## Cryptanalysis

#### **Error Correcting Codes**

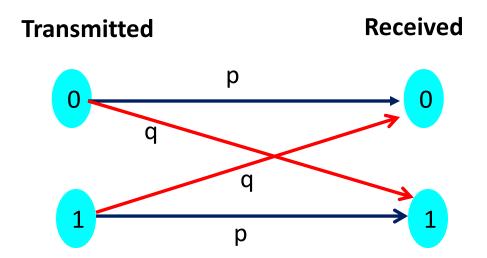
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## Binary symmetric channel (BSC)

 Each bit transmitted has an independent chance of being received correctly with probability p and incorrectly received with probability q=1-p.



 Can we transmit m bits more reliably over this channel if we have spare bandwidth?

#### **Error Detection**

- Suppose we want to transmit 7 bits with very high confidence over a binary symmetric channel. Even if p>.99, we occasionally will make a mistake.
- We can add an eight bit, a check sum, which makes any valid eight bit message have an even number of 1's.
- We can thus detect a single bit transmission error. Now the probability of a relying on a "bad" message is P<sub>error</sub>=1-(p<sup>8</sup>+8p<sup>7</sup>(1-p)) instead of P<sub>error</sub>=1-p<sup>8</sup>. If p=.99, P<sub>error</sub> drops from about 7% to .3%.
- This allows us to detect an error and hopefully have the transmitter resend the garbled packet.
- Suppose we want to avoid retransmission?

#### **Error Correction**

We can turn these "parity checks" which enable error detection to error correction codes as follows. Suppose we want to transmit b<sub>1</sub>b<sub>2</sub>b<sub>3</sub>b<sub>4</sub>.
 Arrange the bits in a 2 x 2 rectangle:

b <sub>1</sub>	b <sub>2</sub>	$c_1 = b_1 + b_2$
$b_3$	b <sub>4</sub>	c <sub>2</sub> =b <sub>3</sub> +b <sub>4</sub>
$c_3 = b_1 + b_3$	$c_4 = b_2 + b_4$	$c_5 = b_1 + b_2 + b_3 + b_4$

- We transmit  $b_1b_2b_3b_4c_1c_2c_3c_4c_5$ .
- The receiver can detect any single error and locate its position.
- Another simple "encoding scheme" that corrects errors is the following. We can transmit each bit three times and interpret the transmission as the majority vote. Now the chance of correct reception is P<sub>correct</sub>=p<sup>3</sup>+3p<sup>2</sup>q>p and the chance of error is P<sub>error</sub>=3pq<sup>2</sup>+q<sup>3</sup><q. For p=.99, P<sub>error</sub>= 0.000298 and P<sub>correct</sub>= .999702.

#### Codewords and Hamming distance

- To correct errors in a message "block," we increase the number of bits transmitted per block. The systematic scheme to do this is called a code, C.
- If there are M valid messages per block (often M=2<sup>m</sup>) and we transmit n>lg(M) bits per block, the M "valid" messages are spread throughout the space of 2<sup>n</sup> elements.
- If there are no errors in transmission, we can verify the message is equal to a codeword with high probability.
- If there are errors in the message, we decode the message as the codeword that is "closest" (i.e.-differs by the fewest bits) from the received message.
- The number of differences between the two nearest codewords is called the distance of the code or d(C).

## Hamming distance

- The best decoding strategy is to decode a message as the codeword that differs least from a codeword. So, for a coding scheme, C, if d(C)=2t+1 or less bits, we can correct t or less errors per block.
- If d(C)=s+1, we can detect s or fewer errors.
- The Hamming distance, denoted Dist(v, w), between two elements v, wεGF(2)<sup>n</sup> is the number of bits they differ by. The Hamming distance satisfies the usual conditions for a metric on a space.
- The Hamming weight of a vector **v**εGF(2)<sup>n</sup>, denoted, ||v|| is the number of 1's.
- If v, wεGF(2)<sup>n</sup>, Dist(v, w)= ||v⊕w||.

#### Definition of a Code

- In the case of the "repeat three times" code, C<sub>repeatx3</sub>, M=1 and n=3. There are two "codewords," namely 111 and 000. d(C<sub>repeatx3</sub>)=3, so d=2t+1 with t=1.
- In general, a C(n,M,d) denotes a code in GF(2)<sup>n</sup> with M codewords with d(C)=d the minimum distance, n is dimension.
- As discussed, such codes can correctly decode transmissions containing t errors or less.
- The rate of the code is (naturally) R=lg(M)/n.
- Error correcting codes strive to find "high rate" codes that can efficiently encode and decode messages with acceptable error.

