

Wants

→ slow rpm, 3 phase, bldc motor.

Have

→ high rpm, bnf sensing, positing arduino,  
performs well.

Dif

→ assume with low rpm that BEMF will  
be too noise.

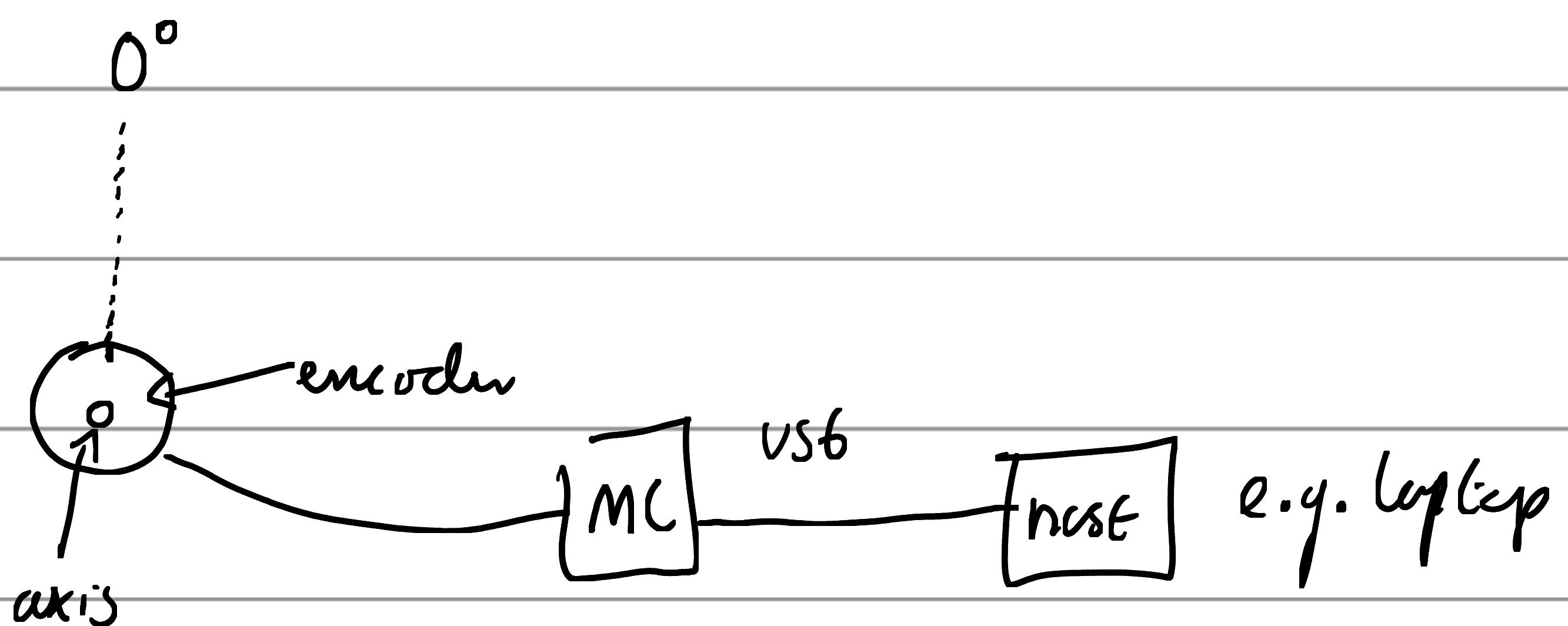
→ Arduino works bnf competuly

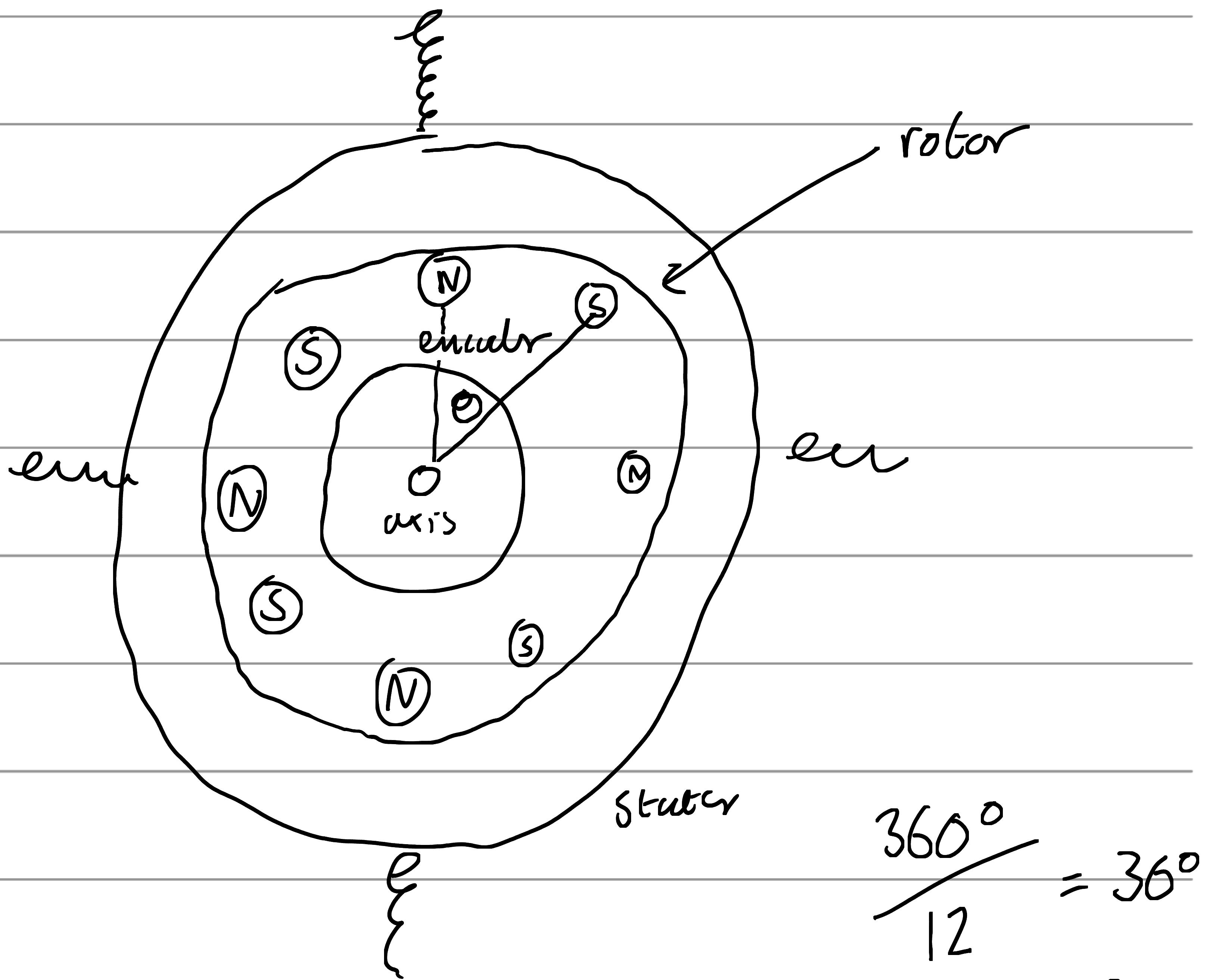
→ Teensy 4.0 work bnf marginal, 100/255 duty

