response of the plant

*Open-loop:*



Output has no effect on the control action

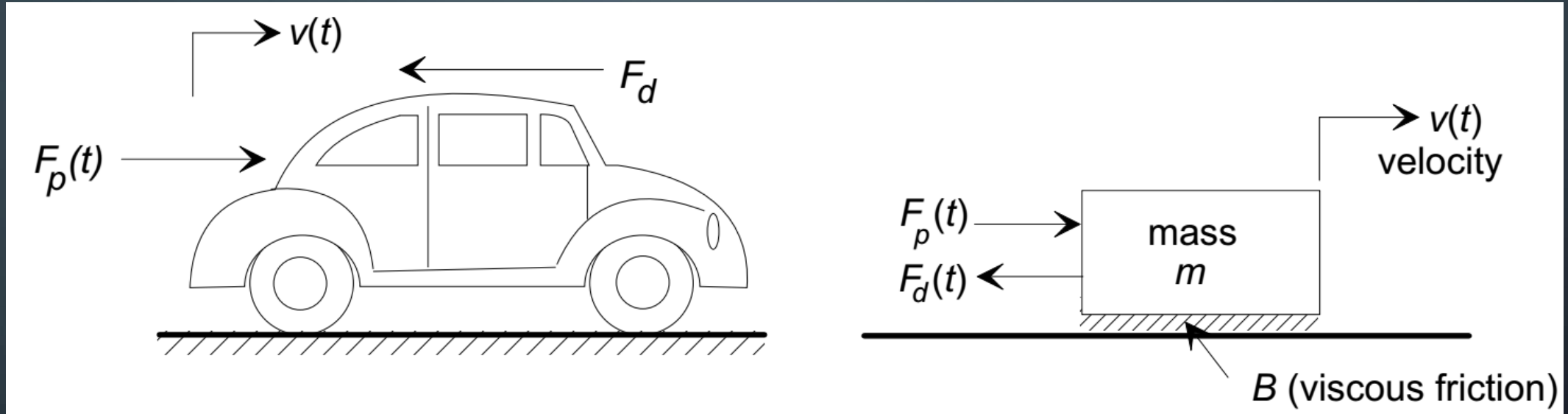
*Closed-loop:*



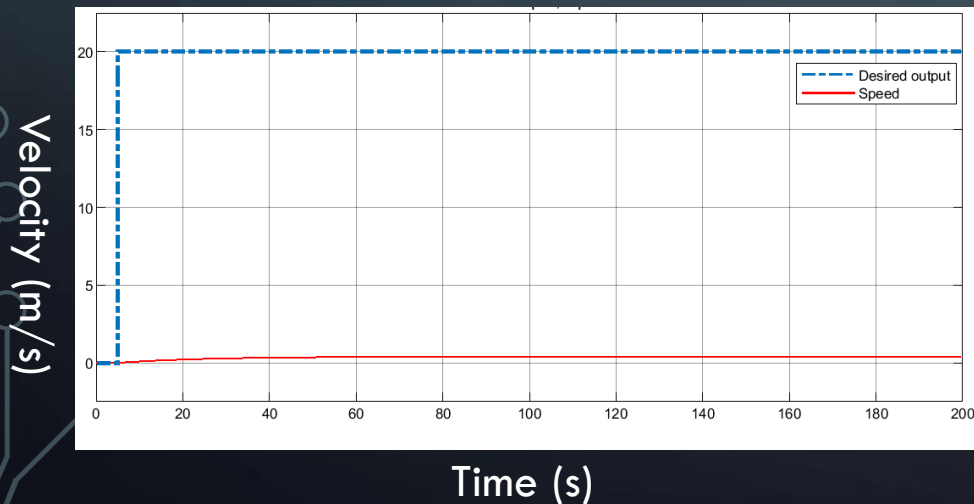
Output has effect on the control action through **feedback**

# FEEDBACK CONTROL EXAMPLE

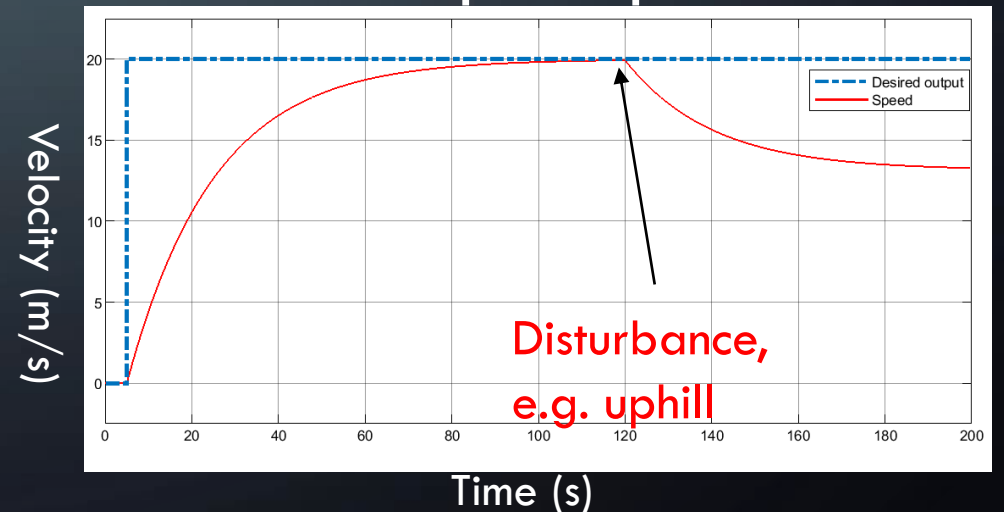
Example: Car cruise control with the desired velocity of 20 m/s



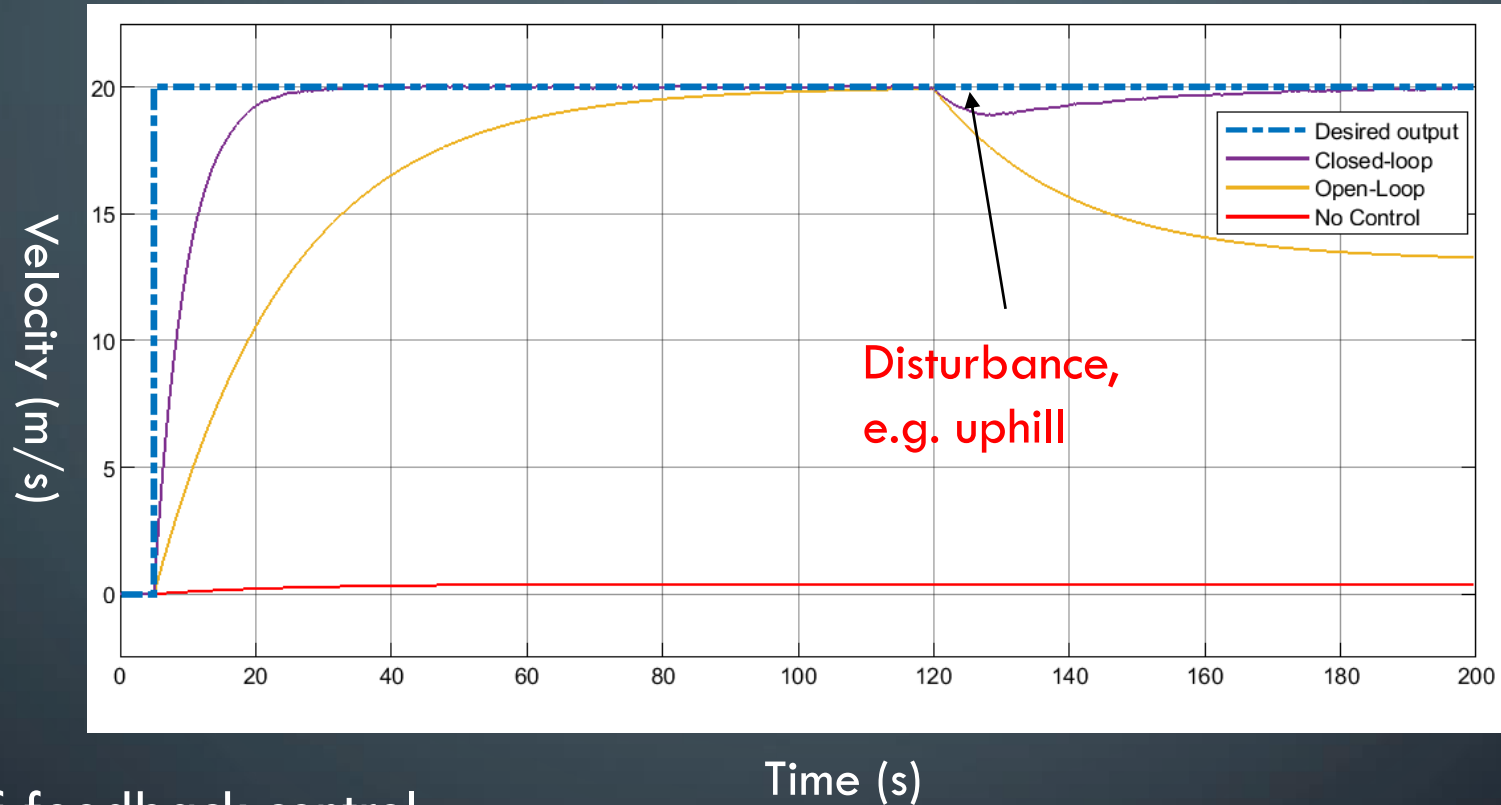
No control



Open-loop



# BENEFITS OF FEEDBACK CONTROL



Benefits of feedback control:

