

Sensitivity of Fields Generated within Magnetically Shielded Volumes to Changes in Permeability

T. Andalib^b, J.W. Martin^{a,b,*}, C.P. Bidinosti^{a,b}, R.R. Mammei^{a,b},
B. Jamieson^{a,b}, M. Lang^b

^a*Physics Department, The University of Winnipeg, 515 Portage Avenue, Winnipeg, MB, R3B 2E9, Canada*

^b*Department of Physics and Astronomy, University of Manitoba, Winnipeg, MB R3T 2N2, Canada*

Abstract

Future experiments seeking to measure the neutron electric dipole moment (nEDM) require stable and homogeneous magnetic fields. Normally these experiments use a coil internal to a passively magnetically shielded volume to generate the magnetic field. The stability of the magnetic field generated by the coil within the magnetically shielded volume may be influenced by a number of factors. The factor studied here is the dependence of the internally generated field on the permeability μ of the shield material. We provide measurements of the temperature-dependence of the permeability of the material used in a set of prototype magnetic shields, using experimental parameters nearer to those of nEDM experiments than previously reported in the literature. Our measurement implies $0.1\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.7\%/K$. Assuming typical nEDM experiment coil and shield parameters gives $\frac{\mu}{B_0} \frac{dB_0}{d\mu} \sim 0.01$ resulting in a temperature dependence of the magnetic field in a typical nEDM experiment of $\frac{dB_0}{dT} = 10 - 270$ pT/K for $B_0 = 1$ μ T. The results are useful for estimating the necessary level of temperature control in nEDM experiments.

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*Corresponding author

Email address: j.martin@uwinnipeg.ca (J.W. Martin)

1. Introduction

The next generation of neutron electric dipole moment (nEDM) experiments aim to measure the nEDM d_n with proposed precision $\delta d_n \lesssim 10^{-27} \text{ e}\cdot\text{cm}$ [1–8]. In the previous best experiment [9, 10] which discovered $d_n < 3.0 \times 10^{-26} \text{ e}\cdot\text{cm}$ (90% C.L), effects related to magnetic field homogeneity and instability were found to dominate the systematic error. A detailed understanding of passive and active magnetic shielding, magnetic field generation within shielded volumes, and precision magnetometry is expected to be crucial to achieve the systematic error goals for the next generation of experiments. Much of the research and development efforts for these experiments are focused on careful design and testing of various magnetic shield geometries with precision magnetometers [11–15].

In nEDM experiments, the spin-precession frequency ν of neutrons placed in static magnetic B_0 and electric E fields is measured. The measured frequencies for parallel ν_+ and antiparallel ν_- relative orientations of the fields is sensitive to the neutron electric dipole moment d_n

$$h\nu_{\pm} = 2\mu_n B_0 \pm d_n E \quad (1)$$

where μ_n is the magnetic moment of the neutron.

A problem in these experiments is that if the magnetic field B_0 drifts over the course of the measurement period, it worsens the statistical precision with which d_n can be determined. For a $\delta d_n \sim 10^{-27} \text{ e}\cdot\text{cm}$ measurement, the average field over one fill cycle ($\sim 100 \text{ s}$) of the measurement cell must be known to the 10 fT level or better, and the field is seldom stable to this level. For this reason, experiments use a comagnetometer and/or surrounding atomic magnetometers to measure and correct the magnetic field to this level [9, 11, 12]. Drifts of 1-10 pT in B_0 may be corrected using the comagnetometer technique, setting a goal magnetic stability for the B_0 field generation system in a typical nEDM experiment.

In such experiments, typically $B_0 = 1 \text{ }\mu\text{T}$ is used to provide the quantization axis for the UCN. The B_0 magnetic field generation system typically includes a

30 coil placed within a passively magnetically shielded volume. The passive magnetic shield is generally composed of a multi-layer shield formed from thin shells of high-permeability material (mu-metal). The outer layers of the shield are normally cylindrical [1, 4] or form the walls of a magnetically shielded room [16, 17]. The innermost magnetic shield is normally a specially shaped shield, where the
 35 design of the coil in relation to shield is carefully taken into account to achieve adequate homogeneity [3, 5, 9].

Mechanical and temperature changes of the passive magnetic shielding [18, 19], and the demagnetization procedure [17, 19, 20], affect the stability of the magnetic field within magnetically shielded rooms. Active stabilization of the
 40 background magnetic field surrounding magnetically shielded rooms can also improve the internal stability [12, 18, 21]. The current supplied to the B_0 coil is generated by an ultra-stable current source [11]. The coil must also be stabilized mechanically relative to the magnetic shielding.

One additional effect, which is the subject of this paper, relates to the fact
 45 that the B_0 coil in most nEDM experiments is magnetically coupled to the innermost magnetic shield. If the magnetic properties of the innermost magnetic shield change as a function of time, it then results in a source of instability of B_0 . In the present work, we estimate this effect and characterize one possible source of instability: changes of the magnetic permeability μ of the material
 50 with temperature.

While the sensitivity of magnetic alloys to temperature variations has been characterized in the past [22, 23], we sought to make these measurements in regimes closer to the operating parameters relevant to nEDM experiments. For these alloys, it is also known that the magnetic properties are set during the
 55 final annealing process [23–25]. In this spirit we performed our measurements on small witness cylinders that were annealed using the typical process.

2. Sensitivity of Internally Generated Field to Permeability of the Shield $B_0(\mu)$

The presence of a coil inside the innermost passive shield turns the shield into a return yoke, and generally results in an increase in the magnitude of B_0 . The ratio of this field inside the coil in the presence of the magnetic shield to that of the coil in free space is referred to as the reaction factor C , and can be calculated analytically for spherical and infinite cylindrical geometries [26, 27]. The key issue of interest for this work is the dependence of the reaction factor on the permeability μ of the innermost shield. Although this dependence can be rather weak, the constraints on B_0 stability are very stringent. As a result, even a small change in the magnetic properties of the innermost shield can result in an unacceptably large change in B_0 .