What we should consider.

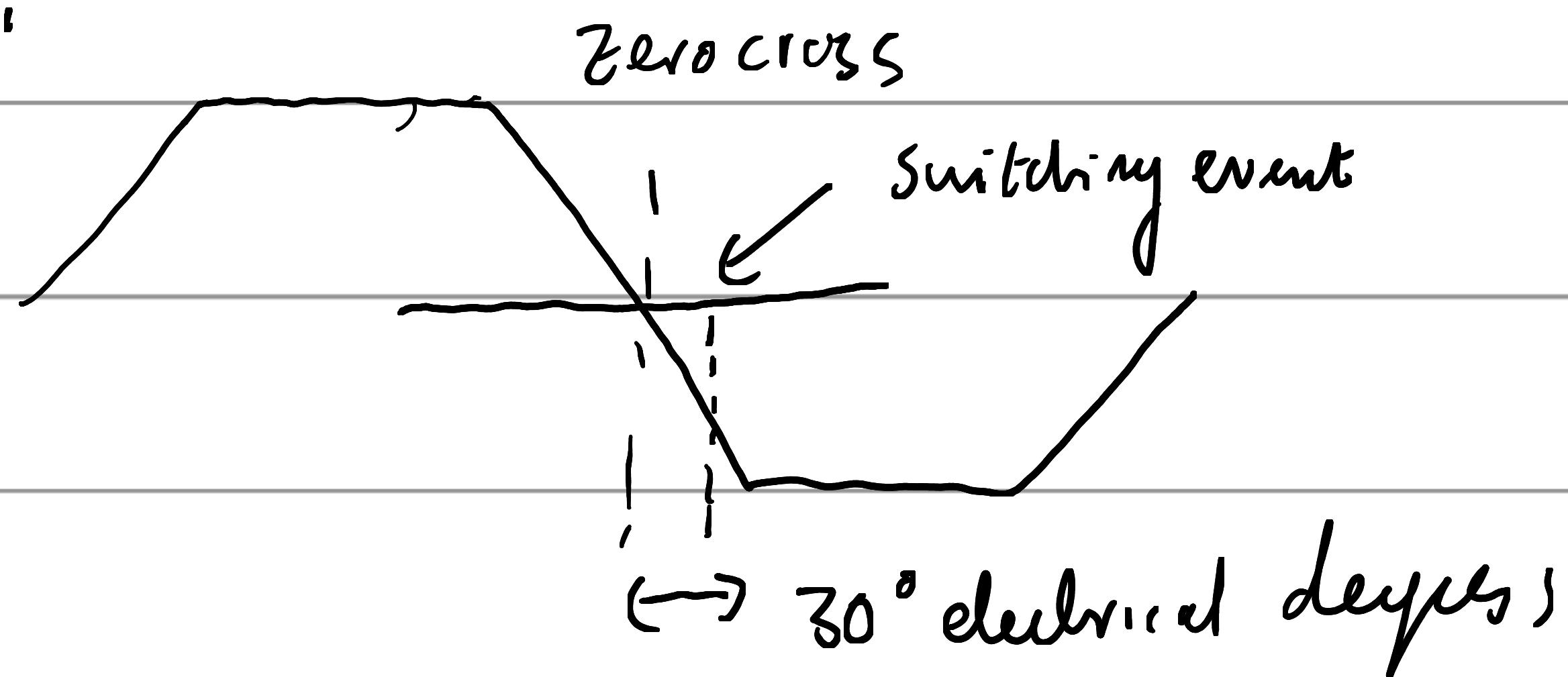
→ BEMF strategy will not work for low rpm, therefore  
we will not know position information reliably.