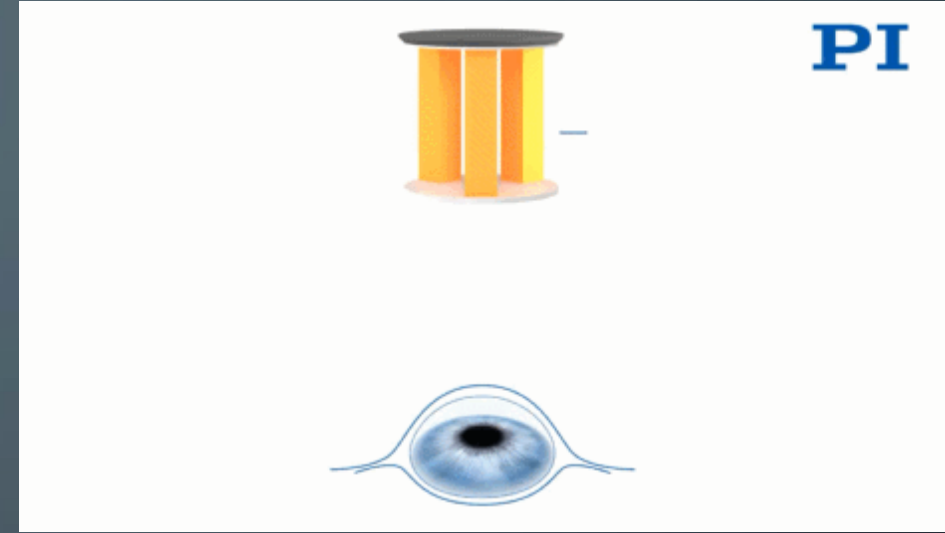
- Stabilize the plant (Stabilization)
- Regulate the output to follow the desired reference (Regulation)
- Better transient performance (Tracking)
- Reduce response to disturbances (Disturbance Rejection)



# APPLICATIONS



Hard disk



Surgery robot



Robot Arm

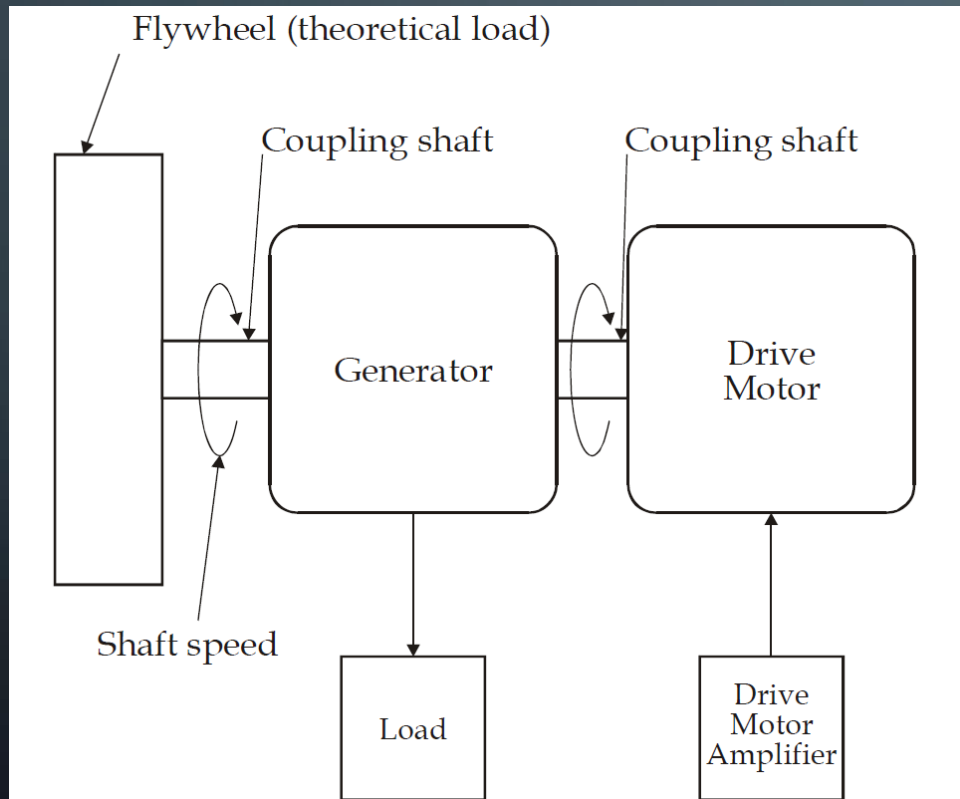


Segway

# FUNDAMENTALS OF FEEDBACK CONTROLS

- Control theory
- Servo trainer modelling
- Velocity control system
- Goal: The objective of this experiment is to study the speed control of a motor with different types of controls. Also, to investigate the effect of proportional controls and PI controls on the **steady state error**.

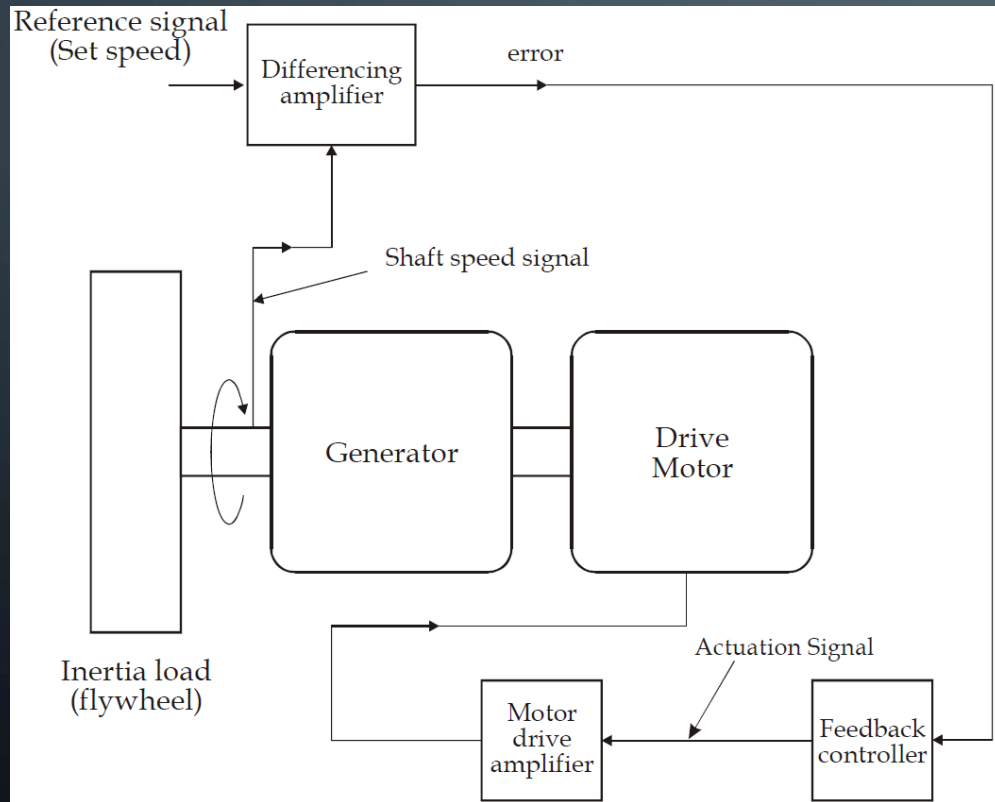
# CONTROL PRINCIPLES



- Manual adjustment
  - Measurement of speed
  - Computation of remedial action
  - Manual effort for load adjustment
- Problems
  - Time consuming and expensive
  - Concentration
  - Response speed