To illustrate, we consider here the model of a sine-theta surface current on a sphere of radius a , inside a spherical shell of inner radius R , thickness t , and linear permeability μ . The uniform internal field generated by this ideal spherical coil is augmented by a factor C in the presence of the shield, but is otherwise left undistorted. The general reaction factor for this model is given by Eq. (38) in Ref. [26]. In the high- μ limit, with $t \ll R$, the reaction factor can be approximated as

$$C \simeq 1 + \frac{1}{2} \left(\frac{a}{R} \right)^3 \left(1 - \frac{3}{2} \frac{R}{t} \frac{\mu_0}{\mu} \right), \quad (2)$$

which highlights the dependence of B_0 on the relative permeability $\mu_r = \mu/\mu_0$ of the shield.

Fig. 1 shows plots of B_0 versus μ_r for coil and shield dimensions similar to the ILL nEDM experiment [9, 28]: $a = 0.53$ m, $R = 0.57$ m, and $t = 1.5$ mm. In addition to analytic calculations, we also include the results of two axially symmetric simulations conducted using FEMM [29] to assess the effects of geometry and discretization of the surface current.

In the first simulation, the same spherical geometry was used as for the analytical calculations. However, the surface current was discretized to 50 individual current loops, inscribed onto a sphere, and equally spaced vertically

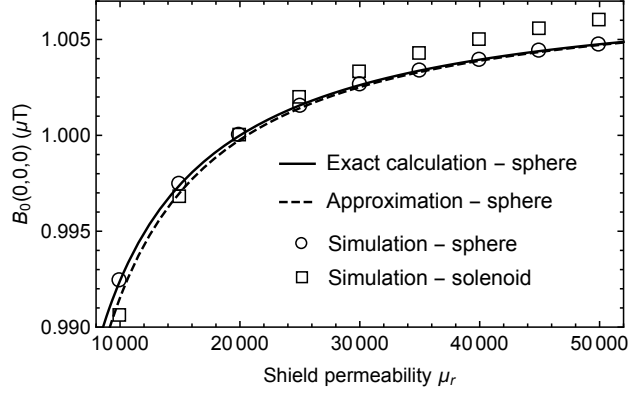


Figure 1: Magnetic field at the coil isocenter as a function of shield permeability for a geometry similar to the ILL nEDM experiment as discussed in the text. The solid curve is the exact calculation for the ideal spherical coil and shield from Ref. [26]; the dashed curve is the approximation of Eq. 2. The circles and squares are the FEMM-based simulations for the spherical and solenoidal geometries with discrete currents. In all cases, current magnitudes were chosen to give $B_0 = 1 \mu\text{T}$ at $\mu_r = 20,000$.

(i.e. a discrete sine-theta coil). A square wire profile of side length 1 mm was used. As shown in Fig. 1, this simulation gave good agreement with the analytical calculation. In the second simulation, a solenoid coil and cylindrical shield (length/radius = 2) were used with the same dimensions as above. Similarly,
90 the coil was modelled as 50 evenly spaced current loops, with the distance from an end loop to the inner face of the shield end-cap being half the inter-loop spacing. In the limit of tight-packing (i.e., a continuous surface current) and infinite μ , the image currents in the end caps of the shield act as an infinite series of current loops, giving the ideal uniform field of an infinitely long solenoid [30, 31].
95 As shown in Fig. 1, the result is similar to the spherical case, though the slope of $B_0(\mu_r)$ is somewhat steeper.

In ancillary measurements of shielding factors (discussed briefly in Section 3.1), we found $\mu_r = 20,000$ to offer a reasonable description of the quasistatic shielding factor. By evaluating the slope of the analytic curve in Fig. 1
100 at $\mu_r = 20,000$, we estimate that the scale of the sensitivity of a generic nEDM

experiment to global changes in the magnetic permeability is $\frac{\mu}{B_0} \frac{dB_0}{d\mu} \sim \frac{3a^3}{4R^2t} \frac{1}{\mu_r} \sim 0.01$.

For a high- μ innermost shield, the magnetic field lines emanating from the coil all return through the shield. This principle can be used to estimate the magnetic field internal to the material B_m , and in our studies gave good agree-
105 ment with FEA-based simulations. For the solenoidal geometry previously described and used for the calculations in Fig. 1, B_m is largest in the side walls of the solenoidal flux return, attaining a maximum value of $170 \mu\text{T}$. If we assume $\mu_r=20,000$, the H_m field is 0.007 A/m . Typically the shield is degaussed
110 (idealized) with the internal coil energized. After degaussing, B_m must be approximately the same, since essentially all flux returns through the shield. However, the H_m field must become significantly smaller, as the material must reside on the ideal magnetization curve in $B_m - H_m$ space. (For a discussion of the ideal magnetization curve, we refer the reader to Ref. [25].) In principle, the
115 H_m field could be reduced by an order of magnitude or more, depending on the steepness of the ideal magnetization curve near the origin. Thus $B_m = 170 \mu\text{T}$ and $H_m < 0.007 \text{ A/m}$ set a scale for the relevant values for nEDM experiments. Furthermore, the field in the nEDM measurement volume, as well as in the magnetic shield, must be stable for periods of typically hundreds of seconds
120 (corresponding to frequencies $< 0.01 \text{ Hz}$). This sets the relevant timescale for magnetic properties most relevant to nEDM experiments.