→ Rotary encoder (absolute)





Phase X:



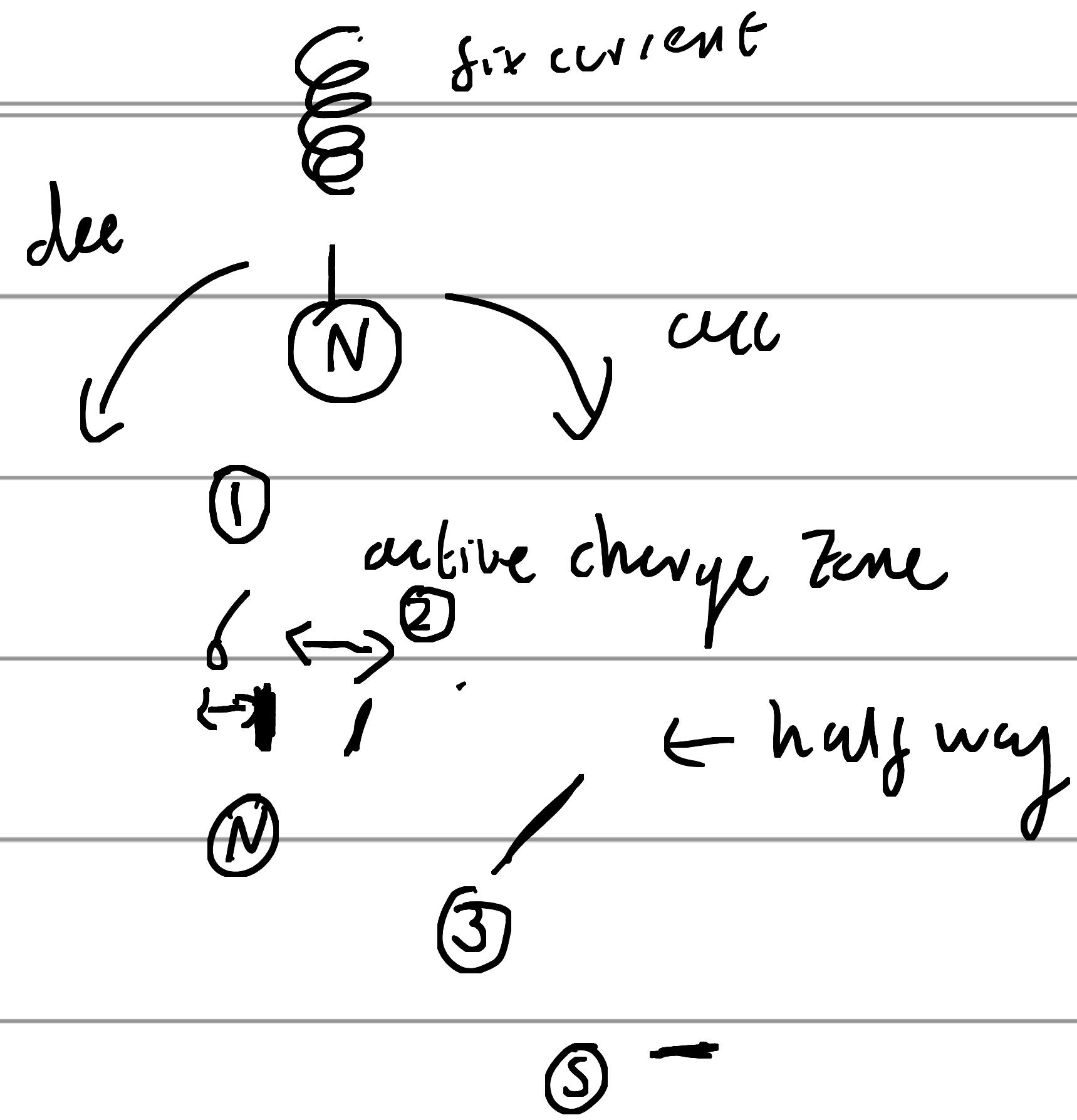
Bldc 6 electrical steps per electrical revolution

I guess you will have n - electrical cycles per mechanical cycle.

$$\text{Mechanical steps per mechanical cycle} = \frac{N \text{ poles}}{2} \cdot 6$$

$$\frac{12}{2} \cdot 6 = 36$$

9 coils      12 poles



Step 1 charge coils, Step 2 discharge coils, step 3  
charge cancellation step A  $\rightarrow$  B

processsing that we charge the relevant coil  
when the encoder has indicated precession beyond the  
central point of the pole (above max flux), then  
we have an interrupt each time the encoder senses a  
rotational displacement change, such that we disable the  
coil charge after some time well before we enter the  
next zone

swap phase, + charge coil, wait for  
displacement discharge coil. [for each mechanical  
step.]

Duty cycle resolution, (impulse of a single pulse of a single electrical/mechanical step)

Arduino

256

reset

Tensy  
40

256, 512, 1024, 8192  
2048

U096

8192

problems

z

switching  
enz

driving  
H bridge

(X)