# CONTROL PRINCIPLES

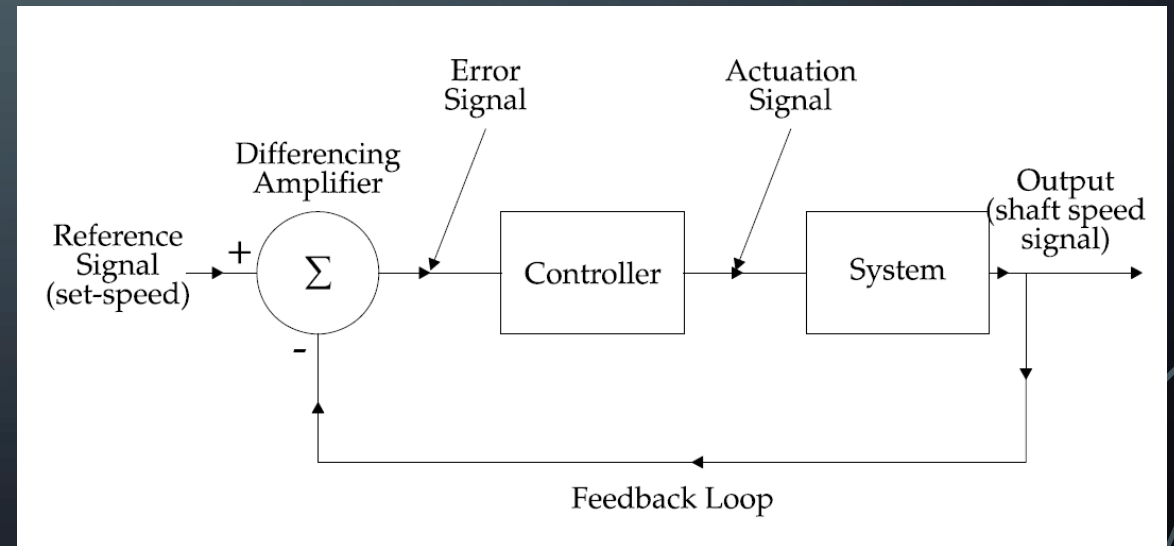


- Error signal

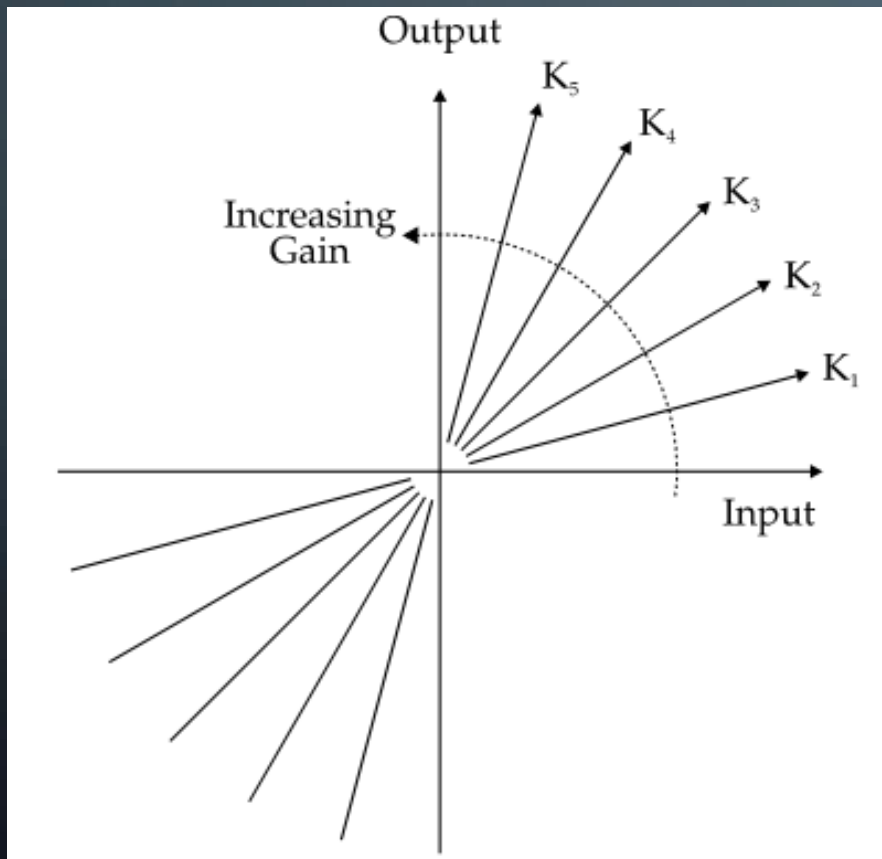
= Measured signal - reference signal  
(Setpoint)

- Feedback

- Closed-loop control system

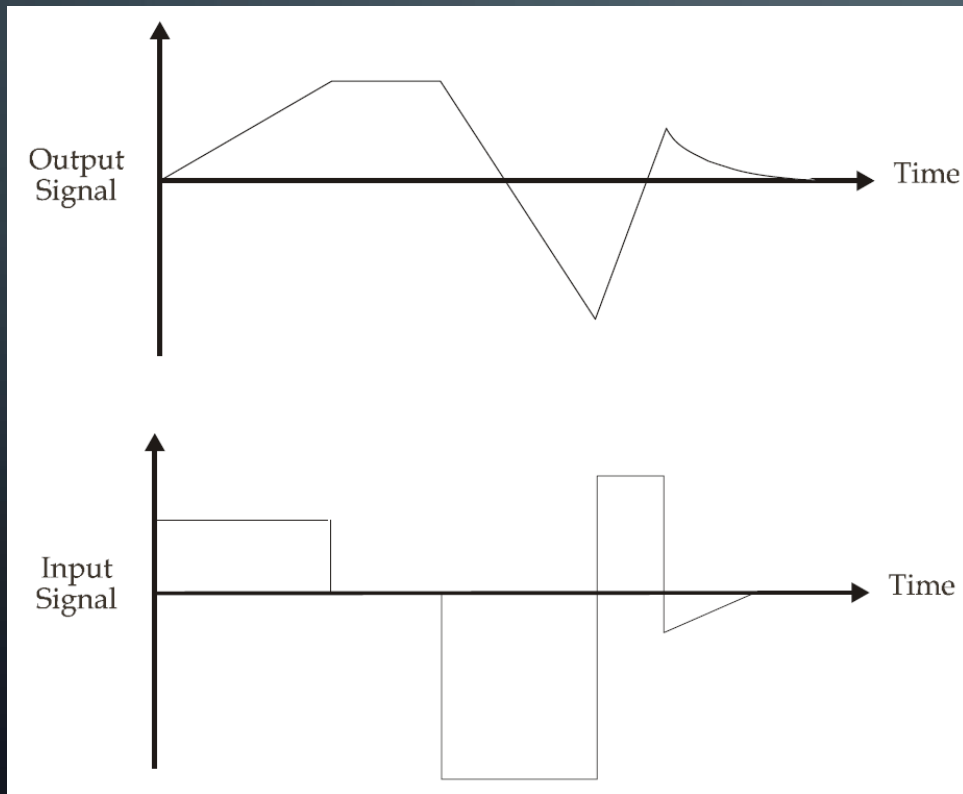


# PROPORTIONAL CONTROL

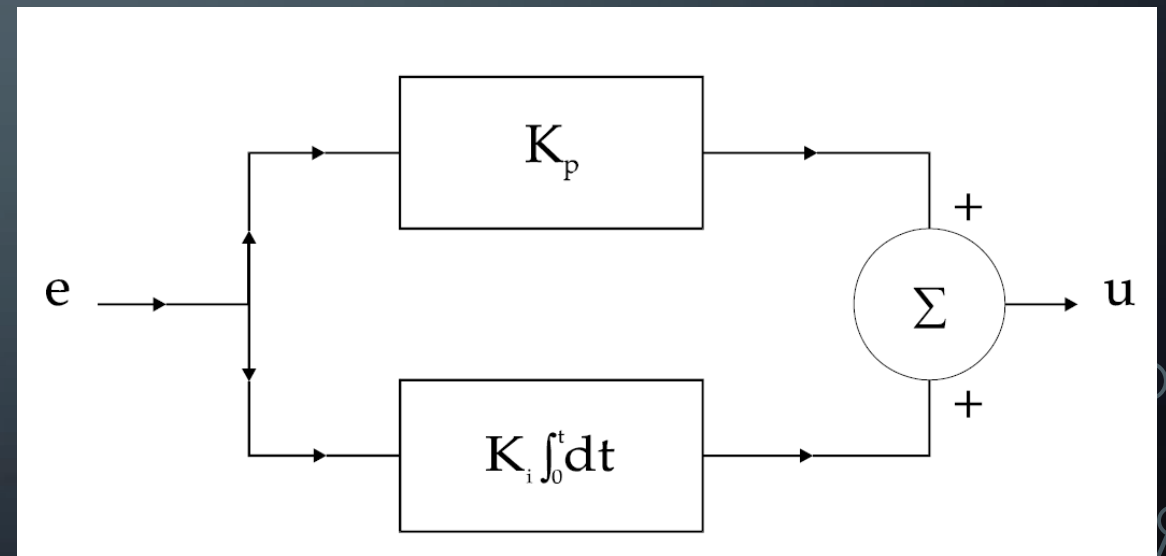


- Steady state error  
 $\equiv \text{actual speed} - \text{set speed}$
- Gain
- Instability