## Example rates and errors

Code	n	M	d	R	p <sub>1</sub>	p <sub>2</sub>	P <sub>1,e</sub>	$P_{2,e}$
Repetition x 3	3	2	3	1/3	3/4	7/8	0.156	0.043
Repetition x 5	5	2	5	1/5	3/4	7/8	0.103	0.016
Repetition x 7	7	2	7	1/7	3/4	7/8	0.071	0.006
Repetition x 9	9	2	9	1/9	3/4	7/8	0.049	0.004
Hamming(7,4)	7	16	3	4/7	3/4	7/8	0.556	0.215
Golay(24,12,8)	24	4096	17	1/2	3/4	7/8		
Hadamard (64,32,16)	64	32	16	3/16	3/4	7/8		
RM(4,2)	16	11	4					
BCH[7,3,4]	7	8	4	3/7				

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#### Shannon

- Source Coding Theorem: The n random variables can be encoded by nH bits with negligible information loss.
- Channel Capacity:  $C = \max_{P(x)}(H(I|O)-H(I))$ . For a DMC, BSC with error rate p, this implies  $C_{BSC}(p) = 1 + plg(p) + q lg(q)$ . So for BSC R= 1-H(P).
- Channel Coding Theorem: For all R<C<sub>max</sub>, ε>0, C(n,M,d) of length n with M codewords: M≥2<sup>[Rn]</sup> and P<sup>(i)</sup><sub>error</sub>≤ε for i=1,2,...,M.
- Translation: Good codes exist that permit transmission near the channel capacity with arbitrarily small error.

#### The Problem of Coding Theory

- Despite Shannon's fundamental results, this is not the end of the coding problem!
  - Shannon's proof involved random codes
  - Finding the closest codeword to a random point is the shortest vector problem, so "closest codeword" decoding is computationally difficult. Codes must be systematic to be useful.
  - The Encoding Problem: Given an m bit message, m, compute the codeword, t (for transmitted), in C(n,M,d).
  - The Decoding Problem: Given an n bit received word, r=t+e, where
     e was the error, compute the codeword in C(n,M,d) closest to r.
  - General codes are hard to decode

#### **Bursts**

- Bursty error correction: Errors tend to be "bursty" in real communications.
- Burst error correcting codes can be constructed by "spreading out codewords". Let cw<sub>i</sub>[j] mean bit j of codeword i. Transmit cw<sub>1</sub>[1], cw<sub>2</sub>[1],..., cw<sub>k</sub>[1], cw<sub>1</sub>[2],... where k is the size of a "long" error.
- Some specific codes (RS, for example) are good at bursty error correction.

# Channel capacity for Binary Symmetric Channel

- Discrete memory-less channel: Errors independent and identically distributed according to channel error rate. (No memory).
- Rate for code, R<sub>C</sub>= Ig(M)/n.
- Channel capacity intuition: How many bits can be reliably transmitted over a BSC?
  - The channel capacity, c, of a channel is c= sup<sub>X</sub>
     I(X;Y), where X is the transmission distribution and Y is the reception probability
  - Shannon-Hartley: c= Blg(1+S/N), B is the bandwidth,
     S is the signal power and N is the noise power.
  - Information rate, R=rH.

## How much information can be transmitted over a BSC with low error?

- How many bits can be reliably transmitted over a BSC?
   Answer (roughly): The number of bits of bandwidth minus the noise introduced by errors.
- Shannon's channel coding theorem tells us we can reliably transmit up to the channel capacity.
- However, good codes are hard to find and generally computationally expensive.

#### Calculating rates and channel capacity

- For single bit BSC, C=1+plg(p)+qlg(q).
- Recall c= sup<sub>X</sub> I(X;Y).
- The distribution P(X=0)=P(X=1)=1/2 maxmizes this.
- c = 1/2 + 1/2 + plg(p) + qlg(q)

#### **Linear Codes**

- A [n,k,d] linear code is an k-subspace of an n-space over F (usually GF(2)) with minimum distance d.
  - An [n,k,d] code is also a (n, 2<sup>k</sup>,d) code
- Standard form for generator is G= (I<sub>k</sub>|A) with k message bits, n codeword bits. Codeword c=mG.
- For a linear code, d=min<sub>u≠0, uεC</sub> {wt(u)}.
  - Proof: Since C is linear, dist(u, w)= dist(u-w,0)=wt(u-v). Since the code is linear, u-v ε C. That does it.
- Parity check matrix is H: v ε C iff vH<sup>T</sup>=0.
- If G is in standard form, H=[-A<sup>T</sup>|I<sub>n-k</sub>]. Note that GH=0.
- Example: Repetition code is the subspace in GF(2)<sup>3</sup> generated by (1,1,1).