Positional encoder

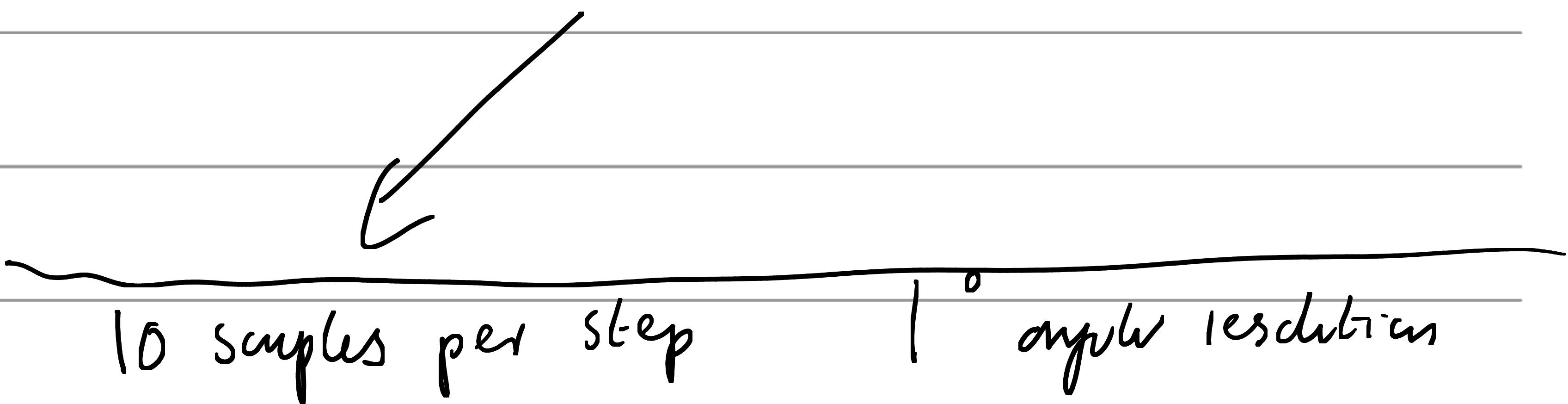
might still be an issue

removes this problem

$360^\circ \rightarrow 36 \text{ steps}$

$10^\circ \text{ per pr step.}$

↑  
mechanical



maybe we want 100 samples per step

$0.1^\circ$

what we want from an rotary encoder

→ absolute position

→ radial resolution ( $0.1 \rightarrow 1^\circ$ )

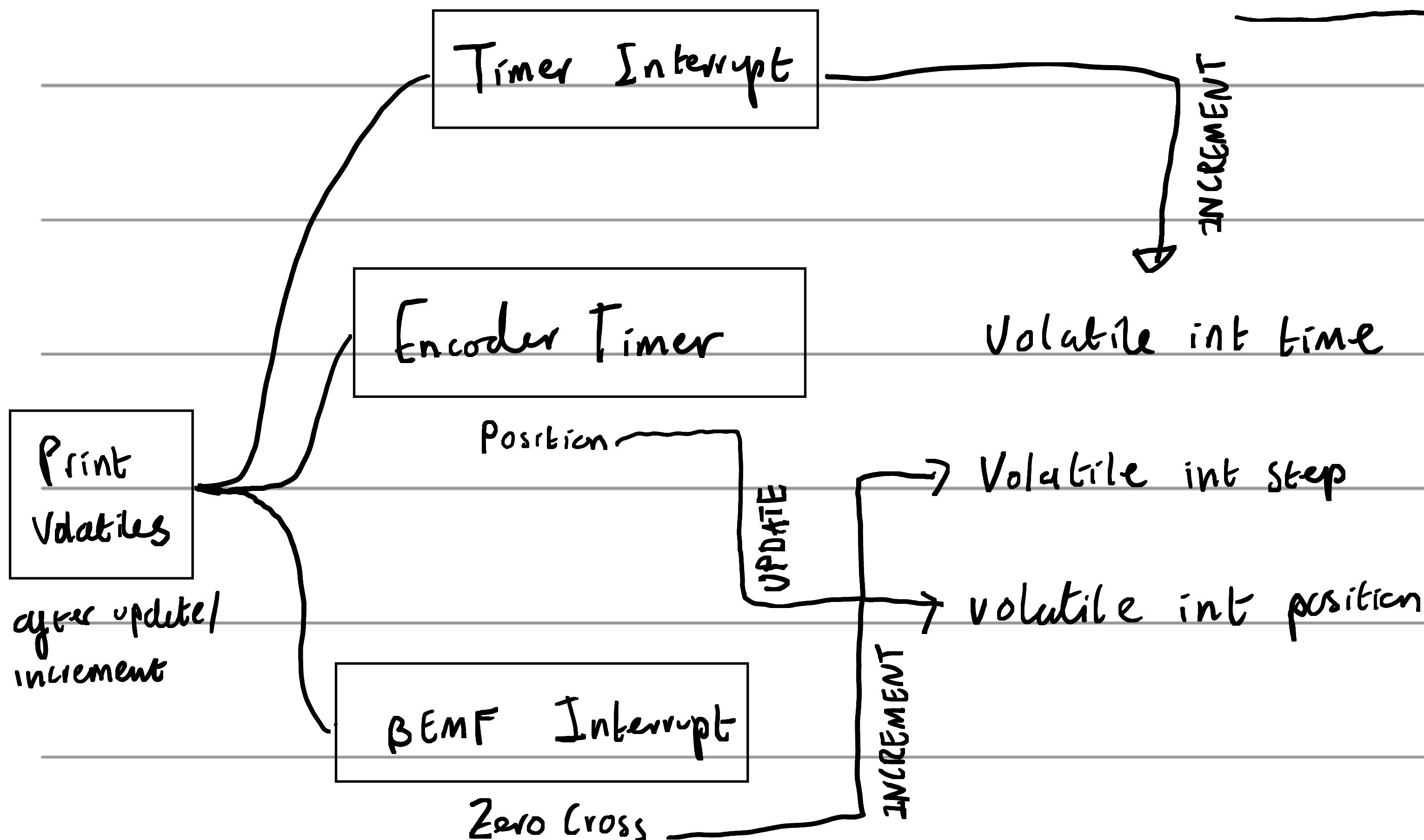
Nice to have

optical (no dust), no interfere bearing, arduino + teensy 40 support.