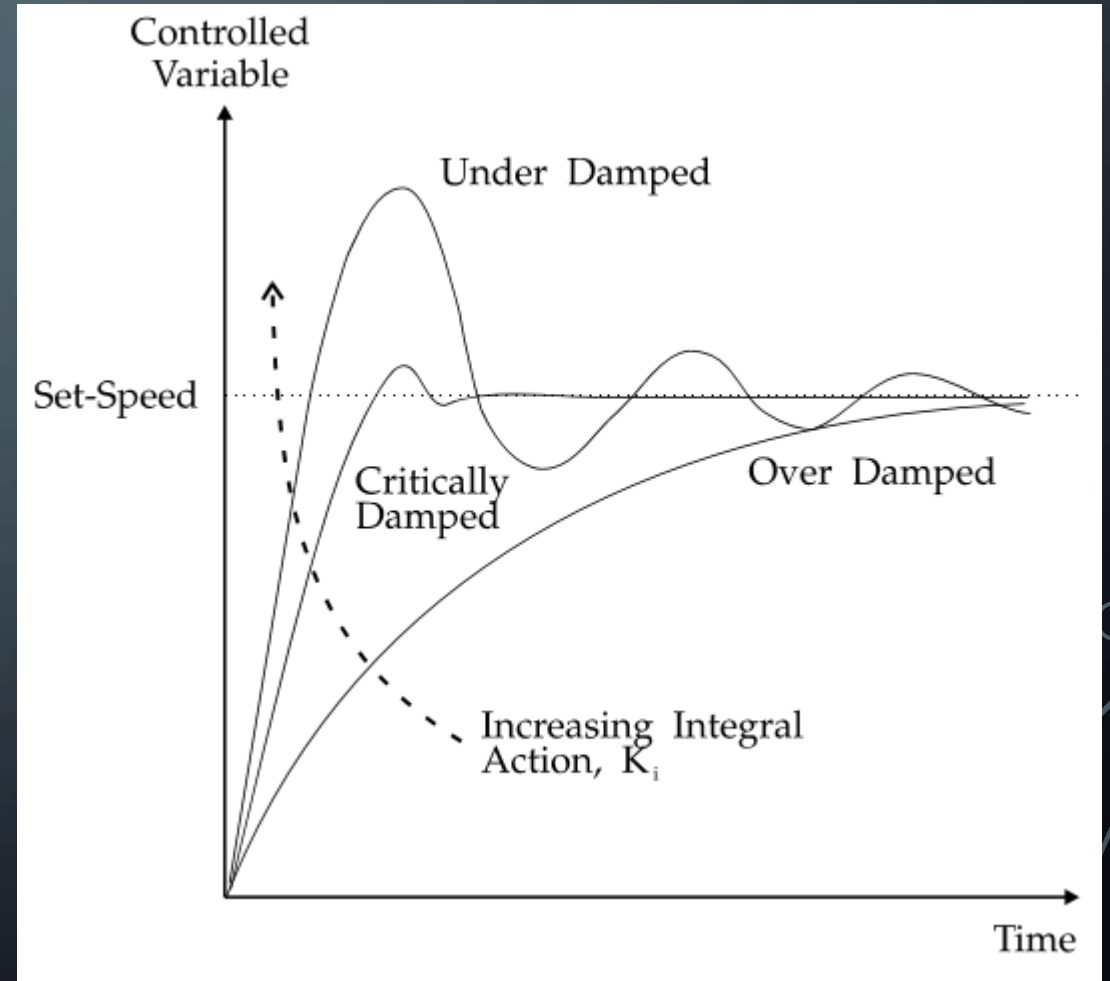
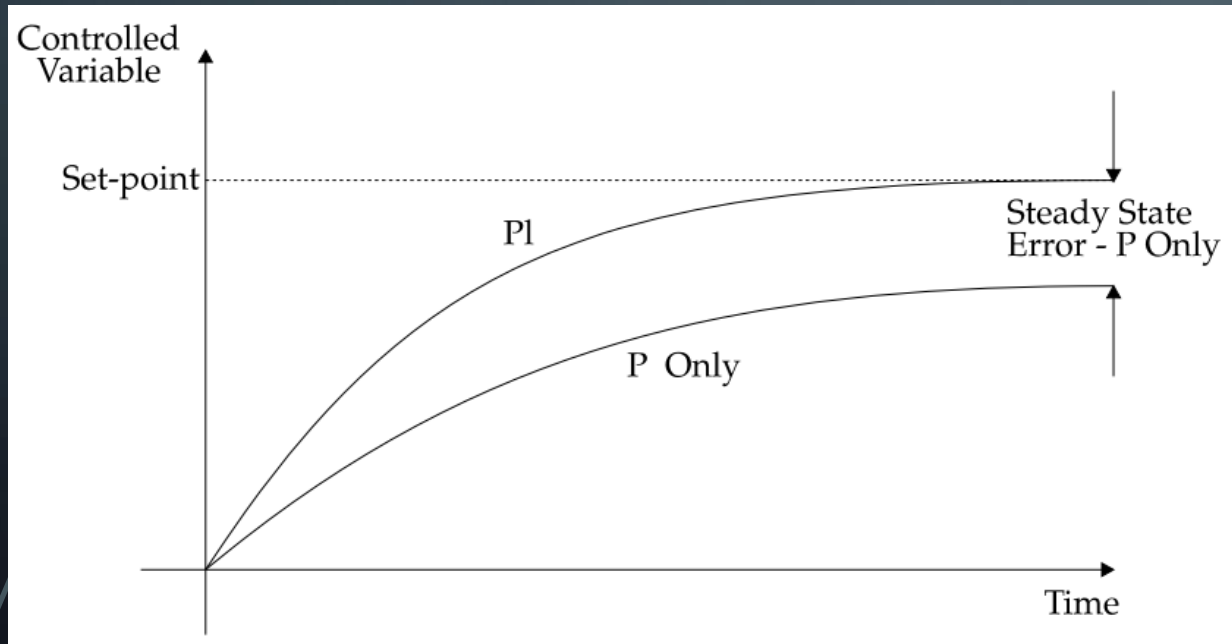
# INTEGRAL CONTROL



- Input: zero, output: constant
- Input: positive, output: ramp upward
- Input: negative, output: ramp downward
- Integration characteristic