## G and H and decoding

- Let r=c+e, where r is the received word, c is the transmitted word and e is the error added by the channel.
- Note codewords are linear combinations of rows of G and rH<sup>T</sup>=cH<sup>T</sup>+eH<sup>T</sup>=eH<sup>T</sup>.
- Coset leader table

Minimum weight

miniminani won	9116					
Coset leader					<b>Error</b>	<b>Syndrone</b>
<b>c</b> <sub>1</sub>	$\mathbf{c_2}$	$c_3$		$\mathbf{c}_{M}$	0	<b>0=0</b> H <sup>⊤</sup>
<b>c</b> <sub>1</sub> + <b>e</b> <sub>1</sub>	c <sub>2</sub> +e <sub>1</sub>	c <sub>3</sub> +e <sub>1</sub>		c <sub>M</sub> +e <sub>1</sub>	<b>e</b> <sub>1</sub>	$\mathbf{e_1}H^{T}$
<b>c</b> <sub>1</sub> + <b>e</b> <sub>2</sub>	$c_2+e_2$	$c_3+e_2$	•••	$c_M + e_2$	$\mathbf{e_2}$	$\mathbf{e_2}H^\intercal$
•••	••••	•••				
c <sub>1</sub> +e <sub>h-1</sub>	$c_2$ + $e_{h-1}$	<b>c</b> <sub>3</sub> + <b>e</b> <sub>h-1</sub>		c <sub>M</sub> +e <sub>h-1</sub>	$\mathbf{e}_{h-1}$	$\mathbf{e}_{h-1}H^{T}$

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## Syndrome and decoding Linear Codes

- $S(r)=rH^T$  is called the syndrone.
- A vector having minimum Hamming weight in a coset is called a coset leader.
- Two vectors belong to the same coset iff they have the same syndrone.
- Now, here's how to systematically decode a linear code:
  - 1. Calculate S(r).
  - 2. Find coset leader, **e**, with syndrone S(**r**).
  - 3. Decode r as r-e.
- This is more efficient than searching for nearest codeword but is only efficient enough for special codes.

## Syndrone decoding example (H[7,4])

- Message: 1 1 0 0.
- Codeword transmitted: 1 1 0 0 0 1 1.
- Received: 1 1 0 0 0 0 1. (Error in 6<sup>th</sup> position)

## Syndrone decoding example (H[7,4])

Coset table (Left)

```
Syn Coset Leader
000
      0000000 1000011 0100101 1100110 0010110 1010101 0110011 1110000
110
      0000001 1000010 0100100 1100111 0010111 1010100 0110010 1110001
101
      0000010 1000001 0100111 1100100 0010100 1010111 0110001 1110010
      011
111
      0001000 1001011 0101101 1101110 0011110 1011101 0111011 1111000
100
      0010000 1010011 0110101 1110110 0000110 1000101 0100011 1100000
010
      001
      1000000 0000011 1100101 0100110 1010110 0010101 1110011 0110000
```

- (1 1 0 0 0 1) H<sup>T</sup>= (0 1 0) which is the syndrone of the seventh row whose coset leader is e= (0 0 0 0 0 1 0).
- Decode message as (1 1 0 0 0 1) + (0 0 0 0 0 1 0) = (1 1 0 0 1 1).

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## Syndrone decoding example (H[7,4])

Coset table (Right)

Syn

#### Bounds: How good can codes be?

- Let A<sub>q</sub>(n, d) denote the largest code with minimum distance d.
- Sphere Packing (Hamming) Bound: If d=2e+1,  $A_q(n,d) \le \sum_{k=0}^{e} {}_{n}C_k(q-1)^k \le q^n$ .
  - Proof: Let I be the number of codewords.  $I(1+(q-1)_nC_1+(q-1)^2_nC_2+...+(q-1)^e_nC_e)$  ≤  $q^n$  because the e-spheres around the codewords are disjoint.
- **GSV Bound:** There is a linear [n, k, d] code satisfying the inequality:  $A_q(n,d) \ge q^n/(1+(q-1)_nC_1+(q-1)^2_nC_2+...+(q-1)^{d-1}_nC_{d-1})$ 
  - Proof: The d-1 columns of the check matrix are linearly independent iff the code has distance d. So  $q^{n-k3}(1+(q-1)_nC_1+(q-1)^2_nC_2+...+(q-1)^{d-1}_nC_{d-1})$
- Singleton Bound: M≤q<sup>n-d+1</sup>, so R≤1-(d-1)/n.
  - Proof: Let C be a (n,M,d) code. Since every codeword differs by at least d-1 positions, q<sup>n-(d-1)</sup>≥M.