3. Measurements of $\mu(T)$

3.1. Previous Measurements and their Relationship to nEDM Experiments

Previous measurements of the temperature dependence of the magnetic prop-
125 erties of high-permeability alloys have been summarized in Refs. [22, 25, 32]. These measurements are normally conducted using a sample of the material to create a toroidal core, where a thin layer of the material is used in order to avoid eddy-current and skin-depth effects [23, 32]. A value of μ is determined by dividing the amplitude of the sensed B_m -field by the amplitude of the driving

130 AC H_m -field (similar to the method described in Section 3.3). Normally the frequency of the H_m -field is 50 or 60 Hz. The value of μ is then quoted either at or near its maximum attainable value by adjusting the amplitude of H_m . Depending on the details of the $B_m - H_m$ curve for the material in question, this normally means that μ is quoted for the amplitude of H_m being at or near the
135 coercivity of the material [22, 23], resulting in large values up to $\mu_r = 4 \times 10^5$.

It is well known that μ measured in this fashion for toroidal, thin metal wound cores depends on the annealing process used for the core. There is a particularly strong dependence on the take-out or tempering temperature after the high-temperature portion of the annealing process has been completed [22,
140 23, 32]. Such studies normally suggest a take-out temperature of 490-500°C. This ensures that the large $\mu_r = 4 \times 10^5$ is furthermore maximal at room temperature. Slight variations around room temperature, and assuming the take-out temperature is not controlled to better than a degree, imply a scale of possible temperature variation of μ of approximately $\left| \frac{1}{\mu} \frac{d\mu}{dT} \right| \simeq 0.3\text{-}1\%/K$ at
145 room temperature [22, 23].

A challenge in applying these results to temperature stability of nEDM experiments is that, when used as DC magnetic shielding, the high-permeability alloys are usually operated for significantly different parameters (B_m , H_m , and frequencies).

150 For example, when used in a shielding configuration, the effective permeability is often measured to be typically $\mu_r = 20,000$ rather than 4×10^5 . This arises in part because H_m is well below the DC coercivity. As noted in Section 2, a more appropriate H_m for the innermost magnetic shield of an nEDM experiment is < 0.007 A/m, whereas the coercivity is $H_c = 0.4$ A/m [23]. The frequency
155 dependence of the measurements could also be an issue. Typically, nEDM experiments are concerned with slow drifts at < 0.01 Hz timescales whereas the previously reported $\mu(T)$ measurements are performed in an AC mode at 50 Hz frequencies.

The goal of our experiments was to develop techniques to characterize the
160 material properties of our own magnetic shields post-annealing, in regimes more

relevant to nEDM experiments.

We created a prototype passive magnetic shield system in support of this and other precision magnetic field research for the future nEDM experiment to be conducted at TRIUMF. The shield system is a four-layer mu-metal shield
165 formed from nested right-circular cylindrical shells with endcaps. The inner radius of the innermost shield is 18.44 cm, equal to its half-length. The radii and half-lengths of the progressively larger outer shields increase geometrically by a factor of 1.27. Each cylinder has two end-caps which possess a 7.5 cm diameter central hole. A stove-pipe of length 5.5 cm is placed on each hole was designed to
170 minimize leakage of external fields into the progressively shielded inner volumes. The design is similar to another smaller prototype shield discussed in Ref. [33]. The magnetic shielding factors of each of the four cylindrical shells, and of various combinations of them, were measured and found to be consistent with $\mu_r \sim 20,000$.

175 In our studies of the material properties of these magnetic shields, two different approaches to measure $\mu(T)$ were pursued. Both approaches involved experiments done using witness cylinders, which are smaller open-ended cylinders (see Fig. 2) made of the same material and annealed at the same time as the prototype magnetic shields. We therefore expect they have the same mag-
180 netic properties as the larger prototype shields, and they have the advantage of being smaller and easier to perform measurements with.

The two techniques used were

1. measuring the AC axial shielding factor of the witness cylinder as a function of temperature, and
- 185 2. measuring the temperature-dependence of the slope of a minor B-H loop, using the witness cylinder as a transformer core, similar to previous measurements of the temperature dependence of μ , but for parameters closer to those encountered in nEDM experiments.

We now discuss the details and results of each technique.

190 *3.2. Axial Shielding Factor Measurements*

In these measurements, a witness cylinder was used as a magnetic shield. The shield was subjected to an AC magnetic field. The amplitude of the shielded magnetic field B_s was measured at the center of the witness cylinder. Changes in B_s with temperature signify a dependence of the permeability μ on temperature. 195 The relative slope of $\mu(T)$ can then be calculated using

$$\frac{1}{\mu} \frac{d\mu}{dT} = - \frac{\frac{1}{B_s} \frac{dB_s}{dT}}{\frac{\mu}{B_s} \frac{dB_s}{d\mu}}. \quad (3)$$

The numerator was taken from the measurements described above. The denominator was taken from finite-element simulations of the shielding factor for this geometry as a function of μ .

This technique is quite different than the usual transformer core measurements conducted by other groups. As shall be described, it offers an advantage 200 that considerably smaller B_m and H_m fields can be accessed. Measuring the temperature dependence of the shielding factor is also considerably easier than measuring the temperature dependence of the reaction factor, since the sensitivity to changes in $\mu(T)$ is considerably larger in magnitude for the shielding factor case where $\frac{\mu}{B_s} \frac{dB_s}{d\mu} \sim -1$ compared to the reaction factor case where 205 $\frac{\mu}{B_0} \frac{dB_0}{d\mu} \sim 0.01$.

3.2.1. Experimental Apparatus for Axial Shielding Factor Measurements

The witness cylinder was placed within a homogeneous AC magnetic field. The field was created within the magnetically shielded volume of the prototype 210 magnetic shielding system (described previously in Section 3.1) in order to provide a controlled magnetic environment. A short solenoid inside the shielding system was used to produce the magnetic field. The solenoid has 14 turns with 2.6 cm spacing between the wires. The solenoid was designed so that the field produced by the solenoid plus innermost shield approximates that of an infinite 215 solenoid. The magnetic field generated by the solenoid was typically 1 μ T in amplitude. The solenoid current was varied sinusoidally at typically 1 Hz.