# PROPORTIONAL INTEGRAL CONTROL

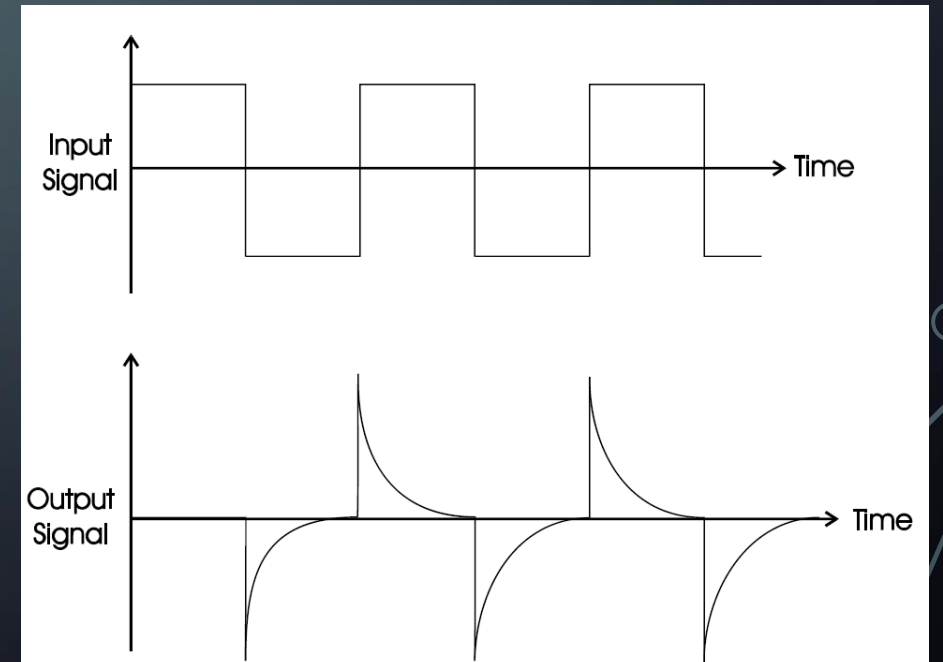
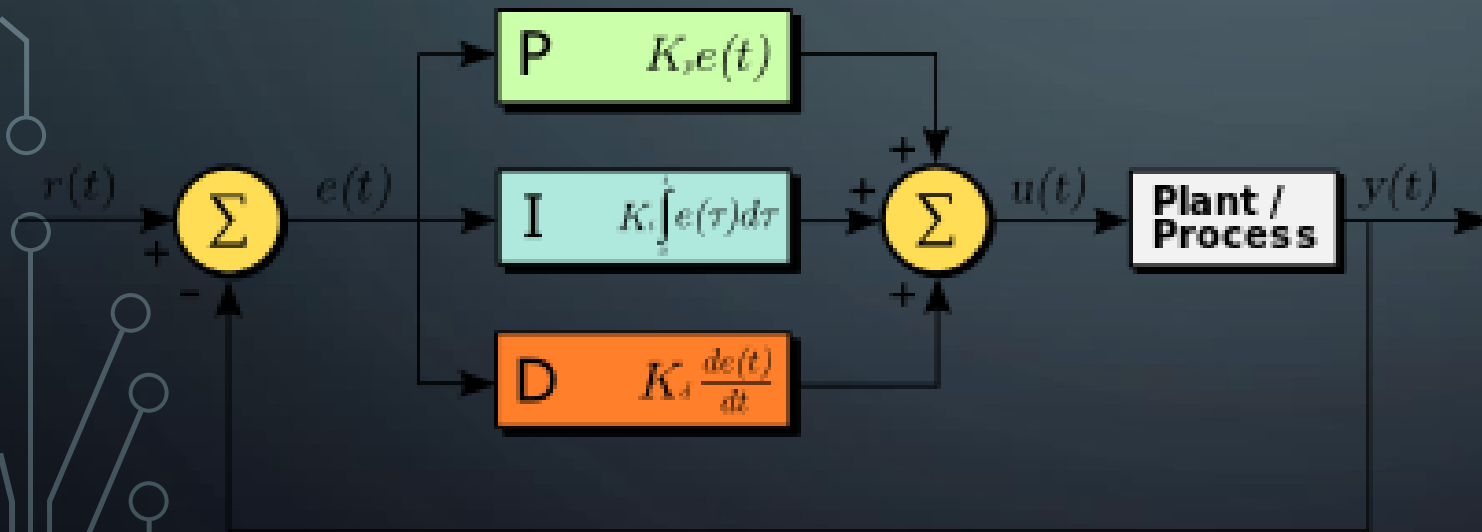