#### **MDS**

- Singleton Bound: M≤q<sup>n-d+1</sup>, so R≤1-(d-1)/n.
- Code meeting Singleton bound is an MDS code.
- If L is an MDS code so it L<sup>⊥</sup>.
- If L is an [n,k] code with generator G, L is MDS iff there
  are k linearly independent columns.
- Binary 3-repetition code is an MDS

## Hamming

- A Hamming code is a [n,k,d] linear code with
  - $n = 2^{m} 1$
  - $k = 2^{m} 1 m$
  - d=3.
- To decode **r=c+e**:
  - Calculate S(r)= rH<sup>T</sup>.
  - Find j which is the column of H with the calculated syndrome.
  - Correct position j.

#### [7,4] Hamming code

• The [7,4] code has encoding matrix G, and parity check H where:

```
1000110 1101100
G= 0100101 H= 1011010
0010011 0111001
```

The code words are:

```
0000000, weight: 0
                         0001111, weight: 4
1000011, weight: 3
                         1001100, weight: 3
0100101, weight: 3
                         0101010, weight: 3
1100110, weight: 4
                         1101001, weight: 4
0010110, weight: 3
                         0011001, weight: 3
1010101, weight: 4
                         1011010, weight: 4
0110011, weight: 4
                         0111100, weight: 4
1110000, weight: 3
                         1111111, weight: 7
```

## Decoding Hamming code

- Message: 1100 → 1100011.
- Received as 1100001.
- 1100001  $H^T$ = 010 which is sixth row of  $H^T$ . Error in sixth bit.
- 1100001+0000010= 110011

#### **Dual Code**

- If C is an [n,k] linear code, then C<sup>⊥</sup> = {u: u·c=0, cεC} is an [n, n-k] linear code called the dual code.
- The parity check matrix, H, of a code, C, is the generator of its dual code.
- A code is self-dual if C= C<sup>⊥</sup>.
- Weight enumerator: Let  $A_i$  be the number of codewords in C of weight i, then  $A(z) = \sum_i A_i z^i$  is the weight enumerator.

# Example: dual code of (7,4) Hamming code

#### Codewords:

000000	0111001
1101100	1010101
1011010	1100011
0110110	0001111

#### **Hadamard Code**

 Hadamard Matrix: H H<sup>T</sup>=nI<sub>n</sub>. If H is Hadamard of order m, J=

> H H H -H

is Hadamard of order 2m.

- Hadamard code uses this property. Generator matrix for this code is G= [H|-H]<sup>T</sup>. For message I, 0≦<2<sup>i</sup> send the row corresponding to i.
  - Used on Mariner spacecraft (1969).
- To decode, a 2<sup>i</sup> bit received word, r, compute d<sub>i</sub>= r×R<sub>i</sub>, where R<sub>i</sub> is the 2<sup>i</sup> bit row i.
  - If there are no errors, the correct row will have  $d_i = 2^{i-1}$  and all other rows will have  $d_i = 0$ .
  - If one error,  $d_i = 2^i-2$  (all dot products but 1 will be  $\pm 2$ ), etc.

#### Hadamard Code example

- Let h<sub>ij</sub>= (-1)<sup>a0 b0 + ... + a4 b4</sup>, where a and b index the rows and columns respectively. This gives a 32 times 32 entry matrix, H.
- H(64, 32, 16): 64=26 bit codewords, 6 messages. First 32 rows:

```
00
                                           0000000000000001111111111111111
                                                                              16
01010101010101010101010101010101
                                  01
                                           0101010101010101101010101010101010
                                                                              17
0011001100110011001100110011
                                  02
                                           00110011001100111100110011001100
                                                                              18
0110011001100110011001100110
                                           0110011001100110100110011001
                                  0.3
                                                                              19
00001111000011110000111100001111
                                                                              2.0
                                  0.4
                                           000011110000111111111000011110000
01011010010110100101101001011010
                                  0.5
                                           01011010010110101010101101010101
                                                                              21
                                                                              22
00111100001111000011110000111100
                                  06
                                           00111100001111001100001111000011
01101001011010010110100101101001
                                  07
                                           01101001011010011001011010010110
                                                                              23
                                  0.8
                                                                              2.4
000000001111111110000000011111111
                                           0000000011111111111111111100000000
                                                                              25
010101011010101010101010110101010
                                  09
                                           010101011010101010101010101010101
00110011110011000011001111001100
                                           00110011110011001100110000110011
                                                                              2.6
                                  10
01100110100110010110011010011001
                                  11
                                           01100110100110011001100101100110
                                                                              2.7
000011111111000000001111111110000
                                  12
                                           00001111111100001111000000001111
                                                                              28
010110101010010101011010101010101
                                  13
                                           010110101010010110100101010101010
                                                                              29
00111100110000110011110011000011
                                  14
                                           00111100110000111100001100111100
                                                                              30
01101001100101100110100110010110
                                  15
                                           01101001100101101001011001101001
                                                                              31
```