The witness cylinder was placed into this magnetic field generation system as shown schematically in Fig. 2. The cylinder was held in place by a wooden stand.

220 A Bartington fluxgate magnetometer Mag-03IEL70 [34] (low noise) measured the axial magnetic field at the center of the witness cylinder. The fluxgate was a “flying lead” model, meaning that each axis was available on the end of a short electrical lead, separable from the other axes. One “flying lead” was placed in the center of the witness cylinder, the axis of the fluxgate being
225 aligned with that of the witness cylinder. The fluxgate was held in place rigidly by a plastic mounting fixture, which was itself rigidly mounted to the witness cylinder.

To increase the resolution of the measured signal from the fluxgate, a Bartington Signal Conditioning Unit (SCU) was used with a low-pass filter set to
230 typically 10-100 Hz and a gain set to typically > 50 . The signal from the SCU was demodulated by an SR830 lock-in amplifier [35] providing the in-phase and out-of-phase components of the signal. The sinusoidal output of the lock-in amplifier reference output itself was normally used to drive the solenoid generating the magnetic field. The time constant on the lock-in was typically set to
235 3 seconds with 12 dB/oct rolloff.

As shall be described in Section 3.2.2, a concern in the measurement was changes in the field measured by the fluxgate that could arise due from motion of the system components, or other temperature dependences. This could generate a false slope with temperature that might incorrectly be interpreted as a change
240 in the magnetic properties of the witness cylinder.

To address possible motion of the witness cylinder with respect to the field generation system, another coil (the loop coil, also shown in Fig. 2) was wound on a plastic holder mounted rigidly to the witness cylinder. The coil was one loop of copper wire with a diameter of 9.7 cm. Plastic set screws in the holder
245 fixed the loop coil to be coaxial with the witness cylinder.

Systematic differences in the results from the two coils (the solenoidal coil, and the loop coil) were used to search for motion artifacts. As well, some

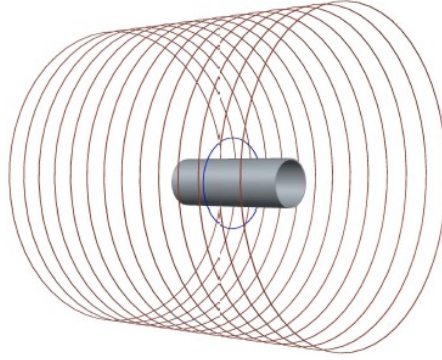


Figure 2: (color online) Axial shielding factor measurement setup. The witness cylinder with an inner diameter of 5.2 cm and a length of 15.2 cm is placed inside a solenoid (shown in red) with a diameter of 30.8 cm and a length of 35.5 cm, containing 14 turns. The thickness of the witness cylinder is $1/16'' = 0.16$ cm. The loop coil (shown in blue) is mechanically coupled to the witness cylinder and has a diameter of 9.7 cm.

differences could arise due to the different magnetic field produced by each coil, and so such measurements could reveal a dependence on the profile of the applied magnetic field. This is described further in Section 3.2.2.

The temperature of the witness cylinder was measured by attaching four thermocouples at different points along the outside of the cylinder. This allowed us to observe the temperature gradient along the witness cylinder. To reduce any potential magnetic contamination, T-type thermocouples were used, which have copper and constantan conductors. (K-type thermocouples are magnetic.)

Thermocouple readings were recorded by a National Instruments NI-9211 temperature input module. The magnetic field (signified by the lock-in amplifier readout) and the temperature were recorded at a rate of 0.2 Hz.

Temperature variations in the experiment were driven by ambient temperature changes in the room, although forced air and other techniques were also tested. These are described further in Section 3.2.2.

3.2.2. Data and Interpretation

An example of the typical data acquired is shown in Fig. 3. For these data, the field applied by the solenoid coil was 1 μ T in amplitude, at a frequency of 1 Hz. Fig. 3(a) shows the temperature of the witness cylinder over a 70-hr measurement. The temperature changes of 1.4 K are caused by diurnal variations in the laboratory. The shielded magnetic field amplitude B_s within the witness cylinder is anti-correlated with the temperature trend as shown in Fig. 3(b). Here, B_s is the sum in quadrature of the amplitudes of the in-phase and out-of-phase components (most of the signal is in phase). The magnetic field is interpreted to depend on temperature, and they are graphed as a function of one another in Fig. 3(c). The slope in Fig. 3(c) has been calculated using a linear fit to the data. The relative slope at 23°C was found to be $\frac{1}{B_s} \frac{dB_s}{dT} = -0.75\%/K$.

Some deviations from the linear straight-line dependence can be seen in the data. For example, when the temperature changes rapidly, the magnetic field takes some time to respond, resulting in a slope in $B_s - T$ space that is temporarily different than when the temperature is slowly varying. This is typical of the data that we acquired, that the data would generally follow a straight line if the temperature followed a slow and smooth dependence with time, but the data would not be linear if the temperature varied rapidly or non-monotonically with time. We also tried other methods of temperature control, such as forced air, liquid flowing through tubing, and thermo-electric coolers. The diurnal cycle followed by the building's air conditioning was found to be the most stable and gave the most reproducible results for temperature slopes.

As mentioned earlier, data were acquired for both the solenoid coil and the loop coil. Repeated measurements of temperature slopes using the loop coil fell in the range $0.4\%/K < |\frac{1}{B_s} \frac{dB_s}{dT}| < 1.5\%/K$. Similar measurements for the solenoidal coil yielded $0.3\%/K < |\frac{1}{B_s} \frac{dB_s}{dT}| < 0.8\%/K$.