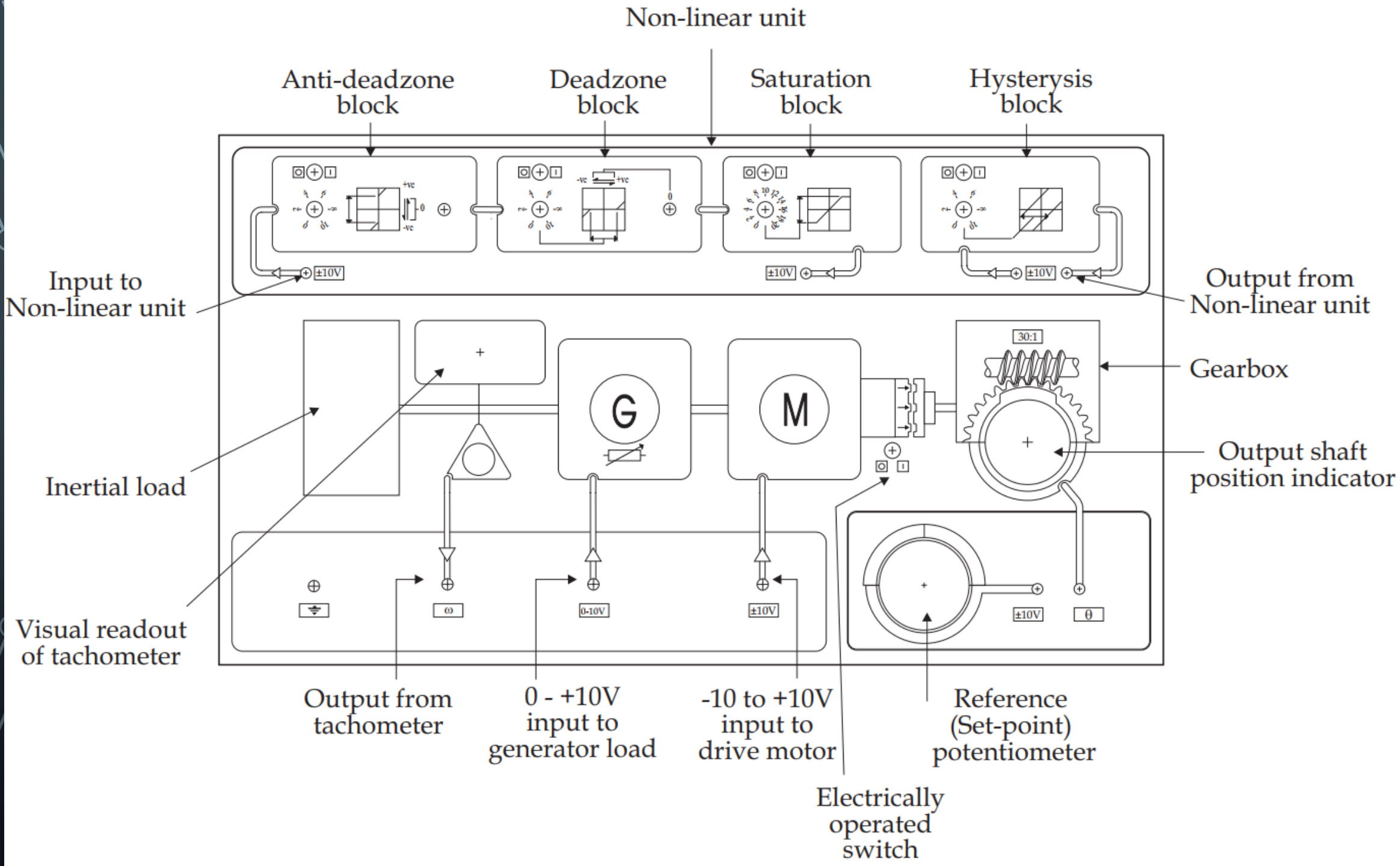
# THREE TERM CONTROLLER

- Fast response with minimum overshoot
- Proportional-integral-derivative

- Differential control

- Input: reverse polarity
- Output: large peak then decay

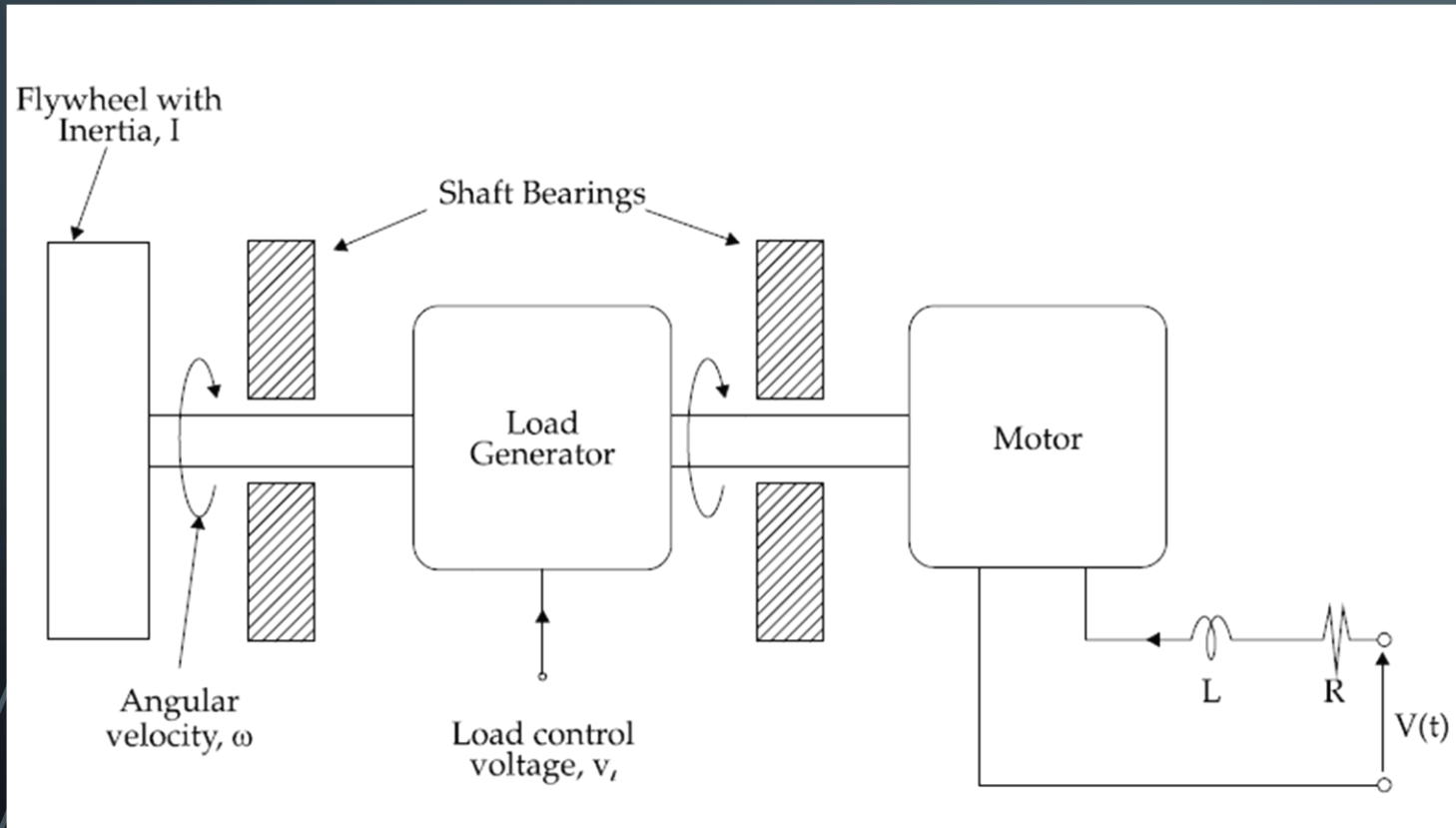








# SERVO TRAINER ANALYSIS



- Torque is proportional to current
$$\tau_m = k_m i$$
- Back EMF voltage is proportional to rotation speed

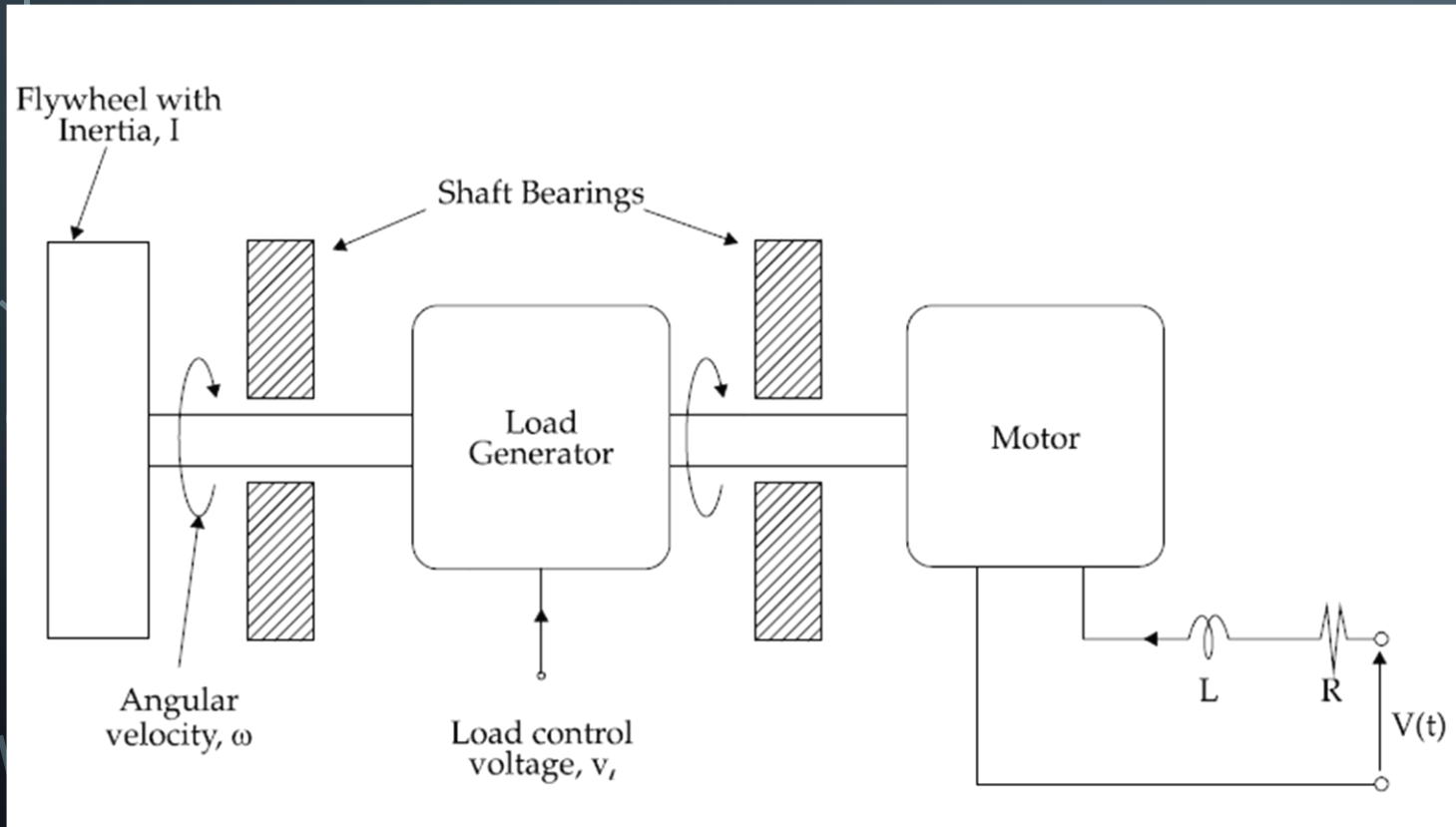
$$v_{bemf} = k_m \omega$$

- System model

$$\tau_m = b\omega + k_l v_l + I \frac{d\omega}{dt}$$

$$V(t) = Ri + L \frac{di}{dt} + v_{bemf}$$

# SERVO TRAINER ANALYSIS



- System Transfer function

$$\omega(s) = \frac{k_m v(s)}{(sI + b)(sL + R) + k_m^2} - \frac{k_l(R + sL)v_l(s)}{(sI + b)(sL + R) + k_m^2}$$