#### Hadamard Code example

#### Last 32 rows:

```
1111111111111111111111111111111111111
                                  32
                                          48
1010101010101010101010101010101010
                                  33
                                          101010101010101010101010101010101
                                                                             49
1100110011001100110011001100
                                  34
                                          1100110011001100001100110011
                                                                             50
1001100110011001100110011001
                                  35
                                          1001100110011001011001100110
                                                                             51
11110000111100001111000011110000
                                          11110000111100000000111100001111
                                                                             52
                                  36
10100101101001011010010110100101
                                  37
                                          10100101101001010101101001011010
                                                                             53
11000011110000111100001111000011
                                  38
                                          11000011110000110011110000111100
                                                                             54
10010110100101101001011010010110
                                  39
                                          10010110100101100110100101101001
                                                                             55
111111110000000011111111100000000
                                          111111110000000000000000011111111
                                                                             56
                                  40
10101010101010101101010101010101
                                          10101010010101010101010110101010
                                                                             57
                                  41
11001100001100111100110000110011
                                          11001100001100110011001111001100
                                                                             58
                                  42
10011001011001101001100101100110
                                  43
                                          10011001011001100110011010011001
                                                                             59
11110000000111111111000000001111
                                          111100000000111100001111111110000
                                                                             60
                                  44
101001010101101010100101010101010
                                  45
                                          1010010101011010010110101010101
                                                                             61
11000011001111001100001100111100
                                  46
                                          11000011001111000011110011000011
                                                                             62
10010110011010011001011001101001
                                          10010110011010010110100110010110
                                                                             63
```

#### Hadamard Code example

- Suppose received word is:
  - 11001100110011000011001100110001
- Dot product with rows of matrix is:

```
00: 002, 01: 002, 02: -02, 03: -02, 04: -02, 05: -02, 06: 002, 07: 002.
08: -02, 09: -02, 10: 002, 11: 002, 12: 002, 13: 002, 14: -02, 15: -02.
16: -02, 17: -02, 18: -30, 19: 002, 20: 002, 21: 002, 22: -02, 23: -02.
24: 002, 25: 002, 26: -02, 27: -02, 28: -02, 29: -02, 30: 002, 31: 002.
32: -02, 33: -02, 34: 002, 35: 002, 36: 002, 37: 002, 38: -02, 39: -02.
40: 002, 41: 002, 42: -02, 43: -02, 44: -02, 45: -02, 46: 002, 47: 002.
48: 002, 49: 002, 50: 030, 51: -02, 52: -02, 53: -02, 54: 002, 55: 002.
56: -02, 57: -02, 58: 002, 59: 002, 60: 002, 61: 002, 62: -02, 63: -02.
```

So we decode as 50 and estimate 1 error.

## The amazing Golay code

- Golay Code  $G_{24}$  is a [24, 12, 8] linear code.
- $G = [I_{12}|C_0|N] = [I_{12}|B]$ 

  - N is formed by circulating (1, 1, 0, 1, 1, 1, 0, 0, 0, 1, 0)
     11 times and appending an row of 11 1's.
- The first row of N corresponds to the quadratic residues (mod 11).
- Note that wt(r<sub>1</sub>+r<sub>2</sub>)= wt(r<sub>1</sub>)+wt(r<sub>2</sub>)-2[r<sub>1</sub>·r<sub>2</sub>], all codewords have weight divisible by 4 and d(C)=8.
- $G_{24} = G_{24}^{\perp}$ . To decode Golay, write G= [I<sub>12</sub>|B] and B<sup>T</sup>= (**b**<sub>1</sub>, **b**<sub>2</sub>, ..., **b**<sub>12</sub>) with **b**<sub>i</sub> a column vector.

## G for G(24,12, 8)

8 9 10 11 12 13 14 15 16 17 18 19 20 21 22 23 24 1 2 3 4 5 

## Properties of the Golay code

- The Golay code G(24,12, 8) is self dual. Thus,  $GG^T=I+BB^T=0$
- Other properties:
  - Non-zero positions form a (24, 8, 5) Steiner system.
  - Weights are multiples of 4.
  - Minimum weight CW is 8 (hence d=8).
  - Codewords have weights 0, 8, 12, 16, 24.
  - Weight enumerator is  $1+(759)x^8+(2576)x^{12}+(759)x^{16}+x^{24}$ .
- Voyager 1, 2 used this code.
- Get G(23,12,7) is obtained by deleting last column. It is a remarkable error correcting code. 7=2x3+1, so it corrects 3 errors. It does this "perfectly."