In general, the slopes measured with the loop coil were larger than for the solenoidal coil. A partial explanation of this difference is offered by the field profile generated by each coil, and its interaction with the witness cylinder. This

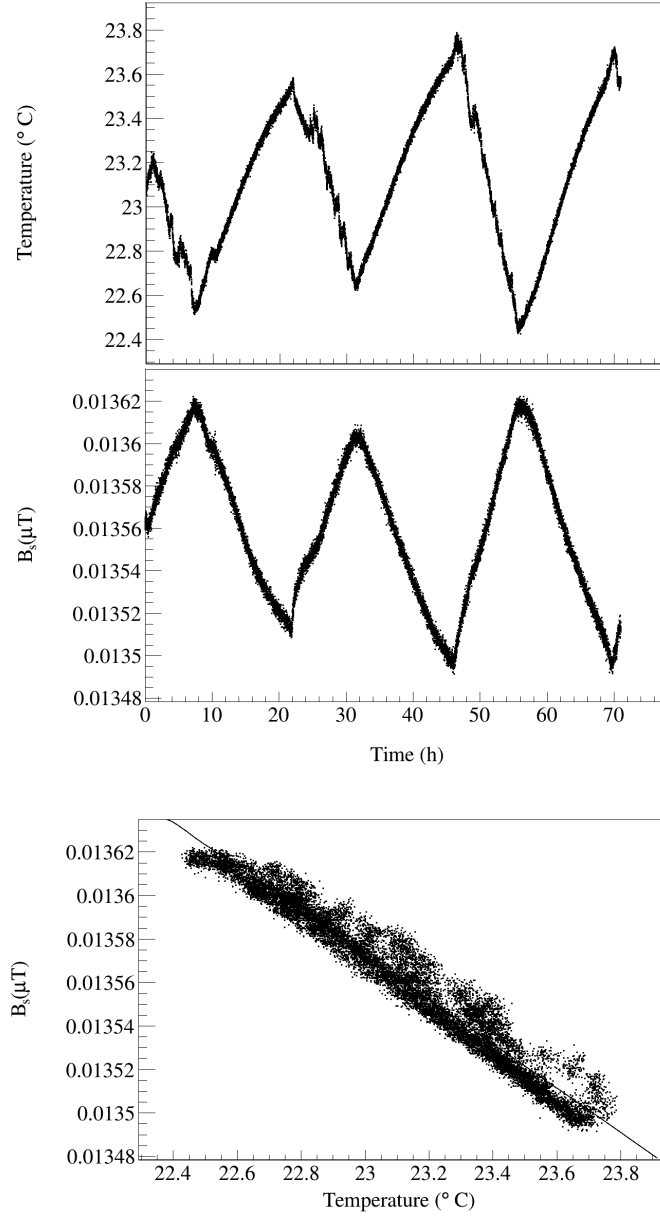


Figure 3: Ambient temperature and shielded magnetic field amplitude, measured over a 70 hour period. (a) temperature of the witness cylinder as a function of time. (b) magnetic field amplitude measured by fluxgate at center of witness cylinder vs. time. (c) magnetic field vs. temperature with linear fit to data. At 23 $^{\circ}\text{C}$, $\frac{1}{B_s} \frac{dB_s}{dT} = -0.75\%/K$.

is addressed further in Section 3.2.3.

The other difference between the loop coil and the solenoidal coil was that the loop coil was rigidly mounted to the witness cylinder, reducing the possibility of artefacts from relative motion. Given that this did not reduce the range of the measured temperature slopes we conclude that relative motion was well controlled in both cases.

Several other possible systematic effects were considered, all of which were found to give uncertainties on the measured slopes $< 0.1\%/K$. These included: thermal expansion of components including the witness cylinder itself, temperature variations of the magnetic shielding system within which the experiments were conducted, degaussing of the witness cylinder, and temperature slopes of various components e.g. the fluxgate magnetometer and the lock-in amplifier.

The stability of the system was also tested by replacing the mu-metal witness cylinder with a copper cylinder of very similar dimensions. The apparatus was then run through its usual experimental cycle over several days, and this was done multiple times for different parameters such as coil current. For all measurements the temperature dependence of the demodulated magnetic signal was $< 0.1\%/K$, giving confidence that unknown systematic effects contribute below this level.

Based on the systematic effects that we studied, we conclude that they do not explain the ranges of values measured for $\frac{1}{B_s} \frac{dB_s}{dT}$. We suspect that the range measured is either some yet uncharacterized systematic effect, or a complicated property of the material.

3.2.3. Geometry correction and determination of $\mu(T)$

To relate the data on $B_s(T)$ to $\mu(T)$, the shielding factor of the witness cylinder as a function of μ must be known. Finite element simulations in FEMM and OPERA were performed to determine this factor. The simulations are also useful for determining the effective values of B_m and H_m in the material, which will be useful to compare to the case for typical nEDM experiments when the innermost shield is used as a flux return.

	$ \frac{\mu}{B_s} \frac{dB_s}{d\mu} $ (simulated)	$ \frac{1}{B_s} \frac{dB_s}{dT} $ (%/K) (measured)	$\frac{1}{\mu} \frac{d\mu}{dT}$ (%/K) (extracted)
Solenoidal Coil	0.42-0.50	0.3-0.8	0.6-1.9
Loop Coil	0.56-0.65	0.4-1.5	0.6-2.7

Table 1: Summary of OPERA and FEMM simulations and shielding factor measurements, resulting in extracted temperature slopes of μ .

For closed objects, such as spherical shells [26, 27], the shielding factor approaches infinity as $\mu \rightarrow \infty$, and $|\frac{\mu}{B_s} \frac{dB_s}{d\mu}| \rightarrow 1$. Because the witness cylinders are open ended, the shielding factor asymptotically approaches a constant rather than infinity in the high- μ limit, and as a result $|\frac{\mu}{B_s} \frac{dB_s}{d\mu}| < 1$ here. From the
325 simulations the ratio $\frac{\mu}{B_s} \frac{dB_s}{d\mu}$ was calculated. A linear model of the material was used where $\mathbf{B}_m = \mu \mathbf{H}_m$ with μ constant.