- Assume small inductance  $L = 0$  and big inertia flywheel

$$\omega(s) = \frac{k'_m v(s)}{Ts + 1} - \frac{k'_l v_l(s)}{Ts + 1}$$

$$T = \frac{IR}{bR + k_m^2}$$

$$k'_m = \frac{k_m}{bR + k_m^2}$$

$$k'_l = \frac{k_l R}{bR + k_m^2}$$

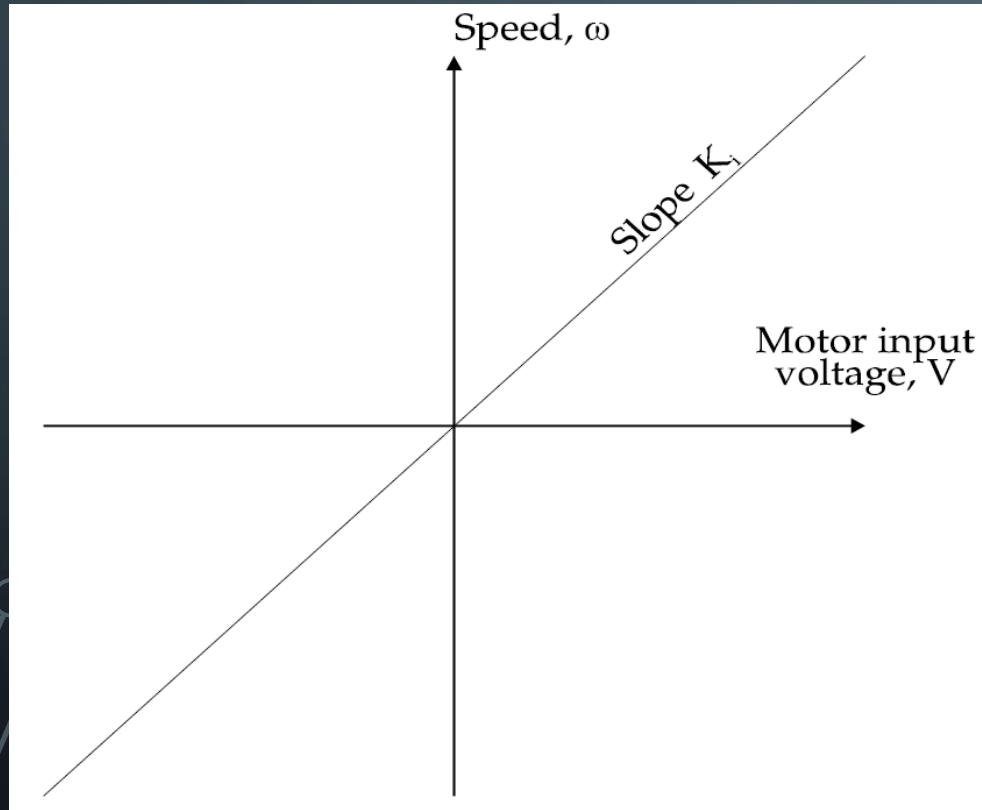
- Assume only inertia load

$$\omega(s) = \frac{k'_m}{Ts + 1} v(s)$$

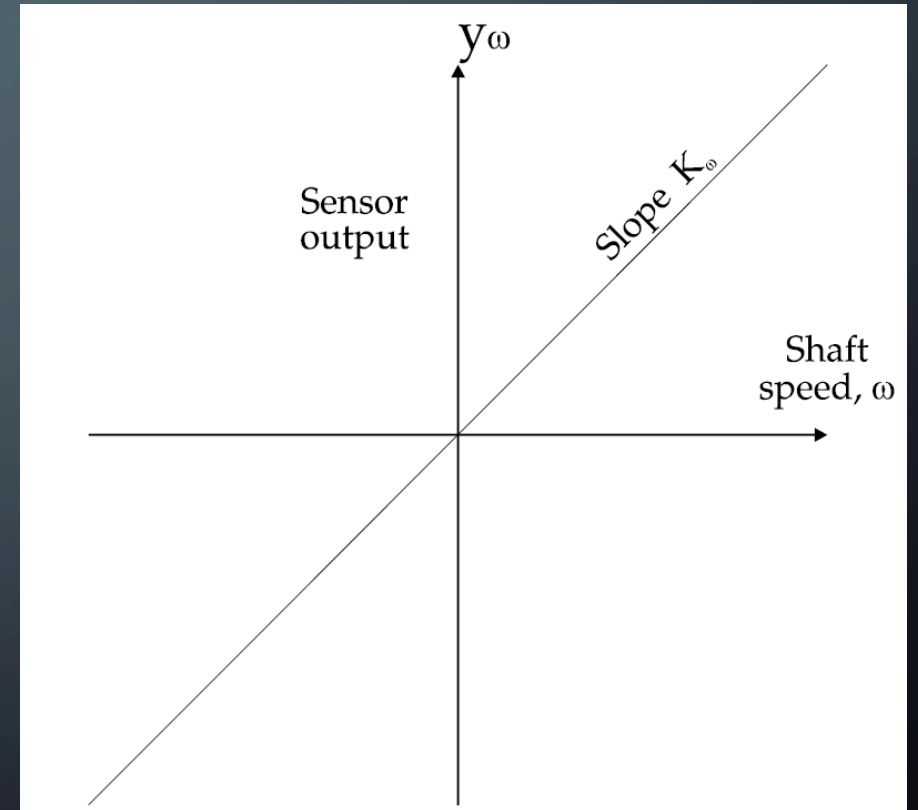
# SERVO TRAINER ANALYSIS

$$\omega(s) = \frac{k'_m}{T_s + 1} v(s)$$

$$\omega = k_i v$$

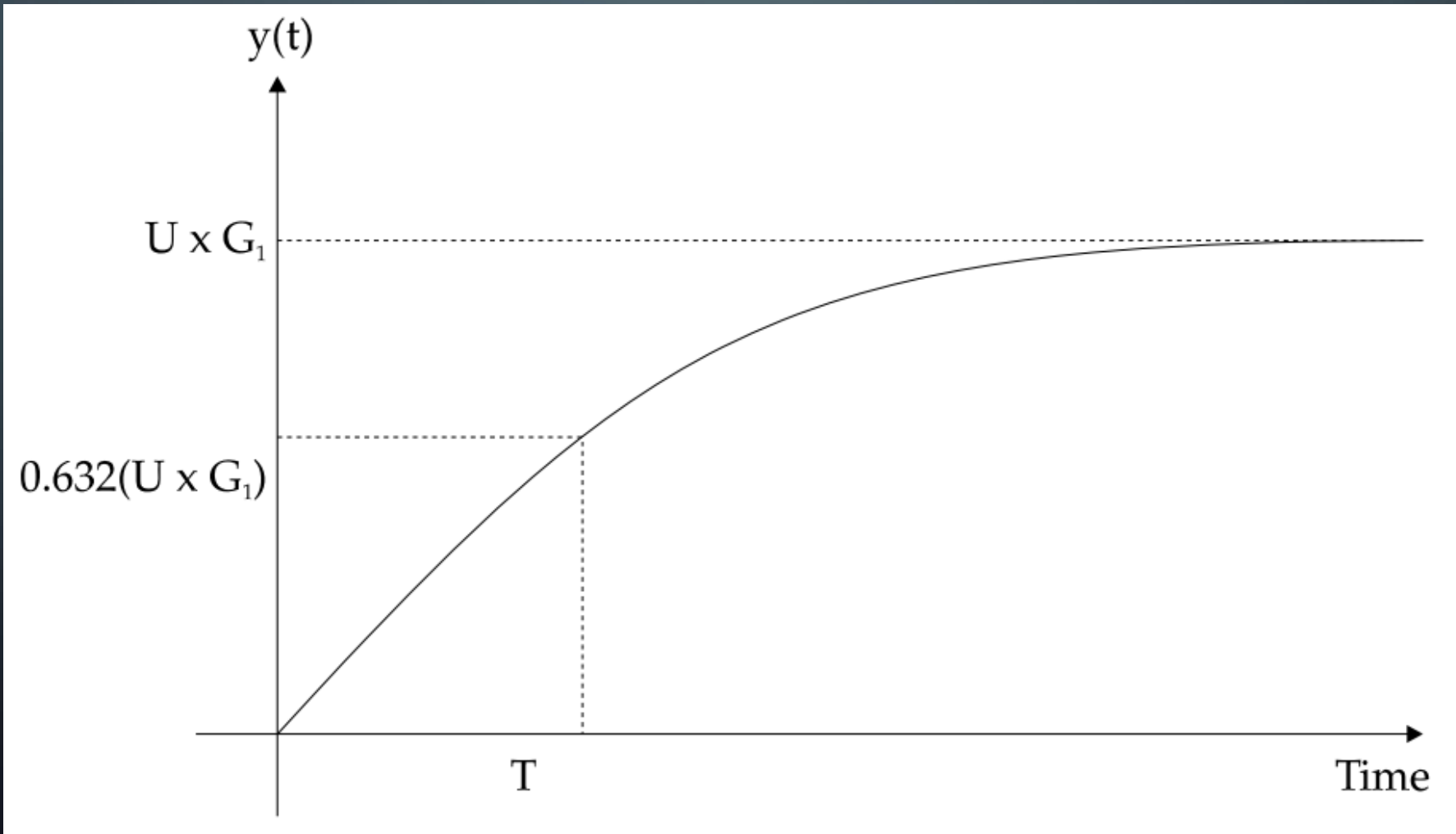


$$y_\omega = k_\omega \omega$$



$$y_\omega(s) = \frac{G_1}{T_s + 1} v(s)$$

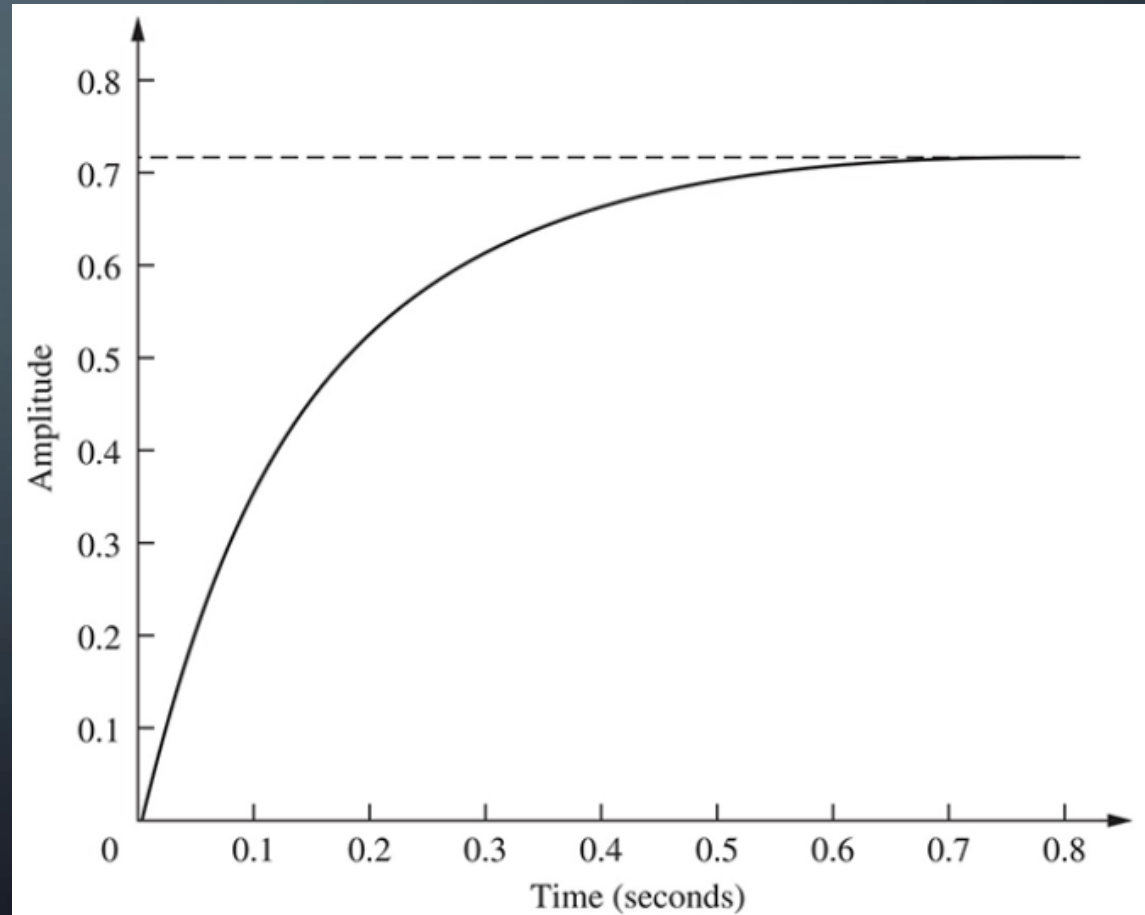
# MEASUREMENT OF SYSTEM CHARACTERISTICS



# FIRST-ORDER TRANSFER FUNCTIONS VIA TESTING

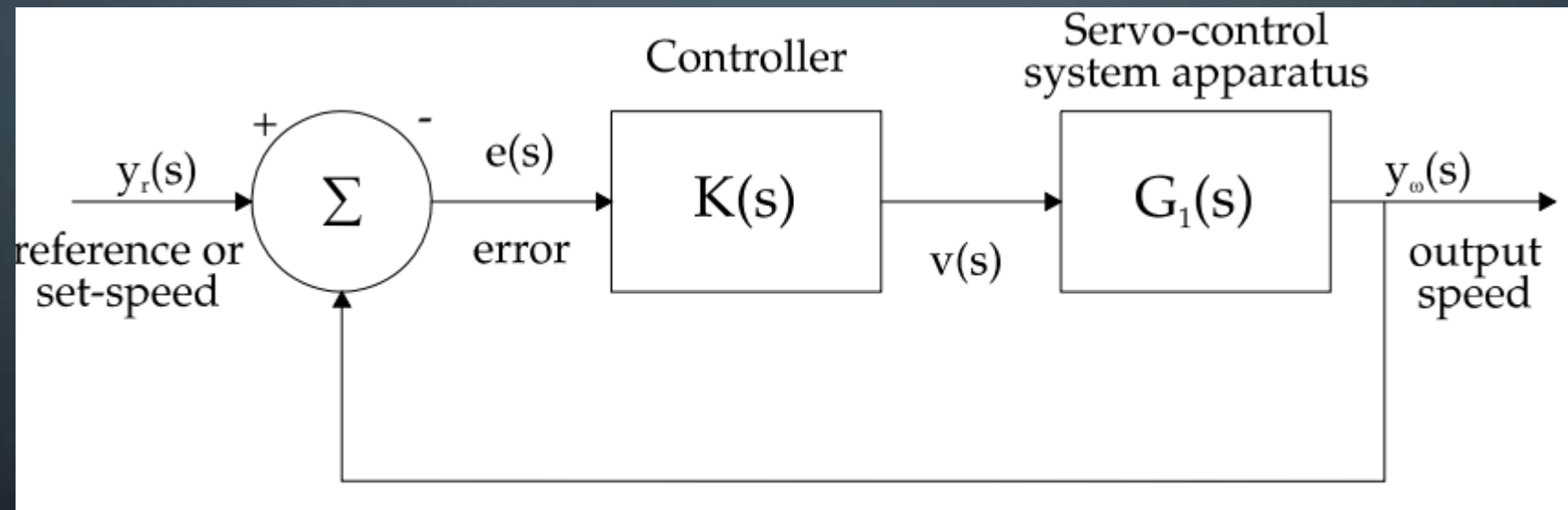
$$C(s) = \frac{K}{s(s+a)} = \frac{K/a}{s} - \frac{K/a}{(s+a)}$$

- Final value: 0.72
- Time constant
  - $0.63 \times 0.72 = 0.45$ , or
  - 0.13 second
  - $\alpha = 1/0.13 = 7.7$
  - $K/\alpha = 0.72$
  - $K = 5.54$
- $G(s) = 5/(s+7)$