# CALIBRATION

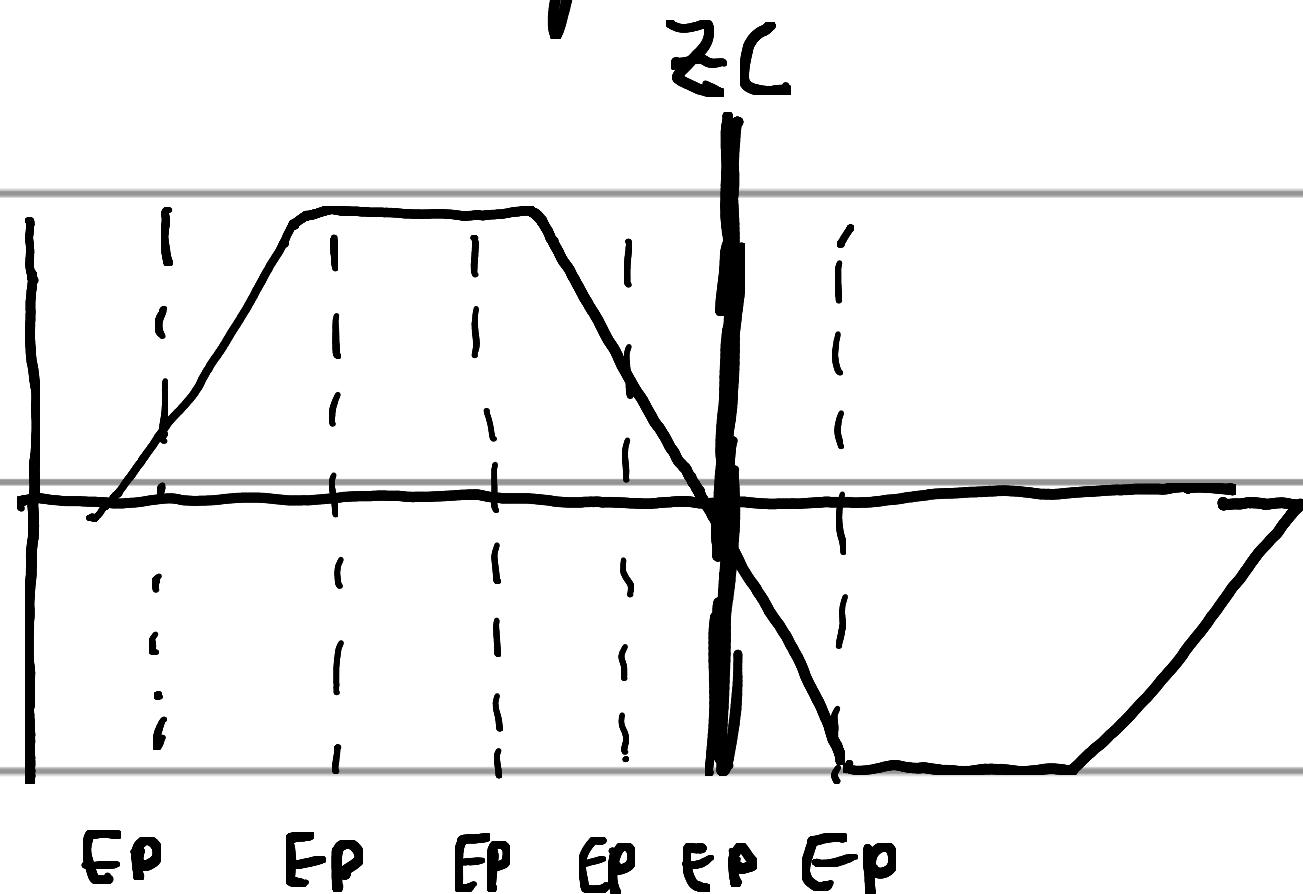
18 / 5 / 22



The idea is to drive the assembly to a minimal speed

such that we get clear Bemf signals while all the time

logging the Bemf step, the rotational position via encoder  
values, and maybe the time.