The simulations differed slightly in their results, dependent on whether OPERA or FEMM was used, and whether the solenoidal coil or loop coil were
330 used. Based on the simulations, the result is $|\frac{\mu}{B_s} \frac{dB_s}{d\mu}| = 0.42 - 0.50$ for the solenoidal coil, with the lower value being given by FEMM and the upper value being given by a 3D OPERA simulation, for identical geometries. This is somewhat lower than the value suggested by Ref. [36] with fits to simulations performed in OPERA, which we estimate to be 0.6. We adopt our value since it is
335 difficult to determine precisely from Ref. [36]. For the loop coil, we determine $|\frac{\mu}{B_s} \frac{dB_s}{d\mu}| = 0.56 - 0.65$, the range being given again by a difference between FEMM and OPERA.

Combining the measurement and the simulations, the temperature dependence of the effective μ (at $\mu_r = 20,000$ which is consistent with our measure-
340 ments) can be calculated by equation (3). The results of the simulations and measurements are presented in Table 1. Combining the loop coil and solenoidal coil results, we find $0.6\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.7\%/K$ to be a reasonable range for the possible temperature slope of μ .

As stated earlier, the simulations also provided a way to determine the typ-

ical B_m and H_m internal to the material of the witness cylinder. According to the simulations, the B_m amplitude was typically 100 μT and the H_m amplitude was typically 0.004 A/m. These are comparable to the values normally encountered in nEDM experiments, recalling from Section 2 that $H_m < 0.007$ A/m for the innermost magnetic shield of an nEDM experiment. A caveat is that these measurements were typically conducted using AC fields at 1 Hz, as opposed to the DC fields normally used in nEDM experiments.

3.3. Transformer Core Measurements

As an alternative method of measuring changes in μ , a method similar to the standard method of magnetic materials characterization via magnetic induction was used. In this measurement technique, the witness cylinder was used as the core of a transformer. Two coils (primary and secondary) were wound on the witness cylinder using multistranded 20-gauge copper wire. The windings were made as tight as possible, but not so tight as to potentially stress the material. The windings were not potted in place.

Three witness cylinders were tested. Data were acquired using different numbers of turns on both the primary and secondary coils (from 6 to 48 on the primary, and from 7 to 24 on the secondary).

The primary coil generated an AC magnetic field as a function of time $H(t)$, while the secondary coil was used to measure the emf induced by the time-varying magnetic flux proportional to $dB(t)/dt$. To a good approximation

$$H_m(t) = \frac{N_p I(t)}{2\pi R} \quad (4)$$

where N_p is the number of turns in the primary, $I(t)$ is the current in the primary, and R is the radius of the witness cylinder, and

$$\frac{dB_m(t)}{dt} = \dot{B}_m(t) = \frac{V(t)}{t\ell} \quad (5)$$

where $V(t)$ is the voltage generated in the secondary, and t and ℓ are the thickness and length of the witness cylinder. For a sinusoidal drive current $I(t)$, and under the assumption that $B_m(t) = \mu H_m(t)$ with μ being a constant, the

voltage generated in the secondary $V(t)$ should be sinusoidal and out of phase with the primary current.

The internal oscillator of an SR830 lock-in amplifier was used to generate $I(t)$. This was monitored by measuring the voltage across a $1\ \Omega$ resistor with
 375 small temperature coefficient in the primary loop. The lock-in amplifier was then used to demodulate $V(t)$ into its in-phase V_X and out-of-phase V_Y components (or equivalently $\dot{B}_m(t)$ being demodulated into $\dot{B}_{m,X}$ and $\dot{B}_{m,Y}$, as in equation (5)). The experiment was done at 1 Hz with $H_m(t)$ as small as possible, typically 0.1 A/m in amplitude, to measure the slope of the minor $B_m - H_m$
 380 loops near the origin of the $B_m - H_m$ space.

The temperature of the core was measured continuously using the same thermocouple arrangement described previously. Measurements of V_Y as a function of temperature would then signify a change in μ with temperature. In general, we used ambient temperature variations for the measurements, similar to the
 385 procedure used for our axial shielding factor measurements.

The naive expectation is that the out-of-phase V_Y component should signify a non-zero μ , and the in-phase V_X component should be zero. In practice, due to a combination of saturation, hysteresis, eddy-current losses, and skin-depth effects, the V_X component is nonzero. It was found experimentally that keeping
 390 the amplitude of $H_m(t)$ small compared to the apparent coercivity (~ 3 A/m for the 0.16 cm thick material at 1 Hz frequencies) ensured that the V_Y component was larger than the V_X component. This is displayed graphically in Fig. 4, where the dependence of $\dot{B}_{m,Y}$ and $\dot{B}_{m,X}$ on the amplitude of the applied $H_m(t)$ is displayed, for a driving frequency of 1 Hz. Clearly the value of $\dot{B}_{m,X}$ can be
 395 considerable compared to $\dot{B}_{m,Y}$, for larger H_m amplitudes near the coercivity. At larger amplitudes, the material goes into saturation. Both $\dot{B}_{m,Y}$ and $\dot{B}_{m,X}$ eventually decrease as expected at amplitudes much greater than the coercivity.

To understand the behavior in Fig. 4, a theoretical model of the hysteresis based on the work of Jiles [37] was used. The model contains a number of adjustable parameters. We adjusted the parameters based on our measurements of
 400 $B_m - H_m$ loops including the initial magnetization curve. These measurements

were performed separately from our lock-in amplifier measurements, using an arbitrary function generator and a digital oscilloscope to acquire them. The measurements were done at frequencies from 0.01 to 10 Hz. It was found that
405 the frequency dependence predicted by Ref. [37] gave relatively good agreement with the measured $B_m - H_m$ loops once the five original (Jiles-Atherton [38]) parameters were tuned.