## The Golay code G(23,12,7) is perfect!

- There are 2<sup>12</sup> code words or sphere centers.
- There are <sub>23</sub>C<sub>1</sub>=23 points in Z<sub>23</sub> which differ by one bit from a codeword.
- There are <sub>23</sub>C<sub>2</sub>=253 points in Z<sub>23</sub> which differ by two bits from a codeword.
- There are <sub>23</sub>C<sub>3</sub>=1771 points in Z<sub>23</sub> which differ by two bits from a codeword.
- $2^{12} (1+23+253+1771) = 2^{12} (2048) = 2^{12} \times 2^{11} = 2^{23}$ .
- 23 bit strings which differ by a codeword by 0,1,2 or 3 bits partition the entire space.
- The three sporadic simple Conway's groups are related to the lattice formed by codewords and provided at least one Ph.D. thesis.

## Decoding G(24,12,8)

Suppose r=c+e is received. G= [I<sub>12</sub> | B]=[c<sub>1</sub>, c<sub>2</sub>, ..., c<sub>24</sub>] and B<sup>T</sup>= [b<sub>1</sub>, b<sub>2</sub>, ..., b<sub>12</sub>].

#### To decode:

- 1. Compute  $\mathbf{s} = \mathbf{r}G^T$ ,  $\mathbf{s}B$ ,  $\mathbf{s} + \mathbf{c}_i^T$ ,  $1 \le i \le 24$  and  $\mathbf{s}B + \mathbf{b}_i^T$ ,  $1 \le j \le 12$ .
- If wt(s)≤3, non-zero entries of s correspond to non-zero entries of e.
- 3. If wt(sB)≤3, there is a non-zero entry in the k-th position of sB if the k+12-th position of e is non-zero.
- 4. If wt( $\mathbf{s}+\mathbf{c_i}^{\mathsf{T}}$ )  $\leq$  2, for some j,  $13 \leq$  j  $\leq$  24 then  $\mathbf{e_j}$ =1 and non-zero entries of  $\mathbf{s}+\mathbf{e_i}^{\mathsf{T}}$  are in the same positions as non-zero entries of e.
- 5. If wt( $\mathbf{s}B+\mathbf{b}_{\mathbf{j}}^{\mathsf{T}}$ )  $\leq$  2, for some j,  $1 \leq$  j  $\leq$  12 then  $\mathbf{e}_{\mathbf{j}}$ =1 and non-zero entries of  $\mathbf{s}B+\mathbf{b}_{\mathbf{j}}^{\mathsf{T}}$  at position k correspond to non-zero entries of  $\mathbf{e}_{\mathbf{k+12}}$ .

# Decoding G(24,12,8) example

- G is 12 x 24.  $G=[I_{12}|B]=(c_1, c_2, ..., c_{24})$ .
- $B^T = (b_1, b_2, ..., b_{12}).$
- $\mathbf{m} = (1,1,0,0,0,0,0,0,0,0,0,1,0).$
- mG=(1,1,0,0,0,0,1,0,1,0,1,1,0).
- $\mathbf{r}=(1,1,0,1,0,0,0,0,0,0,1,0,1,0,0,0,0,1,0,0,0,0,1,0)$ .
- **s**=(011110110010).
- **s**B=(101011001000).
- Neither has wt≦3, so we compute s+c<sub>i</sub><sup>T</sup>, sB+b<sub>i</sub><sup>T</sup>.
- $\mathbf{s+b_4}^T = (0,0,0,0,0,0,0,1,0,1,0,0)$

- $\mathbf{m} = (1,1,0,0,0,0,0,0,0,0,0,1,0).$

#### Cyclic codes

- A cyclic code, C, has the property that if (c<sub>1</sub>, c<sub>2</sub>, ..., c<sub>n</sub>)εC then (c<sub>n</sub>, c<sub>1</sub>, ..., c<sub>n-1</sub>) ε C.
- Remember polynomial multiplication in F[x] is linear over F.
- Denoting U<sub>n</sub>(x)= x<sup>n</sup>-1 we have
- Theorem: C is a cyclic code of length n iff its generator  $g(x)=a_0+a_1x+...+a_{n-1}x^{n-1}|U_n(x)$  where codewords c(x) have the form m(x) g(x). Further, if  $U_n(x)=h(x)g(x)$ , c(x) in C iff h(x)c(x)=0 (mod  $U_n(x)$ ).