EP - encoder pulse

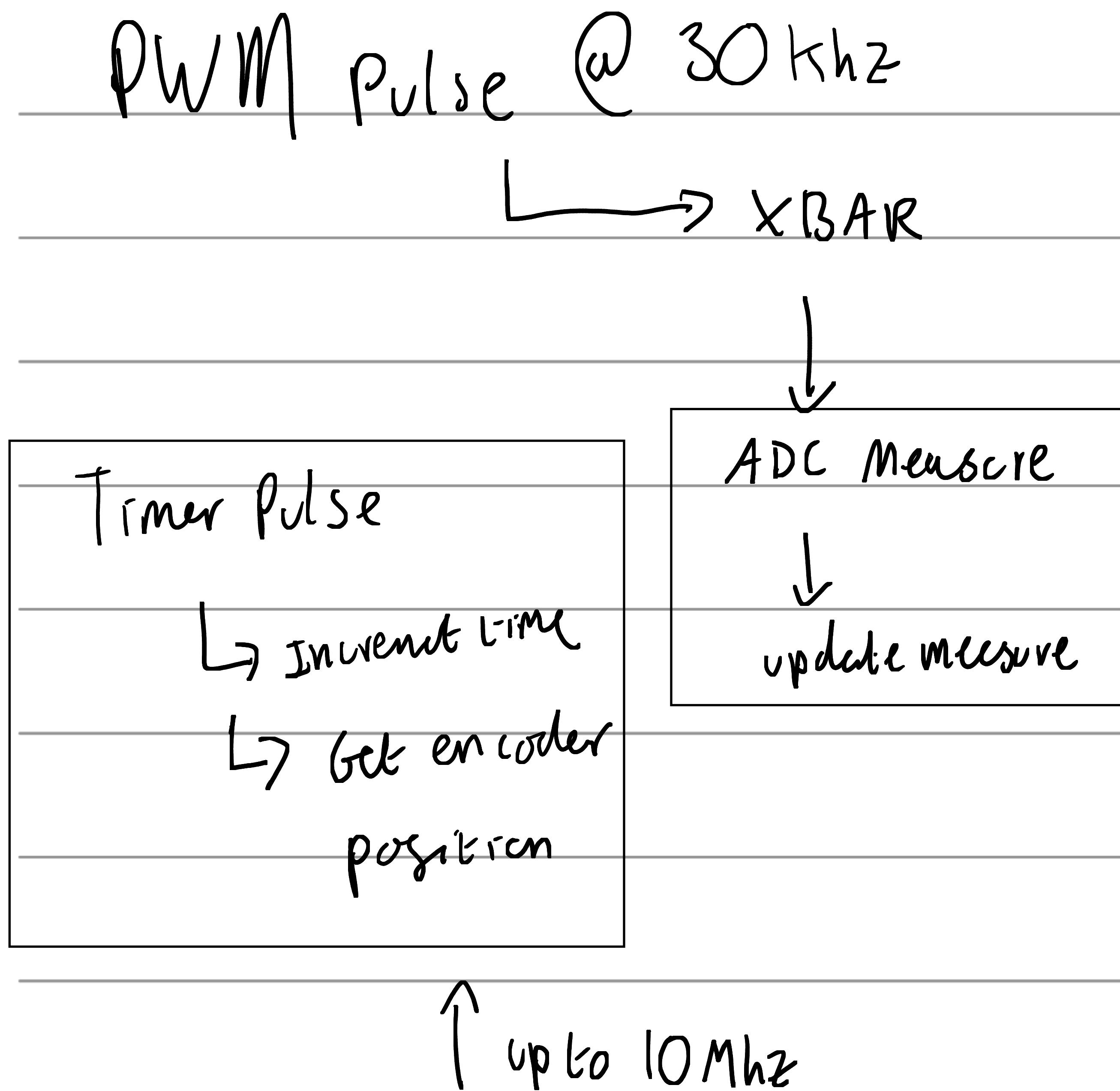
ZC - zero cross

We can find the data via the encoder value the position

I therefore get an average error for the zero-cross angular positions. Ideal switching points are  $30^\circ$  (electrical) after the ZC positions.

One could use the ADC to confirm this.

The way this could work:



21/5/22

Chosen the ADC method as the primary method.

Found 2 good magnetic encoders: Both 3.3 → 5V

→ AS5600 12-bit I<sup>2</sup>C Not for high RPM



x0a6

→ AS5147P 14-bit SPI 28k rpm



↓  
16,384



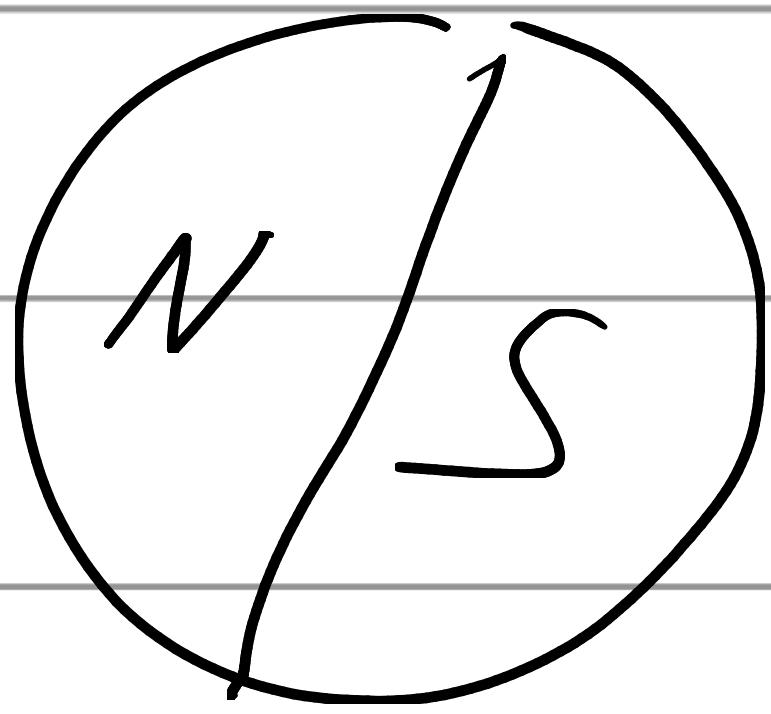
This one is dedicated to motor control

applications like bldc.

AS5147P

With 16,384 angular steps, 9 coils 12 poles thus 36 steps.

$$\frac{16,384}{36} = 455.11 \text{ samples per zone.}$$



$2^{12}$  4096

3601

4096

6.088°

$0.2 \rightarrow 0.7$

optical

## Jerk model

We will be making position & time measurements

$$P_i = [S_i, T_i]$$

$S$  - displacement angular

Euler method

$T$  - time interval

$$\omega_{i+1} = \frac{S_{i+1} - S_i}{T_{i+1} - T_i}$$

$$\omega_{i+2} = \frac{S_{i+2} - S_{i+1}}{T_{i+2} - T_{i+1}}$$

$$\alpha_{i+2} = \dot{\omega}_{i+2} = \frac{\omega_{i+2} - \omega_{i+1}}{T_{i+2} - T_{i+1}}$$

$$= \frac{1}{T_{i+2} - T_{i+1}} \left( \frac{S_{i+2} - S_{i+1}}{T_{i+2} - T_{i+1}} - \frac{S_{i+1} - S_i}{T_{i+1} - T_i} \right)$$

$$= \frac{1}{a-b} \left( \frac{c-d}{a-b} - \frac{d-e}{b-f} \right)$$

$$\Rightarrow \frac{c-d}{(a-b)^2} - \frac{d-e}{(a-b)(b-f)} = \frac{S_{i+2} - S_{i+1}}{(T_{i+2} - T_{i+1})^2} - \frac{S_{i+1} - S_i}{(T_{i+2} - T_{i+1})(T_i - T_{i+1})}$$