For the parameters of the (static) Jiles-Atherton model, we used $B_s = 0.45$ T, $a = 3.75$ A/m, $k = 2.4$ A/m, $\alpha = 2 \times 10^{-6}$, $c = 0.05$, which were
410 tuned to our $B_m - H_m$ curve measurements. For classical losses, we used the parameters $\rho = 5.7 \times 10^{-7}$ $\Omega \cdot \text{m}$, $d = 1.6$ mm (the thickness of the material), and $\beta = 6$ (geometry factor). These parameters were not tuned, but taken from data. For anomalous losses we used the parameters $w = 0.005$ m and $H_0 = 0.0075$ A/m, which we also did not tune, instead relying on the tuning
415 performed in Ref. [37].

These parameters were then used to model the measurement presented in Fig. 4, including the lock-in amplifier function. As shown in Fig. 4, trends in the measurements and simulations are fairly consistent. The sign of $\dot{B}_{m,X}$ relative to $\dot{B}_{m,Y}$ is also correctly predicted by the model (we have adjusted them both
420 to be positive, for graphing purposes). We expect that with further tuning of the model, even better agreement could be achieved.

Jiles' model makes no prediction of the temperature dependence of the parameters. Ideally, the temperature dependence of $\dot{B}_{m,Y}$ and $\dot{B}_{m,X}$ under various conditions could be used to map out the temperature dependence of the parameters. However, this is beyond the scope of the present work.
425

We make the simplifying assumption that temperature dependence of $\dot{B}_{m,Y}$ may be approximately interpreted as the temperature dependence of a single parameter μ , i.e. that

$$\frac{1}{\dot{B}_{m,Y}} \frac{d\dot{B}_{m,Y}}{dT} = \frac{1}{\mu} \frac{d\mu}{dT}. \quad (6)$$

This is justified in part by our selection of measurement parameters (the amplitude of $H_m = 0.1$ A/m and a measurement frequency of 1 Hz) which ensure
430

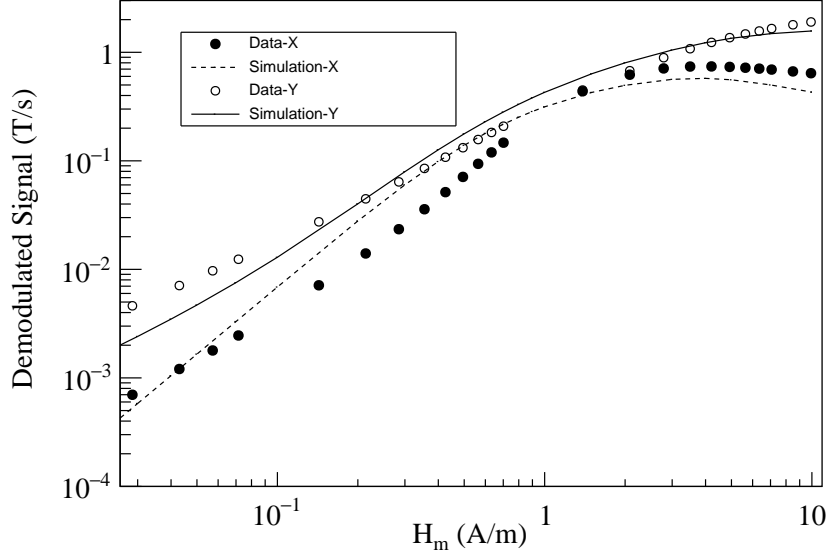


Figure 4: $\dot{B}_{m,X}$ and $\dot{B}_{m,Y}$ as a function of amplitude of the applied H_m field at 1 Hz. Points show the acquired data. Curves display the simulation based on the model described in the text.

that $\dot{B}_{m,Y}$ dominates over $\dot{B}_{m,X}$.

We assign no additional systematic error for this simplification, and all our results are subject to this caveat. We comment further that in our measurements of the axial shielding factor (presented in Section 3.2), the same caveat exists. In
435 that case the in-phase component dominates the demodulated fluxgate signal. In a sense, measuring $\mu(T)$ itself is always an approximation, because it is actually the parameters of minor loops in a hysteresis curve which are measured. In reality, our results may be interpreted as a measure of the temperature-
dependence of the slopes of minor loops driven by the stated H_m .

440 Measurements of $\frac{1}{B_{m,Y}} \frac{d\dot{B}_{m,Y}}{dT}$ as a function of T were made. In general, the data mimicked the behavior of the axial shielding factor measurements, giving a similar level of linearity with temperature as the data displayed in Fig. 3. Other similar behaviors to those measurements were also observed, for example: (a)

when the temperature slope changed sign, $\dot{B}_{m,Y}$ would temporarily give a different slope with temperature, (b) the measured value of $\frac{1}{\dot{B}_{m,Y}} \frac{d\dot{B}_{m,Y}}{dT}$ depended on a variety of factors, most notably a dependence on which of the three witness cylinders was used for the measurement, and on differences between subsequent measurements using the same cylinder.

Based on a number of measurements with different cores and windings, the data showed a range of $0.1 - 2.1\%/K$ for $\frac{1}{\mu} \frac{d\mu}{dT} = \frac{1}{\dot{B}_{m,Y}} \frac{d\dot{B}_{m,Y}}{dT}$, again naively assuming the material to be linear as discussed above. The sign of the slope of $\mu(T)$ was the same as the axial shielding factor technique.