#### Cyclic codes

- Let C be a cyclic code of length n over F, and let  $\mathbf{a} = (a_0, a_1, \ldots, a_{n-1}) \mathbf{\varepsilon}$  C be associated with the polynomial  $p_{\mathbf{a}}(x) = a_0 + a_1 x + \ldots + a_{n-1} x^{n-1}$ . Let g(x) the polynomial of smallest degree over such associated polynomials the g(x) is the generating polynomial of C and
  - 1.g(x) is uniquely determined.
  - $2.g(x)|x^{n}-1$
  - 3.C: f(x)g(x) where  $deg(f(x)) \le n-1-deg(g)$
  - 4. If  $h(x)g(x)=x^{n}-1$ , m(x)C iff h(x)m(x)=0 (mod  $x^{n}-1$ ).
- The associated matrices G and H are on the next slide.

# G, H for cyclic codes

Let g(x) be the generating polynomial of the cyclic code C.

JLM 20101208

#### Cyclic code example

- $g(x)= 1+x^2+x^3$ ,  $h(x)= 1+x^2+x^3+x^4$ ,  $g(x)h(x)= x^n-1$ , n=7.
- Message 1010 corresponds to  $m(x) = 1 + x^2$ .
- $g(x)m(x)=c(x)=1+x^3+x^4+x^5$ , which corresponds to the codeword 1001110.
- G, H are
- Codewords are
  - 1011000 0101100 0010110 0001011 1110100 0111010 0011101 1001110

#### **BCH Codes**

- Cyclic codes; so generator, g(x) satisfies g(x)|x<sup>n</sup>-1.
- Theorem: Let C be a cyclic [n, k, d] code over  $F_q$ ,  $q=p^m$ . Assume p does not divide n and g(x) is the generator. Let a be a primitive root of  $x^n$ -1 and suppose that for some I, d, we have  $g(a^l)=g(a^{l+1})=\ldots=g(a^{l+d})=0$ , then  $d^3d+2$ .
- Constructing a BCH code:
  - 1. Factor  $x^n-1=f_1(x)$   $f_2(x)...f_r(x)$ , each  $f_i(x)$ , irreducible.
  - 2. Pick a, a primitive root of 1.
  - 3.  $x^n-1=(x-a)(x-a^2)...(x-a^{n-1})$  and  $f_i(x)=\prod_t (x-a^{j(t)})$ .
  - 4.  $q_i(x) = f_i(x)$ , where  $f_i(a) = 0$ .  $q_i(x)$  are not necessarily distinct.
  - 5. BCH code at designed distance d has generator  $g(x)=LCM[q_{k+1}(x),...,q_{k+d-1}(x)].$
- Theorem: A BCH code of designed distance d has minimum weight
   3d. Proof uses theorem above.

## Example BCH code

- F=F<sub>2</sub>, n=7.
- $x^7-1=(x-1)(x^3+x^2+1)(x^3+x+1)$
- We pick a, a root of (x³+x+1) as a primitive element.
- Note that  $a^2$  and  $a^4$  are also primitive roots of  $(x^3+x+1)$ , so  $x^3+x+1=(x-a)(x-a^2)(x-a^4)$  and  $x^3+x^2+1=(x-a^3)(x-a^6)(x-a^6)$
- $q_0(x)=x-1$ ,  $q_1(x)=q_2(x)=q_4(x)=x^3+x^2+1$ .
- k=-1, d=3,  $g(x)=[x-1, x^3+x^2+1]=x^4+x^3+1$ .
- This yields a [7,3,4] linear code.

## **Decoding BCH Codes**

- For **r=c+e**:
  - 1. Compute  $(s_1, s_2) = \mathbf{r} \mathbf{H}^T$ ,
  - 2. If  $s_1 = 0$ , no error,
  - 3. If  $s_1 \neq 0$  put  $s_2/s_1 = a^{j-1}$ , error is in position j (of  $p \neq 2$ ,  $e_i = s_1/a^{(j-1)(k+1)}$ ,
  - 4. c=r-e.

## Example Decoding a BCH Code

- $x^7$ -1, a, a root of  $x^3$ +x+1=0. This is the 7-repetition code.
- $rH^T = (1,1,1,1,0,1,1,1) H^T = (a+a^2, a)$
- H= 1, a,  $a^2$ ,  $a^3$ ,  $a^4$ ,  $a^5$ ,  $a^6$ 1,  $a^2$ ,  $a^4$ ,  $a^6$ ,  $a^8$ ,  $a^{10}$ ,  $a^{12}$
- $s_1 = a + a^2 = 1 + a + a^2 + a^3 + a^4 + a^5 + a^6$
- $s_2 = a = 1 + a^2 + a^4 + a^6 + a^8 + a^{10} + a^{12}$
- $s_1/s_2=a^4$ , j-1=4, j=5, **e**=(0,0,0,0,1,0,0).
- $s_1 = e_i a^{(j+1)(k+1)}$
- $s_2 = e_i a^{(j+1)(k+2)}$