What are we measuring  $\dot{\theta}_i$ ,  $\ddot{\theta}_i$ ,  $\dddot{\theta}_i(t)$

$$\omega_i = \dot{\theta}_i = \frac{d\theta_i}{dT_i} \quad \text{angular velocity}$$

$$\dot{\omega}_i = \ddot{\theta}_i = \frac{d\omega_i}{dT_i} = \frac{d^2\theta_i}{dt^2} \quad \text{angular acceleration}$$

$$\dot{\ddot{\theta}}_i = \dddot{\theta}_i = \frac{d\dot{\omega}_i}{dT_i} = \frac{d^3\theta_i}{dt^3} \quad \text{angular jerk.}$$

Jerk model  $X_i = [\overset{\circ}{\theta}, \overset{\circ\circ}{\theta}, \overset{\circ\circ\circ}{\theta}, \overset{\circ\circ\circ\circ}{\theta}]$  in 1D

White noise  $w(t)$  applying to jerk

$$X = Ax_i + Bw(t) \quad a \text{ is a constant}$$

$$\frac{d}{dt} \begin{bmatrix} \overset{\circ}{\theta} \\ \overset{\circ\circ}{\theta} \\ \overset{\circ\circ\circ}{\theta} \\ \overset{\circ\circ\circ\circ}{\theta} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & -a \end{bmatrix} \begin{bmatrix} \overset{\circ}{\theta} \\ \overset{\circ\circ}{\theta} \\ \overset{\circ\circ\circ}{\theta} \\ \overset{\circ\circ\circ\circ}{\theta} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix} w(t)$$

$$\frac{d}{dt} \begin{bmatrix} \theta \\ \dot{\theta} \\ \ddot{\theta} \\ \vdots \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & -a \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta \\ \dot{\theta} \\ \ddot{\theta} \\ \vdots \\ \ddot{\theta} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} w(t)$$

$$\frac{d\theta}{dt} = \ddot{\theta}$$

$$\frac{d\dot{\theta}}{dt} = \dddot{\theta}$$

$$\frac{d\ddot{\theta}}{dt} = \ddot{\theta}$$

$$\frac{d\ddot{\theta}}{dt} = -\ddot{\theta} + w(t)$$

Measurement vector  $Z(k+1) = H X(k+1) + V(k+1)$

$H$  is the system measurement or observation matrix

$V$  is the measurement noise vector.

1D tracking

$$X(k+1) = F(k+1, k) X(k) + u(k)$$

$$F(k+1, k) = e^{\frac{A}{\Delta t} (t_{k+1}, t_k)}$$

$U(k)$  is the discrete white noise

$$F(t) = \begin{bmatrix} 1 & T & T^2/2 & p_1 \\ 0 & 1 & T & q_1 \\ 0 & 0 & 1 & r_1 \\ 0 & 0 & 0 & s_1 \end{bmatrix}$$

$$T = t_{k+1} - t_k$$

$$p_1 = (2 - 2\alpha T + \alpha^2 T^2 - 2e^{-\alpha T}) / 2\alpha^3$$

$$q_1 = (e^{-\alpha T} - 1 + \alpha T) / \alpha^2$$

$$r_1 = (1 - e^{-\alpha T}) / \alpha$$

$$s_1 \sim e^{-\alpha T} = 1 - \alpha T + \frac{\alpha^2 T^2}{2!} - \frac{\alpha^3 T^3}{3!} \dots$$

where  $\alpha T$  is small

$$\lim_{\alpha \rightarrow 0} F(T) = \begin{bmatrix} 1 & T & T^2/2 & T^3/6 \\ 0 & 1 & T & T^2/2 \\ 0 & 0 & 1 & T \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

# Kalman algorithm

## Prediction

- 1) Predict next frame

$$X_t = f(X_{t-1}, U_t) + q_1$$

- 2) Calculate prediction covariance matrix

$$Z_t = h(X_t, U_t) + q_2$$

## Correction

- 3) calculate EKF gain

$$K_t = P_{t|t-1} H_t^T \left[ H_t P_{t|t-1} \circ H_t^T + R_t \right]^{-1}$$

Q4) Update the observation estimated value

$$X_t = X_{t|t-1} + K_t [Z_t - Z_{t|t-1}]$$

S) update error in covariance matrix

$$P_{t|t} = [Z_t - Z_{t|t-1} - K_t H_t] P_{t|t-1}$$

✓ ✓ ✓ ? Q? ✓ ✓

We need f, h, Z, K, X+, H, P

Initialisation, 3 measurements to estimate  $\hat{\theta}_1, \hat{\theta}_2, \hat{\theta}_3$

$\hat{\theta}$  taken to be zero initially.