A dominant source of variation between results in this method arose from properties inherent to each witness cylinder. One of the cylinders gave temperature slopes consistently larger $\frac{1}{\mu} \frac{d\mu}{dT} \sim 1.2 - 2.1\%/K$ than the other two $\frac{1}{\mu} \frac{d\mu}{dT} \sim 0.1 - 0.7\%/K$. We expect this indicates some difference in the annealing process or subsequent treatment of this cylinder, although to our knowledge the treatment was controlled the same as for the other two cylinders. Since our goal is to provide input to future EDM experiments on the likely scale of the temperature dependence of μ that they can expect, we phrase our result as a range covering all these results.

Detailed measurements of the effect of degaussing were conducted for this geometry. The ability to degauss led us ultimately to select a larger number of primary turns (48) so that we could fully saturate the core using only the lock-in amplifier reference output as a current source. A computer program was used to control the lock-in amplifier in order to implement degaussing. A sine wave with the measurement frequency (typically 1 Hz) was applied at the maximum lock-in output power. Over the course of several thousand oscillations, the amplitude was decreased linearly to the measurement amplitude (~ 0.1 A/m). After degaussing with parameters consistent with the recommendations of Refs. [17, 19], the measured temperature slopes were consistent with our previous measurements where no degaussing was done.

Other systematic errors found to contribute at the $< 0.1\%/K$ level were: motion of the primary and secondary windings, stability of the lock-in amplifier

475 and its current source, and stability of background noise sources.

To summarize, the dominant systematic effects arose due to different similarly prepared cores giving different results, and due to variations in the measured slopes in multiple measurements on the same core. The second of these is essentially the same error encountered in our axial shielding factor measurements. We expect it has the same source; it is possibly a property of the
480 material, or an additional unknown systematic uncertainty.

4. Relationship to nEDM experiments

Neutron EDM experiments are typically designed with the DC coil being magnetically coupled to the innermost magnetic shield. As discussed in Section 2, if the magnetic permeability of the shield changes, this results in a
485 change in the field in the measurement region by an amount $\frac{\mu}{B_0} \frac{dB_0}{d\mu} \sim 0.01$.

The temperature dependence of μ has been constrained by two different techniques using open-ended mu-metal witness cylinders annealed at the same time as our prototype magnetic shields. We summarize the overall result as
490 $0.1\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.7\%/K$, where the range is driven in part by material properties of the different mu-metal cylinders, and in part by day-to-day fluctuations in the temperature slopes.

We note the following caveats in relating this measurement to nEDM experiments:

- 495 • Although the measurement techniques rely on considerably larger frequencies and different H_m -fields than those relevant to typical nEDM experiments, we think it reasonable to assume the temperature dependence of the effective permeability should be of similar scale. For frequency, both techniques typically used a 1 Hz AC field, whereas for nEDM experiments the field is DC and stable at the 0.01 Hz level. Furthermore, in one measurement technique the amplitude of H_m was ~ 0.004 A/m and in the
500 other was ~ 0.1 A/m. For nEDM experiments $H_m < 0.007$ A/m and is DC.

- Both measurement techniques extract an effective μ that describes the slope of minor loops in $B_m - H_m$ space. A more correct treatment would include a more comprehensive accounting of hysteresis in the material, which is beyond the scope of this work.

Assuming our measurement of $0.1\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.7\%/K$ and the generic EDM experiment sensitivity of $\frac{\mu}{B_0} \frac{dB_0}{d\mu} \sim 0.01$ results in a temperature dependence of the magnetic field in a typical nEDM experiment of $\frac{dB_0}{dT} = 10 - 270$ pT/K. To achieve a goal of ~ 1 pT stability in the internal field for nEDM experiments, the temperature of the innermost magnetic shield in the nEDM experiment should then be controlled to the $0.1 - 0.004$ K level. This represents a potentially challenging design constraint for future nEDM experiments.

As noted by others [39], the use of self-shielded coils to reduce the coupling of the B_0 coil to the innermost magnetic shield is an attractive option for EDM experiments. The principle of this technique is to have a second coil structure between the inner coil and the shield, such that the net magnetic field generated by the two coils is uniform internally but greatly reduced externally. For a perfect self-shielded coil, the field at the position of the magnetic shield would be zero, resulting in perfect decoupling, which is to say a reaction factor that is identically unity. For ideal geometries, such as spherical coils [40–42] or infinitely long sine-phi coils [43–45], the functional form of the inner and outer current distributions are the same, albeit with appropriately scaled magnitudes and opposite sign. More sophisticated analytical and numerical methods have been used extensively in NMR and MRI to design self-shielded gradient [46, 47], shim [48, 49], and transmit coils [45, 50], and should be of value in the context of nEDM experiments, as well. We are also pursuing novel techniques for the design of self-shielded coils of any arbitrary field profile and geometric shape [51].

5. Conclusion

In the axial shielding factor measurement, we found $0.6\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.7\%/K$, with the measurement being conducted with a typical H_m -amplitude

of 0.004 A/m and at a frequency of 1 Hz. In the transformer core case, we found $0.1\%/K < \frac{1}{\mu} \frac{d\mu}{dT} < 2.1\%/K$, with the measurement being conducted with
535 a typical H_m -amplitude of 0.1 A/m and at a frequency of 1 Hz.

The primary caveat to these measurements is that both measurements (transformer core and axial shielding factor) do not truly measure μ . Rather they measure observables related to the slope of minor hysteresis loops in $B_m - H_m$ space. They would be more appropriately described by a hysteresis model like
540 that of Jiles [37], but to extract the temperature dependence of all the parameters of the model is beyond the scope of this work. Instead we acknowledge this fact and relate the temperature dependence of the effective μ measured by each experiment.

We think it is interesting and useful information that the two experiments
545 measure the same scale and sign of the temperature dependence of their respective effective μ 's. This is a principal contribution of this work.

In future work, we plan to measure $B_0(T)$ directly for nEDM-like geometries using precision atomic magnetometers. We anticipate based on the present work that self-shielded coil geometries will achieve the best time and temperature
550 stability.

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