#### Reed Solomon

- Reed-Solomon code is BCH code over  $F_q$  with n = q-1. Let  $\alpha$  be a primitive root of 1 and choose d:  $1 \le d < n$  with  $g(x) = (x-\alpha)(x-\alpha^2)...(x-\alpha^{d-1})$ .
  - Since  $g(\alpha) = g(\alpha^2) = ... = g(\alpha^{d-1})=0$ , BCH bound shows  $d(C) \ge d$ .
  - Codewords are g(x)f(x), deg(f(x))≦n-d. There are  $q^{n-d+1}$  such polynomials so  $q^{n-d+1}$  codewords.
  - Since this meets the Singleton bound, the Reed Solomon code is also an MDS code.
  - The Reed Solomon Code is an [n, n-d+1, d] linear code for these parameters

#### Reed Solomon example

#### Example:

```
F=GF(2²)={0,1,w,w²}
n=q-1=3, a= w.
Choose d=2, g(x)= (x-w).
G= w 1 0
0 w 1
```

- Code consists of all 16 linear combinations of the rows of G.
- For CD's:
  - $F=GF(2^8)$ , n=  $2^8$ -1=255, d=33.
  - 222 information bytes.33 check bytes.
  - Codewords have 8 x 255 = 2040 bits.

#### Polynomials and RM codes

- R(r,m) has parameters  $[n=2^m, k=1+{}_mC_1+...+{}_mC_r d=2^{m-r}]$ , it consists of boolean functions whose polynomials are of degree  $\leq m$ .
- $RM(r,m)^{-} = RM(m-r-1,m)$ .
- RM(0,m)= {0, 1}, RM(r+1, m+1)= RM(r+1, m)\*R(r, m).
- RM(n,0) is a repetition code with rate 1/n.
- Min distance in R(r,m)= 2<sup>m-r</sup>.

$$G(r+1,m)$$
  $G(r+1,m)$ 
•  $G(r+1,m+1)=$  0  $G(r,m)$ 

# RM(4,0) and RM(4,1)

- $n=2^4=16$ .
- Constants
- Linear

  - 0000 1111 0000 1111, 0000 0000 1111 1111

#### RM(r,4) code example

```
11111111111111111
         0000000011111111
X_4
         0000111100001111
X_3
         0011001100110011
X_2
         0101010101010101
X_1
         0000000000001111
X_3X_4
         000000000110011
X_2X_4
         0000000001010101
X_1X_4
         0000001100000011
X_2X_3
         0000101000000101
X_1X_3
         0001000100010001
X_1X_2
         0000000000000011
X_2X_3X_4
         0000000000000101
X_1X_3X_4
         0000000000010001
X_1X_2X_4
         000000010000001
X_1X_2X_3
         00000000000000001
X_1X_2X_3X_4
```

#### McEliece Cryptosystem

- Bob chooses G for a large [n, k, d] linear code, we particularly want large d (for example, a [1024, 512, 101] Goppa code which can correct 50 errors in a 1024 bit block). Pick a k x k invertible matrix, S, over GF(2) and P, an n x n permutation matrix, and set G<sub>1</sub>=SGP. G<sub>1</sub> is Bob's public key; Bob keeps P, G and S secret.
- To encrypt a message, x, Alice picks an error vector, e, and sends y=xG<sub>1</sub>+e (mod 2).
- To decrypt, Bob, computes  $y_1=yP^{-1}$  and  $e_1=eP^{-1}$ , then  $y_1=xSG+e_1$ . Now Bob corrects  $y_1$  using the error correcting code to get  $x_1$ . Finally, Bob computes  $x=x_1S^{-1}$ .
- Error correction is similar to the "shortest vector problem" and is believed to be "hard." In the example cited, a [1024, 512, 101] Goppa code, finding 50 errors (without knowing the shortcut) requires trying  $_{1024}C_{50}>10^{85}$  possibilities.
- A drawback is that the public key, G<sub>1</sub>, is largest.

# McEliece Cryptosystem example - 1

• Using the [7, 4] Hamming code, G=

m=1011.

## McEliece Cryptosystem example - 2

- $e = (0 \ 1 \ 0 \ 0 \ 0 \ 0)$ •  $y_1 = yP^{-1} = (0 \ 0 \ 1 \ 0 \ 0 \ 1)$
- $\mathbf{x}_1 = (0\ 0\ 1\ 0\ 0\ 1\ 1)$
- $\mathbf{x_0} = (0\ 0\ 1\ 0)$
- $\mathbf{x} = \mathbf{x_0} \mathbf{S}^{-1} = (1 \ 0 \ 1 \ 1)$

# End

JLM 20101208 54