$$\hat{\theta}_3 = M_x(3) \quad \hat{\theta}_3 = \frac{M_{xc}(3) - M_x(2)}{\tau}$$

$$\hat{\theta}_3 = \left[ \frac{M_{xc}(3) - 2M_x(2) + M_x(1)}{\tau^2} \right] \quad \hat{\theta}_3 = 0$$

constant time interval  $\tau$

## Measurement noise

$$M_x(1) = \bar{x}(1) + v(1)$$

$$M_x(2) = \bar{x}(2) + v(2)$$

$$M_x(3) = \bar{x}(3) + v(3)$$

Error in  $\overset{\circ}{\theta}, \overset{\circ}{\theta}, \overset{\circ}{\theta}, \overset{\circ}{\theta}, \overset{\circ}{\theta}$        $\epsilon_1, \epsilon_2, \epsilon_3, \epsilon_4$

$$\begin{aligned}\epsilon_1(3|3) &= \bar{x}(3) - \hat{x}(3|3) = \bar{x}(3) - \bar{x}(3) - v(3) \\ &= -v(3)\end{aligned}$$

$$\epsilon_2(3|3) = \dot{\bar{x}}(3) - \dot{\hat{x}}(3|3) =$$

## Euler's

$$X_j^o(k+1) = F_j X_j(k) + w_j(k)$$

$$Z_j(k+1) = H_j X_j(k+1) + v_j(k+1)$$

$$X_j = \begin{bmatrix} x & \dot{x} & \ddot{x} & \cdots \end{bmatrix}^T$$

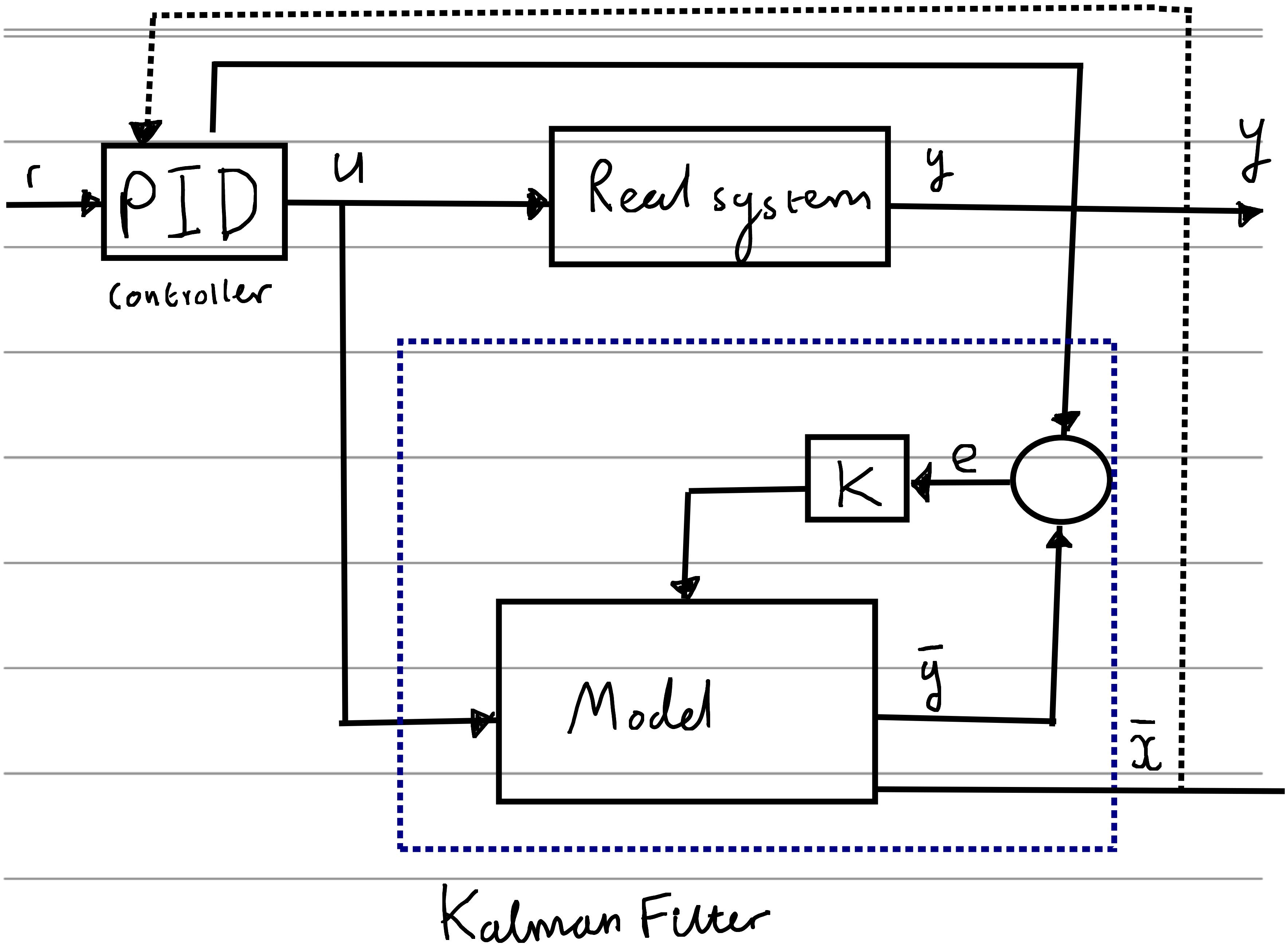
$$w_j = [u_{1j} \ u_{2j} \ u_{3j} \ u_{4j}]^T$$

$$H_j = [1 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0]$$

$$Z_j = [M_{dc}]^T$$

$$v_j = [\text{noise } n_x]^T$$

$$F_j = \begin{bmatrix} 1 & T & T^2/2 & P_1 \\ 0 & 1 & T & Q_1 \\ 0 & 0 & 1 & R_1 \\ 0 & 0 & 0 & S_1 \end{bmatrix}$$



Kalman Filter