# VELOCITY CONTROL SYSTEM



# STEADY STATE ERROR ANALYSIS

- Output and error

$$y_w(s) = \frac{K(s)G_1(s)y_r(s)}{1 + K(s)G_1(s)}$$

$$e(s) = \frac{y_r(s)}{1 + K(s)G_1(s)}$$

- Final value theorem

$$\lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s e(s)$$

# STEADY STATE ERROR ANALYSIS

- Steady state error

$$e_{ss}(s) = \lim_{s \rightarrow 0} \frac{y_r(s)}{1 + K(s)G_1(s)}$$

- Proportional control

$$K(s) = K_p$$

$$e_{ss}(s) = \lim_{s \rightarrow 0} \frac{s y_r(s)}{1 + K_p G_1(s)}$$

- Proportional integral control

$$K(s) = K_p + \frac{K_i}{s}$$

$$e_{ss}(s) = \lim_{s \rightarrow 0} \frac{s y_r(s)}{1 + (K_p s + K_i) G_1(s)}$$

# DYNAMIC RESPONSE ANALYSIS

- Dynamic response

$$y_{\omega}(s) = \frac{G_1}{Ts + 1} v(s)$$

$$y_{\omega}(s) = \frac{K(s)G_1(s)y_r(s)}{1 + K(s)G_1(s)}$$

- Proportional control

$$y_{\omega}(s) = \frac{k_p G_1}{Ts + 1 + k_p G_1} y_r(s)$$

- Proportional integral control

$$y_{\omega}(s) = \frac{(k_p s + k_i)G_1}{Ts^2 + (k_p G_1 + 1)s + k_i G_1} y_r(s)$$